

Consisting of a Fane Model 803 ultra-low resonance 8" speaker unit with PVC cone surround, and a Fane Model 303 high frequency pressure tweeter together with a printed circuit cross-over assembly with ferrite cored coils.

HIGH FIDELITY

LOUDSPEAKER KIT

SPECIFICATION OF UNITS
Madel 803
8"

Model, 303 Overall Diameter 30Hz Natural resonance \*35Hz 1200Hz Main resonance Magnet Pole Diameter Magnet Flux Density 10,000 gauss 2kHz-17kHz 8-15 ohms \*15 wates 13,000 gauss 30Hz=8kHz 8-15 ohms Frequency response Impedance Power Handling +15 wates

Very easy to assemble, this boxed kit comes with acoustic foam damping material, panels, screws, wire, wiring diagrams and instruc-tions to provide a remarkable loudspeaker assembly—a pair is ideal for stereo.

CABINETS for the 'Mode One' are available in Teak veneer finish. SIZE  $15\frac{1}{4}$ " x  $10\frac{1}{4}$ " x 9".

\*When used with above cabinet.

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with an easy-to-replace 'dupont' teflon screw-in nozzle.

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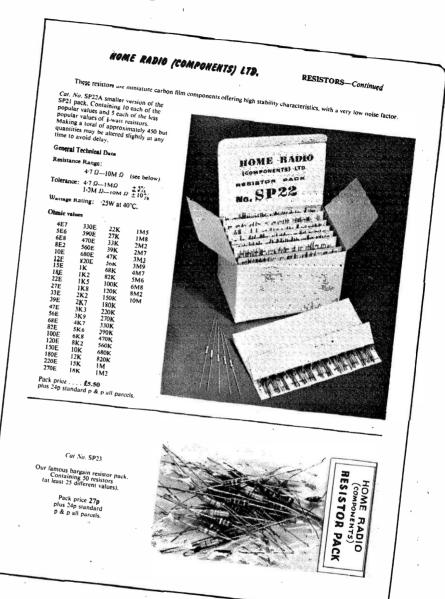
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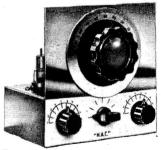
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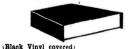
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1			
1	PS 47	Co-Axial Flush	0.14
1	PS 46	Co-Axial Surface	0.0
1	PS 45	Car Aerial	0.0
ı	PS 44	Phono Double	0.1
1	PS 43	Phono Single	0.0
i	PS 42	Jack Stereo Switched	0.20
1	PS 41	Jack 2" Switched	0.1
1	PS 40	Jack 3.5mm Switched	0.10
1	PS 39	Jack 2.5mm Switched	0.0
1	PS 38 DIN		0.16
1	PS 37 DIN	5 Pin 180°	0.10
1	PS 36 DIN	3 Pin	0.10
1	PS 35 DIN	2 Pin (Speaker)	0.06
1	SOCKETS		

Co-Axial Flush	0.14
NE SOCKETS	
D.I.N. 2 Pin (Speaker)	0.13
D.I.N. 3 Pin	0-17
D.I.N. 5 Pin 180°	0.17
D.I.N. 5 Pin 240°	0.17
Jack 2.5mm Plastic	0.10
Jack 3.5mm Plastic	0.18
Jack 1" Plastic	0.24
Jack 1" Screened	0.28
Jack Stereo Plastic	0.22
Jack Stereo Screened	0.32
Phono Screened	0.14
Car Aerial	0.15
Co-Axial	0.17
	D.I.N. 2 Pin (Speaker) D.I.N. 3 Pin D.I.N. 5 Pin 180° D.I.N. 5 Pin 180° D.I.N. 5 Pin 240° Jack 2-5mm Plastic Jack 2-5mm Plastic Jack 4" Plastic Jack 4" Screened Jack 8tore Plastic Jack Stereo Screened Phono Screened Car Aerial

#### DILICE

PS 1	D.I.N. 2 Pin (Speaker)	0.11
PS 2	D.I.N. 3 Pin	0.12
PS 3	D.I.N. 4 Pin	0.15
PS 4	D.I.N. 5 Pin 180°	0.14
PS 5	D.I.N. 5 Pin 240°	0.15
PS 6	D.I.N. 6 Pin	0.15
PS 7	S.I.N. 7 Pin	0.15
PS 8	Jack 2.5mm Screened	0.10
PS 9	Jack 3.5mm Plastic	0.09
PS 10	Jack 3.5mm Screened	0.12
PS 11	Jack 1" Plastic	0.13
PS 12	Jack 1" Screened	0.18
PS 13	Jack Stereo Screened	0.29
PS 14	Phono	0.08
PS 15	Car Aerial	0.15
PS 16	Co-Axial	0.10

C	AΒ	LES	
CP	1	Single Lapped Screen	0.06
$\mathbf{CP}$	2	Twin Common Screen	0.08
CP	3	Stereo Screened	0.08
СP	4	Four Core Common Screen	0.23
CP	5	Four Core Individually Screened	0.30
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CP	9	Speaker Cable	0.04
CP	10	Low Loss Co-Axial	0.10

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	1M, 21 VC 1	Single less Switch	0.14
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Pa	rameter
HARMONIC D	ISTORTION
LOAD IMPED	ANCE
INPUT IMPE	DANCE
FREQUENCY	RESPONSE Œ 3dB
SENSITIVITY	for BATED O/P
DIMENSIONS	

Conditions	Performance
Po - 3 WATTS f -1KHz	0.25°,
~	8 - 16 Ω
f=IKHz	100 k Ω
Po 2 WATTS	50 Hz - 25KH
's=25V. RI=8Ω f 1KHz	75mV. RM8
	3" × 24" × 1"

The above table relates to the AL10, AL20 and AL30 modules. The following table outlines the differences in their working conditions.

Parameter	AL10	AL20	I ALSO
Maximum Supply Voltage	25	30	30
Power output for $2\%$ , $T.H.D.$ (RL = $8\Omega$ f = 1 KHz)	3 watts RMS Min.	5 watts RMS Min.	10 watts RMS Min

## AUDIO AMPLIFIER

AL 10.	3 watts	RMS	£2-19
AL 20.	5 watte	RMS	£2.59
AL 30.	10 watts	RMS	£3.01
-			

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POWER SUPPLIES
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#### TRANSFORMERS

## PA 12. PRE-AMPLIFIER SPECIFICATION

The PA 12 pre-amplifier has been designed to match into most budget stereo systems. It is compatible with the AL 10, AL 20 and AL 30 audio power amplifiers and It can be supplied from their associated power supplies. There are two stereo inputs, one has been designed for use with \*Ceramic cartridges while the auxiliary input will suit most †Magnetic cartridges. Full details are given in the specification table. The four controls are, from left to right: Volume and on/oil switch, balance, bass and treble. Size 152mm × 84mm × 35mm.

Prequency response 20Hz 50KHz (+ 3dB) 2011z 50KHz (- 3dB) Bass control -± 12dB at 60Hz Treble control -± 14dB at 14KHz 1nput 1. Impedance Meg. ohm Sensitivity 309mV 1uput 2. Impedance 30 K ohms Sensitivity 4mV

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## The STEREO 20

The 'Stereo 20' amplifier is mounted, really wired and tested on a one-piece chassis measuring 20 cm × 14 cm × 5·5 cm. This compact unit comes complete with on/off switch volume control, balance, bass and treble controls. Transformer, Power supply and Power amps. Attractively printed front panel and unatching control knobs. The 'Stereo 20' has been designed to fit into most turntable plinths without interfering with the mechanism or alternatively, into a separate cabinet. Output power 20w peak. Input 1 (Cer.) 300mV into 1M. Freq. res. 25Hz-25kHz. Input 2 (Aux.) 4mV into 30K. Hammonic distortion. Bass control ± 1/24B at 60Hz typically 0-25°, at 1 watt. Treble con. ± 1/4dB at 14kHz.



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## THE ALSO

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- \* Load-3, 4, 8 or 16 ohms.
- \* Distortion-better than 1% at
- \* Signal to noise ratio 80dB.

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- \* Supply voltage 10-35 Volts.
- \* Overall size 63mm 105mm × 13mm.

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## STEREO PRE-AMPLIFIER TYPE PA100

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Three switched derec upter, and rumble and scratch filters are features of the PA100 which also has a STEREO/MONO switch, volume, balance and continuously variabless and trethe controls.



SPECIFICATION 

Treble Control
Filters: Rumble (High Pass)
Scratch (Low Pass)
Signal/Noise Ratio Input overload

T00H2
8K Hz
better than - 65dB.
- 26dB
- 35 volts at 20mA
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	3000		159	41 · 25 64 · 54	*
	LO	W V	LTAGE	:	
	Prim	:- 200-	240V. Se	c:- 12 & 24V.	
	Amp	8	Туре	Price	P&P
	12V	24V	No.	£	£
	0.5	$0 \cdot 25$	111	1.00	0.22
	1	0.5	213	1.23	0.22
	2	1	71	1.60	0.22
	4	2	18	2 · 25	0.38
	6	3	70	2.70	0.42
1	8	4	108	8.00	0.52
1	10	5	72	3·55	0.52
	12	6	116	4.50	0.52
4	16	8	17	5.50	0 • 52

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40	20 232	18.30	1.00
60	30 226	28 - 70	1.10
30 V	olts		
Prim:	- 200-240V.	Sec:- 12, 15, 20	24, 30V.
Amps	Туре	Price	P&P
0.5	112	1.20	0.22
1	79	1.84	0.38
2	3	2 · 45	0.38
3	20	3.00	0 · 42
4	21	3.55	0.52
5	51	4 - 40	0.52
6	117	5 - 25	0.52
8	88	6.80	0.67
7.0	00	0.00	0.01

	( 50 Ya	lts			. 60 Vo	its		
)	Prim:-	200-240V.			Prim: - 2	200-240V.		
	Sec:- 18	, 25, 33, 4	0. 50V.			, 30, 40, 4	8, 60V.	
	Amps	Type	Price	P&P	Amps	Туре	Price	P&P
	0.5	102	1.60	0.30	0.5	124	1.60	0.38
	l ","	103			] 1	126	2 - 25	0.38
	,		2.35	0.38	2	127	3 - 55	0.42
	2	104	3 · 25	0.42	1 3	125	5.40	0.52
	3	105	4 · 40	0.52	1 4	123	6-98	0.67
	4	106	5.48	0.52	5	40	8-46	0-67
	6	107	8 · 65	0.67	6	120	8·20	0.82
	-				. 8	121	11 · 60	1.00
	8	118	11 · 27	0.97	1 10	122	15.25	1.00
	10	119	14.15	0.97	12	189	18.43	1.10

	MINIAT	URE AND	EQUIPMENT	•			
ь	Prim 240V	with screen					
	Vol	ts	Milli:	Milliamps		Price	P&P
	Sec. 1	Sec. 2	Sec. 1	Sec. 2	Type No.	£	£
	3-0-3		200		238	1.12	0.10
	0-6	0-6	500	500	234	1.18	0.10
	0~6	0-6	1000	1000	212	1.28	0.22
	9-0-9		100 .		13	0 . 95	0.10
	0-9	0~9	330	330	235	1.28	0.10
	0-8-9	0-8-9	500	500	207	1.70	0.22
	0-8-9	<b>9–8–0</b>	1000	1000	208	2 · 30	0 · 30
	15-0-15	_	40		240	1.28	0.10
	0~15	0-15	200	200	236	1.28	0.10
	20-0-20		30		241	1.09	0.10
	0-20	0-20	150	150	237	1 · 28	0 · 10
	0-15-20	0-15-20	500	500	205	2.16	0.38
-	0-20	<b>0</b> –20	300	300	214	1.36	0.22
	20-12-0-1		700(D/C	3)	221	1 · 21	0.30
	0-15-20	0-15-20	1000	1000	206	3.10	0.38
р.	0-15-27	0-15-27	500	500	203	2 · 38	0.38
•	0-15-27	0-15-27	1000	1000	204	2 · 40	0.38

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H8/2A	3 · 3μF	25V	4p 4p	H7/6	<b>≤</b> 100µF	25V	
H8/3	3μF	50V	4p	H7/7	100µF	25V	5p
H8/3A	4μF	50V	4p	H7/8	125μF		4p
H8/4	4.7μF	25V	4p	H7/8A	100µF	16V 35V	5p
H8/4A	5μF	64 V	4p	H7/9		63V	6p
H8/5	5μF	100	4p	H7/9A	O 100μF	4V	6р
H8/5A	5µF	150V	4p	H7/10		25V	4p
H8/6A	10μF	10V	4p	H7/10A	125μF 160μF	2·5V	6p
H8/7	10μF	70V	4p	H7/11	160µF	25V	3p
H8/8	16µF	35V	4p	H7/11A	150μF	16V	6p
H8/8A	16µF	16V	4p	H7/13A	200µF	25 <b>V</b>	5p
H8/9	20μF	6V	2p	H7/14	<b>Δ.</b> 220μF	50V	8p 10p
H8/9A	20μF	70V	4p	H7/14A	220µF	16V	6p
H8/10	22μF	50V	4p	H7/15	= 220µF	25V	5p
H8/10A	22μF	100V	4p	H7/15A	220	35V	10p
H8/11	25μF	12V	4p	H6/1A	250 v.E	4V	3p
H8/11 A	24μF	275 V	4p	H6/2	250#F	25V	3p
H8/12	32μF	15V	4p	H5/3A	<b>5</b> 320μF	2·5V	3p
H8/12A	30μF	16V	4p	H6/4	√ 320µF	10V	4p
H8/13A	32µF	50V	4p	H6/4A	320.00	16V	5p
H8/14	40μF	25 <b>V</b>	5p	H6/5	330 "E	25V	10p
H8/14A	40μF	16V	4p	H6/5A	A3 330"E	35V	15p
H8/15	47μ <b>F</b>	50V	4p	H6/8	● 470µF	25V	19p
H8/15A	40µF	35 <b>V</b>	4p	H6/8A	<b>170μ</b> 470μ F	35 <b>V</b>	20p
H7/1	50µF	6V	3p	H6/9A	400μF	40V	20p
H7/1A	50μF	10V	4p	H6/13A	1000μF	25V	16p
H7/2A	64 u F	2·5V	2р				
H7/3A	64µF	25V	4p				
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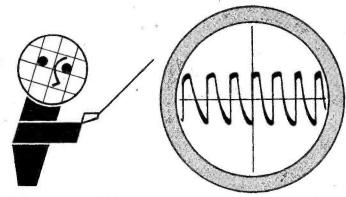
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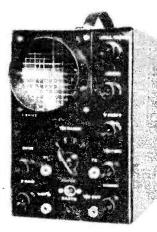
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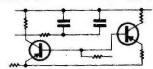
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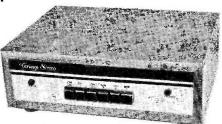
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Size approx. 83" x 61"

Varicap stereo tuner

84" x 24" x 05"

This elegant and practical stereo tuner features push button 'spot-on' tuning with up to five simple pre-set stations (no difficult tuning dial and drive cord). It's easy 'no problem' construction requires only a few simple setting adjustments with a D.C. Voltmeter. It incorporates NEW pre-set modules for R.F. and I.F. circuits, this eliminates the need for the usually difficult circuit alignment.

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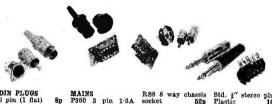
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52n

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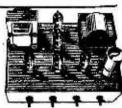
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AC126			30p MJE340	50p	OC44	20p	TIP41C	95p 2N2905	25p
AC122   25p   BCV75   20p   MJE371   36p   OC70   15p   T1550   40p   202907   25p   AC176   25p   BCV71   20p   MJE320   5pp   OC71   17p   ZTX107   35p   Z1896   40p   AC171   25p   BCV72   20p   MJE320   5pp   OC72   20p   ZTX300   15p   20303   30p   AC171   25p   BDV12   25p   BDV12   37p   MJE3055   85p   OC75   25p   ZTX500   15p   20305   40p   AC171   25p   BDV12   37p   MJE3055   85p   OC76   25p   ZTX501   27p   203055   50p   AC171   37p   MPF101   35p   OC77   40p   10659   AC171   40p   AC171			30p MJE370	73 p	OC45	20p	TIP42C		
AC176 25p BCY70 20p MJES20 39p OC71 17p ZTX107 15p ZW2962 10p AC171 25p BCY72 20p MJES251 73p OC72 20p ZTX300 15p ZX303 30p ACY11 25p BCY72 20p MJES251 73p OC72 20p ZTX300 15p ZX303 30p ACY18 25p BD123 75p MJES20551 85p OC75 25p ZTX501 15p ZX303 30p ACY18 25p BD123 75p MJES2055 85p OC76 25p ZTX501 15p ZX303 40p ACY20 25p BD123 75p MM1613 43p OC77 40p 1N659 7p ZX302 64p ACY20 25p BD123 75p MM1612 60p OC81 23p 1N914 80 ZX303 10p ACY20 25p BD131 75p MPF102 45p OC82 25p 1N916 80 ZX303 10p ACY21 25p BD156 75p (ZX5457) 35p OC83 22p 1N4001 80 ZX303 10p ACY21 25p BD156 75p (ZX5457) 35p OC83 22p 1N4002 9p ZX303 10p AD140 55p BD156 75p (ZX5457) 35p OC84 25p 1N4002 9p ZX303 10p AD140 55p BD156 75p (ZX5457) 35p OC64 25p 1N4002 9p ZX303 10p AD162 37p BDY17 £1-50 (ZX5458) 35p OC140 30p 1N4003 9p ZX303 10p AD162 37p BDY17 £1-50 (ZX5458) 35p OC140 30p 1N4004 10p ZX303 10p AD162 37p BDY15 £1-55 MPF105 AD211 £1-50 BF152 20p (ZX5459) 40p OC170 25p 1N4006 15p ZX303 10p AD211 £1-50 BF152 20p (ZX5459) 40p OC170 25p 1N4006 15p ZX303 10p AD211 £1-50 BF152 30p MPSA16 30p OC200 50p 1N4007 15p ZX303 10p AF114 50p BFX84 25p MRSA05 35p OC200 50p 1N4007 18p ZX303 15p AF118 50p BFX85 30p MPSA16 30p OC201 50p 1N4007 18p ZX303 15p AF118 50p BFX85 30p MPSA16 30p OC201 50p 1N4007 18p ZX303 15p ZX303 15p AF118 50p BFX85 30p MPSA16 30p OC201 50p 1N4007 18p ZX303 15p ZX303 15p XX303 15p ZX303 15p XX303 15p XX30	A C127	25p BCY55	60p MJE371	86p	OC70	15p	T1S50	40n 2N2907	
ACY17 ACY18 ACY19 ACY18 ACY19 ACY18 ACY19 ACY18 ACY19 ACY18 ACY19 ACY18 ACY19 ACY19 ACY18 ACY19	AC128	25p BCY70		59n	OC71				
ACY18 25p BD123 75p MM1613 43p OC78 25p ZTX531 27p 2N3055 55p ACY19 25p BD123 75p MM1613 43p OC78 25p ZTX531 27p 2N3055 55p ACY19 25p BD123 75p MM1613 43p OC78 25p ZTX531 80p 2N3702 19p ACY20 25p BD124 75p MM1613 43p OC78 25p 1N916 80 2N3702 19p ACY20 25p BD131 75p MPF102 45p OC82 25p 1N916 80 2N3703 19p ACY21 25p BD156 75p (2N5457) 35p OC84 25p 1N4001 80 2N3703 19p AD149 65p BDV17 £1 50 (2N5457) 35p OC64 25p 1N4002 9p 2N3705 10p AD161 37p BDV17 £1 50 (2N5458) 35p OC140 30p 1N4003 9p 2N3706 10p AD211 £1 50 BF152 20p (2N5459) 40p OC170 25p 1N4006 15p 2N3707 10p AD211 £1 50 BF152 20p (2N5459) 40p OC170 25p 1N4006 15p 2N3707 10p AD211 £1 50 BF195 15p MPSA16 30p OC200 50p 1N4006 15p 2N3709 10p AF114 25p BF195 15p MPSA16 30p OC200 50p 1N4006 15p 2N3709 10p AF118 50p BFX84 25p MRSA56 35p OC200 50p 1N4006 15p 2N3707 22p AF117 25p BFX84 25p MPSA16 30p OC201 60p 1N4047 8p; 2N3712 £2 76 AF116 25p BFX93 30p MPSA16 30p OC201 60p 1N4047 8p; 2N3712 £2 76 AF118 50p BFX85 35p OK201 8p; 2N390 30p AFX86 35p OKX713 35p HR34 4p; 2N390 30p AFX86 35p OKX713 35p HR34 4p; 2N390 30p AFX86 35p OKX713 35p HR34 4p; 2N390 30p AFX86 35p OKX714 30p AFX86 35p AFX86 30p AFX86 3	AC176	25p BCY71				200			
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ASZ221 35p BFY10 35p MKT401 25p TIP33A £1.05 2N708 15p 12N305 25p BA102 30p BFY44 55p MKT404 80p TIP34A £1.55 2N300 20p 12N4058 12p BA112 60p BFY50 25p NKT773 25p TIP35A £2.05 2N1302 25p 2N4059 12p BA114 10p BFY51 20p NKT774 25p TIP36A £3.35 2N1302 25p 2N4059 12p BA166 15p BFY5 22p OA5 20p TIP41A 75p 2N1303 16p 2N4061 12p BC107 10p BFY5 10p OA5 20p TIP41A 75p 2N1303 16p 2N4061 12p BC109 10p BFY9 60p OA47 10p TIP42A 90p 2N1304 20p 2N4061 12p BC109 10p BFY9 60p OA47 10p TIP42A 90p 2N1304 20p 2N4061 12p BC109 10p BFY9 60p OA47 10p TIP23B 5p 2N1305 20p 2N4062 12p BC109 10p BFY9 60p OA47 10p TIP23B 5p 2N1305 20p 2N4062 12p BC109 12p BSW6 60p OA79 12p TIP30B 60p 2N1306 25p 2N4285 15p BC149 12p BSW6 73p OA81 10p TIP32B 20p 2N1306 25p 2N4285 15p BC149 12p BSY95 12p OA90 10p TIP33B £1-60 2N1613 20p 2N4280 15p BC153 13p C428 30p OA200 10p TIP35B £3-64 2N2147 70p 2N4290 15p BC169C 12p BY100 15p OA200 10p TIP35B £3-64 2N2147 70p 2N4871 35p BC169C 12p BY100 15p OA200 10p TIP35B £3-64 2N2147 70p 2N4871 35p BC169C 12p BY100 55p OA200 30p TIP418 83p 2N2160 55p 2N4890 60p			25p MKT222	25p	TIP32A	74 p	2N706A	15p 2N3904	22p
BA102         30p   BFV44         50p   MKT404         80p   TIP34A £1.*55   ZN950         20p   ZN4658         12p	ASZ21	35p BFY10	35p MKT401	25p	TIP33A	£1.05	2N708		
BA112         60p         BFV50         25p         NKT773         25p         T1P35A         £2:05         2718232         25p         218059         12p           BA114         10p         BFV51         20p         NKT774         25p         T1P36A         £3:35         2N1302         16p         2N4061         12p           BC107         10p         BFV5         10p         OA10         20p         T1P42A         30p         20p         2N4061         12p           BC108         10p         BFV9         60p         OA47         10p         T1P24A         90p         20p         2N4062         12p           BC199         10p         BFV9         60p         OA47         12p         T1P308         65p         2N1008         22p         2N4126         17p           BC149         12p         BSW6         60p         OA79         12p         T1P308         67p         2N1008         25p         2N4287         15p           BC149         12p         BSW95         73p         OA81         10p         T1P32B         £1:00         2N1008         25p         2N4287         15p           BC153         13p         C42e         30p		30p BFY44	50p MKT404	80p	TIP34A	£1.55	2N930		
BA114   10p   BFV51   20p   KTT74   25p   T P36A & 3-35   2N1302   16p   2N4060   12p   BA156   15p   BFV5   22p   0A5   20p   T P41A   75p   2N1303   16p   2N4061   12p   BC108   10p   BFV5   10p   0A10   20p   T P42A   30p   2N1304   20p   2N4062   12p   BC108   10p   BFV9   60p   0A47   10p   T P29B   58p   2N1305   20p   2N1456   17p   BC109   10p   BFV9   60p   0A79   12p   T P30B   68p   2N1006   25p   2N4285   15p   BC147   12p   BSW6   75p   0A81   10p   T P31B   70p   2N1307   25p   2N4281   15p   BC149   12p   BSV96   75p   0A81   10p   T P32B & 21p   2N1308   25p   2N4288   15p   BC149   12p   BSV96   32p   0A90   10p   T P33B & 11p   2N1309   25p   2N4289   15p   BC153   13p   C42B   30p   0A200   10p   T P35B & 2-8   2N1306   25p   2N4290   15p   BC159   13p   BV100   15p   0A202   10p   T P35B & 2-8   2N1417   70p   2N4871   35p   BC169C   12p   BY129   55p   0A201   35p   2N4260   60p   30p   3	BA112		25p NKT773	25p	TIP35A	£2.05	2N1132	25p 2N4059	
BA156         15p   BFY5         22p   OA5         20p   TIP41A         75p   ZN1303         16p   ZN40651         12p   CN20651           BC107         10p   BFY5         10p   OA47         10p   TIP42A         90p   ZN1304         20p   ZN4062         12p   ZN40651		10p BFY51	20p NKT774	25p	TIP36A	£3 · 35	2N1302	16 p 2N4060	120
BC107         10p BFV5         10p 60p 60p 60p 60p 60p 60p 60p 60p 60p 6	BA156	15p BFY5	22p OA5	20 p	TIP41 A	75p	2N1303		
BC108         10p BFY9         60p 0 A47         10p 1 P192B         58p 2 N1305         20p 1 N16126         17p 1 N162B         15p 2 N106B         25p 2 N142B         15p	BC107	10p BFY5	10p OA10	20 p	TIP42A		2N1304		
BC109 10p   BLY10 & 3-00   OA70   12p   TIP30B 66p   2N1006   25p   2N4225   15p   BC147   12p   BSW6 60p   OA70   12p   TIP30B 70p   2N1307   25p   2N4225   15p   BC148   12p   BSW6 75p   OA81   10p   TIP32B & 71p   2N1307   25p   2N4288   15p   BC149   12p   BSY95   12p   OA90   10p   TIP32B & 11p   2N1309   25p   2N4288   15p   BC157   13p   C111   40p   OA91   10p   TIP32B & 1-60   2N1613   20p   2N4290   15p   BC158   13p   C428   30p   OA200   10p   TIP34B & 1-60   2N1613   20p   2N4290   15p   BC159   13p   BY100   15p   OA200   10p   TIP35B & 2-81   2N1171   25p   2N444 & 1-90   BC169C   12p   BY122   55p   OA202   35p   TIP428   3-76   2N12147   70p   2N4871   35p   C169C   12p   BY122   55p   CA202   35p   TIP41B   35p   2N2160   55p   2N4920   60p   60p	BC108	10p BFY9	60p OA47	10n	TIP29B				170
BC147         12p BSW6         60p OA79         12p TP318         70p 2N1307         25p 2N4287         25p 2N4287         15p CM28	BC109	10n BLY10		120	TIP30B				
BC148 12p   BSW6 75p   OA81 10p   TIP32B 82p   2N1308 25p   2N4288 15p   BC149 12p   BSW6 12p   OA90 10p   TIP33B 81:12   2N1309 25p   2N4289 15p   BC157 13p   C111 40p   OA91 10p   TIP34B 81:60   2N1613 20p   2N4290 15p   BC159 13p   C428 30p   OA200 10p   TIP34B 81:60   2N1613 20p   2N4290 15p   BC159 13p   BY100 15p   OA202 10p   TIP35B 82:81   2N1711 25p   2N4481 35p   BC169C 12p   BY122 55p   OA420 210 35p   TIP41B 83p   2N2160 55p   2N4920 60p	BC147	12p BSW6		120		70p	2N1307	25n 2N4287	
BC149         12p   BSY95         12p   OA90         10p   TIP33B         £1:12   DN1309         25p   ZN4290         15p   DN159           BC157         13p   C114         40p   OA91         10p   TIP34B         £1:60   DN1613         20p   ZN4290         15p   DN1613           BC159         13p   BY100         15p   OA200         10p   TIP35B         £2:81   ZN1711         25p   ZN4484         £1:90           BC169         12p   BY122         55p   CA4201         35p   TIP41B         33p   ZN2160         55p   ZN4920         60p	BC148					82 n	2N1308		
BC157 13p   C111 40p   OA91 10p   TIP34B £1-60   2N1613 20p   2N4220 15p   BC159 13p   C428 30p   OA20 10p   TIP34B £2-81   2N1171 25p   2N4424 £1-50   BC159 13p   BY100 15p   OA202 10p   TIP36B £3-64   2N2147 70p   2N4821 35p   BC169C 12p   BY122 65p   OA202 10p   TIP36B £3-64   2N2147 70p   2N4821 35p   C429   C42	BC149	12p BSY95							
BC159 13p BY100 15p OA200 10p TIP35B £2.81   2N1711 25p   2N444 £1.90 BC159 13p BY100 15p OA202 10p TIP35B £3.64   2N2147 70p   2N4871 35p BC169C 12p BY122 65p   OA210 35p   TIP41B 83p   2N2160 55p   2N4920 60p	BC157			100	TIP34B				
BC159 13p BY100 15p OA202 10p TIP36B £3 64 2N2147 70p 2N4871 33p BC169C 12p BY122 65p OA210 35p TIP41B 83p 2N2160 55p 2N4920 60p	BC158	13n C426		10p	TIP35B				
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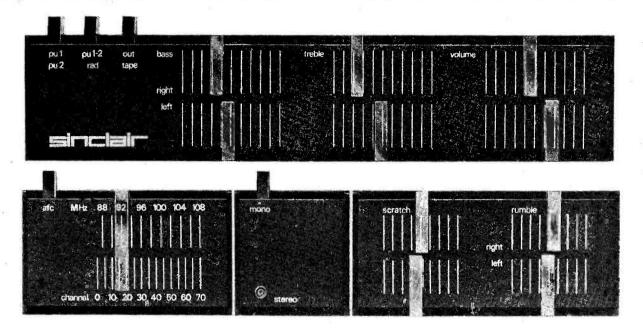
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# Sinclair Project 80 exciting

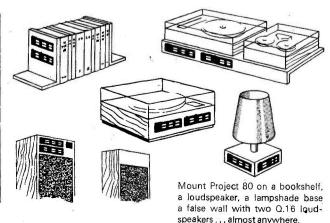


## only 3" deep x 2" high

Living with hi-fi takes on new meaning with Sinclair Project 80. The electronics of these revolutionary new modules are all contained within elegantly designed matching cases no more than three-quarters of an inch deep. They are designed for mounting on any appropriate flat surface by means of 6BA bolts extending from the rear of each module and which pass through suitably drilled holes. Connections are taken away out of sight in a similar manner. The possibilities opened up by Project 80 are endless – superb hi-fi systems can be installed in ways hitherto only dreamed about and never before made practical. No more cutting out and shaping to put modules in position. A few holes drilled with the aid of templates supplied and the job is done. Now you need never again be faced with problems of keeping the hi-fi from clashing with carefully thought-out furnishing schemes. (That will surely please wives!) Slider controls have been introduced in place of knobs and all modules in the range incorporate new up-dated circuitry with emphasis on performance standards and built-in protection against overload and shorting. The aim was to re-think modular construction completely – to make it infinitely more versatile, even simpler and more reliable – the result – Project 80 – another triumph for Sinclair, and the most exciting construction modules ever.

## the slimmest, most elegant hi-fi modules ever made

System	The Units to use	Units cost
Simple battery record player	Z.40	<b>£5·45</b> +54p V.A.T.
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# new thinking in modular hi-fi

Stereo 80 pre-amplifier and control unit For P.U., radio and tape Tape monitoring switch Simplest ever fixing

Each channel has its own separate tone and volume controls operated by sliders, enabling ideal environmental matching to be obtained. A virtual earth input stage forms part of the up-dated circuitry that ensures the finest possible quality from all signal sources. Generous overload margins are allowed on all inputs. Clear instructions with template are supplied.

TECHNICAL SPECIFICATIONS

Size  $-260 \times 50 \times 20$ mm  $(10\frac{1}{4} \times 2 \times \frac{3}{4}$ ins)

Finish - Black with white indicators and transparent sliders

Inputs - Magnetic pick-up 3mV RIAA corrected; Ceramic pick-up 300mV

Radio 300mV; Tape 30mV Signal/noise ratio - 60dl

Frequency range - 20Hz to 15KHz ± 1dB: 10Hz to 25KHz ± 3dB

Power requirements – 20 to 35 volts
Outputs – 100mV + AB monitoring for tape
Controls – Press button for tape radio and P.U. Sliders for volume, bass (+ 12dB to - 14dB at 100Hz) treble (+ 11dB to - 12dB at 10KHz)

R.R.P. £11.95 +£1.19 V.A.T.

Project 80 FM tuner

and stereo decoder





Twin dual varicap tuning: 4 pole ceramic filter: switchable A.F.C.

 On the decoder, solid state stereo indicating beacon.

Making the Project 80 F.M. tuner and decoder available separately gives a wider choice of systems and saves money where stereo reception may not be required. The tuner is a triumph of electronic design and assures excellent performance. The decoder gives a 40dB channel separation with 150mV output per channel. Both units may be used with other than Project 80 systems.

TECHNICAL SPECIFICATIONS OF TUNER

Size  $-85 \times 50 \times 20$ mm  $(3\frac{1}{2} \times 2 \times \frac{3}{4} lns)$ Tuning range -87.6 to 108 MHz Detector -1.C. balanced coincidence for good A.M. rejection

One I.C. equal to 26 transistors

Distortion - 0.2% at 1 KHz for 30% modulation 4 pole ceramic filter in I.F. section

Aerial impedance - 75 Ω or 240-300 Ω

Sensitivity - 4 microvolts for 30dB quieting

Output - 300 mV for 30% modulation

Power requirements - 23 to 33 volts

**DECODER** 

Size  $-47 \times 50 \times 20$ mm ( $1\frac{7}{8} \times 2 \times \frac{3}{8}$ ins)

One 19 transistor I.C.

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Sinclair Radionics Ltd. London Road, St. Ives, Huntingdon PE17 4HJ

Z.40 & Z.60 power amplifiers totally short-circuit proof





Intended for use in Project 80 installations, these modules readily adapt to an even wider range of applications. Both incorporate built-in protection against short circuiting and risk of damage from mis-use is greatly

Z.40 TECHNICAL SPECIFICATIONS

Size – 55 × 80 × 20mm (2½ × 3½ × ¾ins) 9 transistors input sensitivity – 100mV

Output - 15 watts RMS continuous into 8 Ω (35v)

Frequency response - 10Hz - 100KHz ± 1dB Signal/noise ratio - 64dB

Distortion – at 10 watts into 8 Ω less than 0·1% Power requirements – 1·2 to 35 volts

Z.60 TECHNICAL SPECIFICATIONS

Size  $-55 \times 98 \times 15$ mm ( $2\frac{1}{8} \times 3\frac{3}{4} \times \frac{3}{4}$ ins) 12 transistors Input sensitivity -100-250mV

Output - 25 watts RMS continuous into 8 Ω (45V)

Distortion - typically 0.03%

Frequency response – 10Hz to more than 200KHz  $\pm$  1dB

Signal/noise ratio - better than 70dB

Built-in protection against transient overload and short circuiting Load impedance - 4 Ω min; max, safe on open circuit

## Z.40 R.R.P. £5.45 + 0.54 V.A.T.; Z.60 R.R.P. £6.95 + 0.69p V.A.T. Project 80 active filter unit

Makes a highly desirable part of any worthwhile system where inputs may be from record, radio or tape. As with Stereo 80, separate controls applied to each channel make it easier to obtain ideal stereo balance.

TECHNICAL SPECIFICATIONS Size  $-108 \times 50 \times 20$  mm  $(4\frac{1}{4} \times 2 \times \frac{3}{4}$ ins) Voltage gain - minus 0.2dB

Frequency response - 36Hz to 22KHz controls minimum Distortion – at 1 KHz – 0·03% using 30V supply HF cut off (scratch) – 22KHz to 5·5KHz, 12dB/oct. slope L.F. cut off (rumble) – 28dB at 20Hz, 9dB/oct. slope

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 For scratch and rumble control

Transistorised active circuitry

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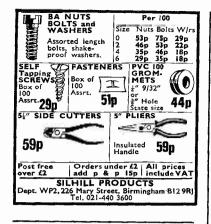
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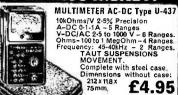
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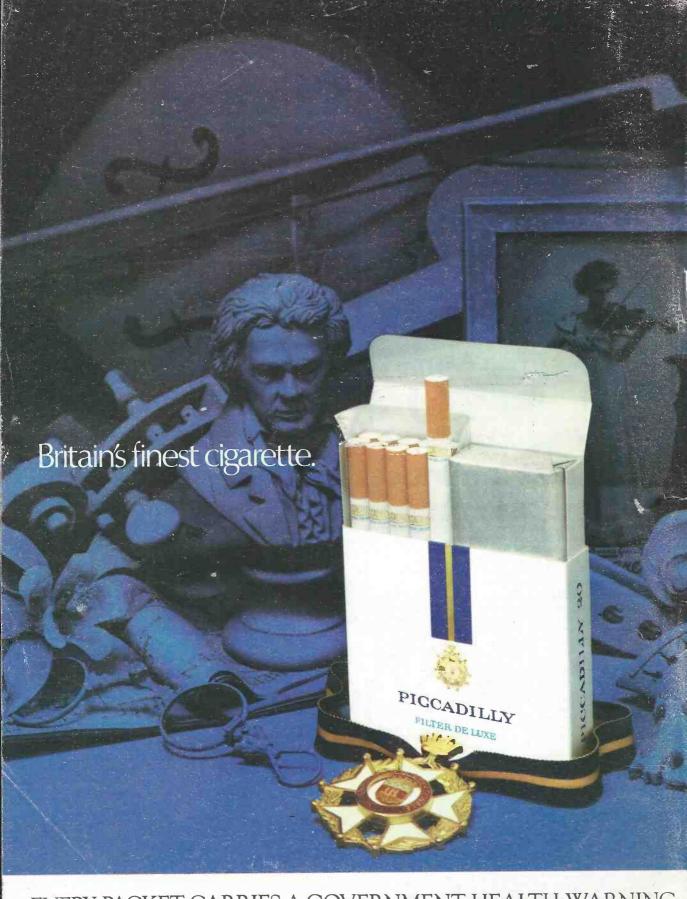
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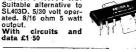
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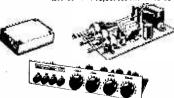
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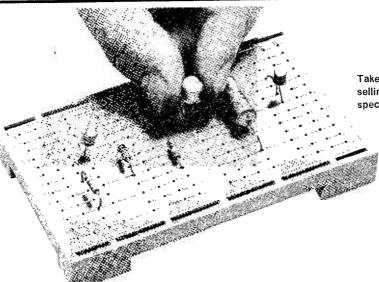
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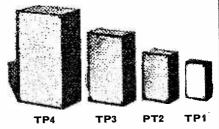
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1N4002

1N4004

1N4005

1N4006

1N4007

16½p 22p

16½p

16 p

20p

22p

38n

RECTIFIERS

1 amp

6.p

90

11p

16 gp

100V

200V

1000V

PILV

50

100

200

400

600

ROO

1000

BY100

BY103

BY105 BY126

BY127

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64p 64p 7
64p 14p 7
64p 23p 13p 7
64p 23p 134p 7
64p 23p 7
64p 25p 2½in × 1in 2½in × 3¾in 2½in × 3¾in 3¾in × 3¾in 3¾in × 5in 17in × 2½in 17in × 3¾in 17in × 5in 19p 41p 57½p 82½p 10p 10p

DIP Breadboard. 4-15in × 6-15in, 110p. Verostrip, 8-5in × 1-5in, 26½p. Spot-Face Cutter, 40p. Pin Injection tool (state 0-1in or 0.15in). 52p. Terminal pins, 20p (packs

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512p 1 DIP10/2 for



SL403D DIP14/5 for EA1000 16p





# electronics

DIODES

TIC44
TIC226D
TIP41A
TIP42A
TIP42A
TIS43
TIS43
TIS73
TIS97
2N/106
2N/106
2N/108
2N/108
2N/103
2N/1304
2N/1305
2N/1306

2N3054 2N3055 2N3391A 2N3566 2N3566 2N3702 2N3703 2N3706 2N3706 2N3708 2N3709 2N3709 2N3711 2N3771

2N3772 2N3773 2N3619 2N3623 2N3903 2N3904 2N3905

### RESISTORS

Rating	Type	Tok	erance	Range	Eac
+ watt	Metal Glaze	±5%	E12 Series	51Ω to 100KΩ	5p
watt	Carbon Film	+ 5%	E24 Series	51Ω to 330KΩ	11p
watt	Metai Oxide Film	± 2%	E24 Series	$10\Omega$ to $1M\Omega$	43p
watt	"Cermet" Thick Film	± 2%	E12 Series	56Ω to 150KΩ	8p
watt	Carbon Film	±5%	E24 Series	10Ω to 10MΩ	1 p
watt	Carbon Composition			·3\O; 3.9\O; 4.7\O	41 p
watt	Carbon Composition	±0.5Ω	5·6Ω: 6·8Ω: 8		5 D
1 watt	Carbon Film	± 10%	E12 Series	10Ω to 10MΩ	51p 3p
	Carbon Film	± 5%	E12 Series	10Ω to 10MΩ	6p
2 watt			E12 Series	0.22Ω to 0.47Ω	10p
2∳ watt	Wire Wound	± 5%			
2+ watt	Wire Wound	± 10%	E12 Series	1Ω to 270Ω	10p
5 watt	Wire Wound	±5%	E12 Series	0-5Ω to 8-2KΩ	10p
10 watt	Wire Wound			0.5Ω to 6.8KΩ	11p
10 watt	Wire Wound	±5%	1.0KΩ: 15KΩ:	20ΚΩ: 25ΚΩ	17p
15 watt	Wire Wound		ţ-, -, -, -, -, -, -, -, -, -, -, -, -, -	10Ω to 6.8KΩ	13 <u>∤</u> p

and

BF177 BF178 BF179 BF180 BF181 BF184 BF185 BF194 BF195 BF196 BF197 BF200 BF220 BF2448

BF262 BF263 BF272 BF392 BFS98 BFW10 BFW11 BFW87 BFX29 BFX30 BFX88 BFY51 BR100 BRY39

115 pp 241 pp 128 pp 174 pp 128 pp 244 pp 128 pp 248 pp 248 pp 248 pp 25 pp 16 pp 16

28pp 2330 pp pp pp pp pp pp pp 25p2 pp 15pp pp pp pp pp pp 25p2 pp 25p BSX20 BSX21 BSY9SA BU105 BU105/02 D13V MJ4090 MJ4091 MJ62955 MPE103 MPF103 MPF104 OC28 OC28 OC26 OC26 OC72 OC72 OC72 OC72 OC72 OC72 OC73 OC74 OC75 OC61 OC76

TRANSISTORS

BC268 BC328 BC328 BC772 BC210 BC211 BD115 BD131 BD132 BD133 BD133 BD133 BD136 BD137 BD138 BD140 BD201 BD202 BD203 BD204 BD203 BD204 BD205 BD205

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5p; 0·2μF, 63p.

MULLARD C280 (250V POLYESTER)

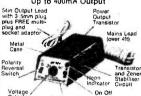
-0·01μF. 0·015μF, 0·022μF,
0·033μF, 0·047μF, 3½p; 0·068μF,
5½p; 0·033μF, 7p; 0·47μF, 9p; 0·68μF,
12p; 1·0μF, 14½p; 1·5μF, 22p;
2·2μF, 26½p.

# 2N4871 2N5129 BA114 BA115 BA116 BA128 BA129 BA143U BA143V BA144 BA145 BA146 BA154 2N5129 2N5172 2N5459 2N5777 2N6076 3N140 3N141 3N152 3N153 40360 OA10 OA47 OA79 OA81 OA85 OA90 OA91 OA200 OA200 OA211 IN914 IN916 IN4148 36 p 47 p 14 p 1 - 01 89 p 101 p 50 p 50 p 55 p £1 - 39 £2 - 17 55 p 10 p 40361 40362 40408 40430 40432 40673 AA119 AA129

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AC1026 AC1026 AC1026 AC1026 AC1027 AC AF239 AS729 AS729 BC107 BC108 BC114 BC1147 BC148 BC148 BC157 BC168 BC168 BC177 BC178 BC177 BC178 BC184 BC185 BC184 BC184 BC185 BC186 BC186

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	ZTX313	11p	ZT X504	43p	ZS142	33p	Zen	
9p	ZTX314	13p	ZTX510	17P	ZS170	10p	Seri	es'
8p	ZTX320	31p	ZTX530	21p	ZS171	13p	as	
11p	ZTX330	15p	ZTX531	22p	ZS172	16p	BZY	88
14p	ZTX331	16p	ZTX550	17p	ZS174	17p	Seri	es
	ZTX341	22p	ZTX551	17p	ZS176	23p	11	
14p	ZT X382	13p	-17001		ZS178	38p	ead	
15p	ZTX383		DIOD	E C	ZS270	11p		
12p		15p			ZS271	16p		
13p	ZTX384	18p	ZS120	8p	2021	.op		

		E	LECT	ROLY	TICS			
ıF	4V	6 3 V	10V	16V	25V	40V	63V	100V
1							6 ½ p	8p
1 5							6 p	
1 5 2 2 3 3 4 7							6 p	
3 3			1 .				6 p	_
4 7			<i>i</i>				6½p	8p
6.8	,		-			6½p	6 ½ p	
10			J		6½p		6-jp	9p.
10 15 22 33				6½p		6½p		
22			6½p		6½p		6 ½ p	11p
33		6½p		6½p		6 <u>1</u> p		
47	6-±p		6-2p		6 p	<b>6</b> ½p	8p	
68		6-2p	-	6 <u>+</u> p			10p	
100	6½p		6 ½ p		6 ½ p	8p	11p	23p
150	1	6-jp		6½p	6 p	10p	13p	
220	6½p		6 <u>‡</u> p	8р	10p	11p	19p	31p
330	6 ½ p	•	8p	_			22p	39p
470	1	6-1p	10p		13p	19p	26p	48p
680		6½p 10p		13p	20p	23p		59p
1000	10p		110	17p	22p	25p	44p	79p
1500	1.	12p	20p	23p				
2200		18p	20p		36p	44p	79p	149p
3300		20p	-,-					
4700	29p			44p	68p	79p	149 p	252p
6400				75p	93p			
10000	1						299p	333p

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Log or	lin less swi	tch (and 1kΩ lin) itch	13p 26p 44p

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E298ED A262	11p	VA1040	11p
E298ED A265	110	VA1053	11p
F298ED P268	110	VA10555	11p
E298ZZ 05	13p	VA1056S	11p
F298ZZ 06	110	VA1066S	11p
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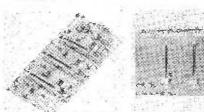
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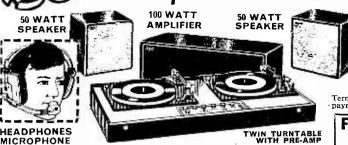
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For Lead Guitar, Mic., Gram, Radio, Tape (Not for use with Bass instruments), Inc. 3 inputs and 2 vol. controls plus Treble, Bass and Presence. Output Jack for additional 15 ohm Speaker. Attractively finished in black with silver-finished fascia and trimmings. Compact size. Fitted carrying handle Terms: Deposit 28 95 and 12 monthly payments 24-97 (Total 268-59), Carr. 21-50 ONLY £59-95 SAE for leaflet

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All power ratings are R.M.S. continuous. 2 YEARS' GUAR ANTEE High flux ceramic magnets. ALL CARRIAGE FREE. 'POP' 100 | 'POP' 50 | 'POP' 50

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Two extra heavy duty 12in. 50w

★ Two extra heavy duty 12in. 50w Loudspeakers.
★ Four Jack inputs and two Volume Controls for instant use of up to four pick-ups or "mikes". Bass and Treble controls. Sea S.A.B. for leaflet. Credit Terms: Deposit £15-50 £72-50 £72-50



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A 4 input, 2 vol. control Hi-Fi 30 watt unit with Separate Bass Bass Current output ratio and Treble controls, s. Peak output

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**FAL PHASE 100/2 AMPLIFIER** 

**AMPLIFIERS** 



FAL TWIN CHANNEL

Solid State, 4 sep. controlled inputs plus master volume control. Individual Bass and Treble controls. Output for speakers 3-30 ohms. Protective circuitry against damage

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300-0-300v. 100mA, 6-3v. 4a, 0-6-6-8v. 3a, 22-45
300-0-300v. 130mA, 6-3v. 4a, 0-5-6-3v. 3a, 22-45
300-0-300v. 130mA, 6-3v. 4a, 0-5-6-3v. 3a, 22-45
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350-0-

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250-0-250v. 100mA, 6.3v. 2a., 0-5-6 3v. 2a. 21-49
250-0-250v. 100mA, 6.3v. 2a., 6.3v. 1a. 21-71
250-0-250v. 80mA, 6.3v. 2a., 6.3v. 1a. 21-75
250-0-360v. 80mA, 6.3v. 4a., 0-5-6.3v. 2a. 21-32
250-0-250v. 100mA, 6.3v. 4a., 0-5-6.3v. 2a. 21-32
250-0-250v. 100mA, 6.3v. 4a., 0-5-6.3v. 3a. 22-45
350-0-360v. 130mA, 6.3v. 4a., 0-5-6.3v. 3a. 22-85
250-0-360v. 100mA, 6.3v. 4a., 0-5-6.3v. 3a. 22-85
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OUTPUT TRANSFORMERS Push-Pull 8 watts EL84 to 3  $\Omega$  or 15  $\Omega$ . Push-Pull 10 watts 6V6, ECL86 to 3-5-8-

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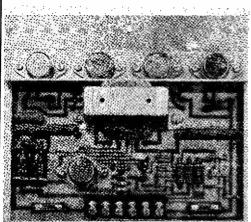
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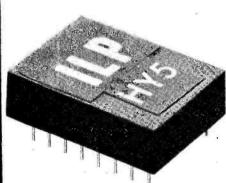
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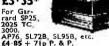
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3 poles	44p	44p	44p	44 <sub>0</sub>	770	77p	770	£1.04	£1.04
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5 poles	44p	44p	77p	770	£1.04	£1.04	£1.04	£1.60	£1-80
6 poles	44p	770	77p	77p	£1.04		£1.04		£1.87
7 poles	77p	770	77p	£1.04	£1.32	£1.82	£1 32	£2·15	£2·15
8 poles	770	77p	77p	£1.04	£1-32	£1.82	£1.32	£2.42	£2.42
9 poles	77p	77p	£1.04	£1.04	£1.60	£1.60	£1.60	£2·70	£2.70
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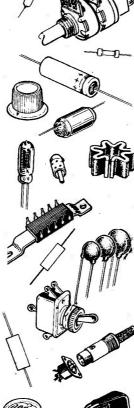
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10				_	8p	9p	8p	8p.
22		_	8р	<u> </u>	9p	8p	8p	10p
47	8р		9p	8p	9p	8p	10p	13p
100	9p	8р	8p	8p	9p	10p	12p	190
220	8p	8p	9p	10p	10p	Ϊįρ	i7p	28p
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1,000	Hp	I3p	13p	17p	20p	25p	41p	
2,200	15p	18p	23 p	26p	37p	4ip		_
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Code .	1-9	10-9	9 1	00 up*
000000	9   	(see note   8   0   9   1   2   3   7   7   9	below) 7 - 5 9 0 - 7 9 0 - 7 9 0 - 9 1 - 6 6 6	5 nett 5 nett 5 nett nett nett
Codes: (	carbon fi	lm, high stabi	lity low noise	

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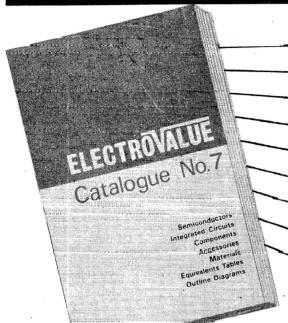
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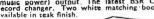
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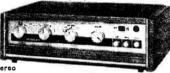




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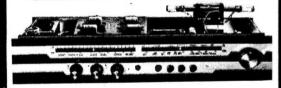
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13AGT   65   6BR   85   6R76   60   20L1   88   5702   30   BF80   3.9   18BGT   55   6BS7   410   6R7M   75   20P1   55   5763   2.00   BF83   4.9   18BGT   55   6856   4.00   EBF83   4.9   68   68   68   68   68   68   68   6		T .50		. 95	6Q7(M)	.55	20F2	.75	807	.59	EBC81	
1B36T   5.5   6B87   1.40   6R7M   75   20P1   55   5763   2.00   EBF83   4.114   1.4   6BW7   66   68C6TGT   33   20P4   88   AC/PEN   98   EBL21   1.5     1R56T   66   6EZ6   49   68ET   44   20P5   1.30   AC/PEN   98   EBL21   1.5     1R5   38   6CB64   40   68ET   44   20P5   1.30   AC/PEN   98   EBL21   1.5     1R5   38   6CB64   40   68ET   44   20P5   1.30   AC/PEN   98   EBL21   1.5     1R5   38   6CB64   40   68ET   42   20A6G   38   AC/PEN   98   EBL21   1.5     1R5   38   6CB64   40   68ET   42   20A6G   38   AC/PEN   20E   20E   42     1R5   38   6CB64   40   68ET   42   20A6G   38   AC/PEN   20E		T .65	6BR8	.85		.60		.88	5702			. 3
1H4GT   65   68W6   .85   68A7   .44   20P3   .50   .7475   .70   .8F89   .3     1L4   1.14   68W7   .66   68C6TGT   .33   20P4   .88   .AC/PEN   .98   .BEL21   .5     1N5GT   .65   68Z6   .49   68BT   .44   20P5   .70   .AC2/PEN   .80   .81     1R5   .38   6CB6   .40   68E7GT   .44   20A6   .38   .AC2/PEN   .80   .81     1U4   .48   .62   .12   .25   .6veg   .72   .2veg   .70   .70   .70     1U5   .90   .6CD6   .80   6VeGT   .38   .25Z4G   .33   .AL60   .10   .00   .6CC3   .9     2D21   .49   .6CG8   .75   .75   .74   .30   .825   .80   .70   .70   .00   .6CC3   .9     20K5   .55   .6CH6   .60   .6XeGT   .28   .25Z4G   .33   .AL60   .10   .00   .6CC3   .9     20K5   .55   .6CH6   .60   .6XeGT   .28   .25Z4G   .33   .AL60   .00   .6CC3   .9     20K5   .55   .6CW6   .75   .76   .75   .7		T .52	6BS7	1.40	6R7M	.75	20P1	.55	5763			. 4
114						.44	20P3	.80				. 3
INFGCT		.14	6BW7	.66	68C7GT	. 33	20P4	.88	AC/PE	N .98	EBL21	1.5
184		T ,65	6BZ6	.49	68H7	.44	20P5	1.30	AC2/P	EN/	EC86	
184					6SK7GT	.44	25A6G					. 5
174				. 28	6SQ7GT	.38	25L6G		AC6PE	N.38	EC92	. 4
105				1.25	6 V 6 G				AC/TH	1 £1	ECC33	1.6
2D21					6V6GT	.38	25Z4G		AL60	1.00	ECC35	9
SA4						.30	25Z5	-80	ATP4			.7
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SCG8   55   6EW6   75   9D7   65   30FL1   75   DAF91   30   BCC189   5			BDTSA		724	.80	30F5		CY31			
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6A8G         1.25         6F24         .85         10P14         2.06         0									DK40			
6A8G         1.25         6F24         .85         10P14         2.00         1.00         BMP6         55         ECH21         1.6         6AG5         6AG5         .27         6F28         .70         12AG6         68         30P19/         30         DL92         .33         ECH55         .66         6AG5         .28 D6         .68         30P19/         .75         DL96         .44         ECH45         .66         6AG1         .75         12AE6         .68         30P11         .88         DM71         .44         ECH81         .30         .64         .68         12AT6         .85         30PL1         .88         DM71         .44         ECH81         .30         .64         .48         30PL1         .88         DM71         .69         .60         EGC1         .60         .60         .60         .75         12AG6         .48         .30PL1         .88         DM71         .20         ECH84         .48         .6AK5         .36         .30PL1         .80         DY802         .33         ECL84         .48         .6AK5         .40         .30PL1         .88         .80P1         .20         .6CL82         .34         .6AK8         .49         .615GT         .32				. 85	10P13	. 70				.70		2.10
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0AR6         .04         .04         .05         12AT6         .35         30PL13         .75         DY80f         .30         ECRH4         .46         30PL14         .90         DY80f         .30         ECRE4         .46         30PL15         .95         E80CC         1.66         ECLE2         .38         6AN8         .49         .35DT         .5         E80CC         1.66         ECLE3         .5         .6AQ5         .40         .36         .28A         .38         .35A         .6         E80C         1.20         ECLE3         .5         .6AQ5         .40         .36         .28A         .7         .28E         .2         .28E         .38         .35D5         .75         .83F         .12         .20         .20EL8         .6         .80F         .20         .2CL23         .2         .2         .2ERE6         .38         .35D5         .75         .83F         .2         .2ERE6         .38         .35D5         .75         .83F         .2         .2ERE6         .38         .35EV         .38         .2EVE5         .6         .38         .3         .2EVE5         .38         .2EVE5         .38         .3         .2EXE7         .38         .3         .2EXE7         <					12AE6	. 65	30PL1	- 66	DM71	2.00	ECH83	. 44
6AM8A									DY87/6	.30	ECH84	.44
6ANS         49         6J5GT         32         12BA6         30         38A3         85         E80P         1.20         ECCT28         6AG5         40         6AG6         40         6AG6         83         35D         75         B3B7         1.20         ECCL28         6AG6         6AG7         6AG7         30         12BH7         27         35LGGT         55         E88CC         80         ECL85         66         6AG7         6AG7         12K6         95         35Z4GT         42         E180F         1.00         ECL85         6AG6         6AU6         30         6K7G         13         12K7GT         35         35Z5GT1         00         E182C1         100         EF22         1.5           6AV6         .30         6K7G         .19         12K7GT         .35         35Z5GT1         .00         E182C1         .00         EF41         .7           6AV6         .30         6K7G         .19         12K7GT         .35         .35C5GT1         .00         E182C1         .00         EF41         .7           6AV6         .30         6K7G         .19         12K7GT         .35         .00C6         .45         EA49         .35         .7 <td></td> <td></td> <td></td> <td>. 75</td> <td></td> <td></td> <td></td> <td></td> <td>DY802</td> <td>. 88</td> <td>ECL80</td> <td>.40</td>				. 75					DY802	. 88	ECL80	.40
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6AS7         1-00         647(M)         38         12J7GT         55         35W4         83         E99CC         1.00         ECL86         40           6AU6         .30         6K7G         .19         12K7GT         .38         3524GT         42         E180F         1.00         EF22         1.06           6AV6         .32         6K8G         .95         12K7GT         45         50B5         .85         81149         .53         EF41         .76           6AW8A         .65         6L1         2.00         12SCT         .50         60C9         .45         6A0         .27         EF42         .56           6AX4         .55         6L6GT         .58         19SGT         .38         50CD6C         1.25         EA76         1.00         EF33         1.55           6B8G         .30         6L7         2.00         12ST7         .35         50EH5         .55         EBBC90.30         EF80         .30         EF80         .30         EF80         .30         EF80         .30         EF80         .30         .50         .60         .60         .60         .75         .60         .60         .60         .60         .60					12BE6	.38		.75		1.20	ECL84	. 60
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6AU6 .33 6JU8A .75 12K5 .95 3574GT .42 £180F 1.00 EF22 1.56 6AU6 .30 6KTG .19 12K7GT .38 3525GT1.00 E182GC1.00 EF640 .66 6AV6 .32 8K8C .35 12Q7GT .45 50BE .45 £1148 .53 £F41 .77 6AW8A .65 6L1 2 .00 12SCT .50 50CE .45 £A50 .27 £F42 .56 6AW8A .55 6L1 2 .20 12SCT .38 50CD6G 1.25 £A76 1.00 £F73 1.55 £B8G .30 6L7 2 .00 12SCT .38 50CD6G 1.25 £A76 1.00 £F73 1.55 £B6G .30 6L7 2 .00 12SCT .44 50L6GT .65 £A8C9 .30 £F89 .36 £B8G .30 £F89 .37 £B8G .30 £F89 .38 £B8G .30		1.00		.38				. 83	E92CC	1.00	ECL86	.40
6AVG         .30         6KTG         .19         12KTGT         .38         32ZGGTI         .00         E182CCI         .00         E74         .6         .85         E114B         .58         E14B         .53         E74         .77         .6         AWS         .85         6114B         .53         E74         .77         .6         AWS         .65         .65         .00         .6         .45         .6         .6         .72         .72         .72         .72         .72         .72         .72         .72         .72         .72         .72         .72         .73				.75	12K5	.95	35Z4GT	.42	E180F	1.00	EF22 1	
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6AWA .65 6LT 2.00 128C7 .50 30C5 .65 EA50 .27 EF42 .56 6AX4 .55 6L5GT .58 128G7 .88 50CD6C1.25 EA76 1.00 EF73 1.55 6B8G .30 6L7 2.00 128H7 .35 50EH5 .55 EABC80.30 EF80 .28 6L18 .49 128L7 .44 50L6GT .65 EACG1 .75 EF33 .66				. 35	12Q7GT			- 85	E1148	. 53	EF41	.70
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6BA6 28 6L18 49 12SJ7 44 50L6GT .65 EAC91 .75 EF93 .66						. 38	50CD6G	1.25	EA76	1.00	EF73 1	. 50
6BA6 .28 6L18 .49 12SJ7 .44 50L6GT .65 EAC91 .75 EF93 .60						. 35	50EH5	. 55	EABC8	0.30	EF80	. 26
DBCS -60 6L19 2.00 12SQ7 -65 85A2 .60 EAF42 .50 EF85 .34				.49	128J7	. 44	50L6GT	.65	EAC91	.75		
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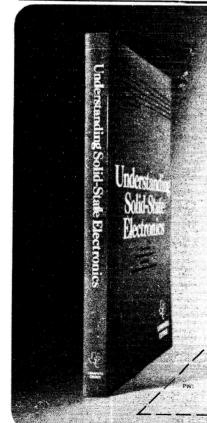
	EF92 .30	KT2 .50	DI 84 99	00104 .38	AC154	-28	BC118	-25	OA91	.10
	EF98 .80		PL84 .33	UM80 .44	AC156	.22	ECV10	-50	OA95	-18
	EF183 .29	KT41 .98 KT44 .75	PL504/500	UY41 .49	AC157	.28	BCY12	-55	OA200	-10
•	EF184 .32	KT66 2.40	-67	UY85 .38	AC165	.28	BCY33	.22	OA202	.11
3	EH90 .45		PL508 .90	U10 1.00	AC166	-28	BCY34	25	OC22	.42
			PL509 1.15	U12/14 1.00	AC168	.42	BCY38	25	OC23	42
==	EL32 .50 EL34 .54	KTW61 £1	PL802 .95	U19 1.73	AC176	·61	BCY39	-28	OC24	42
. 55		KTW62 £1	PY33/2 .50	U25 .65	AC177	.31	BCZ11	-42	OC25 -	
. 25		MHLD6 £1	PY81 .31	U26 .60	ACY17	.28	BF158	-32	OC35	-35
. 15		P61 1.00	PY82 .25	U191 .70	ACY18	.22	BF159	.28	OC36	47
. 55		PABCeo .88 PC86 .60	PY83 .33	U251 .80	ACY19	21	BF163	-22	0C44	·īí
. 39			PY88 .33	U301 -58	ACY20	.20	BF173	42	OC45	·12
.48	EL84 .23		PY500 .80	U403 .75	ACY21	.21	BF180	.33	OC46	.17
.30			PY500A .80	U404 .49	ACT22	17	BF181	-44	OC70	.14
.50			PY800 .31	U801 .76	ACY28	.20	BF185	.44	0071	.12
		PC900 .45	PY801 .31	U4020 .55	AD140	·40		.17	OC72	12
. 59	EL91 .38	PCC84 .40	PZ30 .48	VP23 .75	AD149	.55	BFY50	-25	OC74	-25
. 59	EL95 .39	PCC85 .44	QQV03/10	VP41 .75	AD161	-50	BFY51	-21	0075	12
. 45		PCC88 .60	1.20	VU133 35	AD162	-50	BFY52	- 22	0076	-17
. 50	ELL80 1.00 EM80 .40	PCC89 .50	Q875/20 £1	W107 65	ADT140	.69	BY100	.20	OC77	-30
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. 29	EM81 .40 EM83 .55	PCF80 ·28	QV04/7 £1	X65 .65	AF114	28	BY114	.20	OC78D	.17
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. 20			TP2620 .98	& Diodes	AF121	.33	BYZ10	.28	OC82	.12
. 86		PCF200 .67	UABC80 .83	2N404 ·20	AF124	.28	BYZ11	-28	OC82D	.12
		PCF801 .48	UAF42 .55	2N966 ·58	AF125	·19	BYZ12	.28	QC83	-22
85		PCF802 .50	UBC41 .53	2N1756 ·55	AF126	.20	BYZ13	.28	OC84	.26
.44	EY81 .40 EY83 .54	PCF806 . 70	UBC81 .45	2N2147 94	AF139	72	CG12E	.22	OC123	.25
. 53 . 60	EY84 .70	PCH200 .70 PCL82 .32	UBF80 .38	2N2297 ·25	AF178	-75	GD9	.22	OC139	.25
.00			UBF89 .35	2N2369 24	AF180	-53	GET113	-22	OC169	.25
34	EY87/6 .33 EY88 .40	PCL83 .54 PCL84 .38	UBL21 .77	2N2613 ·43	AF186	·61	M1	.17	00172	-89
			UCC84 .75	2N3053 -36	BA115	.15	OA5	81	OC300	.24
34		PGL805/85	UCC84 .75	2N3121 2.75	BA116	.28	OA9	.14		-42
75		PCL86 .47	UCC85 .38	2N3703 ·21	BA129	14	OA10	47	OC503	-88
10		PCL86 .47 PD500 1.44	UCF80 .65	2N3709 ·22	BA130		OA47	·11	OC204	-33
	GY501 .75		UCH21 .66	2N3988 ·55		17	OA70	.17	OC205	.47
	GZ32 .54		UCH42 .65	28323 -55		14	OA73	.17		
	GZ32 1.00		UCH81 .88	AA119 ·17	BC108		OA79	10		
	GZ34 .57	PEN453DD	UCL82 .38	AA120 ·17	BC113	28	OA81	·10		
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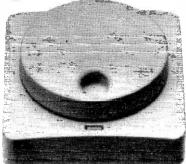
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capacitors /							
	4 VO		16 VO		40 VO		
	$47\mu F$	6 ½ p	$15\mu F$	6½p	$47\mu F$	64p	
	$100\mu F$	6 <u>‡</u> p	$33\mu F$	6½p	100µF	9p 1	
	$220 \mu F$	6½p	$150\mu F$	6½p	68µF	10p	
	$330\mu F$	6≟p	150µF	8p	$220 \mu F$	11p	
	1000µF	13p	220µF	9p	$470 \mu F$	19p	
	4700j.F	29p	680µF	17p	680µF	25p	
			1000µF	17p	1000µF	25p	
	6-3 VC		$1500 \mu F$	25p	2200µF	44p	
	$33\mu F$	6½p	2000µF	43p			ı
	$68\mu F$	61p					ı
	$150\mu F$	. 6½p	25 VO				ı
	$470\mu F$	11p	$10\mu F$	₿₽p			ı
	$680 \mu F$	13p	$22\mu F$	6½p	63 VO	LT	ı
	1500µF	18p	47µF	6½p	$1\mu F$	6±p	ı
	$2200 \mu F$	18p	100µF	8p	2·2μF	6½p	ı
	3300µF	26p	150µF	8p	4.7µF	6½p	ı
	10 VO	IT	$220\mu F$	10p	6.8µF	6≟p	l
		6 <u>₹</u> p	470µF	13p	10µF	6½p	ı
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	$220 \mu F \\ 330 \mu F$	8p 10p	5000µF	68p	150µF	13p	ı
	330μF 470μF	10p	40 VC	N T	220µF	19p	ı
	1000.00	11p		6 <u>}</u> p	330µF	22p	ı
	$1000 \mu F \\ 1500 \mu F$	20p	15μF	6≟p		26p	ı
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working on both negative and positive earth vehicles. It covers the full medium and long wave bands. It is permeability tuned and sturdily constructed. Output is a full 2.5 watts into an 8 ohms speaker. But the Tourist PB will operate into any loud-speaker from 8 to 15 ohms. Apart from the output stage, which is an integrated circuit, the only other electronic components that need soldering are some capacitors, resistors, etc. The kit includes a pre-built RF tuner unit, and fully modulised IF stages which are pre-aligned before despatch. As well as electronic components this kit also contains 2 diamond-spun aluminium knobs, elegant ponents this kit also contains 2 diamond-spun aluminium knobs, elegant matching front panel, dial, washers, screws and wire.

The Tourist PB can be mounted in any standard size dash panel and it has an illuminated tuning scale. Chassis size is: 7in wide, 2in high and  $4\frac{5}{16}$  in deep. Circuit diagram and comprehensive instructions 55p free with parts. Fully retractable and lockable car aerial £1-37 post paid.

CAR RADIO KIT £6.60 p. and p. 55p. Speaker with baffle and fixing strips £1.65, + 23p p. & p. post free if bought with the kit. Send stamped addressed envelope for leaflet. If you can solder on printed circuit board, you can build this push-button car radio kit. It's simple—just follow the step-by-step instructions.



### PE TAPE LINK **CONSTRUCTORS**

Suitable 3 speed tape deck, less heads. Caters up to 5\frac{3}{4}ins. spools. 240V AC mains. Unused but store soiled hence no warranty. £4.00



e e e

ONLY £7.64 + 55p. p & p

For the man who wants to design his own stereo—here's your chance to start, with Unisound—pre-amp, power start, with Unisound—pre-amp, power amplifier and control panel. No soldermynner and control panel. No soldering—just simply screw together. Watts per channel into 8 ohms. Inputs: 120mV (for ceramic cartridge). The 120mV (for ceramic cartridge). The heart of Unisound is high efficiency I.C. monolithic power chips which ensure very low distortion over the audio spectrum.



# IN-GAR ENTERTAINMENT

With this elegant stereo 8 track add on unit, audio enthusiasts now have the opportunity to extend their systems to include the playing of 8 track cartridges. Simply select your channel, by push button, four digital lamps indicate channel selected. The Viscount III, the

fabulous Stereo 21 and the Unisound Modules £10.60 + 90p. p. & p. will accept this unit, simply connect up.

# DIAGNOSIS

Of all the constructional projects that you can possibly think of, which do you think would attract the least interest? Before you answer, take a look at our contents page this month and you will very likely find a clue. It is probably understandable that test equipment does not reach the realms of glamour in most constructors eyes, perhaps because it does not perform an entertaining function. But if you are always looking for entertainment you will be missing out on the "nitty-gritty" of the true do-it-yourselfer.

What does test equipment conjure up in your mind on first impact; a box of knobs, dials and meters, a standard performance by which your other projects are compared, or a dull boring necessity to the spectrum of your fullest activities? All of these are confirmed to us by readers who write to us asking for guidance in fault-finding or setting-

up.

If you have test gear, that's fine; if you bought it ready made, you should know what it can do for you. Most of all its performance and/or limitations must be compatible

with your requirements.

There are areas, however, that are either too specialised or require only occasional application of test gear, making the cost of outright purchase too high. This is where the

constructor comes into his own.

This issue of *Practical Wireless* takes a look at some test equipment and next month starts a series that will help you to obtain the most from oscilloscopes. Some simple add-on circuitry is often useful and we will be publishing some ideas later with constructional details. Also in this issue is the second part of the "Trouble Tracer" which will be valuable for those who like to cure faults on a wide range of radio and audio equipment,

Of special interest is the stabilised 30 volt power supply, designed for transistor and i.c. circuitry; this is invaluable as a workshop "tool" or as a standby supply in the event of failure in other equipment. We also have a neat little unit that can be used for testing Zener diodes, and next month's issue will include a signal generator providing

square, pulse and triangular waveforms.

Test equipment presents an art of its own, the brick wall on which much other equipment likes to lean. It is security to the "home-brew" and peace of mind to the constructor. It has even been described as the difference between furrowed brows and inspired genius—possessing a magical potion that cannot be obtained on doctor's prescription. Probably the greatest asset of test gear is its usefulness as an electronic doctor, diagnosing symptoms and dispensing strange "noises". You, the constructor, are the surgeon.

M. A. COLWELL—Editor.

The March issue will include P.W. CRATA; a waveform generator providing square, pulse and triangular waveforms; a second channel interference rejector (or pre-selector) for 10 to 30 MHz; P.W. Datacards nos. 7 and 8; a crystal calibrator and the start of our series on using oscilloscopes. We apologise that this latter feature could not be started in this issue due to an overwhelming demand on our space.

Further details on page 967

# NEWS...

# Tandy's here!

massive American operation called Tandy is now installed at their new warehouse premises just north of Birmingham. Tandy aims to acquire retail outlets and issue franchises to sell its own branded audio, electronic components and kits.

Their optimism in the face of the existing competition is to be admired, but has already caused some component suppliers to beware of the power of the financial backing of Tandy. For readers of this magazine, they offer a very wide range of components, and goods in the domestic electronics field were packed in wooden crates originating from Taiwan. The shop at the front of the warehouse was large and a browser's paradise. Our reporter. however, arrived with some degree of concern for the British businessmen; even the launching festivities did not deter him from talking to the Area Sales Manager in the shop. Whilst the components side of the business was our main interest, we were surprised to find that "bubble-pack" cards bearing American prices had also been priced in Sterling. A glossy catalogue shows only a part of the range, and readers who acquire one will be surprised to find component prices so much higher than in most British concerns.

Our reporter concluded that Tandy must belatedly study the British market or think again.

## Sonex 74—Latest

THE Sonex, high fidelity exhibition, will be at the Post House Hotel, London Airport, from March 29 to 31st.

The Post House is ideally situated to the M4 Motorway and the airport giving easy access to overseas and British visitors. The organisers, British Audio Promotions Limited, confidently expect that they will be able to provide adequate car parking facilities close by.

## IMPORTANT MESSAGE TO ALL READERS

Readers of "Practical Wireless" and most other magazines will be aware of the restrictions imposed by the Government and as a result of Trades Union action in various industries. To ensure that you receive your copy of "Practical Wireless" at the earliest possible date, we have to rely on effective communications and delivery through the post and by railway. We also depend on the operation of electrical equipment and printing machinery. Petrol and diesel oil are vital to the needs of our job in guick and effective processing of

editorial and advertising material.

We are sure that you will understand the many problems that confront all of us. We therefore request your patience and understanding during any temporary period when publication may be delayed.

It would help you and us enormously if you make sure of ordering your copy of "Practical Wireless" in advance. We try to help you by telling you a little of what will be published in the next issue and what you can expect in future issues of P.W.

# Electronic aid for the disabled

ABOUT five years ago, Toby Churchill—a qualified mechanical engineer working for Lucas—contracted an unidentified virus disease which left him with a number of disabilities. These include a complete loss of the power of speech and a paralysed right arm.

He conceived the idea of a portable electronic device which would enable him to communicate with others easily and set about the task of putting his idea into practice. After reviewing the electronic components available, he knew that his ideas were not just a pipe dream but a practical possibility. The Engineering Department at Cambridge University heard of Toby's ideas and agreed that they would form the basis of a worthwhile project.

Very quickly the ideas became reality. A typewriter-like keyboard was coupled to a Burroughs 'self-scan' display system. Circuits were designed to allow the unit to be powered from rechargeable batteries.

Toby now talks to people using the keyboard: the letters, words and numerals appear in a very easily-read form on the self-scan display panel.

A number of people assisted with the development of the unit—which has been called the Lightwriter—in many different ways. Burroughs, keyboard manufacturers, Cambridge University and Burroughs' UK agents, Walmore Electronics Ltd., all played their parts.

As a result of pressure from friends and acquaintances with similar disabilities, a company—Toby Churchill Ltd.—has been set up to manufacture the Lightwriter. The company has financial backing and production facilities, and it is expected that the first units will become available in the first half of 1974.



The heart of the Lightwriter is the self-scan display panel which is manufactured by Burroughs Corporation in America and available through the UK agents, Walmore Electronics Ltd. It was found that the self-scan display was the only display device available which successfully met all the Lightwriter's requirements.

The display unit consists of a long matrix of special gas-discharge tubes set in cavities. The display used in the Lightwriter has sufficient cavities to form 32 properly spaced alpha-numeric characters, each character being presented in a 5 x 7 dot matrix format.

# **British quad success**

BRITISH research team has developed a quadrophonic sound recording and reproduction system that could be an answer to the complex requirements of broadcasting quad while being adaptable to existing commercial systems. The work has been carried out at the University of Reading in collaboration with IMF, the broadcast monitor loudspeaker company, under sponsorship of the National Research Development Corporation.

In a communication from John Wright of IMF, the system describes a recording technology compatible with existing disc, tape and f.m. broadcasting and will be publicly demonstrated at the Sonex exhibition at the Post House Hotel, Heathrow Airport, in March. The system, called "ambisonics", is aimed at giving the listener the experience not only of the spacial disposition of the performers, but also of the directional qualities of the reverberant sound, thus extending the stereo medium beyond currently used stereo and quadraphonics.

A full account of information so far available has been written for *Practical Wireless* and appears in this issue.

# Mid-Lanark A.R.C.

THE Mid-Lanark Amateur Radio Club have sent us details of their future programme. On 4th January they will have a lecture entitled "The Oscilloscope"—how it works and how to use it. A demonstration will be given by GM8DRQ.

On 1st February there will be an Amateur TV demonstration by GM6ADR/T.

Meetings are held at 7.30 p.m. in the Wrangholm Hall Community Centre, Jerviston Road, Motherwell.

Visitors will be made especially welcome and should 'phone D. H. Plumridge (Hon. Sec.) on Hamilton 28759 if further information is required.



S its title suggests, this instrument was designed as an electronic alternative to a conventional stop-watch. The main design requirements were:—

(a) Good accuracy and reproducibility

(b) Simplicity

(c) Low cost

The low cost requirement ruled out a digital timer, so an analogue circuit was used. The final circuit gives three ranges, 0-5, 0-15 and 0-50 seconds, and the complete instrument including case and meter can be built for about £7.

### **THEORY**

For a capacitor C charged to a voltage V, the charge Q is given by:—

Q=CV.....(1)
However, assuming the capacitor was previously discharged, the charge Q is also equal to the integral of the current I with respect to time t:—

 $Q = \int I dt.$  (2) If the charging current is held constant, this simplifies to:—

Q=It.....(3) Combining equations (1) and (3) gives:—

It=CV and, rearranging,  $t = \frac{C}{I} \times V$ 

If C and I are made numerically equal, the time in seconds can be read directly as a voltage. This forms the basis of the instrument.

# CONSTANT-CURRENT GENERATOR

The basic circuit of a constant-current generator is shown in Fig. 1. With S1 open, no base current flows, and the current shown on the meter is the leakage current, which is so low as to be negligible.

When S1 is closed, current flows through the zener diode and the  $1k\Omega$  resistor, holding the base 5.6V below the supply voltage. Since Tr1 is now conducting, there is a drop of about 0.6V across the emitter-base junction, so the emitter is held constant at 5V

below the supply voltage. This produces a current of  $500\mu A$  in the  $10k\Omega$  resistor and, neglecting the base current, this is also the collector current, which is effectively constant in spite of changes in the supply or collector voltage.

If the meter is replaced by a capacitor, Trl will deliver a constant charging current.

# HIGH IMPEDANCE VOLTMETER

If the voltage across a capacitor (say  $250\mu F)$  is measured using a conventional voltmeter (say  $20k\Omega/V)$  the capacitor will discharge so rapidly that accurate measurements will be impossible. If a very much larger capacitor is used the discharge is less rapid but leakage currents become a problem. The solution is to use a voltmeter having a very high input impedance.

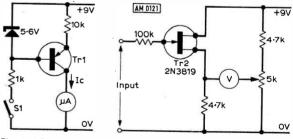


Fig. 1, left: Basic circuit of a constant-current generator. Fig. 2, right: Circuit of high impedance voltmeter.

The circuit is shown in Fig. 2. The FET gives an extremely high input impedance, of the order of  $500M\Omega$  so that the time-constant with a  $250\mu F$  capacitor is about 35 hours! VR1 is used to set the voltmeter to zero when the input is short-circuited.

An interesting feature of this circuit is that the voltmeter does not indicate the absolute value of the input voltage but a proportion of it. However, since all measurements have this constant multiplier (about 0.9 in the prototype) no corrections need be made.

R1 has no effect during normal operation but gives some protection against switching transients.

The capacitor used must be large enough to allow the constant-current transistor to operate at a reasonable collector current but not so large that leakage currents are a problem. In the final circuit a value of  $250\mu\text{F}$  was chosen. Then, with the current at  $250\mu\text{A}$ , the voltage will reach 5V in 5 seconds.

## FINAL CIRCUIT

The final circuit is shown in Fig. 3a. S1 is a 3-pole 4-way switch and is used as an on/off switch and a range selector. S1a connects the battery to the circuit while S1b allows three different currents to be preset. S1c, in conjunction with S3, replaces the capacitor C1 with a fixed resistor so that a calibration check may be made.

S2 connects the base circuit of Tr1 and allows a charging current in C1 to flow while it is closed. (Alternative methods of starting and stopping the timer are given later in the text.)

The diodes D1 and D2 act as a low votage zener diode and hold the base of Tr1 at about 1.2V below the supply voltage as part of the constant current generator.

S4 short-circuits the input to the voltmeter circuit and discharges the capacitor C1 if S3 is "Normal". While S4 is operated VR4 is adjusted so that the voltmeter reads zero.

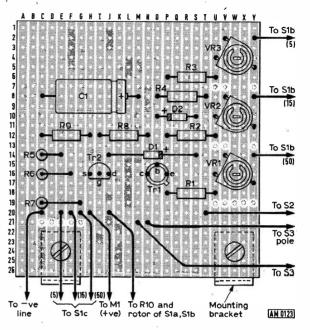


Fig. 3b: Layout of components on 0.1in matrix standard Veroboard. Remainder of components are located on front panel.

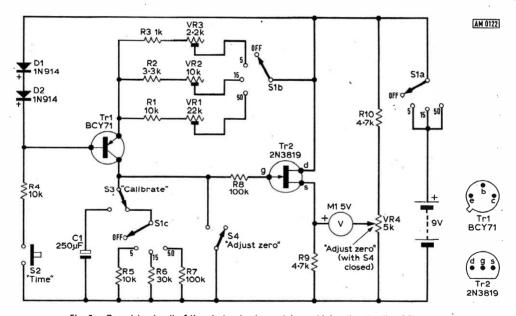


Fig. 3a: Complete circuit of the electronic stop-watch combining the circuits of Figs. 1 and 2.

### CONSTRUCTION

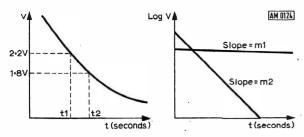
Layout is not critical, and the method of construction may be varied considerably. The prototype used an Elf instrument case which provides a smart and convenient housing.

The meter should be chosen for accuracy and readability. It need not be mounted in the case: a multimeter set to the 5V range and connected by flying leads would work perfectly well, although this arrangement might be rather cumbersome. With a 5V meter, the maximum overload that can occur to the meter is about 50%, which should not cause any damage.

## CALIBRATION

C1 must first be checked for an acceptably low leakage current. S1 should be set to range 1 and the capacitor charged to about 4V. S2 should now be released and a note made of the voltage and the time. The voltage should be checked again 15 minutes later: it should be at least 90% of its original value. If the voltage has fallen below 90% the leakage current is too high and C1 must be replaced (although it may still be suitable for other circuits).

There are two methods of measuring the value of C1 given here, as alternatives.



Figs. 4 and 5: Graphs to illustrate the two methods of determining the value of C1.

Simple method

R11 should be temporarily wired across C1, and the capacitor charged to about 4V. When the charging current is stopped the voltage should fall faster than before. Plot a graph of voltage against time, taking readings every 15 seconds for 10 minutes.

Draw a smooth curve, Fig. 4, through the points and note the time in seconds at  $2 \cdot 2V$  and  $1 \cdot 8V$ ,  $t_1$  and  $t_2$  respectively.

The value of C is given by  $C = 5(t_2 - t_1)\mu F$ .

**Example.** If  $t_1=212$  seconds and  $t_2=280$  seconds, C is 5  $(280-212)=340\mu F$ .

For the mathematically minded, the theory of this method is as follows:—

At any instant  $t_1$  the current through the resistor is given by  $V/1M\Omega = V\mu A$ .

For small changes in voltage, the exponential discharge may be considered linear with time, so that if the voltage falls from  $V_1$  to  $V_2$  the average current through R11 is  $V_1+V_2/2\mu A$ .

Hence the charge lost by C1 is  $V_1+V_2/2 \times t$ . But this is also equal to  $C(V_1-V_2)$ , therefore:—

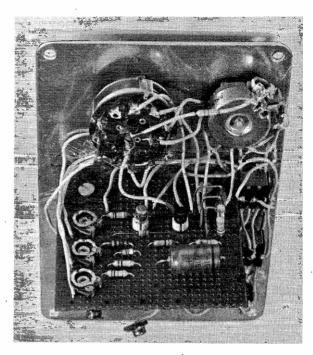
$$C = \frac{V_1 + V_2}{2(V_1 - V_2)} \times t \, \mu F.$$

Although the purist may note that the average voltage is less than

$$\frac{V_1+V_2}{2}$$

# ★ components list

2 28 72 2			333
Resistors:		4)	
R1 10kΩ 5%	R5 10kΩ 29	% R9 4 · 7kΩ 5 % R10 4 · 7kΩ 5	%
R2 3·3kΩ 5%	R6 30kΩ 29	% R10 4 · 7kΩ 5	%
R3 1kΩ 5%	R7 100kΩ 29	% R11 1MΩ 2	0%
R4 10kΩ 5%	R8 100kΩ 5 9	%	Ź
All resistors 1W			ds
largely on the ac			
use 2% or better.			
does not appear			
VR1 22kΩ			
	All skeleton pre-		*
VR4 5kΩ potent			
	J.1.101.01.)		
Semiconductors		2	
Tr1 BCY71	Tr2 2N3819	D1/2 1N914	
6.4			
Miscellaneous			
Capacitor C1, 2	250µF 16V, see	text. Meter, 5VI	DC
FSD (G. W. S	Smith Model S	D640). Switch	<b>S1</b>
4 pole 3 way, b	reak before mak	ce. S2 SPST, pus	sh-
		toggle. S4, SP	
	The same of the sa		
	Perchaard Ratte	ry PP3 and termi	na
push-to-make. \			
push-to-make. \	Instrument c	ery PP3 and terminase (West Hy	



Photograph of prototype showing Veroboard bolted to front panel and interconnecting wiring.

the effective discharge resistance is rather less than  $1M\Omega$ , due to leakage, and these two errors tend to cancel.

### Accurate method

For this method a table of natural logarithms is required. If a capacitor C is charged to a voltage Vo and allowed to discharge through its leakage R, the voltage at time t is given by:—

$$V = Vo e \frac{-t}{CR}$$

Taking logarithms to the base e:-

$$\log_e V = \log_e V_0 - \frac{t}{CR}$$

Plotting loge V against t gives a straight-line graph

of slope 
$$m_1 = \frac{-1}{CR}$$
 (see Fig. 5).

R here represents the parallel combination of the capacitor leakage resistance and the input impedance of the FET voltmeter. A second graph is plotted with R11 in parallel with C1.

The effective resistance is now:-

$$\frac{R\times 1M}{R+1M}$$
 and the slope  $m_2 = \frac{-(1+R)}{CR}$ 

where R is in  $M\Omega$  and C is in  $\mu F$ .

Simple algebra shows that  $C = \frac{1}{m_1 - m_2}$ 

(Note that both  $m_1$  and  $m_2$  are negative, and that  $m_1$  is the correction factor for leakage. The value of the leakage resistance is given by  $R=(\,\frac{m_2}{m_1}\,-1)M\Omega.$ 

Whichever method is used, at least three "runs" should be plotted and an average taken to give the value of C. R11 should now be removed.

Next the constant-current ranges must be set. If the capacitor was measured as  $390\mu F$ , as in the prototype, the current on range 3 (0-5 seconds)

should be set to  $390\mu A$ . Switch S3 to CAL and adjust VR3 until the voltmeter indicates 3.90V. Now switch to range 2 and adjust VR2 until the voltmeter again reads 3.90V. Repeat using VR1 on range 1.

Finally operate S4 and check that the meter still reads zero. If the zero has drifted it must be readjusted using VR4, when VR1, VR2 and VR3 may all have to be reset.

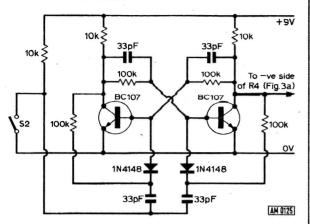


Fig. 6: Additional bistable switch circuit for long intervals, in place of \$4.

As it would be rather inconvenient to hold S4 operated during timing, a bistable can be used so that S4 is pressed once to start, and again to stop timing. The circuit is given in Fig. 6.

It is worth remembering that electrolytic capacitors normally have a tolerance range of -50% to +100% so that the nominal 250 $\mu F$  capacitor may have an actual value from 125 $\mu F$  to 500 $\mu F$ 

# LIGHT OPERATION OF TIMER

Another idea that was used with the prototype was a bistable operated by light-dependent resistors, Fig. 7.

Two LDR's are used, type ORP12. Interrupting the light falling on LDR1 operates the bistable so that Tr2 is conducting, which turns the timer on.

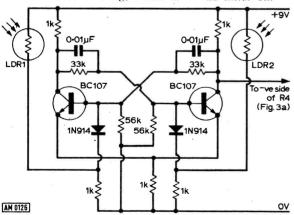


Fig. 7: In this circuit the bistable is operated by LDR's.

Interrupting the light to LDR2 restores the bistable so that Tr2 turns off. R8 may be altered to a lower value which will affect the sensitivity. It may be necessary to increase R6 and R7 to, say  $100k\Omega$  if low gain transistors are used.

# TELEVISION

# IN THE FEBRUARY ISSUE

### CONVERTING FOREIGN RECEIVERS

Every so offen you are likely to be approached by someone wanting a foreign set converted for use in the UK. Just what can and carl't be achieved? Now that we are on 625 lines quite a number of conversions can be satisfactorily carried out without too much trouble. Keith Cummins outlines the approach to be adopted and also takes a look at a type of portable not offen encountered in the UK—the type with a compactron valve line up.

### SIMPLE FET VOM

Even a 20k $\Omega/V$  meter can give misleading readings when transistor stages are being checked. A really high-impedance meter is thus a great help. The use of a field effect transistor provides the solution: Bob MacClay describes a simple meter with an input impedance of  $10M\Omega/V$  ( $20M\Omega/V$  on the 1V range).

### SERVICING TV RECEIVERS

The next chassis to be reported on by Les Lawry-Johns is that used in the Decca MS2000/MS2400 series. This is a hybrid (valves/transistors/i.c.) chassis with several pitfalls for the unwary.

### THE SILICON VIDICON

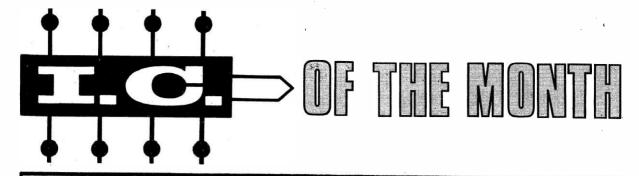
The latest phase in TV camera tube technology is the development of the silicon diode array vidicon. The silicon vidicon has high sensitivity and is not damaged by over-exposure. The basic silicon-diode array will also feature in the all solid-state cameras at present being developed. Ian Sinclair reports.

### SERVICE NOTEBOOK

More hints and reports from George Wilding's TV servicing experiences.

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Number 43

Motorola MC1566L Constant-Current Source

HEN a constant-current source is required and the various advantages offered by the use of IC's are to be exploited, an input voltage limit of 40 or, possibly, 50V is normally necessary if the IC's are not to be damaged.

Neil Wellenstein, an applications engineer working in Motorola's Arizona laboratories, discovered a means of obtaining a variable constant-current supply with input voltages as high as 750V using a standard regulator IC. In fact, the input voltage is limited only by the breakdown voltage of the seriespass transistors employed.

# **Development**

The IC used by Wellenstein was the MC1566L which has the ability to "float" on its own output voltage. However, when used conventionally, a voltage sensitive error occurs in the constant-current mode which is large enough to prevent the device from being used as a precision constant-current source. Normally the constant-current feature of the MC1566L would only be used to provide short circuit protection when the device is employed as a voltage regulator. The magnitude of the current error is small enough to be of no consequence in this application.

The MC1566 contains a current sensing and a voltage sensing amplifier which "float" on the output voltage and which are supplied from an on-chip regulator. The on-chip regulator receives its input from an auxiliary 25V supply external to the chip.

When used conventionally a constant 1mA flows from pin 3 through a resistor to earth to establish the reference voltage for the voltage sensing amplifier. The error voltage appears between pins 8 and 9. When the device goes into the current limit mode (short circuit conditions) part of the 1mA output from pin 6 can flow through a diode to pin 9 thereby upsetting the error voltage and producing a voltage sensitive output current error.

## **Practicalities**

Wellenstein discovered that by reversing the roles of the voltage and the current sensitive amplifiers, he could eliminate this problem altogether. The accompanying drawing shows the circuit, Fig 1. The net effect is that any portion of the reference current that appears in the load must pass through the current sensing resistor R9 which cannot be bypassed as was previously the case.

The maximum input voltage to the circuit is

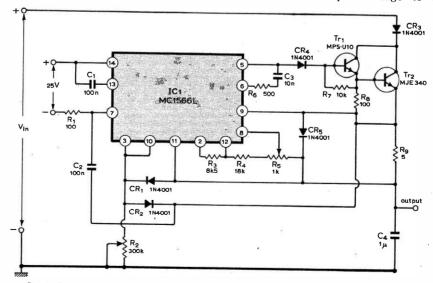


Fig. 1: Practical circuit of the constant-current source fed from a high voltage supply.

To record, without addition or loss, the total directional reverberant information of a live performance inherently needs a microphone technique responsive to the direction of arrival of the sound and capable of delivering this directional information in a format suitable for recording. Such microphones do not at present exist in a commercial form, but have been improvised using tetrahedral arrays of cardioid microphones in promising experiments by Michael Gerzon of the Mathematical Institute, University of Oxford.

An integrated microphone exhibiting these properties is under development with the assistance of Calrec Audio. Such a microphone will sense sounds from all directions, including those from above as well as from all around horizontally.

Sounds having a vertical component cannot be avoided, and the microphone must be designed to respond in a suitable manner so that this vertical information can either be retained or rejected electronically later in the system according to requirements. The signals generated from such a microphone could be amplified and fed directly to four loudspeakers, but *not* with these speakers positioned in the conventional square array.



Similarly the signals, for purposes of information storage, may be fed to a four-channel tape recorder, but again the resultant tape would not be directly suitable for conventional playback.

By the use of suitable decoders such a tape could be replayed through any number of loudspeakers purposefully arranged; these arrangements specifically including the commonly proposed four speaker array or two speaker stereo or even true mono.

This four track tape, or the signals direct from the microphone, may be encoded onto two channels of information in such a way that allows the original directional information to be in large measure decoded for multi-speaker playback. The two channel signal, be it on disc, cassette or f.m. radio, is directly suitable without decoding for stereo presentation a la Blumlein and with no more than the usual adverse compatibility to mono when the two channels are directly paralleled.

# THREE CHANNEL ENCODING

The original four signals can also be encoded onto the three audio channels potentially available

within the bandwidth of f.m. broadcasts, to provide marginal improvement particularly in phase relationship and mono/stereo compatibility. The use of a third channel on disc, possibly of reduced bandwidth, has been advocated by Prof. Duane Cooper of Illinois for similar reasons.

Finally, encoding the four signals onto four tracks of tape or a multiplexed disc such as the JVC-CD4 makes optimal use of the capabilities of four genuine audio channels and enables the full recovery of all information, including that of height, without phase anomalies. Demonstrably there are more versatile ways of employing four independent channels of information than merely feeding them to four loudspeakers.

# COMPATIBLE QUAD

Certain of the commercial 'quadraphonic' systems have features consistent with ambisonics, without however exhibiting all its essential properties. Most notably the Nippon-Columbia UMX system of Cooper & Shiga comprises a compatible series of 2-, 3- and 4-channel codings basically consistent with ambisonics although hitherto confined to pan-potting in horizontal angle.

The Japanese 'Regular Matrix' defines a potentially ambisonic 2-channel system, but is vague about the distinction between internal and overhead sounds. Recordings made on any of these systems could be played back through an ambisonic system, with limitations in what could be reproduced but without glaring anomalies, by suitable switching between the basic units of the ambisonic decoders.

It is apparent that true compatibility between different systems, numbers of communication channels, and numbers and configurations of the loud-speakers the listener may use, can come only from the ambisonic approach of considering how to record the directional character of all the sound that may reach the microphone, and then considering how this full directional information can be collapsed down successively to horizontal-only, stereo and mono presentation.

# RICHER GAMUT

Experiments with systems aimed at reproducing the true directional quality of original sound shows how restricted are the methods of synthesis and panpotting in current use. Ambisonics provides a much richer gamut of possibilities.

In so far as it is possible to imitate synthetically the signals due to any sound that the system could record naturally, the possible effects include a previously recorded or synthesised signal being made to appear to recede into the distance or rush up to the listener, changing the character of its reverberance and 'atmosphere', swooping around the listener or even looping the loop.

In addition there is the possibility of creating sounds that never existed in nature, including the so-called 'internal' or 'in-the-head' sounds. At the present stage of development no limits can be placed on the possibilities thus opened up for future developments of the art in synthesised music of 'pop' effects.

In what is called 'serious' music, ambisonics can be the means of reversing the current trend of elaborate recording techniques coming between the listener and the performance. It represents a return to the concept of creating a sound at a good listening position at the live performance, and then recording as accurately as possible the characteristics that give this sound its live quality; surely what 'high fidelity' reproduction is all about.

## **PARAMETERS**

The physical basis of these developments can be expressed in terms of wave-equation of sound. The sound-field at a point is completely determined by knowledge of the pressure p and the three resolved cartesian components of fluid velocity  $v_x$ ,  $v_y$  and  $v_z$ . These four parameters can evidently be transmitted along four communication channels each of audiobandwidth.

A natural way of doing so has been discussed, with much other relevant theory, by Gerzon, in terms of signals representing the four spherical harmonics of order zero and unit, namely 1, x, y, z (where x, y and z are direction cosines and 1 represents an omni-directional component). Note that this or equivalent methods are optimal in the use made of four channels, and are quite distinct from the inferior way the four channels are used in 'quadraphonics'.

If attention is confined to progressive sound-waves, without any standing-wave component, knowledge of the three velocity components  $v_x$ ,  $v_y$  and  $v_z$  determines the pressure p uniquely except for an ambiguity of sign caused by the inability to distinguish two waves of opposite phase travelling in opposite directions.

Thus knowledge of p is partially redundant, and this redundancy can be used to compress the information into three channels each of audio bandwidth, if some limitations are accepted. With respect to sounds arriving horizontally, the transformation between four and three channels can be made completely reversible, so that for this application three and four channel systems are strictly equivalent.

# TO TWO CHANNELS

The possibility of compression (with some further compromise) into two channels can be expressed in terms of representing the direction of arrival of sound by the two parameters of horizontal angle  $\Theta$  and vertical angle  $\Theta$  (i.e. azimuth and altitude angles). Two degrees of freedom are available in a two channel signal, after normalisation to represent the intensity of the sound, namely the relative amplitudes and the relative phases of the signals in the two channels. Thus by using phase information, the three-dimensional sound field can be correctly identified.

Developments in the realisation of the above theoretical possibilities are the subject of current patent applications in the U.K., Canada, Denmark, France, Germany, Italy, Japan, Netherlands, Switzerland and the U.S.A.

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limited by the series-pass transistor. In the case of the MJE340 shown, the maximum input voltage is 300V. The circuit provides a constant-current output which is adjustable from  $200\mu A$  to 100mA; above 10mA take care not to exceed the ratings of the MJE340. At both the  $200\mu A$  and the 1mA settings, output impedance exceeds  $200M\Omega$ .

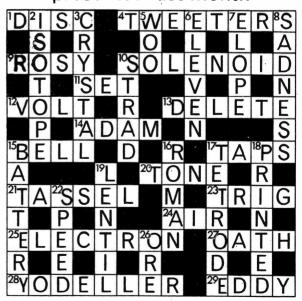
The  $1\mu F$  capacitor C4 is necessary to ensure circuit stability. However, it limits the rate at which the voltage across the load can change in response to a sudden change in load resistance. This response time can be found by multiplying the capacitor value by the final load resistance. The instantaneous load current is found by dividing the instantaneous output voltage by the final value of load resistance.

The MC1566L is made to a military specification, the 'civilian' equivalent being the MC1466L. The MC1466L costs £3·30, the MPSU10 is 75p and the MJE340 (or BD158) is 41p, all obtainable from Jermyn Home Electronics, 112 Vestry Estate, Sevenoaks, Kent. Add 25p for P/P plus 10% VAT to total cost.

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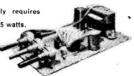
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ECH84 40n CV24 80n PL509 £1:10 UAF42	55p UL84 40p 3D6 40p 6AB7 45p 6B7	55n 6F33 £3-25 12BH7 50n 956	40p £6.00 80p WL417A								
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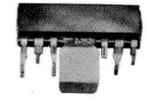
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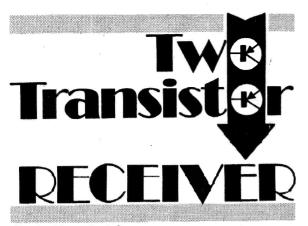
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# R.A.Penfold

ALTHOUGH this receiver uses only two transistors, it is capable of receiving local broadloudspeaker. In most areas it will also receive several foreign stations after dark. There is also provision for a crystal earpiece.

The set is self contained, having an internal 9 volt battery (PP4),  $2^3_4$ in. dia. loudspeaker and ferrite rod aerial. It tunes the medium wave band only and measures  $6 \times 3^1_8 \times 1^5_8$ in., excluding control knobs.

As the circuit uses only the minimum of components, it is both inexpensive, and easy to construct and has proved to be stable and reliable. It is a feature of the receiver that once it has been constructed, no special alignment is required before it can be used.

### How it works

The circuit consists of a reflexed stage with controlled regeneration, coupled to a single transistor operated as a class A output stage.

The circuit diagram of the receiver is shown in Fig. 1. Tr1 stage provides the majority of the circuit

gain. L1 is the tuned winding of the ferrite aerial, and this is tuned by VC1. L2 couples the r.f. signal to the base of Tr1. Tr1 is biased by R1, and C5 plus R2 provide the supply decoupling. R1 is taken to the junction of T1-R2 so that no a.c. negative feedback is introduced, but a certain amount of d.c. feedback is, and this has a stabilising effect on the biasing of Tr1.

The primary winding of T1 forms an a.f./r.f. load for Tr1, and an amplified r.f. signal will appear at Tr1 collector. From here it is coupled via C4 to D1. The r.f. signal is detected by D1. It is important that D1 has the polarity shown.

The audio signal present at the junction of D1-C3 is coupled to the base of Tr1 via L2 winding. Tr1 now operating as an audio amplifier, with T1 primary forming an a.f. load.

In order to increase sensitivity, and also selectivity, regeneration is applied to Trl stage. This is the process of feeding some of the r.f. signal at Trl collector back to the base of Trl (via L1/L2) for further amplification.

Trl inverts the r.f. signal, and from Trl collector the r.f. signal is coupled via C2 to VR1 slider, and from here via C1 to L1/L2. VR1 controls the regeneration, and maximum sensitivity is obtained with this adjusted just below the point at which the circuit breaks into oscillation. It would appear logical to connect C2 to the top of VR1, and C1 to the slider, but when this was tried the circuit was rather unstable.

Trl and Tr2 are coupled by transformer Tl, C6 providing the necessary d.c. blocking between the base of Tr2 and the d.c. path to earth through Tl secondary. Tl is a miniature driver transformer for two OC72's. The centre tap on the secondary is ignored. Various transformers were tried and despite differences in impedances and ratios, they all gave virtually identical results.

Tr2 operates as the output transistor. This is biased by R3. The loudspeaker forms the collector load for Tr2 when the loudspeaker is in use, and R4 forms the load when the earpiece is used. The jack socket for the earpiece has a break contact which is used to cut out the loudspeaker from circuit when the earpiece is plugged in. As R4 has a relatively high value it can be left in circuit when the loudspeaker is being used, without having a detrimental effect on performance. C7 introduces a certain amount of negative feedback at the high audio frequencies which helps to improve the audio response.

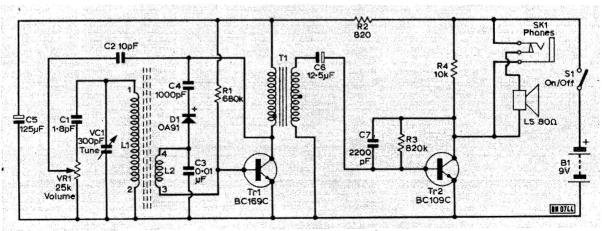


Fig. 1: Circuit of the two transistor receiver.

### Ferrite aerial

This is wound on a  $4^12 \times ^38$  in ferrite rod. This size should be readily available. Enamelled or d.c.c. wire is used, the exact gauge not being too important. Something in the region of 30-32 s.w.g. is a good choice as this is fairly easy to use, and gives a reasonably compact winding. Fig. 2 shows the construction of the aerial.

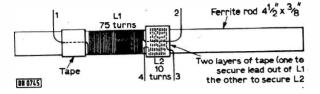


Fig. 2: The ferrite rod aerial.

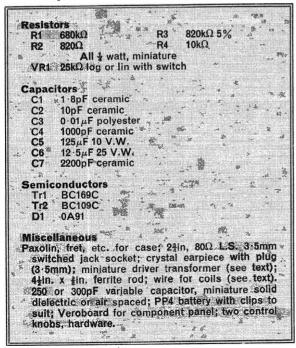
As can be seen from Fig. 2, L1 winding consists of 75 turns of wire. This wound in a single layer, and should be kept as tidy as possible, avoiding any overlaps. The leads at each end of the coil are taped to the rod, using insulation tape, so as to hold the coil in place. The coil is wound slightly off centre, so that when the aerial is mounted in the case it can be positioned slightly away from the speaker.

L2 winding consists of 10 turns of wire wound over one of the pieces of insulation tape, with a further layer of tape placed on top of this in order to hold it in position. The coil has a single layer, and again it

should be free from overlaps.

The case is made from hin. sheet paxolin, or a similar material. A cut-out for the speaker is made using a coping or fret saw. This and the other holes should be made before the various parts are

# \* components list



assembled. The way in which the parts fit together can be ascertained from the diagram. A good quality household adhesive, such as Araldite is recommended.

Two small blocks of wood approx.  ${}^5_4 \times {}^5_8 \times 1$ in. are glued to the sides of the case (as shown in Fig. 4). The rear of the case is held in place by two wood screws which pass through two holes already drilled in the back and then screw into the two small blocks of wood.

The method of mounting the ferrite rod can be seen in Fig. 4. The two blocks of wood on which it is mounted measure approx.  $1 \times {}^5 8 \times {}^1 4$ in. and  $1 {}^1 8 \times {}^5 8 \times {}^1 4$ in. The rod is mounted as far to the rear of the case as possible. A  ${}^3 8$ in. dia. hole is drilled in each of the blocks of wood, and the rod is pushed into these (it need not be glued). The blocks are then glued to the case.

Some of the initial wiring can now be commenced. This is shown in Fig. 4. This also shows the position of the battery and Veroboard component panel.

## Mounting the loudspeaker

A piece of speaker fret is glued behind the speaker cut-out, and the speaker is then in turn carefully glued to this. A sticky backed plastic material is used to cover the case, and so give a smart finish.

VC1, VR1, and the jack socket can then be mounted. VC1 may have a  $^3$ <sub>8</sub>in. dia. single hole fixing, but many types have a two or three hole screw fixing. In these cases it is best to glue the capacitor to the front panel.

# Veroboard panel

With the exceptions of Cl, and R4, all the small components are mounted on a small piece of 0·15in. matrix Veroboard. This has the copper strips running lengthwise. The component layout of this panel is shown in Fig. 3.

If T1 is of a type, which has a mounting clip and flying leads, the clip should be removed and the leads cut short so that it can be treated as a printed circuit type.

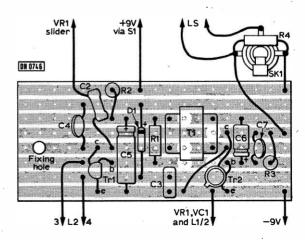


Fig. 3: Veroboard panel component layout.

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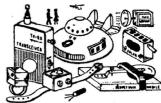
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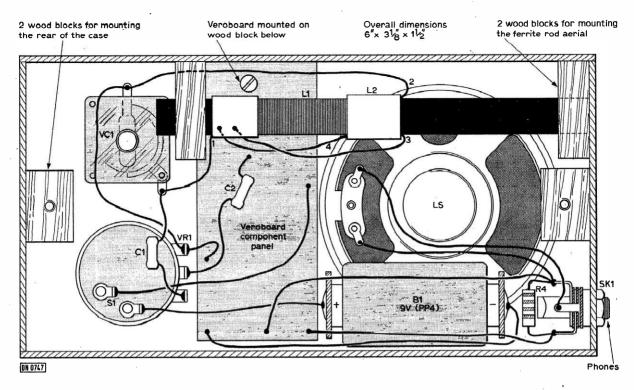


Fig. 4: General construction and layout details.

The 10pF capacitor C2, should have its lead out wires left full length. These can be insulated with sleeving, but this is not essential. The free end of C2 should not be connected to VR1 slider until the Veroboard panel has been mounted

The method of mounting the board is quite simple. A piece of wood approx.  $1 \times^{3}_{4} \times^{1}_{4}$  in. is glued to the inside of the front panel, at the top between the speaker fret and VC1. The Veroboard panel is then positioned as shown in Fig. 4, and a small wood screw is placed through the mounting hole and screwed tight into the block of wood.



Fig. 5: Transistor base connections.

## Using the Set

For initial testing it is suggested that the earpiece is used. If VR1 is adjusted to turn the receiver on, and VC1 is then adjusted, it should be possible to receive two or three stations, although these will probably be rather weak. If VR1 is rotated further clockwise the stations should become louder. If VR1 is turned too far the receiver will begin to oscillate, and whistles will be heard from the earpiece as the set is tuned across the band. If the receiver fails to oscillate, try reversing the connections of L2 winding. The receiver is most sensitive with VR1 set just below the point at which oscillation occurs.

As the ferrite aerial is directional, the set should be rotated for optimum signal.

If the set works properly using the earpiece it can be tried using the speaker. For satisfactory speaker reception it will probably be found necessary to always adjust VR1 for maximum sensitivity and it will need re-adjustment each time the set is retuned.

## Conclusion

The receiver may suffer from hand capacity effects. This is where putting ones hand near VC1 tends to slightly alter the tuning. This is a common failing of simple circuits of this type where a single tuning capacitor with a non-metallic chassis is used. It can be eliminated by placing a sheet of aluminium between the front panel and VC1-VR1. The aluminium panel should be connected to the negative supply.

Current consumption of the unit is approximately 11mA when using the loudspeaker and 4mA when using the earpiece.

## OSCILLOSCOPE TECHNIQUES

Part 1 of this series due to commence in this issue has been held over until the March issue because of pressure on editorial space.

# practically wireless HENRY commentary by

TELL it not in Gath. Certain of our exalted boffins are dreaming up new ways of overcoming noise. No, not polishing up Dolby—those lads are quite capable of effecting refinements themselves; you should see the latest Japanese version of the Dolby-B circuit, shorn of all but its essentials!

There are alternatives. Now that f.m. broadcasting has become and will continue to become, an everyday part of our lives, some means of bettering signal-to-noise ratio is needed. What we suffered from mono, because we could make a direct and favourable comparison with a.m., is intolerable when 'we listen to stereo v.h.f. broadcasts.

This is partly psychological. In this country, some parts at any rate, stereo radio is very new: in some parts, still a vague promise. We tend to want our money's worth. We listen more intently. We wait, with bated breath, for some evidence that the studio engineer has his wires crossed and is allowing us to eavesdrop on a prompt from the wrong side of the wings.

The trouble, however, goes deeper. We now have the artistes aware of anti-noise possibilities; feeling they may have been cheated out of their full whack of exposure unless the engineering department has done its damnedest. Can you imagine



Evesdrop on a prompt.

someone like Arthur Garratt chatting scientifically to us without the full benefit of every decibel available; or James Burke, let alone Raymond Baxter, doing a 'what's to come' piece without the full armoury of technics at their backing?

Henry was particularly amused by the Larry Adler comment that Woman's Own would not let him retain his title for an article: 'What kind of noise annoys an Oistrakh?' They etiolated it to: 'No Noise is Good Noise.'

Beneath the levity is a serious question. What sort of noise would annoy an Oistrakh? Certainly not always the same sort of noise that perturbs ordinary mortals like you and me. Noise is a subjective thing. In an iron foundry, Thor could bash on regardless, but while I am tapping out this piece, the clack of a pair of platform soled shoes beneath my window can be utterly distracting. Well, I mean, they have a new au pair just down the road . . .

In the middle of the night, noise is the breathing of a wayward moth. To Patrick Moore, 1420MHz is probably the bête noir frequency\*, while to sundry girl operatives in a factory, constant vibrations at 37Hz caused genuine excuses for sick headaches.

Probably the noise that would annoy an Oistrakh would be a second fiddle slightly out of tune in the opening of the Bach Double Violin Concerto. You can't do much about that with Dolby.

Trouble with the f.m. broadcast situation is that everybody has to be in on the act, or there is no point in applying signal tailoring. The other trouble, less often mentioned, is that current systems act on signals below a certain arbitrary level and within certain frequency bands. Which is all very well, but what about the background noise that accompanies some high level sounds and can still be heard?

No. 102

No noise



Breathing of a wayward moth

Studio types will tell you that "compandor" systems carry that particular penalty. Get a few of those big bass drum thwacks in the Panufnik "Heroic Overture" overriding a musing woodwind, and the processing inserts a 'noise swish' that has become so much a familiar phenomenon to us that we think the live performance lacks something.

Just like the old lady who complained after a concert that the orchestra did not sound as 'mellow' as on her pre-transistor radiogram.

Even the best BBC boffins will not be able to reinsert her "lost" frequencies in the right proportion. They cannot even do much to restore power lost to you by your distance from the receiver. It is a point not aired often enough by those hi-fi salesmen who insist (metaphorically) on assuring customers that they can get good stereo on a discarded clothes-line! That is, transmitter power will not of itself determine the range of broadcast reception. This is determined by the height of the aerial, the topography of the earth, humidity, temperature and other atmospheric mysteries. What the extra power does is precisely what we are here con-cerned with: it improves the signal-to-noise ratio.

\* 1420MHz is the radiation frequency of the neutral hydrogen that results from the collision of hydrogen atoms: the intergalactic wavelength.

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	I · OµF	35V	10µF	25V	100uF	3V	

VEROBOARD  21 × 32 22p  23 × 5 24p  32 × 5 24p  33 × 5 27p  17 × 21 75p  17 × 32 (plain) —  17 × 32 (plain) —  21 × 5 (plain) —  21 × 5 (plain) —  21 × 32 (plain) —  21 × 32 (plain) —  21 × 32 (plain) —	Dack   Dack
Pin insertion tool 52p Spot face cutter 42p Pkt, 50 pins 20p	52p 42p BATTERY ELIMINATOR £1.50 9V mains power supply. Same size as PP9 battery.

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0·01μF 10p	0·047μF 13p	0·22μF <b>20</b> p
0·022μF 12p	0·1μF !3p	0·47μF 22p
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3 Pin	5A	Chassis Plug Line Socket	16p 15p	2 Pin	5A	Line Plug	20p

Line Socket 15p	2 Pin 5A Line Plug 20p
THERMISTORS  VA1005 15p  VA1026 15p  VA1033 15p	ROTARY MAINS SWITCH D.P. 2A 32p
VA1055S 15p VA1066S 15p VA1077 15p R53 £1 35	LINEAR IC's 709 14 pin DIL 40p 741 8 pin DIL 40p 741 14 pin DIL 38p
WAVECHANGE SWITCH 23p Ip 12W, 3p 4W, 2p 2W, 2p 6W, 4p 3W	723 14 pin DIL 95p 747 14 pin DIL 85p 748 8 pin DIL 45p DIL Sockets 14 pin and 16 pin 16p

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# transiter BEGGGGR. R.WILDING

RANSFORMERLESS complementary output stages are widely used in transistor receivers and amplifiers, since they confer the advantage of push-pull operation without the disadvantages of wound components.

The bases of the output pair are directly fed from the collector of the preceding driver stage and it only becomes necessary to include a low value resistive component between driver load and collector to develop the required p.d. for establishing their nosignal forward bias.

When germanium output transistors are used, requiring a base/emitter p.d. of about 0.2V, a silicon diode will develop the voltage to bias them both, but if silicon transistors are employed, two silicon diodes will be necessary.

Alternatively, a low value fixed or variable resistor may be used, and shunted by a miniature thermistor so that as ambient temperature increases, the latter's resistance decreases to reduce forward bias and thus stabilise the mean operating current. In some receivers a diode-connected transistor may be used and, if of similar thermal characteristics to the output pair, will compensate for temperature changes in similar fashion to a thermistor.

Current designs employ diodes, thermistors, fixed and variable resistors in many permutations and some of the more common ones are shown in Fig.1 together with a typical basic transformerless output stage. R1 constitutes the driver collector load with the forward biassed diode developing the required potential for Class B operation, R2 and R3 stabilising output transistor working and minimising their spreads in characteristics.

When the driver collector voltage rises, b/e potential of Tr2 rises to increase its conductance, but the b/e potential of Tr3 is reduced to lower its conductance still further from the Class B position, and into cut-off.

When driver collector voltage reduces, the reverse happens, so that each transistor virtually only amplifies one half-cycle of the applied signal. By carefully selecting the no-signal operating point just above cut-off and therefore the most curved section of the transfer characteristic, highly efficient working with minimum crossover distortion can be achieved.

In the latest GEC 2541/Sobell 1541 portables, the amplifying properties of a transistor are utilised to give particularly good bias compensation against variations in amplifier supply voltage and ambient temperature.

The circuit is shown in Fig.2, with the output from a BC148 pre-amplifier being capacitively fed to audio amplifier Tr64.

This latter stage also DC stabilises the output

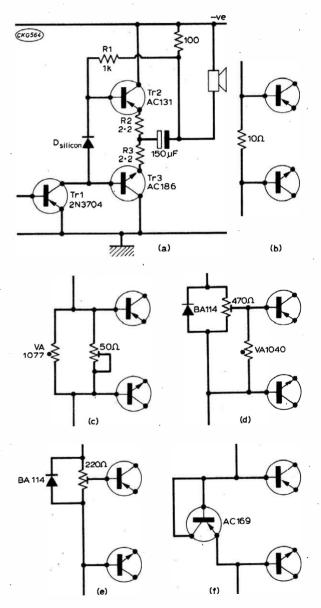


Fig. 1: (a) Basic complementary push-pull output stage, obtaining bias for the matched transistors from the voltage developed across diode. For germanium transistors, one silicon diode will develop the required p.d., but for silicon transistors, two silicon diodes in series will be necessary. Fig. 1 (b) to (f): Alternative methods of providing forward bias, showing typical values, exact values depending on transistors used and supply voltage.

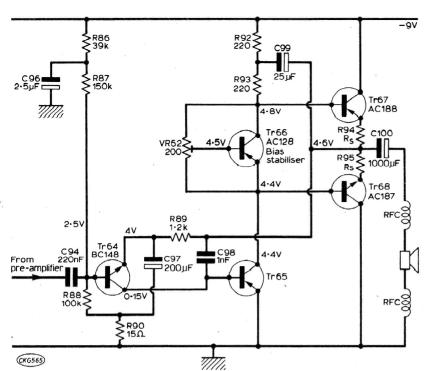


Fig. 2: DC coupled driver and output stages used in GEC 2541/Sobell 1541 six waveband portables. The complementary pushpull output stage is stabilised in two ways, by Tr64 which equalises their mean operating voltages via Tr65, and by Tr66, which compensates their forward bias against variations in supply voltage and amblent temperature.

Fig. 3: AF circuit used in several BRC portables, utilising a saturated transistor Tr10 in series with the supply to provide forward bias to the push-pull output stage. With Tr6 being DC coupled to Tr7, variations in the former's emitter current due to supply or temperature changes produce changes at the junction of R36/37 which effect compensating alterations to Tr10's bias and the voltage developed across it.

transistors, for with its emitter resistor being returned to the junction of their emitter resistors, while its base is held at a fixed potential by the R86/87/88/90 chain, any variation in the voltage at the output transistors emitters produces collector current changes in Tr64, which, due to DC coupling being maintained throughout the circuit, alters the individual Tr67/68 biasing of equalises their working conditions.

Normally, the voltage at the junction of R94/95 is -4.5V, but if this tends to rise due to inequalities in the conductivity of Tr67/68, the emitter voltage of Tr64 will also be raised to increase its forward bias.

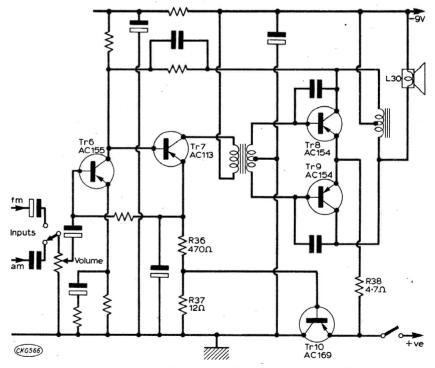
Tr64's collector current will therefore increase and as this is also the base current of driver Tr65, the latter's col-

lector current will increase but decrease its collector voltage applied to the bases of the output transistors.

Net bias to Tr67 will therefore reduce, but increase to Tr68 and result in a lowering of the voltage at the junction of R94/95 to its original value.

Bias compensation is achieved by utilising the voltage developed across Tr66 and VR62 to forward bias the output transistors.

Normally, collector voltage of Tr66 is -4.8V and emitter voltage is -4.4V, and as the junction of



R94/95 is -4.6V, ignoring the small voltage drop across resistors, the output transistors are each forward biassed to 0.2V.

Due to the collector current of Tr64 being the base current of Tr65, the latter transistor becomes very sensitive to small changes in supply voltage and temperature. Increases in Tr65's collector current produce proportional increases in the voltage developed between the slider of VR62 and Tr66's emitter to reduce the c/e potential and therefore

forward bias to the output stage.

Bias for the transistors in conventional push-pull transformer circuits is usually obtained from a potential-divider across the supply, invariably including a thermistor for compensation against thermal changes.

However, several BRC receivers incorporating this type of circuit employ a transistor specifically for bias compensation whether from changes in temperature or supply voltage.

A typical example of such a circuit is shown in Fig. 3 where the compensating transistor, Tr10, is placed in series with the positive supply line to chassis.

The emitters of the output transistors are returned to positive by the  $4.7\Omega$  resistor R38, while the bases are taken directly to chassis via the centre-tapped secondary of the driver transformer.

The chassis is slightly negative to positive supply by the voltage developed across Tr10 which in turn is regulated by the forward bias tapped from the junction of R36/37 in the emitter lead of Tr7.

As Tr6 is DC coupled to Tr7, variations in the former's collector current, due to changes in supply voltage or ambient temperature, become amplified by the latter and produce proportional changes across R37 to vary the bias applied to Tr10 and thus its collector/emitter p.d.

Although worked in a highly saturated condition, Tr10 dissipates only a small wattage since its collector-emitter voltage is only a fraction of 1V.

## SINGLE STAGE FEEDBACK

The simplest method of introducing negative feedback to one transistor stage is to omit the emitter resistor's decoupling capacitor, so that as emitter voltage follows base voltage, the effective b/e potential is reduced to reduce stage gain.

When the emitter resistor is fully decoupled, ie, by a capacitor large enough to hold emitter voltage constant at the lowest frequency handled, full gain is obtained.

A less widely used but very effective arrangement

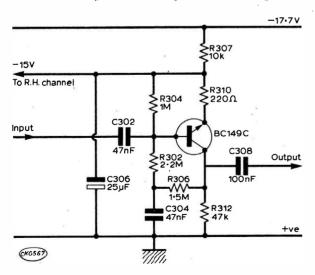


Fig. 4: Low noise audio input stage of GEC six + six stereo unit, forward biased from the collector, but with the AF signal largely filtered out by decoupler C304. R306/302 therefore DC stabilise the transistor while the undercoupled emitter resistor R310 introduces negative feedback.

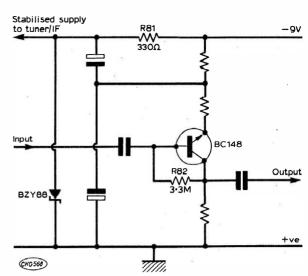


Fig. 5: AF pre-amplifier used in the GEC|Sobell range of portables forward biased by R82 directly from the collector, and applying both negative feedback and DC stabilisation.

is to supply the base bias feed from the transistor's collector instead of from the lt rail.

This can be done whether emitters or collectors stem from chassis, and as well as introducing signal negative feedback, also stabilises the transistors DC working conditions, since an increase in I<sub>c</sub> for any reason increases the voltage drop across the collector resistor to reduce forward bias and restore the original position.

For DC stabilisation without signal negative feedback, it is only necessary to split the base bias resistor and decouple the junction with a high value electrolytic capacitor.

An example of such an arrangement is given by both of the low noise audio input stages of the GEC 6+6 stereo unit, as shown in Fig.4.

R312 is the collector load stemming from positive chassis, while R306 and R302 provide forward bias directly from the collector.

Both being of very high value, practically all the

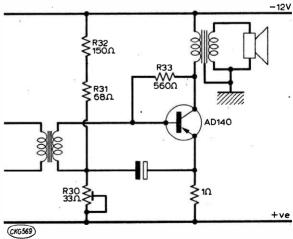


Fig. 6: Single-ended output stage used in many Philips car radio portables. R33 provides signal negative feedback, but not DC stabilisation, due to the primary resistance of the output transformer lonly being one ohm. R30 is set to produce a no-signal collector current of 550 mA.

## TAKE 2® DAVID ANDREWS

## No. 57 NEON FLASHER

## A series of simple transistor projects, using not more than twenty components.

T must be admitted that the possible applications for this circuit are rather few and far between but it is, nevertheless, offered as a novelty that some readers might like to exploit. The author recalls seeing an advertisement a few years ago for a present for the "man with everything"—it was a heated toilet seat with a built-in flasher to act as a beacon in the dark! This circuit could possibly be used in the latter function without risk of electrocution or perhaps it could be used as a sleep inducer in the child's bedroom!

## **Operation**

The function is to make a small neon tube flash at a rate of about one flash per second, the driving source being a nine volt battery. Current consumption is very small, approximately 1 to 2 mA (average), thus a small battery could power the unit for considerable periods. The unijunction transistor Tr1 Fig. 1 forms a relaxation oscillator that produces a positive-going pulse at base b1 approximately once a second. Frequency of operation is set by R2 and C1. For the above quoted rate R2 ought to be  $10k\Omega$  but slower rates can be obtained by increasing its value to  $100k\Omega$ . Should faster rates be required C1 can be reduced in value proportionately.

The positive-going pulses from the oscillator are fed to the emitter follower Tr2 that provides suffi-

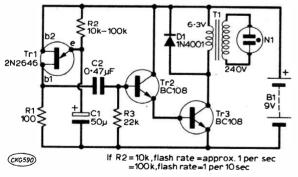


Fig. 1. Circuit of the neon flasher. R2, which, with C1, controls the flash rate could be made variable using a potentiometer. Additionally, C1 could have switched values.

## \* components list

R1	100 ohn	n 🧓	w. 7	Tri	2N2646	A
	10kΩ (s		· ·		BC108	
. R3	22kΩ				BC108	
C1	50μ <b>F (6</b> )	V) 💮			80V nec	
C2	0·47μF 1N4001	(polyes	iter)	T1 _	240V/6·	3V .
וט	TIMACOT	****				, <b>6</b> ,

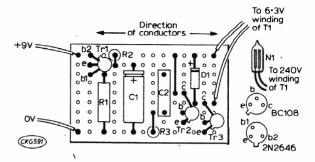


Fig. 2. The circuit can be constructed on veroboard as shown above. The board, transformer and neon tamp could be fitted into a suitable aluminium or plastic box.

cient drive to cause Tr3 to go into saturation. The rapid rise in current through Tl's primary induces a high voltage across its secondary windings and this causes the neon to strike. This is repeated every time Tr1 goes through the conductive part of its oscillating cycle.

D1 should not be omitted as it protects the base collector junction of Tr3 from high reverse voltages when Tr3 switches off. The transformer can be any small mains one that has a 6.3V output; note, however, it is connected into the circuit with the 6.3V winding as the collector load and the neon connected across the 240V winding.

## TRANSISTOR BIAS & FEEDBACK

—continued from page 963

AF signal is filtered to chassis through the decoupler C304.

In the latest GEC/Sobell 6 waveband portables, negative feedback and DC stabilisation is applied to the pre-amp stage by a  $3.3M\Omega$  resistor from collector to base, further feedback being developed across the unbypassed resistor in its emitter lead, Fig. 5.

In common with the other AF stages, this stage is operated from the negative 9V rail, R81 and zener diode BZY88 providing a stabilised supply for the tuner and i.f. stages.

Most transistor receivers employ a push-pull output stage of one form or another, with a feedback loop extending from the speaker to the driver stage, but in some Philips car radio models, a single AD140 is used, signal feedback being provided by a  $560\Omega$  resistor from collector to base.

A typical example is shown in Fig. 6, but no DC stabilisation is provided by the resistor in this case since the DC resistance of the output transformer primary is only  $1\Omega$ , but effective signal feedback is achieved since the output transformer's impedance to AF is comparatively high.

Forward bias is therefore provided by R32/R31, and to a lesser extent by R33, since it is much higher in value, while R30 sets the correct no-signal collector current at 550mA.



## THE TRANSMITTER

The transmitter, Fig. 6, consists of just two NE555 IC's and a few resistors plus capacitors. More resistors may be switched in as required to provide more tones.

The transmitter uses four channel tone encoding, but the receiver only employs two of these. One for 'ON', one for 'OFF'—the other two being spare for future use.

As stated earlier, the simplest of the command functions is accomplished by detecting the presence of the ultrasonic carrier, IC2 is programmed to provide the ultrasonic carrier at the frequency of the transducer. In the authors's case, this was 40kHz. The output of the 555 is a square wave fed directly to the transducer.

The next step is to frequency modulate the ultrasonic carrier. This is very simply achieved by IC1 feeding the appropriate tone into pin 5 of IC2. The tone frequencies can again be selected from Table A and switched by means of a multiway push button unit.

The PCB layout is detailed in Fig. 7, and this board fits neatly inside a Norman Rose AB7 aluminium box, along with battery and clips. The transducer is glued on the outside of the case. No doubt a more pleasing arrangement could be evolved without much effort.

since aluminium boxes always betray a certain amateurish technique. A multipurpose plastics case will shortly be made available to accommodate this

requirement.

When wielding the soldering iron around the circuit there is not much to be said, apart from "put the 555's in the right way round." Experience (bitter), shows that the 555 is not one of the hardier IC's, and can be made defunct with very little effort. The author finds a desoldering tool of some kind is one of the most invaluable items in the constructor's toolkit.

The switch on the tone channel in the prototype was soldered directly to the board so when designing the board, try and accommodate the switch in a similar fashion, since it can save much time, and should mean that wiring mistakes will not occur. Remember that one pole of the switch on each tone channel is used to switch the supply.

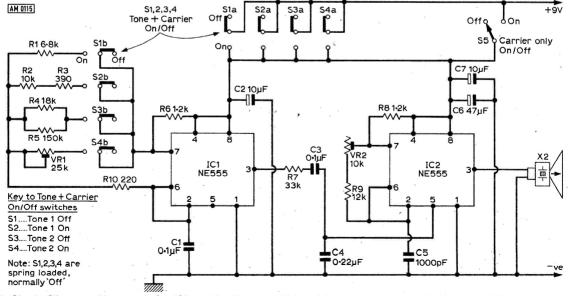
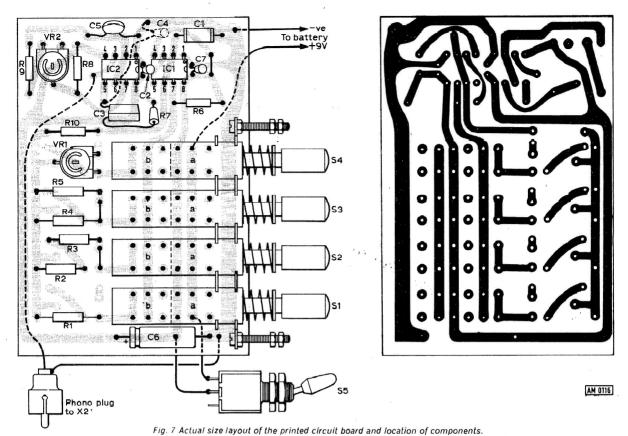
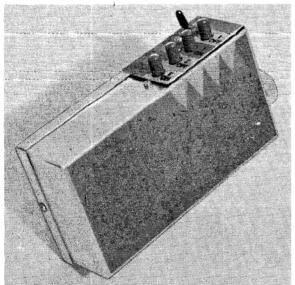


Fig. 6. Circuit of the transmitter or controller, IC1 providing the tones which modulate the carrier produced by IC2 Note: -R10 was inadvertently omitted from components list.

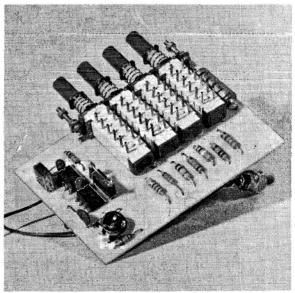




The finished controller, the switch providing "carrier only" control when required.



1. Operation of the carrier transmitter, IC2, is checked by supplying 9V to the system. Place the transmitter and receiver close together and monitor the output at pin 9 or 7 of the SN76660N. Now tune the transmitter preset on IC2 until the meter reading deflects, indicating

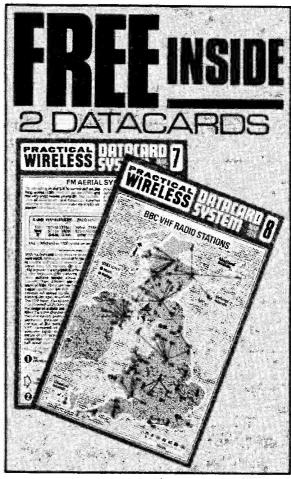


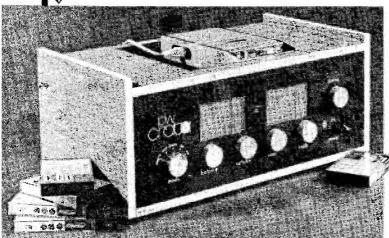
Completed board ready for mounting into case.

the presence of the correct ultrasonic carrier, usually around 40 kHz.

2. Align the ultrasonic tone decoder module as set out in the section on that unit. It may be useful to monitor the output by means of an LED, to indicate output status. The limiting resistor, Rx, should be chosen to suit the available LED, a typical value is 820Ω for a TIL209, Fig. 3.

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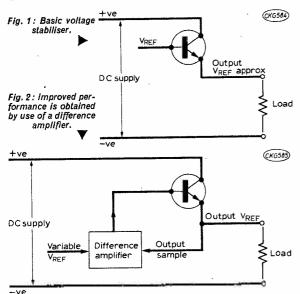
of one kind or another and so the experimenter needs a variety of DC levels. Power supply stability is unimportant for some circuit designs, but many amplifiers perform better if driven by a stabilised supply and, if it has a really low impedance output, decoupling may be unnecessary. In particular, DC amplifiers require stabilised supplies.

This stabilised supply aims at a professional performance, but uses low cost components readily available to the constructor. The output is fully floating or may be used in twin pack applications (centrepoint earthed) described at the end of this article.

## BASIC PRINCIPLES

The majority of stabilised power supplies use the emitter follower principle where the emitter takes up a potential slightly less than that applied to the base of a transistor, Fig. 1. A reference voltage from a battery or a zener diode applied to the base of this transistor will yield an output voltage quite well stabilised against load and supply variations. An NPN transistor is shown in Fig. 1, but a PNP will perform equally well in the negative supply line.

Unfortunately, such a simple circuit is seldom adequate, especially if one requires a variable output.



## DESIGNED AND DEVELOPED BY THE IPC TECH

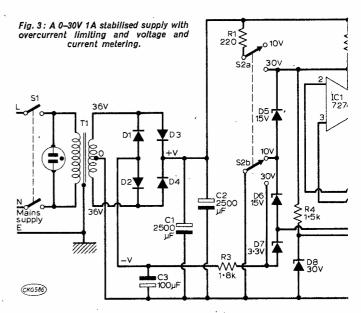
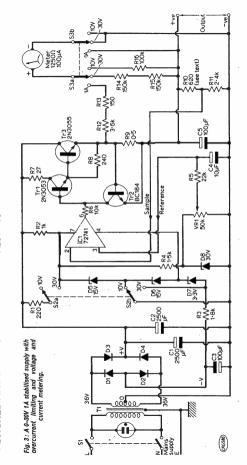


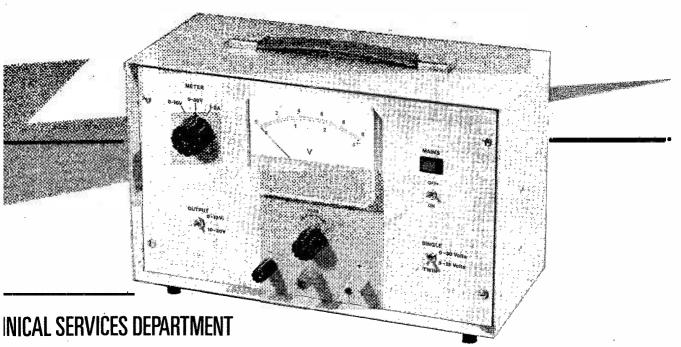
Fig. 2 shows how to modify the basic circuit by adding a difference amplifier to improve stability. A sample of the output voltage passes to the difference amplifier, which compares it with a reference voltage. The difference amplifier has a very high gain, so that if the output voltage increases or decreases compared with the reference, the transistor base voltage immediately moves in a direction which compensates for this change. In addition, because this action is effective up to ripple frequencies, the circuit also provides extra smoothing of the output.

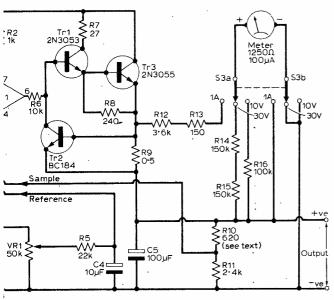
## PRACTICAL CIRCUIT

This power supply is based on the circuit of Fig. 2, but incorporates features to overcome practical limitations of the components used. Fig. 3 shows the full

# DESIGNED AND DEVELOPED BY THE 1PC TECHNICAL SERVICES DEPARTMENT







circuit in which a transformer giving 36-0-36V RMS supplies the basic voltage. The transformer voltage is not very critical, windings giving between 32 and 38V would be adequate. For convenience, a bridge rectifier operating as two full-wave rectifiers provides the positive and negative voltage rails, but four separate diodes may be used if preferred. The main positive supply uses two  $2.500\mu F$  smoothing capacitors, but a single  $5.000\mu F$  capacitor can be used if available. Negative bias for the reference chain is provided by D1 and D2 and smoothed by C3.

The series regulator (emitter follower) uses a 2N3055 transistor Tr3, connected as a Darlington pair with Tr1 a 2N3053 to reduce the current needed from the difference amplifier IC1. The resistors R10/R11 which sample the output voltage are connected across the output terminals together with C5. R10 may need

selecting to ensure that the supply will give its maximum 30-31V output. It is important that R10, R11 and C5 are mounted directly at the output terminals. The output voltage sample from R10/R11 passes to the inverting input of a 741 operational amplifier, which is used because of its low cost and ready availability.

On range 1, which gives 0-10V, the 741 takes its power from  $\pm$  15V supplies derived from the main voltage rails by R2, D5 and R3, D6. D7 has no function on this range. The reference voltage is set by potentiometer VR1 which is connected between the  $\pm$  15V rails. When switched to range 2, the whole IC1 supply rises positively by 18 3V (D6+D7) and provides a new reference voltage for operating over the 10-30V range. This arrangement is necessary because the 741 integrated circuit will swing between  $\pm$  10V satisfactorily, but not reliably outside these limits. However, tests show that the stabilised ranges are in fact rather wider than specified.

Transistor Tr2, whose emitter-base junction is connected across a  $0.5\Omega$  resistor R9 in series with the

## **★** Specification

## Voltage ranges

0-10V range 1 10-30V range 2 ±5-15V twin pack range 3

## **Current capability**

1A all ranges. Short circuit protection incorporated

## Voltage stability

0-3V zero to full load 0-015V per °C

1.5% for 10% mains change

## Ripple

1mV off load 3mV full load

## Pulse response

Full recovery in 25μS

output, provides current limiting. As the output current rises above 1A, the voltage developed across R9 switches on transistor Tr2. This shunts the control voltage applied to the base of Trl, reducing the output voltage to a low level. Transistor Tr2 is not critical and a spare lying in the junk box will probably serve for this. One can test the current limiting by supplying 5-10V at 1A and then increasing the voltage until the current stops increasing; if this occurs in the region of 1 to 1.4A the unit will be satisfactory. The current of the design model limits at 1.2A. It is, of course, necessary to use an external ammeter for this test.

## **METERING**

To make a really professional job of this supply. a metering circuit has been incorporated. However, as meters are expensive one can omit this circuit and instead fit a calibrated dial to the potentiometer VR1. This will be suitable for applications not requiring accurate voltage settings. To use a meter which differs from the one specified calculate the appropriate series multiplier resistors.

The following examples, using a  $100\mu A$  1,250 $\Omega$ meter, shows how to calculate the values of the resistors using the formula:-

$$R_{\text{TOTAL}} = R \text{ meter} + R \text{ series}$$

$$= \frac{\text{Voltage range required} \times 1000}{\text{Meter F.S.D. in milliamps}}$$

10 volt range (R16) 
$$R_{TOTAL} = \frac{10 \times 1000}{0 \cdot 1} = 100 k\Omega$$

Strictly one should subtract the meter resistance from this figure, but in practice the effect is insignificant and may be neglected.

30 volt range (R14+15)

$$R_{\text{TOTAL}} = \frac{30 \times 1000}{0.1} = 300 \text{k}\Omega$$

The most satisfactory way of producing this resistance is to use two  $150k\Omega$  resistors in series: again one may neglect the meter resistance.

Current range. Since R9 is directly in series with the output current (apart from a small load which R10 and R11 take), it will drop 0.5V at 1A. The meter has to read this voltage to measure the current, using a series resistance obtained from the calculation:-

$$R_{TOTAL} = \frac{0.5 \times 1000}{0.1} = 5,000\Omega$$

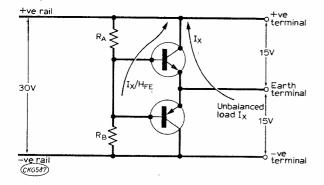
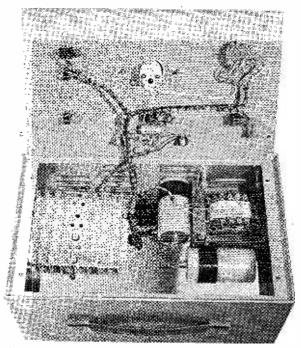


Fig. 4: Basic voltage divider using two emitter-followers.



View with front panel open to show wiring to controls with C5 and R10/11 mounted at the output terminals.

In this case the meter resistance is a significant part of the total and must be subtracted to give the value for R12 & 13 as  $3,750\Omega$ . One can make this up from a  $3.6k\Omega$  in series with a  $150\Omega$  resistor. Close tolerance resistors (2% or better) are essential for the meter circuits.

The current range determines which meter can be used. For example, if one used a 5mA meter the above calculation gives a total resistance of  $100\Omega$ , so the internal resistance of the meter would have to be not more than this.

## TWIN PACK SUPPLIES

Frequently one needs a positive and a negative supply rail to drive DC amplifiers, operational amplifiers and similar circuits. For these applications one may use two power supplies of the type described, because a floating supply may be connected with either its positive or negative rail earthed. Commercial twin packs usually duplicate the stabiliser circuits and have additional windings on the mains transformer. If a suitable transformer is available, there is no reason why you should not use the same system. However if one is content with lower voltages, then the floating supply can be converted into a twin pack for an additional cost of about £3.

If one connects two equal resistors across the floating output and earths their centre point, the negative rail will be below earth potential and the positive rail will be above earth by an equal amount.

In practice, two resistors are unsatisfactory because the two supply rails will not remain balanced with varying loads, but one can use instead two emitter followers, Fig. 4. In this arrangement the bases of the transistors are connected to a voltage half way across the 30V supply and the emitter follows this voltage. By connecting the emitters to earth one obtains a twin pack supply.

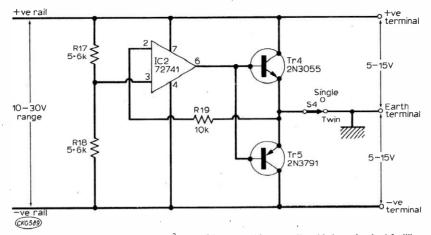


Fig. 5: Adding this circuit across the output of Fig. 3 provides an optional balanced output facility.

The simple circuit shown in Fig. 4 is also unsatisfactory for three reasons. First, short circuits on either of the outputs will 'blow' the opposite transistor. Second, the wattage ratings of the transistors limits the out-of-balance current, so that if one employs low wattage transistors, one must balance the loads carefully. Third, because changes in the out-of-balance current cause an excessive voltage drop in resistor RB, the rails will not be very stable.

Fortunately, one can cure all these faults quite simply by using high power transistors (mounted on a heat sink) driven by a feedback amplifier as shown in Fig. 5. In this arrangement IC2 constantly compares the output of the emitter followers with the reference input, and maintains the balance under all conditions. Because IC2 will not operate at very low levels the voltage range available from this twin pack is limited. However, the range provided (±5-15V) is satisfactory for most applications.

Some difficulty may be experienced in obtaining the 2N3791, used in the prototype. The Motorola MJ2955 (also in a TO3 package) is specified as a PNP complement of the 2N3055 and should be a satisfactory alternative.



View of rear of unit showing circuit board mounted on back panel.

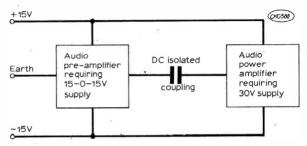


Fig. 6: Powering balanced and unbalanced equipment from a single supply.

In practice, because the very low output impedance makes each rail 'earthy' as far as AC signals are concerned, one can use the full 30V when the supply is wired in this fashion. For example, it is quite in order to connect a 30V amplifier, which is isolated from earth, across the terminals and to use a capacitor from a pre-amplifier connected to the ±15V supplies to drive it, Fig. 6.

## CONSTRUCTION

A standard instrument case measuring  $12^{1}_{4} \times 7^{1}_{2} \times 5^{1}_{2}$  in. provides a suitable unit in which to house the power supply. Fig. 7 shows the front panel marked out ready for drilling.

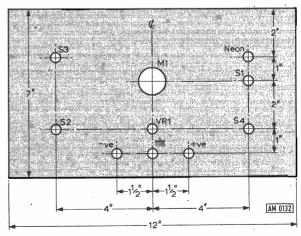


Fig. 7: Drilling dimensions for the front panel. Hole sizes should suit the components used.

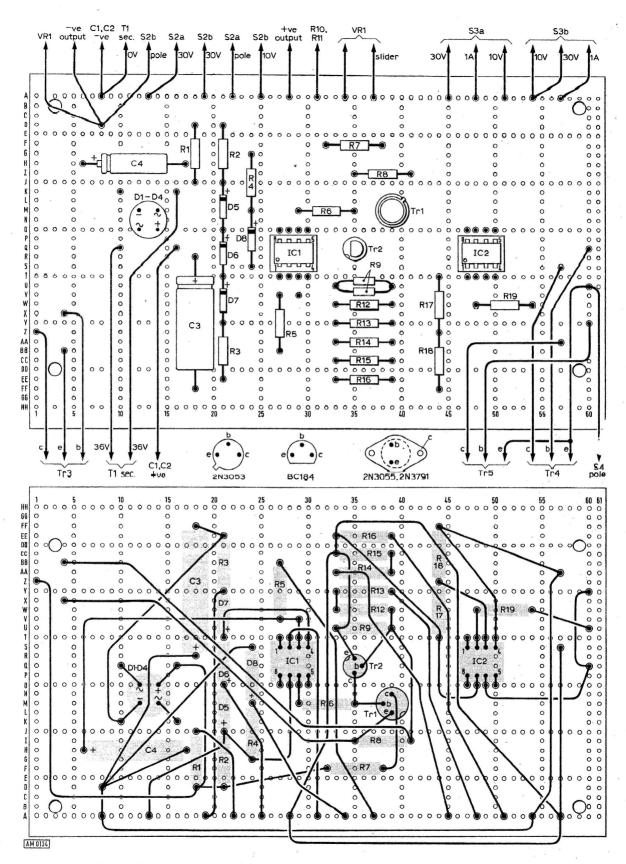


Fig. 8: Layout of component and wiring sides of circuit board. All components are mounted on terminal pins.

## \* components list

Resistors R1 220Ω ½W 5% R2 1kΩ 1W 5% R3 1 8kΩ 1W 5% R4 1 5kΩ ½W 5%	R9 0 5Ω 1W 5% R10 620Ω ‡W 5%* R11 2 4kΩ ‡W 5% R12 3 6kΩ ‡W 2%
R5 22kΩ ½W 5% R6 10kΩ ½W 5% R7 27Ω ½W 5% R8 240Ω ½W 5% VR1 50kΩ ½W Linear	R13 150Ω ½W 2% R14 150kΩ ½W 2% JR15 150kΩ ½W 2% R16 100kΩ ½W 2% * See text
Capacitors C1 2500μF 63V C2 2500μF 63V C3 100μF 63V	C4 10µF 63V C5 100µF 63V
Semiconductors Tri 2N3053 Tr2 BC184* Tr3 2N3055 IC1 SN72741 * See text	D1-D41A 100PlV bridge* D5 BZY88/C15 D6 BZY88/C15 D7 BZY88/C3V3 D8 BZY88/C30
3 way wafer, Meter 10 MR315). Fransformer powents "Univ. Rec. VAC). Plain Verobo	2 DPDT Toggle. S3 2 pole 10μA 1250Ω (RS Components in 36-0-36V 1-1A (RS Components in 36-0-36
Additional componen Resistors R17 5-6k() R18 5-6k	nts for Twin Pack output.
All ½W 5%  Semiconductors Tr4 2N3055 Tr5 2N3791 or MJ295	IC2 SN 72741

63p and 73p respectively, plus VAT, from Jermyn

Home Electronics, 112 Vestry Estate, Sevenoaks,

S4 SPST Toggle. TO3 heatsink (Chromasonic Electronics). TO3 Mica washers and insulating

Kent. Post and packing 25p per order.

bushes (2 sets). IC socket.

Miscellaneous

With the exception of the two  $2500\mu F$  capacitors, which are attached with insulating "P" clips to the transformer, all other components are assembled on a 6 x  $33_4$ in. 0.1 matrix plain veroboard. This is mounted on the back panel using 6BA spacers to give the required clearance.

The layout of the board Fig. 8 does not pose any problems. One can work logically through the circuit diagram starting with the bridge rectifier at one end of the board, and progressing through to the output of transistor Trl at the other end. Links between components on the circuit board and points elsewhere in the unit should be made via terminal pins on the edge of the board. Flexible wiring connects these pins and the external components.

## **ASSEMBLY**

First the transformer and heat sinks are mounted inside the case, Fig. 9. Depending upon its size the transformer may foul the output switch mounted on the front panel, in which case it must be moved back. The two heat sinks carrying the transistors Tr3, Tr4 and Tr5 are bolted together and attached to the side of the case using a simple bracket. Mounting these transistors so that their base and emitter connections are accessible when the front and back panels are removed makes wiring and testing easier.

Finally, the interconnecting leads have to be attached to the panel, transformer and power transistors. The transformer wiring is kept separate and wired directly to the mains switch, indicator neon and bridge rectifier connections. The three wires from the output terminals, namely positive, negative and sample, are critical and are loomed together to run directly to their respective circuit board connections.

The remaining interconnecting wiring is also loomed together for neatness, and in consequence wire identification is necessary. The number of wires involved makes colour coding impracticable, so a simple method of wire identification or masking tape marked with letters can be used instead. As a precautionary measure wire lengths should be kept to a minimum, but they should be long enough to allow the back and front panels to be laid flat.

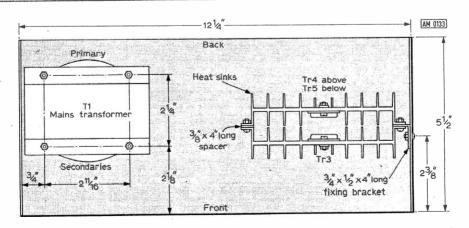


Fig. 9: Plan view showing mounting details of mains transformer and power transistor heatsinks.

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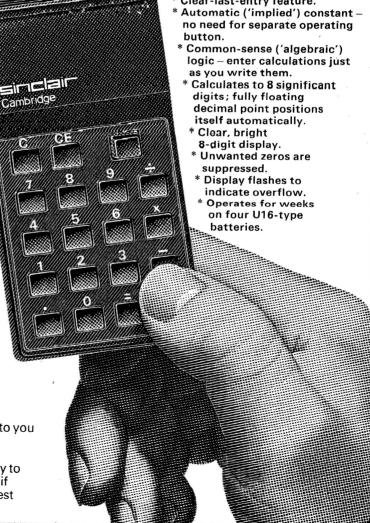
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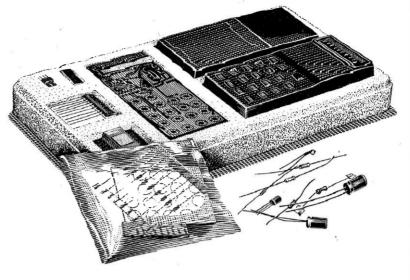


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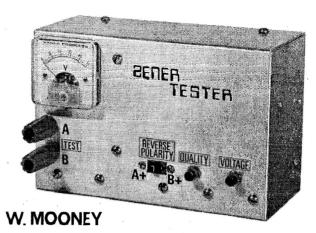
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## ZENER TESTER



THIS zener diode tester is an inexpensive pushbutton device which quickly gives the stabilising voltage of a zener diode and an indication of its quality in a matter of seconds. The tester may also be used to determine the polarity of the diode, which is sometimes difficult to establish from visual checks. The instrument may also be used as a "go"-"no-go" tester for normal diodes.

Because of its speed and ease of use, this piece of apparatus is therefore a must for the "Unmarked and untested" device enthusiast and a very useful tool for any electronics workshop.

Although mainly for bench use, battery operation is still desirable so the tester is therefore powered by an internal PP9 battery.

## THE CIRCUIT

Most zener diodes in use lie in the range 2V to 30V. Many such diodes will be outside the range of a single 9V battery so some means of high voltage generation must be available. It is possible to obtain the voltage required by the use of two or three 9V batteries in series but this is expensive in the long run and the replacement of batteries can become tiresome. A single transistor/inverter is therefore used to provide the required voltage from a 9V supply.

The inverter consists of a power transistor in an oscillator circuit using a ferrite ring on which is wound a high voltage secondary giving about 200VAC off load. Although the dissipation in the

transistor is only about <sup>1</sup><sub>2</sub>W a 2N3055 power transistor is used as these are readily available and will take a lot of abuse during setting up, without failure.

A ferrite ring is used in the transformer resulting in low external radiation and high efficiency. The AC voltage developed across the secondary winding is rectified using a silicon bridge. In the setup described the DC generated can reach around 250V if there is no diode connected to the test terminals.

As can be seen from Fig. 1 the oscillator is very simple in design. On pressing S1, the "Voltage" switch a current flows in the collector winding caused by the bias resulting from R1. The increasing flux in the core causes a feedback current in the base circuit in the correct phase to sustain oscillation. The frequency is a function of the value of R1 and C1 and the collector winding inductance, and greatly depends also on the load applied to the secondary winding. The frequency resulting is in the region of 50kHz and the current in the collector approximates to a squarewave.

The rectified output is applied to C3 via R2, with the zener to be tested connected to terminals A and B. The voltage on C3 builds up until the zener begins to conduct, at its reference voltage. The switch S3 is used to reverse the voltage applied to the zener should it be connected in the forward biased direction. A voltmeter connected across C3 will thus read the zener voltage. Depression of S2, the "Quality" switch, increases the current in the diode to about twice its original value. An increase in the voltmeter reading will result due to the dynamic resistance of

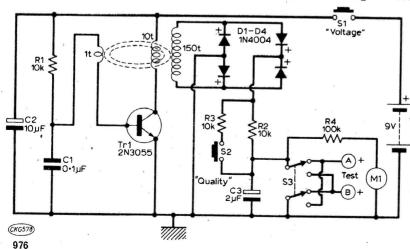


Fig. 1: Circuit of the diode tester. Rectifiers D1/D4 could be a single bridge unit. Layout of components is not critical. Panel layout can follow that used by author, as in the heading photograph.

## \* components list

Resistors R3 10kΩ ½W R1 10kΩ 1W R2 10kΩ ½W R4 100kΩ 5% ½W Capacitors C2 10µF 12V C1 0.1µF polyester C3 2µF 400V **Semiconductors** 2N3055 D1-D4 1N4004 Miscellaneous S1.S2Pu sh-to-make switch. S3 DPDT switch, Meter 500µA FSD. Terminals (2). Case, 6 x 4 x 2½in. approx. Battery PP9 (9V) and terminal clips. Tag strips. Ferrite rings (2) FX1593 (Hawnt Electronics Ltd. 112 Pritchett St., Birmingham B6 4EN, 25p a pair inc.). Insulating kit for 2N3055.

the diode but in a good zener diode the increase should hardly be visible on the meter. The voltmeter consists of M1 and Rm giving a FSD of 50V.

## COMPONENTS

The oscillator transistor is a 2N3055 and a specimen with a DC current gain of at least 30 should be chosen. Most devices will meet this requirement but the author has come across devices with gains as low as 2.

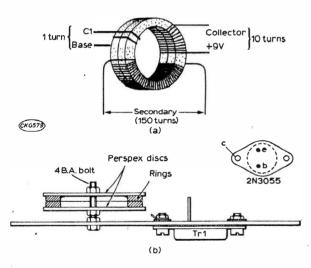


Fig. 2: Details of the construction of the oscillator transformer and its mounting on plate with the transistor. Two ferrite rings are used, taped together.

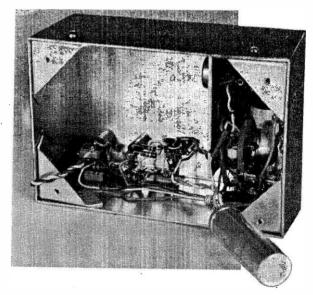
The ferrite rings have plenty of space for easy winding. There are many rings available which are of similar dimensions and although the magnetic properties may not be known, in a simple application like this, most rings will suffice. A little experimentation with the number of turns in the feedback winding should result in success as the setup is far from critical, Fig. 2. The windings on the ring are all pilewound. The feedback and collector windings are made from thin insulated connecting wire. The secondary consists of 150 turns of 30 SWG enamelled copper wire.

## CONSTRUCTION

As the layout is in no way critical most constructors will have their own ideas on positioning components. The only important criterion; for right-handed people the "voltage" and "quality" buttons should be on the right hand side of the cabinet and the terminals on the left, the meter being unobscured somewhere in between. In this type of instrument the front panel is best in the horizontal position as in the case of testmeters.

The transistor does not require a heatsink but may be mounted on an aluminium bracket for convenience using the usual insulating mica washer. The transformer may also be mounted on this bracket. A screened cabinet is desirable as severe radio interference can result in the low frequency end of the RF spectrum from harmonics, when the oscillator is running. The original unit was constructed using a ready-made chassis size 6in x 4in x 21 $_2$ in deep. The front panel layout is shown in photographs. The components were mounted on tagstrips which were bolted to the chassis at convenient points.

A base plate was cut from hin aluminium and four rubber feet attached, the plate held to the cabinet with four self-tapping screws. The battery was held in place with a wad of foam rubber. The PP9 is quite heavy, however, and would best be held in place with an aluminium bracket made for the purpose.



Rectifier diodes D1/D4 are mounted on a tag strip below the oscillator transformer. Remainder of small components are on a tag strip in lefthand compartment.

## **OPERATION**

A zener diode may be considered as a perfect zener in series with a resistor as in Fig. 3. The lower the value of Rd the better the stabilisation will be. A graph of current against voltage for the resistive part, the zener part and a combination of the two is shown in Fig. 4a, b and c. It can be seen that an increase in current from A to B results in an increase in reference voltage from a to b in Fig. 4c. If the

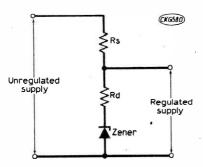
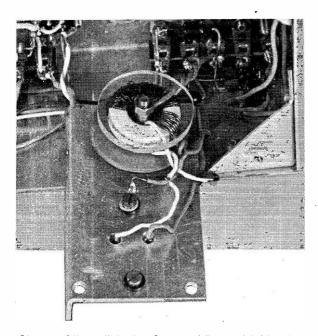


Fig. 3: Circuit of a perfect zener diode with a series resistance approximates the zener diode in practice.

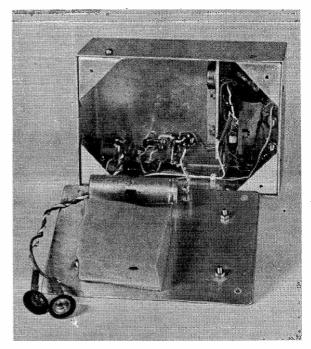
resistive component is very small the plot becomes as in Fig. 4d, a similar increase in current now causes only a small change in reference voltage. In a normal diode the situation is in Fig. 4d.

The portion of the graph marked k in Fig. 4b is called the knee and is usually less than 1mA. The instrument described supplies a current of over 1mA in all cases, therefore this effect should not normally be encountered.



Close-up of the oscillator transformer and its associated transistor removed from case. The one turn base winding is clearly seen.

Assuming that the diode polarity is not known, it is connected across the terminals A and B and the "voltage" button pressed. If the voltmeter reading is very low, say about 0.2V then this is probably the forward biased direction, in which case the polarity should be reversed by operation of S3, and the voltage button pressed again. The meter will (a) rise to the zener voltage, in which case the "quality" button should be pressed, or (b) remain low, in which case the diode is useless, or (c) the voltage will continue to rise and probably go off scale in which case it is an ordinary diode (or a zener with a higher voltage than 50V which is unlikely).



Capacitor C3 is held by a clip to the bottom of the box. A smaller sized capacitor could be fitted inside the box.

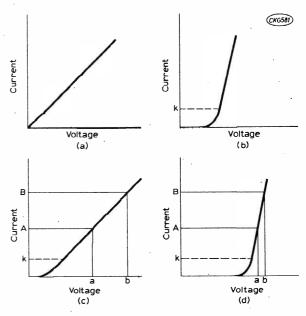
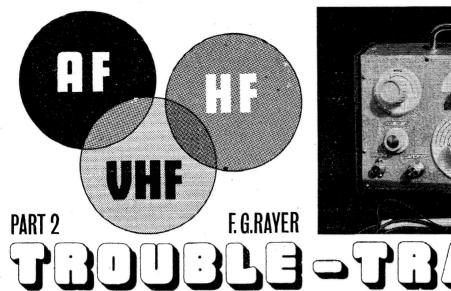


Fig. 4: Four graphs to illustrate the operation of a zener diode as described in the text.

Some diodes which are damaged read a high voltage in both directions, acting as a resistor, pressing the quality button causing a substantial increase in the meter reading.

Base emitter junctions of some silicon transistors have low reverse breakdown voltages, in the region of 5V, and act as excellent zeners. To test these the collector lead is left open circuit and the base and emitter leads connected to the instrument as in the case of a normal diode.



いこういいいし

RIEF details of the way in which the Trouble-Tracer can be used should prove of help. In some cases a faulty stage may be located with equal speed by using either the generator or the tracer facility. But in other cases one method will be better than the other, so the manner in which faults can be located by either means is outlined. The same general method of working will apply with circuits other than those which are given as examples.

The VHF and HF oscillators produce an unmodulated radio signal which produces no audio output when tuned in with a radio receiver. This unmodulated or CW signal will, however, operate a receiver tuning meter and is used for adjusting some circuits where modulation is not required (e.g., a crystal filter).

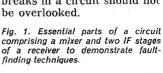
For almost all general tests and similar purposes, the tone generator must be switched on. The radio frequency signal, when tuned in with a receiver, then produces an audible tone, which is heard in the receiver speaker. The RF output is also modulated in this way when it is applied to IF stages and an audible indication is required from the radio receiver or equipment being checked.

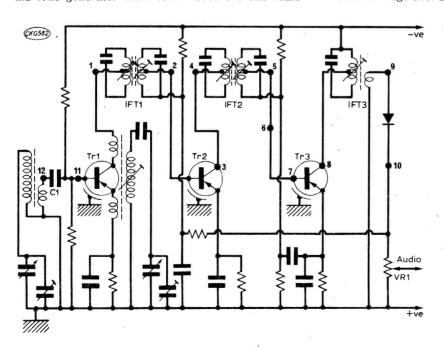
## RF/IF TRACING

Fig. 1 shows typical mixer and IF stages of a receiver. Assuming no results are obtained, a meter has shown that the correct voltage is present from the negative to the positive supply lines, but no audio signal is found at VR1. Therefore the fault must lie in the mixer or IF stages. With the RF probe applied at point 1, one or two local stations ought to be heard on tuning round. If not, the fault is sought in the mixer stage Tr1. Check by testing aerial and oscil-

lator coil windings for continuity and by testing resistors and capacitors as well as Tr1. (More detailed investigation of this stage is possible with the generator.)

If signals are heard at point 2 IFT1 is probably in order. If signals disappear at point 3, Tr2 is not working. So resistors etc. in this stage would be tested. If necessary, detailed check can be madè along any circuit. When transferring the prod from the primary to secondary, such as from 4 to 5 of IFT2, a reduction in volume would be usual, but loss of signals altogether would indicate that IFT2 is probably faulty. Possible localised breaks in a circuit should not be overlooked.





For example, if signals are found at the actual pin of IFT2, point 5, but not at the lead or foil conductor 6, then the soldered joint here needs investigating. Also, if the signal is present at 6, but not at the emitter of Tr3, there is a crack in the foil, or other circuit interruption.

When bringing in a new stage, such as moving from 2 to 3, or from 7 to 8, a considerable increase in volume is to be expected. A test at 9 shows if IFT3 is working. An audio signal should then be found at 10. If not, test D1 and associated circuits.

It will be seen that the method of working is logical and very simple. When the part of the circuit which is defective is brought into use, signals cease.

## USING THE GENERATOR

Assuming that the receiver audio section is working, the faulty stage in Fig. 1 may be localised by injecting RF from the generator. In this case, point by point working is carried out **backwards** through the circuit. If the receiver has a 455-470kHz IF, inject

done, stray capacitance from the prod and lead will upset alignment.

## TRACING IN AF CIRCUITS

Fig. 2 is a typical audio amplifier, or audio section of a receiver. With an ordinary prod or preferably a prod on a screened lead with earthing clip taken to point 1, signals should be heard on the tracer speaker. If not, the tuner, pick-up, or other source of audio signals needs investigating.

The signal should similarly be present at 2, and should be much amplified at 3. If not, check Tr1 and associated resistors etc. The source of a trouble, such as distortion, may sometimes be quickly found by this means. For example, if quality is normal at 2, but poor at 3, Tr1 is probably not operating correctly. Its base, emitter and collector resistors should be checked first, including the possibility of a break in the circuit such as at point 4.

Moving to point 5 tests C2 and its conductors and joints. Even more volume should be found with the

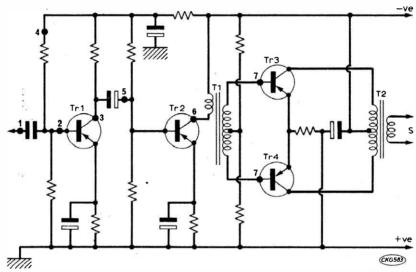


Fig. 2. Basic circuit of a three stage audio amplifier referred to in the text.

at point 8 with a prod from the RF output socket and tune the generator, on the correct range, until the modulated signal is heard. Then transfer the prod to points 7, 5, 4 and 3 systematically. In this way, a stage or IFT can be checked, as well as foil or other conductors. For example, if the generator signal is heard with the prod on IFT2 pin 4, but ceases when the prod is on point 3, the conductor from 4 to 3 needs examining.

Stage by stage checks at intermediate frequency will proceed to point 11, mixer base. With the IF applied here, all the IFT cores (five in Fig. 1) should peak correctly. If not, and results are poor, suspect the IFT which will not tune properly.

For RF tests, tune to the appropriate frequency. Band coverage can be checked with the generator, using very loose coupling from the output lead. Either take it to a loop of a few turns near the ferrite aerial, or place the lead near the aerial. When checking tuned circuits, a prod should not be applied directly to them when they are being trimmed. As an example, a prod can be taken to 4, to check IFT2, but if IFT2 is being adjusted, the prod should be taken to 2, IFT1, or an earlier point. If this is not

prod at point 6. If not, check T1 primary, and other items in this stage. Either point 7 should give equal volume. If not, test T1. If the signal is found here, but the set's speaker is not working, check the output stage Tr3/Tr4, T2 and associated items.

## GENERATOR TESTS AT AF

Such circuits can be checked also with the audio tone generator. Point by point tests are then made backwards through the equipment, as described for IF circuits. For example, if injecting a signal at 6 Fig. 2 produces an output in the equipment speaker, but not when the prod is moved to 5, then the stage containing Tr2 is faulty.

When a faulty stage has been located, a more detailed check is made to find the resistor, capacitor, or other defective item or to locate the exact fault. It is possible to use both the generator and tracer simultaneously, to check any stage, and this can be useful when there is more than one fault. As an example, injecting AF at 5 and testing for AF at 6 shows if Tr2 is working.



## 

## SHORT WAVES by DAVID GIBSON, G3JDG

APPY New Year, and I trust that there's a goodly crop of Santa-delivered receivers all eagerly swishing their antennae about with a view to sending in a luscious log. AR88 owners will, no doubt, have received a crane for ease of

Each year, most of the Amateur fraternity make huge numbers of New Year resolutions-how many did you keep from last year? More seriously, 1974 should be a "good" year for Amateur Radio and it really is well worth while making a few resolutions. How many receiving stations, for example, have an oscilloscope? They're not difficult to make, and for an s.w.l. it doesn't have to be a complex instrument. Commercial 'scopes are easily available and they offer an invaluable aid. One useful application is to couple a 'scope into the i.f. of the receiver. In this way, you can actually see the signal you are receiving. If you transmit, say, c.w., then by using the receiver/oscilloscope arrangement as a monitor you can observe the keying waveform.

Perhaps the best resolution for the s.w.l. is simply "Better reports for 1974". An s.w.l. QSL card can look very pretty, but remember that it's the information on it that is of interest to the transmitting Amateur. If your card simply has a rubber stamp approach-RST/date/time/mode/please QSL OM, then your chances of getting a QSL in return are minimal. Put as much information as you can into the report. Don't be afraid to write additional notes or even a letter if the report warrants it. Don't forget the conditions on the band at the time of the report. If you hear a station in, say, Brazil, then it will be of interest if you noted that other stations were fading or that Brazil was (or seemed to be) the only S. American area coming in but that many VE stations had been heard at an average strength

of RST 559 or whatever. Projects; we all build these-at least I hope we do. How about planning your projects this year. Plan it and then do it before you move on to the next thing. A grid dip oscillator (GDO) or a solid state version is invaluable and extremely simple

to build-and inexpensive. How about making one and playing antennas this summer? You can plan various types of aerials to experiment with. For a wire antenna, ordinary cheap bell wire will do for summer experiments. You can check the resonance of lengths of coax and numerous other things all with a simple g.d.o.

For the transmitting Amateur the resolution must be "Better use of the Bands". Four metres is a must. Don't forget RAEN (Radio Amateur Emergency Network). This organisation has various Nets and carries out exercises on 70MHz. This might make interesting listening for s.w.ls too. Information on this from the RSGB, 35 Doughty Street, London, WC1N 2AE. While you're at it, why not join the

Society?

In answer to impassioned pleas in the post from flat dwellers and unfortunate s.w.l. brothers in "No aerials allowed" environments some words of comfort. A length of flex round the picture rail can work wonders and you don't even need a back garden! In a room only 8ft. square, you can put up a twenty metre dipole (16ft. each half and fed in the middle with coax). Don't forget the loft. Quite ingenious antenna systems have been constructed up there by many Amateurs. A four metre beam can be used in some lofts. Even in the smallest, it's possible to wind some flex round the beams and feed the end with an aerial tuner unit (a.t.u.) on all bands from 160 metres to ten metres. So how about an evening's thought to some serious Amateur Radio New Year resolutions?

## Readers' Logs

Stanley Sharred (Birmingham), tells exciting tales of happenings of which the following is passed on; listen 2330hrs for KV4FZ on top band. Stanley's best on 160 metres; DK2QL, DK3BJ, DL0PG, HB9ANW, OE3SGA, PA0HIP all on s.s.b. while on c.w. (same band); OK2PEW, VO1KE, W3ZOW. K1NOL, PY6APM, SV1DO, UQ2GCT, K2ANR, OK1FCW. OH3NB/OHS. OK1AYY. KV4FZ, OK1MCW.

P. Barber (Co. Durham) gives a list of stations who are transmitting slow scan television on 14MHz. Frequency quoted is 14230kHz and callsigns heard include: F3RT, F6AZO, F6BIG, F6BKB, F9IB, G2BAR, G3LIV, I1PRQ, I3HDC, I8PSX, I8TMY, JA7FS, OD5HC, OZ4EDR, VE3FCN, W4LAS, W8BT. Connections to the receiver from the slow scan television monitor are made via the phone jack. How about a resolution to look into s.s.t.v. this year?

Glyn Fisher (Rutland), HRO tuning 4-6MHz m.o.s.f.e.t. converter, sends in a list of goodies heard high up on 144MHz. The antenna is a Vee dipole in the bedroom: DC1EU, DK1IE, DL3TR, F1BCD, F1CBH, F1CCP, F1CEC, GD2HDZ, GW8FTA/ P, ON4PB, ON5GF, PA0QC, PA0VV. Glyn wants to know of any modern receiver which covers 4-6MHz continuous. At present he can only think of an HRO or Canadian 52 set-anyone any ideas?

## BROADCAST BANDS

Short Wave Reports by 15th of the month to Malcolm Connah, 59 Windrush, Highworth, Swindon, Wiltshire, SN6 7DT.

Medium Waves Logs to Charles Molloy, 132 Segars Lane, Southport, PR83JG.

VHF/FM Reports to Simon David, c/o Practical Wireless, Fleetway House, Farringdon Street, London, EC4A 4AD.

## AMATEUR BANDS Short Wave/VHF

Logs in alphabetical order please by 15th of the month to David Gibson, G3JDG, 12 Cross Way, Harpenden, Hertfordshire.





## VHF/FM DXING

## by SIMON DAVID

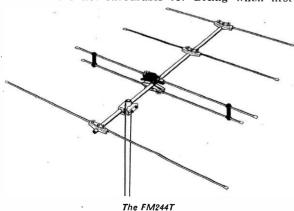
ROOURAGING news from the BBC is that more than 99 per cent of the U.K. population can receive v.h.f. broadcasts. Of these, two-thirds are within the service areas of transmitters carrying Radio 2, Radio 3 and Radio 4 stereo programmes. This should provide considerable encouragement for those not yet converted to f.m.

Tuner sensitivities these days are such that you could get by temporarily with the proverbial "wetstring" aerial if you are within good striking distance of the transmitter. One of the P.W. authors, Roland Perry, has done this in his second floor apartment room at Cambridge. Roland writes: "The aerial is three feet of wire draped along my desk and at one time or another I have heard BBC Radios Solent, Oxford, London, Nottingham, Humberside, Sheffield, as well as Capital Radio and LBC". Radio 4 transmissions from as far away as Rowridge have also been heard. His cascode f.e.t. preamp is obviously largely responsible.

In Lisburn, Co. Antrim, Mr. R. Montgomery proudly received Radio 3 in stereo for the first time on 91 3MHz. Interesting point here is that he uses a vertically polarised J-Beam 'H' aerial. He also receives stereo from Southern Ireland, Radio Telefis

Eirean and Radio Na Gaeltachta.

Talking of "wet-string" aerials I have tried something of this sort with a clip attached to my wall shelving rack. Results some 40 miles North-West of Wrotham were quite good and even produced some surprises from commercial radio. I have been trying out the new FM264T 6-element array from Antiference. This is a combination of the well known "Mushkiller" array with a "Trumatch" dipole. Conditions were not favourable for DXing when first



installed, but the mild spell in early December improved matters considerably.

It is easily installed and when lined up in the direction of South-East I have first-class reception of all the Wrotham and Croydon transmissions plus Radio Medway. Also pulled in were Lille 88 8MHz and 98 1MHz, Rheims 98 7MHz, although the latter was not a very strong stable signal.

Reducing this array to 4-elements had a marked effect on signal strength and the weaker stations tended to be almost impossible to get. On a more sensitive receiver, no doubt a 4-element array would be worthwhile. A photo of the 4-element FM244T array is shown here. The important point about the FM264T and FM244T is the uniform broad-band impedance characteristic, with a claimed variance of ±0.5dB over 88 to 100MHz.

Incidentally the FM264T was fitted in the loft about four feet above the joists. Of course, I have a large loft which is essential, but better results are obtainable when mounted outside on the chimney stack. Incidentally, I found it very useful to run two or three cables up to the loft; I can monitor the receiver output while positioning the aerial. The third (4-core) cable is for a rotator. I tend to prefer loft installations for convenience and certainly to avoid the unsightly clutter of outside installations on top of the roof.

The Editor has told me that *P.W.* is giving a pair of Datacards in the March issue, specially devoted to f.m. reception. I shall certainly make sure I don't miss them.

## MEDIUM WAVE BROADCASTS by CHARLES MOLLOY

BRIAN MURRAY (Edinburgh) has been trying the medium waves with his Astrad VEF204 10 transistor receiver using its internal ferrite rod aerial. He reports hearing programmes in English from the American Forces Network, Frankfurt on 872kHz; Radio Tirana, Albania 1394kHz at 2200hrs; Radio Portugal 755kHz at 2310hrs; Trans World Radio, Montecarlo 1466kHz at 2325hrs. Brian has also logged the BBC local radio station at London on 1457kHz at 1740hrs.

Stephen Mason (Loughton, Essex) has an Ultra 8 transistor receiver which he uses with a 7 metre horizontal wire aerial located 10 metres above ground level. His log includes programmes in English from Radio Sweden on 1178kHz; Radio Portugal 755kHz; Milan, Italy 899kHz; Voice of America, Munich 1196kHz; Radio Berlin International 1511kHz; Radio Tirana, Albania 1394kHz; and in Spanish from Radio Espana de Madrid EAJ2 which is a commercial station on 917kHz. Stephen has received verifications (QSLs) from Sweden, Portugal, Warsaw, Milan, Prague and the Voice of America, for reception on the medium waves.

lan Gordon (Birmingham) has been busy again with his Codar CR70A, aerial tuning unit and 25m longwire antenna. He reports hearing the Radio Peking relay in Albania on 1457kHz sign-on at 1730hrs with interference from BBC Radio Birmingham on the same frequency and Sud Radio, Andorra in French on 818kHz. Ian reports that the IBA (London) has been heard in Birmingham on 719kHz.

Several readers have asked if the writer would include a log of his own MW DX giving details of the

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DM70	0.68	EF80	0.46	PC97	0.45	PL508	1.05	30C17	1.00
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DY86/7	0.42	EF85	0.48	PCC84	0.52	PL802	0.96	PCF805	0.90
DY802	0.45	EF86	0.90	PCC88	0.80	PY33	0.66	30F5/PF8	
EABC80	1.00	EF89	0.81	PCC89	0.65	PY81/800	0.50		1.10
EB91	0.88	EF91	1.30	PCC189	0.65	PY82	0.55	30FL1/	
EBC81	0.75	EF92	1.40	PCF80	0.51	PY88	0.60	PCE800	0.75
EBF80	0.60	EF95	1.85	PCF82	1.80	PY500A	1.05	30FL2	0.75
EBF83	0.62	EF183	0.60	PCF86	0·65	PY800	0.50	30FL12	1.05
EBF89	0.58	EF184	0.60	PCF200	0.85	PY801	0.50	30FL14	0.85
EC86	0.75	EH90	0.56	PCF201	0.87	U26	1.00	30L1/PC0	284
EC88	0.77	EL34	0.95	PCF801	0.65	U191	1.00		0.52
ECC81	0.45	EL36	1.05	PCF802	0.71	U193	0.50	30L15/PC	VC805
ECC82	0.48	EL81	0.80	PCF806	0.66	UABC80	0.90	00220,2	1.05
ECC83	0.45	EL84	0.47	PCH200	0.88	UBC81	0.70	30L17	0.90
ECC84	0.55	EL85	0.55	PCL82	0.54	UBF89	0.60		
ECC85	0.58	EL86	0.95	PCL83	0.66	UCC85	0.80	30P4MR	1.30
ECC88	0.75	EL95	0.70	PCL84	0.59	UCH81	1.00	30P12/PC	
ECC189	0.71	EL91	1.31	PCL85	0.63	UCL82	0.70		1.05
ECF80	0.56	ELL80	1.80	PCL86	0.68	UCL83	0.70	30P19/PC	802
ECF82	0.78	EM84	1 18	PCL805/8	5	UF89	0.80		1.00
ECF86	0.71	EM87	1.18		0.68	UI.84	0.95	30PL1/PC	T.801
ECH81	1.00	EY51	0.83	PD500	1.55	UY85	0.50		0.95
ECH83	1.00	EY86/87	0.42	PFL200	0.80	6/30L2/		30PL13/	
ECH84	0.78	EY88	0.64	PL36	0.88	ECC804	1.00	PCL800	1.20
ECL80	0.53	EZ80	0.51	PL81	0.75	6F23/EF8			1 20
ECL82	0.61	EZ81	0.40	PL81A	0.88		1.05	30PL14/	1 05
ECL83	0.68	GY501	0.90	PL82	0.50	30C1/PCF	80 .	PCL88	1.25
ECL86	0.63	GZ34	0.78	PL83	0.96	Į.	0.51	30PL15	1.05
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						UCC85	0.45	68N7GT	0.48
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						UCH81	0.40	6U5G	0.75
JAE	10	r lists	۶.			UCL82	0.35	6V6G	0.40
						UCL83	0.70	6V6GT	0.45
AZ1	0.60	EF80	0.25	PC86	0.60	UF41	0.75	6X4	0.40
AZ31	0.55	EF85	0.35	PC88	0.60	UF89	0.40	6X5G	0.40
CBL31	1.20	EF86	0.30	PC97	0.50	UL41	0.85	6X5GT	0.45
CL33	1.50	EF89	0.28	PC900	0.48	UL84	0.48	7B6	0.75
CY31	0.50	EF91	0.87	PCC84	0.40	UY41	0.48	7B7	0.70
DAF91	0.30	EF92	0.50	PCC88	0.55	17Y85	0.40	7C5	1.30
DAF96	0.50	EF95	0.40	PCC89	0.50	VR105-30		7C6	0.75
DCC90 DF91	1.35 0.30	EF98	0·75 0·30	PCC189 PCF80	0-80 0-80	VR150-30		7H7	0.70
DF96	0.50	EF183 EF184	0.85	PCF82	0.35	OA2	0.40	7R7	0.75
DK91	0.45	EL32	0.60	PCF86	0.60	OB2	0.40	787	2.25
DK00	0.70	EL32	1.75	PCF801	0.50	1R5 185	0·45 0·30	7¥4 12AT6	0·75 0·40
DK92 DK96	0.60	EL34	1.75 0.50	PCF802	0.50	1T4	0.30	12AT7	0.40
DL92	0.40	EL36	0.50	PCF805	0.90	384	0.40	12AU6	0.45
DL94	0.48	EL37	2.50	PCF806	0.75	3V4	0.70	12AU7	0.88
DL96	0.55	EL41	0.90	PCF808	0.90	5R4GY	0.80	12AX7	0.33
DY86/7	0.36	EL42	0.90	PCL82	0.35	5U4G	0.40	12BA6	0.45
DY802	0.87	EL84	0.28	PCL83	0.66	5V4G	0.50	12BE6	0.50
EABC80	0.38	EL95	0.40	PCL84	0.45	5Y3GT	0.45	30C1	0.80
EAF42	0.75	ELL80	1.00	PCL85	0.50	5 <b>Z4</b> G	0.45	30C15	1.05
EB91	0.22	EM80	0.45	PCL86	0.45	6/30L2	0.90	30C17	1.10
EBC33 EBC41	1.00 0.75	EM81	0.60	PCL805/8	0.50	6AK5	0.40	30C18	0.90
EBC41	0.88	EM84 EM85	0·35 1·00	PD500	1.30	6AM5	0.90	30F5	1.10
EBF80	0.40	EY51	0.40	PEN45	0.75	6AQ5 6AS7G	0·45 0·85	30FL1 30FL2	0·80 0·75
EBF83	0.40	EY86	0.40	PL36	0.55	6AT6	0.45	30FL14	0.90
EBF89	0.82	EZ40	0.75	PL81	0.50	6AU6	0.80	30L15	1.05
EBL31	1.50	EZ41	0.75	PL82	0.45	6BA6	0.28		
ECC81	0.40	EZ80	0.28	PL83	0.45	6BE6	0.35	30L17	0.95
ECC82	0.33	EZ81	0.29	PL84	0.40	6BH6	0.75	30P4MR	1.30
ECC83	0.33	GY501	0.80	PL500	0.75	6BJ6	0.55	30P12	1.05
ECC84	0.80	GZ30	0.45	PL504	0.75	6BQ7A	0.55	30P19	1.00
ECC85	0.40	GZ32	0.50	PL508	0.90	6BR7	1.00	30PL1	0.95
ECC88	0.40		. 0.65	PL509	1.55	6BS7	1.35	30PL13	1.20
ECH35	1.25	GZ37	1.25	PL802	0.95	6BW6	0.90	30PL14	1.25
ECH42	1.00	HN309	1.50	PX4	3.50	6BW7	0.90	35W4	0.50
ECH81	0,80	KT61	1.75	PX25	3.50	6C4	0.85		
ECH83	0.45	KT66	2.50	PY33	0.68	6CD6G	1.30	35Z4GT	0.70
ECL80	0·55 0·85	KT81 (70		PY81	0·30 0·35	6CH6	1.40	50CD6G	1.20
ECL82	0.70	KT81	1·30 1·75	PY82 PY83	0.38	6CW4 6F23	1.00 1.05	807	0.50
ECL83 ECL86	0.40	KT88	2.90	PY88	0.40	6F25	1.00	813 ITT	£13
ECLL800		KTW61	1.00	PY500	1.00	6F28	0.70	813 USSI	
EF37A	1.20	MU14	1.00	PY81/800		6J5M	0.65		£5.75
EF39	1.20	N78	9.75	PY801		6J5G		866A	1.00
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AA119 AAZ13	0.7	BD132 BF115	0.22	Quan	tity	dis	cou	nts	on
AAZ15	0.10	BF167	0.25	annli	rati	on.	Sen	d S	SAE
AC107 AC126	0·35 0·25	BF173 BF179	0·28 0·33	for fu	.11	into	J C 11	<b>.</b>	// LM
AC127	0.25	BF180	0.35	ior it	# III   1	1515.			
AC128 AC176	0·20 0·25	BF181 BF194	0·35 0·15	,					
AC187	0.20	BF195	0.13	OC16	1.00	ZTX500	0.15	2N-2646	0.50
AC188	0.20	BF197 BF200	0·15 0·32	OC20 OC23	1·25 1·25	ZTX501 ZTX503	0·15 0·16	2N2904 2N2904	0·20 A 0·25
ACY21 ACY39	0·22 0·65	BFS61	0.25	OC25	0.40	ZTX531 ZTX550	0·25 0·18	2N2905	0.26
AD140	0.50	BFS98	0.25	OC28 OC35	0.70	ZTX550 1N914	0.18	2N2905	A 0.28
AD149 AD161	0·50 0·39	BFX29 BFX88	0·28 0·22	OC36	0·55 0·65	IN914 IN4001	0·08	2N2906 2N2926	
AD162	0.39	BFY50	0.20	OC42	0.40	IN4002	0.7	2N3053	0.20
AF115 AF116	0·25 0·25	BFY51 BFY52	0·20 0·20	OC44 OC45	0·18 0·18	IN4003 IN4004	0.8	2N 3055 2N 3525	
AF117	0.20	BFW10	0.61	OC71	0.15	IN4005	0.10	2N3614	
AF186	0.40	BFW10 BY100	0.15	OC72	0.25	IN4006	0.12	2N3615	0.65
AF239 ASY27	0.40	BY126 BY127	0.14	OC76 OC77	0·40 0·55	IN4007 1N4009	0·12 0·06	2N3702 2N3703	0·11 0·12
ASY28	0.25	BZX61 se	ries	OC77 OC81	0.28	1N4148	0.06	2N3704	0.14
BA102 BA115	8-25 0-10	BZY88 se	0.20	OC81D	0.28 0.45	18921 182033	0.07 0.20	2N3705 2N3706	
BC107	0.12	BELOG SC	0.10	OC81Z OC83	0.25	182051A	0.10	2N3707	0.13
BC108	0.12	CR81-05	0.30	OC140 OC170	0·65 0·25	IS2100A IS3010	0.20 0.25	2N3708 2N3709	
BC109 BC113	0·12 0·16	CR81-40 CR83-05	0·45 0·40	OC171	0.30	2N696	0.15	2N3710	
BC117	0.21	CR83-40	0.55	OC200	0.55	2N697	0.15	2N3711	0.11
BC143 BC147	0·30 0·12	MJE340 MJE370	0.50 0.68	OC201 OC202	0.80 0.90	2N706 2N706A	0·10 0·12	2N3819 2N3820	
BC148	0.10	MJE520	0.65	OC203	0.55	2N1131	0.25	2N3823	0.50
BC169C	0.14	MJE2955 MJE3055	1·10 0·75	OCP71 ORP12	1.00 0.55	2N1132 2N1302	0.25 0.18	2N3903 2N3904	0.15
BC182 BC182L	0·12 0·12	MPF102	0.40	ORP60	0.45	2N1303	0.18	2N3904	0.25
BC184L	0.13	MPF103	0.36	T1005	0·20 0·29	2N1304	0.22	2N3908	
BCY32 BCY33	1·20 0·38	MPF104 MPF105	0·35 0·46	TIC44 TIC226D	1.00	2N1305 2N1306	0.28	2N4058 2N4059	
BCY34	0.45	NKT404	0.60	TIL209	0.25	2N1307	0.28	2N4060	0.13
BCY70 BCY71	0·15 9·20	OA5 OA10	9-40	TIS43 ZTX 107	0·26 0·12	2N1308 2N1309	0.25	2N4061 2N4062	
BCY72	0.15	OA79	0.10	ZTX108	0.10	2N1613	0.20	3N141	0.81
BCZ11	0.65	OA81	0.10	ZTX300 ZTX301	0.14	2N1614	0.45	40360	0.40
BD121 BD124	1·00 0·80	OA91 OA200	0.08	ZTX301 ZTX302	0·14 0·18	2N2147 2N2160	0.75 0.61	40361 40362	0·45 0·50
BD131		OA202	0.10	ZTX304	0.24	2N2369A	0.16	40430	0.85
BN7400	0.20	SN7425	0.37	SN7473	0.44	SN74107	0.51	SN7415	7 1.09
SN7401	0·20 0·20	BN7427	0·37 0·43	8N7474 8N7475	0·48 0·59	8N74110 8N74111	0·57 0·86	8N7417 8N7417	0 2·88 4 1·80
SN7402 SN7403	0.20	BN7428 BN7430	0.20	SN7476	0.45	SN74111	1.00	SN7417	
SN7404	0.20	BN7432	0.37	SN7480	0.80	8N74119	1.92	BN7417	6 1.44
SN7405 SN7406	0.20 0.40	8N7433 8N7437	0·43 0·45	SN7482 BN7483	0·87 1·20	8N74121 8N74122	0·57 0·80	8N7419	
BN7407	0.40	BN7438	0.43	SN7484	1.00	SN74123	1.44	SN7418	2 2.30
SN7408	0·25 0·33	8N7440 8N7441A	0.20	BN7486 BN7490	0·50 0·75	8N74141 8N74145	1·00 1·44	BN7419 BN7419	
BN7409 BN7410	0.20	DIT (441A)	0.85	SN7491A	N	BN74150	2.30	8N7419	
SN7411	0.28	SN7442	0.85		1.10	SN74151	1.15	BN7419	6 1.58
8N7412 8N7413	0·28 0·30	8N7450 8N7451	0·20 0·20	SN7492 SN7493	0·75 0·75	SN74154 SN74155	2·30 1·15	BN7419 BN7419	
<b>BN7416</b>	0.30	BN7453	0.20	SN7494	0.85	SN74156	1.15	SN7419	9 2.88
BN7417 BN7420	0.80 0.80	SN7454 SN7460	0·20 0·20	BN7495 SN7496	0·85 1·00	DIL		14 5	n 15p
SN7422		SN7470 SN7472	0.33	SN7496 SN7497	4.32	SOCKE	re	14 pi	113p
SN7423	0-40	8N7472	0.38	SN74100	2.16	JUCKE		10 01	11/P

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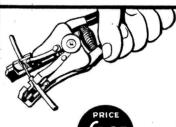
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equipment used. The main receiver is a Marconi Mercury marine communications receiver which covers 15kHz to 4MHz while the standby is a BC946 Medium Wave Command receiver. Two aerials are in use-a standard MW loop antenna and a 90ft longwire at 20ft above the ground. The OTH is Southport in Lancashire. Stations heard regularly in the evening are Radio Na Gaeltachta Conamara, Eire on 539kHz; Baghdad Iraq on 760kHz; Ouazvin, Iran 841kHz; Kermanshah, Iran 985kHz; Kuwait 1345kHz, The BBC Eastern Mediterranean relay in Cyprus is now on 1322kHz and is heard relaying the BBC World Service in English at 2305hrs. North American reception has been good this winter with CJON St. John's, Newfoundland, on 930kHz and WNEW New York City 1130kHz conspicuous at 2330hrs. Others heard before midnight are CBN St. John's on 640kHz: CHER Sydney, Nova Scotia 950kHz; WINS New York City on 1010kHz; CBA Moncton New Brunswick on 1070kHz; Radio St. Pierre, St. Pierre et Miquelon

(near Newfoundland) in French on 1375kHz. Two South Americans—ZYD66 Rio de Janeiro 940kHz in Portuguese and YVRS Radio Margarita, Venezuala on 1020kHz have been logged frequently along with PJA6 Radio Victoria which is located on the Dutch island of Aruba, near Venezuela and broadcasts religious programmes in English on 925kHz at 2315hrs. On the Longwaves Ankara, Turkey is now heard on 182kHz during the evening along with Azilal, Morocco in Arabic on 209kHz and Tipaza, Algeria on 251kHz in French.

Brian Richardson (Nottingham) asks about Whites Radio Log which gives details of all North American MW stations. This log used to appear in three parts in consecutive issues of an American magazine but it is no longer being published. Currently in the UK there is the Guide to Broadcasting Stations by Butterworth which is available in many bookshops. It lists all MW stations in Europe and a number of the more powerful ones in other parts of the world.

## SHORT WAVE DX by MALCOLM CONNAH

ow that the festivities are over we can all get down to some serious listening, and possibly constructing as well.

This time of year is ideal for trying to catch some of the stations which have eluded you in the past. The only way to catch them is to concentrate all available efforts to the task. One suggestion would be to make a list of all those stations which should be audible on your equipment and then systematically listen for them.

Radio New Zealand, for instance, should soon be audible again. Last year the frequencies were 9540 and 11780 and the time was 0800 GMT.

I wish you all the best with your listening and hope to receive your logs in the near future.

## Readers' Logs

A very large number of logs have been received this month and we start with a couple from our overseas readers.

**Bruce A. Laird** of Greensborough in Victoria, Australia has heard the following stations:

9625 R. Canada International in English at 0900.

9675 Radio Japan in English at 1100.

9715 R. Nederland in English at 0800.

9745 HCJB, Quito, Ecuador, English at 0800.

11775 Swiss B.C., in English from 0700 to 0930.

11875 Radio Japan, English from 0930 to 1030.

11890 FEBC, Philippines in English at 0930.

11940 R. Japan in Chinese and Vietnamese at 1000. 15235 R. Japan in English at 0930.

Steven Phillips of Durban in South Africa used his Hamerstein Hi-Fi Stereo 30 receiver and 30 odd feet of aerial wire to hear:

6160 R. Australia in English at 1500.

11730 R. Nederland via Madagascar at 1400.

11770 BBC, Ascension Is. relay at 1700.

11815 NHK, Japan in English at 1300.

15420 BBC, East Med. relay at 0400.

15840 R. Nederland in English at 1230.

21600 Deutsche Welle noted at 1200.

We return to the U.K. for a log from **Harold Emblem** of Mirfield in Yorkshire who has a Lafayette HA63 receiver and 18 metre long-wire antenna. This

combination was used to hear:

9545 R. Accra, Ghana from 2104.

11905 Austrian Radio at 1900.

15185 WINB, Red Lion, U.S.A. at 2115. 15410 United Nations Radio noted at 2200.

15440 WNYN, New York at 2100.

17855 United Nations Radio, Tangiers, 1830.

25790 RSA, South Africa noted at 1420.

Simon Auger of Cheshunt in Hertfordshire writes—"After mourning for the injury to my transistor superhet, I blew the dust off my homebrew 2 valve regen.; attached it to my 50 foot end-fed aerial and an A.T.U. and accomplished the following in fourteen afternoons":

4965 RSA, South Africa in English at 1330.

6070 Radio Sofia, Bulgaria at 1930.

9005 Voice of Iran, Tehran, English at 2000.

9630 Radio Sweden noted at 1230.

9760 R. Nacional d'Espana at 1900.

9912 AIR, Delhi in English at 2000.

11775 Radio Bucharest, Rumania at 1500.

15340 Radio Cairo noted at 1800.

K. Goldsworthy of Hounslow in Middlesex has a Russion-made VEF204 receiver which, when connected to a 10 metre antenna, produced the following:

6025 R. Portugal in English at 2130.

7185 RSA, South Africa in English at 2000.

15185 Radio Finland in English at 1200. 17750 Havana, Cuba in English at 2130.

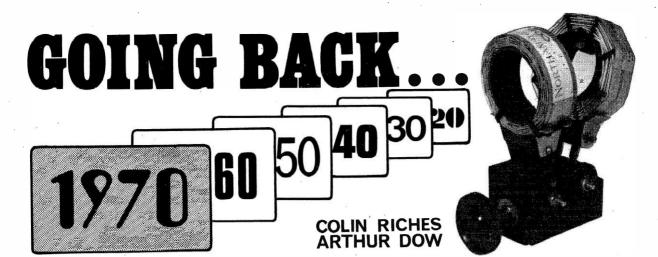
17820 R. Canada International, English at 2230.

17825 RSA, South Africa, English at 1600.

Another point which is worthy of mention at this time of year is the fact that the new edition of the 'World Radio TV Handbook' is due for publication. This book contains details of all Broadcast stations throughout the world. The entry for each station shows a list of all frequencies used; a current programme schedule; details of the Interval Signal; the station announcement; the address of the station and its QSL policy.

The introduction includes general hints on shortwave listening and the last section of the book is a list, by frequency, of all the stations. This list is very useful in identifying particular stations, especially when the frequency to which the receiver is tuned is accurately known.

All in all the handbook is an invaluable asset to all serious DXer's and I would recommend it to anyone regarless of their level of experience.



## S-T's Neutrodyne

N our August 1972 Going Back we gave an alltoo-brief account of some of the contributions made to the art of wireless in the early days by John Scott-Taggart. We mentioned, among other matters, that he had sold many patents to the big wireless manufacturing companies including one to the Hazeltine Corporation (USA). Our attention has been drawn to an article published in the July 31st, 1926 issue of "Wireless" wherein much is made of the fact that the Neutrodyne (neutralising) circuit was patented by S.T. on January 2nd 1923 while the date for a practically identical patent by Professor Hazeltine in the USA was April 5th, 1923, some three months later.

"Wireless" sported headlines such as "Who invented the Neutrodyne?", "We should in England call it the Scott-Taggart Neutrodyne" and "New facts about a great invention". We only wish that we could be allowed to reproduce the three pages that "Wireless" published on this most interesting story if only to ensure that present day readers of PW are made aware that the British habit of throwing away potential moneymaking inventions is no recent acquisition! Remember the Hovertrain, among others?

To further quote Wireless "The utmost interest is being shown in the whole question of the Neutrodyne circuit. Although such circuits have from time to time been incorporated in various receivers, the extraordinary success of the "Elstree Six" has persuaded the wireless public that for selectivity, range, signal strength and non-radiation the Neutrodyne stands supreme . . . Although in America the manufacturers and public immediately appre-

## WHO INVENTED THE NEUTRODYNE?

"We should in England call it the Scott-Taggart Neutrodyne"

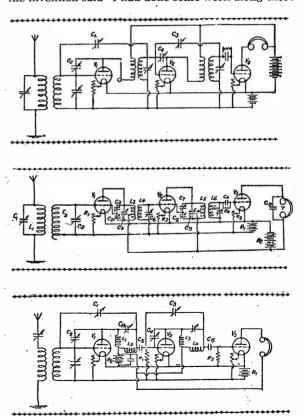
-Professor Hazeltine

NEW FACTS ABOUT A GREAT INVENTION

How it Came to be Sold to America

ciated the merits of the Neutrodyne, yet neither have hitherto fully done so in Great Britain. The wide publicity and dozens of demonstrations of the "Elstree Six" have, after three and a half years, made the Neutradyne 'catch on' '

Professor Hazeltine himself stated that he did not contemplate a wireless receiver, his idea being to use neutralising in land-line telephony. Only later did he develop the principle in wireless sets. At a luncheon at the Savoy Hotel, in front of over one hundred guests, Professor Hazeltine paid generous tribute to the work of Scott-Taggart and discussing the invention said "I had done some work along those



Three of the actual circuits included in Patent 217971 of January 2nd, 1923, in the name of Scott-Taggart.

Headlines from "Wireless" of July 31st, 1926 exposing the story behind the invention of the Neutrodyne.

lines generally and the result was the receiver known as the Neutrodyne. Similar work was being done by Mr. Scott-Taggart and I feel that while we in America call it the Hazeltine Neutrodyne we should in England call it the Scott-Taggart Neutrodyne".

S.T.'s patent had been published in detail in Wireless Weekly in June 1923, and although British industry was fully aware of the inventor's claims not a single firm approached S.T. As Wireless noted "this was not to be wondered at; as recently as a year ago probably the largest manufacturer of broadcast receivers declared publicly that the Neutrodyne was dying out in America and that it could never become popular in this country". After this turn down by the trade here the S.T. patent was bought by the Hazeltine Corporation although S.T. himself was unaware, at the time, of the identity of the purchasers since the arrangements were made through agents.

The ironical position developed in which a first class British patent had been sold to America who exported sets to us here which were licensed under the S.T. patent. The Hazeltine Corporation sold licence rights on the patent to no less than fourteen leading manufacturers in the USA. From Wireless again . . . "The value of the invention was appreciated from the first by the Americans and their enterprise is in striking contrast to our own . . . the Neutrodyne receiver in America has been a colossal success. No other invention has had such an extraordinary vogue. Up to date (1926) £7,000,000 worth of licensed Neutrodyne receivers have been sold. Professor Hazeltine and his associates draw patent licence fees to the extent of £120,000 per annum!"



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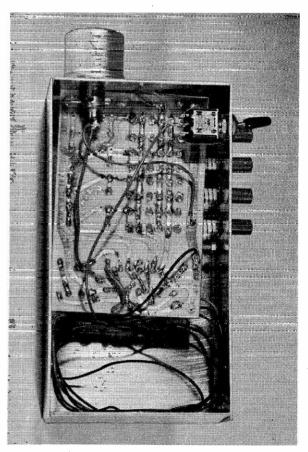
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## ULTRASONIC REMOTE CONTROLLER-

continued from page 966

## **OPERATION AND APPLICATIONS**

The prototype system operates at a range of 60 to 80 feet. Bearing in mind the medium of transmission, the best results are obtained under 'line of sight' operation. Ultrasonics behave in a similar fashion to light as far as propagation is concerned, so this means that ultrasonics will be reflected when they strike a hard surface. The prototype system was surprisingly omnidirectional under most conditions, and the user will soon acquaint himself with the possibilities and limitations of ultrasonic command.



The board is fitted into case with the two bolts on the switch unit. The 9V battery fits into the space at the bottom of the case.

Applications are numerous and one that immediately springs to mind is switching lights on and off. This may sound a little mundane to many readers, but how often have you fallen over the garden fork when fumbling about for the garage light switch? With the same transmitter in your pocket, you can turn on entrance lights when you drive in at night, and turn on the radio for the news.

One application of the carrier-only system, that has not been investigated at length in this article, is that of burglar alarms. One American firm now produces a level sensing IC and tone alarm, that will be set off if any one of four inputs indicates a change of state of 25% or so. The change of state can be brought about by interference with the ultrasonic carrier, breaking of light beams etc.

## ON RECENT DEVELOPMENTS

## DODGY FOOD TESTER

EVERY once in a while one reads or hears of a major scare over tinned foods. It's all the fault of those grotty little bacteria and the technical name is Botulism. The "normal" way in which tinned food is tested is to grow a culture and then to turn this over to highly skilled personnel for microscopic analysis. Trouble is that all this culture growing and analysing takes a lot of time and skill.

This is where the "Bactometer" comes into its own. At present it is undergoing various tests, but the great advantages claimed are that the equipment will make the necessary results available in less than an hour, and it doesn't need skilled people to operate it—it's automatic.

How does the Bactometer work? Basically, it is a very sophisticated bridae which measures impedance. The test material is prepared in a liquid culture and a "reference" is made using a medium which is sterile. By putting these two separate cultures into a favourable growing environment the impedance of each is plotted. If there is no growth in the test sample, then the impedance will not differ greatly. However, if growth does take place, indicating contamination, then the impedance drops giving a positive sign. It would appear that bacteria have no resistance to the advance of science! Incidentally, once the samples have been put into the Bactometer, the entire process from that point on is automatic and thus no skill is required to operate the equipment. A chart recorder plots the impedances giving a written printout proof. Perhaps a monitary Bactometer could be devised for bank managers to detect overdrafts?

## JAP REPORT

A piece of information which surprised me was that the Japanese home colour television market is approaching saturation point. Apparently some 80% of Japanese households now have colour television and manufacturers are talking about a push to get people thinking about a second set! The scene is an interesting one. Originally, the Japanese manufacturers considered that the smaller colour set was the trend for Japan. Now, Aiwa, a Japanese Company, are to import large screen (24-26in). American models. This has caused quite a stir.

On the Japanese market itself, the accent is now on power saving and this has been underlined with the recent energy availability problems which have become almost international. Meanwhile, Philips, the Europeanbased Dutch giant, have a 26in. colour tube which heats to emmission in only five seconds and yet uses 20% less power than earlier designs. The Japanese finding was that people like to switch on the set and see results almost immediately. The public has been weaned to this by the use of solid state.

## **VIDEO SYSTEMS**

Video recording systems have been in the news, and still are. One favourite method is to record the video signal using a laser beam. On playback, another small laser is used to detect the signal. This approach has been proposed for other markets too, such as mass memories, audio-visual and industrial control. Now, just as everyone is getting used to the idea, a Company has come up with a video playback system which uses the humble 25W light bulb. The bulb needs only about £30 of standard parts to form the basis of the playback system or video record player.

A laser is used to make the video disc by "printing," a series of dots whose individual diameters and densities vary according to the amplitude of the signal being recorded. Some 60 minutes of playing time is provided, the analogue signal being compacted by some 6:1. The train of dots form a spiral which roughly follows the path

that a groove would take in an ordinary audio record.

Playback is achieved on await for it—transparent turntable. The 25W light bulb is located beneath the turntable so that it's light shines through both turntable and record. The light is detected by three light sensors or photodiodes. Two of these are responsible for tracking and keeping the lens system in line with the train of dots spiral. The third sensor detects the variations in diameter/density of the dots and this is converted into an amplitude modulated signal. This is mixed with a very small r.f. signal and the resultant a.m. carrier is then fed directly to the aerial terminal of a standard television receiver, Initial experiments have shown that the system should be suitable for digital. audio and analogue recordings.

Please note, the system is experimental and will not be on the market for some time. This column describes state of the art things, and this is one of them.

## **BOOM TIME**

The last year has been a iremendous boom time for the electronics industry. So much so that materials are running out and suppliers just cannot keep pace with the demands of industry. This means that components are becoming scarcer. It is not unusual to find a supplier quoting a 72-week delivery. Couple this with the oil problem and it looks more serious. Plastics are tied in with oil. Although plastic packaged semiconductors are cheaper than ceramic ones, could it be the other way round soon? One thing is easily possible—the price of semiconductors and particularly i.cs may rise soon. It could be the old story of supply and demand, perhaps even a black market. Psst, wanna a buy an OC71 guv?





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POCKET FIVE

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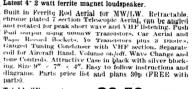
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## LEARNING BY PRACTICAL PROJECT STEPS

## PART 5—EMITTER FOLLOWER

T is natural to think that the role of an amplifier is simply to take small voltages and turn them (somehow) into high voltages but this is by no means the end of the story.

Apart from voltage amplifiers we often come across circuits that fulfil some role of amplification that is not always so obvious or, for that matter, easily measurable. To name a few; we have current amplifiers, power amplifiers, impedance converters (sometimes called buffer stages), inverters and operational amplifiers. The transistor can give us all these options in different forms of circuit and we shall explore some of these obvious and not-so-obvious applications.

The transistor is basically a current sensitive device so it would not be unreasonable to look, firstly.

at its role as a current amplifier. The circuit we shall consider is the emitter follower (briefly mentioned in a previous part). The name current amplifier implies that we require a circuit that needs only a very small input current generated by a given voltage to give a larger current in an output circuit with more or less the same sort of voltage swings. If we say that the input and output voltage swings are about the same but impose the condition that the input current is small, it immediately means that the resistances involved in the input circuitry are greater than those we are likely to see in the output (which has larger current variations for the same voltage changes). Fig. 29 is a test circuit that will demonstrate what we mean.

VR1 is used simply to provide a source of variable

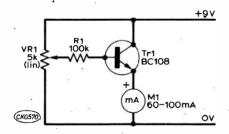


Fig. 29. The emitter current measured is proportional to the base current through R1. The latter is set by VR1.

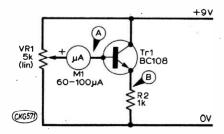


Fig. 30. The transistor draws only the amount of base current necessary to make the potential at B rise to that of A (less the Vbe of the transistor).

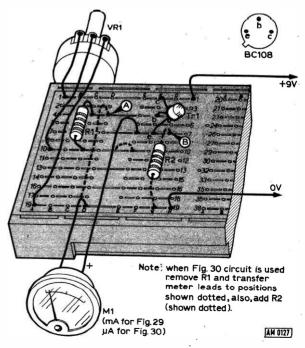


Fig. 31. Layout for Figs. 29 and 30.

voltage that is used to drive base current into Tr1. If you assume the internal resistance of the testmeter (set to current) is zero it is a simple calculation to work out the base current supplied.

The maximum will be when VR1 is set to the top of its track (+9V as a source). The base current is

therefore 
$$\frac{9-Vbe}{100k\Omega}=\frac{8\cdot 4}{100}$$
 mA =  $0\cdot 084mA$ .

The current you will measure in the emitter circuit will be as high as 30mA for this setting of VR1. The increase in current flow is up by a factor of about 300. One could, in fact, have calculated the emitter current from the current gain of the transistor ( $h_{\text{FE}}$ ):

 $collector\ current = h_{FE} \times base\ current$ 

emitter current = collector current + base current  $\therefore$  Ie = (h<sub>FE</sub> × Ib) + Ib

In other words  $Ie = Ib \times (h_{FE} + 1)$ 

Because the hFE for a BC108 is approximately 200 to 300 one can see that the emitter current we measure is about two to three hundred times the base current. An immediate problem that springs out from this calculation is that you can only calculate the result if you know a precise value for  $h_{\text{FE}}$  but this varies a lot from one transistor to another-even though they may have the same type number. Thus the precise currents you measure in the emitter circuit will be dependent on the particular transistor you are using. Later you will see that steps have to be taken in many circuits to compensate for hFE variations from one transistor to another. It is clear, then, that we can get an amplification of current by using a transistor and if you vary VR1 you should see that to all intents and purposes the emitter current is directly proportional to the base current supplied.

We can now carry out a variation on the same theme by using the circuit of Fig. 30. At first glance you might think that the base current will be very much greater than  $60\mu A$  because all that seems to be limiting the base current is R2 in the emitter circuit. Firstly try it out and you should find that by adjusting VR1 the measured base current goes from zero up to a maximum of about 30 to  $60\mu A$  (depending on your own particular transistor). When VR1 is set to maximum it is tempting to think that the base current is approximately 9V divided by the  $1,000\Omega$ of the emitter resistor (about 9mA). If you make this guess you are overlooking the fact that as soon as you start to apply base current into Tr1 we will cause emitter current to flow in R2 and this causes a voltage drop across it. The voltage at B will rise. The

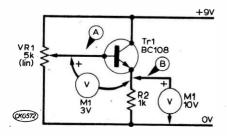


Fig. 32. At all settings of the slider the potential difference between A and B is about 600mV showing that B "follows" A.

base current is thus controlled by the new potential difference between points A and B (NOT between point A and ground!).

You can see the voltage at point B rising by switching the meter to volts (meanwhile connecting point A directly to the base of the transistor, Fig. 32). Notice that there is very little difference between the voltages set at A and those that are given at B. The difference is basically 600mV which is the forward base emitter voltage drop. Return to Fig. 30 and change R2 for a  $10k\Omega$  resistor. You will see that the base current drawn by the transistor reduces by a factor of ten. Reduce R2 to  $100\Omega$  and the base current increases by about a factor of ten from its original value. The transistor thus gives an output voltage that follows the input voltage and the "following" effect forces the transistor to draw only sufficient base current to make the voltage at B approach that at A.

The input current is self limiting and makes the transistor look as if it has a very high base resistance. We say it presents a high input resistance to the voltage source and the output is more or less the same voltage as the source but the emitter current is very much greater than the input current.

## Touch sensitive switch

Another way of looking at this circuit is to consider a limited current coming from a voltage source—Fig. 33—as one might find with a touch sensitive switch where the very small current flowing through the fingers is needed to cause a higher current in another circuit. Bridging the two contacts with the fingers causes base current to flow and the emitter voltage rises towards the voltage of the source (+9V). Depending on how good the skin contact is and how good a conductor you are you may pass sufficient base current to make point B rise to +9V.

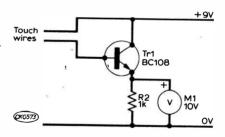


Fig. 33. The small current flowing through the fingers generates emitter current and this is shown as a potential difference across R2.

We have selected a value for R2 so that you may not be able to produce a low enough skin resistance to pass sufficient base current and hence the voltage at B may not follow the base exactly and the voltage you measure will vary depending on how hard you press the contacts with your fingers. To make the voltage at B go to +9V with ease you must reduce the base current requirement and this is done by reducing the amount of emitter current needed to make point B follow the input. This is done by increasing the value of R2 to  $10k\Omega$ .

A more precise way of obtaining an output voltage at the emitter of Trl proportional to the body's resistance is to make sure that the emitter current, to be controlled, is not limited by the  $h_{\rm FE}$  of the transistor. You can then use a circuit that relies on the potential divide effect of the body resistance and R1 (Fig. 34) to give a potential at the base of Tr1 that is reflected in the potential at its emitter (less only the base emitter drop). The current available in the collector/emitter circuit of the transistor is still many times greater than the current flowing through the body and can be used as a signal into other circuits. Fig. 36 is such a circuit. When the contacts are bridged with the fingers a small current flows into the base of Tr1; this produces a higher current in the collector/emitter circuit of Tr1 some, of which,

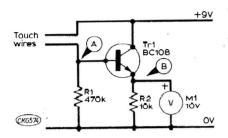


Fig. 34. The potential at A is set by the potential divide of R1 with skin resistance and this is reflected by the potential measured at B in the lower resistance emitter circuit.

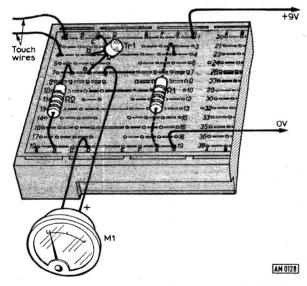


Fig. 35. Layout for Fig. 33. To carry out experiment in Fig. 34 change R2 to  $10k\Omega$  and insert R1.

passes into the base circuit of Tr2 where it is further amplified to provide base drive to the inverter Tr3 which needs quite a high base current to turn on the lamp. Bridging the contacts turns on the lamp. If you think about it R2 serves no useful purpose—on the contrary it wastes useful emitter current of Tr1 which could be used as base current for Tr2; thus it can be removed and the circuit still works. The combination of one emitter follower feeding into another emitter follower is often seen when a circuit requires a very high input resistance and the combination is called a super alpha pair.

## Super alpha pair

The maximum current you can control in the emitter circuit is given by the base current multiplied by the hFE for the transistor. Thus the current gain of a typical single stage using a BC108 is about 200 to 300. For a super alpha pair the gain would be  $200\times200=40,000$ . In our circuit of Fig. 36 the base current required by Tr3 to turn on the lamp is about 0.2mA. We thus need only one fortythousandth of this as base current into the first stage i.e. about 0·005μA. In reality more than this has to flow through the skin because we have the  $470k\Omega$  of R1 shunting the base of Tr1 to ground. It is possible to remove R1 and the circuit will become even more sensitive -the only problem is that it may become too sensitive and trigger by capacitively coupled pick up by Trl's base. In theory R3 could be omitted but this is unwise; it is there to protect the base emitter junction of Tr3 from too high a current.

The emitter follower principle is often used as the active element in voltage regulators where a variable voltage output is needed from a power supply at quite high currents.

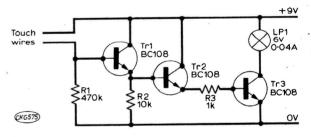


Fig. 36. A touch sensitive light switch. In practice R2 can be omitted as can R1 but removal of the latter might make the circuit over sensitive. R3 should not be omitted.

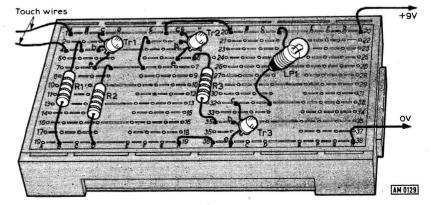


Fig. 37. Layout for Fig. 36.

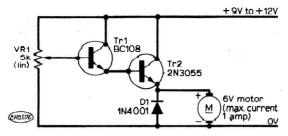


Fig. 38. Simple voltage adjustment to control the speed of low voltage d.c, motors. D1 is included to prevent high reverse voltages from the brushes damaging the transistor. The output is not short circuit protected.

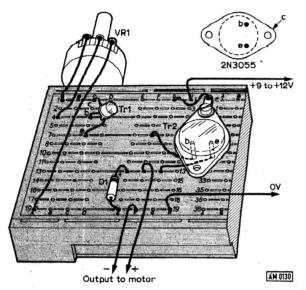


Fig. 39. Layout for Fig. 38. The short leads on Tr2 should be extended to connect to the T-Dec and the collector connection made via a solder tag bolted to Tr2.

It is expensive to use a high power potentiometer but a small potentiometer and a power transistor will do the job very nicely. Provided the output current does not exceed the rating of the transistor and is less than the Ib  $h_{\rm FE}$  product the output voltage of Fig. 38 will be approximately equal to the potential set at the wiper of the potentiometer (less the transistors forward voltage drops). With the power transitors

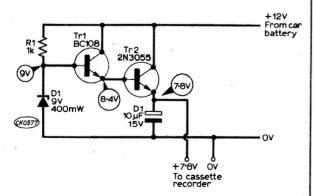


Fig. 40. A simple voltage dropper and regulator for running a cassette recorder from a car battery. Not short circuit protected.

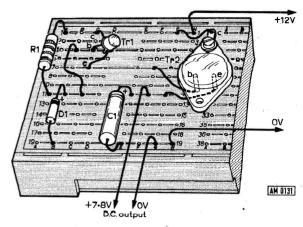


Fig. 41. Layout for Fig. 40.

sistor shown the circuit would act as a crude speed controller for small model motors. The wiper of the potentiometer provides the reference voltage. This could be generated by means of a zener diode and thus a stabilised power supply could be made to run, say, a cassette tape recorder from a 12V car battery (Fig. 40). C1 is included to suppress ignition noise.

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500V AC         £2.30         -         -         £2.80         -         £2.65         -	
S Meter 1mA	5 -
VU Meter         £2.65         £2.70         £3.60         £3.70         £4.55         £3.65         £3.70         £4.30         £2.90         £3.15         £3.50         £3.50         £3.61         £3.70         £4.30         £2.90         £3.15         £3.50         £3.50         £3.61         £3.90         £2.60         —	+-
1A AC	
5A AC	5 -
10A AC	<del>  -</del>
20A AC - £2.40 £2.50 £2.60 £3.90 £2.60	-
	+
	=
30A AC - £2.40 £2.50 £2.60 £3.90 £2.60	=
50A AC £2.60	=
50mA AC £2.60	=
100mA AC £2.60	-
200mA AC £2.60	=
500mA AC	=
50mV DC £2.90	=
100mV DC 12.30	=
500mA/5A DC	- - - - - - - - - - - - - - - - - - -
1/15A UC	            
5/15V DC	
5/50V DC	            

## **GINFORMATION**

SEW PANEL	. METERS-SI	IZES AND	FIXING
	Front	Panel Hole	Fixing
Model 38P	42 x 42mm	32mm dia.	4 studs
Model 45P	50 x 50mm	38mm dia.	4 studs
Model 52P	60 x 60mm	48mm dia.	4 studs
Model 65P	86 x 78mm	57mm dia.	4 studs
Model 85P	120 x 110mm	98mm dia.	4 studs
Model 65	80 x 80mm	64mm dia	4 studs
Model S80	80 x 80mm	65mm dia.	4 studs

	Front	Panel Hole	Fixing
Model SW100	100 x 80mm	65mm dia.	4 studs
Model SD460	59 x 46mm	38mm dia.	4 studs
Model SD640	85 x 64mm	45mm dia.	4 studs
Model SD830	110 x 83mm	58mm dia.	4 studs
Model PE70	90 x 34mm	70 x 31mm	2 holes
Model ED107	Size: 100 x 90	x 150mm hig	h
including to	erminals.		



P\$200 Regulated POWER SUPPLY UNIT

aurrly UNIT Solid state. Variable output 5—20V DC up to 2 Amp. Inda-pendent meters to monitor voltage and current. Output 220/240V AC. Size: 190 x 136 x 98mm.



OUR PRICE £19.95

PS100B Regulated POWER SUPPLY UNIT

Solid stete. Output 6, 9 or 12V DC up 10 3 Amp. Meter to monitor current. Input 220/240V AC. Size: 100 x 82 x 159mm. OUR PRICE £11.97



P&P 25p

MCA220 Automatic Voltage Stabiliser

Input 88-125Y AC or 176-250 V AC. Output 120V AC or 240V AC. 200V A rating. **OUR PRICE** 



## TO3 Portable OSCILLOSCOPE

3" tube. Y amp sensitivity 0.1Vp-p/ CM, Bandwidth: 1.5Hz-1.5MHz. Input Impedence: 2 MOhms, 25pF. X amp. sensitivity: 0.9Vp-p/CM. Bandwidth 1 5Hz-



O.SVP-p/CM.
Bandwidth 1.5H-z
900kHz. Inpedence 2 Meg
10Hz. to 300kHz. Syn-kronisation.
Internal or external. Illuminated scale
140 x 215 x 330mm. Weight 15%ibs.
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with handbook.

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OSCILLOSCOPE USCILLUSCUPE
5 MHz pass band.
Separate Y1 and Y2
amplifiers. Rectangular 5" x 4" CRT.
Calibrated triggered sweep from 0.2usec. to 100 milli-sec/cm.
Free running time Free running time base, 50Hz-1MHz. Built-in time base



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## CI5 PULSE OSCILLOSCOPE

C15 PULSE OSCILLO: For display of pulsed and periodic wave-forms in electronic circuits. VERT. AMP Bandwidth: 10MHz. Sensitivity at 100kHz VRMS/nm: 0.1—25. HOR. AMP. Band-worstito-VRMS/nm: 0.3—25. Presst triggered sween •



LAFAYETTE HAGOO Receiver



General coverage 150-400kHz, 560 kHz-30MHz. FET front end 2 mechanical filters, product detector. Noise limitor, variable BFO, S Meter, Bandspread and RF gain. 220/240V AC or 12V DC. Brand new with instructions. OUR PRICE £50.00 P&P 50p

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Nine wave-bands covering 1.8-29.7 MHz, 144-146MHz and 10MHz WWV SSB, CW, AM and FM. AF output is more than 1 watt. S Meter. Squelch control. BFO. Variable AF and RF controls. 4-16 ohm output and jack for phones. Power requirement 100, 240V AC, 12/14V DC. Size: 270 x 140 x 310mm.

OUR PRICE £132.50 Carr. paid

## TRIO 9R59DS RECEIVER



Four bands covering 550kHz to 30 Paul Dands Covering 550kHz to 30 MHz continuous and electrical bendwitz continuous and electrical bendwitz continuous and electrical bendwitz continuous and phone jack. SSB—CW Author and phone jack. SSB—CW Author and spreads parate band spreads policy and spreads band spreads provided the parate out 1% wast. Variable RF and AF gain controls. 115/250V AC, with instructions. OUR PRICE £42.50 Carr. paid

TRIO JR310 SSB RECEIVER



Covers 3.5, 7, 14, 21, 28, 28.5 and 29.1MHz bands and WWV 15MHz. SSB, AM and CW. AF output more than 1W. Crystal controlled BFO for SSB. SMeter, ANL etc. AC 110/120–220/240V. Size: 330 x 179 x 310mm. OUR PRICE £75.00 Carr. paid

TRIO TR2200 RECEIVER VHF transceiver Will transmit and Will transmit and receive on six channels between 144—146MHz. 1 watt transmitter. 12V DC internal or external supply. Built-in charger for Ni-Cad cells. Power/volume



Power/volume switch, squelch control, channel selector, mic. socket, earphone/external speaker socket. Complete with microphone and 144,48, 144,72 & 145,32 crystals. Size: 134 × 58 x 180mm.

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Four bands covering 560 KHz/30 BFO, Built-in speaker. Operates 220/240V AC. Brand new with

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Solid state mobel transcever for 12 volt DC neg. The transcerver for 12 volt DC neg. The transcerver for 12 volt DC neg. The 12 volt DC neg. The 14 and 146MHz. Power sistast 10W and IW sylichable. Control of 14 of 15 volt neg. Squetch and channel selector. Internal 3" speaker. Complete volt ynamm in 12 volt neg. The 15 volt neg. Virginia volt neg. Pl I switch, three volt neg. The 15 volt neg. The 15 volt neg. Virginia volt neg. OUR PRICE £75.00 . P&P 50p

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Built-in driver unit. Impedence 16 ohms. Power rating 10W. Response 380— 7000Hz. Size app. 6" x 6". Weather and shock protected.

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