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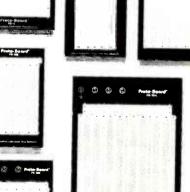
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Panasonic® RF-3100

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ON THE COVER

A completely self-contained PROM programmer for storing your programs in reusable 2716 EPROM's. This project is a must for experimenting with microprocessors and designing microprocessor based projects. Construction details start on page 45.

ANNUAL INDEX JANUARY-DECEMBER

1981

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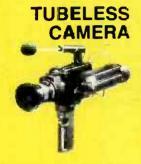
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VIDEO ELECTRONICS

DAVID LACHENBRUCH CONTRIBUTING EDITOR



Video cameras have entered the all-solid-state age in the United States with the start of marketing of Hitachi's MOS camera, which uses a single ³/₂-inch metal oxide silicon IC as an image sensor. The MOS sensor has 180,000 picture elements, providing a claimed 260-line horizontal resolution. The camera weighs 3.96 pounds.

The most noticeable aspect of the camera is its small size and light weight—indeed, the lens and viewfinder combined are larger than the camera itself. The picture differs from any made with a Vidicon pickup tube in that there is no lag, smear, or burn-in even when the camera is aimed directly at a light. Hitachi claims that the MOS is good for 100,000 hours, as compared with 5,000 hours' average life of a Vidicon. Among its advantages are low power consumption (5.3 watts), rapid start (picture appears in a half second), reliability, and ruggedness. The suggested list price is \$1,995.

MINI-VHS

While Japanese manufacturers were meeting to standardize on the portable VCR of the future using quarter-inch tape, it was revealed that JVC has developed a miniature cassette compatible with the VHS system that can be in production as early as next spring.

It will record for up to an hour at the slowest VHS speed in a new ultra-compact portable VCR developed by JVC that weighs about 5.5 pounds and is about 20% smaller than the Technicolor VCR, which uses non-standard quarter-inch cassettes.

The mini-cassettes can be played back through the portable recorder. or, with an adaptor, through any VHS recorder. JVC is understood to be talking with other companies in the VHS group to adopt its new system as a standard for portables. The new mini-VHS development isn't expected to deter Japanese companies from working on a new quarter-inch standard, but it could take some of the urgency out of their efforts.

NEW MINI-VCR

The smallest and lightest VCR to date made its debut in Germany this past September, and will be headed this way soon. Ironically, it's made in Japan, not Germany, but will bear the Grundig label in this country. The new recorder is made by Funai Electric and is compatible with the small recorder Funai makes for Technicolor, but it has been completely redesigned. Weight and battery has been trimmed to five pounds (four pounds without battery) as compared with Technicolor's seven pounds, and the entire recorder is about the size of a cigar box. It provides slow and fast speeds, freezeframe, and records for up to an hour on a quarter-inch cassette only slightly larger than an audio cassette. The price will be about \$1,000 in the U.S., including power supply.

HIGH-DEFINITION TV

There have been plenty of proposals for high-definition television, most of them centering on 1,000-plus-line standards requiring about 30 MHz of bandwidth per channel. Now, along comes Imagevision, developed by Compact Video, which requires 10 MHz, is claimed to be "ready for delivery," and uses modified conventional equipment. Imagevision was demonstrated before the Society of Motion Picture and Television Engineers, where it received favorable comment.

Imagevision uses a new television standard called "PALAF" (for Phase Alternate Line Alternate Frame) and has 665 lines, claimed to provide the definition of 1,250 lines in other systems, with 24 non-interlaced frames per second. All the equipment used (such as cameras and videotape recorders) was commercially available gear modified for the system. The most successful demonstration of Imagevision was the projection of part of a movie being made using the system. The movie had been transferred to 35-mm film and was projected on a theater screen. Most observers agreed it looked every bit as good as a movie made directly on 35-mm film. Somewhat less successful were satellite-relayed video pictures projected by light-valve projection-TV systems—but that may be because the projectors became the limiting factors.

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FEBRUARY 1982

WHAT'S NEWS

Educators confer on microcomputers

Microcomputer Week '82, the third annual conference cosponsored by the Center for Advancement of Teaching and Learning (CATALYST) of Jersey City State College, will be held at the college from the 3rd to the 7th of March 1982.

The conference centers on microcomputers in education. Some 66 all-day seminars or short courses will be offered for professional development and improved performance. Each will include eight hours of instructions. Some seminars will extend over two or three days. There will also be 24 informal evening sessions—during dinner and continuing after dinner—for groups with special interests and problems.

Cost of the seminars range from \$95 for one day to \$73 per day for the whole week. A brochure with complete details, registration forms, etc., has been prepared. Further information is obtainable by phone from 201-434-2154 or 201-547-3094, or by mail from CATALYST Conference, H 112, Jersey City State College, 2039 Kennedy Boulevard. Jersey City, NJ 07305.

TV standards converter for video recorders

Early international television broadcasts were hampered by differing television standards. Systems using the 525-line frames of the United States, Japan, and other countries were incompatible with the 625-line standard of most of Europe. (At one time, France had an 815line system.)

Those difficulties were overcome in a number of ingenious ways for the first few broadcasts, and very soon "system converters" were designed to input the signals of one system and change the number of lines, modulation method, sync signals, etc., to those of the system to which the signal was being relayed.

Color, with its NTSC, PAL, and SECAM systems, brought new difficulties, which were also quickly ironed out with new converters. Video recording raised the problem anew, an unfortunate circumstance, since videotapes can be shipped more easily than international broadcasts can be set up.

But now a company called Instant Replay (2980 McFarlane Rd., Coconut Grove, FL 33133), has put on the market an Image Translator, in two models. The SMP-1 permits a viewer with a PAL recorder and receiver to play NTSC tapes, witout modifying the receiver. The SMP-2 permits a viewer to play back tapes recorded in the PAL format in full color on an NTSC VCR. using an American TV receiver, without modification. It can be used with either direct-view or large-screen projection TV. Thus international video broadcasting's objective of cultural exchange is being extended to video recording.

The Image Translator is an extension of the publishing efforts of the manufacturer, Instant Replay Video Magazine, "the world's first video magazine." a two-hour tape published bimonthly at a price of \$59.95 per issue. Believing the television image to be the most nearly perfect medium of international cultural exchange, and the videotape the most nearly universal unit for such exchange, owner and editor Chuck Azar developed the Translator to internationalize the videotape. To promote the internationalization further, a printed list of all Image Translator owners will be sent to all subscribers, to enable them to swap tapes between countries.

Improved approach in auto-focus camera

A new Pentax 35-mm SLR camera, the *ME-F*, expected on the market in early 1982, introduces a new computerized through-the-lens electronic focusing system. Measuring the actual focal contrast at the center of the field of vision, in a manner similar to that of the human eye, the system is claimed to be far more accurate than those that cover the whole field, or range-finder electronic systems that measure the image-field distance.



PENTAX ME-F single lens reflex camera with new 35-mm-70-mm Zoom lens.

The new camera focuses correctly on complex patterns, subjects behind glass, and over the whole depth-of-field range.

The *ME-F* lets the photographer know when the subject is in focus by means of a dualsignal indication: An LED lights in the viewfinder at the point of optimum focus and a beep signal sounds.

The list price of the *ME-F* with zoom lens—is expected to be about \$1,000.

New VideoDisc features demonstrated in France

Dr. Jon K. Clemens, co-recipient of the Rhein Prize 1979, and Director of VideoDisc Systems at RCA's research laboratories. demonstrated the advanced capabilities of the new RCA 'CED' (Capacitance Electronic Disc) system at the recent Vidcom convention in Cannes, France. Dr. Clemens showed such features as programmable random high-speed visual access. search, repeat picture, and automatic repetition of selected program segments.

In demonstrating the visualsearch feature, Dr. Clemens used the prototype system's two search speeds, $16 \times$ and $120 \times$. He also programmed the player to repeat sections of the disc, demonstrated its repeat-picture capability, and proved its wearfree feature by playing a still picture from a single groove for 20 minutes—amounting to 9,000 replays of the same groove.

No date was set for the introduction of those advanced VideoDisc player features to the market.

FAA renovating nation's air traffic control

The Federal Aviation Administration is undertaking a program to upgrade air-traffic control computers throughout the country. The program is aimed at meeting projected air-traffic demands of the 1990's and beyond. It will call for replacement of the 23 Air Route Traffic Control Center computers with an advanced system.

RCA has been awarded a \$4 million contract to provide system engineering and technical analysis in the program. RCA's work will cover five areas: requirements specification and allocation; cost, schedule, and *continued on page 12*

NEW Here's a professional, portable DC-20 MHz Oscilloscope with build-it-yourself Heathkit savings



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You Explore the New TRS-80 Model III **Inside and Out**

NRI training is hands-on training, with practical experiments and demonstrations as the very foundation of

vour knowledge. You don't just program vour computer, vou go inside it...watch how circuits interact...interface with other systems...gain a real insight into its nature.

Training includes the new TRS-80 Model III microcomputer, 6-function LCD Beckman multimeter, and the NRI Discovery Lab with hundreds of tests and experiments.

You also work with a precision 26-scale, 6-function LCD Beckman multimeter featuring full portability and a $3^{1/2}$ digit display. Using it along with the exclusive NRI Discovery Lab and your TRS-80, you perform over 60 separate experiments in all. You learn how to troubleshoot and gain greater understanding of the information Lab

your tests give you.

Advanced Technology Microcomputer Is Yours to Keep

As part of your training, NRI sends you the new stateof-the-art TRS-80 Model III microcomputer. This functional unit is complete with 65-key keyboard and 12" display in one desk-top unit. It features

high-speed cassette loading, built-in interface for parallel printer, and provisions for optional disk drive. Its 4K RAM is internally expandable to 16K or 48K and its BASIC language is compatible with most Model I software.

Along with your multimeter and the NRI Discovery Lab, this latest concept in advanced microcomputers is yours to learn with, yours to keep and use for your own personal programs, business use, and other applications.

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Get all the details on this exciting course in NRI's free, 100-page catalog. It shows all equipment, lesson outlines, and facts on other electronics courses such as Complete Communications with CB, TV and Audio Servicing, Digital Electronics, eleven different interest areas in all.

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(TRS-80 is a trademark of the Radio Shack division of Tandy Corp.)

WHAT'S NEWS

continued from page 6

risk analysis: performance analysis. simulation, and modeling: feasibility and trade studies, and technical management standards and practices.

Super circuit chips speed data tenfold

"Super chips," that Hughes Aircraft call "electronic circuits that resemble an array of 100 Los Angeles street maps on a thumb tack," are now being developed at Hughes' laboratories. Called Very High Speed Integrated Circuits (VHSIC), they will be more reliable and require less power than integrated circuits now in use.

The chip technology used in the program includes CMOS (Complementary Metal Oxide Semiconductor) and SOS (Silicon On Sapphire). The SOS technology is especially significant for military systems, because it is inherently hardened against radiation.

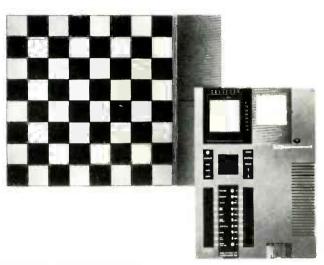
Hughes has also received an Army contract to develop a highspeed electron-beam lithography system that will focus beams of electrons to write circuit patterns in submicron dimensions. Those circuit patterns will be converted into transistors and interconnects smaller than any now in production.

Yankee chess computer is new world champion

Chess Champion Mark V, made by SciSys Computer Inc. of New York City. emerged as the winner in the Commercial Division. World Micro Computer Chess



PHOTO-LITHOGRAPHY MACHINE that will be used to etch VHSIC circuits on chips the size of a thumbtack. Hughes and five other firms are in the program conducted by the Defense Department to develop the high-speed superchips. The machine is being operated by Hughes technician Margaret Sanchez.



THE CHESS CHAMPION MARK V.

Championships, held last Fall in Travemunde, Germany. The tournament was held under the auspices of the International Chess Association and the World Chess Federation. It was open to all chess programs executed by readily available onechip microprocessors.

The Chess Champion Mark V is an AC-powered chess computer with an integral LCD chessboard. Its 32K memory is many times more powerful than the average home computer. and its variable time control enables it to play any form of chess, from Speed to Tournament. It can play 12 games simultaneously against humans, other computers, or even against itself. On request, it can analyze, comment, or advise on the game in progress, and can provide a complete game history or predict the outcome.

Sony Corp sets up a U.S. laboratory

Sony Consumer Electronics Laboratories (SOCEL) a division of Sony Corp of America, will be located in the Century Plaza in Paramus. NJ, reports Sony Corp. executive vice president Kenji Tamiya. The laboratories will conduct research, development. and design work. particularly the areas of CATV systems and terminals, receivers for direct satellite broadcasts. and videotext/teletext systems and terminals. "The laboratories will offer a unique opportunity for interaction between American and Japanese engineers." said Mr. Tamiya.

The new laboratory is the latest in a number of Sony "firsts" in America. In 1972 Sony opened the first U.S. color TV plant owned by a Japanesebased company, in San Diego, CA. In 1977, it established what is now one of the largest magnetic tape plants in the world, in Dothan, AL. The Sony Technology Center in Palo Alto, CA, carries out broadcast equipment research, development, and manufacturing, and was the first Sony facility to be located outside Japan.

Electronic newspaper debuts in Chicago

NITE-OWL, a full-channel teletext publication, is now being broadcast on WFLD-TV, Channel 32, Chicago. Designed for late-night information, it supplies up-to-the-minute news. sports, weather, and other information, and advertising.

Information is presented in text form (teletext) on the screen. Teletext was first developed in Great Britain and is now being offered commercially in the United States by Field Electronic Publishing. **R-E**

LETTERS

SENDING GOOD CW

Fie on Herb Friedman, and a fig in his ear! He seems to be showing an infirmity of old age—intolerance—and maybe even a lack of good sense in "Communications Corner," **Radio-Electronics**, October 1981.

As a ham, from my earliest days, I tried diligently to send good CW, first on a straight key, then on a swiper (homebrew) and then on the ultimate (for those days): a bug (also homebrew). And, I, too, share his view that handwrought CW is a pure performing art.

First ticketed in 1933, I am now (after a long layoff) a reconstituted ham. Once I got back on the air, I was thunderstruck by the bad straight key fists, the miserably misadjusted and misrun bugs, and that ultimate abomination: the key running its operator instead of the other way around. Do you ever hear a "6" sent with four dots any more?

Then, I, too became aware of the beautiful CW coming from microprocessor keyboards, whether at 20 or 60 wpm. What a joy! As with good writing, I believe that CW should be sent with clarity and intelligibility, and to hell with dialects or accents. If it takes a machine to do it, so be it! It's all right with me.

So come on, Herb. Don't display the length and condition of your teeth. Get a good keyboard and join the fun. I have one (homebrew) and it's better than crackerjack.

C.H. FRERES, KJ6G, San Diego, CA

NIKOLA TESLA

I would like to associate myself with the thoughts expressed in the letters by Marc Seifer and Vince Marasco in the June 1981 **Radio-Electronics** about Nikola Tesla.

Tesla is one of my favorites. I can turn on my Edison light bulb and visualize, in some vague manner, Tesla's rotating magnetic field at some remote point miles away. Yes, it is truly Tesla's rotating magnetic field—the one he discovered before he invented a device wherein the field could reside and produce rotating power: the induction motor. And subsequently, he was forced to invent a device to generate that magnificent field—the polyphase alternator.

Had Edison's proposal for electric power distribution prevailed, we would now have direct-current generating power plants spaced about every mile where electricity is in use today. And we would know no such thing as the sealed motor-compressor system used in modern refrigeration and air-conditioning systems. Edison's DC voltage could not be stepped up or down with transformers, and the losses in his system would be great.

What about electric-power transmission without wires? Some large company —I would suggest Westinghouse—should embark upon a research project to rediscover that lost art.

I could go on and on. I am interested in Tesla and his discoveries. A machinist friend of mine is also fascinated, as are others in the electronics field. Had Nikola Tesla never existed, nevertheless, civilization would have required one like him just the same.

Knowledgeable people have been made aware of most of the other great men of the past—so why not the unique Nikola Tesla? ALFRED C. POWELL.

Pensacola, FL

Why not, indeed? Well, you're doing your part.

LED VU METER ENCLOSURE

Reading through the May 1981 **Radio-Electronics**, I was struck by the appearance—in the article, "Build This LED VU Meter"—of an enclosure produced by our client, PacTec Corp., subsidiary of LaFrance Corp. I checked carefully for a reference to the PacTec Enclosure, to no avail. In fact, I noted that the kit for the VU-1 available from BFA Electronics actually includes no case at all.

Your readers might like to know that the case seen in the article is available from PacTec, Enterprise and Executive Avenues, Philadelphia, PA 19153. It is one of the series CM enclosures, and appears specifically to be a CM 5-200.

We have many prototyping kits in our line. The kits are particularly useful to designers and hobbyists, and are available in a wide range of sizes. I hope that you and your readers will find this material useful.

JON H. CLINCH,

Larwin/Livers Associated, Inc.

RADAR DETECTORS

In reading the "Letters" department in **Radio-Electronics**, and being the owner of one of the best "Super Het" radar detectors on the market, I can only feel sorry for those people who must constantly challenge the right to own and operate such devices. It is a shame that, for once, the foxes were "outfoxed themselves" by allowing a hole in the law big enough to *continued on page 22*



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Introducing the Sinclair ZX81

If you're ever going to buy a personal computer, now is the time to do it.

The new Sinclair ZX81 is the most powerful, yet easy-to-use computer ever offered for anywhere near the price: only \$149.95* completely assembled.

Don't let the price fool you. The ZX81 has just about everything you could ask for in a personal computer.

A breakthrough in personal computers

The ZX81 is a major advance over the original Sinclair ZX80-the world's largest selling personal computer and the first for under \$200

In fact, the ZX81's new 8K Extended BASIC offers features found only on computers costing two or three times as much. Just look at what you get:

Continuous display, including moving graphics

Multi-dimensional string and numerical arrays

* Plus shipping and handling. Price includes connectors for TV and cassette, AC adaptor, and FREE manua



NEW SOFTWARE:Sinclair has published pre-recorded programs on cassettes for your ZX81, or ZX80 with 8K BASIC. We're constantly coming out with new programs, so we'll send you our latest software catalog with your computer.

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- IK of memory expandable to 16K

The ZX81 is also very convenient to use. It hooks up to any television set to produce a clear 32-column by 24-line display. And you can use a regular cassette recorder to store and recall programs by name.

If you already own a ZX80 The 8K Extended BASIC

chip used in the ZX81 is available as a plug-in replacement for your ZX80 for only \$39.95, plus shipping and handling-complete with new keyboard overlay and the ZX81 manual.

So in just a few minutes, with no special skills or tools required, you can upgrade your ZX80 to have all the powerful features of the ZX81. (You'll have everything except continuous display, but you can still use the PAUSE and SCROLL commands to get moving graphics.)

With the 8K BASIC chip, your ZX80 will also be equipped to use the ZX Printer and Sinclair software.

Order at no risk**

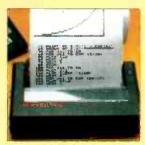
We'll give you 10 days to try out the ZX81. If you're not completely satisfied, just return it to Sinclair Research and we'll give you a full refund.

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ZX81 MANUAL: The ZX81 comes with a comprehensive 164-page programming guide and operating manual designed for both beginners and experienced computer users. A \$10.95 value, it's yours free with the ZX81.



ZX PRINTER: The Sinclair ZX Printer will work with your ZX81 or ZX80 with 8K BASIC. It will be available in the near future and will cost less than \$100.



computer, the ZX81 is expand able. Sinclair's 16K memory module plugs right onto the back of your ZX81 (or ZX80, with or without 8K BASIC). Cost is \$99.95, plus shipping and handling.

"The \$99⁹⁵ personal computer.

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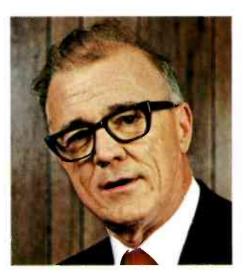
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y father always told me that there were certain advantages to putting all your eggs in one basket. "John," he said, "learn to do one important thing better than anyone else, and you'll always be in demand.

I believe he was right. Today is the age of specialization. And I think that's a very good thing.

Consider doctors. You wouldn't expect your family doctor to perform open heart surgery or your dentist to set a broken bone, either. Would you?

For these things, you'd want a specialist. And you'd trust him. Because you'd know if he weren't any good, he'd be out of business.

Why trust your education and career future to anything less than a specialist?

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Specialists aren't for everyone.

I'll tell it to you straight. If you think electronics would make a nice hobby, check with other schools.

But if you think you have the cool-and want the training it takes to make sure that a sound blackout during a prime time TV show will be corrected in seconds-then answer this ad. You'll probably find CIE has a course that's just right for you!

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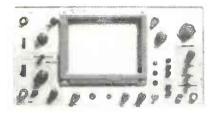
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continued from page 13

squeeze my Q-1000 detector through.

I don't believe for one moment that the policeman operating the radar unit is primarily interested in saving my life on the roadways, or is instructed and knowledgeable enough to use the device without error. When I see a policeman just *sitting* behind a bush, operating those objectionable devices, I see my state and local tax money being wasted.

With crime the way it is today, I think that speeding is the lowest rung on the ladder when it comes to breaking the law. If as much time went into combatting other areas of crime as is spent in playing "cat and mouse" with the motorist, I think we would see a sharp decline in crime in those really important areas. Besides, if it were not for the high revenue collected by ticketing motorists for their deeds, I doubt whether we would see policemen just sitting around, relaxing in their cars with their radar units.

To say that every radar-detector owner has it solely for the purpose of speeding without getting caught is like saying that anyone who owns a gun is planning a bank job or an assassination. What the complainers need is to be tagged by radar; then their attitudes would change. I have seen so many radar traps that were cleverly set up, that I hate to think about how much money was collected that way.

Let us face the facts: As long as radar is used the way it is, and radar detectors are legal—as they should be—then the game will continue on the highways. I, for one, will continue to plug in my detector for protection. M.J. RYBICKI,

Berwyn, IL

THE PIANOCORDER

We read Warren Baker's article about the Pianocorder in your November 1981 issue with much interest. It is certainly a fascinating development, and we are pleased to have licensed Superscope to reproduce many of our piano rolls on their cassettes.

I do object to a few minor points, however. The so-called "old-fashioned" pneumatic players are still being manufactured by the thousands; and so are the rolls, which are not nearly as "fragile" as you might be led to believe. Some of our products are 80 years old, and still performing well.

In addition, rolls bear not only the music but the printed lyrics as well, positioned to appear simultaneously with their corresponding notes. Grouping around an "old-fashioned" player, and singing along with the rolls, is a pleasure that no electronic update has yet surpassed.

, QRS MUSIC ROLLS, INC., Robert J. Berkman, Adm. Assistant

BACKSCATTER

I'd like to comment on the article by Mr. Stanley Leinwoll, "Why Radio Moscow is continued on page 103



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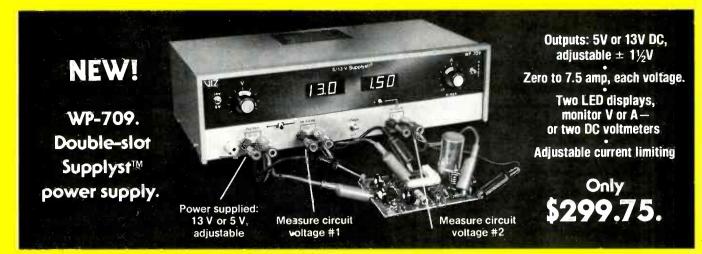
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SATELLITE/TELETEXT NEWS

GARY ARLEN CONTRIBUTING EDITOR

TELETEXT LOOK-ALIKE SERVICES

Although line 21 of the vertical blanking interval is used primarily for closed captions for people with impaired hearing, a number of new activities are making that channel look more like teletext. Dozens of programs on ABC, NBC, and PBS are transmitted with those closed captions each week; a special TeleCaption decoder is required.

But in the past few months, several projects have gotten underway to add more text services. For example, early in the year the networks began transmitting "InfoData," which is a text sequence, used intially to announce which programs will be captioned during the coming week. More recently, KCET-TV Channel 28 in Los Angeles began carrying "Newsline," a version of the Associated Press news "ticker" used by cable TV systems. In that service, electronic versions of the news wire scroll across the screen of TV sets keyed into the captioning system. WHA-TV in Madison Wisconsin, KWCM-TV in Appleton, Minnesota and WNET in Newark, NJ are planning similar news and information systems on line 21. All are public-TV stations.

TEXT VIA SATELLITE

Several teletext packages, intended primarily for cable-TV usage, are going aloft via satellite during the coming months. The most ambitious is one that will supplement "The Weather Channel," a new all-day weather-information service due to get under way in 1982. The teletext portion will include local weather reports transmitted from the Weather Channel's Atlanta headquarters to "Weather STAR" addressable teletext receivers in various cities. Meanwhile, another 24-hour text-only weather-information service, "View Weather," is being transmitted as part of the CableText package on Satcom | Transponder 6.

Perhaps the most substantial teletext-via-satellite project is the one now under way at Time Inc. About 4,000 pages of information are being sent over a full transponder from Time Inc. headquarters in New York to Southwestern Cable TV in San Diego (owned by a Time subsidiary) as part of a major teletext test.

AROUND THE SATELLITE CIRCUIT

Blonder-Tongue Laboratories, a long-time respected supplier of cable-TV hardware, is getting into the earth-station business with a new dish and earth-station modulator.

Forget any ideas about using satellites to collect solar power and beam it down to earth as the next big wave of satellite activity after communications satellites. The U.S. Department of Energy gave a tentative approval to such a concept, which would probably take 50 years to put into place. But a few months ago the National Research Council recommended against such a plan, saying it would be far too costly-as much as 3 trillion dollars. Furthermore, such a solar-power satellite would probably create interference with communications satellites and there would be other technical and political problems in developing such a system. But, just in case, NASA is going to continue keeping an eye on the idea.

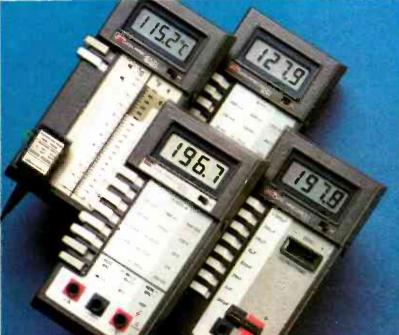
Metromedia is now sending the Merv Griffin Show to TV stations; Merv's show is transmitted on Westar III every morning, to be taped by stations which run it later in the day.

JUDGE **APPROVES** PRIVATE TERMINALS

CANADIAN A British Columbia judge has found an apartment-house manager "not guilty" of accusations that he operated a private satellite dish to provide video programming to tenants. The decision is seen as a victory for backyard satellite-receiver users in Canada, where the federal government has been trying to crack down on use of such private terminals. The ruling, however did not resolve the question of reception of video transmissions via satellite from across the border.

Meanwhile, prosecuters in other Canadian provinces are continuing their cases against other private-terminal owners-both commercial operators (such as the apartment-house manager) and personal users. The Canadian Radio-Television Commission requires earth-station operators to have a technical certificate and a broadcast license. B-F

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MXR Model 156 CX Decoder

MXR INNOVATIONS, INC. (740 DRIVING Park Avenue, Rochester, NY 14613) offers an interesting line of audio accessories. Their latest one is a *CX* decoder, the *model 156*.

As you must be aware by now, CX is a compatible companding system for CX-encoded records. Without a decoder, the records sound normal. With a decoder, they attain a dynamic range of about 85 dB, with the extension being at the "quiet" end. That means that the noise floor drops by 20 dB, resulting in virtually noise-free playback and a much more natural-sounding reproduction of the recorded material.

My first experience with the *model* 156 decoder was uncanny. I cued the turntable arm, stepped back, and waited for the music to begin. Nothing happened. Thinking that maybe I had switched my speakers off, I went up to check them. As I was looking, the music began. There'd been no noise when the stylus contacted the record, and no noise from the lead-in groove. Nothing!

And the quality of the sound! Loud passages were, of course, loud—CX doesn't concern itself with headroom. What was amazing was that during quiet passages, I could hear things that I would otherwise have missed because of record surface-noise. And, between cuts, there was nothing to indicate that

a record was on the turntable—just dead silence. It was like being present at a performance in a hushed room. The only thing you heard was what you were *supposed to* hear, and then, you heard it all. It's something that has to be experienced to be believed.

About the time you read this, CBS is expected to have about 50 *CX*-encoded records on the market, and a number of other recording companies have signed agreements with CBS to also use the *CX* companding system.

The MXR decoder

EQUIPMENT REPORTS

The model 156 is extremely simple to connect and use. Two sets of audio cables connect it to the tape-recorder loop of your preamp or receiver; it obtains its power from a 117-volt power line. (My early-model decoder did not have provisions for restoring the use of the jacks that it occupied; MXR tells me that this will be remedied in later models. In any event, the fact is that the decoder can be placed just about anywhere in the loop with other equipment, thus still allowing your sound system full flexibility.)

All the controls are located on the small unit's front panel (the jacks are on the rear apron). A 7-inch calibration record is included with the decoder, and calibration is simply a matter of playing the record and adjusting the REFERENCE LEVEL control until neither one of the two LED's mounted in the front panel is on. That's all there is to it. The adjustment isn't even critical—an error of as much as 6 dB will not be noticeable.

After the unit has been calibrated, nothing has to be adjusted until a different phono cartridge or amplifier is brought into the system.

Aside from the REFERENCE LEVEL control, there are only two others—one for DISPLAY that brings the LED's into the circuit, and one to switch the decoder into or out of the system. There's no power switch—the device draws only 2.4 watts when in use and the assumption is made that it will probably be connected to a switched outlet on your equipment, anyway.

The frequency response of the decoder is given as 20 Hz-20 kHz, ± 0 , -0.5 dB. THD is less than 0.09% over that range, with 0.05% typical at 1 kHz.

Total dynamic range available is specified as 100 dB.

The manual, while short, explains everything you'll need to set up and use the decoder.

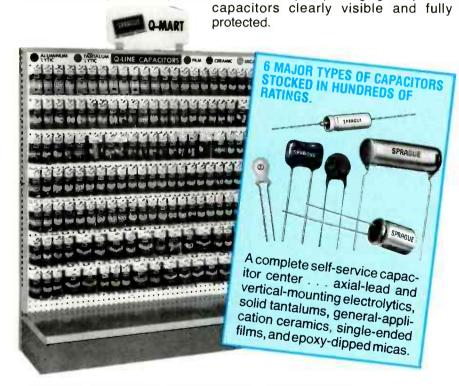
One word of caution: The decoder works *only* for *CX*-encoded records. If a *CX*-encoded record is recorded on tape and played back through the decoder, only the characteristics of the material on *that record* will be modified; any tape hiss that was present before will still be there.

Finally, while *CX* records are on the shelves, they may be hard to locate the fact that a recording is *CX*-encoded is shown on the *back* of the jacket, not the front. (That may be due to the fact that, since the records can also be played on ordinary equipment without a decoder, the record companies don't want to scare away record buyers who don't know about the system and don't have special equipment.) Once *CX* gets a foot in the door, though, the *CX* logo will probably move to the front, where it can easily be seen.

The MXR model 156 decoder is a good example of how simple it is to improve the pleasure you can get from a good recording. It has a suggested list price of \$99.95...and don't forget to watch for the CX logos on the back of the record jackets. **R-E**

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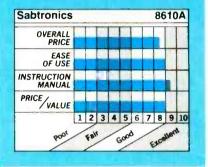


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Sabtronics Model 8610A 600-MHz Frequency Counter



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FOR ANYONE—WHETHER HE'S A TECHnician, ham, or designer—who has to measure signal-frequencies, a frequency counter is a necessity. For a long time I tried getting along without one. I'd borrow one from friends or, more usually, take the equipment to be checked out usually an amateur radio handie-talkie or a piece of digital equipment I'd put together—to someone who had a counter.

Eventually that became too much. The stuff I was building was getting too awkward to carry, and frequently required more than a simple yes-or-no frequency check. So, I finally broke down and purchased a counter of my own.

My choice was the Sabtronics *model* 8610A 600-MHz counter. Although it has a smaller brother (the *model* 8110A, good to 100 MHz) and a bigger one (the *model* 8000B, that goes up to 1 gigaherz). the 600-MHz version offered just what I needed, with a little bit to spare.

It covers all the radio bands I use and has a low-frequency limit of 20 Hz; some of my projects involve audio frequencies and the *model 8610A* is ideal for that end of the spectrum, too.

Putting it together

The counter is available either as a kit or assembled. Naturally, I chose the kit form. I was very pleasantly surprised with the quality of the components and of the manual—that came with the kit. The main PC board is silk-screened and solder masked. The 600-MHz prescaler board is not, but very clear diagrams make it difficult to go wrong.

The construction manual is excellent. There are plenty of diagrams and (literally) step-by-step instructions with continued on page 32

RADIO-ELECTRONICS

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EQUIPMENT REPORTS

continued from page 30

a space for you to make a mark indicating that you've successfully completed that step.

There are also a number of construction hints along the way, at places where a novice builder might find himself wondering exactly what he was supposed to do. There's even an introductory section on how to make good solder connections. The manual includes a table showing what readings should be obtained with different switch settings and, finally, provides troubleshooting hints. Sabtronics leaves nothing to chance!

Calibration of the unit is easy. An internal SENSITIVITY control is adjusted using a voltmeter, and a small variable capacitor trimmed to bring the counter right on frequency. That calibration can be performed in two ways. The first is conventional-input a 10-MHz signal from a known-to-be accurate source. The second method requires a shortwave receiver, but is just about as accurate. A receiver tuned to WWV at 10 MHz is used as a standard, and the counter, placed close to the receiver, adjusted to read the same frequency. The accuracy of the receiver isn't critical-when the National Bureau of Standards says it's transmitting on 10 MHz, who's to doubt it?

Construction went quickly, and the unit worked the first time it was turned on

Features

The model 8610A has an 8-digit LED display. The decimal point, which adjusts its position automatically, also serves to indicate the gate time: 0.1. 1. or 10 seconds, by flashing at that rate. There is also an overflow indicator on the display to indicate an out-of-range condition.

Three switches on the front panel select frequency range, gate time, and turn the counter on and off. A single BNC connector is used for input. Power is supplied either by four "C"-type nickel-cadmium rechargable cells (optional) or from a power supply/charger, also available as an option. The counter can be used while the batteries are being charged.

Earlier, I mentioned a low-frequency limit of 20 Hz. I should clarify that by stating that the guaranteed operating range is 20 Hz-600 MHz. Typical operating range, though, is specified as being from 10 Hz to 750 MHz. Resolution is claimed to be 0.1 Hz up to 10 MHz, 1.0 Hz to 100 MHz, and 10 Hz to 600 MHz.

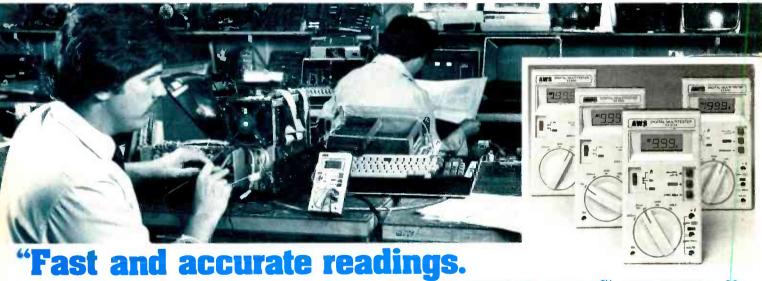
The counter features input protection -from 150-volts RMS up to 10 MHz to 4-volts RMS above 100 MHz. Its sensitivity is 10-mV RMS up to 100 MHz, 70-mV RMS up to 450 MHz, and 150mV RMS to 600 MHz.

The unit uses a temperature-controlled crystal oscillator and, while there is some drift-under 200 Hz during the first few minutes of operation-the counter is essentially rockstable after less than half an hour. Temperature stability is given as 0.1 ppm (Parts Per Million) per °C, with an aging rate of less than 5 ppm per year. According to the specifications, setability is claimed to be ± 2 ppm: I found it even better.

The counter measures $8 \times 6.5 \times 3$ inches and weighs 1.2 pounds without batteries. A bail is provided to tilt the unit up for better visibility. There's quite a lot of empty space in the enclosure, which leads me to wonder what other items Sabtronics has in mind to put in there.

I feel that the model 8610A frequency counter is one of the best equipment choices I could have made for my bench. It's available from Sabtronics International, Inc., 3709 N. 50th Street. Tampa, FL 33610 for \$119.00 in kit form and \$149.00 assembled, plus \$5.00 for shipping and handling. R-F

continued on page 34



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WITH THE CASUAL MONITORING OF VHF/ UHF frequencies at an all-time high, it is not surprising that many law-enforcement agencies, among others, are concerned with the privacy of their communications. While many of those agencies prefer to use a simple "10" code, or a variation of it, still others prefer the additional protection of a voice privacy system.

There are many ways to garble a voice transmission including speech inversion, rolling code, spread spectrum, frequency hopping, and voice digitizing. The new model *DSC-2* from Grove Enterprises restores intelligibility to speech-inverted communications.

In speech inversion, high voice-frequencies and low voice-frequencies are transposed by the mixing action of a local audio-oscillator. A descrambler simply takes the inverted speech and restores it by passing it through a circuit that is identical to the one that originally scrambled it. That is what this unit does.

Since an agency that wishes to scramble its transmissions can choos. any oscillator frequency it wants, the unit has a control to "fine tune" the restored speech. Often, a speech-inverted signal will also have an audio tone superimposed over it for increased security. That technique, called tone masking, reduces the intelligibility of even restored speech. The *Code Breaker* has a tunable audio notch-filter to remove such an audio tone, leaving the restored speech out in the clear. The unit can also be used with shortwave receivers to remove heterodyne interference that is often heard on broadcast transmissions as well as when monitoring single-sideband and CW signals.

The unit requires 12-volts DC for operation. There is an optional power supply available from the manufacturer, or you can use any mobile power supply. The power jack is of the recessed coaxial variety, eliminating the possibility of a short circuit while plugging in the power supply. There is also an internal diode that is used to prevent damage if the power-supply connections are accidentally reversed. Although designed for 12-volt operation, in our tests (more on that later) we noticed that the unit operated at power inputs ranging from 8 to 16 volts without any noticeable degradation of the audio signal. Incidentally, all of the required interconnection cables are included with the unit.

The unit has an internal speaker. If you wish, an external speaker, earphones, or a recording device can be connected using the rear-panel-mounted external speaker jack. Other rear-panel jacks allow you to connect up to two receivers to the descrambler.

Our test

The notch-filter control allowed for considerable rejection (up to 60 dB at midrange) over the 800-3000-Hz frequency range. Tuning was sharp, permitting the removal of distracting tones without destroying the audio character of the speech. Incidently, the audio quality of the speaker seemed quite good; that may be because the rugged metal cabinet provides a favorable acoustical housing. The local oscillator for the speech descrambler was tunable from 2600 to 4000 Hz, defeating all types of single-inversion speechscrambling systems.

Quiescent current during idle (no input driving the unit) was measured at 55 mA. The current increased to approximately 200 mA during high-level audio transients. Since the audio circuitry is designed around a 2.5-watt LM380N, no appreciable heating or distortion was evident.

At normal volume levels, a 2N3904 twin-T oscillator injects a sine-wave tone into a Motorola MC1496P balancedmixer IC. The output is passed through the notch filter to the audio stage. Signals can be processed by the audio filter even when the unit is not in the descramble mode. allowing for greater flexibility.

The decoder sells for \$97.95, plus \$2.25 shipping, with the AC adaptor; \$89.95, plus \$2.25 shipping, without. It is available from Grove Enterprises, Brasstown, NC 28902. **R-E**

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NOLOGY TODAY

CAR TELEPHONES Cellular Technology Promises More Channels

Long waiting lists for phones and a lack of channels have limited the usefulness of the mobile-telephone system. A sophisticated new system is about to change all of that.

AN EXECUTIVE HEADS FOR HIS CAR parked in a downtown Chicago officebuilding garage, hoping to make it home to suburban Lake Forest in time to entertain dinner guests. Weighing heavily on his mind are sensitive negotiations taking place in California with which he must be in constant touch during his drive up the expressway.

Still in the garage, he dials the number of his Los Angeles contact on the pushbutton mobile phone, just as he would from his office upstairs. In a couple of seconds the phone in L.A. is ringing, and our busy executive starts his drive. During his hour-long mobile conversation, neither he nor his associate will notice that their full-duplex communications channel will have shifted frequency three times, with the mobile call going through three separate base-station sites.

The entire process is channelled through one of the most sophisticated land-based communications systems available to the general public, cellular mobile telephone. A cellular mobiletelephone system divides a metropolitan area into a mosaic of small cells. each with its own low-power transceiver, so that the same channel can be used for different conversations in nonadjacent cells. The result is more available channels for mobile-phone customers. To accomplish that, cellular telephone uses a remarkable system of computer-controlled radios that "hand off" a mobile call from one cell to the next as a vehicle travels across town. The hand-off is done so smoothly that the caller will probably not even notice it

In a conventional mobile-telephone setup, with a high-power base transmitter covering an entire metropolitan area, our executive would have just a 50% chance of getting an open channel at some time during his hour drive. If he did manage to get on the air. he would be occupying one of the 23 valuable channels in the Chicago area. (There are only 12 such channels in New York City!) But with the cellular system, there is a 99% chance that his call would go through on the first try. And, as he moves from one cell to another, our friend's telephone call shifts to a different channel on that other cell's transmitter/receiver, and the or ginal channel opens up for another caller.

Additionally, the cellular system makes portable telephones feasible.

This is not a science fiction dream, but a real system proven in two trials in congested areas: Chicago. sponsored by an AT&T subsidiary called AMPS (Advanced Mobile Phone Service), and Washington, D.C./Baltimore, sponsored by Motorola and operated by American Radio-Telephone Service Incorporated. Because of the success of those tests last year, the FCC, in a preliminary action, allocated two 20-MHz segments of the 800-MHz band for cellular-telephone communications. If final approval is granted (a decision was



FIG. 1—CELL-SITE BUILDING is dwarfed by the 150-foot antenna tower. The two antennas below the platform are used for reception; the antenna above the platform is used for transmission.

expected just after this issue went to press early last December), within two years the first fully operational systems could be in place; the system should be available in 70 cities within 5 or 6 years. When that happens, a cellular network will allow so many more people to have mobile phones in their automobiles and trucks, that the usual waiting list for those phones will all but disappear.

Basic cell systems

A cellular mobile phone system consists of a deceptively simple network of remote duplex transceivers. Cells are connected by landlines (telephone lines) to a central MTSO (*Mobile Telecom*munications *Switching Office*). It is at the MTSO that mobile calls are automatically patched into the regular telephone network, without the need for mobile operators.

A typical cell site, like AMPS' Lyons, Illinois site, is little more than a small building (about 20 feet by 20 feet at Lyons) to house the equipment racks, and a 150-foot tower for the antenna system (see Fig. 1). In more urban areas, cell sites can be located on building roofs.

The base station's receiving antennas are two vertical monopoles set up in a diversity reception mode. Two antennas are used in diversity reception so that a moving vehicle can maintain contact with the base station at all times. In that mode, the base-station's receiver is able to choose between the stronger of the two signals coming from the antennas; if a car is temporarily shielded from one antenna, the receiver simply

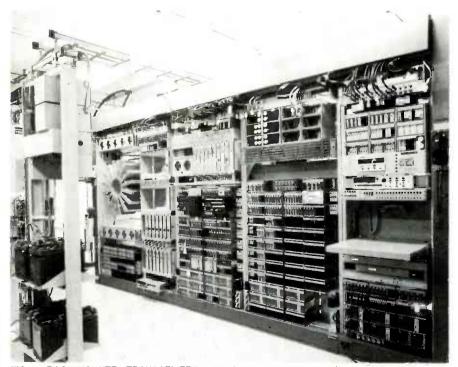


FIG. 2—RACK-MOUNTED TRANSCEIVERS and logic components are totally automated. Self-testing is done by the equipment on the right and the battery back-up system in the foreground is provided in case of a power failure.

switches over to the other.

The base station's transmit antenna, which is also a vertical monopole, is capable of handling up to eight transmissions on different frequencies at once. Since line losses at 800 MHz can be incredibly high with conventional cables, all antennas are fed with Heliax gas-filled coaxial cables that are maintained at a constant pressure.

Equipment in the small building is completely automated. One typical site that we visited recently with an AMPS official had two racks of transceivers, each rack with eight channels in operation (see Fig. 2). Small fans on the 45watt air-cooled amplifiers hummed quietly. Each cell site performs repeated self-testing of its equipment and reports irregularities to a central AMPS office. Only then is a service technician dispatched to a site. In addition, an outside transmitter tests the voice quality of each channel throughout the system by sending a 1-kHz tone and comparing the signal coming through the radios and phone lines to the original. That, too, is automatic, with a central printer logging the results of each test at approximately one-minute intervals.

In the event of power failures at the site, a wet-battery back-up power supply can operate the communications gear for about eight hours. Part of that back-up is also a power inverter to keep the automatic test gear in operation.

Mobile installations

Vehicle installations are quite simple affairs. A one-piece trunk-mounted unit is a combination 800-MHz duplex FMtransceiver and sophisticated logic unit. Installation in most vehicles takes about 41/2 to 5 hours, including complete testing. The vehicle's phone number is entered into the radio by way of a PROM chip that is programmed at the installation site. Because of the 45-MHz separation between the transmit and receive frequencies, cellular mobile phones use two 2.5-dB gain antennas at the car; one each for transmit and receive. The antennas are precisely tuned for the proper segment and can be either roof or trunk-lid mounted; roof-mounted antennas are 5/8 wavelength, while the longer trunk-lid antennas are 1/2 wavelength center-fed verticals.

Control heads vary somewhat with manufacturer. In the Chicago test, for example, mobile equipment was supplied by Motorola, Oki, and E.F. Johnson. Some heads have a *Touch-Tone* pad and resemble a standard desk phone (see Fig. 3). The deluxe model has a pushbutton pad and digital display built into the modern-looking angular handset.

Costs for cellular phone privileges have not been set, but the rates used in the Chicago test will give you an idea of how much is involved. Mobile equipment is leased for \$45-60 per month, depending on control-head style and options. For air time, there is a basic charge of \$25 per month, which includes 120 minutes of air time. Additional minutes are 25 cents each

The Chicago test divided are 2100square-mile metropolitan area into ten cells as shown in Fig. 4. While the downtown cell had the smallest geographical area, it also had the highest number of voice channels (26). Mobile units used 10-watt transmitters, but when the operational system is permanently set up, there will be more and smaller cells, allowing the mobile units to communicate comfortably using 3 watts. In the test, cells at opposite ends of the city used the same nine channels.

Growing cells

Cell sites are assigned a number of channels, based on the projected number of calls within that cell. More congested sites, such as the central business district of a major city, will have perhaps three or more times as many channels as an outlying cell. Because the base transmitters run at a low power-level of about 40 watts, the signal does not cover the entire metropolitan area—channels can be used by several cells as long as those cells are far enough apart. Motorola uses a four-cell cluster, while AT&T perfers a sevencell cluster before reusing a channel.

Moreover, as the number of users increases within a cell, or throughout the system, cells can be subdivided into smaller cells with lower-power transmitters and directional antennas allowing channels to be used even more times within the system. Theoretically, a cell can be as small as one mile in diameter. With the number of channels currently allocated to cellular telephones, a large city with 500 one-mile cells could serve a quarter of a million users!

Typically, a cell will go through three stages to accommodate greater traffic. In the first stage, a transceiver and omnidirectional antenna are centrally located within a cell. The second, or transitional stage, establishes additional cell sites, with directional antennas for some of the channels. The new cell sites will service a more localized area with a lower-power transmitter. More channels are switched over to the new cells until the original cell becomes a final-stage cellular system, consisting of a complete network of directional, low-power cell sites. In practice, the growing system will always be in a transition stage because service needs in any area will always be changing. And since the mobile units are capable of switching to any channel, as instructed by the MTSO, the user will be unaware of any of the changes in cell locations and frequencies.



FIG. 3—HUMP-MOUNTED CONTROL HEAD with handset and Touch-Tone keypad. A deluxe model (not shown here) also has a readout of the last number dialed.

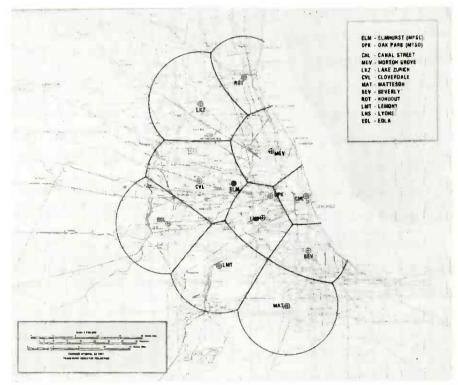


FIG. 4—THE CHICAGO TEST of the cellular mobile phone system divided the metropolitan area into 10 cells. Future networks will have smaller cells and use lower-power equipment.

How cellular phones work

The secret to the cellular system lies in two-way digital-data communications between car-and-cell, and cell-and-MTSO, plus one central computer at the MTSO keeping everything under control.

Perhaps the best way to describe all the intricacies of the system is to follow

a couple of telephone conversations from start to finish, with a cell hand-off added for good measure. This example is based on calls through the Chicago developmental system, operated since 1978 by Bell Laboratories and Illinois Bell Telephone, and used by about 2000 willing customers from a cross-section of professions and businesses.

You get in your car and turn on the engine as you always do. Then you switch on your deluxe telephone-control head mounted on the transmission hump. Using the Touch-Tone buttons, you key in your own three-digit security code that makes sure you're the only one making calls from your mobile phone. Nothing is coming through the speaker, but the logic unit of the trunkmounted transceiver scans through the control channels of all cell sites in your metro area and determines by Cell A's set-up channel's signal strength tht you are in Cell A. The set-up channel is a data-only frequency pair, and is a critical link in cellular telephone. Your radio is then kept on Cell A's set-up channel as long as the data signal there stays strong enough. At the cell site, there is a back-up setup channel transceiver if one should fail. The setup process takes less than a second.

As you drive along, you want to call your friend to see if he needs your help with a personal-computer problem he's been having. Without picking up the headset, you enter his telephone number on the *Touch-Tone* pad. An audible "beep" comes from the head's built-in speaker when each button is pushed, and the number appears in a display by the buttons. A quick glance at the display confirms you have properly entered the number, so you press the SEND button.

The following sequence takes place about as quickly as if you were calling your next door neighbor from your kitchen: The mobile logic unit sends out a data burst on the setup channel that identifies your radio, sends the number you're calling, and alerts Cell A that you have an outgoing call. Cell A's receiver passes the information over a landline to the central MTSO, where the Western Electric computer assigns an unused voice channel for the call. That frequency is passed back through Cell A to the mobile radio. The mobile transceiver obeys, and shifts frequency to the specified channel. When the MTSO senses the mobile unit's signal on that channel (the information is again relayed via Cell A's receivers), it places the call over the landline phone system.

The next sound you hear over the console speaker will be the ringing of your friend's phone. If the called number is busy, no problem. You will hear a busy signal, as you would expect. Simly press END, which alerts the MTSO to "hang up." Your channel opens immediately for other callers, as your radio shifts back to silently monitor the setup channel. Since that last number you dialed is still shown by the display, pressing SEND will automatically re-dial the call. The same instantaneous data exchange and frequency shift, though perhaps on a different channel, will take place.

Well, this time your friend answers the phone, and you become engaged in a long conversation. Unknown to you or him, your car is about to go from Cell A to Cell B. Here's what happens at a hand-off: The MTSO computer is busy monitoring the signal strength of every mobile signal coming through every cell site. It "sees" that your signal is starting to fall off from Cell A (though you don't realize it on your end) and it must find an open channel for you in Cell B. Since channels are not shared by adjacent cell sites, your hand-off involved a shift in both transmit and receive frequencies in addition to a change in cell sites.

The computer spots an open channel for you in Cell B. When your signal drops to a specified level, the MTSO sends a data burst (at a data rate of 10 megabytes-per-second) to your radio's logic unit via Cell A. In that one-quarter second, the audio is interrupted as the radio gets its instructions to shift channels—which it does without your even suspecting that brief "blank and burst" sequence.

When you hang up the handset, your radio goes back to the setup channel—but a different set-up channel, because you are now in Cell B. As you continue your drive, you decide that a quick call home may save you from having to go out later for milk. So you press a two-digit memory code to speed dial your home phone number, one of several frequently called numbers you previously stored. You find out that you are to pick up your children at a softball game on the way home.

At the field you find that the game is just about over, so you decide to park the car nearby and watch the last inning. Meanwhile, your wife finds that she does need milk, and wants to let you know before you get home. She dials your mobile phone number—a regular seven-digit number assigned to your telephone through an exchange located at a conventional (landline) telephone switching office.

The call is automatically sent to the central MTSO, whose job is to find you. It sends your identity code through every cell site in the service area on every set-up channel. Your radio, which you purposely left on, is constantly listening to the setup channel. When it hears your number, it tells Cell B via the setup channel that you are in Cell B. The MTSO then assigns a voice channel in Cell B for your wife's call. Your radio shifts to that channel, and your phone rings. Fortunately, you select the option that honks your horn when the phone rings.

As in that example, business users in the AMPS Chicago test found that the system saved them time and gas. A survey found that 84% of them were "very satisfied" with the service provided by the cellular telephone system.

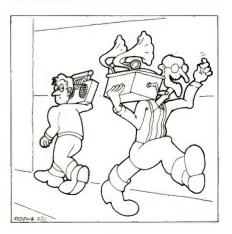
Our test

To test for ourselves the voice quality of the AMPS system, we took a ride in a car with a specially equipped mobile telephone. In addition to the telephone control head, that unit had a little box under the dash called a System Access Monitor, which gives a readout of the channel currently being used and has a signal-strength meter.

Local and long distance calls were as easy to make as if from a desk. Audio quality of the mobile phone was equal to any land-line, except when going through very low-elevation tunnels and extended viaducts. Even then, a couple of static crackles at the mobile end were at a very low level-perhaps inaudible to someone who did not know what he was listening for. None of those we called suspected we were talking on a mobile phone until we let them in on the secret. Communication was consistently flutter-free, even when the System Access Monitor meter showed that there were extreme changes in signal level.

The next test was to force a hand-off to an adjacent cell to detect the blankand-burst used to shift frequencies. As the channel readout changed on the System Access Monitor, we heard a slight electronic "click," and that was all. Again, we were listening for it with a tuned ear; otherwise we would have missed it. And since normal installations don't have the Monitor readout or meter, you would never know when to expect a hand-off—or when one had occurred.

As the demand for mobile communications increases, especially with the growth of microwave digital telecommunications, a system is needed that will make every megahertz count. Cellular mobile telephone is such a system, and it will soon provide many of us with high-technology, reliable, twoway voice communications that will be just as convenient as dialing your home or office telephone. **R-E**



RADIO-ELECTRONICS

BUILD THIS

ROBERT N. BEABER

THE 2716 EPROM (ERASABLE PROGRAMmable Read Only Memory)—currently available for ten dollars or less—is ideal for microprocessor-based projects. It permits storage of 2K (2048 bytes) of user program and operates from a single 5-volt supply. Because it can be erased by strong ultraviolet light, its contents can be changed if a modification must be made to the program it contains, or it can simply be erased for reuse in another project.

The programmer described here allows the project designer to program the 2716, using hexadecimal code, at each memory location by entering the required data with two thumbwheel switches and pressing a single PROGRAM switch. After programming a given memory location, the device automatically selects the next highest one. It also has a SINGLE STEP OF CONTINUOUS RUN option to provide access to address locations at a rate determined by the user. Those options provide the control needed to verify the data contained by the EPROM. Both the memory location addressed and its contents are indicated by 7-segment LED's.

Circuit operation

Figure 1 is a schematic of the EPROM programmer. Refer to it as we discuss how the circuit works.

The programming sequence begins when the PROGRAM switch, S3, is pressed. That switch is debounced by IC11-c, and the resulting pulse is used to clock one-shot IC7-b. The output of the one-shot enables NAND gate IC8-c and allows the 40-Hz oscillator—IC9-b —to clock counter IC14.

Initially, the counter's output is set at 0. As it is clocked by IC9-b its outputs go to a logic-high (+5 volts) state one at a time, in sequence. Note that only five of the ten possible outputs are used—pins 1, 4, 5, 7 and 11; they correspond to counts of 5, 2, 6, 3 and 9, respective-ly. As the outputs of IC14 change state.

The second secon

they control the actions of latches IC13-a and IC13-b.

When pin 4 of IC14 goes high, pin 1 of IC13-a is latched high. That, through buffer IC2-c, brings pin 20, CS (Chip Select—sometimes referred to as "data enable") of SO1, which contains the EPROM, high, and readies the EPROM for programming.

The output of IC14's pin 4 also turns on Q1 and Q2, which provide the programming voltage (25.5 volts) to pin 21 of SO1, and enables the three-state drivers, IC1 and IC2. The inputs of those two IC's are connected to S1 and S2, the BCD-output thumbwheel switches used to provide the programming data. The outputs of the two IC's are connected to pins 9-11 and 13-17 of SO1, which correspond to the eight data lines of the 2716 EPROM.

After 25 milliseconds—1/40-second, equal to one period of the 40-Hz clock —pin 7 of IC14 goes high and IC13-b is also latched high. That, in turn, takes pin 18 (CE/PGM—CHIP ENABLE/PROGRAM) of SO1 high and the data from the thumbwheel switches is programmed into the EPROM. The address programmed usually starts with 0, but can be selected by S4. (SINGLE STEP), and S5 (RUN).

When 50 milliseconds have passed, pin 1 of IC14 goes high. That resets IC13-b, removing the logic-high from pin 18 of the EPROM. After another 25 milliseconds, pin 5 of IC14 goes high, resetting IC13-a. That removes the CS voltage, removes the 25.5 volts from pin 21 by turning Q1 and Q2 off, and disables IC1 and IC2, thus removing the thumbwheel-switch data from the EPROM's data lines.

When pin 11 of IC14 goes high, it resets one-shot IC7-b, disabling NAND gate IC8-c and cutting IC14 off from the 40-Hz clock. Pin 11's going high also resets IC14 to 0 and advances counter IC3 to the next address to be programmed.

The remainder of the logic circuitry provides RESET, SINGLE STEP. and RUN functions.

Switch S6 (RESET) is used to reset the cascaded address counters IC3. IC4, and IC5. to zero. Switch S4 provides SINGLE STEP capability—every time the switch is pressed. IC7-a, a one-shot, supplies a pulse to counter IC3, causing the memory location being addressed

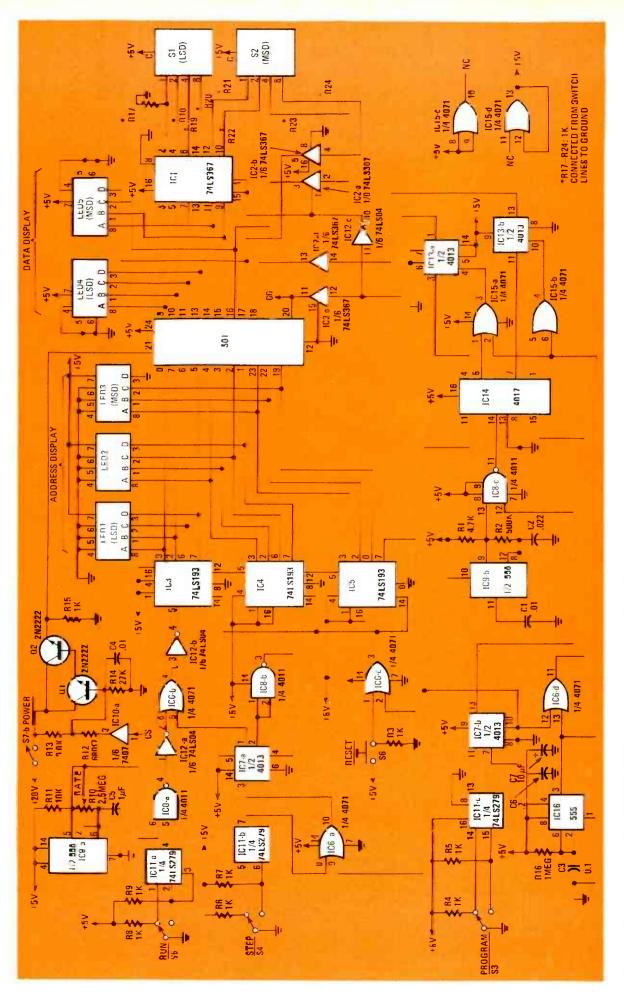


FIG. 1—WHILE ONLY R17 IS SHOWN, resistors R17-R24 are connected from the eight output-lines of S1 and S2 to ground.

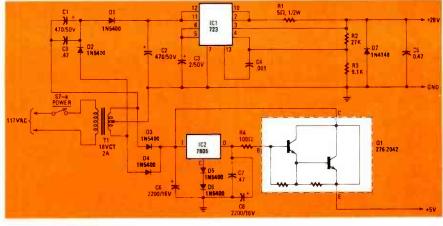


FIG. 2—POWER SUPPLY provides both the 5-volts needed for the logic circuitry and the 25.5-volts used to program the EPROM.

to be incremented by one. Finally, S5 (RUN) allows the variable-frequency oscillator formed by IC9-a and controlled by R10 to clock counter IC3 at a rate determined by the user.

The three switches provide the user with several means for accessing EPROM memory addresses to enter or verify data. The LED displays on the address and data lines provide a continuous indication of the memory address being accessed and of its contents.

Power supply

The power supply shown in Fig. 2 provides all the voltages required by the programmer. Power switch S7 has two sections—one for 117-volts AC (Fig. 2) and one for 28-volts DC (Fig. 1). That is necessary because when the AC voltage is cut off by S7-a, an induced current in the secondary of the transformer will cause a pulse to appear on the 28-volt DC line. Since that pulse can be transmitted to pin 21 (V_{pp}) of the EPROM—a pin used in the program-

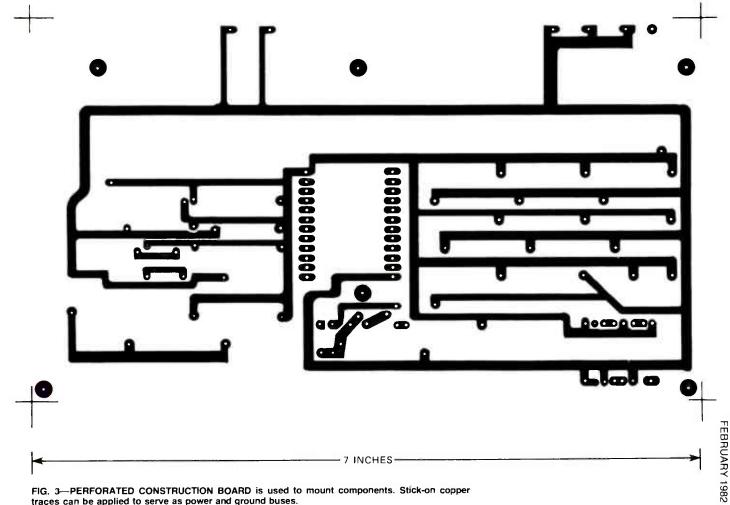
PARTS LIST—POWER SUPPLY

All resistors 5%, ¼-watt unless otherwise specified R1-5 ohms, 1/2-watt R2-27,000 ohms R3-9100 ohms R4-100 ohms Capacitors C1, C2-470 µF. 50 volts, electrolytic C3-2 µF. 50 volts, electrolytic C4-.001 µF, ceramic disc C5-0.47 µF, ceramic disc C6, C8-2200 µF, 16 volts, electrolytic C7-0.1 µF, ceramic disc Semiconductors IC1—LM723N adjustable voltage regulator IC2—7805 5-volt regulator Q1-high-power Darlington, NPN-type (Radio Shack 276-2042 or similar) D1-D6-1N5400, 3 amps, 50 PIV T1-18 volts, center-tapped, 2 amps Miscellaneous: TO-3 heat sink, PC board or perforated construction board, 14pin IC socket, hardware, etc.

ming process—erroneous data may be entered into the IC. Therefore, S7-b is used to cut off the 28-volt supply at the same time the main 117-volt supply is turned off.

Construction

The 2716 EPROM programmer can be constructed on perforated construction



CIRCUITS

REMOTE VOLUME ATTENUATOR

This IC can be used to make a great remote volume-control. But what's even better is that the device can be operated by either analog or digital means.

MARTIN BRADLEY WEINSTEIN

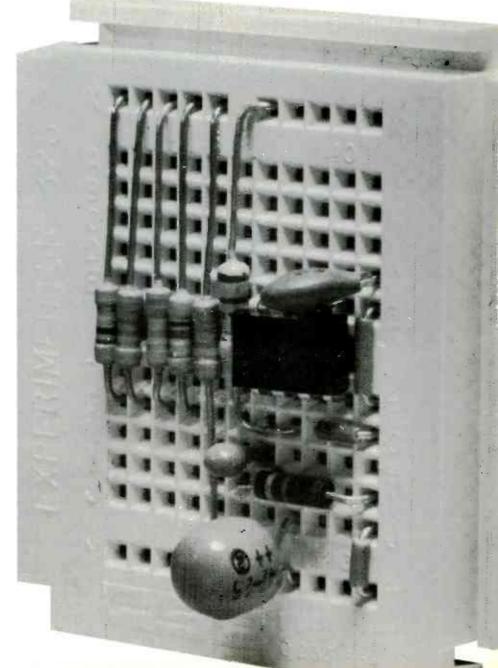
THERE'S AN INTEGRATED CIRCUIT AVAILable from Motorola, the MC3340P, capable of effectively attenuating an audio signal by 90 dB or introducing up to 13 dB of gain (or anything in between) while exhibiting a total harmonic distortion of about 0.6%. Attenuation can be controlled by either a DC voltage (between 3.5 and 6.0 volts) or a resistance (0-33 kilohms) applied to its control pin and referenced to ground. Suggested for use with approximately 100-millivolt RMS input-signals, it can be used with signals up to 0.5-volt RMS. The supply-voltage range is wide—9-18 volts DC. And it's priced in the \$1.00-\$1.50 range.

This little wonder can accomplish a number of remarkable wired-remotecontrol functions. For example, a radio receiver in the trunk of your car feeding a rear-deck speaker can be easily modified by using the circuit shown in Fig. 1 in place of its volume control; a single wire to a dash-mounted pot then provides complete control of its volume. But there's an even better way to use this device.

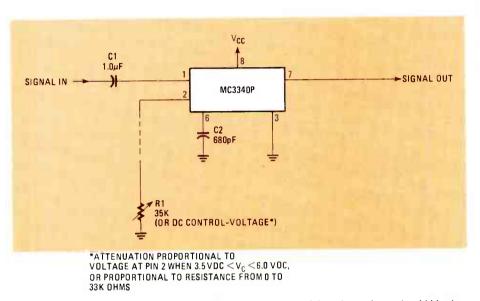
Digital control

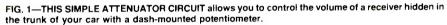
The circuit shown in Fig. 2 gives you complete control over attenuation using a set of resistors arranged in a network. Grounding any combination of four (or fewer) lines (labelled A, B, C. and D) provides a specific degree of attenuation.

Getting the circuit to perform properly is harder than it appears. The control characteristics of the MC3340P are far from linear. Another complication is getting a reasonably linear analog control characteristic starting with binary values and off-the-shelf resistors.



PARTS LIST





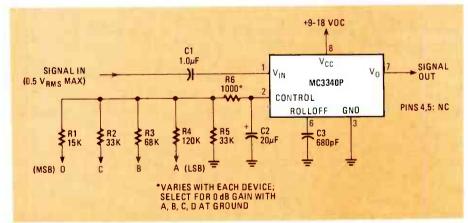


FIG. 2—SCHEMATIC DIAGRAM of a digitally-controlled attenuator circuit. Lines A, B, C, and D, can be driven by the simple circuit shown in Fig. 4.

The technique required is common to a number of digital-to-analog conversion schemes. You can get a fairly linear response by using resistors with a 1-2-4-8 (powers-of-two) relationship. But there are no combinations of standard resistors with that relationship, although the 15K-33K-68K-120K-ohm combination used here (for R1, R2, R3, and R4) comes very close. Trial-and-error had a great deal to do with establishing the final component values, but there were also a few educated guesses involved. The parallel combination of R1, R2, R3, and R4, for example, is itself in parallel with a 33K resistor (R5) because that value is the maximum control resistance and corresponds to the maximum attenuation of the IC. It took a very long night with
 Resistors ¼ watt, 5% or better

 R1—15,000 ohms

 R2, R5—33,000 ohms

 R3—68,000 ohms

 R4—120,000 ohms

 R6—5100 ohms (see text)

 Capacitors

 C1—1.0 μF

 C2—20 μF, electrolytic

 C3—680 pF

 Semiconductors

 IC1—MC3340P (Motorola)

TABLE 1

D	С	в	Α	Net resistance
0	Ø	Ø	Ø	7650Ω
Ø	0	Ø	1	8040Ω
Ø	Ø	1	Ø	8370 Ω
Ø	Ø	1	1	8860Ω
0	1	Ø	Ø	9330Ω
Ø	1	0	1	9950 Ω
Ø	1	1	Ø	10500 Ω
0	1	1	1	11310Ω
1	Ø	Ø	Ø	12960 <u>Ω</u>
1	Ø	0	1	14280Ω
1	Ø	1	Ø	15510Ω
1	Ø	1	1	17500Ω
1	1	Ø	Ø	19750Ω
1	1	Ø	1	23220Ω
1	1	1	Ø	26880Ω
1	1	1	4	34000Ω

a scientific calculator to double-check the calculated performance of the circuit, and that's why R6 is there—to adjust the performance so that it is slightly more linear. Table 1 gives the calculated net resistance for each control code and Fig. 3 plots both net control-resistance and calculated attenuation values against the binary control codes.

Unfortunately, errors can occur almost anywhere in this design. For one thing, the manufacturer's curves, while accurate enough when you're controlling the IC with a potentiometer, aren't very precise as far as determining actual attenuation vs. resistance performance when you're using fixed values. Another problem is that even 5%-tolerance resistors can vary enough to cause jarring nonlinearities in their combined performance in this network. For example, one combination can easily yield a lower net resistance than another combination representing a higher binary value. The solution is either expensive precision resistors

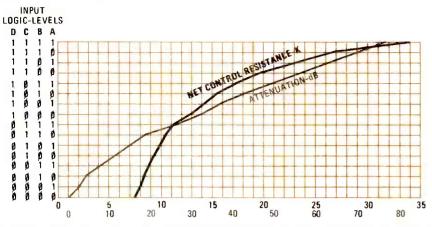


FIG. 3—NET CONTROL RESISTANCE and calculated attenuation plotted against four-bit control signals. Note that the calculated attenuation curve is nearly, but not quite, linear.

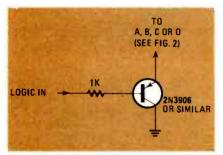


FIG. 4—THIS CIRCUIT can be used to drive a digital volume attenuator's control lines. Almost any PNP transistor can be used.

or—as you'll probably choose—handselection of 5% resistors.

Even with selected resistors, the performance of the breadboarded circuit was not as predicted by the graph in Fig. 3 when input and output signals were measured and actual attenuation calculated. The problem turned out to be R6: the value I had chosen was too small and the circuit had some gain when it should have had 0-dB attenuation. That made the whole system's performance nonlinear. Increasing R6 to 5100 ohms helped performance significantly, but note that the value of R6 will vary from device to device. You wouldn't think a circuit this simple could have caused so many problems. But the end result was worthwhile.

Driving the control lines

There are several ways to provide the digital control-signals the circuit requires. Most computers and digital control-systems can easily drive a PNP transistor (Fig. 4), possibly using a latch. A hexadecimal thumbwheel switch (or any 4-bit binary-encoded switch) can be used without drivers. A four-channel optical system could be used where isolation was required.

A SIPO (Serial In, Parallel Out) shift register could also be used. A serial Baudot or ASCII code, for example, following a specific control-character on a teletypewriter circuit could raise the volume of an alert tone-generator or monitor receiver in a network or a remote-control application.

Perhaps more exciting is the prospect of full keyboard control of a communications system, including volume levels, with or without the assistance of a computer. A single dashboard-mounted keypad could select synthesizer frequencies or channels, set volume and squelch levels, adjust mike gain, and more—even for several transceivers. All the valuable hardware could then be secured in the trunk.

Most exciting of all is the fact that this simple circuit makes computer coordination of our living and working environments possible. With the attenuators installed and the proper control bussing routed, the "turn-it-up" and "turn-it-down" nuisances in life might be eliminated. If the phone should ring, for example, the computer might use the circuit to turn down the volume automatically on the TV or stereo. If your clock radio wasn't quite rousing you, the computer could make it louder; or at night, it could fade the music down as you fell asleep. You could also program maximum volume levels for any number of devices and vary them with the time of day-to let the kids blast the stereo from, say, 4 to 5:30, but to keep it under control at other times. (The prospects for fading in romantic mood-music are also promising.)

Basically, this circuit is just an elaborate volume control (although the MC3340P responds very well to signals up to 600 kHz, and fairly well through 2 or 3 MHz)—a potentiometer with "bells" and "whistles." I'm sure you can think of quite a few places to use something like this. **R-E**

SERVICE QUESTIONS

TUBE BURNOUT

In a GE M704 (a Canadian model) the horizontal-output tube went out too often. I wrote to you and you said that the problem could be caused by the failure of the horizontal oscillator to start, resulting in overload of the output stage. You suggested that I change the capacitors across the oscillator coil. I did, and that did the trick. Thanks.— D.H., London, Ontario

SNOW, NO AGC

I've got a CTC-43A with far too much snow on all channels. The RF AGC control has no effect at all. The plate of the 2DS4 RF-amp tube reads +250 volts and

the grid reads -30 volts. The 16K resistor isn't shorted, and a new tube didn't help. ---P.B., Ewa Beach, HI

I think you've got your finger on it. Note that the plate voltage on the RF tube is the same as the supply voltage— +250 volts. You're not drawing any current. That must be due to the excessive grid voltage. The normal RF-AGC voltage with no signal should be only +0.79 volts. Check the AGC keyer collector; also the 10-megohm resistor that feeds it from the +270-volt supply. Also check the 2.2-megohm resistor between that point and the tuner.

I'd be inclined to suspect a leaky AGC transistor. Check to see whether the IF-AGC voltage works. If so, something is wrong in the resistors in the RF-AGC circuit and I am inclined to suspect the 10-megohm one (we've had problems with that resistor changing value before).

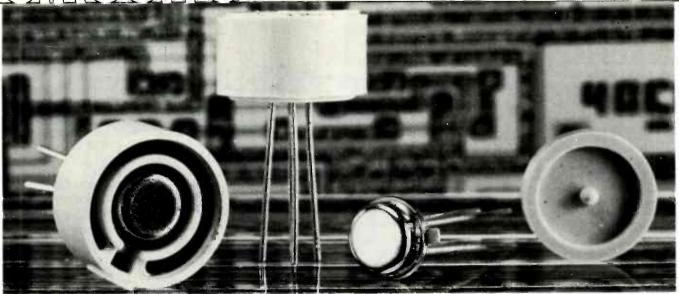
DAMPER DIODES

This RCA EP454W portable works fine for a while, then the horizontal-output tube gets red hot and blows the damper diode. I suspected the oscillator stopping, but every time I check, it's running. I am totally confused, do you have any ideas? —J.G., Mentone, IN

I think you've got one thing reversed. Probably, the damper diode shorts and *that* blows the tube. Check the solidstate damper. If it's a single unit, Part No. 125932, it's the original. That part has been known to cause problems. RCA recommends replacing it with a dual "piggy-back" type, Part No. 120818.

By the way, the parts list you're using is incorrect. It recommends using a GE-504 as a replacement. That's a plain sine-wave type, but you must use a fastrecovery type here and in all horizontalfrequency circuits. **R-E**

CIRCUITS



Precision Voltage References

An electronic instrument or circuit can only be as accurate as the reference used to calibrate it. Build your own low-cost, high-precision voltage references using the information presented here.

PRECISION VOLTAGE-REFERENCE CIRcuits are constantly needed by anyone who works with electronics. They are needed in an electronics workshop, for example, to calibrate voltmeters and oscilloscopes. In the laboratory, they can be used to troubleshoot, test, and calibrate instruments such as pHmeters. Computer applications include A/D and D/A converters; most of the popular circuits that are used for those applications require an external voltage reference to make them reasonably accurate. In addition, projects such as electronic thermometers, and anything based on the Wheatstone bridge (and that includes quite a few electronics instruments) will require a precision voltage-reference.

Commercially available voltage references can be very expensive, but we can make references that are precise enough for all but the most demanding applications for a small cost. The technology that is available today allows us to do things that a hobbyist could only dream about a decade ago. In this article, we will discuss some of the more common precision voltage-references, and offer both finished projects and design notes for those who would

JOSEPH CARR

like to "roll their own."

Why does a hobbyist need a precision voltage-reference? Let's say, for example, that you want to build an A/D or D/A converter. We all know that the greater the bit-length of those converters, the better the accuracy and resolution of the measurements made. or voltage produced. But no data converter is better than the analog voltagereference used to make the conversion. All data converters produce an output that is the product of a binary word and an analog reference voltage (or current, in some cases). In practical terms, that means that an error in the voltage reference will foul up the accuracy of the conversion.

Let's consider two data converters that a hobbyist might use: 8-bit and 10bit D/A converters. The resolution of the 8-bit converter is 0.391%; the resolution of the 10-bit converter is 0.098%. Those figures are the maximum possible resolution based only on bit-length. If the reference voltage for both is 10.00 volts, then the actual reference voltage must fall within the range of 10.00 volts \pm 391 mV in the 8-bit case and 10.00 volts \pm 98 mV in the 10-bit case. A 12bit D/A converter, which is now available at relatively low cost, requires an analog reference of 10.00 volts ± 24 mV.

Clearly, as the bit length of the converter increases, the voltage reference must become more precise. Reference voltage sources for 8-, 10-, and 12-bit D/A converters are easily constructed at a low cost using components available today.

Zener diodes

One of the most important voltage references used in electronics is the Zener diode. Figure 1 shows the basic circuit for a Zener-diode regulator; a graph of the voltage-versus-current characteristic for a typical Zener diode is shown in Fig. 2.

The Zener diode acts much like an ordinary PN-junction diode as long as the applied voltage is less than a specified value of reverse bias (V_Z) , or when the diode is forward biased. In the range between V_Z , and a forward bias of approximately 700 mV (the actual figure is usually 600 to 700 mV), the current through the diode is a small reverse current called leakage. The diode is essentially turned off in that region. When the voltage exceeds a forward bias potential of 700 mV, the

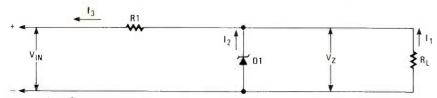


FIG. 1—THE ZENER DIODE is often used as a voltage reference or as a simple voltage regulator. A basic circuit for a Zener diode regulator is shown here.

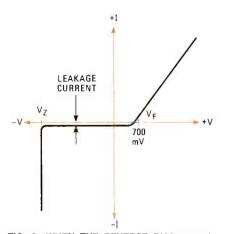


FIG. 2—WHEN THE REVERSE BIAS exceeds V_Z the negative current through the Zener increases rapidly.

positive current increases in the same manner as in an ordinary PN junction. The Zener acts quite differently, however, when the reverse bias exceeds V_Z . When that voltage is reached, the negative current rapidly increases. No matter how much more negative the applied voltage becomes, the voltage across the diode remains relatively constant.

But the Zener diode is not the best voltage reference for many applications. The Zener's voltage rating is nominal, not absolute. When the package says "6.2 volts." it may mean some potential in the neighborhood of 6.2 volts. It is also true that the Zener's voltage tends to change a little bit with changes in temperature—hardly something that you would want a voltage reference to do!

Figure 3 shows a circuit that overcomes the temperature-drift problem of Zener diodes. Two Zeners are used in that circuit; the output is the difference between their two V_z potentials. The

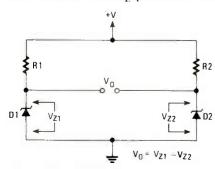


FIG. 3—THIS CIRCUIT shows one way to overcome the problem of temperature drift in Zener diodes.

circuit works reasonably well as long as both Zener diodes are kept in the same thermal environment. It is common practice to bond the two diodes together and coat them with silicone grease in order to keep them at the same temperature.

Several manufacturers once offered four-terminal "Zener diode" packages that used four to eight Zeners, in two series circuits, to form the output potential. Those devices were arranged like the circuit in Fig. 3, except for the number of diodes and the fact that all were on a common substrate, thus tracking each other in temperature.

Op-amp voltage-reference sources

One problem with Zener references, even precision Zener's that have been trimmed to a specific, accurate voltage, is that they are limited to values selected by the diode manufacturer—often not the voltage required in some particular application. We can, however, use an op-amp circuit to scale the voltage to a needed value. Such a circuit will also keep the load across the Zener at a constant, low level.

Almost any op-amp will do for that type of circuit, but it is generally best to use a premium-grade type. Some of the best are BiMOS and BiFET op-amps, such as the RCA CA3140 device.

The basic properties of an operational amplifier tell us that applying a voltage to one input will cause a voltage to appear at the output. In the circuit of Fig. 4 we apply the voltage from a Zener diode to the non-inverting input of the op-amp. That means that the output voltage will have a value equal to that voltage times the gain of the operational amplifier, or:

 $V_0 = V_Z[(R2 + R3)/R1 + 1]$

If we make the gain larger than one, then the output voltage will be greater than the Zener voltage. Similarly, if we make the gain one (i.e. if we short out R2 and R3, causing the output to be connected directly to the inverting input), then the output voltage is the same as the Zener voltage.

We can trim the value of the output voltage by making the feedback resistance variable. To keep tight control over the feedback resistance, a series combination of a fixed resistor and a potentiometer (R2 and R3) is used. It is generally good practice to make the resistance of potentiometer R3 approximately 10 percent of the total resistance (R2 + R3), and to make it a 10-, or 15-turn trimmer pot.

Special devices

In the past couple of years, semiconductor manufacturers have brought out several different types of voltage references. Some of those are two-terminal devices that act like diodes in the circuit, even though they are complete integrated circuits in themselves.

An example of such a device is the LM199 from National Semiconductor (2900 Semiconductor Dr., Santa Clara, CA 95051); the functional block diagram of that IC is shown in Fig. 5. That reference is a four-terminal device that combines a 6.95-volt Zener diode with an on-chip heater element. The Zener diode is maintained at the same temperature as the heater because it is buried on the same I.C. die (silicon chip) as the heater element.

The heater is not an actual resistance element, but is a Class-A amplifier with the input shorted. One property of such an amplifier is that it will dissipate a constant power, so that the amplifier will maintain a constant heat level after the die comes to equilibrium.

Burying the Zener in the same die as the heater element has many benefits including lower noise operation and thermal stability. Ordinary discrete Zener diodes have a drift specification on the order of 5 mV/°C, while for the LM199 devices the drift is specified in terms of microvolts per degree.

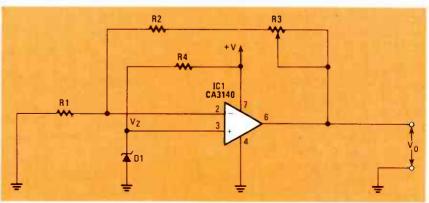


FIG. 4—AN OPERATIONAL AMPLIFIER can be used to scale the voltage from the Zener diode to whatever value is needed.

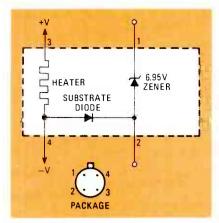


FIG. 5—A ZENER DIODE and an on-chip heater element are combined in the LM-199. The resulting IC is a highly stable voltage-reference device.

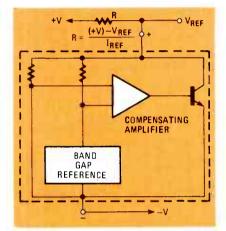


FIG. 6—THIS TWO-TERMINAL REGULATOR from Ferranti Electric can be used just like a Zener diode. It has a stability as great as 10 ppm per 1000 hours of operation.

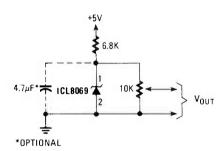


FIG. 7—A 1.2-VOLT temperature-compensated voltage reference, the Intersil ICL8079 used in this circuit is well suited to high-speed A/D and D/A converter projects.

The LM199 is rated to have an initial operating voltage of ± 2 percent of the rated nominal voltage, and the device is very stable over periods of time. The long-term stability is rated at 20 ppm, while the short-term stability is 1 ppm.

In most cases, the LM199 device is operated with a potential of 9- to 40volts DC across the heater terminals (pins 3 and 4), while the Zener is hooked up in a circuit in the normal manner.

Ferranti Electric (87 Modular Ave., Commack, NY 11725) offers a twoterminal regulator. Those devices can be used in the same manner as ordinary Zener diodes, but they use band-gap technology to produce a stable output voltage. Ferranti offers versions with output voltages of either 2.45 or 1.26 volts. The internal circuitry for the different types is essentially the same, and is shown in Fig. 6.

The Ferranti band-gap devices are available in five versions: ZN404, ZN458, ZN458A, ZN458B, and ZN423. The first four of those are 2.45-volt reference devices, while the last is a 1.26volt reference device. All of the devices except the ZN423 are capable of sinking anywhere from 2- to 120-mA DC; the ZN423 can sink 1.5- to 12-mA DC.

The long-term stability of the Ferranti devices is 10 ppm per 1000 hours. Temperature coefficients range from 30 ppm to 200 ppm, depending on the device.

As in ordinary Zener-diode operation, a single series current-limiting resistor is used with the Ferranti regulators (see Fig. 6 for the formula to calculate the value of that resistor). For example, using a 2.45-volt regulator (V_{REF} =2.45), a +5-volt DC power supply (+V=5), and a current of 3.75 mA (I_{REF} =.00375), the resistance required would be 680 ohms.

The Intersil (10710 N. Tantau Ave., Cupertino, CA 95014) ICL8069 diode shown in Fig. 7 is another of a band-gap reference source. That device is a 1.2volt temperature-compensated voltage reference that features very low-noise operation. As a result, it is suited well to applications in high-speed data conversion projects. at Zener currents up to 50 microamperes.

Various versions of that device are made with differing temperature coefficients and temperature ranges. Of course, the selection of a particular device depends upon the nature of the application it is to be used for. A device that would be suitable for an 8-bit A/D converter would not be suitable for a 12-bit converter.

Two applications for the ICL8069 are shown in Figs. 8 and 9. The application shown in Fig. 8 is similar to the operational amplifier circuit shown earlier. That circuit uses the ICL8069 in the feedback loop of a premium-grade LM108 op-amp. A potentiometer is used to trim the output to exactly 10.00 volts. That circuit can be used in a wide variety of applications. The circuit in Fig. 9 uses the ICL8069 device in a data-converter application. While an Intersil ICL7107 is used in that specific circuit, the ICL8069 can be used with any similar device whose output voltage is within the operating range of the diode.

Precision Monolithics, Inc. (1500 Space Park Dr., Santa Clara, CA 95050) offers two different IC voltage references, the REF-01 and REF-02. The REF-01 is a 10.00-volt device. It is housed in an eight-pin TO-99 case

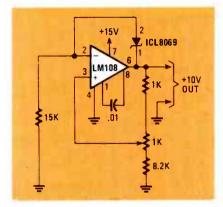


FIG. 8—A ICL8069 IS PLACED in the feedback loop of an LM108 op-amp in this voltage reference circuit. The potentiometer is used to trim the output to exactly 10.00 volts.

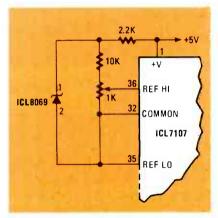


FIG. 9--USED HERE in an A/D converter circuit, the ICL8069 is well suited for such an application.

although just four of the pins are used. The REF-02 is a 5.00-volt device. It is also housed in a TO-99 case, but in this instance five of the pins are used. The fifth pin is a temperature-dependent terminal so that the REF-02 can be used as a temperature sensor.

The simple circuit shown in Fig. 10 contains a potentiometer that lets you set the output of either the REF-01 or REF-02 to exactly 10.00 volts. Either device can be used in a number of A/D or D/A converter circuits.

Selectable output supplies

All of the circuits presented thus far supply just a single output voltage. Even in the devices and circuits that use a potentiometer, it is used to adjust the output to a single value and then left alone. Sometimes however, a multipleoutput voltage reference is needed. An example of that might be a voltage calibrator used with oscilloscopes and voltmeters.

The simplest way to make a multipleoutput voltage reference is to place a resistance network between the output of one of the circuits presented in this article and the load. A typical network is shown in Fig. 11. In that network, each tap is a decade higher in voltage than the next lower tap. The lowest value shown is 10 ohms, but that might

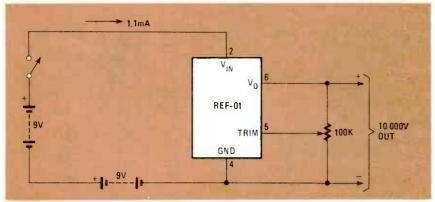


FIG. 10—A SIMPLE POTENTIOMETER is used to trim the output of the PMI REF-01 voltage reference to exactly 10.00 volts.

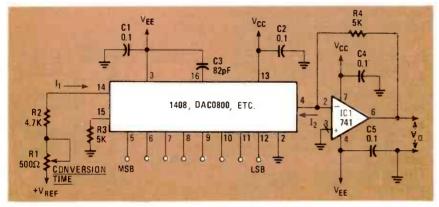


FIG. 12—A DIGITALLY SELECTABLE voltage reference. Once the input current (I_1) to the D/A converter is set, the output from the circuit (V_0) is determined by digital input (pins 5-12) to the D/A converter.

be too low for some applications. In that case, make the lowest value 100 ohms, and increase each of the other resistances by a factor of ten (i.e. make the sequence 100-900-9000-90,000-900,000 etc.). With the resistances used in Fig. 11, the voltages found at the taps would be as follows:

A-10.00 volts B-1.00 volts C-100 millivolts D-0 volts We can also make a digitally selectable reference source by using a D/A converter. Figure 12 shows how such a converter can be used to control the output voltage (V_0) of a reference supply.

The output current of any D/A converter is the product of the input current (I_1) and the binary word input, divided by the maximum possible value of the binary word:

 $I_2 = I_1 \times (N/2^n)$ Where: I_1 is the reference current.

Solid State News

Power FET's data sheets

Power MOS Field-Effect Transistors is the title of a 30-page booklet from Motorola. It includes data sheets, crossreferences, and operating theory for the company's enhancement-mode Nchannel silicon-gate TMOS FET.— Motorola Semicondutor Products, Box 20912, Phoenix, AZ 85036.

RCA COS/MOS product guide

COS/MOS Digital Integrated Circuits (COS-278H) is a 30-page product guide written to aid the user in selecting the optimum types of devices for specific applications. It provides maximum ratings, recommended operating conditions, static electrical-characteristics, selection charts, and functional diagrams for the complete line of RCA A-Series (15-volts maximum rating) and B-Series (20-volts maximum rating) COS/MOS devices. Devices covered include gates, buffers. inverters. encoders, decoders, multivibrators, flip-flops, and triggers. Also included are IC's intended for use in microprocessor, memory, and timekeeping circuits.—RCA, Solid State Division, PO Box 3200, Somerville, NJ 08876.

RF diodes catalog

RF Signal-Processing Diodes Selector Guide is a 6-page listing of all Motorola tuning diodes for frequency control, hot-carrier diodes for mixing and detec-

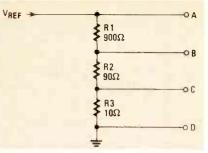


FIG. 11—A RESISTOR NETWORK can be used to get a multiple-output voltage reference. The voltage at each tap is 10 times higher than the voltage at the tap below it.

 I_2 is the output current. N is the decimal value of the binary word input, and n is the bit-length of the D/A converter.

Try this problem out as an example: Find the output current (I₂) when V_{ref} is ± 10.00 volts, and R1 is adjusted so that I₁ is exactly 2 mA. Assume that the binary word applied to the putput is 10010110 (i.e. decimal 150), and that the D/A converter is 8-bits (as in Fig. 12).

The solution would be found as follows:

 $l_2 = (2mA) (150/2^8)$ $l_2 = (2mA) (150/256)$ $l_2 = (2mA) (0.586)$ $l_2 = 1.17 mA$

The output voltage from the circuit (V_0) in Fig. 12 is the product of the D/A converter current and the feedback resistors, so:

- $V_0 = (I_2)$ (R4)
- $V_0 = (1.17 \text{ mA}) (5.00 \text{ kilohms})$
- V₀=5.895-volts DC

Once the reference current I_1 is set, the output from the D/A converter is determined by the digital input to the converter. It is a simple matter to vary that input to give us precisely the reference voltage that is needed. There are many ways to do that, including connecting the D/A converter's digital input to the output port of the microcomputer or to binary-output thumbwheel switches. **R-E**

tion, and PIN diodes for switching. Included are design curves and package outlines.—Motorola Semiconductor Products, PO Box 20912, Phoenix, AZ 85036.

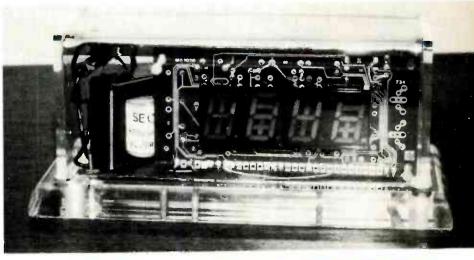
RF transistor catalog

Transistor Designers 1981 Catalog is a 159-page VHF/UHF/microwave transistor selection-guide. Included are detailed data on low-noise, general-purpose, and linear power silicon-transistors. and low-noise and linear power GaAs FET's for use between 60 MHz and 18 GHz. Charts show performance curves and typical contours for the constant-gain and noise-figure functions. Some devices are available both packaged and as unmounted chips.—Avantek, 3175 Bowers Ave., Santa Clara, CA 95051. R-E

BUILD THIS

AN EYE-CATCHING AND USEFUL PROJECT need not cost a great deal of money. The device described here is a combination digital clock/outdoor thermometer; the total cost of project as shown is less than fifty dollars.

To make things more interesting visually, the project was housed in a clear case—an acrylic butter dish—so that the entire circuit can be seen. Three reed switches are used to set the time or turn on the thermometer; that is done by bringing a magnet near the case, close to where the appropriate switch is located. Although standard subminiature panel mounted switches can be used, we choose the reed switches because it simplifies the construction and gives the completed



THERMOMETER

project a modern appearance. As a thermometer, the project displays temperature in either °C or °F over a range of -40°C to +90°C or -40°F to +193°F. The temperature sensor does not require a separate power supply and is not affected by lead length.

About the circuit

The heart of the circuit is a National Semiconductor MA1026 digital LED alarm clock/thermometer module as is shown in Fig. 1. That module features a 0.7-inch LED display, snooze alarm, 24-hour alarm, 12/24 hour mode, 50/60-Hz select, fixed/flashing colon, and sleep cycle. The sleep cycle can be used, with external circuitry, to turn off a radio or appliance after a time period of 59 minutes or less.

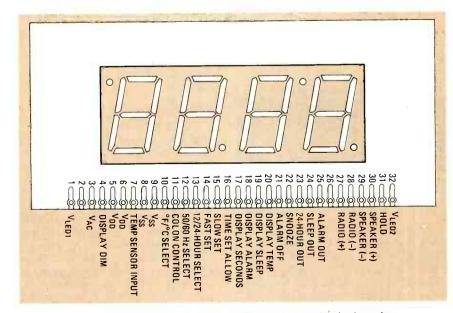
For this project, the module can be treated as a "black box," requiring just a transformer for input power, switches for display selection, and sensors for temperature measurement. The transformer has three windings-a 120-volt primary, a 6-volt center-tapped secondary (125-milliampere current-handling capability on either side of the tap), and a 10.5-volt secondary capable of supplying 25 milliamperes. Power consumed by the module is minimal, therefore the power-handling capacity of the switches is not important. An LM334 adjustable current source is used as the temperature sensor; that three-legged device looks like a small plastic transistor. A 220-ohm resistor, R1, is used to scale the output of the LM334 so that the IC will supply a current of one microampere-per-°C. Because the output from the sensor is a current, lead length is not critical. Note that a low temperature-coefficient resistor should Know the time and temperature instantly with this easy-tobuild project. The clear, acrylic case makes it an interesting conversation piece.

MICHAEL RIGSBY

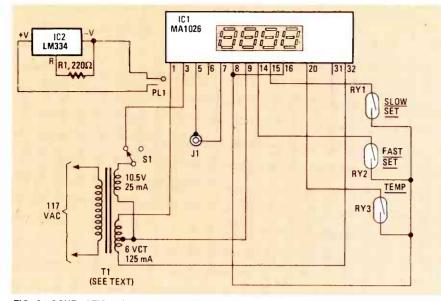
be used for R1 (either a metal-film or, if you can find it, wire-wound unit), and that it should be part of the sensor package so that it will change temperature with the IC. To make the sensor package, mount the LM334 and the resistor on a small piece of perforated construction board.

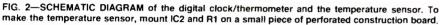
In the basic clock/thermometer shown in Fig. 2, pin 8 is tied to pin 16, letting you set the time using the FAST SET and SLOW SET switches. Closing the TEMP switch puts the device in the "thermometer mode." In that mode, if pin 10 on the clock module is left unconnected, the temperature readings will be in $^{\circ}F$; if pin 10 were tied to pin 8, the temperature readings would be in $^{\circ}C$.

The basic clock/thermometer can also be adapted for battery operation. If you want to do that, the schematic shown in Fig. 3 should be used. In that circuit, a



FIG, 1—THE COMPLETE PIN-OUT for the National MA1026 clock module is shown here.





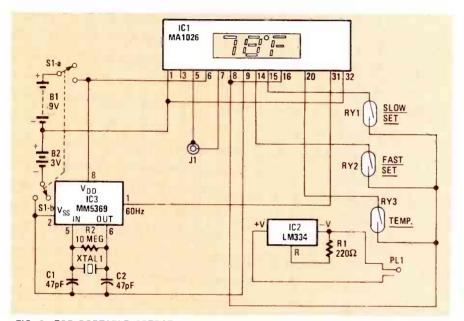


FIG. 3—FOR PORTABLE OPERATION, build this version of the clock/thermometer project. The oscillator circuit draws just 1.5 milliamperes,



FIG. 4—THIS IS HOW the project will look when it is finished. The clear case—an acrylic butter dish—makes the clock more interesting visually.

crystal-controlled oscillator is used to drive the clock. In the oscillator, the MM5369 CMOS oscillator/divider takes the output from a 3.58-MHz crystal (that crystal is used in color television sets) and divides it using an internal 17stage binary counter to produce an accurate 60-Hz output. Typically, the oscillator draws less than 1.5 milliamperes.

Construction

While such things as wire size and lead length are not critical, certain precautions should be taken with component placement. Traces on the module must not be shorted by the transformer case, temperature-sensor input jack, fastening bolts, or uninsulated wires. Since the reed switches are operated magnetically, they should be placed as close to the edge of the acrylic case as possible. Also keep in mind that the transformer generates a magnetic field, and could affect the operation of the reed switches if they are placed too near it.

When drilling the case, it is very easy

PARTS LIST

Resistors ¼ watt, 5% R1—220 ohms, metal film or wire wound R2—10 megohms Capacitors C1, C2—47 pF, ceramic disc Semiconductors IC1—MA 1026 clock module (National) IC2—LM334 adjustable current source/ temperature sensor IC3—MM5369 oscillator/divider J1—miniature phone jack, panel mount PL1—miniature phone plug RY1-RY3—reed switches T1—dual secondary; 10.5 volts at 25 mA, 6 VCT at 25 mA

XTAL1-3.58 MHz crystal

Miscellaneous: acrylic case, small piece of perfboard, wire, solder, hardware, epoxy glue, etc.

NOTE: If you are unable to obtain IC1, T1, or XTAL1 from your usual dealer, they can be ordered from Digi-Key Corporation, Hiway 32 South, PO Box 677, Thief River Falls,MN 56701

to scratch or shatter the acrylic material unless you are careful. To make holes, place the case on a cloth and drill a "pilot" hole using a 1/16-inch bit. When you enlarge the hole, it is best to reverse the direction of the drill just before breaking through; use the friction of the moving bit to melt the remaining plastic. That technique avoids the danger of a snag and the resulting damage.

Fast-drying epoxy is clear and forms an excellent bond for mounting the module and other components to the acrylic—do not use glue on the face of the module as it will appear as a trapped drop of water. Avoid "super" glues; their fumes will create a permanent cloudiness in the clear plastic.

Before closing up the case, you will need to calibrate the module so that it reads the correct temperature. The temperature-sensor assembly should be stabilized in a known environment. To do that, prepare an ice-point bath of water and ice. Place a mercury thermometer in the bath, and watch the temperature until it drops to 32°F (or 0°C). Once that happens, the temperature of the bath should remain constant until all of the ice melts. Before then, however, place the sensor in the bath, and adjust the module potentiometer (the blue knob on back of the module) until the temperature reading is precisely 32°F (or 0°C). Once that is done, remove the sensor from the bath and close up the case. If you use epoxy, use only small drops-you might want to open up the case again someday. The completed project is shown in Fig. 4.

Options

As we mentioned earlier, the MA1026 module is loaded with features and continued on page 107

RADIO-ELECTRONICS

TECHNOLOGY TODAY

Part 3 EARlier in this series we discussed the data that's transmitted on the vertical-blanking interval lines, preparing the videotex decoder for the video data that is to follow it, but did not talk about that information itself.

The actual display data-the alphanumerics and graphics -is placed on a video line on a "spaceavailable" basis, in accord with the North American Broadcast Teletext Standard proposal. That's the reason for the bytes in the headers that indicate where the next information is to be found. The data contained in the headers also allows the control information to be multiplexed with the display information.

The videotex decoder separates the two and stores them in RAM. At the appropriate time, the alpha-graphic (display) information is called up by the decoder and acted upon by the control information to determine charactersize, color, etc. That system is known as the parallel-attribute format.

Prestel-which has also filed with the FCC for acceptance-does it differently. As mentioned earlier, Prestel data is transmitted sequentially-the first character is sent first, and the last one last, and that's how they appear on the screen. The character-attribute data is sent along with the alpha-graphic data. That means that when an attribute (such as color), is changed, a pixel (picture element) is sacrificed and nothing is seen in the space that would otherwise be occupied by a character. Because one character or control element follows the other, the system is known as serial-attribute transmission.

(Editor's note—This is known as Level-1 Prestel. There are a total of five levels of Prestel, each more advanced than the previous one. Level 1, while the most "primitive," is also the most common at this stage. The proposal before the FCC includes all five levels.)

Alpha-graphics are sent using the ASCII (American Standard Code for *President, Richard Larratt and Associates Inc.

Videotex for TV-



How it works

Videotex has several levels of alphanumeric and graphics capabilities. Here's a discussion of how they work, as well as a look at videotex's future.

RICHARD LARRATT, M.A., P. Eng.*

Information Interchange) code—the one that is used by most computers. In it, each character is represented by a 7-bit word. For example, the letter "A" is represented by the decimal value 65, and a space by the decimal value 32. Since the information is transmitted in digital form, the "A" becomes "1000001," and the space "0100000," Table 1 shows videotex alphanumeric characters and their ASCII values. A switchable control byte allows the same numbers to be used to obtain other characters.

In the simplest forms of videotex (see Fig. 5 in Part 1 of this series in the November 1981 issue of **Radio-Electronics**) the ASCII data is sent to a character-generator ROM. That IC contains the information needed to display all the letters and graphic symbols

required by the system. When it is fed an ASCII value, it responds by outputting the dot pattern needed to form the character on the video display.

DRCS

The next step up in videotex technology calls for the use of a DRCS (Dynamically Redefinable Character Set). The DRCS allows the use of more graphics elements than are contained in the character-generator ROM. A videotex decoder equipped for DRCS's must contain some RAM. That memory stores the DRCS downloaded from the videotex computer over the air (or cable) and in effect acts as a secondary charactergenerator ROM. An example of videotex graphics using а

DRCS is shown in Fig. 15.

In addition to providing enhanced graphics, DRCS's allow the use of special alphabets, such as Greek or Hebrew, to transfer mathematical data or for foreign-language captioning.

PDI's

PDI's (*P*icture *D*escription *I*nstructions) allow the supplier of videotex information even more flexibility. Table 2 shows the PDI's described in the North American Broadcast Teletext Standard proposal. Once they are downloaded into the decoder's memory, they make possible very complex graphics with very little effort. For instance, the LINE PDI could be used to instruct the decoder to draw a "...LINE from *here* to *there...*".

Once the PDI's are loaded into memory, they make possible a wide variety of high-quality graphics, such as the illustration shown in Fig. 16.

PDI's also require RAM (typically 16K) but can make a PDI-equipped decoder worth its expense when graphics are of the essence.

Alpha-photographic

An example of alpha-photographic videotex is shown in Fig. 17. Alpha-photographic videotex allows the transmission not only of text, but also of still

				TABLE 1				
Decimal	Binary	Character	Decimal	Binary	Character	Decimal	Binary	Character
32	0100000	SPACE	96	1100000		64	1000000	@
33	0100001		97	1100001	а	65	1000001	@ A
34	0100010	"	98	1100010	b	66	1000010	B C
35	0100011	#	99	1100011	С	67	1000011	С
36	0100100	\$	100	1100100	d	68	1000100	D
37	0100101	%	101	1100101	е	69	1000101	E
38	0100110	&	102	1100110	f -	70	1000110	F
39	0100111	1 1 1	103	1100111	g	71	1000111	G
40	0101000	(104	1101000	h	72	1001000	н
41	0101001)	105	1101001	i	73	1001001	1
42	0101010	* * *	106	1101010	j j	74	1001010	J
43	0101011	+	107	1101011	k	75	1001011	К
44	0101100		108	1101100	1.0	75	1001100	L L
45	0101101		109	1101101	m	77	1001101	M
46	0101110		110	1101110	n	78	1001110	N
47	0101111	- 1	111	1101111	0	79	1001111	0
48	0110000	0	112	1110000	р	80	1010000	Р
49	0110001	1	113	1110001	q	81	1010001	Q
50	0110010	2	114	1110010	r	82	1010010	R
51	0110011	3	115	1110011	S	83	1010011	S
52	0110100	4	116	1110100	t	84	1010100	т
53	0110101	5	117	1110101	u	85	1010101	U
54	0110110	5 6	118	1110110	v	86	1010110	V
55	0110111	7	119	1110111	w	87	1010111	W
56	0111000	8	120	1111000	×	88	1011000	X
57	0111001	9	121	1111001	у	89	1011001	Y
58	0111010		122	1111010	z	90	1011010	Z
59	0111011		123	11,11011	1	91	1011011	[
60	0111100	<	124	1111100	1	92	1011100	A STATE
61	0111101	=	125	1111101	1	93	1011101	-]
62	0111110	>	126	1111110	~	94	1011110	Ň
63	0111111	?	127	1111111	DELETE	95	1011111	_

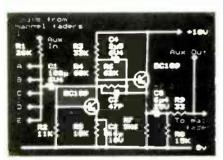


FIG. 15—A DYNAMICALLY REDEFINABLE character set (DRCS) makes the presentation of pictures such as this schematic relatively easy.

video frames—something that will be very important when shopping-byvideotex becomes widespread and the buyer wants to see not a drawing, but a real picture of the item he is interested in.

Figure 10 in the December 1981 issue of Radio-Electronics shows an alphaphotographic system. Several layers of RAM are overlaid, and their contents taken as a whole to provide varying brightness levels, which are not available in less sophisticated forms of videotex. As an example, let's look at bit-1 of each of the four planes of RAM shown in that figure. If we consider the four planes together, we can build values ranging from 0 to 15, depending on whether the bits are "1" or "0"'s. That means that each pixel can be assigned one of sixteen brightness levels, if the values are converted into

TABLE 2 TEXT: Present a string (group) of characters. POINT: Set the position for a point and, if requested, draw it. LINE: Draw a line using its endpoints as starting and stopping coordinates. ARC: Draw a circular arc using three sets of coordinates (any circle or part of a circle can be defined by three points). **RECTANGLE:** Draw a rectangle of a specific height and width. POLYGON: Draw a polygon based on the coordinates of its vertex points. BIT: Produce an image using bit-mapped graphics. CONTROL: Provide control over the modes of the drawing commands.

voltages that will control the electron beam of the video display.

Scanned sequentially, the contents of the bit-plane memory will appear as a series of brighter or darker pixels—a photograph! Even color is possible.

Of course, there is a tradeoff involved. Depending on the baud rate at which the data is sent, it can take several seconds for the picture to be completed. (Actually, that process would not be apparent to the user because the data would be loaded into RAM and not displayed until everything was present.) That's one reason why the "photograph" in Fig. 17 fills only part of the screen—a small picture requires less memory and can be transmitted and displayed more quickly (about two seconds for the one shown). Despite its apparent disadvantages, though, alpha-photographic videotex can be quite useful.

Which is best?

We've presented several levels of videotex—alpha-mosaic, alpha-geometric, and alpha-photographic. Which one is going to win out?

That's difficult to say. Eventually, of course, we'll be using the alphaphotographic mode, and every level below it. For the immediate future, however, we are faced with a conflict between a rather simple (although technically sophisticated) alpha-mosaic system, typified by Level-1 Prestel, and a more expressive but more complex and, initially, more expensive—alphageometric system of the Antiope/ Telidon-type.

Proponents of both systems have petitioned the FCC for acceptance of their method as the North American standard. Those backing the Prestel system claim that it is simple, inexpensive, and has a proven history.

Those in favor of the Antiope/Telidon

system say that its flexibility outweighs its apparent complexity, and that component prices will decrease rapidly once it is accepted and decoders are produced in volume.

At the moment, the decision lies with the FCC. Recent developments in FCC policy, however, indicate that the ultimate decision may be up to the user: both systems may be allowed, and the winner will be the one with the most subscribers. It's even possible that the two systems will coexist, one in some areas, and one in others.

That, however, invokes memories of "quad" sound, and the controversy and confusion that surrounded it before it more or less disappeared. Whatever the outcome, videotex—in one form or another—will be with us, and will make life a bit simpler and more rewarding.

Using videotex

Let's take a look at a not-so-hypothetical videotex system. We'll assume that the system used is that proposed to be the North American Broadcast Standard and that the videotex data is being sent over the air on two lines of the vertical-blanking interval. The equipment involved is an ordinary television receiver equipped with a videotex decoder and a small numeric keypad like a *Touch-Tone* pad. Also attached to the decoder is a modem, through which information can be sent back by telephone to the videotex computer.

A good videotex information-system will be constructed like an onion—the outside layer of information will be general. More detailed information can be obtained by requesting pages indicated by the first level. Still *more* detailed information will be available by selecting pages indicated by the second level. Also available will be an alphabetical directory—it might tell you that the page numbers for categories beginning with the letters "VAA" through "ZZZ" can be found on page number such-andsuch.

Let's assume that you want to make an airline reservation from New York to San Francisco. The first thing you will do will be to find out what page carries airline information. Upon entering the page number, a directory to various airlines will appear on the screen within about 10 seconds.

When the airline you select appears on the screen, it may present a menu showing the cities served, with each city identified by a number. By keying in that number, you will find another page showing the line's flights to that city, and possibly the status of each flight ("BOOKED FULL," "STAND-BY ONLY," or even "SEATS AVAIL-ABLE").

By keying in the number associated with the flight you want, a menu will be presented for you to "fill out" to make

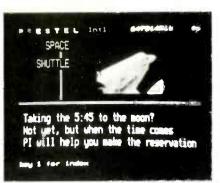


FIG. 16—PDI's (Picture Description Instructions) allow very complex graphics to be generated.



FIG. 17—THE ALPHA-PHOTOGRAPHIC mode allows photographs to be included as part of the videotex material. The picture shown took about two seconds to transmit.

your reservation and, at that time, you will be instructed to set up your telephone link to transit that information to the videotex or airline computer.

You'll be asked for your videotex account number, which may be reconfirmed on the screen. That not only automatically tells the airline just who is buying the ticket, but can also be used for direct-billing purposes. You'll then fill in the blanks on the screen to indicate how many people will be traveling, what class of service is desired, departure date, etc.

Finally, you will be given the option of committing yourself to that reservation, or—and sometimes more important—simply telling the computer to forget the whole matter.

Getting detailed news would also be done through a series of ever-more-detailed menus. The first page would be an index by subject ("WORLD." "NATIONAL." "SPORTS," "COM-ICS." etc.) and would also contain the latest headlines. By unpeeling layers of the information "onion," you could scan the items of interest to you, or read about them at length.

Who pays for it?

There's no such thing as a free lunch. Videotex equipment is not cheap, and the time that goes into programming the system also costs money. Where does the money come from?

In some instances, the company selling a service or product (such as an airline) pays out of its own pocket—it's probably less expensive than running a sales office. In other cases, such as financial reports like the Dow Jones index, a charge of 10 or 25 cents would be made to your account each time you accessed the service, or you would be billed for the time that you used.

Finally, to encourage the use of the system, some services, like news headlines or weather, might be provided without charge. After all, if you're going to use your videotex terminal for that, you'll also eventually find yourself using it for services that you (or someone else) pays for.

How we'll use it

Technically, videotex is quite an achievement. What, though, will it do for you? Some ideas were presented above; here are a few more.

Initially, interactive videotex—where the user can communicate with the service—will be limited. Broadcast videotex, using a full video-frame occupying all of a TV channel will predominate, presenting a continuously scrolling display of information. But the viewer will have no opportunity to take advantage of it, except as a non-selective information source.

At the beginning, interactive videotex will probably be available at selected public terminals (in airports, offices, and other places where its capabilities can justify the initial expense).

Our world, though, is changing. We are, more and more, becoming a "wired society." Right now that pertains primarily to telephone service—you can speak to anyone in the world for what would have been considered, a few years ago, a ridiculously low price.

Now, cable-TV is becoming a part of our lives and, if one channel were used to receive videotex information, and another to transmit multiplexed viewer response (at a relatively low-data-rate), interactive videotex over cable could easily become a reality.

In the near future we'll see two more developments that will further the interactive aspect of videotex.

The first is fiber optics. By using light, rather than RF, as a transmission medium, it will be possible to have an almost unlimited number of channels for information interchange. Fiber optics, using hair-thin transparent lightguides instead of cables, is their logical replacement. We'll be seeing a lot more of that technology soon.

The second could very well be DBS (Direct Broadcast Satellites). They could replace any wired—or light-piped —information service and, although it may take a while, they may become two-way. Not only would you be able to receive the programs or data they sent, but a relatively-low-power transmitter might transmit your information into the system.

continued on page 103

HOMITIMORKS

SPEECH KARL SAVON SYNTHESIS Techniques

SEMI-CONDI

FR

Electronic speech-synthesis hs come a long way from the Voder, which took seven people to operate when it was demonstrated at the 1939 N.Y. World's Fair. Now all you need are a couple of IC's.

SPEECH SYNTHESIS—THE ARTIFICIAL generation of that series of sounds known as "speech"—has advanced beyond the novelty stage to become a real alternative to the simple audible and visible indicators and displays common to so much of our society. There is very little you can do with a light or buzzer that can't be done better with a "spoken" word. "Toys" such as Texas Instruments' Speak & Spell have been recognized as effective learning tools. Through the electronics mouths of those machines, children are exposed to new words in an exciting interactive way. Talking calculators and timepieces have expanded the horizons for the blind; pilots and drivers are relieved of the need to watch their meters and gauges continuously; alarms can be given with instructions as to what actions should be taken. Speech-synthesis devices are finding a myriad uses in communications, appliances, automotive applications, clocks, instrumentation, language translators, and annunicators. It is estimated that the speech-synthesis market will grow to hundreds of millions of dollars in the next five years.

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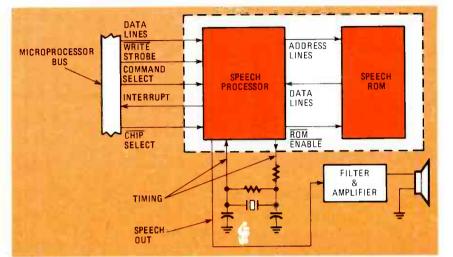


FIG. 1—A TYPICAL IC SPEECH-SYNTHESIZER uses a ROM containing speech data, circuitry to process the data and turn it into analog form, and an amplifier and speaker.

Fundamentals

Human speech begins in the cortex of the brain but, so far, that *creative* process has been imitated only crudely. The *mechanics* of speech, though, have been simulated with uncanny precision.

The human speech process begins in the lungs and proceeds through the larynx to create the noises and tones that correspond respectively to unvoiced and voiced human speech output. The wideband hissing "s", "f," and "sh" are unvoiced sounds, created by the random flow of air particles through the voice tract, but the voiced consonants like "b" and "l" are constructed of discrete narrowband frequencies: the vibrating vocal cords generate the voiced sounds. Fricatives like "v" and "z" combine the two types of sources and the *plosives*, such as "p," "t" and "k" are produced by the controlled release of air from the mouth.

The speech process is completed by articulation—changing the shape of the throat, and the shape and positioning of the teeth, tongue, palate and lips. Together, those parts of the mouth and respiratory system form an adaptive filter that alters the resonances of the vocal tract.

While there are more similarities than differences among the approaches to speech synthesis, they can be separated into three general groups: direct waveform-digitization, phoneme synthesis, and linear-predictive coding.

Direct waveform-digitization

Direct waveform-digitization is the process of taking a live or recorded speech-waveform and passing it through an analog-to-digital converter. The final quality of the reconstructed speech (obtained by passing the digitized data through a digital-to-analog converter) depends on several factors, including the rate at which the speech waveform is sampled (how many times each second the waveform is "looked at" and its value digitized) and the number of amplitude levels into which it is segmented. As the number of levels increases, more information must be processed and stored in memory; as the number of levels decreases the memory and processing requirements decrease -at the expense of intelligibility

It is necessary to sample the waveform at a frequency that is at least twice as high as the highest frequency in the original waveform. If the sampling frequency is too low, *aliasing*, which causes the higher frequencies to appear falsely as lower frequencies, takes place.

Playback of the digitized signal is (in theory) a simple matter of recalling the sequences of data stored in memory and processing them with a digital-toanalog converter.

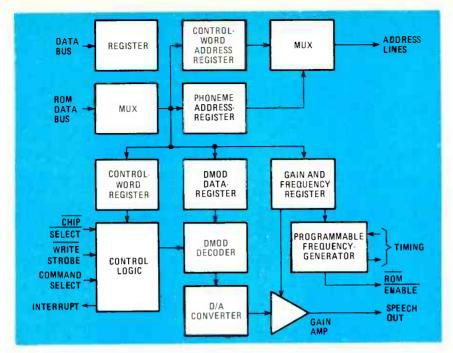
Direct waveform-digitization of human speech is highly redundant; it wastes a lot of memory storing, for example, drawn-out sounds like "000000h." Methods used to reduce the memory requirements and rate of processing of redundant information may actually cause this synthesis technique to be classified in one of the other simulation groups.

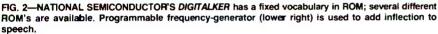
National Semiconductor (2900 Semiconductor Dr., Santa Clara, CA 95051) uses a direct-digitization method that reconstructs the speech using pulsecoded modulation. It calls its system the *Digitalker*. Compressed speech data is stored along with frequency and amplitude information in a read-onlymemory (ROM). Figure 1 shows a typical configuration connected to a microprocessor bus complete with an output filter, amplifier, and speaker. Figure 2 shows a block diagram of the speech processor chip, referred to as the SPC.

It cannot be overemphasized that the key to the practicality of this, as well as the other systems, is speech-compression-coding of the signals that minimizes the redundant information. National points out that its system produces speech quality far better than the crude sound you may associate with early demonstrations of speech digitization. In terms of memory space, the result is that male voices require about 100 bits per word and female voices somewhat more. The system is more like a digital recorder that digitizes actual voices, stores, and then plays back, than the other methods which model the vocal tract.

The compression method combines three techniques. First, redundant pitch periods are removed. Then, adaptive delta-modulation, which uses the difference between two successive sampling points rather than the values at those points, is used to conserve memory space by storing only the difference information. Phase adjustments and half-period zeroing, remove the "direction" component of the waveform: that information is not needed for intelligibility, so can be safely done away with. Computer processing is used to accomplish the compression. National claims that its compression scheme, combined with waveform digitization, is very competitive with other systems, including linear-predictive coding.

The *Digitalker* is programmed with control information that instructs it how many times to repeat a specific waveform. A programmable frequencygenerator is used to add inflection. The system is easy to use because it requires only a start pulse and an 8-bit address to trigger any message. Simple switches can be used if a microproces-





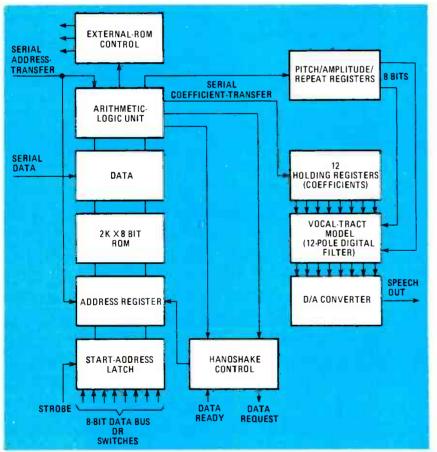


FIG. 3—A PHONEME-BASED synthesizer, such as General Instruments' SP0256, uses the sound elements that make up words to provide an almost unlimited vocabulary.

sor controller is not justified. One *Digitalker* device can produce up to 256 different messages.

An evaluation board is available which includes the NMOS speech-pro-

cessor IC, two speech ROM's containing 138 words, and an erasable programmable ROM (EPROM). The *DT1000* board can be connected to an external control-system through a 22pin edge connector. A demonstration package is available with a vocabulary aimed at several industrial markets. The 5 \times 6-inch *DT1000* evaluation board is priced at \$495. The *DT1050* three-piece IC set is \$85. Write directly to National Semiconductor for more information.

Phoneme synthesis

Phoneme synthesis works by combining basic sound elements (phonemes) that are made up into complete words and sentences. In theory, any spoken word can be synthesized by stringing phonemes together. The quality of the resulting speech depends on the extent of the phoneme library. To cover a wide range of inflection requires a correspondingly large number of phonemes. Again, compromise is necessary to put together a practical system: redundancy-elimination coding techniques are used.

The phoneme method is particularly suitable when the extent or type of vocabulary required is not fixed: When there is no definite vocabulary list, speech cannot be constructed from stored words and phrases but must be generated from the more elemental phoneme sounds.

General Instruments (600 W. John St., Hicksville, NY 11802) has taken the approach just described, combining phoneme synthesis with a digital filter. The design is versatile enough to operate in a linear-predictive filter mode (more about that shortly). GI has released a product specification that describes the SP0256 speech-processor IC; it is an LSI n-channel metal-gate device that can synthesize up to 256 sound sequences.

Figure 3 shows a block diagram of the processor. The speech process is started by addressing the ROM location that contains the phoneme desired. Up to 256 phonemes can be stored in the 16K bits of the on-board ROM, but that can be extended to as many as 3825 phonemes (or, more usually, complete words or phrases) through the addition of up to 491K bits of ROM.

The device includes a controller and a vocal-tract model (VTM). The VTM is a digital filter similar to that used in linear-predictive coding. The system generates complex sound-sequences under the control of 15 slowly varying parameters including: repeat count, pitch period, source amplitude, and 13 digital filter-coefficients. The controller is a sequential processor that gets its instructions and data from the ROM and alters the controlling the VTM.

One-byte inputs specify 256 entry points. The entry points in the internal ROM are spaced by 8-byte increments, meaning that each phoneme can be defined by 8 bytes (64 bits). Expanding

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the input format to 2 bytes increases the number of entry points to 3825, assuming that the full complement of ROM has been added. The controller has 16 executable instructions and supports one level of subroutine nesting. JMP (JuMP) and JSR (Jump to Sub-Routine) instructions are included to allow chaining of program segments and the reuse of code sequences.

External ROM interfaces to the speech-processor device through serial input and output lines. Sixteen-bit serial addresses are used for addressing. Serial-to-parallel conversion is handled internally by the SP0256. The analog output is developed by a 7-bit pulsewidth modulation, digital-to-analog converter. The output is passed through a 5-kHz-cutoff low-pass filter and is then amplified externally. The single-unit price for the SP0256 is \$26.70. The price will probably be even lower from a General Instruments distributor.

Votrax's phoneme system

Votrax, Div. of Federal Screw Works (500 Stephenson Highway, Troy, MI 48084) offers another phoneme-based system. Votrax uses a switched capacitor filter in its CMOS-technology SC-01 LSI integrated circuit. Phonemes require an average of 6 bits and each phoneme is 40 to 200 milliseconds long. The device has a 64-phoneme library accessed by a 6-bit code. Typical speech-data uses 70 bits per second. Votrax plans to provide a text-tophoneme translator algorithm that will let the user do his own programming. Figure 4 shows a block diagram of the system. The price of the SC-01 is \$55.00 each in quantities of five (single units may be available from distributors).

Linear-predictive coding

Linear-predictive coding—used extensively by Texas Instruments (P.O. Box 225012, Dallas, TX 75265)—is a combination of techniques that model the vocal tract electronically. Noise and tonal sources generate signals that are processed by the LPC filter. The method derives its name from the fact that it predicts the parameters of the next speech sample using a linear combination of the proceding speech samples. That results in a major reduction in the amount of memory required for the storage of speech data.

Speech-synthesis techniques using that method require as few as three integrated circuits: a digital latticefilter, a ROM, and a controller. Present systems are based on the 10-stage filter shown in Fig. 5; its output is set by pitch, amplitude, and filter coefficients.

The lattice-filter structure includes multiplication, summation, and delay blocks. Digital filters are constructed of memory-register delay components, and of summers and multipliers con-

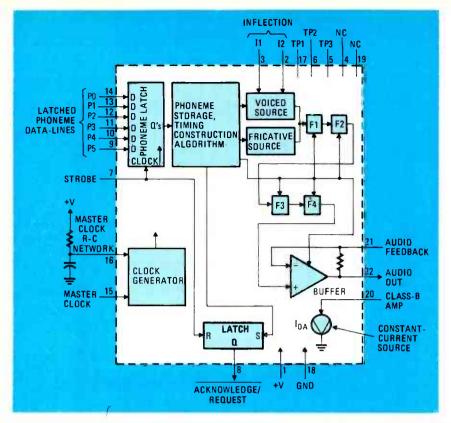


FIG. 4—VOTRAX'S SC-01 is a single-IC, phoneme-type, speech synthesizer intended for computer control.

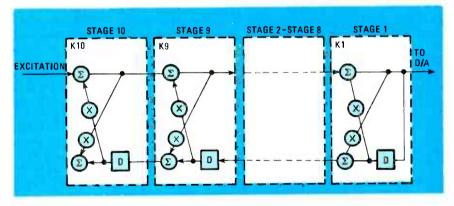


FIG. 5—TEN-STAGE LATTICE FILTER used in Texas Instruments' linear-predictive-coding synthesizer simulates the resonant effects of the mouth and nasal cavities.

nected in either feedback or nonrecursive, direct path configurations. Multiplication takes longer than addition, so pipeline techniques are used to synchronize the filter's operation to the sampling rate of the system. The allpole filter is described by the equation:

$$H_{Z} = \frac{G}{10a_{k}^{Z-k}}$$
$$1-\sum_{k=1}^{K-1}$$

The a_k terms are the ten filter-coefficients, G is a gain factor, and the z^{-k} terms represent time delays.

The voice-synthesis processor performs about 400,000 multiplications and additions each second. A frame (complete set of speech data) is supplied to it about every 25 milliseconds. Data from adjacent frames is interpolated about every 3 milliseconds to generate a smooth output. Once again a tradeoff is made among flexibility, fixed or variable vocabulary, speech quality, and costper-second of speech output.

A computer program is used to produce an optimal set of coefficients for the vocal tract filter, the 8-bit energy and pitch information, and to generate the one-bit repeat codes. A technician corrects any audible deficiencies.

Texas Instruments' TMS5100 speechprocessing computer, which includes a 50-milliwatt power amplifier, is a key component in its speech-synthesis line. The TMS5220 is designed for easy 8-bit data bus interface. It is used in the *TI*-99/4 home computer speech-synthesizer peripheral. Single-unit prices for the TMS5100 and TMS5220 are \$32.00 and \$48.00 respectively. **R-E**

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A subwoofer allows your sound system to reproduce very-low bass, such as that found in organ recordings. This SAFE subwoofer may be small, but it's mighty where it counts.

GEORGE PAPPANIKOLAOU

IN THE DECEMBER 1981 ISSUE. **RADIO-ELECTRONICS** PUBLISHED plans for constructing a speaker that used the SAFE (Symmetrical Air Friction Enclosure) principle (U.S. patent no. 4, 168,761, other patents pending) to obtain good sound from a small enclosure. This article deal with a companion subwoofer using the SAFE principle that provides extended bass response from a relatively small enclosure.

The heart of the system is an 8- or 10-inch speaker in a cabinet with a volume of under six cubic feet. It uses no special devices, such as passive radiators, to increase the effective size of the speaker, and the SAFE design provides an extremely rigid enclosure.

The SAFE principle

The SAFE principle is an outgrowth of the labyrinth-type enclosure invented by Benjamin Olney in the 1930's. A labyrinth-type enclosure is simply a long tube placed behind a speaker and into which the low-frequency back wave from the speaker is propagated. The length of the tube is used to provide "reflex action," causing the back wave to emerge from the tube in phase with the front wave. This boosts the low-frequency output of the system over a specific range of frequencies. The longer the effective length of the tube, the lower the frequencies that will be boosted.

The reason for using the term "effective length" is that friction between the air set in motion by the speaker and the internal surface of the labyrinth causes the velocity of the back wave to be decreased and making it take longer to pass through the tube than would be the case if there were no friction. The fact that it takes the sound wave longer to transverse the labyrinth than it normally would, makes it appear that the wave has traveled a longer path than it actually did.

Two methods commonly used to increase effective length of the labyrinth are to line its interior with acoustic padding such as fiberglass, or to completely fill it with the same material.

Many designs usually fold the tube (run it back and forth instead of in straight line) to keep the size of the enclosure

R-E TESTS IT

LEN FELDMAN CONTRIBUTING HI-FI EDITOR

WE PUT AN EIGHT-INCH SAFE SUBwoofer through extensive listening tests in our laboratory and also subjected it to several measurements. We drove it using an amplifier rated at about 40-watts-per-channel and a passive crossover network.

Based on our sensitivity measurements, we would classify this speaker as being low-to-medium in efficiency (an electronic crossover would have helped). Nonetheless, we noted a fundamental output from the subwoofer down to 25 Hz. We measured relatively uniform response from the system from 30 Hz up to the 100-Hz cutoff frequency and were able to produce sound-pressure levels in excess of 100-dB SPL before any significant doubling or distortion was observed.

The unit was tested using both single-tone signals and musical material. In reproducing the music, the improvement in bass reproduction was apparent only when the program source contained a fair amount of low-bass energy. Not too many program sources do, but when you find one that *does*, the added impact of the low, low bass contributed by the subwoofer made the project decidedly worthwhile.

One word of caution. Trying to reproduce ultra-low bass in a small

room can prove to be a frustrating experience. Standing waves set up in the room are responsible for that, and it is possible to move about the room and hear thunderous bass in one location while hearing virtually none in another spot, where virtually complete cancellation of bass frequencies takes place.

We used close-miking techniques in our sound-pressure measurements, placing the microphone both at the upper back-wave ports of the enclosure and at close proximity to the driver itself. Below 40 to 50 Hz, most of the energy, as might have been expected, came from the backwave ports rather than from the front of the woofer. **R-E**

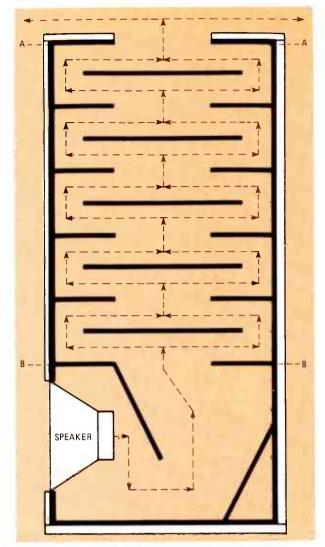


FIG. 1—SIDE VIEW OF SAFE ENCLOSURE showing speaker back-wave path. Path is symmetrical between points A and B.

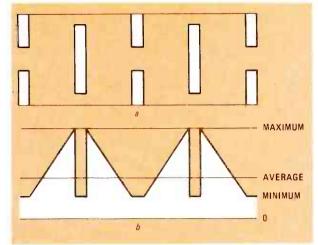


FIG. 3—PARTITIONS USED IN SAFE ENCLOSURE (a) affect its effective cross-sectional area (b).

down. Also, the effective length of the tube and the free-air resonance of the speaker are chosen so that the speaker is damped at its free-air-resonance frequency, avoiding objectionable excess output at that frequency.

There are, however, two other factors that can be used to increase air friction, and thereby the effective length of the labyrinth. These factors are what's involved in the SAFE principle. First, the internal surface area—not the length—of the tube can be increased; second, the back-wave path can be made longer by increasing the number of bends in it.

A SAFE enclosure uses a number of small chambers in series (see Fig. 1) to both increase the internal surface-area and to insert more bends in the path of the speaker backwave. The small chambers are constructed by alternating two types of partitions within the tube as shown in Fig. 2. The first type completely blocks the tube, except for a square hole in its center. The second is smaller and is centered so that the area between it and the inside perimeter of the tube is equal to that of the hole in the first type. This forces the back-wave from the speaker to be alternately spread out and combined as it passes through the tube. (It is obvious that this design increases the internal surface-area of the tube and

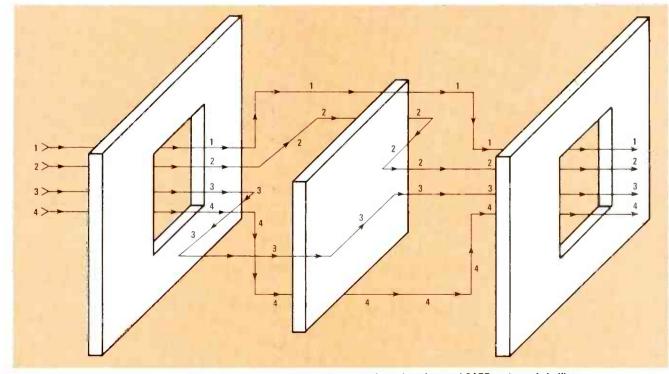


FIG. 2—SPEAKER BACK-WAVE is alternately combined and separated as it passes through and around SAFE enclosure's batfles.

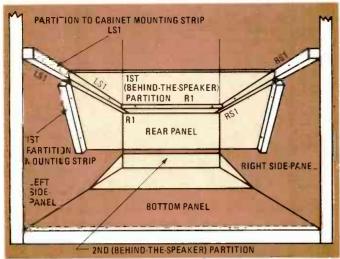


FIG. 4-EXAGGERATED PERSPECTIVE VIEW of first chamber. Note mountings of first and second behind-the-speaker partitions.

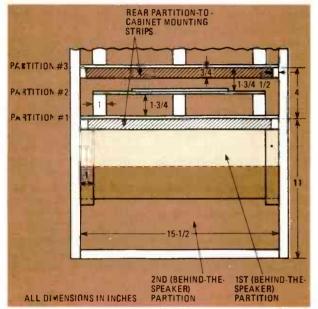


FIG. 5-NON-PERSPECTIVE VIEW from front of enclosure. Note construction of partition 1/partition 2 assembly.

	PARTS	LIST-ENCLOSURE
ze	(inches)	Description

Qty.	Size (inches)	Description
2	151/2 × 151/2 × 3/4	Front and rear panels
2	$17 \times 34 \times \frac{3}{4}$	Side panels
1	$15\frac{1}{2} \times 15\frac{1}{2} \times \frac{3}{4}$	Bottom
ľ	$15\frac{1}{2} \times 17 \times \frac{3}{4}$	Тор
1	151/2 × 151/2 × 1/2	Partition 11
1 5	$15\frac{1}{2} \times 15\frac{1}{2} \times \frac{1}{4}$	Partitions 1, 3, 5, 7, 9
5	$13\frac{1}{2} \times 13\frac{1}{2} \times \frac{1}{4}$	
40	$1 \times 1 \times 1^{3/4}$	Strips for mounting
		odd-numbered partitions to even-numbered ones
12	$\frac{1}{2} \times \frac{3}{4} \times \frac{151}{2}$	Partition-to-cabinet mounting strips
12	$\frac{1}{2} \times \frac{3}{4} \times 14\frac{1}{2}$	"
1	$15\frac{1}{2} \times 7 \times \frac{1}{4}$	First behind-the-speaker partition
1	15½ × 8 × ¼	Second behind-the-speaker partition
2	$6\frac{1}{2} \times 1 \times 1$	Mounting strips for first behind- the-speaker partition

Speaker-JBL LE8T (8-inch) or LE10A (10-inch) or similar Miscellaneous: Two binding posts, screws, hot-melt glue and glue gun, acoustic insulation, crossover network, etc.

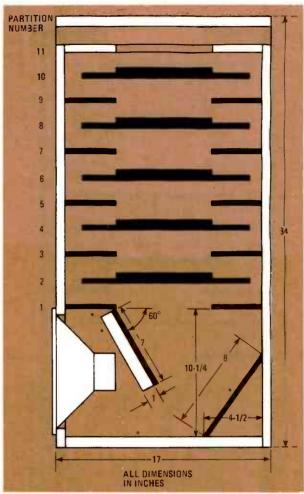


FIG. 6-SIMPLIFIED CUTAWAY VIEW of enclosure showing mounting details of first and second behind-the-speaker partitions.

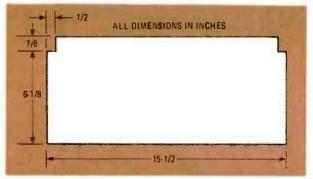


FIG. 7-FIRST BEHIND-THE-SPEAKER PARTITION has cutouts to accommodate mounting strips.

the number of bends in the back-wave path.) Traversing this path causes the back-wave to be compressed and expanded alternately (Fig. 3), slowing the wave and increasing the effective length of the tube.

Other factors

There are two other factors that have to be considered when designing a speaker enclosure: the free-air resonant frequency of the speaker used, which limits the low-frequency response, and the reflex action of the enclosure, which boosts the low-frequency output of the system.

The two factors are interrelated and an enclosure that does not increase the free-air resonant-frequency of the speaker (desirable) but doesn't provide any reflex action (undesirable). or one that does increase the speaker's free-air resonant frequency (undesirable) but does provide reflex action (de-

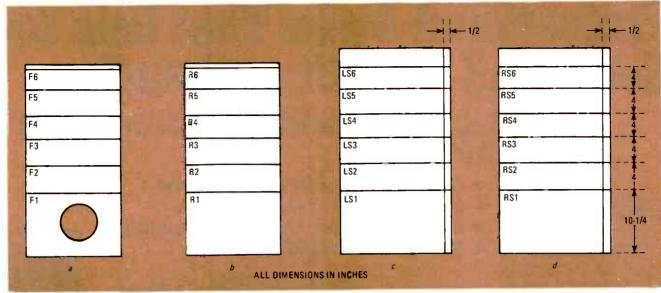


FIG. 8—PARTITION MOUNTING STRIPS for the front are shown in (a), rear are shown in (b), left side are shown in (c) and right side are shown in (d).

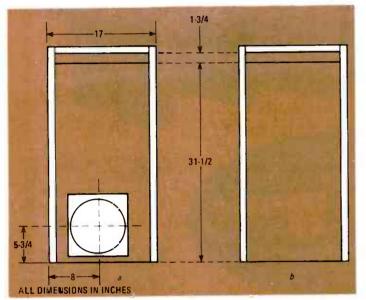


FIG. 9—SPEAKER SHOULD BE MOUNTED slightly off-center to prevent standing waves. Center figure shows distance between partition 1 and top of enclosure.

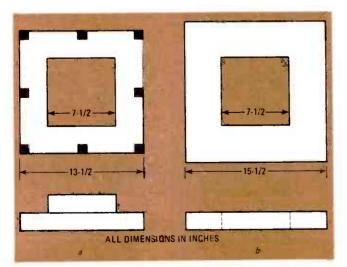


FIG. 10-EVEN-NUMBERED PARTITIONS (a) have "top hat" structure caused by using center piece cut out of odd-numbered partitions.

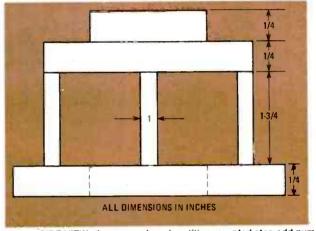


FIG. 11—SIDE VIEW of even-numbered partition mounted atop odd-numbered partition using partition mounting strips. Each strip is located as shown in Fig. 10 and measures $1 \times 1 \times 134$ inches.

sirable), is not as effective as one that provides reflex action *and* does not increase the speaker's free-air resonance.

The SAFE subwoofer enclosure is designed so that it does not significantly restrict the action of an 8-inch speaker (which would increase its free-air resonant frequency) while providing good reflex action. A 10-inch speaker can also be used. A 10-inch speaker would offer the advantage of increased power-handling capacity and, perhaps, better efficiency.

Construction

The choice of a speaker is not critical. Almost any 8- or 10inch open-frame woofer will do; an 8-inch unit should have a ceramic magnet weighing about 20 ounces. JBL's *LE8T* is a good choice for an 8-inch system; the same firm's *LE10A* should do nicely in a 10-inch one.

While detailed step-by-step instructions will not be provided, the figures and the captions accompanying them should be studied closely; they will allow you to construct the enclosure with little difficulty. (The only tricky part—the area behind the speaker—will be explained below.)

Like all subwoofers, the SAFE enclosure should be constructed from solid, heavy material, such as $\frac{3}{4}$ -inch particle board, to prevent cabinet vibrations that will decrease the output of the system. The partitions can be made from $\frac{1}{4}$ inch plywood, with the exception of partition 11, which should be $\frac{1}{2}$ -inch thick.

Use plenty of screws and glue! The former will add mechanical integrity and the latter will make the enclosure continued on page 106

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FEBRUARY 1982

HOMITIMORKS

STEREO AUDIO FOR TV

Stereo and multichannel TV programming is already a reality in several countries. Here's the latest on what's happening here.

LEN FELDMAN CONTRIBUTING EDITOR

BY THE TIME YOU READ THIS, ALL THE tests should have been completed. Even the report prepared by the Electronic Industries Association's Broadcast Transmission Standards (BTS) Committee may already be in the hands of the FCC, and we will be a step closer to having stereo audio for television broadcasts in the U.S. And, with two (or more) audio channels available, not only will stereo be able to be broadcast, but the second channel can also be used for the transmission of a second monophonic program-perhaps a soundtrack or commentary in a second language, for viewers whose native tongue is not English.

No matter how soon a multichannel audio system is adopted for U.S. TV, and no matter which system is chosen, the U.S. will still be a late starter in adopting stereo for TV. In Japan, that type of service has been available since 1978. In fact, of the three basic systems being proposed for use in the United States, one is the Japanese system that was developed and tested over a period of four years by the Technical Research Laboratories of NHK, the Japanese Broadcasting Corporation.

The two other systems currently being considered for use in this country are by Telesonics Systems, Inc. and by the Zenith Radio Corporation. The Japanese system under consideration is officially known in this country as the EIAJ system, since it is being sponsored in the U.S. by the Electronic Industries



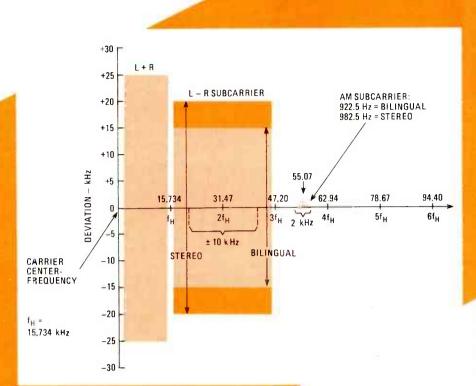
Association of Japan. Finally, a fourth system, developed—and just being introduced—in Germany, may have some influence on what method is finally adopted. (That system will be discussed in detail soon in a future issue of **Radio-***Electronics*).

Further complicating the decisionmaking process is the fact that, in addition to the three basic transmission-systems just referred to, there are also three noise-reduction systems being proposed for use with them, making a total of nine possible combinations. The incorporation of more audio channels in the bandwidth now assigned to a single TV audio channel will degrade the audio signal-to-noise ratio, regardless of which of the three stereo systems is selected. That is what has prompted Dolby, dbx, and most recently, CBS, to propose noise-reduction encoding/decoding systems. Those, coupled with any of the proposed multichannel TV-audio systems, would reduce background noise to acceptable levels and allow broadcasters to maintain the quality of fringe-area signals.

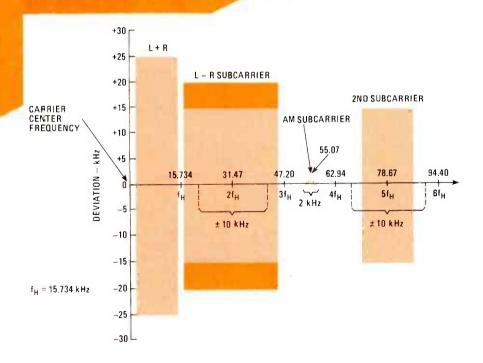
Since multichannel TV-audio will eventually be with us, it would be a good idea to understand how the three basic transmission-systems that have been proposed work. The principles involved in audio companding are familiar to most readers of **Radio-Electronics** so we will not include a description of the three noise-reduction schemes in this discussion.

The EIAJ system

The components of the signal used by the Japanese system are shown in Fig. 1. Note that the "x" axis (horizontal) serves two purposes—it shows both the audio-frequency response of each channel and the frequency of the







FIG, 2—U.S. VERSION of the EIAJ system extends frequency-response of the subcarrier to 15 kHz. In addition, a second subcarrier, with an 8-kHz bandwidth, is provided for.

subcarrier and pilot tone *relative* to that of the main audio carrier. The "y" axis (vertical) shows the frequency deviations of the main and sub-carriers.

The main audio carrier-used for monophonic transmissions. for one

channel of multichannel transmissions, or for the L+R component of a stereo transmission—has a maximum deviation of ± 25 kHz (just like the NTSC system used in the U.S.) and a frequency response from 50 Hz to 15 kHz.

The subcarrier, located at a frequency of 31.47 kHz (twice the TV horizontal line-frequency) above the main carrier, is used for the second channel of a multichannel transmission. or for the L-R component of a stereo transmission. That frequency was chosen because it minimizes interference between the video signal and the subcarrier in intercarrier-type receivers. When frequency-modulated, the subcarrier has a maximum deviation of 10 kHz on either side of the center frequency. In Japan, the frequency response of that channel is from 50 to about 12 kHz, but the system has been modified for use in the U.S. to extend the response to 15 kHz.

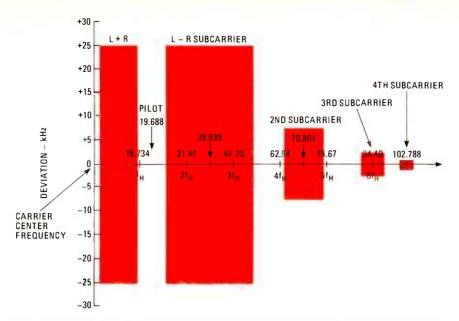
The subcarrier is used to modulate the main carrier—a technique known as *injection*. When stereo is being transmitted, the amount of injection is ± 20 kHz, giving the main carrier a total deviation of ± 45 kHz; in the multichannel mode, the amount of injection is ± 15 kHz, producing a total deviation of ± 40 kHz.

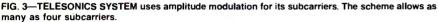
An amplitude-modulated control signal having a bandwidth of 2 kHz is transmitted at a frequency 55.07 kHz (3¹/₂ times the horizontal line-frequency) above the main carrier. Its purpose is to inform the decoding equipment at the receiver which mode is being used. A 982.5-Hz tone indicates that a stereo program is being transmitted; a 922.5-Hz tone indicates that a multichannel program is being broadcast. The absence of a tone means that the program material is monophonic. The control signal can also be used to activate a visual indicator to inform viewers/listeners which audio mode was being used with the video they were watching.

Tests made by the developers of the Japanese TV multiplex-system have shown that a second subcarrier can be added without making it necessary to change the parameters of the first subcarrier. In tests conducted by the EIA last summer in Chicago, such a system-as shown in Fig. 2-was used. The characteristics of the second subcarrier were the same as those of the first, with the exception that the audiofrequency response was restricted to an upper limit of 8 kHz. It is expected that that subcarrier would be used for services where limited frequency-response would not pose a problem.

The Telesonics system

The Telesonics system, whose arrangement is shown in Fig. 3, uses a double-sideband, suppressed-carrier. AM subcarrier similar to that used in stereo-FM broadcasting in the United





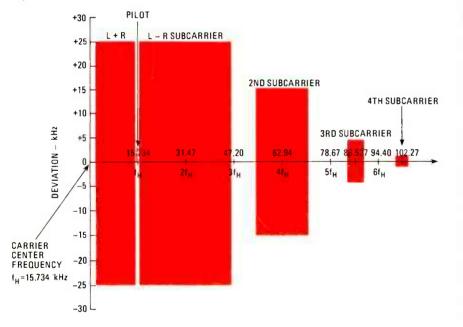


FIG. 4—ZENITH RADIO CORPORATION'S system uses the horizontal line-frequency, 15.734 kHz, for the pilot tone.

States and many other countries. (The EIAJ system, with the exception of the control signal, is all-FM.) Such an arrangement requires a pilot signal, which is used at the receiving end to restore the suppressed carrier. In the Telesonics system, the pilot signal has a frequency of 19.668 kHz, or 1.25 times the TV horizontal line-frequency. The doublesideband suppressed-carrier used for transmission of L-R information is 39.939 kHz, or 2.5 times the horizontal line-frequency, away from the audio carrier center-frequency. Tests of the system have been proposed using three different levels of deviation of the main carrier by the subcarrier: 11.25 kHz, 25 kHz, and 33.75 kHz. Tests using additional FM subcarriers at other frequencies, in order to provide a second subchannel for sound transmission; a third carrier for ENG (*Electronic News Gathering*), telemetry, or another service, and a fourth subcarrier for telemetry alone have also been suggested. The additional FM subcarriers are impressed upon the main carrier at very low levels and with highly restricted audio-frequency ranges, as shown in Fig. 3.

The Zenith Radio system

Frequency allocations and subcarrier arrangements used in the Zenith Radio Corporation system are shown in Fig. 4. Like the Telesonics system, the Zenith system also uses a double-sideband, suppressed-carrier AM subcarrier. Instead of using a separate pilot signal, the Zenith system uses the horizontal line-frequency itself for that purpose, with the center frequency of the suppressed subcarrier falling at twice the horizontal line-frequency, or 31.47 kHz. Provision for a separate audio program (such as a second-language summary of the news) is made using an FM subcarrier having its center frequency at four times the video horizontal-line rate, or 62.94 kHz, while other FM subcarriers at 5.5 and 6.5 times the horizontal frequency can also be included for telemetry or other telecommunications purposes.

Signal-to-noise ratios

As we noted earlier, all the stereo/ multichannel TV-audio systems involve some degradation of the signal-to-noise ratio. In Japan, where the system employed uses a frequency-modulated subcarrier for transmission of the L-R information (or the second-channel information), results at station JOAX (the anchor station of the national network. based in Tokyo), using an 85 kW audio carrier, show that the signal-to-noise ratio is about 60 dB on both the main and sub-channels at receiving points within the city, fairly close to the transmitter site. In fringe areas, of course, that figure would be worse.

Experiments have already shown that the systems using AM subcarriers (Telesonics and Zenith, so far) for their difference (L-R) information or secondchannel transmissions suffer a greater reduction in signal-to-noise ratio than do all-FM subcarrier systems. That's especially true in fringe areas, where the main audio-carrier is of insufficient strength to send the receiving circuitry into full FM-limiting. It is for that reason that the EIA Broadcast Television Standards Committee deemed it advisable to incorporate the companding system proposals mentioned earlier in the test recently completed.

Several critical listeners, none of whom knew which companding systems were being used (or, for that matter, that they were being asked to judge the merits of three companding systems), were asked to listen to a variety of recorded material and to judge which "sounded best." Having monitored some of the tests myself on behalf of the EIA, I can report that, while each companding system was judged effective for certain kinds of music, the difference in background noise between all the uncompanded transmissions and those using noise reduction was obvious, regardless of which system was used for the test. That was particularly true when "fringe area" reception conditions were simulated

Audio enthusiasts have complained for many years about the poor quality *continued on page 102* HLELSTERED



Designed to provide the ideal cassette tape bias level for signals at any frequency, this new version of the Dolby HX system overcomes the problems of the original and adds some interesting refinements.

JUST ABOUT TWO AND A HALF YEARS ago, Dr. Ray Dolby, of Dolby Laboratories, introduced a new signal-processing scheme that he called *Dolby HX*; the HX stood for "headroom expansion." That circuit was Dolby Laboratories' first attempt at signal processing that had nothing to do (at least directly) with noise reduction. While Dolby noise reduction has been very widely accepted by tape-deck manufacturers, the *Dolby HX* system was not greeted with the same sort of enthusiasm.

At least one manufacturer, however, must have been fascinated by the ideas behind *Dolby HX*—enough so to have developed a version of the headroomexpansion idea that overcomes earlier objections to the HX approach, and provides benefits that were not part of Dolby's original system. That manufacturer is the Danish firm of Bang & Olufsen, and, with the unqualified blessing of Dolby, their version is called *Dolby HX Professional*.

In order to appreciate the improvements in this new version of Dolby HX. it will be helpful to review the concepts behind Dolby's original version of the headroom-expansion system. The underlying concept is based on the fact that there is an ideal level of tape bias for every audio frequency. As can be seen in Fig. 1, the bias level that provides the highest output level and acceptably low distortion levels for a midfrequency tone of 333 Hz is quite different from the optimum bias level needed when recording a 10-kHz signal. Figure 1 also shows that as we attempt to record higher and higher frequencies. lower and lower levels of bias are required. Generally, most makers

of cassette decks choose a bias level that, of necessity, is a compromise one that covers reasonably well as much of the audio range as possible. Usually, that bias level is set so that it is a bit lower than optimum for best low-frequency/mid-frequency maximum output level (MOL), and for a reasonable output level at high frequencies. Since such compromise settings of bias tend to favor low and mid frequencies, the phenomenon known as high-frequency tape saturation is common in cassette decks.

The earliest version of *Dolby HX* worked on the assumption that, if the bias level of a recorder is made variable and allowed to change dynamically with changing input-signal content, it should become practical to record high-frequency signals at higher levels by reducing the bias levels when high-level,

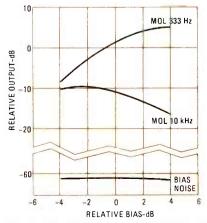


FIG. 1—THE BIAS LEVEL that provides the highest output with acceptably low distortion for a 333-kHz tone is quite different than the bias level required by a 10-kHz tone.

LEN FELDMAN CONTRIBUTING HI-FI EDITOR

high-frequency signals dominate the program content being recorded. Furthermore, since the headroom expansion takes place entirely during the recording process, no "decoding" or correction circuitry is needed during playback to get the benefits of such a variable bias system.

A block diagram of the original system is shown in Fig. 2. For the system to work, there must be a control signal that is sensitive to amplitude and frequency content. But such a control signal already exists in any *Dolby B*-equipped deck. All *Dolby B* noise-reduction circuits generate a voltage signal that depends on the strength and frequency content of the stronger of the two input channels. That voltage is used to control the amplifiers in the *Dolby B* noise-reduction circuits.

But, while the Dolby B control signal may be ideal for noise-reduction companding, it is less than ideal for bias level control. In Dolby B, the control signal is derived without regard to the recording pre-emphasis curve. That curve shapes the frequency response of the signal that, after processing for noise reduction, reaches the recording head. As the bias level changes, the frequency response of the system also changes. The problem of varying frequency response during recording was dealt with in the original *Dolby* HX system by allowing the control signal to alter preemphasis at the same time that it alters bias. If everything was adjusted perfectly, the bias and pre-emphasis changes were supposed to balance each other out to provide the flat response and expanded headroom for which the system was designed.

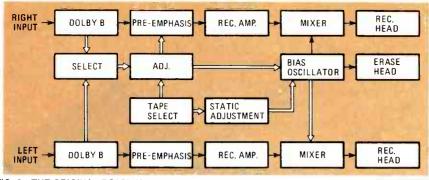


FIG. 2—THE ORIGINAL DOLBY HX system is shown in this block diagram. For the system to work, a control signal that is sensitive to amplitude and frequency content is required.

Self biasing and mutual biasing

While studying how *Dolby HX* works and the dynamics of magnetic recording in general, the engineers at Bang & Olufsen noted two phenomena. They noted that any signal fed to a recording head acts as bias for the signal to be recorded, with high-frequency signals being more effective as a bias than low-frequency signals. A signal even acts as a bias for itself (self-biasing), though that effect is quite small for audio signals. Nevertheless, self bias may be a small part of the reason why high-frequency signals require a much lower bias setting than low-frequency signals.

Much more important, however, is the biasing effect of the high-frequency content of an audio signal on the lowfrequency part of the same signal. While that effect is only a small part of the total bias applied to the recording head, it is enough to alter the bias conditions for low frequencies. That is shown dramatically in Figs. 3 and 4.

Figure 3 is a spectrum analysis of a single 300-Hz tone, recorded at a level of 250 nanowebers-per-meter (nWb/m). That is about 2 dB above the Dolby

calibration level-not an exceptionally high level. The bias on the recorder used was adjusted to a conventional level: slightly below optimum for the best low-frequency MOL. The tall spike at the left is the 300-Hz signal. The spike at around 900 Hz is the thirdorder distortion component and its level is about 46 dB below that of the 300-Hz signal, which corresponds to a thirdorder harmonic distortion level of 1.58%. Those are not unusual values for a good-quality cassette recorder. (In Figs. 3 and 4, each vertical division is 10 dB, while the horizontal scale is logarithmic and runs from 20 Hz at the left to 20 kHz at the right; key frequency points are shown at the top of the scope display).

In Fig. 4, the 300-Hz signal has been kept at the same input level, but a pair of tones with frequencies of 9 and 10 kHz have been mixed into the signal at a level some 50 dB below that of the lower-frequency signal. You'll note that there is a dramatic reduction in third-order distortion of around 6 dB, while the recorded level of the 300-Hz signal actually rises about 2 dB. The net third-order distortion is now 8 dB lower

than before (relative to the 300-Hz fundamental), or 0.63%. Both the reduced third-order distortion and the increased level of the recorded signal are caused by the additional biasing effect of the high-frequency signals; that additional biasing has raised the total *effective* bias level for the low-frequency signal to near its optimum value. The biasing of one part of an audio signal that is being recorded by another part of the same signal can be called mutual biasing.

It follows that the flat frequency-response obtained under static test conditions (using a single-tone test signal) may not be so flat when complex musical signals are recorded on tape. When such musical signals are recorded on a machine with fixed bias, the frequency response changes continuously due to mutual biasing, depending upon the instantaneous high-frequency content of the overall signal. Such dynamic changes in frequency response are present not only in cassette recorders, but in any tape recorder using high-frequency bias, such as studio tape decks and high-speed duplicating machines used to produce prerecorded tapes.

Dolby HX Professional

Dolby HX Professional was developed jointly by Bang & Olufsen and Dolby Laboratories. Although the principles of Dolby HX and Dolby HX Professional are similar, the aims of the two systems are different. In Dolby HX, the chief aim was to allow high-frequency signals to be recorded at higher levels on cassette tape. Dolby HX Professional has the more fundamental aim of keeping the active or effective bias constant, regardless of signal content. Active bias is defined here as the effective bias seen by each frequency in the audio spectrum. In keeping the active

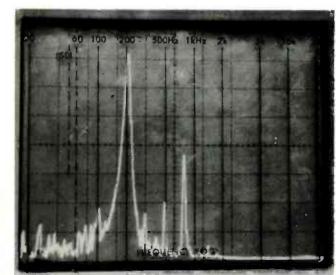


FIG. 3—SPECTRUM ANALYSIS of a single 300-Hz tone recorded at a level of 250 nanowebers-per-meter.

O LOU ZO I SOOKS HAYR ZA

FIG. 4—MUTUAL BIASING is demonstrated here. Two low-level, higher-frequency (9 and 10 kHz) tones were mixed with the 300-Hz tone.

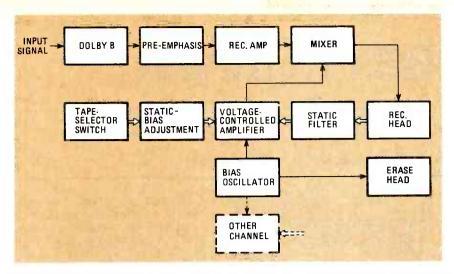


FIG. 5—BLOCK DIAGRAM of the Dolby HX Professional system. Unlike Dolby HX, this system measures the sum of the audio signal, including any preprocessing, and the bias at the record head.

bias constant, all of the recording parameters that would normally be affected by a change in bias under static conditions, will remain stable under the dynamic conditions that prevail during the recording of more complex music signals. For example, the frequency-response changes shown in Fig. 4, which were caused by mutual biasing, would not occur if active bias were kept constant for all audio frequencies.

Dolby HX Professional and Dolby HX are similar to the extent that both systems change the level of the bias signal dynamically, but the reason for doing it and the method used are different. As things worked out, all of the original aims of Dolby HX are fulfilled by Dolby HX Professional, but that is a secondary benefit of the latter system. Dolby HX Professional has other advantages over Dolby HX. It is a dedicated system: one that is completely independent of Dolby noise-reduction modules or other electronic circuits within the recorder. (Remember, the original Dolby HX system uses the same control voltage that was developed by the Dolby B noise-reduction circuits.) That makes the system suitable for professional applications, such as in studio recorders, as well as for use in high-quality cassette decks.

Figure 5 is a block diagram which shows how Dolby HX Professional works. Unlike Dolby HX, which only measures the audio signal to determine the bias compensation required. Dolby HX Professional measures the sum of the audio signal, including any preprocessing, and the bias at the tape head. The signal from the recording head is fed to a processing circuit consisting of an accurately designed static filter. Once that filter has modified the signal. the signal is rectified to form a control voltage that is an accurate replica of the total biasing effect of the processed audio signal, plus the bias at the record-

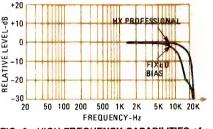


FIG. 6—HIGH-FREQUENCY CAPABILITIES of a cassette recorder equipped with *Dolby HX Professional* versus a conventional machine of the same quality.

ing head, or the "active bias." By measuring those signal components at the recording head, the system insures that the control voltage is always an accurate representation of the flux across the air gap in the recording head, regardless of signal strength or of any signal processing that may have taken place in previous stages. All processes and signals are carefully accounted for when the control voltage for bias compensation is derived from Dolby HX Professional. The control voltage is compared with a reference voltage that is adjusted for the static bias needed for the particular type and brand of tape being used, and a correction signal is generated. That signal is used to adjust a voltage-controlled amplifier that in turn alters the amplitude of the bias voltage supplied to the recording head. That maintains a constant active bias for the signal recorded. Once the circuit is correctly designed and installed, there is no need for any further adjustments for any tape or any signal-processing circuitry. Furthermore, since effective bias remains constant, no changes in pre-emphasis with changing bias are required; with Dolby HX, such changes in pre-emphasis were needed. Dolby HX Professional can be used with any noise-reduction circuitry currently available, or, for that matter, no noise-reduction circuitry at all.

If an audio signal having only low frequencies is fed to a recorder equipped with Dolby HX Professional, the circuit will "recognize" that fact and the bias level will remain fixed; since there are no high frequencies present in the signal, no bias-level change is required. If high frequencies are added, the Dolby HX Professional circuitry reduces the bias signal from the bias oscillator. The amount of reduction is determined by the passive filter, which monitors the entire signal applied to the tape head. That reduction in bias level will be almost exactly equal to the mutual biasing caused by the high frequencies. Thus, low frequencies see a constant bias level, but bias is reduced for high frequencies since they receive no added bias from the audio signal; low frequencies and high frequencies both see their optimum bias levels.

Should the signal consist of high frequencies only, the *Dolby HX Professional* circuit will reduce the bias from the oscillator even more; to a level that is very close to ideal for that type of signal. In that way, the original purpose of *Dolby HX*—namely, substantially increased headroom for high-frequencies —is also a characteristic of *Dolby HX Professional*. With *Dolby HX Professional*, the bias setting is adjusted to give the lowest possible distortion at low- and mid-frequencies, and the best MOL—compromises are no longer required.

Figure 6 shows the difference in highfrequency capability between a fixedbias machine and one with *Dolby HX Professional*. The machine equipped with *Dolby HX Professional* delivered an output that was attenuated only 1.4 dB with respect to our 0-dB reference (250 nWb/m) at 13.5 kHz. A conventional machine of equal quality, on the other hand, delivered an output that was attenuated 8.6 dB at the same 13.5 kHz frequency, or fully 7.2 dB poorer response at that relatively high frequency. The same type of tape was used in both machines.

Dolby HX Professional includes another refinement. It functions totally independently on each of the two stereo channels of a tape deck. That means that bias for each channel is always optimally adjusted for the signal being applied to *that channel*, even though the signal applied to the other channel may have a completely different content, requiring a totally different bias level.

At the moment, Bang & Olufsen is the only company making use of that refinement of the *Dolby HX* circuit. The new circuitry appears in their *Beocord Model 8002* cassette recorder. If the circuit proves to be as effective as claimed, you are likely to see it used by quite a few manufacturers before too long. **R-E**

HOBBY CORNER

This little IC can produce a great deal of sound.

EARL "DOC" SAVAGE, K4SDS, HOBBY EDITOR

IT IS ALMOST UNBELIEVABLE THAT SUCH a variety of sounds can be created by one small 28-pin IC. The Texas Instruments 76477 can generate everything from simple tones to very complex sounds. Best of all is the fact that you can create any sounds you desire by just adding a few resistors and capacitors

Before we get into how to make some of those sounds, perhaps we should mention some of the ways you can use the 76477. There are, of course, sound effects. While I can't guarantee it, I believe that this IC can produce any sound you have ever heard. Further, I know that it will produce many that you have never heard before!

The more mundane uses for the 76477 are obvious. They include such things as chimes, musical "instruments," door bells, noise generators-in fact, just about anything that requires sound. Now, let's see how to make the IC perform.

For now, we'll treat the 76477 as a "black box." To explain what actually goes on inside the IC would require a book, so we'll largely ignore the "how" and discuss the "what" of the IC.

First, there is a matter of the power supply. Texas Instruments' has made this easy by providing two ways to power the IC. A voltage regulator has been built into the 76477 so you can apply anything from 7.5- to 9-volts DC to pin 14. If you do that, you can also take 5-volts DC at up to 10 mA from pin 15 for use with external circuits! The other way to power the IC is to apply a regulated 5-volts DC to pin 15. In either method, pin 2 is the ground connection.

The next thing is the audio output. A small audio amplifier is built into the IC, but further amplification is needed to bring the signal level up to a useful level. The manufacturer suggests the simple two-transistor amp shown in Fig. 1. For a permanent installation, that will do nicely. However, I prefer a different approach for experimentation (and you will be doing a lot of experimenting to find new sounds). What I did was to use a small all-purpose amplifier such as the Radio Shack 277-1008, and connect it to the IC as shown in Fig. 2.

There is also a master switch for the IC: pin 9. If pin 9 is connected to

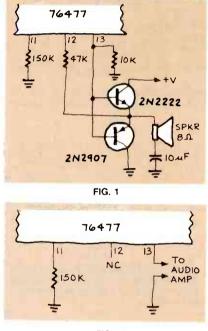


FIG. 2

ground, the IC is on; if it is connected to 5-volts DC, the IC is off.

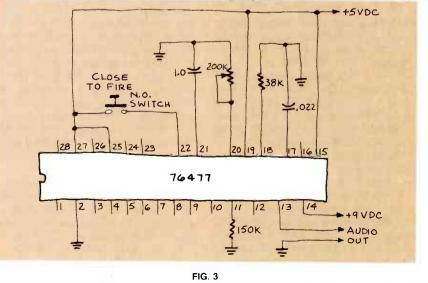
Now that we've discussed the power, audio, and master switch pins, that leaves us with just 21 more pins. Those 21 are the ones that are used to program the sounds you want. Seldom, if ever. will you use all of them at once but you do need to know what all of them do.

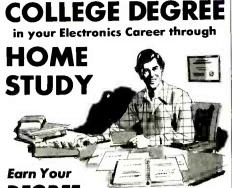
To make things a little easier, we'll group those 21 pins into functions.

Table 1 contains those functions and the pin numbers associated with each function. Note that most pins have an

TABLE 1

0	
	w Frequency Oscillator (SLF) 7500 ohms to 1 megohm
	500 pF to 100μ F
Voltage C	Controlled Oscillator (VCO)
18R	
170	100pF to 1µF
19	Duty-cycle (normally +5-volts DC)
22	Switch $(+5-volts DC = SLF;$
	0 = pin 16)
16	External (High voltage =
	Low frequency)
Noise	
5R	7500 ohms to 1 megohm
6C	150pF to .01µF
4R	Switch (43,000 ohms)
Mixer	
	6, and 27 See text
One Shot	
24R	7500 ohms to 1 megohm
23C	0.1 to 50 µF
Envelope	
1 and	28 See text
Attack/De	ecay
10R	Attack (7500 ohms to 1
	megohm)
7R	Decay (7500 ohms to 1
8C	megohm) Timing (.01 to 10µF)
00	
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appended letter (R or C) that indicates whether a resistor or capacitor is to be connected to it. In all cases, the component is connected from the indicated pin to ground. In addition, some of the pin numbers are followed by the normal range of values for the components that are to be connected to it.

The Super Low Frequency oscillator (SLF) has a normal range of 0.1 to 30 Hz, but it will operate at a much higher frequency if needed. The value of the resistor and capacitor attached to the appropriate pins determines the frequency. The SLF can also be used to 'modulate'' other IC functions such as the VCO and/or mixer.

The Voltage Controlled Oscillator (VCO) produces a tone. The basic pitch of the tone is determined by a resistor connected to pin 18 and a capacitor connected to pin 17. Remember, each of these components are connected between the pin and ground. In addition to the two components, the pitch of the tone is also determined by either the output voltage from the SLF or by a voltage applied to pin 16. Pin 22 is the switch: connecting it to +5-volts DC causes the SLF to be the controlling voltage; grounding that pin lets you control the frequency by applying a voltage to pin 16. In addition, the duty cycle of the output waveform can be controlled by a voltage applied to pin 19. Normally, pin 19 is connected to +5volts.

The noise source is switched by pin 4. Leaving that pin unconnected keeps the noise source off; shunting pin 4 to ground through a 43,000-ohm resistor turns it on. The resistor connected to pin 5 and the capacitor connected to pin 6 determines the sound quality of the noise. An external noise source can also be input at pin 3.

The mixer does just what you would expect-it allows you to mix the generated sounds in any combination. It is programmed by a binary control signal on pins 25, 26, and 27. The three pins gives you up to eight possible combinations, ranging from binary 0 (000) to binary 7 (111). A logic high (1) is +5volts DC and a logic low (\emptyset) is ground. The mixer code is as follows: $\emptyset \emptyset \emptyset =$ VCO: $\emptyset \emptyset 1 = SLF$: $\emptyset 1 \emptyset = Noise$: $\emptyset 1 1 =$ VCO/Noise: 100 = SLF/Noise: 101 =SLF/VCO/Noise: 110 = SLF/VCO:and |1| = inhibit all. For example, $\emptyset | 1$ (a binary 3) will cause the VCO and noise source to be mixed and passed on while inhibiting the SLF.

Let's pause for a minute and take stock. The SLF can generate a frequency-even a sound-that can be fed into the VCO or the mixer. The VCO can generate a sound that can be modulated by the SLF or an external voltage and fed into the mixer. The noise generator can also be fed into the mixer. continued on page 83

WE TAKE

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EBRUARY 1982

COMMUNICATIONS CORNER

Here's a window-mounted CB antenna that really performs.

HERB FRIEDMAN, COMMUNICATIONS EDITOR

BECAUSE OF THE MORE FREQUENT REstrictions on the use of rooftop antennas, many hams, SWL's, and CB'ers have had to resort to the use of socalled "portable antennas:" usually an inductively loaded short antenna that is intended to clip on to a window sill or ledge. Unfortunately, if it isn't some eagle-eyed busybody who can spot No. 50 wire 100 feet in the air and run to the building's manager screaming they have found the source of all TV interference (even when there isn't any), it's the lack of ready access to a window ledge that prevents many "cliff dwellers" from using some form of 'window antenna.''

The past summer, however, I discovered an effective "window antenna" for CB, one that really works. While it's designed specifically for CB, any hobbyist that's reasonably handy with a soldering iron can fabricate a model for a VHF ham band.

The little gem that's the subject of this month's column is the Gold Line *Intenna*. Before I get into the details, let me tell you how I discovered it. since this will give you a better picture of what it can do.

I was standing in a CB/auto shop in a 4-block tourist town just chatting, and commented that when I pulled into a motel I can't keep in contact with the rest of the family when they take the car down the road to some stores and shops to pick up supplies. One of the customers breaks in to say that he had the same problem until he purchased a CB antenna that sticks to the window. even those that don't open, and he usually can get coverage up to one mile or so depending on what floor his room is on and which direction the window faces. You can just bet I had to try this gadget, but unfortunately, it wasn't the easiest to locate because CB is not the market it once was. But I finally did get hold of one, and it really works.

The Gold Line Intenna is essentially a half-wave loaded (dipole) antenna combined with a reflected power indicator of a standard VSWR meter. The main section is a lightweight metal cabinet that measures approximately $5 \times 1.5 \times 2$ inches. On the back are two rubber suction cups that permit the box to be attached to the window.

Coming from each end of the box is a

32-inch long antenna element, which is actually a flexible wire with a suction cup on the end. On the bottom end of the box is a UHF (coaxial) connector to which the output of a CB transceiver is connected. The box is first attached to the window and then the antenna elements are stretched out and also attached to the window.

On the control box are two knobs labelled TUNE and the meter, which is arbitrarily calibrated from one to five. The knobs are actually mounted on the tuning slugs of variable inductors that are used as "loading" coils for the antenna elements.

The complete circuit is shown in Fig. 1. The dashed lines represent the metal cabinet. The two tuning coils are L1 and L2. The reflected power indicator is made up of L3, D1, R1, C1, and M1. Inductor L3 is really an RF pickup coil that takes a sample of the reflected power from the transmission line. Notice that loading coil L2 is actually between the antenna element and the metal cabinet. There might be some inclination to eliminate L2 and convert the design into a loaded 1/4-wave vertical. It could be done, but the tuning would be even more critical and the antenna would not be as efficient for keeping the signal low to the ground.

Using the Intenna

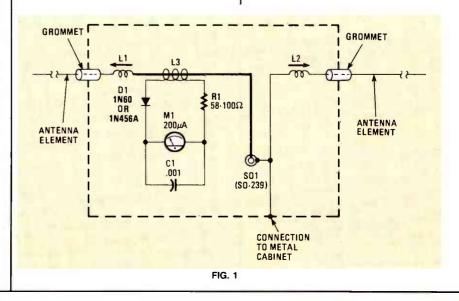
To use the *Intenna*, the CB rig is first connected to the control box with a

length of coax cable long enough to let you keep the rig at least 10-feet from the antenna. (If you build a similar device for some other frequency, try to keep the transmitter at least ¼-wavelength from the antenna. I realize this is impossible at the lower VHF frequencies, so all you can do is try and hope. Sometimes it will work, but other times the RF feedback will do strange things to the signal—particularly to the modulation.)

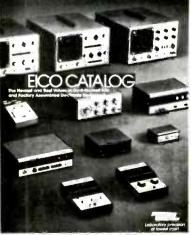
Then the top antenna element is secured to the highest point on the window using the suction cup. The wire is stretched out and the control box is secured to the window. Finally, the lower element is allowed to hang down and is then secured to the window. If the element is longer than the window, the wire is simply left hanging free or it can be taped to the wall.

The transmission line must be oriented at right angles to the antenna, or as close as possible to a right angle for at least 1/4-wavelength, or whatever you can squeeze in. A few strips of masking tape will easily support the RG-58 cable on a wall.

When the antenna is in place have someone key the rig. (You could do it yourself but it might prove somewhat difficult to adjust the antenna if you're also keying the transmitter.) Adjust each of the two tuning controls for the lowest meter reading. Step away from the antenna when you check the meter







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108 New South Road Hicksville, N.Y. 11801 reading (your body could affect the tuning). It will take a few tries to get it right, but eventually the meter will indicate near zero, or somewhere below 1/3 scale. That's it, the *Intenna* is ready for use.

As I said, CB gear isn't that easy to locate, so if you want information on pricing and availability of the *Intenna* write to Gold Line, P.O. Box 115, West Redding, CT 06896. **R-E**

HOBBY CORNER

continued from page 81

Since the output of the mixer can be any combination of those, you can see that it can produce very complex sounds. Even so, we can still process the sound further before it gets out of the 76477.

The signal from the mixer is fed to an envelope generator where it is shaped. Control of the envelope is determined by the binary signal on pins 1 and 28. Since there are only two pins, we are limited to just four control signals: $\emptyset \emptyset =$ VCO; $\emptyset 1 =$ mixer; $1\emptyset =$ one shot; and 11 = VCO alternating. For example, $1\emptyset$ (binary 2) would cause the one shot to shape the output from the mixer. Incidently, the pulse duration of the one shot is determined by the resistor/capacitor combination on pins 23 and 24; its maximum duration is about 10 seconds.

Before the signal gets to the audio amp, there is one more opportunity to process it! You can set the speed of the leading edge (attack) and the trailing edge (decay) of the envelope by connecting resistors to pins 10 and 7 and a timing capacitor to pin 8.

Finally, the signal is fed to the builtin audio amplifier and out of the 76477. With all of those possible combinations, is it any wonder that this little IC can produce very complex sounds?

Of course, there is much more that could be said but you have enough to go on to create your own sounds using the 76477. To give you a start, the circuit in Fig. 3 can be used to generate a "laser gun" sound. **R-E**



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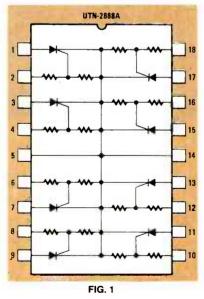
CIRCLE 33 ON FREE INFORMATION CARD

STATE OF SOLID STATE

SCR arrays and bar-graph displays ROBERT F. SCOTT, SEMICONDUCTOR EDITOR

LET'S BEGIN THIS MONTH WITH A LOOK at two new SCR arrays from Sprague. They are the UTN-2886B and UTN-2888A monolithic multiple-SCR arrays designed for coupling microprocessors, demultiplexers, and similar devices to high-current loads such as relays, solenoids, and lamps. The devices feature low input-current; TTL, LS-TTL, and CMOS compatibility; 35-volts minimum forward-blocking-voltage, and a 2-amp in-rush-current capability. They can be used with either full-wave or half-wave power sources.

The UTN-2886B consists of six individual SCR's and two pairs of paralleled SCR's (pins 1–16 and 8–9). Each separate SCR can handle 250 mA—the paralleled units at pins 9 and 16 handle 500 mA—at $+50^{\circ}$ C ambient temperature. Power dissipation can be increased by attaching an external heat sink. The UTN-2888A array contains eight isolated SCR's (see Fig. 1), each capable of continuous and simultaneous operation at $+50^{\circ}$ C.

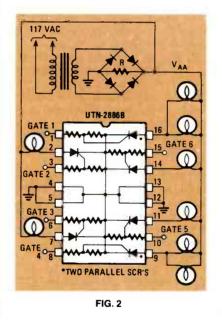


Characteristics of any individual SCR (maximum limits are given unless otherwise specified) are: Forward blocking-current. 50 μ A; gate-to-anode leakage. 250 μ A; forward ON-voltage. 1.2 volts at +25°C and 1.15 volts at +55°C; gate trigger-voltage. 2.5 volts; gate trigger-current. 300 μ A; gate OFF-current. 10 μ A; holding current, 10 mA at 0°C and 5.0 mA at +55°C, and anode

OFF-voltage, 400 mV minimum.

The devices operate from unfiltered DC from half-wave or full-wave-rectified sources. They cannot be used with raw AC. In operation, the SCR is triggered on by applying a positive voltage to its gate. It continues to conduct, even though the gate voltage is removed or made slightly negative, until the anode-cathode voltage is reduced to below 400 mV.

The diagram in Fig. 2 shows a typical application, using the SCR's as lamp



drivers. When using multiple SCR's and a common V_{AA} supply, gate-toanode leakage currents (I_{GA}) can hold the supply voltage above the anode OFF voltage, $V_{AK(OFF)}$, and prevent proper turn-off. Resistor R is connected across the power supply to insure proper SCR operation.

The maximum value of the resistor can be calculated from the following formula:

$$\frac{400 \text{ mV}}{(n-1) 250 \ \mu \text{A}}$$

where n is the number of SCR's being used in the system. Be sure to add 2 to the SCR count if you use the paralleled devices connected to pins 9 and 16 of the UTN-2886B.

For further information, contact Sprague Electric Co., Semiconductor Division, 155 Northeast Cutoff, Worcester, MA 01606.

Circuit of the Month

Bar-graph displays are popular as replacements for analog meters in radios, audio equipment, etc. The bar-graph circuits are simple and easily adapted for use with existing equipment. The analog signal needs little processing before being fed to the bar-graph display.

The display may consist of a number of discrete LED's, arranged in a linear array, or you can use a 10-element device such as the HDSP-4820, HDSP-4830 or HDSP-4840 series from Hewlett-Packard. In those devices, the LED's have been matched for color and brightness. A cut-away drawing of a 10-element bar-graph is shown in Fig. 3. The light from each LED element is "stretched" by diffusion and reflection to form individual elements. The array series is available in standard red (HDSP-4820), high-efficiency red

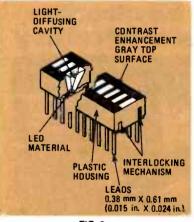


FIG. 3

(HDSP-4830), and high-efficiency yellow (HDSP-4840); the pin-outs for all three are the same.

There are a number of rather simple display-drivers for use between an analog or digital source and the LED arrays. Some display drivers or decoder/drivers provide a logarithmic display with 2- or 3-dB increments. Other drivers provide a linear display with either fixed or variable voltage increments. The drivers are made by, among others, National, Texas Instruments, Exar, Siemens, and AEG-Telefunken. *continued on page 86*

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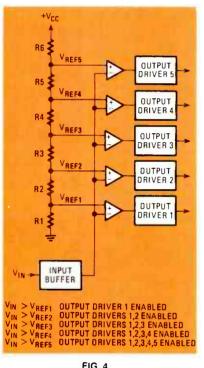
CIRCLE 14 ON FREE INFORMATION CARD

STATE OF SOLID STATE

continued from page 84

A typical bar-graph decoder consists of a set of voltage comparators and a precision voltage-driver that supplies reference voltages to the individual comparators. Figure 4 shows a representative analog-input 5-stage bargraph-display decoder. The decoder drives the display in a logarithmic or linear manner, depending on the scaling of the reference-voltage network.

Some decoders provide displays in either bar or dot-position modes. In the bar mode, the decoder lights all the LED's whose comparator voltagethresholds are below the level of the applied input signal. In the dot-position mode, the display shows a moving dot





whose positive is determined by the highest-threshold-voltage comparator triggered by the applied signal voltage.

Figure 5 shows a circuit that displays signal levels from zero to five volts in either a dot or bar pattern. It is based on the Hewlett-Packard 10-element bargraph display and National Semiconductor's LM3914 dot/bar display driver. In this circuit, the display is a moving dot when the driver's MODE pin (pin 9) is tied to pin 11 and a bar graph when pin 9 is tied directly to $+V_{CC}$ at pin 3. (The National Data Book recommends leaving pin 9 floating for a dot display. Try it both ways.)

The incremental voltage-steps and thus the full-scale voltage, are determined by the voltage developed by the

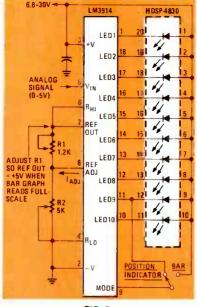


FIG. 5

internal reference source between pins 7 (REF OUT) and 8 (REF ADJ). That constant voltage forces a constant current through R1 and through an adjustable resistor. R2. Therefore the total voltage between the REF OUT pin and ground can be easily determined by using the following formula:

 $V_{OUT} = V_{REF} (1 + R2/R1) + I_{ADJ}(R2)$; where $V_{REF} = 1.25$ volts and I_{ADJ} is 75 μ F typical. 150 μ A maximum.

The current drawn from the voltagereference pin determines the LED current. Approximately 10 times this current passes through each lighted LED. The LED current, and thus the brightness, can be adjusted by resistor R2. With a 7-volt supply, the LM3914 dissipates approximately 110 mW in the DOT

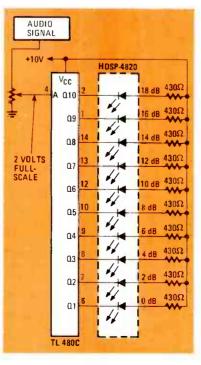


FIG. 6

mode, and around 720 mW in the BAR. mode.

A low-cost VU meter is shown in Fig. 6. The bar-graph display is driven logarithmically by a Texas Instruments TL480C 10-step logarithmic analog level-detector. The device uses ten comparators to detect the level of the analog signal applied to the analog input (pin 4). Output Q1 is switched to a logiclow, turning on the 0-dB indicator LED when the input signal reaches approximately 218 mV. The comparators switch

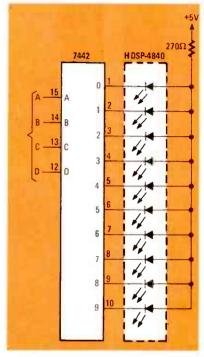


FIG. 7

in 2-dB increments until they are all low at an input value of approximately 1.732 volts.

In some systems, a BCD (*Binary* Coded Decimal) signal is the most convenient way to drive a bar graph. Figure 7 shows a 7442 BDC-to-decimal (1 to 10) decoder connected to drive a 10-element bar-graph display in the DOT mode. In the case of that circuit, the LED's light singly to indicate the level of the BCD signal.

The material for these circuits came from National Semiconductor's Linear Databook and Hewlett-Packard's Bar Graph Array Applications. If you want to see other applications such as "chaining" for dot or bar displays of 20 or more LED elements, or interfacing to a 6800 or 8080A microprocessor, we recommend those two publications. The Linear Databook is available for \$9.00 from National Semiconductor Corp., Literature Distribution, 2900 Semiconductor Drive, Santa Clara, CA 95051. The Hewlett-Packard literature is available from Hewlett-Packard Components, 640 Page Mill Road, Palo Alto, CA 94304; ask for Application Note No. 1007 and the 10-Element Bar Graph data sheet. R-E



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COMPUTER CORNER

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LES SPINDLE*

WHEN A DISCUSSION ABOUT COMPUTERS of the future occurs, the conversation often discloses bizarre fears about a dehumanized society. In these cynical scenarios, students acquire their education awkwardly because they interact with a machine rather than a human teacher; patients have difficulty recovering from illnesses because their diagnosis is done by an unfeeling computer rather than a sensitive doctor, and social interaction suffers because people seldom have to leave their home for any purpose. Such predictions of doomsday are, of course, ridiculous. If anything, a computerized society is a more efficient society-with more leisure time for people to spend on social and recreational pursuits. A world linked via computer terminals has to do with more communication among more people-not social repression.

Computer data bases will provide much of the impetus for these improved communications by enabling a world of information to be instantaneously ac-

*Associate Editor, Interface Age Magazine

cessible to just about anyone. Whether practical or frivolous, vocational or recreational, social or clinical, information is immediately available from a data base. A computerized data base is essentially an electronic information file that is accessible to consumers that have a computer terminal and the necessary data communications peripherals.

They are currently over 600 commercial data bases in the U.S. alone, providing students, lawyers, farmers, computer hobbyists, stockbrokers, home consumers, and all types of business people with a wealth of vital information. The essential benefit of accessing data this way is obvious: it eliminates the time and hassel involved in searching for information sources, scanning through the sources to find specific information, and copying or transcribing that information. With a terminal and a data base, the desired information is retrieved and printed out within minutes after the appropriate information has been keyed in.

To illustrate how a data base can aid the average computer user, let's con-

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•	PERSONAL CALENDAR & NOTEBOOK
	FIG. 1

sider two popular data-base networks geared towards the home user or small businessman: The Source, provided by Source Telecomputing Corp. (a subsidiary of Telecomputing Corporation of America, 1616 Anderson Rd., McLean, VA 22102) and CompuServe, Inc. (formerly MicroNet), 5000 Arlington Center Blvd., Columbus, OH 43220.

Both services were introduced in 1979 and both essentially thrive on the same premise: providing sophisticated networking capabilities to the average consumer at an affordable cost. The technology is scarcely new. Computer networking and timesharing have been used as methods of doing business for large corporations and government agencies for several years. What is revolutionary about the home-computer network concept, however, is that it demonstrates the potential for a world linked up electronically in the most widespread mass-communication network conceivable. Like television, radio, the telephone, and every other mass-communication technology that has come along, the home computer will not become a true social phenomenon or booming economic industry until it is affordable by the average consumer. The computer is now affordable to a large number of consumers, but the telecommunications technology must catch up.

What makes CompuServe and the Source work is the fact that it is offered at a discount rate during hours that are not marketable for "nine-to-five" business transactions. During peak hours, the mainframe computers are heavily in demand by corporate clients. By tapping the heretofore little used evening and weekend hours, which coincide with the ideal time for consumer usage, the networks can offer the mainframe computer capabilities—and the special consumer data bases—at a much lower price.

The CompuServe network is available between 6 PM and 5 AM weekdays, all day weekends and holidays. Cost is \$5 per hour. At all other times, the rate is \$22.50 per hour. For The Source, a onetime registration fee of \$100 is required. From 7 AM to 6 PM on Monday through Friday, the fee is \$15 per hour. From 6 *continued on page 92*



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CIRCLE 82 ON FREE INFORMATION CARD

COMPUTER CORNER

continued from page 88

PM to 12 AM on weekdays and all day on weekends, the cost is \$4.25 per hour. From midnight to 7 AM daily, the cost is \$2.75 per hour.

For hookup, the basic requirements for both networks include a terminal, a 300-baud modem (communications interface) and a telephone. Upon registration, the user receives an information packet including a user's manual, an index of data bases and services, an account number, and a private password. The system is accessed by dialing an assigned phone number, waiting for the tone to commence, placing the receiver in the modem, then entering the I.D. number and password. The modem translates electrical signals into sounds, and vice versa, to communicate with the host computer. or other computers plugged into the network. When a set of instructions appears, the user is told how to proceed in order to access the services or data bases he desires. Now he is ready to begin his interaction with the network.

By providing the possibility of terminal-to-terminal communication among the various subscribers, the network





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has the potential to achieve a communication method with mass appeal. Beyond providing a fun communication source among users in true CB-radio tradition, the network can function as a virtual electronic library and shopping center. The subscriber can browse through merchandise catalogued into specific programs, and even order the products directly through the terminal. If computer art is of interest to him, for example, he can browse through CompuServe's art-gallery listing and order a computer-generated picture for future delivery. Also available are programs on stock market data, filing capabilities for information storage, financial advice, games, educational aids, theater and movie reviews, news, weather, and sports.

The advantages of such a time-sharing network for the home user or small business are numerous and varied. As the programs and subscriber services become more diverse and useful, it is not hard to imagine that having a personal computer in every home, hooked up to major mainframe data bases, will eventually be as commonplace as a television or a telephone.

Electronic mail, for instance, is no longer a service strictly for major corporations. Exciting new directions for inexpensive mass communication at all levels of business and consumer usage are now evident. Even personal correspondance can be less expensive through a computer network than by a long distance phone call, although the personal interaction, in some cases, is obviously preferable.

The process of ordering merchandise can be streamlined. The home consumer can do tedious shopping for groceries or household goods without ever leaving the home. Such services, of course, are still somewhat limited. but expansion in this area is certainly inevitable as more homes become equipped with computer terminals.

Both CompuServe and The Source emphasize their extensive stock market services. CompuServe's *MicroQuote* includes information on over 32,000 stocks. A recent survey by The Source indicated that a large percentage of its subscribers sign up for the stock market facilities alone.

Because new data bases are constantly being added, both networks distribute a monthly news magazine to subscribers, detailing the latest services. The master index for CompuServe lists over 350 separate data bases with a wide range of subjects: adventure games, child care, want ads, tax advice, gold prices, text from various newspapers including the *Washington Post*, and photography. Among services described in the latest issue are a guide to admissions information and entrance *continued on page 102*



Solve your perplexing paging problems...

No one microphone is correct for every paging situation—different circumstances require different and sometimes unique solutions. That's why Shure has drawn on its unparalleled knowledge of microphone design to create a series of paging microphones specifically designed to reduce or eliminate the most common paging problems.

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use Model 414B; • an excelent limited space model that can stand up to extreme abuse is Model 507B; • anc if these won't solve your problem, read on.

... by the book

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CIRCLE 11 ON FREE INFORMATION CARD

NEW IDEAS

Automobile ignition substitute

HERE'S A DEVICE THAT CAN HELP YOU find out what's wrong when suddenly one morning your car refuses to start. The ignition substitute described here can even be used to verify that repair work done to your engine by someone else has been done correctly.

Basically, the ignition substitute provides a constant power-source for the ignition coil. Its frequency (0.5-1.0 kHz) is that used by an 8-cylinder engine with an idling speed of 650 RPM, and the unit provides a rapid spark at a 17% duty cycle, while nonetheless staying within the power-dissipation limits of the components.

Construction is straightforward, and any method can be used. The circuit, shown in Fig. 1, consists of a 555 timer IC configured as an astable free-running multivibrator that is used to drive a high-current NPN transistor, such as a 2N6384. (That transistor should be heavily heat-sinked because it may be drawing several amps over quite a long period of time.)

The coil ballast can be from 0.68 to 6.5 ohms, depending on what's available. The 2.5-ohm, 20-watt ballast shown in Fig. 1 works well. All the other resistors can be either $\frac{1}{4}$ - or $\frac{1}{2}$ -watt devices, and the capacitor between pins 1 and 5 of the 555 can range from 0.01 to 0.05 μ F. Do not omit the 100-volt, 0.05 μ F capacitor across the transistor; it prevents voltage spikes from damaging the device. Use 4-footlong clip leads to obtain power directly from the automobile's battery; that length is suggested for convenience.

You can use either your car's own

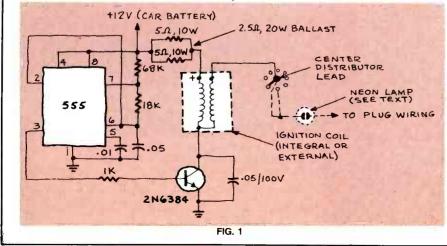
ignition coil, or a separate one. If you choose the latter, be sure to disconnect the one in the vehicle. A good coil will produce a spark between the high-tension lead and ground about 1/4- to 1/2inch long, and a strong bright spark across a plug with a gap smaller than 0.040-inch. That, by the way, is the first test for ignition problems.

To determine whether there's a problem with the car's distributor, supply a spark derived from the ignition substitute to the center distributor lead leading to the rotor and slowly rotate the distributor cap. Crank the engine and, at some point, the engine should catch and run. If the engine cannot be started, but seems to be trying to, the problem is probably in the timing chain, valves, camshaft, or elsewhere. If the engine doesn't even try to start, inspect the rotor, cap, wires, and plugs for damage. Once the ignition problem has been found and corrected, the normal procedure for setting the timing and dwell should be followed.

Do not attempt to adjust the distributor using the ignition substitute. That can not be done because the spark the substitute produces is slightly different from that produced under normal conditions.

Although designed for an 8-cylinder engine, this device can be used with other types. In addition, a neon bulb can be added to the circuit to verify the presence of a spark, and, in fact, can be used as a timing light if placed close to timing marks that have been painted white with fingernail polish.

—Stan K. Stephenson II



NEW IDEAS

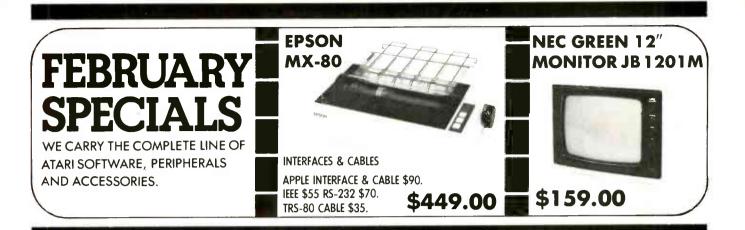
This column is devoted to new ideas, circuits, device applications, construction techniques, helpful hints, etc.

All published entries, upon publication, will earn \$25. In addition, Panavise will donate their model 324 Electronic Work Center, having a value of \$49.95. It combines their circuit-board holder, tray base mount, and solder station (see photo below). Selections will be made at the sole discretion of the editorial staff of **Radio-Electronics**.



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SERVICE CLINIC

Solid-state vertical-sweep circuits

JACK DARR. SERVICE EDITOR

TUBE-TYPE VERTICAL-SWEEP STAGES ARE familiar to all the old-timers, and to most of the younger gang. They generally consisted of a high-mu triode and power pentode (often in the same 'bottle''), a matching transformer, and

fairly-high-impedance yoke. The a tubes were coupled by a feedback loop. so they served both as oscillators and as output stages.

Solid-state sweep stages perform the same functions, of course, but their designs may vary from manufacturer to manufacturer. Furthermore, the symptoms that they show when something goes wrong are not usually what you used to see in tube-type circuits-about the only familiar one is the thin white horizontal line indicating a total lack of vertical sweep.

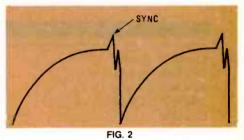
A solid-state sweep circuit uses several small-signal transistors as oscillators, wave-shapers, drivers, etc. The output stage is a type familiar to anyone who has worked on audio equipmenta two-transistor "stacked totem-pole," or complementary-symmetry type. The yoke is fed from the junction of the two power transistors; it's usually a fairlylow-impedance type, driven directly through a very large electrolytic coupling capacitor.

A variation on that is used in some sets-they use a complementary-symmetry pair, with a dual-polarity power supply. Figure 1 shows a circuit of that type, used in a General Electric QB chassis. The output pair there is Q267 and Q268. Note that there is no large coupling-capacitor. The reason for that is that the junction of the two transistors is at ground potential, or very close to it

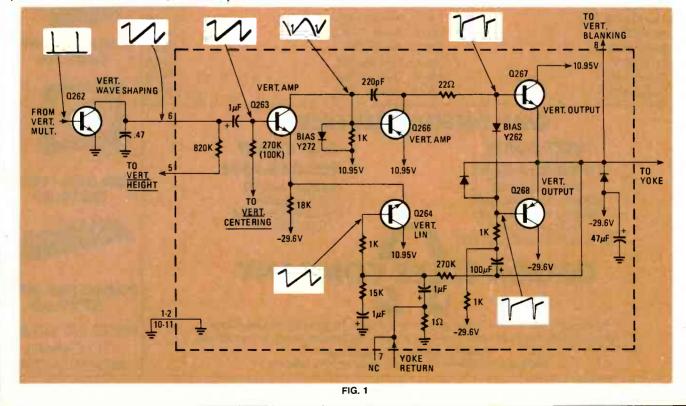
Both circuit-types do the same thing they deliver voltage, in the form of a square wave, to the yoke. When you feed such a square wave into an inductance (like the yoke) you get a sawtooth wave, which is what's required.

Two-transistor circuits seem to be almost universal these days. In some early sets, though, a single transistor was used, with a "loading choke" for matching and for the DC supply, and a blocking capacitor. That circuit operated in the class-A mode, like the tube types; the two-transistor circuits operate in class-B fashion, with each device amplifying one half-cycle of the driving signal.

An oscilloscope is invaluable for working with transistor sweep-circuits. Begin with the oscillator, make sure it's running-and on frequency. There may be several amplifier (and other) stages between the oscillator and the output stage, so just follow the signal through them to the yoke.



You'll find one waveform at some points that you'll never see in tube-type circuits-a perfect sawtooth. With tubes, the signal on the oscillator grid (Fig. 2) shows the familiar rising curve with a spike at the top. Solid-state oscillators, on the other hand, show a



series of pips, as can be seen in Fig. 1. A sawtooth wave is developed from them by a stage called a "wave shaper." The normal drive waveform on the bases of the output transistors is almost the same as in tube sets—a trapezoidal wave. In late-model sets you may find all of the stages, except for the output transistors, on a single IC. When checking this type of circuit, be sure that all the waveforms are correct, that all of the external components are OK. and that the proper DC supply-voltages are present before changing the IC.

Yoke returns

The vertical-yoke winding must return to ground, or to a level defined as a common ground. In quite a few circuits, a large blocking-capacitor is used between the bottom end of the yoke and ground. If it opens, there's no sweep, so always suspect it in such cases.

In many sets using circuits like the one in Fig. 1, you'll find the yoke return coming back to the module and running to ground through a 1-ohm resistor (R288). You must see a sawtooth waveform at the top end of that resistor; it's developed by the sawtooth current flowing through it. That current is fed to the base of a separate transistor, Q261, labelled "Vert. Lin." The reason for that is that all solid-state stages of that type are slightly non-linear! They develop a slight bend in each half of the waveform, causing what is known as "S-bending." The extra transistor and the feedback network are used to correct for that, and the end result is a perfectly linear sawtooth wave. If you run into a linearity problem, always remember to check that feedback resistor. Leakage and other defects there can cause some weird problems.

The feedback from that stage is applied to the emitter of the first vertical amp. The troubleshooting logic is clear: If the problem is linearity, go to the stage that's supposed to correct it.

Dual-polarity DC power-supplies can cause a symptom that would not be believed by a strictly tube-type technician. I know that the first time I saw it I said, "That *can't* happen!" The symptom is a perfectly good raster, but with only the top half of the picture present. That's all—there's no severe compression or other effect, as there would be in a tube set with a similar problem. The bottom half just *isn't there*.

The problem is either a loss of the DC supply to the "bottom" transistor (Q268 in Fig. 1) or the loss of the -29-volt supply to it. One other possibility exists: In the set where I first saw the problem, the vertical-sweep circuit was on a small module. The output transistors were clamped to one end of it and their leads went into individual socketpins on the chassis. That was fine, ex-

cept for the fact that the emitter lead of the "bottom" transistor was not in its socket. It was leaning on the outside of the pin, and making contact intermittently. Plugging the module (and the transistor) in properly restored things to normal. You may find that type of situation in sets with bad solder joints at one of the pins of the output transistors, so check for it.

In quite a few popular circuits, you'll find the signal passing through little low-voltage electrolytic coupling capacitors. If you're getting no signal to the output stage, check to make sure that one of them isn't open. You can do that instantly with a scope: If a signal's going in but not coming out, you've found at least part of your problem.

Here's another problem that's also common in audio circuits: If your symptoms include distortion at the center of the screen—compression or stretching of the picture—check to make sure that the bias diodes in the output stages are good. Failure of one or more of them can cause crossover distortion in audio, and also in TV sweep circuits. **R-E**

SERVICE QUESTIONS

VCR PLAYBACK PROBLEM

I have a Toshiba V-8000 VCR and an RCA CTC-25. If I record a program off the air, it will play back nicely. However, if I try to play back a pre-recorded tape, I can't get a stable picture. I had the set checked, and it's OK. Tried out the VCR with some newer sets, and it played fine. What's going on?—D.B., Memphis, TN

I've heard of that problem before. Apparently in older sets the horizontal AFC time-constant was considerably longer than the one used today, to make the set a little less susceptible to such things as airplane flutter, etc. However, the sync on pre-recorded tapes requires the shorter time constant. As a fix, try varying the size of C82/R124, or C23/R126 on the grid of the AFC tubes.

(Feedback: We tried both capacitors, and wound up taking them out completely. When we did that, the tapes played back perfectly, but we lost the sync on a broadcast signal after about five minutes. We took care of that by connecting the capacitors to a DPST switch so that they can be switched in and out of the circuit. Everything works fine now—thanks!)

COLOR BARS

This Magnanox T-982-12 makes its own color bars! The bars—blue, green, and red from left to right—divide the screen into thirds and are present all the time. When I





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This new 15MHz dual-trace mini-scope was designed by B&K-PRECISION engineers to respond to the special needs of field engineers... a *mini-scope* with *lab-scope* features.

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The rugged 1420 features dual-trace operation and an honest 15MHz response, with useful response beyond 20MHz. An efficient rectangular CRT displays waveforms with good readability under all field service conditions.

There is no sacrifice of features or performance for compact size. The 1420 has 18 sweep ranges from 1 μ S/div. to 0.5S/div. in a 1-2-5 sequence; variable between ranges. Sweep magnification is X10, extending the maximum sweep rate to 100nS/div. For use with computer terminals or video circuits, a video sync separator is built in. Automatic selection of chop and alternate sweep modes is provided, as is front-panel X-Y operation.

The Model 1420 measures only 4.5 X 8.5 X 12", weighs 7.75 lbs., with batteries and comes with two 10:1 probes.

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FEBRUARY 1982

use a color-bar generator, all of the colors are wrong. Nothing I've tried, including purity, will get rid of the bars. Incidently, the waveforms out of the demodulator module are normal. Can you shed some light on what's going on?—M.D., Aberdeen Proving Ground, MD

I ran into something like this in a Philco. In that case half the screen was blue and the other half gold. Eventually found the source to be an open bypasscapacitor on the screen grid of the second color-bandpass amplifier. Since you've checked the DC-power supply with a scope, try checking all of the bypassed points in and around the coloramplifier stages, bandpass, etc. There should be no signal on any bypassed point, so look for a signal where it does not belong.

Since the three color-amplifier stages are separate, the cause of this should be something that is common to all. For example, check to see if there's anything odd in the video feed to those stages.

MORE ON "REDUNDANT" CAPACITORS

Here's a test that was shown to me by a technician for Hollander and Co. of St. Louis, Mo. This is for those Zenith chassis that use several parallel



shunt-capacitors in the collector circuit of the horizontal-output stage. If the high voltage measures too high, one or more of those capacitors may be bad. To test, connect new capacitors in parallel with the ones already in the set. The high voltage should drop to normal, or lower. Now, disconnect the original capacitors one-by-one until you find one that has no effect on the high voltage—that's one of the bad ones. That procedure is almost as fast as using a capacitor tester, and it can be done under "normal" operating conditions.

Most of those capacitors are Zenith 22-5001's. Those are .0018 μ F, 1600-volt units, but they use a special dielectric; ordinary 1600-volt units won't work. For an inexpensive replacement, use a Sprague PP16-D18. The Sprague catalog lists a lot of those; they are *Metfilm* and *Filmite* types.

DEAD SHORT

This CB is driving me crazy. It's got a dead short somewhere but I can't find it. My tests show no short to ground anywhere. Any ideas that you might have would be appreciated—W.G., Kingston, NY

We've run into something like this before. Check the "protective diode" used in the crowbar circuit; it's a Zener. It will test out OK on an ohmmeter, but when voltage is applied it becomes a "0-volt Zener!" A crowbar circuit is exactly what the name implies: If anything goes wrong, it is supposed to break down and short the DC supply to ground. That one may be getting a little bit too eager and breaking down too quickly.

FLYBACK SUBSTITUTE

I asked you about a replacement transformer for a Broadmoor 6911-C, and you said you didn't have any information on it. I wrote to Thordarson-Meissner, and they had one! (It's a FLY-677, if anyone else needs one.)— James E. Higley, Hanover Park, IL.

LEAN ON IT

I wrote you a while ago about a Sharp TV-96P with weak audio, and thought you'd like to know that I fixed it. I tried everything I could think of, and everything *you* suggested, without any luck. Then, while checking around the second IF tube, the sound suddenly came in loud and clear.

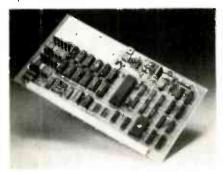
It seems that this happened when I inadvertently leaned on the chassis, which suggested a bad solder-joint. Even though the joints looked good. I reworked about 15 of them and turned the set back on. It worked perfectly. A simple—but not immediately obvious—solution.—L.P., Potomac, MD R-E

with some our destands -

NEW PRODUCTS

For more details use free information card inside back cover.

COLOR VDG CARD, the ColoRAMa, is a color video display generator/controller for SYSTEM-50 (SS-50, SS-50C) 680x computers. It permits software selection of any of 11 different display formats, including eight-color semigraphics, two-and four-color graphics, and two-color alphanumerics.



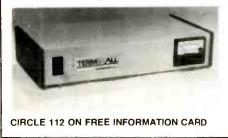
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Two- and four-color displays may also be switched between primary and complementary color sets, either under software control or from the keyboard. Resolution of full-graphic displays ranges from 64×64 picture elements (pixels) to 256×192 pixels.

All of the 64 characters of the standard ASCII character subset are generated by the on-card character generator. The *ColoRAMa* is configured with one kilobyte of display RAM. One-kilobyte of RAM accommodates alphanumeric displays, eight-color semigraphic displays and two full-graphic displays modes. The *ColoRAMa* is priced at \$219.95. — **Percom Data Company, Inc.,** 211 N. Kirby, Garland, TX 75042.

INTEGRATED HARDWARE/SOFTWARE

SYSTEM, Terminall (Model T1 and Model T3), converts the TRS-80 microcomputer (Model I or Model III) into a state-of-theart communications terminal. It includes all the necessary computer-interfacing, audio-demodulating, AFSK tone-generating, and transmitter-keying hardware integrated in one cabinet. That reduces equipment interconnection to a minimum and allows the operator to be on the air



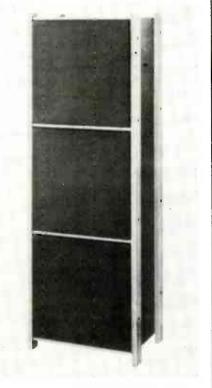
receiving and transmitting Morse code or RTTY in minutes.

The user can plug it into the receiver headphone jack and copy Morse code, Baudot, or ASCII; plug it into the CW key jack and send Morse code, or attach a microphone connector and send Baudot or ASCII using audio tones (AFSK). In addition, the *Terminall* is expandable: disk-based (mailbox) RTTY software may be added at any time.

Terminall model T1 requires Model I TRS-80, 16K RAM and Level II BASIC. Terminall model T3 requires Model III TRS-80, 16K RAM, and model III BASIC.

Terminail model T1 or model T3 are priced at \$499. — Macrotonics, 1125 N. Golden State Blvd., Suite G, Turlock CA 95380.

SUBWOOFER, model AS-1320, is designed to add bass to an existing small system, but will enhance the bass response of even 10-inch and 12-inch systems. By extending the low-frequency range of the audiophile's present system, the model



CIRCLE 113 ON FREE INFORMATION CARD

AS-1320 will reduce the bass distortion that occurs when the wing speakers try to reproduce the low-bass notes. A ported cabinet alignment uses computer-aided



In general, spring reverbs don't have the best reputation in the world. Their bassy "twang" is only a rough approximation of natural room acoustics. That's apity because it means that many people will dismiss this exceptional product as "just another spring reverb". And it's not. In this extraordinary design Craig Anderton uses double springs, but much more importantly "hot rod's" the transducers so that the muddy sound typical of most springs is replaced with the bright clarity associated with expensive studio plate systems.

Kit consists of circuit board, instructions, all electronic parts and two reverb spring units. User must provide power $(\pm 9 \text{ to } 15 \text{ v})$ and mounting (reverb units are typically mounted away from the console).

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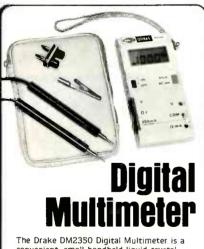


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design techniques to maintain an extended low-frequency response with as flat a response as possible. A six-inch tuned vent is loaded into the floor for even better response. The driver is mounted near the floor so that the first boundary cancellation is above the frequency limits of the subwoofer.

The model AS-1320 requires either the Heathkit model ASA-1320-1 Passive Crossover (mail-order priced at \$44.95) or the Heathkit Electronic Crossover, model AD-1702, (mail-order priced at \$194.95). A complete step-by-step instruction manual makes the subwoofer kit a one-evening project. The model AS-1320 is mail-order priced at \$299.95. — Heath Company, Benton Harbor, MI 49022.

DESOLDER PUMP, model DP-1, features all-metal construction with precision components for maximum reliability and ease of operation. Compact size facilitates comfortable, one-hand operation. Suction is precisely regulated for efficient solder removal without damage to



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delicate circuitry. Self-cleaning at each stroke, the model DP-1 is quickly disassembled without special tools for maintenance or repairs. The rugged Teflon tip is easily replaced. The model DP-1 is priced at \$10.95. —**OK Machine and Tool Corporation**, 3455 Conner Street, Bronx, NY 10475.

STEREO MIXER/PREAMPLIFIER, model DM500, has two stereo phono/line inputs; one microphone input; pro-style slide controls for fading and cueing; a talk switch that attenuates the phono/line



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volume 14 dB for making smooth, professional voiceovers without touching any volume settings, and an easy-access headphone output for monitoring. The built-in, low-noise preamp handles virtually any signal source, including magnetic phono cartridges and low-impedance microphones, and is powerful enough to drive almost any power amplifier. It can also be plugged into the auxiliary input of any receiver or integrated amplifier. The *model DM500* is priced at \$149.00. — **Numark Electronics Corp.,** 503 Raritan Center, PO Box 493, Edison, NJ 08817.

ANTENNA, model PD-1610, is a fiberglass collinear antenna of the "Super-Stationmaster" design. The frequency range is 806-896 MHz and gain is 3.0 dBd. Maximum input power is 500 watts; bandwidth, for 1.5:1 VSWR, is 60 MHz, and vertical beamwidth, ½-power points, is 36 degrees.

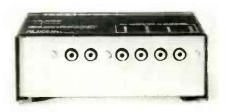


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The overall length of the *model PD-1610* is 44 inches. The fiberglass housing is 1.5 inches in diameter, and the aluminum support pipe is 12 inches long and 1.66 inches in diameter. Copper alloy is used as the radiating element material. The *model PD-1610* attains a resistance to rated wind velocity of 200 miles per hour, with a lateral thrust at rated wind of 48 pounds. Bending moment 1 inch below the top of the support pipe is 47 foot pounds, and lightning protection is direct with an 18-inch, flexible extension of RG-393/U; mounting hardware is supplied.

The *model PD-1610* is priced at \$175.00. — Phelps Dodge Communications Company, Route 79, Marlboro, NJ 07746.

ELECTRONIC CROSSOVER, model 6000, is housed in a compact aluminum chassis with its own power supply, and is fully compatible with both tube- and transistor equipment. A choice of 16 different cross-over frequencies is standard, with custom frequencies available at additional charges. The model 6000 crosses-over from 200 through 10,000 Hz, contains two tweeter-level controls, and features very low distortion (.002%) and noise (-90 dB). The 12 dB/octave slope is suitable for



CIRCLE 117 ON FREE INFORMATION CARD

both horn and dynamic drivers.

The model 6000 can be used either for bi-amping or tri-amping, and features a plug-in frequency module that allows changing frequencies readily. For triamping, a second model 6000 is needed to provide the bandpass filter for the midrange speakers. The model 6000 is priced at \$156.00. — Ace Audio Co., 532 5th Street, East Northport, NY 11731.

GRAPHIC EQUALIZER, model SE-9, is a microprocessor-controlled, octave-band stereo graphic equalizer designed for either room equalization or recording. It features 4-curve memory storage, spectrum-analyzer display, and a built-in pinknoise generator; in addition, it comes supplied with an external electret condenser microphone.



CIRCLE 118 ON FREE INFORMATION CARD

A motorized fader-setting system, operating under CPU direction simplifies the digital-processing requirements; all 16 of the frequency controls are dualslide potentiometers, one section of which can vary the boost or cut within its audio band by ±12 dB. The other section produces a varying DC voltage that is fed to comparator circuits that determine when the motors have positioned the sliders correctly. When the vertical motor has placed one slider properly, the horizontal motor shifts the adjustment mechanism to the next potentiometer, in turn. Using the automatic adjustment procedure minimizes fader-to-fader interaction, and requires only 30 seconds overall. Manually-derived settings, or factory-preselected equalizations, may also be stored in the 4-curve memory and recalled at any time.

The model SE-9 is available in brushed aluminum and is priced at \$700.00 (Also available as model SE-9B in matte black with rack mounting, at the same price.) — **Sansui Electronics Corp.**, 1250 Valley Brook Ave., Lyndhurst, NJ 07071.

SPEAKERS, model 5300, are 5-inch wedges, shaped for mounting in car doors and other surfaces. They feature a



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5-inch woofer, separate 1³/₄-inch tweeter, 3-ounce ceramic magnet, and cloth-roll air-suspension. They have a 15-watt RMS rating, 100–15,000 Hz frequency response, and an impedance of 8 ohms.

The model 5300 speakers are made of

molded ABS plastic and are sold in kits containing two speakers and cables. The suggested retail price for the kits is \$19.95 — **BP Electronics**, 855 Conklin Street, Farmingdale, NY 11735.

SOUND-ENHANCER, the *Imager 801-A*, can fit almost any size car or truck dashboard, and uses very little current. By connecting it to an auto, truck, or van's



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speaker system, the speakers' performance is improved dramatically. The sound appears to be coming from many speakers, surrounding the listener with an extra dimension of depth, clarity, and realism. It comes complete and with full instructions, so that it should not be difficult for the average do-it-yourselfer to install. The *Imager 801-A* is priced at \$149.95.—Omisonix, Ltd., P.O. Box 430, Rt. 17, c/o Carrano Bldg., Northford, CT 06492. R-E

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STEREO AUDIO FOR TV

continued from page 76

of TV sound. In fact, the NTSC system has always been capable of providing "high fidelity" FM sound, with response up to 15 kHz. Small-speaker TV sets, together with TV-station-owner apathy, has resulted in a vicious circle that has kept us from enjoying the kind of sound we get from FM radio and other high-fidelity program sources.

The coming of stereo-TV sound may change all that, if only for the reason that the TV tuners and adaptors needed to supply the two channels of audio will permit us to connect our stereo component-systems directly to the line outputs of those new products. In that way, we will finally be able to bypass the sound systems contained in our TV sets. When will that happen? Best estimates suggest that the FCC's decision will not come before mid-1982, and most likely, it will arrive later than that. that.

I am indebted to my friend William S. Halstead. who provided me with much of the background material used for this description of the stereo-multiplex TV systems being considered by the EIA and the FCC. **R-E**



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COMPUTER CORNER

continued from page 92

procedures for various colleges, and the first magazine to distribute its monthly articles via electronic transmission: *Better Homes and Gardens*. A banking-by-home project is described in an experimental program being tried through CompuServe by the United American Bank, Knoxville, TN.

The program provides checking and savings information, financial advisory services, paying of bills, sending of messages, and bank loan applications. (The only thing it can't do, presumably, is eject dollar bills upon request, as the computerized teller stations do, because they can't be transmitted via electronic signal.)

The Source offers over 1,200 services, including the American stock exchange, etiquette, Black studies, New Mexico news and sports, ham radio, French wines, and unidentified flying objects. (A sample partial listing from the Source is shown in Fig. 1.) A recent news magazine details such added services as Legi-Slate, a listing of Congressional records and a career-placement center that matches job seekers with prospective employers.

Both networks offer computerized flea market and want-ad type services, linking consumers across the nation with the most sophisticated and comprehensive resources imaginable for trading services, selling or buying merchandise, or doing the sundry things that want ads traditionally do.

The list of consumer services goes even further: restaurant guides; travel agent transactions; receiving the latest UPI news breaks as soon as they occur. Of special interest to computer hobbyists is the ability to load and unload complete programs.

Which service is the best? The Source is, at least, the biggest, with a much larger index of services offered and access in over 300 U.S. cities. Purchased last year by the Reader's Digest Corporation and having been introduced several months earlier than Compu-Serve, The Source would seem to have the edge in versatility and growth potential. CompuServe, on the other hand, is not far behind, currently available in over 250 cities, with highly visible market exposure--it's sold through Radio Shack stores. Both companies have passed the 10,000 subscriber mark. Both services have their admirers and detractors. It is probably wise to request literature from each network and peruse the list of data bases offered and to study the specific hookup requirements. That way, you can decide which would be least expensive and R-E easier for you to access.



Winning the DB War," in your December 1981 issue of Radio-Electronics.

If we are to assume that that is, indeed, the technique used by the Russians, then Mr. Leinwoll is giving them more credit than they deserve. The concept of backscatter, and its relationship to propagation, is common knowledge. It would appear that the Russians merely expanded the concept.

The sensing or sampling of the backscatter wave would be somewhat akin to over-the-horizon radar, with computer evaluation as the key to that technique. CHARLES DERAPELIAN, Port Neches, TX

SINGLE BOARD COMPUTER

I found Jorma Hyypia's article on "Learning About Microprocessors" (May 1981) very interesting, even though it left out the assembly-language computer that I'm using and that I think your readers might be interested in learning about, the Paccom 8085AKT.

It uses the 8085 CPU and comes either as a kit or assembled. Memory includes 1K of RAM, 1K or PROM, and 1K of EPROM that contains the system monitor. (Two EPROM sockets are provided, and either 1K- or 2K-IC's can be used.)

This computer can be used either as a trainer or as a controller and is supported by several very good books (that come

VIDEOTEX FOR TV

continued from page 61

However it works out, eventually you will be tied into the videotex network. You'll be able to do your shopping and banking, and plan your day using videotex. Even your work may be affected.

Much of the time that many people spend in the office concerns work that could just as well be done at home. That includes: phone calls, correspondence (assuming the user can type), and preparing reports. Imagine being able to do that without having to commute three hours each day to and from work!

Along other lines, farmers could request the latest weather information, or—perhaps more important—the selling price of their produce in a particular market, or find the market offering the best price. Sales could even be made using videotex.

The ability to communicate with the rest of the world—not only be voice, but by visual information—is something that we not only desire, but will *need*.

Whatever form of videotex we use, it can only stand to benefit us. **R-E**

with it). It is especially good for control purposes because, of 44 lines brought out to the card edge, 36 can be wired to meet the user's specific needs.

Programming is done using a 24-key keypad and 9-digit LED display. The computer can be programmed in either octal or hexadecimal (some of us *still* prefer octal, you know).

The price of the 8085AKT kit is \$249.95 and more information on it can be obtained from: Paccom, 14905 N.E. 40th, Redmond, WA 98052. ALLAN EGAN New York, NY

TEMPERATURE SCALES

I wish to point out two errors in Joseph J. Carr's discussion of temperature scales in the November 1981 **Radio-Electronics.** Firstly, the absolute zero temperature is -273.15° C, relative to the freezing point of water, which is 0°C. The triple point of water, the point where gas, liquid, and solid exist in equilibrium at 1-atmosphere pressure, is defined as 273.16°K and the freezing point of water is found to be 0.01°C lower.

Second, all molecular motion does not cease at 0°K. All molecules in a potential field, such as exists in a crystal of the solid, retain a zero-point vibrational motion due to the uncertainty principle, which prevents both the particle momentum and position from being known exactly at the same time. DAVID G. HENNINGS, Ph.D.,

Omaha, NE

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NEW BOOKS

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HOW TO BUILD YOUR OWN WORKING MICROCOMPUTER, by Charles K. Adams. Tab Books, Inc., Blue Ridge Summit, PA 17214. 308 pp including glossary, appendices, and index; $5\frac{1}{8} \times 8\frac{1}{4}$ inches; softcover; \$9.95. (A hardcover edition is also available at \$16.95.)

Compared to the monstrous computers of a decade or more back, the microcomputer is simple, inexpensive, and easy to use—facts well proven by its popularity and ubiquity. Microcomputers now control almost everything from toys to automobile combustion-efficiency; from games to microwave ovens, and from stop lights to numerically-controlled machines. And the microcomputer boom is just beginning.

This book is for the electronics hobbyist who would like to build and program a simple microcomputer, but doesn't know just how or where to begin. It is an easy-to-read manual, showing the way to get started...the *right* way. Whether one is a beginner or a seasoned electronics buff, building a microcomputer from scratch will give one a taste of the electronics world that can't be experienced any other way. The book shows how a computer works, and provides simple (and more advanced) programming exercises for use with a self-built computer after presenting a step-by-step outline of every aspect of construction, assembly, and testing. Diagrams and tables are presented clearly, as well as parts lists, layout, power-supply sources, keyboard and display circuits, and even plans for an EPROM programmer.

The book may be used in any one of three ways: as a general reference book; as a guide to building a computer, or as a guide to building and programming the specific computer described. CIRCLE 151 ON FREE INFORMATION CARD

ELECTRICAL INTERFERENCE IN ELECTRONIC SYSTEMS: Its Avoidance within High-Voltage Substations and Elsewhere, by R. E. Martin, C. Eng., D.F.H., F.I.E.E., F.I.E.R.E. Consultant. Research Studies Press, PO Box 92, Forest Grove, Oregon 97116. 198 pp including references, bibliography, and appendicies—no index; $5\frac{1}{2} \times 8\frac{1}{2}$ inches; hardcover; \$31.00.

Semiconductors used in modern electronic equipment and in computers are very sensitive to electrical interference, and due to their low thermal inertia, they are easily destroyed by overloads lasting even one millisecond. Many such devices normally operate at microwatt or milliwatt power-levels, so that even if destruction is avoided, false operations can easily be caused by interference. Computers are one of the most difficult devices to protect against interference, because the spectrum of signals they use is extremely wide.

The pulses used by the computer may have durations of a few tens of nanoseconds and rise times of a few nanoseconds; they will thus represent frequencies extending to 100 MHz or more.



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Very short interference-pulses can easily be interpreted by the comptuer as instructions, and that may corrupt a program in storage, or inhibit the running of a program.

This book is intended to be used in conjunction with other documents referenced, to aid immunity from interference by isolation, screening, filtering, and earthing—a scheme designed for use with electronic systems in electrically noisy environments. It is particularly concerned with the use of electronic equipment in high-voltage electricity substations and outlines the ways in which harmful interference may enter electronic systems within an electricity station, as well as indicating appropriate measures for controlling the effect of such interference.

Although there is no index, the table of contents, which lists nearly every page, should suffice to enable the reader to find any particular material easily. There are many clearly printed diagrams, a section on the scope of the book and on how to use it, and a glossary of terms.

CIRCLE 152 ON FREE INFORMATION CARD

MUSICAL APPLICATIONS OF MICROPROCESSORS, by Hal Chamberlin. Hayden Book Company, Inc., 50 Essex Street, Rochelle Park, NJ 07662. 661 pp including appendix, bibliography, and index; 6¼ × 9¼ inches; hardcover; \$24.95.

This book covers digital-microprocessor sound and music synthesis comprehensively, featuring previously unpublished techniques that are practical only with microprocessors. Standard linear techniques for microprocessor applications are discussed and musical applications for the newer and more powerful 16-bit microprocessors are explained in non-mathematical language. All phases of waveform shaping and filtering, as applied by digital devices to music generation, are also covered.

In the first section, "Background," chapters 1 to 5 cover the fundamental material necessary for full appreciation of the sections to follow. The second section, chapters 6 through 11, deals with computer-controlled analog synthesis, and in section three, chapters 12 to 18 cover the principals of digital synthesis and sound modification.

Throughout the discussions, mathematics is kept to a minimum, and even then is limited to elementary algebra and trigo-

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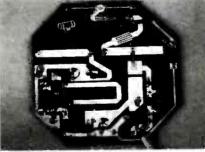
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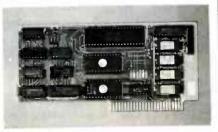
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The Microbuffer II is an intelligent buffered parallel printer interface for the Apple II computer. Includes 16K of on-board RAM. Buffer memory size is user expandable to 32K. Compatible with Applesoft, CP/M, and Pascal. Includes advanced high resolution graphics print routines for the Epson MX-80 with Graphtrax, Anadex, IDS Paper Tiger, NEC Prowriter. Price: \$259. Contact Practical Peripherals, Inc., 31245 La Baya Dr., Westlake Village, Calif. 91362 (213) 706-0339. CIRCLE 41 ON FREE INFORMATION CARD

FEBRUARY 1982



nometry. Important concepts are illustrated through charts, graphs, and, at times, computer programs in BASIC. CIRCLE 153 ON FREE INFORMATION CARD

ELECTRICAL AND ELECTRONICS DRAWING, Fourth Edition, by Charles J. Baer and John Ottaway. Gregg Division, McGraw-Hill Book Company, 1221 Avenue of the Americas, New York, NY 10020. 473 pp including appendices, glossary, and index; 61/2 × 91/2 inches; hardcover; \$16.25.

This new Fourth Edition provides a complete course in electrical and electronics drawing that reflects the latest drafting techniques, technology, and the National Electrical Code.

Presented in a logical, easy-to-master sequence, the text is designed to expand students' knowledge of electronics functions and systems. An up-to-date picture of electrical drafting is provided with basic "how-to" information, plus coverage of electronics, automation, micro-electronics, electric power, and architectural wiring. Included are new chapters on computer-aided design (CAD), microprocessors and industrial controls, and new material on flow and logic diagrams, laying out circuits, using templates and drawing printed circuits.

The book is illustrated extensively and presents questions and problems at the end of each chapter.

CIRCLE 154 ON FREE INFORMATION CARD

airtight. A hot-melt glue gun will make construction easier. The rather elaborate construction of the initial chamber, behind the speaker, serves two purposes. First, it directs the back-wave from the speaker into the labyrinth. Second, it prevents standing waves from building up. Three views of that first chamber are provided in Figs. 4,

5, and 6. It contains two angled partitions. The first behindthe-speaker partition, whose outline is shown in Fig. 7, is attached to mounting strips on the side panels; the second one is attached to the enclosure at all four of its edges.

The only acoustic insulating-material used in the system is in the first chamber, where indicated by asterisks in Figs. 5 and 6. It should be loosely-fluffed 1/4- to 1/2-inch fiberglass.

Two binding posts will be needed to connect the subwoofer to your amplifier. Locate them at the side or rear of the enclosure, and make sure that there is an airtight seal where they pass through the panel.

Prepare the front, rear, and side panels by attaching partition-to-cabinet mounting strips at the positions shown in Fig. 8, on what will become their insides. Cut out the hole for the speaker, bearing in mind that it is offset slightly from the middle of the front panel. Refer to Fig. 9 as you do this.

Prepare partitions 1-11 as shown in Fig. 10 and attach the even-numbered partitions to the odd-numbered ones as shown in Fig. 11. Now, join the sides at rear of the enclosure, leaving the front panel off for the time being. Install the first and second behind-the-speaker partitions, doing the second one first. Don't forget the acoustic installation where indicated. Attach the speaker to the front panel and connect it to the binding posts on the rear panel. Then attach the front panel to the rest of the enclosure and install partitions 1-10 on their mounting strips. Finally, add partition 11, and the top of

The SAFE subwoofer is connected directly to your amplifier's output through a crossover network with a cutoff frequency of about 100 Hz. While an L-C-L "pi" network can be used, an electronic crossover would be better.

With much program material, the benefits of the subwoofer will not be noticeable-music just doesn't contain that much low-frequency information. When the low frequencies are present, however, you'll hear your music as you've never R-E

RADIO-ELECTRONICS

DIGITAL THERMOMETER

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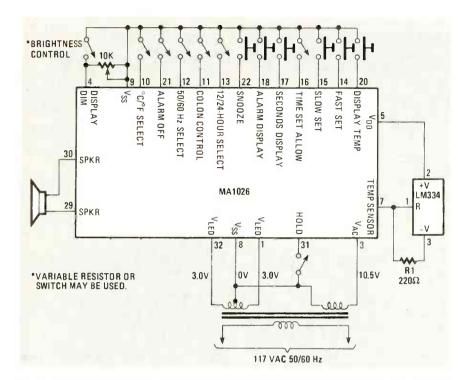


FIG. 5-EXPAND THE BASIC CLOCK/THERMOMETER project and make a full-function alarm clock with snooze alarm, using this circuit.

functions. While the clock/thermometer described in this article does not make use of some of them, there is no reason you can't expand the project to include them. For instance, with the appropriate additional circuitry you can turn the project into a full-feature alarm clock with a snooze alarm such as the one shown in Fig. 5. If you have a lowwattage soldering iron and a good eye. you might want to try accessing the traces going to individual segments of each digit. With a little ingenuity, doing that would let you build a time-of-day clock and temperature monitor for a computer. If you are interested in exploring the potential of the MA1026 further, be sure to order the data sheet and applications notes when you order the module (note that there is sometimes a small additional charge for those items). The module is simple and economical, but it has many features available to those who want to use it fully R-E

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INTENSITY MODULATION		Over	5V p-p	
VERTICAL DEFLECTION				
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Non Distorted Max Amp	More than 4 div	More than 8 div	More than 4 div	More than 8 div
Display Mode	CH 1, CH 2, DUAL, ADD, DIFF		FF	
HORIZONTAL DEFLECTION				
Sweep Mode	A	UTO, NORM	, TV (+), TV (-	-)
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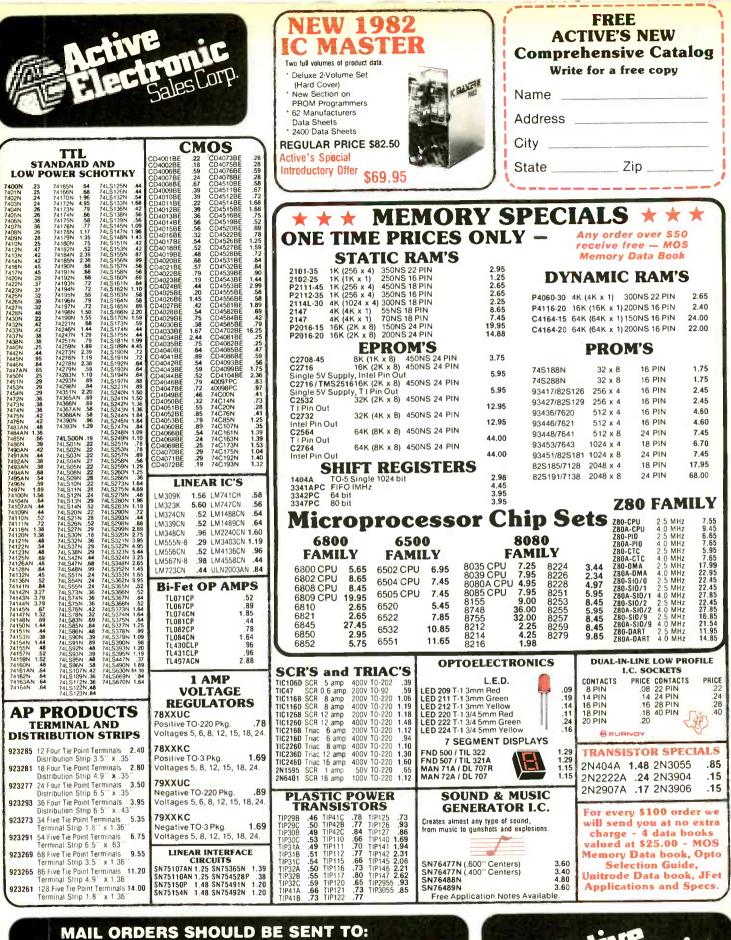
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DIGITS 525 MHz \$99⁹⁵

SPECIFICATIONS:

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7 DIGITS 500 MHz \$79 95 WIRED

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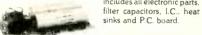
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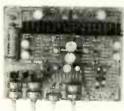
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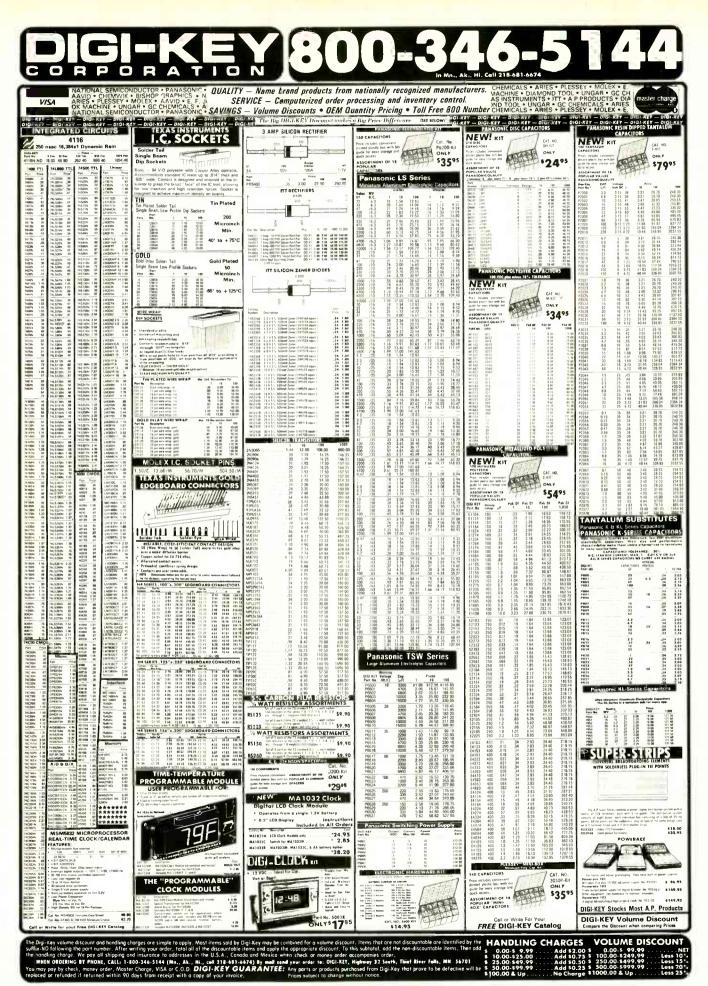
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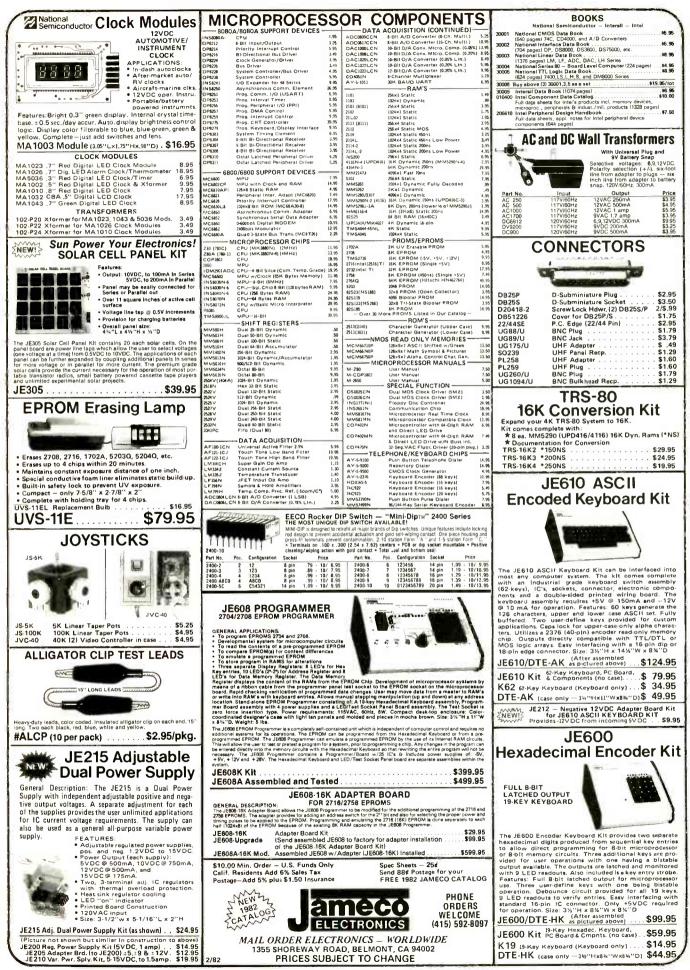




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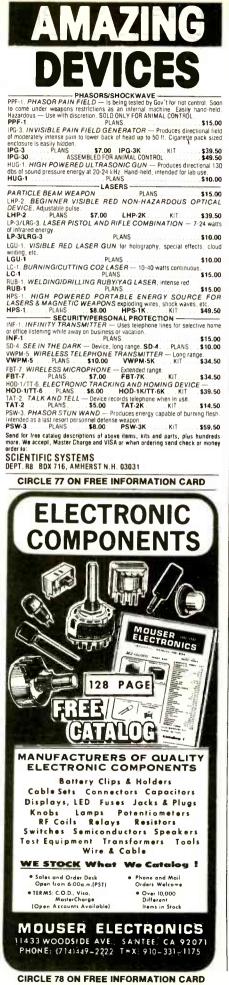
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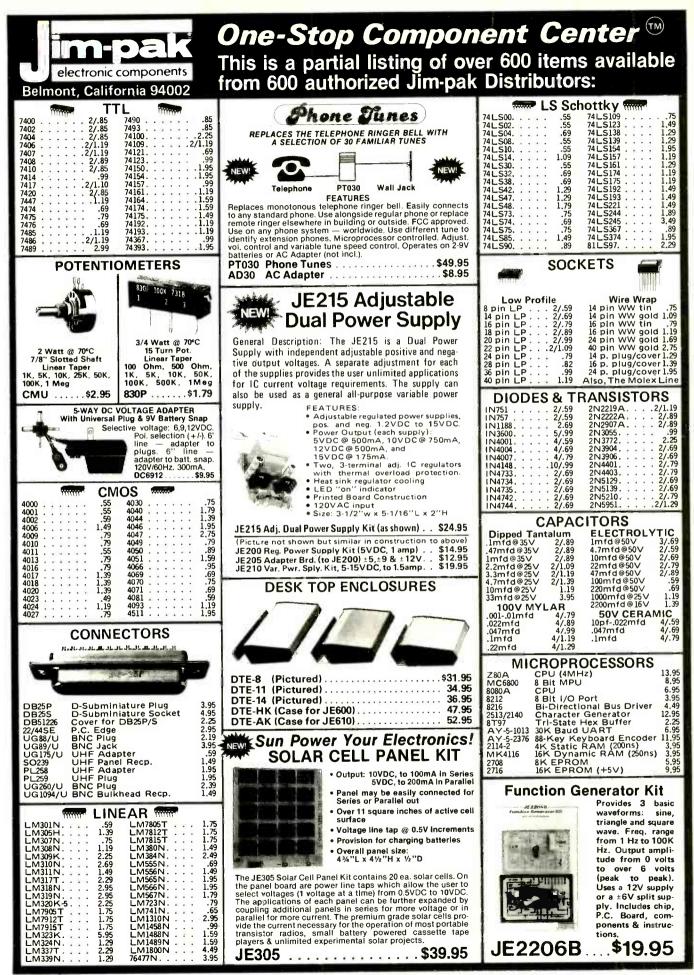
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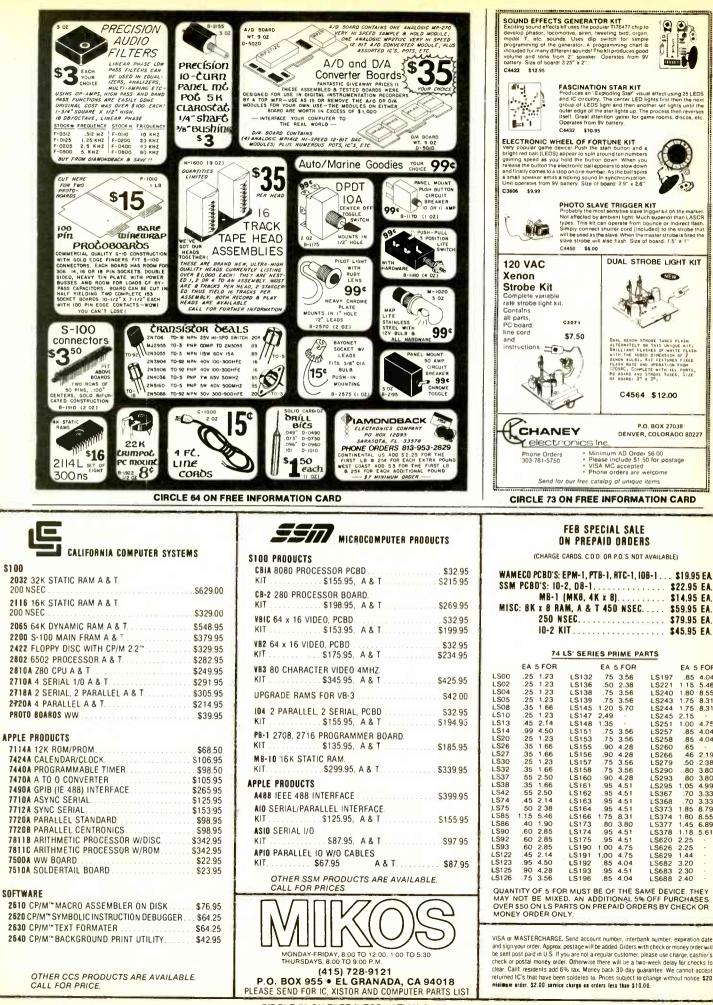






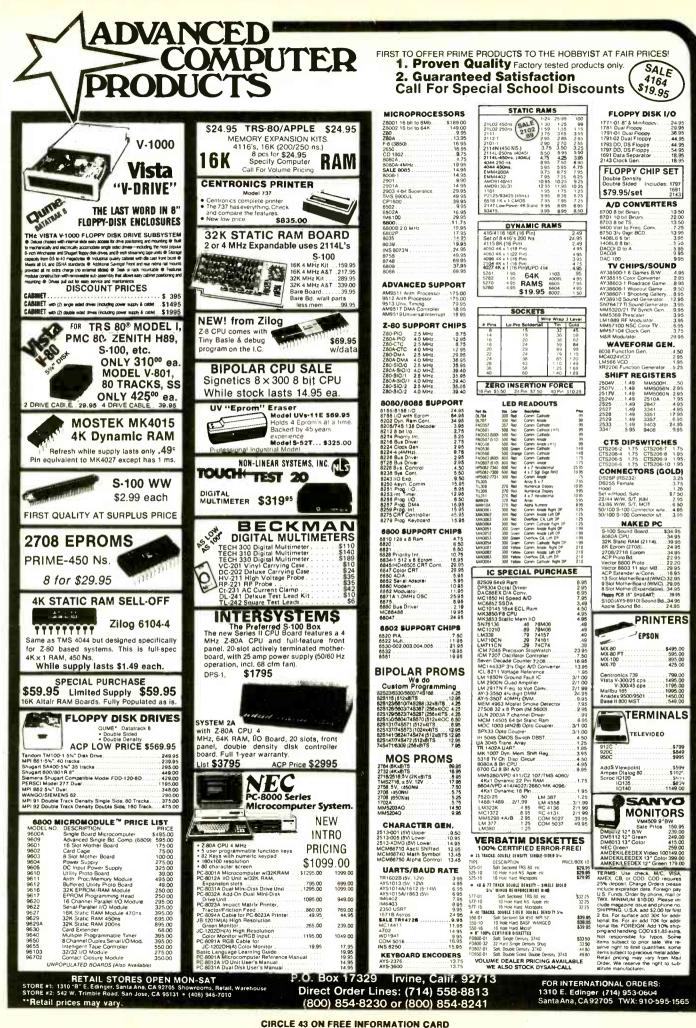
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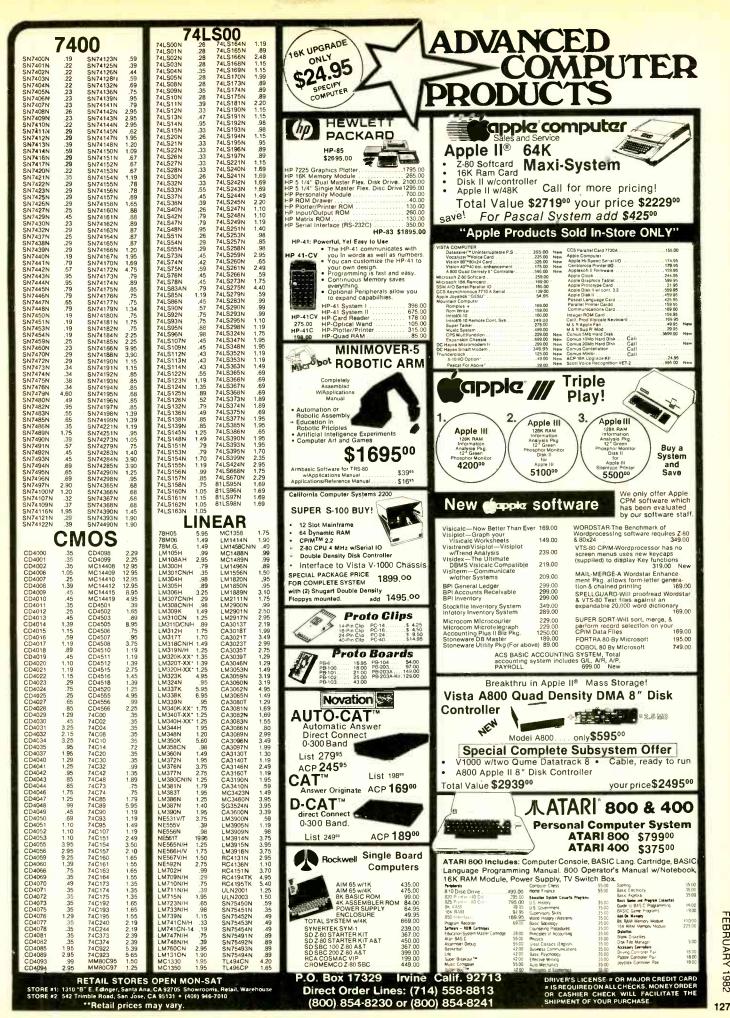


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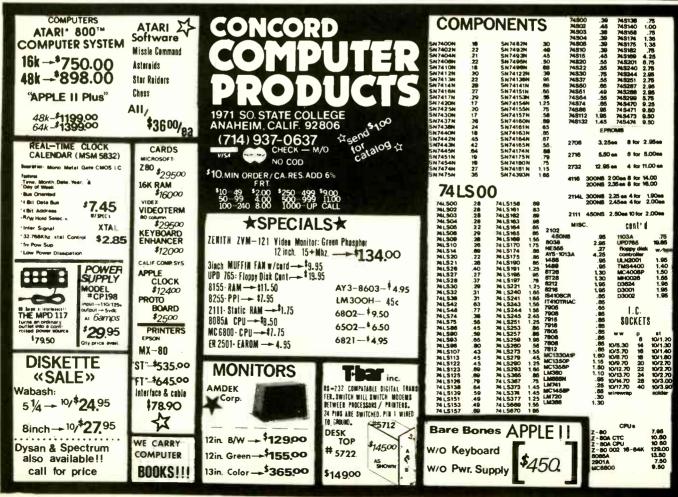
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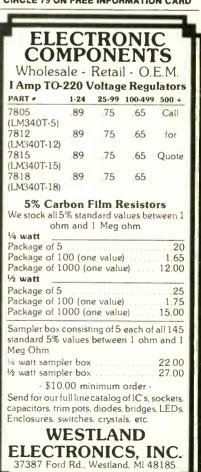
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