

ELECTRONICS TODAY

electronics today international

JULY 1977

35p

NERVOUS TENSION?

LEARN TO RELAX
WITH OUR
GSR MONITOR

8-PAGE

DATA

SUPPLEMENT

TV GAMES RIFLE

ACTIVE FILTERS

TACHOMETER

MICROAMP

MPU CALCULATOR REVIEW



WIRING PEN OFFER

NEWS . . . CONSTRUCTION . . . DEVELOPMENT . . . AUDIO

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POLYESTER CAPACITORS: Axial lead type. Values are in -F)
400V: 0.001 0.0015 0.0022 0.0033 8p; 0.0047 0.0068 0.01 0.015 0.018 9p; 0.022 0.033 10p;
0.047 0.068 14p; 0.1 15p; 0.15 0.22 22p; 0.33 0.47 39p; 0.68 45p;
160V: 0.039 0.15 0.22 13p; 0.33 0.47 22p; 0.68 1.0 29p; 1.5 33p; 2.2 38p; 4.7 47p;
DUBILIER: 1000V: 0.01 0.015 16p; 0.022 18p; 0.047 16p; 0.1 34p; 0.47 43p; 1.0 85p

POLYESTER RADIAL LEAD: Values in -F) 250V:
0.01 0.015 6p; 0.022 0.027 7p; 0.033 0.047 0.068 0.1 8p; 0.15 12p; 0.22 0.33 14p; 0.47 16p; 0.68 20p; 1.0 24p; 1.5 27p; 2.2 31p

ELECTROLYTIC CAPACITORS: Axial lead type. Values are in -F)
250V: 100µF 50p; 100V: 20µF 63V: 0.47 1.0 1.5 2.2 2.5 3.3 4.7 6.8 8 10 15 22 9p; 47 32 50 12p; 63 100 27p; 50V: 1.0 9p; 50µF 100 220 25p; 470 50p; 1000 2200 68p; 40V: 22 33 9p; 100 12p; 330 62p; 35V: 3.3 7p; 3.3 40 470 50p; 1000 48p; 25V: 10 22 47 6p; 80 100 160 8p; 220 250 13p; 470 640 25p; 1000 27p; 1500 30p; 2000 34p; 3300 52p; 4700 54p; 16V: 10 40 42 6R 7p; 100 125 8p; 470 16p; 1000 150p; 2000 220p; 34p; 10V: 4.7 10p; 540 10p; 100 14p.

TANTALUM BEAD CAPACITORS: 35V: 0.1-F 0.22 0.33 0.47 0.68 1.0 2.2 3.3 4.7 6.8 25V: 1.5 1.0 20V: 1.5 16V: 1.0 1.0 F 13p each
100V: 22-F 33 47 6V: 47 68 3V: 10V: 20p each

MYLAR FILM CAPACITORS: 100V: 0.001 0.002 0.005 0.01-F 5p 0.015 0.02 0.04 0.05 0.056-F 6p 0.1-F 0.15 0.2 7p. 50V: 0.47-F 10p

CERAMIC CAPACITORS 50V: Range: 0.5pF to 10 000pF
0.015-F 0.022-F 0.033-F 0.047-F 4p 0.1-F 6p.

SILVER MICA: (Values in pF) 3.3 4.7 6.8 10 12 22 33 47 50 68 75 82 85 100 120 150 220 9p each 250 300 330 360 390 600 820 15p each 1000 1800 2000 2200 20p each

CERAMIC TRIMMER CAPACITORS: 2.2pF 5.1pF 8.2pF 8.3pF 20p

MINIATURE TYPE TRIMMERS: 2.2pF 5.1pF 8.2pF 8.3pF 20p

COMPRESSION TRIMMERS: 5.0pF 8.0pF 20.140pF 190pF 25p 50.200pF 100.500pF 1250pF

JACK PLUGS: Screened chrome 2.5mm 12p 3.5mm 15p MONO 23p STEREO 31p

SOCKETS: 2.5mm 8p 3.5mm 8p MONO 13p STEREO 15p

SLIDE 250V: 1A DP C 14p 1A DP 16p 4 pole 2 way 24p

PUSH BUTTON: Spring loaded SPST on/off 55p SPDT C-over 65p DPDT 6 tag 85p

SWITCHES + PUSH BUTTON: Miniature Non-Locking Push to Make 15p Push to Break 25p

ROCKER: (white) 10A 250V SP changeover centre off 28p

ROCKER: (black) on/off 10A 250V 23p

ROCKER: Illuminated white Lights when on 3A 240V 52p

ROTARY: (ADJUSTABLE STOP) 1 pole 2.12 2 pole 2.6W 3 pole 2.4W 4 pole 2.3W 4.1p

ROTARY: Mains 250V AL 4 Amp 42p

DI L SOCKETS: (Low Profile - Texas) 8 pin 10p; 14 pin 12p; 16 pin 13p; 24 pin 30p; 28 pin 45p; 40 pin 58p

FERRIC CHLORIDE: Anhydrous Tit bag 65p + 30p p&p

DALO ETCH Resist Pen + Spare Tip 75p*

COPPER BOARDS: Fibre Glass Single Sided 6" x 12" 115p Double Sided 6" x 6" 78p 6" x 12" 148p

VOLTAGE REGULATORS: 723C 45p TBO25B 95p

TTL 74*

7400	14	7480	50	74161	116	4021AE	105	702	75	MC1403L	148
7401	14	7481	114	74162	116	4022AE	95	709C	14	pin 30	260
7402	16	7482	82	74163	116	4023AE	20	710*	36	MC1310P	185
7403	16	7483	95	74164	121	4024AE	85	741C	8	pin 22	195
7404	20	7484	125	74165	120	4025AE	19	74705	36	MC1458B*	90
7405	22	7485	125	74166	161	4026AE	185	748C	22	MC1710CG	101
7406	22	7486	36	74167	362	4027AE	60	753	150	MFC6040*	97
7407	44	7489	390	74170	186	4028AE	95	8038CC*	345	MK50255*	550
7408	22	7490	43	74172	590	4029AE	109	AY 10212	670	MK50265*	775
7409	22	7491	80	74173	175	4030AE	58	AY 1 5051	125	MM2112N*	350
7410	15	7492	53	74174	175	4031AE	145	AY 1 67216	191	MM50326*	750
7411	25	7493	40	74175	90	4032AE	195	AY 3-8500*	792	NC390	160
7412	15	7494	85	74176	120	4040AE	110	AY 5 1724*	349	NE518A	210
7413	40	7495	73	74177	116	4042AE	85	AY 5 1230*	490	NE555*	41
7414	74	7496	82	74180	108	4043AE	100	AY 5 3500*	810	NE5560B*	99
7415	35	7497	262	74181	299	4044AE	95	AY 5 4007*	650	NE560*	325
7416	39	74100	135	74182	85	4045AE	145	BX1	249	NE582*	395
7417	39	74104	48	74184	184	4046AE	135	CA3011*	82	NE585*	395
7418	29	74105	48	74185	146	4047AE	99	CA3014*	137	NE585A*	175
7419	24	74107	33	74188	650	4048AE	58	CA3018	72	NE586*	170
7420	30	74109	84	74190	194	4049AE	60	CA3020	170	NE587*	182
7421	30	74110	54	74191	165	4050AE	55	CA3023	170	RAM2102-2*	247
7422	35	74111	76	74192	120	4054AE	120	CA3028A*	102	ROM2513*	750
7423	35	74116	193	74193	120	4055AE	120	CA3035	170	SG5402*	255
7424	38	74118	115	74194	120	4056AE	135	CA3036	180	SL414*	220
7425	18	74120	105	74195	95	4060AE	115	CA3043	190	SN72373	125
7426	28	74121	34	74196	118	4067AE	380	CA3046	80	SN76003	240
7427	48	74122	50	74197	118	4068AE	22	CA3048	220	SN76013	150
7428	33	74123	73	74198	248	4069AE	22	CA3075	80	SN76023	150
7429	33	74125	89	75150	000	4070AE	55	CA3080	80	SN76033	235
7430	17	74126	65	75491	80	4071AE	23	CA3081	190	SN76115	215
7431	74	74128	81	75492	90	4072AE	23	CA3089E	210	SN76227	175
7432	74	74132	75	CMOS*	-	4076AE	155	CA3090A	40	TAA621A	238
7433	130	74136	56	(RCA)	-	4077AE	60	CA3123E	200	TAA661A	155
7434	127	74141	72	4000AE	15	4078AE	22	CA3130*	94	TAA700	353
7435	94	74142	315	4001AE	15	4081AE	22	CA3140	95	TAA960	300
7436	116	74143	314	4002AE	17	4082AE	23	LM300H	170	1AD100	150
7437	82	74144	314	4006AE	115	4502AE	135	LM301A	39	1AD110	170
7438	80	74145	90	4007AE	18	4507AE	55	LM3018	140	TBA120S	90
7439	17	74147	275	4009AE	60	4510AE	135	LM318	195	TBA5400	220
7440	17	74148	160	4010AE	60	4511AE	175	LM319	195	TBA5500	355
7441	7	74150	128	4011AE	19	4512AE	130	LM380	98	TBA641BK1*	225
7442	17	74151	79	4012AE	18	4516AE	140	LM381	170	TBA651	170
7443	17	74153	82	4013AE	55	4518AE	130	LM382	125	TBA800	90
7444	32	74154	150	4015AE	99	4519AE	65	LM3900	65	TBA810S	110
7445	28	74155	76	4016AE	55	4520AE	130	LM3909A	70	TBA820	80
7446	32	74157	95	4017AE	104	4528AE	140	M257AA*	85	TBA9200	350
7447	36	74158	80	4018AE	104	4532AE	128	M258*	197	ZN2020	320
7448	36	74159	225	4019AE	60	4539AE	120	MC724*	95	ZN11	110
7449	36	74160	116	4020AE	125	4585AE	105	MC845P	150	ZN424	125

TRANSISTORS

AC117	35	BC160*	32	BF185*	30	OC28*	70	ZTX303	22	2N3706	10
AC125*	22	BC167	12	BF194	10	OC29*	70	ZTX304	24	2N3707	11
AC127*	18	BC169C	12	BF195	10	OC35*	70	ZTX311	15	2N3708	12
AC128*	18	BC170	11	BF196	10	OC36*	53	ZTX312	15	2N3709	12
AC141*	22	BC171	11	BF198	15	OC42*	17	ZTX341	20	2N3710	12
AC141K*	38	BC172	11	BF199	15	OC43*	25	ZTX501	15	2N3715*	245
AC142*	38	BC177*	18	BF224A	15	OC45*	17	ZTX502	19	2N3721*	164
AC176*	18	BC179*	18	BF248	30	OC46*	35	ZTX503	19	2N3772*	175
AC187*	18	BC182	10	BF256*	34	OC71*	23	ZTX504	28	2N3775*	275
AC188*	18	BC182L	12	BF258*	38	OC72*	27	ZTX505	28	2N3819	22
AC189*	18	BC183	10	BF259*	48	OC77*	56	ZTX506	40	2N3820	44
AC191*	18	BC183L	11	BF261	30	OC78*	56	ZTX507	40	2N3823*	49
AC192*	24	BC184	12	BF269	11	OC81*	25	ZTX508	15	2N3824*	39
AC193*	24	BC184L	12	BF273	11	OC81D*	28	ZTX509	35	2N3866*	90
AC194*	24	BC186	22	BF274	25	OC82*	25	ZTX510	45	2N3904	18
AC195*	16	BC186*	23	BF275	23	OC82D*	35	ZTX511	17	2N3905	18
AC196*	16	BC187	12	BF276	25	OC83*	28	ZTX512	19	2N3906	18
AC197*	16	BC187L	13	BF277	11	OC83D*	44	ZTX513	40	2N4037*	39
AD149*	54	BC213L	13	BF283*	38	OC123*	115	ZTX514	18	2N4058	15
AD161*	38	BC214L	16	BF285*	40	OC139*	140	ZTX515	27	2N4061*	255
AD162*	38	BC207B	16	BF286*	40	OC140*	125	ZTX516	34	2N4191*	66
AF114*	20	BC232	14	BF288*	14	OC170*	40	ZTX517	51	2N4236	47
AF115*	20	BC238*	21	BF289*	24	OC171*	42	ZTX518	61	2N4289	24
AF116*	20	BC241*	37	BF290*	28	OC201*	125	ZTX519	61	2N4859	35
AF117*	23	BC246*	47	BF291*	23	OC202*	135	ZTX520	23	2N4871	32
AF121*	20	BC247*	20	BF292*	25	OC204*	150	ZTX521	25	2N5136	12
AF124*	30	BC248*	20	BF293*	20	OC205*	150	ZTX522	27	2N5137	12
AF125*	30	BC249*	20	BF294*	17	OC206*	150	ZTX523	27	2N5138	12
AF126*	30	BC250*	20	BF295*	17	OC207*	150	ZTX524	27	2N5139	12
AF127*	30	BC251*	20	BF296*	17	OC208*</					

electronics today

international

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EDITORIAL AND ADVERTISEMENT OFFICES

25-27 Oxford Street
London W1R 1RF
Telephone 01-434 1781/2

Telex 8811896

HALVOR W. MOORSHEAD
Editor

RON HARRIS B.Sc
Assistant Editor

GARY EVANS
Editorial Assistant

JIM PERRY
Specials Editor

TONY ALSTON
Project Development

JULIAN ZINOVIEFF
Production

PAUL EDWARDS
Technical Drawing

SANDRA ZAMMIT-MARMARA
Subscriptions

MARGARET HEWITT
Administration

DAVID LAKE (Manager)

BRENDA GOODWIN

KIM HAMLIN
Reader Services

For Advertising Enquiries ring
MARK STRATHERN
on 434 1781/2

INTERNATIONAL EDITIONS

AUSTRALIA Collyn Fivers
Publisher
Les Bell
Assistant Editor

HOLLAND Anton Kriegsman
Editor-in-chief

CANADA Mike Kenward
Editor

FRANCE Denis Jacob
Editor-in-chief

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SEMICONDUCTORS - COMPONENTS

TRIACS

2 Amp			10 Amp		
Volts	TO5 Case	Price	Volts	TO48 Case	Price
100	TR12A/100	£0.31	100	TR110A/100	£0.77
200	TR12A/200	£0.51	200	TR110A/200	£0.92
400	TR12A/400	£0.71	400	TR110A/400	£1.12

6 Amp			10 Amp		
Volts	TO66 Case	Price	Volts	TO220 Case	Price
100	TR16A/100	£0.51	400	TR110A/400P	£1.12
200	TR16A/200	£0.61	DIACS		
400	TR16A/400	£0.77	BR100	£0.23	D32 £0.23

SUPER UNTESTED PAKS

PAK No.	Description	Order No.	Price
U50	100 Germ. Gold bonded OA47 diode	16130	£0.60
U51	150 Germ. OA70/81 diode	16131	£0.60
U52	100 Silicon Diodes 200mA OA200	16132	£0.60
U53	150 diodes 75mA 1N4148	16133	£0.60
U54	50 Sil Rect Top Hat 750mA	16134	£0.60
U55	20 Sil Rect Stud Type 3 Amp	16135	£0.60
U56	50 400mW Zeners D07 Case	16136	£0.60
U57	30 NPN Trans BC107/B Plastic	16137	£0.60
U58	30 PNP Trans BC177/178 Plastic	16138	£0.60
U59	25 NPN TO39 2N697/2N1711 sil	16139	£0.60
U60	25 PNP TO59 2N2905 silicon	16140	£0.60
U61	30 NPN TO18 2N106 silicon	16141	£0.60
U62	25 NPN BF50 51	16142	£0.60
U63	30 NPN Plastic 2N3906 silicon	16143	£0.60
U64	30 PNP Plastic 2N3905 silicon	16144	£0.60
U65	30 Germ. 0071 PNP	16145	£0.60
U66	15 Plastic Power 2N3055 NPN	16146	£1.20
U67	10 TO3 Metal 2N3055 NPN	16147	£1.20
U68	20 Unijunction trans 1N543	16148	£0.60
U69	10 1 amp SCR 1039	16149	£1.20
U70	8 3 amp SCR 1066 case	16150	£1.20

Code No's mentioned above are given as a guide to the type of device in the pak. The devices themselves are normally unmarked.

COMPONENT PAKS

Pack No.	Qty.	Description	Order No.	Price
C1	200	Resistor mixed value approx (Count by weight)	16164	£0.60
C2	150	Capacitors mixed value approx (Count by weight)	16165	£0.60
C3	50	Precision resistors. Mixed values	16166	£0.60
C4	80	1/4W Resistors mixed preferred values	16167	£0.60
C5	5	Pieces assorted ferrite rods	16168	£0.60
C6	2	Tuning gangs MW 1W VHF	16169	£0.60
C7	1	Pack wire resistors assorted colours single strand	16170	£0.60
C8	10	Reed switches	16171	£0.60
C9	3	Micro switches	16172	£0.60
C10	15	Assorted pots	16173	£0.60
C11	5	Metal jack sockets 3 x 3.5mm 2 x standard switch types	16174	£0.60
C12	30	Paper condensers preferred types mixed values	16175	£0.60
C13	20	Electrolytics trans types	16176	£0.60
C14	1	Pack assorted hardware - Nuts / bolts / grommets etc	16177	£0.60
C15	5	Mains side switches ass	16178	£0.60
C16	20	Assorted lag strips and panels	16179	£0.60
C17	15	Assorted control knobs	16180	£0.60
C18	4	Rotary wave change switches	16181	£0.60
C19	2	Relays 6 - 24V operating	16182	£0.60
C20	1	Pak copper laminate approx 200 sq cms	16183	£0.60
C21	15	Assorted fuses 100mA-5 amp	16184	£0.60
C22	50	Metres PVG sleeving assorted size and colour	16185	£0.60
C23	60	1/2 watt resistors mixed preferred values	16186	£0.60
C24	25	Presets assorted type and value	16187	£0.60
C25	30	Metres stranded wire assorted colours	16188	£0.60

SLIDER PAKS

Pack No.	Qty.	Description	Order No.	Price
S1	6	Slider potentiometers mixed values	16190	£0.60
S2	6	Slider potentiometers all 470 ohms	16191	£0.60
S3	6	Slider potentiometers all 10k lin	16192	£0.60
S4	6	Slider potentiometers all 22k lin	16193	£0.60
S5	6	Slider potentiometers all 47k lin	16194	£0.60
S6	6	Slider potentiometers all 47k log	16195	£0.60

CERAMIC PAKS

Containing a range of first quality miniature ceramic capacitors Unrepeatable value

Order No.	Price	
MC1	24 miniature ceramic capacitors 3 of each value - 22pf 27pf 33pf 39pf 47pf 68pf 82pf	16160 £0.60
MC2	24 miniature ceramic capacitors 3 of each value - 100pf 120pf 150pf 180pf 220pf 330pf 390pf 470pf	16161 £0.60
MC3	24 miniature ceramic capacitors 3 of each value - 470pf 560pf 680pf 820pf 1000pf 1500pf 2200pf 3300pf	16162 £0.60
MC4	24 miniature ceramic capacitors 3 of each value - 470pf 560pf 680pf 820pf 1000pf 1500pf 2200pf 3300pf	16163 £0.60

ORDERING

Please word your orders exactly as printed not forgetting to include our part number

V.A.T.

Add 12 1/2% to prices marked * Add 8 1/2% to others excepting those marked † these are Zero

CARBON POTENTIOMETERS

SINGLE GANG with wire end terminations 6mm x 50mm plastic shaft 10mm bushes supplied with shake-proof washer and nut. Tolerance ±20% of resistance

LINEAR TRACK

Value	No.	Price
1K	1831	£0.22
2K	1832	£0.22
4K7	1833	£0.22
10K	1834	£0.22
22K	1835	£0.22
47K	1836	£0.22
100K	1837	£0.22
220K	1838	£0.22
470K	1839	£0.22
1M	1840	£0.22
2M2	1841	£0.22

LOG TRACK

Value	No.	Price
4K7	1842	£0.22
10K	1843	£0.22
22K	1844	£0.22
47K	1845	£0.22
100K	1846	£0.22
220K	1847	£0.22
470K	1848	£0.22
1M	1849	£0.22
2M2	1850	£0.22

DUAL GANG. These high-quality pots are fitted with wire end terminations 6mm x 50mm plastic shaft 10mm bushes supplied with shakeproof washer and nut. Track tolerance ±20% but matched to within ±2dB of each other

LINEAR TRACK

Value	No.	Price
4K7	1851	£0.68
10K	1852	£0.68
22K	1853	£0.68
47K	1854	£0.68
100K	1855	£0.68
220K	1856	£0.68
470K	1857	£0.68
1M	1858	£0.68
2M2	1859	£0.68

LOG TRACK

Value	No.	Price
4K7	1860	£0.68
10K	1861	£0.68
22K	1862	£0.68
47K	1863	£0.68
100K	1864	£0.68
220K	1865	£0.68
470K	1866	£0.68
1M	1867	£0.68
2M2	1868	£0.68

SINGLE GANG SWITCHED. Fitted with double pole on-off switches. The switch action is incorporated within the rotary action of the pot. Switch rating 1.5 amps at 250V AC

LINEAR TRACK

Value	No.	Price
4K7	1870	£0.48
10K	1871	£0.48
22K	1872	£0.48
47K	1873	£0.48
100K	1874	£0.48
220K	1875	£0.48
470K	1876	£0.48
1M	1877	£0.48
2M2	1878	£0.48

LOG TRACK

Value	No.	Price
4K7	1879	£0.48
10K	1880	£0.48
22K	1881	£0.48
47K	1882	£0.48
100K	1883	£0.48
220K	1884	£0.48
470K	1885	£0.48
1M	1886	£0.48
2M2	1887	£0.48

VEROBOARDS

DRILLED COPPER P.C.B.

.1 Pitch			.15 Pitch		
Size	No.	Price	Size	No.	Price
2.5 x 5"	2201	£0.46	2.5 x 17"	2209	£1.13
2.5 x 3.75"	2202	£0.39	2.5 x 5"	2210	£0.42
2.5 x 17"	2203	£1.42	2.5 x 3.75"	2211	£0.31
3.75 x 5"	2204	£0.52	3.75 x 17"	2212	£1.51
3.75 x 3.75"	2205	£0.46	3.75 x 5"	2213	£0.57
3.75 x 17"	2206	£1.82	3.75 x 3.75"	2214	£0.42
4.75 x 17.9"	2207	£2.34	2.5 x 1"	2216	£0.52
2.5 x 17"	(pack of five)	2208			

DRILLED PLAIN P.C.B.

.1 Pitch			.15 Pitch		
Size	No.	Price	Size	No.	Price
3.75 x 17"	2217	£1.18	2.5 x 17"	2221	£0.78
3.75 x 2.5"	2218	£0.26	3.75 x 17"	2222	£1.00
5 x 3.75"	2219	£0.42	2.5 x 5"	2223	£0.26
			2.5 x 3.75"	2224	£0.21
			5 x 3.75"	2225	£0.36

VERO SPOT-FACE CUTTER

Order No 2231 Price £0.68

PLUGS AND SOCKETS

PLUGS

No.	Price
P1	DIN LSP 2-pin speaker 1689 £0.08
P2	DIN 3-pin 1690 £0.12
P3	DIN 5-pin 1691 £0.16
P4	DIN 5-pin 180 1692 £0.14
P5	DIN 5-pin 240 1693 £0.15
P6	DIN 6-pin 1694 £0.20
P7	DIN 7-pin 1695 £0.20
P8	Jack 2.5mm screened 1696 £0.12
P9	Jack Plug 3.5mm plastic 1697 £0.10
P10	Jack Plug 3.5mm screened 1698 £0.15
P11	Jack Plug mono plastic 1699 £0.14
P12	Jack Plug mono screened 16100 £0.28
P13	Jack Plug stereo screened 16101 £0.32
P14	Phono 16102 £0.10
P15	Car aerial 16103 £0.16
P16	Coax free TV 16104 £0.16
P17	Right angle jack 16105 £0.12
P18	Jack 2.5mm plastic 16106 £0.12
P19	Jack stereo plastic 16107 £0.20
P20	Phono free screened 16108 £0.14
P21	D.C. 2 pin plug 16109 £0.12
P22	D.C. 2.5mm plug 16110 £0.12
P23	2 pin AC plug U.S. type 16111 £0.18
P24	AM aerial 16112 £0.13
P25	Cassette mains plug 16113 £0.15
P26	FM 300 ohms plug 16114 £0.13

CHASSIS SOCKETS

No.	Price
CS1	DIN 3.5 pin loudspeaker 1652 £0.08
CS2	DIN 3 pin 1653 £0.10
CS3	DIN 5 pin 1654 £0.10
CS4	DIN 5 pin 240 1655 £0.12
CS5	Jack 6mm 1656 £0.06
CS6	Jack 4.5mm 1657 £0.06
CS7	Jack Mono switched 1658 £0.15
CS8	Jack Stereo switched 1659 £0.18
CS9	Phono single 1660 £0.08
CS10	Phono double 1661 £0.10
CS11	Coax surface 1662 £0.21
CS12	Coax 1663 £0.25
CS13	Jack switched Mono 1664 £0.20
CS14	Jack socket DPDT switch 1665 £0.32
CS15	Car aerial 1666 £0.10
CS16	AC mains US type 1667 £0.16
CS17	Phono 4 way 1668 £0.16
CS18	D.C. power 1669 £0.18
CS19	AC switched 1670 £0.32
CS20	Phono 8 way 1671 £0.32

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Manufacturer's Fall Outs which include Functional and part Functional Units. These are classed as out-of-spec from the maker's very rigid specifications, but are ideal for learning about IC's and experimental work

U721 30 ASSORTED LINEAR TYPES

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5 IC's 76110 Eqv to MC1310P-MA767 Data supplied with pak
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U76A AUDIO POWER OUTPUT AMPLIFIERS

8 Assorted types SL403 76013 76003 Etc Data supplied with pak
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etc Order No 16224 £1.20

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7470-72 73-74-76-104-109 Etc
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74M 30 MSI Assorted Types 7441-47-90-154

Etc Order No 16226 £1.20

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C280 CAPACITOR PAK

BI-PAK

High quality modules for stereo, mono and other audio equipment.



NEW

PUSH-BUTTON STEREO FM TUNER

OUR PRICE ONLY
£20.45

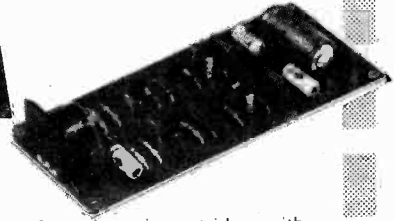
Fitted with Phase Lock-loop Decoder

The 450 Tuner provides instant program selection at the touch of a button ensuring accurate tuning of 4 pre-selected stations, any of which may be altered as often as you choose, by simply changing the settings of the pre-set controls. Used with your existing audio equipment or with the BI-KITS STEREO 30 or the MK60 Kit etc. Alternatively the PS12 can be used if no suitable supply is available, together with the Transformer T538. The S450 is supplied fully built, tested and aligned. The unit is easily installed using the simple instructions supplied.

- ★ FET Input Stage
- ★ VARI-CAP diode tuning
- ★ Switched AFC
- ★ Multi turn pre-sets
- ★ LED Stereo Indicator

Typical Specification:
Sensitivity 3µ volts
Stereo separation 30db
Supply required 20-30v at 90 Ma max.

MPA 30



Enjoy the quality of a magnetic cartridge with your existing ceramic equipment using the new M.P.A. 30, a high quality pre-amplifier enabling magnetic cartridges to be used where facilities exist for the use of ceramic cartridges only. It is provided with a standard DIN input socket for ease of connection. Full instructions supplied.

£2.85

STEREO PRE-AMPLIFIER



PA 100 OUR PRICE £13.75

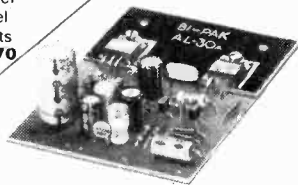
A top quality stereo pre-amplifier and tone control unit. The six push-button selector switch provides a choice of inputs together with two really effective filters for high and low frequencies, plus tape output.

Frequency Response + 1dB 20Hz 20KHz Sensitivity of inputs
1. Tape Input 100mV into 100K ohms
2. Radio Tuner 100mV into 100K ohms
3. Magnetic P.U. 3mV into 50K ohms
P.U. Input equalises to R1AA curve with 1dB from 20Hz to 20KHz
Supply — 20-35V at 20mA.

Dimensions —
299mm x 89mm x 35mm

MK. 60 AUDIO KIT: Comprising 2 x AL60's, 1 x SPM80, 1 x BTM80, 1 x PA100, 1 front panel and knobs. 1 Kit of parts to include on/off switch, neon indicator, stereo headphone sockets plus instruction booklet. **COMPLETE PRICE £29.55** plus 85p postage.

TEAK 60 AUDIO KIT: Comprising: Teak veneered cabinet size 16 3/4" x 11 1/2" x 3 3/4", other parts include aluminium chassis, heatsink and front panel bracket plus back panel and appropriate sockets etc. **KIT PRICE £10.70** plus 85p postage.



NEW AL30A 10w R.M.S. AUDIO AMPLIFIER MODULE

The AL30A is a high quality audio amplifier module replacing our AL20 & 30. The versatility of its design makes it ideal for record players, tape recorders, stereo amps, cassette and cartridge players. A power supply is available comprising a PS12 together with a transformer T538, also for stereo, the pre-amp PA12.

SPECIFICATION:

- Output Power 10w R.M.S.
- Load Impedance 8 to 6ohms.
- Sensitivity 90mv for full output.
- Frequency Response 60Hz to 25KHz ± 2db.
- Supply 22 to 32 volts.
- Input Impedance 50K.
- Total Harmonic Distortion Less than .5% (Typically .3%).
- Max. Heat Sink Temp 80° c.

● Dimensions 90 x 64 x 27mm

ONLY £3.60

**VAT
ADD
12 1/2%**

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STEREO 30 COMPLETE AUDIO

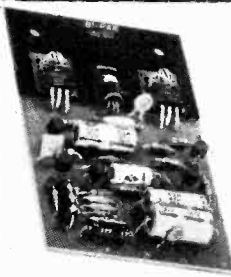
7+7 WATTS
R.M.S.



£16.25

The Stereo 30 comprises a complete stereo pre-amplifier, power amplifiers and power supply. This, with only the addition of a transformer or overwind will produce a high quality audio unit suitable for use with a wide range of inputs i.e. high quality ceramic pick-up, stereo tuner, stereo tape deck etc. Simple to install, capable of producing really first class results, this unit is supplied with full instructions, black front panel knobs, main switch, fuse and fuse holder and universal mounting brackets enabling it to be installed in a record plinth, cabinets of your own construction or the cabinet available. Ideal for the beginner or the advanced constructor who requires Hi-Fi performance with a minimum of installation difficulty (can be installed in 30 mins).

TRANSFORMER £2.45 plus 62p p & p
TEAK CASE £5.25 plus 62p p & p.



AL 60 25 Watts (RMS)

- ★ Max Heat Sink temp 90C.
- ★ Frequency response 20Hz to 100KHz
- ★ Distortion better than 0.1 at 1KHz
- ★ Supply voltage 15-50v
- ★ Thermal Feedback
- ★ Latest Design Improvements
- ★ Load — 3,4,8, or 16 ohms
- ★ Signal to noise ratio 80db
- ★ Overall size 63mm. 105mm. 13mm.

Especially designed to a strict specification. Only the finest components have been used and the latest solid-state circuitry incorporated in this powerful little amplifier which should satisfy the most critical A.F. enthusiast.

£4.35

NEW PA12

Frequency Response 20Hz-20KHz (-3dB). Bass and Treble range, 12dB. Input Impedance 1 meg ohm. Input Sensitivity 300mV. Supply requirements 24V .5mA. Size 152mm x 84mm x 33mm.

NEW PA12 Stereo Pre-Amplifier completely redesigned for use with AL30A Amplifier Modules. Features include on/off volume, Balance, Bass and Treble controls. Complete with tape output.

£6.70

PS12

Power supply for AL30A, PA12, SA450, etc.

£4.35

**OUR PRICE
£1.30**

Input voltage 15-20v A.C. Output voltage 22-30v D.C.
Output current 800 mA Max. Size 60mm x 43mm x 26mm.
Transformer T538 £2.30

Stabilised Power Supply Type SPM80

SPM80 is especially designed to power 2 of the AL60 Amplifiers, up to 15 watts (R.M.S.) per channel simultaneously. With the addition of the Mains Transformer BMT80, the unit will provide outputs of up to 1.5A at 35V. Size: 63mm. 105mm. 30mm. Incorporating short circuit protection.

**Transformer BMT80
£2.60 + 62p postage**

£3.75

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news digest

HANDY METER



Teletonic Altair UK announced the new M22 Digital Multimeter from Data Tech at this year's All Electronics Show in London last month. The M22 has five full functions, 0.1% accuracy, handheld portability, 200 hours battery life, ac or dc current to 20 A.

It measures dc volts from 100mV to 1 kV, ac volts from 100mV to 750V, resistance from 0.1 ohm to 20 megohms, and dc or ac current from 100 nanoamps to 20 A.

The Model 22 can be carried by hand, used on a bench stand or worn like a wrist watch, for two handed probing and big wristed people. The reading hold feature should be used when measuring difficult to reach points. The reflective liquid crystal display gives clear, concise information and is easily readable from wide viewing angles. *Teletonic Altair UK, 2 Castle Hill Terrace, Maidenhead, Berks, SL6 4JR.*

BIRTHDAY PRIZEWINNERS

Well, we said it wasn't easy! Our April competition attracted hundreds and hundreds of entries. Unfortunately, just less than 10% were correct! Having thus insulted nine tenths of our readers, here are the CORRECT answers! A) Skeleton preset pot. B) Soldercon IC sockets. C) Compression trimmer. D) Fuse holder. E) Phono plug. F) Fuse holder cap and fuse. G) Potted bridge rectifier (R.S.) H) Slider control knob.

The five stalwarts who emerged as winners were: *R.W.Hearn, 10 Speedwell Close, Pakefield, Lowestoft, Suffolk NR3 37DU. T.Davey, 2 Northernhay Square, Exeter, Devon, EX4 3ES. A.P.Thomas, 40 Rowley Road, Tottenham, London, N15. A.R.Stickland, 18 Rupert Road, Newbury, Berks. R.P.Mildren, 95 Elm Close, Bristol BS12 6RS.* To all those who entered and failed - thank you for entering, better luck next time.

VOLUNTEERS NEEDED

Deaf-Fax is a research group making and distributing Tele-text decoders to the deaf and the hard of hearing to enable them to pick up Ceefax and Oracle. The units are rented out to the people for a nominal sum. Deaf-Fax have applied for charitable status, and make not a bent penny from the exercise.

However, someone has to build the units in the first place, and that's where you come in. If you're feeling like helping someone, and are in need of something challenging to put together, contact the Secretary Mrs Sheena Carter at 99 Wantage Road, Wallingford, London.

GRAPHICALLY CHEAP

Compelec Electronics have available a circuit board called the VTI which provides all the electronics to turn a television receiver or monitor into a VDU. In addition, the card provides an 8-bit parallel port for interface to a standard ASCII keyboard.

The card is designed to plug into any microcomputer system employing the S-100 bus structure. To the microcomputer the board 'looks like' 512 (1 024 as an option) memory locations. Data can be stored in or retrieved from this block of display memory by the microcomputer just like standard RAM. The character generator ROM contains the full ASCII upper and lower case alphabet, the Greek alphabet, a set of graphics characters and various symbols and punctuation marks.

In kit form the VTI with 1 024 bytes of RAM, providing a display of 16 lines of 64 characters, costs £210 plus VAT from *Compelec Electronics, 310 Kilburn High Road, London NW6.*

DEE TO EH?

Precision Monolithics have introduced a new series of 2-digit BCD monolithic multiplying D/A converters with fast settling time, complementary high compliance current outputs and universal logic inputs. Direct interfacing with CMOS, TTL, DTL, ECL and HTL is provided by an optional logic threshold adjustment. The DAC-20 is the first BCDDAC to interface directly with the high impedance NMOS and PMOS outputs of microprocessor RAMs. *Bourns (Trimpot) Limited, Hodford House, 17/27 High Street, Hounslow, Middx.*

I'M SORRY I'LL PRINT THAT AGAIN!

System 68, May 77:- Link between common pin of Reg 3 and earth omitted on PCB layout.

SINTEL-ATING NEWS!

A rather special ad appears on page 74 this month. Whilst not an actual ETI Reader Offer, it will be of great interest to most readers. Sintel have chosen ETI through which to make a whole list of specially priced components available. You won't find these in any other electronics magazine should you be foolish enough to read any!

CRAFTY VIDEO?

Videocraft have released a Teletext decoder that costs less than half that of comparable decoders presently available. It is supplied as a complete kit comprising an (assembled and fully tested) Texas Tifax Teletext decoding module, power supply, interface module (supplied as a kit), an assembled and tested cable connected remote control, and complete instructions for installation in most (common) television receivers. Due to the compact nature of the Tifax module and interface module, installation within most receiver cabinets is no problem.

Facilities include seven colours, upper and lower case characters, graphics, time coded display and newflash and subtitle inserts in the TV picture. To enable them to supply the correct interface module and instructions, they must know your television set make, model and, if possible, chassis type. The kit costs £180 + 8% VAT. Videocraft, Assetts House, Elverton Street, London SW1P 2QR.

I.T.T. v. THE REST

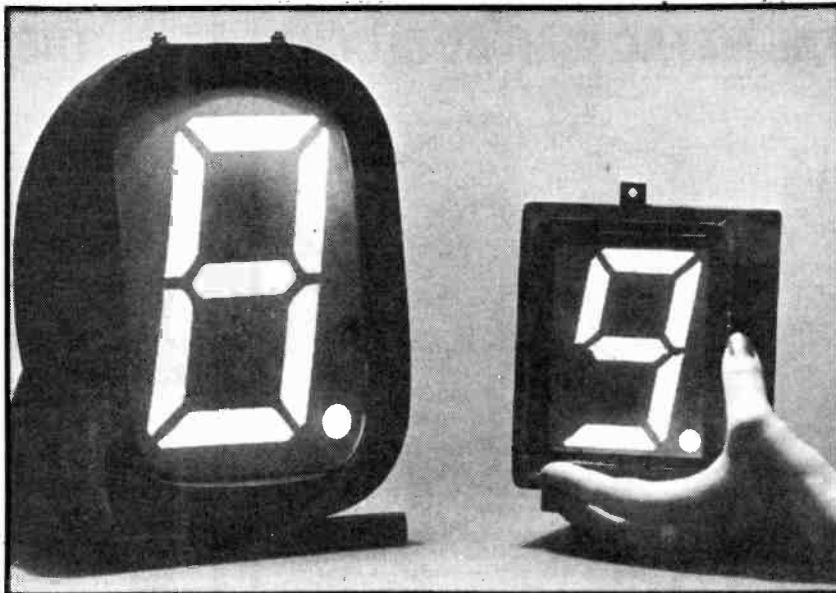
It may interest you to know that ordering a 74xx Series TTL chip is not as simple as it might at first seem. It has come to our notice that certain numbers in this series can refer to chips with completely different logic functions (nasty).

Although for most of the range it does not matter from which manufacturer the devices come, the table below shows that for a few chips I.T.T. have gone it alone in the allocation of numbers:

I.T.T.	
74 135	Quad 2 I/P nand Schmitt.
74 138	Quad 2 I/P OR power driver (30 V - 100 m/A).
74 139	Quad 2 I/P OR power driver (15 V - 100 m/A).

THE REST	
74 135	Quad EX OR/NOR gate.
74 138	3 - to - 8 line decoder/demultiplexer.
74 139	Dual 2 - to - 4 line decoder/demultiplexer.

WHAT A CHARACTER!



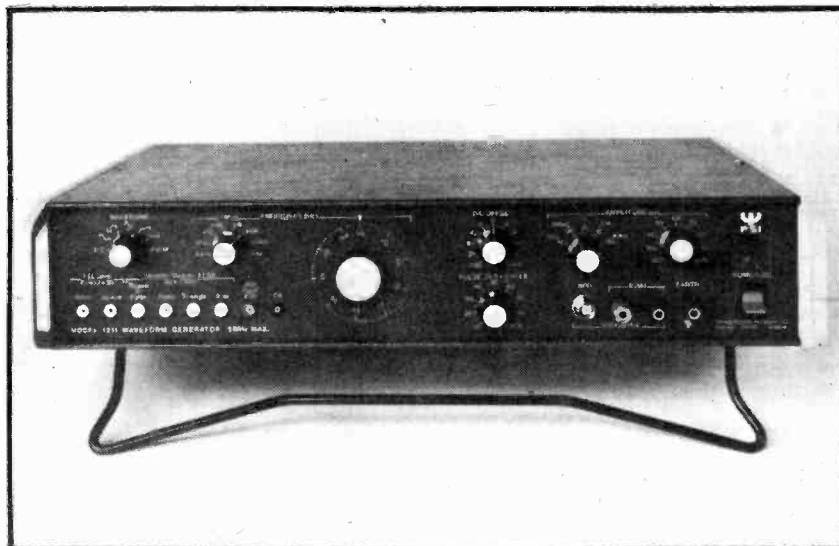
The English Electric Valve Company has developed a high brightness character display tube which operates on a different principle to the more usual stuff. The display can be manufactured in sizes up to about 25 inches diagonal and can display special characters.

The tube operates in a manner similar to a cathode ray tube; a front faceplate carries a high luminance, high-efficiency phosphor energised by a flood beam of electrons from the

cathode. Placed between the cathode and the faceplate is a seven-segment mask with a lead from each of the segments brought out separately so they can be switched at high speeds using a control voltage of 5 V at 10 MΩ input resistance.

The supply voltage is 12 V dc. The power consumption is 2 W total for the 4-inch (E727) and 3 W total for the 8-inch tube (E728). *EEV, Chelmsford, Essex, CM1 2QU.*

GENERATES FUNCTIONING BRITAINS?

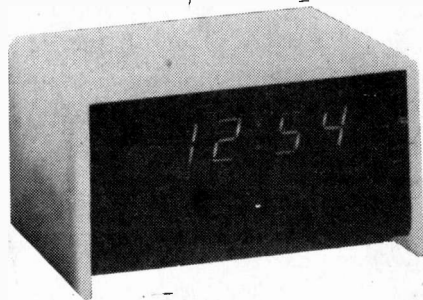


The Procser Scientific Instruments 1211 Function Generator. Sine, square, triangle, ±ramp, and variable duty cycle pulses. Frequency range 0.04 Hz to 5 MHz. And most important it's 100% British! A versatile unit with full sweep facility, and all the other bits and bobs these things are supposed to have. It is designed to compete with the likes of the H.P. 3310A, but at a cost of £425 is considerably cheaper. PSI Ltd., Lady Lane Industrial Estate, Hadleigh, Ipswich, IP7 6DQ.

Metac SPRING BARGAINS

THE METAC DIGITAL CLOCKS

★ COMPLETE KIT ★



- Pleasant green display · 12/24 Hour readout
- Silent Synchronous Accuracy · Fully electronic
- Pulsating colon · Push-button setting
- Building time 1 Hr. · Attractive acrylic case
- Easy-to-follow instructions · Size 10.5 x 5.7 x 8 cm
- Ready drilled PCB to accept components

A professional product for the home constructor. It has been designed by engineers using the most modern techniques and components. It will appeal both to the confirmed hobbyist and to the man who simply wants to 'have a go'. The kit contains everything except a mains lead. The only tools required are a small soldering iron, solder, screwdriver and wire cutters.

KIT PRICE £9.60
+ 76p VAT P&P inc.

ALARM CLOCK KIT

As above plus:

- ★ 24-Hour Alarm
- ★ 10 Minute Repeater
- ★ Mercury Tilt Switch

KIT PRICE £11.60
+ 92p VAT

BRITAIN'S TOP SELLING DIGITAL

2-YEAR GUARANTEE **ALARM CLOCK** **THOUSANDS SOLD**
★ Fully Built, Tested and Guaranteed ★
WITH SLEEP-OVER FEATURE



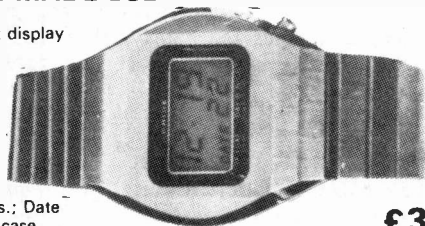
- Bright Orange 12 Hour Display or Soft Green 24 Hour Display Please state choice
- 24 Hour Alarm
- 10 Minute Repetition
- Alarm Set Indicator
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- British Designed and Built
- 5 in. across, 3 1/2 ins. deep

COMPLETE CLOCK
£13.43
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THE GRUEN EXECUTIVE AMERICAN MADE LCD

Continuous 6 digit display with backlight

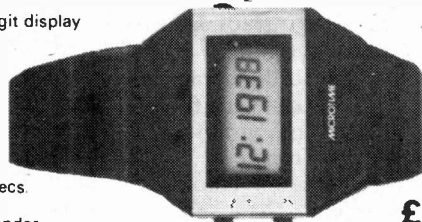


Hours; Mins.; Secs.; Date
Solid heavy metal case
A superb man's watch

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THE BLACK WATCH by BULER SWISS PRECISION MADE LCD

Continuous 6 digit display with backlight

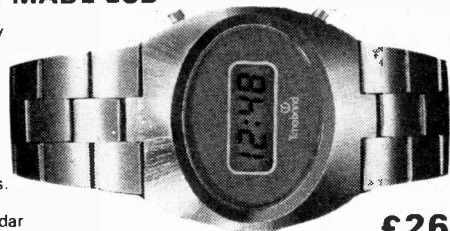


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Month; Date
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Glass-filled epoxy black case

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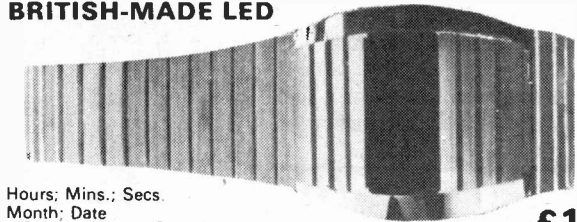
Continuous display with backlight



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Month; Date
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Superb metal bracelet

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THE SILVER BAND by TIME-MICRO BRITISH-MADE LED



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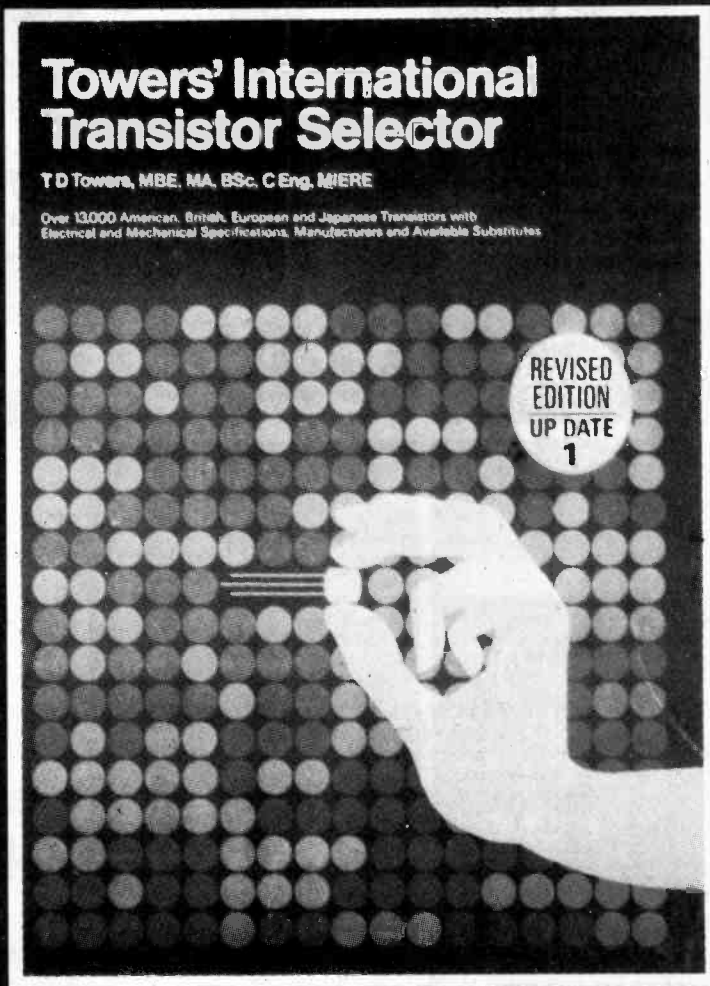


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UPDATE 1 increases the coverage of transistors by about 30% on the first edition.

European Proelectron Standard devices increase by 1500 and now ensure a very comprehensive coverage of Philips, Siemens, Telefunken and Texas Instruments in Europe.

The Japanese 2S Standard Devices coverage has been increased from 1,850 to 3,000 and a further 800 American 2N devices have been selected.

The original 10,000 entries have been fully edited and this new production has been clearly printed from new computer setting.

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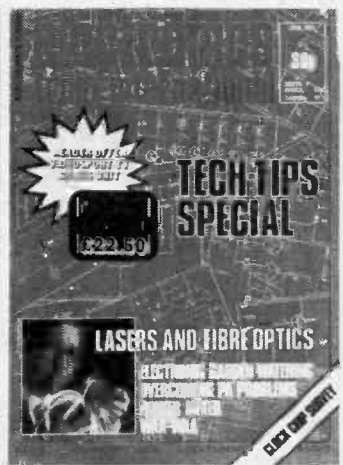
NAME _____

ADDRESS _____

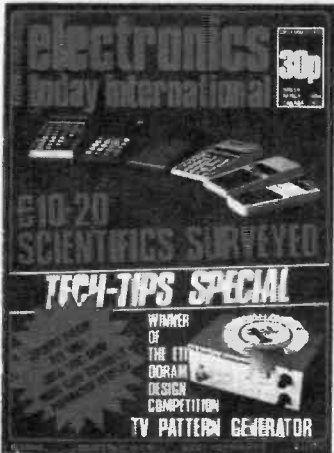
ET14



Gaps?



Gaps?



Gaps?



Gaps?



It can be a nuisance can't it, going from newsagent to newsagent? "Sorry squire, don't have it - next one should be out soon."

Although ETI is monthly, it's very rare to find it available after the first week. If it is available, the newsagent's going to be sure to cut his order for the next issue - but we're glad to say it doesn't happen very often.

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LEARN TO REDUCE TENSION WITH ETI'S

GSR MONITOR

This galvanic skin response monitor provides a means of measuring the minute variations in skin resistance which research has linked to the emotional state of a person

THE BEST WAY to start experimenting with biofeedback is to use a galvanic skin response monitor, a device which measures changes in skin resistance. In March 1977, we published an article which covered the background and theory of biofeedback and we discussed the various types of biofeedback instruments which are available. The GSR monitor is the most simple to use, the electrodes can be simply attached to the fingers with straps and the technique of using the machine can be quickly learned.

Skin resistance changes with changes of emotional state. When tension increases, the skin resistance falls — when tension decreases there is an increase in skin resistance. (Some biofeedback instruction manuals speak in terms of conductivity rather than resistance and state measurements in mhos, and the meter we use gives a positive deflection for decreasing resistance.)

The connection between skin resistance and tension is not fully understood. Tension affects sweat glands and with the changes in the sweat glands there is a change in the membrane permeability of the skin and this change in permeability is the major cause of changes in electrical activity.

Almost a century ago, a scientist named M. Ch. Fere discovered the resistance of the skin to a small electric current changed in response to aroused emotions. This information has since been used in various ways; one obvious example is the polygraph, or lie detector, which responds to the tension generated when a person is lying.

It was not until 1961 that Dr. J. Kamiya, whilst conducting a series of



GSR MONITOR

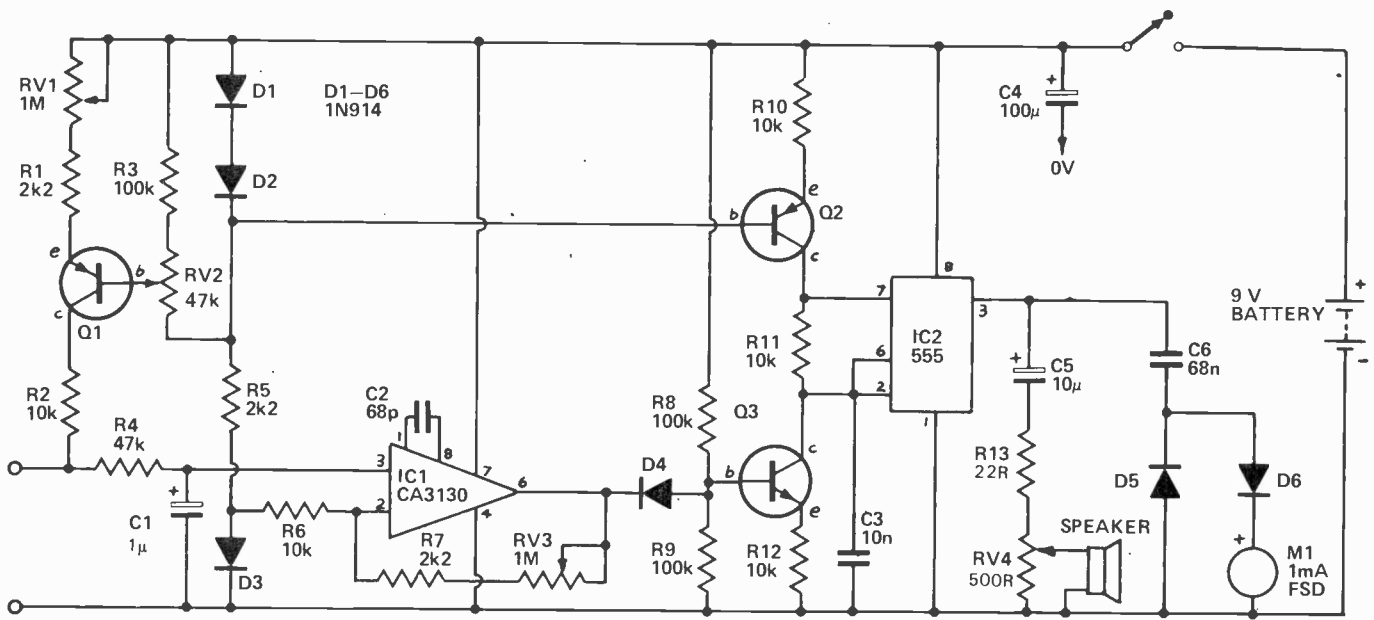
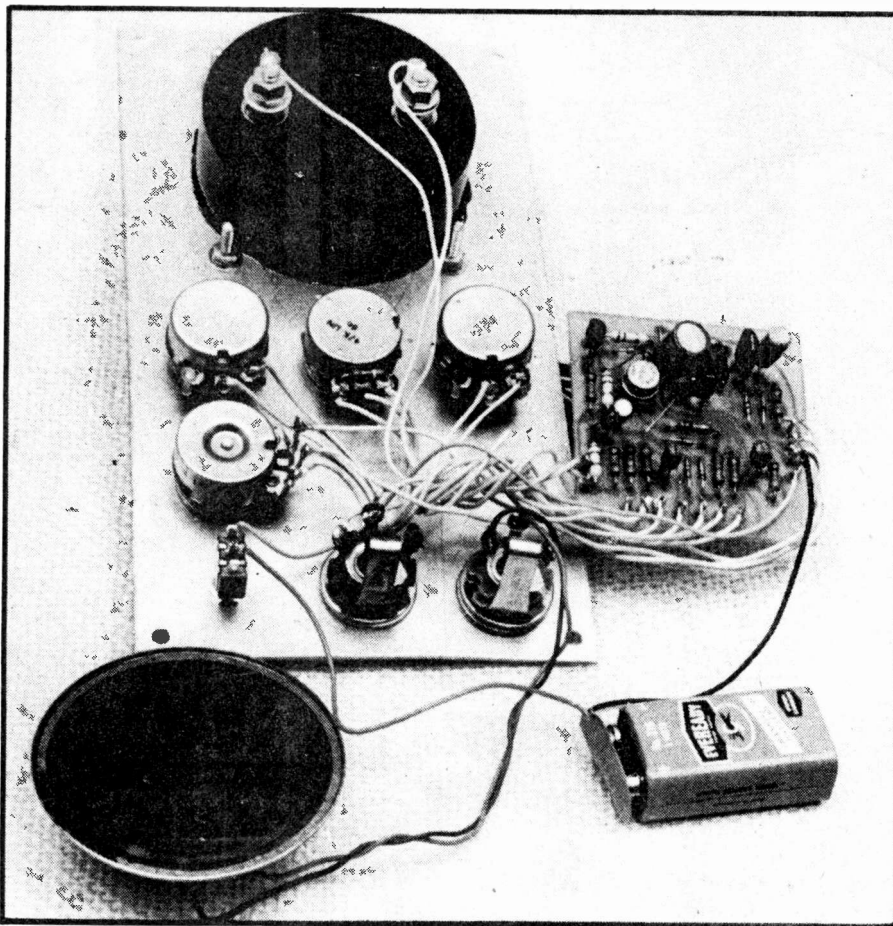


Fig. 1. Circuit diagram of the GSR monitor.

WARNING: THE GSR MONITOR SHOULD NOT BE USED WITH A MAINS POWER SUPPLY.



The picture above shows the internal layout of the GSR monitor. The wiring from the PCB to the front panel, loudspeaker and meter is clearly shown.

HOW IT WORKS

Transistor Q1 acts as a constant current source - the actual value can be varied over a large range by RV1 and over a limited range by RV2. These act as the coarse and fine level controls. This current is passed via R2 to the probes. The voltage developed across the probes is proportional to the skin resistance and is fed to the input of IC1. This amplifies the signal with reference to 0.6 V (drop across D3) and the gain is variable by RV3.

The second IC is an NE555 oscillator where Q2 provides a constant current (about 60 μ A) to the capacitor C3. When the voltage on C3 reaches 6 V the IC detects this and shorts pin 7 to ground, discharging C3 via R11. This continues until the voltage reaches 3 V at which point the short on pin 7 is released, allowing C3 to recharge. The output of the oscillator is connected to a speaker via the volume potentiometer RV4 and the meter via C6 and the diodes D5 - 6. The meter operates in reverse sense to usual, a low resistance gives full scale (or high tone) and high resistance gives zero (or low tone).

We vary the frequency of the oscillator and the meter reading by robbing some of the current supplied by Q2 into Q3. In this way the frequency can be lowered and actually stopped. Transistor Q3 is controlled by IC1 completing the connection between the probes and the output.

Getting hold of components

Nothing here to trouble the constructor

Cost of construction

Should be about £5.00 excluding boxing and meter.

experiments with brain waves, found that with feedback his subjects developed the ability to produce 'Alpha waves' at will.

Dr. Kamiya's experiments created considerable interest and started investigations into whether other bodily functions could be brought under conscious control. Since that time it has been demonstrated that with feedback it is possible for people to control heart beat, blood pressure and temperature — all previously considered to be automatic bodily functions mostly beyond conscious control.

Of course it should be stated that various mystics and yogis have previously demonstrated this type of ability but the fascination of biofeedback is the speed and ease with which this type of control can be learned.

Biofeedback has exciting medical possibilities. GSR machines are being used by therapists for the treatment of many disorders related to tension. The average person will find a GSR machine mainly useful for relaxation training. With the GSR machine it is possible to recognise tension and learn how to decrease tension levels. This type of training is so effective that the machine quickly becomes unnecessary.

However not everyone suffers from tension. The biofeedback machine can be a fascinating toy to play with.

Discovering that you can bring an internal bodily function under conscious control with the same ease that you can twitch your nose is most interesting. And of course you can then perfect this ability just as you perfect your ability at a game like tennis. For many people this is reason enough to build this machine.

What you do with it once you have built it

The ETI GSR monitor has an on/off switch, a sensitivity control and fine and coarse level controls. The machine also has a connection for headphones.

To start relaxation training, you'll need a comfortable chair, low lighting and no distractions. Taking any type of drug can interfere with your ability to relax. This applies to alcohol and cigarettes. Attach the electrodes to the fleshy part of the first two fingers on one hand — firm but not too tight (the non-dominant hand is recommended). Set the sensitivity control to minimum and the 'fine' level control to mid-range. Turn the volume control to minimum. Now you have to set the level with the

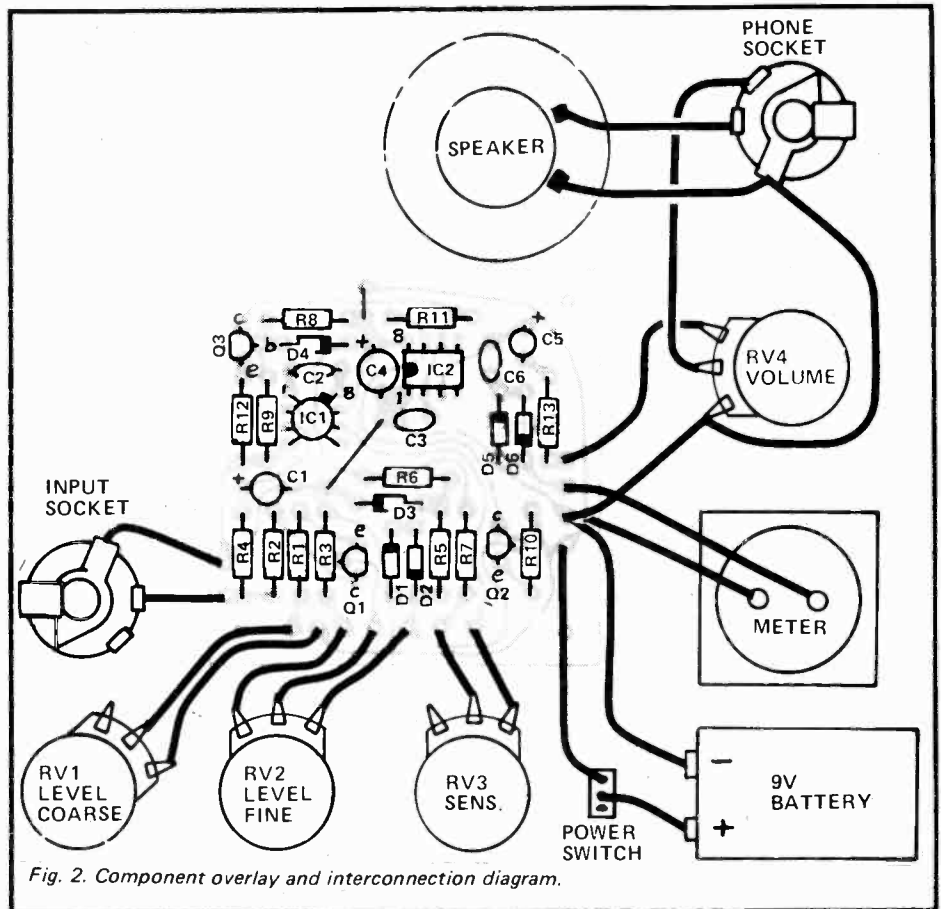


Fig. 2. Component overlay and interconnection diagram.

PARTS LIST

RESISTORS (all ½ W 5%)

R1 2k2
R2 10 k
R3 100 k
R4 47 k
R5 2k2
R6 10 k
R7 2k2
R8,9 100 k
R10-12 10 k
R13 22 R

POTENTIOMETERS

RV1 1 M log
RV2 47 k lin
RV3 1 M log
RV4 500 R lin

CAPACITORS

C1 1 u 16 V electrolytic
C2 68 p ceramic

C3 10 n polyester
C4 100 u 16 V electrolytic
C5 10 u 16 V electrolytic
C6 68 n polyester

SEMICONDUCTORS

D1-6 1N914
Q1,2 BC559 or BC179
Q3 BC549 or BC109
IC1 CA3130
IC2 NE555

MISCELLANEOUS

PC board ETI 546
Meter 1 mA FSD
Box 196 x 113 x 60 mm
Two phone jacks
Four knobs
Small speaker
PP3 or 9 V battery + holder
Pickup probes

'coarse' level control (when the sensitivity is set low the 'fine' level control need not be used). Start with the 'coarse' control at full anticlockwise and turn it up until the meter needle starts to move. Carefully set the needle to mid-range. Now the instrument is set-up in its minimum sensitivity position.

Having mastered setting up with minimum sensitivity try to set the GSR monitor with the sensitivity set halfway. It will require delicate adjustment of the 'coarse' level control. Now the effect of the 'fine' level control can be seen. This control enables you to set the level on a high sensitivity setting.

GSR MONITOR

Although the GSR machine measures minute changes in skin resistance, the level of skin resistance varies considerably from person to person so a wide range of settings is provided.

Now turn up the volume and observe that the meter reading is accompanied by a medium pitched tone. (A convention has developed to link high-pitched tone with tension increase and low pitched tone with a decrease in tension.) Now you relax and bring the tone down and the needle back to zero.

How? Basically you are supposed to find this out for yourself. After watch-

ing the needle for some time you will notice it move up or down. Something has happened to cause a change in your skin resistance. You would be barely aware of what had caused the change but aware enough to try to reproduce the effect. Eventually your awareness grows and so does your ability to control your tension. Many people find that relaxation of the stomach muscles makes the difference. It varies from person to person.

There are several relaxation techniques which work very well. One method is to tense all the muscles of the body as hard as possible, hold them tense for several seconds then very deliberately relax all muscles. There are several books and cassettes available which describe relaxation techniques. The techniques work. The biofeedback machine makes it possible to monitor progress.

As you relax, the needle on the meter and the audible tone will decrease. When the needle reaches zero, reset it again towards the fsd end of the scale and repeat the procedure.

Twenty minutes is the recommended time for a training session. After about one or two weeks of daily relaxation training, it should be possible to pro-

duce the same level of relaxation without using the machine and the machine can simply be used occasionally as a reference.

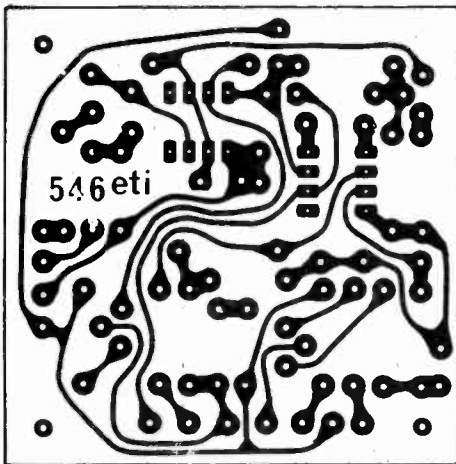
Construction

Construction is not critical although we recommend you use the pc board as it makes things easier. Before soldering the components made sure they are orientated correctly. External wiring can be done with the aid of the overlay-wiring diagram.

Probes

Probe construction and electrical contact is not nearly as critical as with most other biofeedback machines.

Commercial GSR machines use a pad of soft steel wool which is held firmly onto the finger by a short length of Velcro strap (Band-Aids work fine!). However, any method ensuring a firm contact between probe leads and the fleshy part of the finger will do. One method which works very well is to bind tinned copper wire around a guitar finger pick (or solder to a steel pick). Two probe connections are of course required — one for each of the first two fingers.



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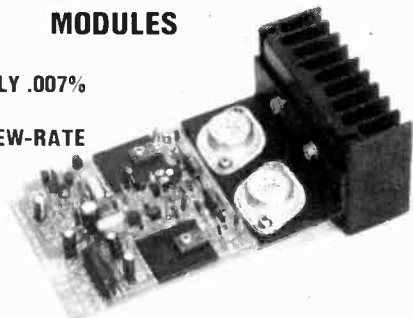
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The

Sinclair

Cambridge

Programmable



PROGRAMMABLE CALCULATORS have always had their market wherever a task must be repeated over and over with different data. Examples in electronics would include filter design - optimising component values for given allowed parameter variations, calculation of resistor networks - for attenuators or to achieve a given R using paralleled units and perhaps even Fourier Analysis.

There are two types of calculator with this ability to remember a sequence of keystrokes, card programmables - epitomised by the HP67 reviewed in ETI last month, and keyboard programmables, of which the CBM PR 100 is a good example. Price has, up until now, been a fairly high hurdle between the mass of potential users and the machines.

Until now.

The Sinclair Cambridge Programmable could well be the stepladder to solve that problem! Selling at around £15 discount including mains unit, its 36 step program memory becomes accessible to even the most impercunious student.

Stepping into the limelight

Housed in the same case - almost - as its predecessor, the Programmable is outwardly little different. Only the 'GT style' bulge on the battery compartment (to accept a PP3) and the actual word 'programmable' coyly added to the front panel artwork attest to the vital difference!

As a normal scientific, the new Cambridge has more functions on the

keys than its Daddy.. Nearly all the buttons are three-function in fact, making a double shift key necessary. The advantage of the small case size wages a fierce war on keyboard convenience and display visibility, which means the Cambridge is fiddly to use in this generation as the last.

But this is not a normal scientific - far from it. This is a 36 step programmable calculator with conditional branching and step facilities, altogether a different animal.

The nature of the beast

Along with the machine comes a desk stand, instruction manual and sample program booklet. Full back-up is provided by Sinclair, in the form of a 4 volume program library containing almost 300 programs. This is available at £4.95 complete.

Sinclair's library gives the basics of the calculation required, and then lists the program in a three column table, giving step number, keystroke and 'check code'.

Scattered within the library pages are blank program sheets to enter your own creations for storage. We have a suggestion to make to Sinclair about that!

It seems to us that large number of

potential users will be putting the machine to work 'in the field' as it were, and these people will like as not be unwilling or unable to hump around a book the size of a program library volume for reference to program listings. Why not a book of blank program sheets the same size as the calculator which could then fit within the carrying case supplied?

Within the case are just three components, MPU chip, display driver and a capacitor! Both chips are from National Semi, as is the display. The MPU is the 4-bit MM5799. The on chip memory of this MPU has made it possible for Sinclair to hold the price as low as they have. The mask program was written in America, where Sinclair concentrate their software expertise. According to them, they only just made it all go into the space available, which means that they don't give anyone else a prayer of getting near the price for a long time.

It will be interesting to see how soon anybody does begin to compete or indeed if anyone bothers!

Using a Mallory PP3 gives around 30 hours use, less if you keep the machine in program mode. With that desk stand under its body, the Cambridge is not at all bad for use

THIS NEW MPU CALCULATOR IS AMAZINGLY LOW PRICED BUT HOW GOOD IS IT? RON HARRIS INVESTIGATES

Cambridge Programmable

as a desk machine, although naturally its size doesn't help any!

Putting it in to get it out

Using the Cambridge as a programmable machine takes perhaps an hour or so to get used to, aided by the excellent manual and examples provided. Simply the best method to learn to program this calculator is to play with it! Once familiar with its little ways and means, (i.e. the four key sequence to get into program mode, and a further two to get it to listen to what you're telling it!) entering sequences is no problem at all.

We doubt that anyone short of Dr Who could simply pick up the little white box and get it to run programs, but within half an hour all should be well. This is a complex instrument of course, and it would be unreasonable to expect otherwise.

Once entered, a program is checked by 'stepping' through and comparing a list of 'check-codes' with the displayed characters. The codes are written on the keyboard too, just so you can be sure.

The greatest advantage of such an instrument is perhaps the fact that once programmed up, operation is by use of a single 'RUN' key. This means that untrained operators can use it to good advantage. As an example we did out a short program to calculate an invoice amount, in which a certain number of magazines are sold at a discount of say 30%.

The calculations needed for each bill are thus: (nos. of magazines) x (price), 30% of this and then total less 30%. On a normal calculator this could involve the following sequence.

(enter number) x 35 = (total amount) x 30 / 100 = (discount amount) and then (total amount) - (discount amount) = invoice total.

This would be shortened with memory or percentage key functions we realise, and 30% could be calculated as 0.3 more easily - but bear with us we're only trying to make things clear!

In good voice

Now if only one invoice is to be made up, then no advantage is gained by use of the program. However

most businesses use this kind of invoice procedure many times in any given period, and once the number exceeds three the program will save the time it takes to enter it.

Our programme is listed out below, with the check codes next to it. Each 'step' is simply the button to operate at that point in the sequence.

To use this listing to work out invoices for you you would simply go through the sequence, check it, and then clear the display.

Enter the number of magazines to be sold on the first invoice. Press RUN. Displayed now is the first total to be typed in. Press RUN again. The display now gives you the 30% discount on the previous total. Type this and then give RUN a third jab. The number now sitting there is the final total.

On 20 such invoices, using the Programmable is shorter by over 600 keystrokes, thus removing 600 possible errors, and making a tedious task far less time consuming.

Game for a few reservations

We found the Sinclair an excellent introduction to programmable calculators. As an ordinary scientific we have reservations, but surely

few people would use it as such. The key spacing is quite tight, and the display does have a limited viewing angle to it.

One beautiful aspect to the business of running programs is the fact that you can get the machine to play games with you! Sinclair give three games programs in the library, including a decent version of moon-landing. The four arithmetic function, keys all do something different if operated twice, and this can be used to advantage within a program. For example the sequence x x will square a number.

If you intend to buy a Cambridge, don't begrudge the £4.95 for the program library, it is worth far more and extends the usefulness of the machine immeasurably.

Go to conclude!

All in all then an amazing presentation for the money, and a machine which will probably create its own market. (We suspect Sinclair themselves will be surprised at the reaction this calculator may illicit).

For our part we can recommend it enthusiastically despite the reservations expressed earlier. It is worth its asking price for the games its capable of playing!

CODE	STEP
.	X
3	#
3	3
5	5
-	=
2	sto
0	stop
5	rcl
.	X
3	#
3	3
0	0
G	÷
3	#
1	1
0	0
0	0
-	=
0	stop
F	+
5	rcl
-	=
0	stop
A	▼
2	go to
0	0
0	0

INVOICE			
N	Magazines at 35p each		
	Less 30% discount		
	Invoice total		

The example program referred to in text is shown on the left, and the type of invoice it works out is illustrated above. The 'CODE' column shows what appears on the display at each step of inserting the program.
Using this check code enables the program listing to be checked before running.

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2N696	0.35	2N3704	0.15	40362	0.48	BC161	0.50	BD135	0.37	8FV89	1.25	CA3020	1.78	LM3301N	0.85	TAA661B	1.32
2N697	0.30	2N3705	0.15	40363	1.20	BC167	0.12	BD136	0.35	8FV90	0.34	CA3020A	2.29	LM3302N	1.40	TAA700	3.91
2N698	0.62	2N3706	0.16	40406	0.58	BC168	0.12	BD137	0.38	8FV51	0.38	CA3028A	1.01	LM3401	0.70	TAA930A	1.00
2N699	0.55	2N3707	0.18	40447	0.45	BC169	0.12	BD138	0.38	8FV52	0.36	CA3028B	1.29	LM3900	0.75	TAA930B	1.05
2N706	0.21	2N3708	0.16	40408	0.65	BC170	0.16	BD139	0.40	8FV53	0.34	CA3030	1.24	LM3905	1.60	TAD100	1.95
2N707	0.12	2N3709	0.18	40409	0.65	BC171	0.14	BD140	0.40	8FV90	1.37	CA3030A	1.88	LM3909	0.68	TBA120	0.65
2N708	0.21	2N3710	0.16	40410	0.65	BC172	0.12	BD239	0.40	8RY39	0.50	CA3045	1.40	MC1035	1.75	TBA400	1.50
2N709	0.50	2N3711	0.18	40411	2.85	BC177	0.20	BD240	0.45	BSK21	0.32	CA3046	0.89	MC1303	1.47	TBA500	2.21
2N718	0.27	2N3712	0.10	40594	0.75	BC178	0.20	BD241	0.45	BU105	1.50	CA3048	2.23	MC1304	1.85	TBA500Q	2.30
2N718A	0.50	2N3713	2.30	40595	0.85	BC178	0.20	BD242	0.45	BU205	2.20	CA3049	1.66	MC1305	1.85	TBA510	2.21
2N720A	0.80	2N3714	2.45	40673	0.45	BC178	0.20	BD243	0.60	ME0402	2.20	CA3052	1.62	MC1306	1.00	TBA510Q	2.30
2N814	0.35	2N3715	2.55	40126	0.37	BC182L	0.14	BD244	0.62	ME0404	0.15	CA3053	0.60	MC1310	1.91	TBA520	2.21
2N916	0.30	2N3716	2.80	AC127	0.44	BC183	0.11	BD245	0.65	ME0404	0.15	CA3080	0.68	MC1312	1.98	TBA520Q	2.30
2N918	0.38	2N3771	1.85	AC128	0.37	BC183L	0.14	BD246	0.66	ME4102	0.10	CA3080A	1.88	MC1327	1.54	TBA530	1.98
2N929	0.25	2N3772	2.00	AC151V	0.35	BC184	0.11	BD246	0.66	ME4104	0.10	CA3086	0.51	MC1330	0.92	TBA530Q	2.07
2N930	0.26	2N3773	2.90	AC152V	0.50	BC184L	0.14	BD246	0.66	MJ480	1.35	CA3088	1.59	MC1350	0.75	TBA540	2.21
2N1132	0.60	2N3789	2.90	AC153	0.49	BC207	0.12	BD520	1.13	MJ481	1.55	CA3089	2.52	MC1351	1.20	TBA540Q	2.30
2N1613	0.35	2N3791	3.10	AC176	0.40	BC208	0.11	BF115	0.38	MJ481	1.55	CA3090	3.80	MC1352	0.97	TBA550	3.13
2N1711	0.37	2N3792	3.50	AC176K	0.60	BC212L	0.14	BF117	0.70	MJ490	1.35	CA3130	0.94	MC1357	1.45	TBA560Q	3.22
2N1893	0.38	2N3794	2.00	AC187K	0.55	BC213	0.17	BF121	0.55	MJ255	1.25	CA3130A	0.65	MC1458	0.91	TBA570	1.29
2N2102	0.60	2N3819	0.36	AC188K	0.55	BC214	0.16	BF123	0.55	MJ255	1.25	LM301A	0.65	NE555	0.53	TBA570Q	1.38
2N2108	0.33	2N3820	0.35	AD161	0.85	BC214L	0.17	BF125	0.25	MJE370	0.58	LM304	2.45	NE555	1.05	TBA641B	2.50
2N2119	0.30	2N3823	0.25	AD162	0.85	BC253	0.22	BF154	0.25	MJE371	0.60	LM307N	0.65	NE566	1.65	TBA651	1.80
2N2219	0.30	2N3904	0.21	AF106	0.55	BC257	0.14	BF159	0.35	MJE520	0.45	LM308C	1.82	NE567	1.80	TBA700	1.52
2N2219A	0.32	2N3906	0.22	AF109	0.75	BC238	0.12	BF160	0.60	MJE295	1.40	LM308N	1.17	SAS560	2.50	TBA700Q	1.61
2N2220	0.35	2N4036	0.67	AF124	0.65	BC239	0.15	BF166	0.40	MJE3055	0.85	LM317K	3.00	SAS570	2.50	TBA720Q	2.30
2N2221	0.22	2N4037	0.55	AF125	0.65	BC253	0.22	BF167	0.38	MP8111	3.35	LM318N	2.25	760101N	1.57	TBA750	1.98
2N2221A	0.26	2N4058	0.20	AF126	0.65	BC257A	0.17	BF173	0.38	MP8112	0.40	LM323K	6.40	76003N	2.55	TBA750Q	2.07
2N2222	0.25	2N4059	0.20	AF127	0.65	BC258A	0.17	BF177	0.30	MP8113	0.45	LM339N	1.75	76008K	2.50	TBA800	1.20
2N2222A	0.25	2N4060	0.20	AF129	0.69	BC258A	0.17	BF178	0.35	MP8113	0.45	LM348N	1.91	76013N	1.70	TBA810	1.16
2N2368	0.25	2N4061	0.17	AF186	0.50	BC259B	0.18	BF178	0.35	MPSA05	0.23	LM348N	1.91	76013ND	1.57	TBA820	1.03
2N2369	0.25	2N4062	0.18	AF200	0.70	BC261A	0.21	BF179	0.35	MPSA06	0.24	LM348N	1.91	76013ND	1.57	TBA920	1.79
2N2369A	0.25	2N4126	0.17	AF239	0.74	BC263C	0.24	BF181	0.40	MPSA12	0.35	LM371N	2.25	76023ND	1.57	TBA920Q	1.99
2N2646	0.75	2N4240	0.85	BC279	0.80	BC300	0.45	BF182	0.45	MPSA15	0.24	LM372N	2.15	76033N	2.55	TBA940	1.62
2N2847	1.40	2N4919	0.65	AF279	0.80	BC301	0.45	BF182	0.45	MPSA15	0.24	LM372N	2.15	761101N	1.46	TCA160C	1.85
2N2904	0.36	2N4920	0.70	AF280	0.85	BC303	0.45	BF184	0.38	MPSU05	0.50	LM374N	2.25	76114N	1.87	TCA160B	2.25
2N2904A	0.37	2N4921	0.50	BC107	0.15	BC307	0.20	BF185	0.35	MPSU06	0.56	LM377N	1.75	76131N	1.30	TCA280A	1.30
2N2905	0.37	2N4922	0.55	BC108	0.15	BC307	0.20	BF185	0.35	MPSU06	0.56	LM377N	1.75	76131N	1.30	TCA290A	3.13
2N2905A	0.38	2N4923	0.70	BC109	0.15	BC309C	0.25	BF194	0.14	MPSU55	0.55	LM379N	3.95	76227N	1.51	TCA290A	1.84
2N2906	0.28	2N5190	0.60	BC113	0.17	BC317	0.14	BF196	0.14	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA370	3.22
2N2906A	0.25	2N5191	0.70	BC115	0.19	BC317	0.14	BF196	0.14	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA740	2.76
2N2907	0.21	2N5192	0.75	BC116	0.19	BC317	0.14	BF196	0.14	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA750	2.30
2N2907A	0.22	2N5195	0.90	BC116A	0.20	BC327	0.20	BF198	0.12	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA760	1.38
2N2924	0.15	2N5245	0.35	BC117	0.22	BC328	0.19	BF200	0.35	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA760	1.38
2N2926	0.13	2N5294	0.40	BC118	0.16	BC338	0.21	BF244	0.35	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA760	1.38
2N3019	0.55	2N5295	0.40	BC119	0.30	BC347	0.12	BF245	0.34	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA760	1.38
2N3053	0.40	2N5296	0.40	BC121	0.45	BC348	0.13	BF245	0.34	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA760	1.38
2N3054	0.60	2N5298	0.40	BC132	0.30	BC348	0.13	BF245	0.34	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA760	1.38
2N3055	0.70	2N5447	0.15	BC134	0.15	BC349	0.12	BF254	0.24	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA760	1.38
2N3390	0.25	2N5448	0.15	BC135	0.15	BC350	1.03	BF254	0.24	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA760	1.38
2N3391	0.25	2N5449	0.19	BC136	0.19	BC351	1.06	BF257	0.37	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA760	1.38
2N3391A	0.25	2N5457	0.32	BC137	0.14	BC352	1.03	BF257	0.37	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA760	1.38
2N3392	0.16	2N5458	0.33	BC140	0.40	BC353	1.20	BF257	0.37	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA760	1.38
2N3393	0.15	2N5459	0.29	BC141	0.45	BC354	1.00	BF257	0.37	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA760	1.38
2N3394	0.15	2N5484	0.34	BC142	0.30	BC358	2.00	BF257	0.37	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA760	1.38
2N3439	0.88	2N5486	0.38	BC143	0.30	BC358	2.00	BF257	0.37	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA760	1.38
2N3440	0.64	2N6027	0.53	BC147	0.12	BC358	2.00	BF257	0.37	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA760	1.38
2N3441	0.85	2N6101	0.65	BC148	0.12	BC358	2.00	BF257	0.37	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA760	1.38
2N3442	1.35	2N6107	0.42	BC149	0.13	BC358	2.00	BF257	0.37	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA760	1.38
2N3638	0.16	2N6109	0.42	BC153	0.27	BC358	2.00	BF257	0.37	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA760	1.38
2N3638A	0.16	2N6121	0.38	BC154	0.27	BC358	2.00	BF257	0.37	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA760	1.38
2N3639	0.30	2N6122	0.41	BC157	0.12	BC358	2.00	BF257	0.37	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA760	1.38
2N3641	0.20	2N6123	0.43	BC158	0.11	BC358	2.00	BF257	0.37	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA760	1.38
2N3702	0.17	2N6126	0.45	BC161	0.50	BC358	2.00	BF257	0.37	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA760	1.38
2N3703	0.15	40361	0.45	BC160	0.50	BC358	2.00	BF257	0.37	MPSU56	0.60	LM380B	0.90	76227N	1.51	TCA760	1.38

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A CONSIDERATION OF THE METHODS USED TO APPLY DIGITAL CIRCUIT

THE WEALTH OF EXPERIENCE possessed by engineers and technicians involved with digital hardware can be enormously valuable to them in their transition to microprocessor based designs. The thought of programming may distress the hardware man; however, to write efficient assembly language programs demands an intimate familiarity with the hardware structure of the particular microcomputer being used.

The flowchart forms the basis of program writing by ensuring that the logical sequence of events has been crystallised. Consider as an example a process control situation depicted in Fig. 1. After a controlled start, system initialisation can commence by processing the input data to check given interlock requirements. Satisfactory results allow the process to begin, otherwise the interlock failure is annunciated and a system stop ordered.

As the process continues towards a designated goal, periodic status checks of the system are required so that control action can be implemented. To ensure that actuators operate correctly the response to an action command is fed back for the system to monitor. This outline scheme typifies the use of flowcharts. Of course each block could be examined further resulting in a more detailed diagram.

Program Power

The real power of programs is their ability to make decisions. Examples of assembly language conditional instructions are:

JUMP and
INCREMENT SKIP IF ZERO

Jump instructions can either be mandatory, thus directing the program to an address which is accessed in all cases, or jumps can be conditional as illustrated by the following examples:

JZ could mean Jump if accumulator=0
JC could mean Jump if accumulator carry=1
JCN could mean Jump if a given CPU pin=1

The number and type of jump instructions provided depends upon the particular microprocessor in use. The Increment Skip if Zero (ISZ) is

useful for decision making because it can distinguish between zero and non-zero contents of index registers. A designated path can be followed in each case.

Interesting Routine

A program sector which is frequently used during a process forms a prime candidate for a SUBROUTINE. In Fig. 1, the block performing

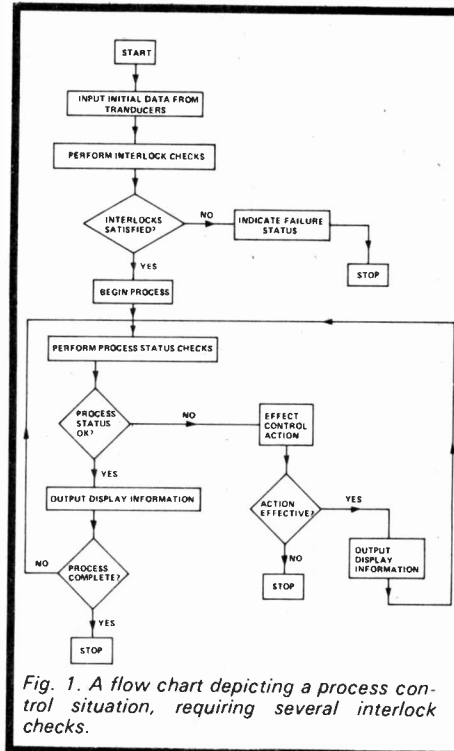


Fig. 1. A flow chart depicting a process control situation, requiring several interlock checks.

process status checks would clearly be used repeatedly. Two of the three paths leading to this block are the result of decisions made during the on-going process. The subroutine would therefore be called into action by an instruction such as JMS 70 (Jump to subroutine at address 70).

To perform process status checks, data has to be input from transducers, stored, and then a number of successive decisions made based upon the data. These requirements can be met by using NESTED SUBROUTINES. A number of subroutines are written each one called by a preceding member of the set. Fig. 2 shows subroutines nested to a depth of three.

After servicing a subroutine the micro processor needs to know the

re-entry point in the program. This information is normally provided by a section of memory store known as a push-down stack. As each further level is entered in a nested subroutine system the latest return address is placed at the top of the stack thus causing previously entered return addresses to be pushed-down.

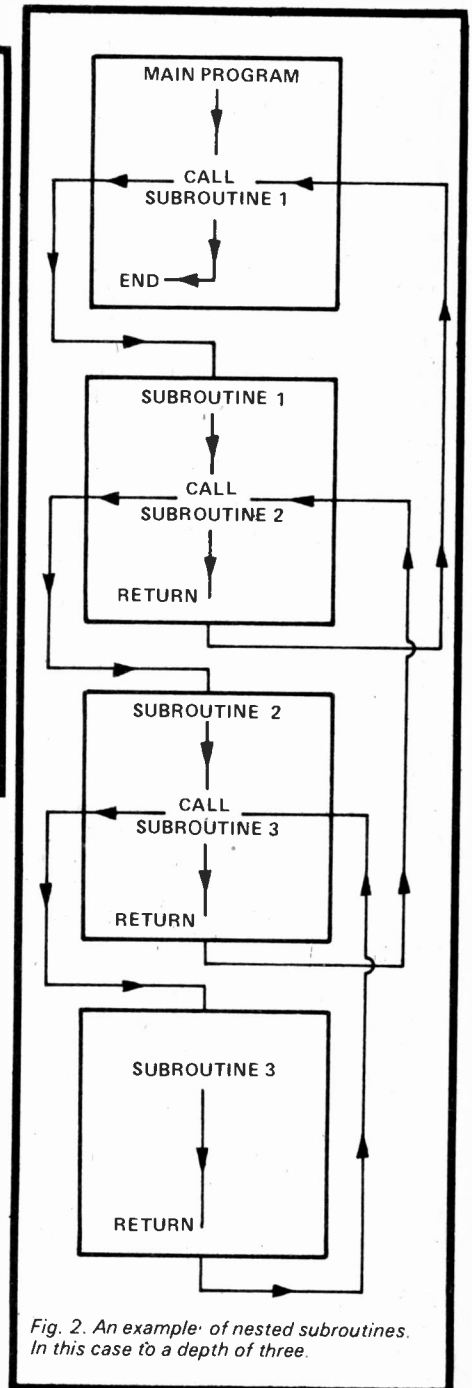


Fig. 2. An example of nested subroutines. In this case to a depth of three.

DESIGN TECHNIQUES TO SOFTWARE ROUTINES BY R. WILSON MSc

Displaying Versatility

A main application area of micro-processor systems is in replacing hard-wired logic by a stored program. A multiplexed seven-segment LED system for numeric display is usually implemented by interconnecting integrated circuits. The function of some of the circuits can be duplicated by a sequence of assembly language instructions.

Flowing Charts

A flowchart for a subroutine which controls the indications of four BCD digits by seven-segment displays is shown in Fig. 3. When this subroutine is called, a four-bit index register which acts as a loop counter is set to 1100 in binary (12 in decimal). By using the ISZ instruction the decision can be programmed simply. The loop counter is incremented each time this instruction is executed. After three loops the index register would contain binary 1111 (15 in decimal). When incremented this becomes 0000 ensuring that after four loops the alternative path is selected by the "skip if zero" part of ISZ.

Routine Subs

The subroutine multiplexes the four displays by successively sending 0001, 0010, 0100 and 1000 as digit drive data during the progression of the program from the first to the fourth loop. The BCD digits represented as segment drive data are thus automatically routed to the correct display from a common highway.

Each BCD digit was stored by four-bits of random access memory (RAM). The read/write facility was essential as variable data was being processed. When programming, for dynamically changing data the locations of the various data sets in RAM must be constantly reviewed. A RAM map, Fig. 4, is a straightforward visual-aid which makes this task easier. A diagram is drawn showing the empty memory locations then as the instructions are written, the space in RAM can be thoughtfully allocated, modified and updated as necessary.

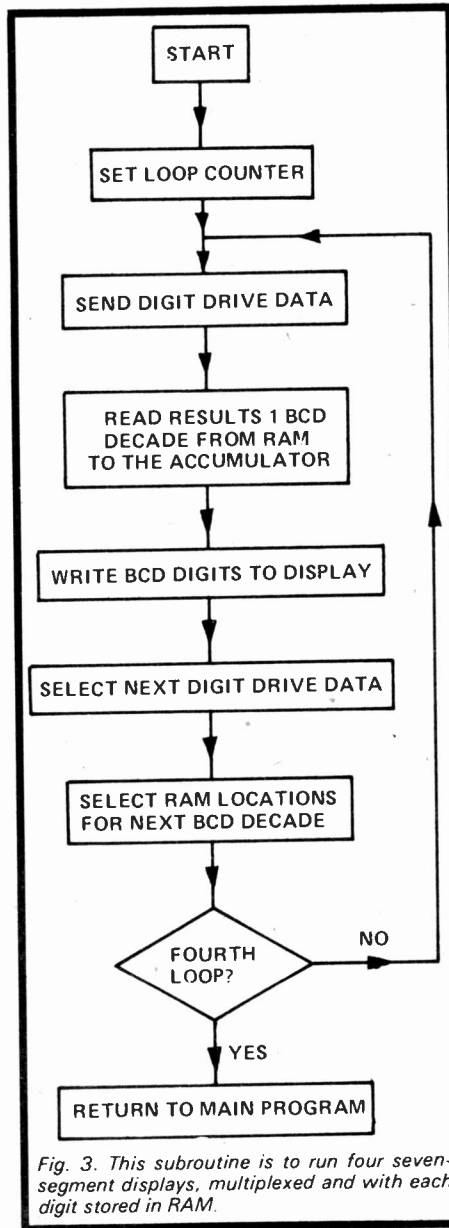


Fig. 3. This subroutine is to run four seven-segment displays, multiplexed and with each digit stored in RAM.

Finding Bugs

Engineers know that the phrase "nothing ever works first time" usually applies to hardwired designs. It also applies to software designs. When a comprehensive program has been written it will contain errors therefore a means of examining the program operation is required. The main sections of a microcomputer hardware structure, the CPU, memory, I/O, and clock can be simulated either by software in the form of a computer package or by

special purpose hardware. Software simulators are available through various commercial time-share networks whereas hardware simulators can be obtained direct from microprocessor distributors.

The source program written in memonics needs to be translated into numeric patterns. The resulting data, called object code, is generated from the source code by using an assembler. Again hardware and software assemblers are available. The object code version of the program can then be presented to the simulator for testing.

Program testing is not a simple task but the following suggestions might be helpful. If possible, assemble the whole program from the outset, or at least assemble substantial segments such as subroutines. This ensures that the actual program is examined rather than a simplified version. By mentally working through the selected program segment using chosen test data the expected outcome can be predicted before beginning a test-run. The RAM map is often useful during this procedure. If a teletype is used to communicate with the simulator during a program test-run a printed record of the test process can be preserved for analysis later. As each section is tested, modified and eventually verified as correct, further segments can be processed. The aim is to commit the whole of the validated program to PROM.

The hardware man can be comforted in his venture into software by remembering that a microcomputer is, after all, an engineer's computer.

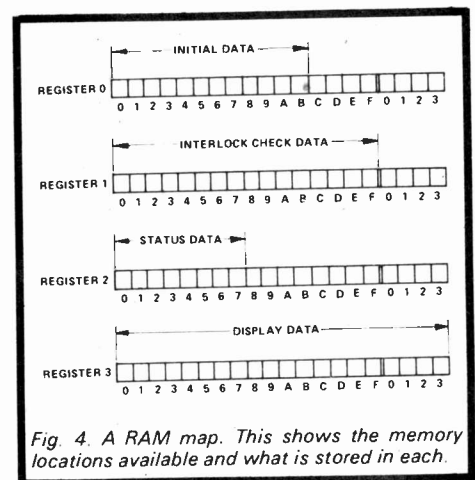


Fig. 4. A RAM map. This shows the memory locations available and what is stored in each.

BE A CATHODE RAY COWBOY WITH OUR

TV GAMES RIFLE

FROM CIRCUIT AND TEXT BY WATFORD ELECTRONICS

THE TV GAMES UNIT featured in our May '77 issue has provision for the addition of two rifle games. In both games a target appears on the TV screen and the object is to 'hit' this with the TV rifle. When the trigger of the rifle is pulled a shot counter is incremented by one and if the rifle is on target the hit counter is incremented, a hit noise produced and the target blanked for a while.

The difference between the two game options is that in one the target moves randomly about the screen and in the second the target traverses the screen from left to right under the control of the manual serve button.

Unlike the other games, the score does not appear on the screen during the game since this might confuse the player, instead the score appears after 15 shots. The score is displayed with the number of shots (i.e. 15) on the left and the number of hits on the right.

Seeing the light

The rifle uses a photodarlington to detect the target on the TV screen but relies on careful construction of the optics involved to ensure adequate sensitivity is obtained from the unit. We also arranged for the rifle to ignore any sources of light other than the target on the TV screen.

Though we called this project a TV rifle game we finally settled on what might more accurately be described as a pistol. The general method of construction used in the pistol is shown in Fig. 1.

Getting started

The butt is made from a fairly hard wood and after being fashioned to the shape shown in

the drawing, the top was dished with a half round file to accommodate the barrel. The next step is to drill a hole vertically through the butt to take the connecting cable between the pistol and games unit. The trigger switch is mounted by drilling two holes, one above the other, and chiselling out the remaining wood to form an oblong hole. A small aluminium plate was then drilled to accept the switch and two small wood screws used to secure the plate to the butt.

Roll out the barrel

The barrel is made from 28mm dia. metal tube 8½in. in length. To fit it to the butt two holes were drilled at one end, and two further holes drilled diametrically opposite the first pair. By passing a screwdriver through the top holes the barrel may be secured to the butt.

The lens used in the pistol had a focal length of 2½in. and came from an old jeweller's eye glass,

this was mounted in a B9A valve screening can. The photodarlington should be carefully positioned at the focal point of the lens. We mounted the photodarlington on a piece of veroboard which enabled us to slide it back and forth until its position was correct.

The assembly may then be mounted in the barrel.

End in sight

The front sight was formed from a 4BA bolt which was filed to provide a sight tip. The rear sight is formed from a ⅜in. wide strip of aluminium about 2½in. long. One end is bent up and a 'V' slot filed with a needle file.

At this stage the p.c.b. may be assembled. The only thing to note about the components used is that they must be as small as possible, we used tantalum beads for C3, C5 and sub miniature electrolytic types.

When the board has been assembled wire it to

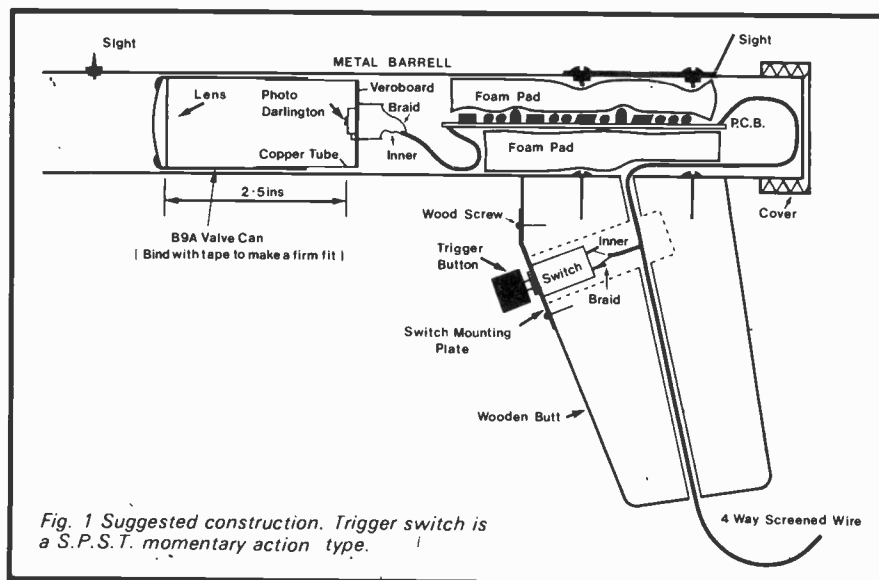
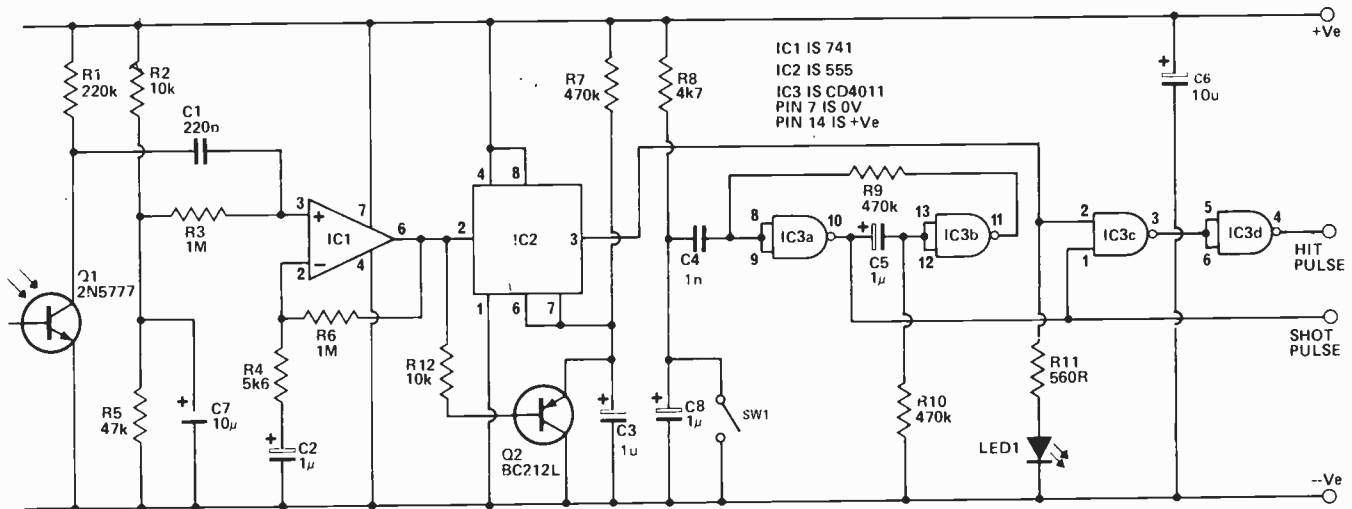


Fig. 1 Suggested construction. Trigger switch is a S.P.S.T. momentary action type.



Full circuit diagram of the TV games rifle.

PARTS LIST

RESISTORS (all 1/2 W 5%)

R1 220 k
R2 10 k
R3 1 M
R4 5k6
R5 47 k
R6 1 M
R7 470 k
R8 4k7
R9 470 k
R10 470 k
R11 560 R
R12 10 k

CAPACITORS

C1 220p polystyrene
C2 1μ 16 V electrolytic
C3 1μ 35 V tantalum
C4 1 n polystyrene
C5 1μ 35 V tantalum
C6 10μ 16 V electrolytic
C7 10μ 16 V electrolytic
C8 1μ 16 V electrolytic

SEMICONDUCTORS

LED1 TIL209
Q1 2N5777
Q2 BC212L
IC1 741
IC2 555
IC3 CD4011

MISCELLANEOUS

PCB
Trigger switch - see text
Lens - 2 1/2" focal length, 1" diameter
Metal tube 8 1/2" long
Wood for butt
4 core screened cable
Misc. hardware for sights etc.
A complete kit is available from Watford Electronics, including lens, tube and butt.

HOW IT WORKS

Q1 is the photodarlington detector. Signals appearing at its collector, due to the target on the TV screen, are coupled by C1 and R1 from the high pass filter required to reject spurious light sources. IC1 has a gain of about 200 at high frequencies, but its gain drops to unity at dc, due to the effect of C2 included in its feedback loop.

The no signal level of the output is set by R2 and R5 to about one volt lower than the supply rail.

When the target is detected, the output of IC1 falls and triggers IC2 which is arranged to operate as a monostable. Pin 3 of IC3 will go high. Q2 is included to make IC2 re-triggerable; its out-

put will remain high until one period after the input signal returns high.

The output of IC2 is fed to IC3c and to the on-target LED.

The trigger switch SW1 is coupled to a monostable formed by IC3a and IC3b. The output of IC3a is normally low and goes high for a few milliseconds when SW1 is operated. This output is the 'shot pulse' and is fed to IC3c, where it is 'NANDed' with the on-target pulse, resulting in a hit pulse which is inverted by IC3d.

The hit pulse is fed to pin 27 of the AY3 - 8500 and the shot pulse is fed to pin 26 of the games chip.

C6 decouples the supply rail.

GETTING HOLD OF COMPONENTS

Watford Electronics will supply a complete set of parts including lens, barrel and butt.

COST OF CONSTRUCTION

The cost of the Watford Electronics kit for this project is £10.50.

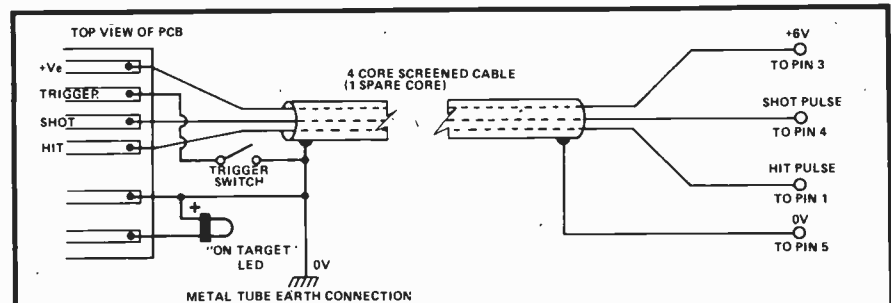
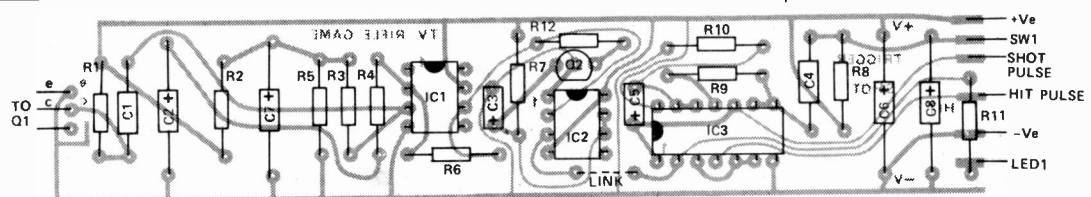
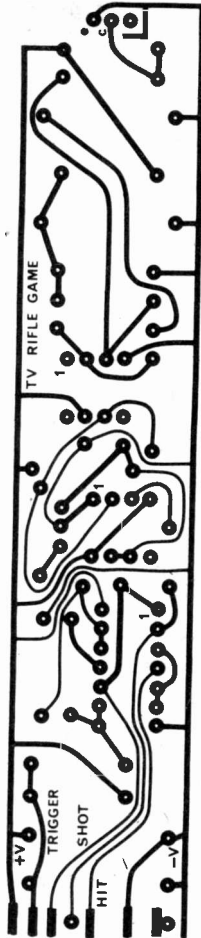


Diagram showing the connections between the PCB and ETIs TV games unit. Pin numbers shown correspond to those on the games unit PCB.

Component overlay for games rifle board shown full-size.



TV GAMES RIFLE



the trigger switch and to the connecting cable. It may then be mounted at the rear of the barrel by gluing it to a foam pad which may in turn be stuck to the barrel.

The connecting wire we used was four core screened cable and should be connected to the games unit as follows: (via DIN plug)

0V	to pin 5
6V	to pin 3
Hit pulse	to pin 1
Shot pulse	to pin 4

Testing

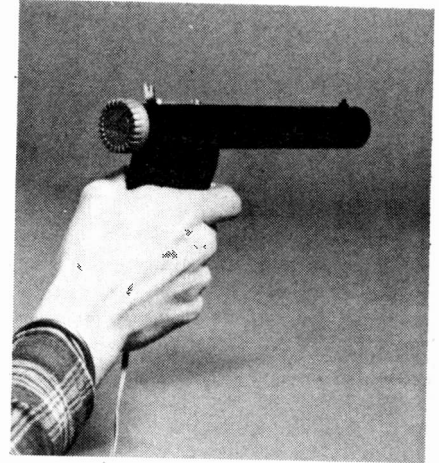
The games unit should now be connected to your TV and the brilliance control of the TV adjusted until the target is bright and the background just visible. If the rifle is now aimed at the target, at close range initially, the 'On Target' LED should glow and a blip should be heard from the speaker when the trigger is pressed.

To adjust the sights use the score display which appears at the end of the game. Block off all the screen except one digit of the score

The sight may then be adjusted so that the hit LED is on when the pistol is aimed at the score.

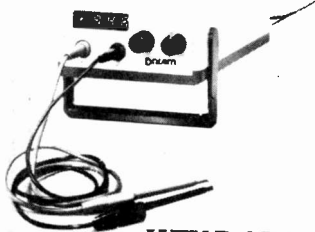
To play

Select the rifle option required with SW2 and press the reset and serve buttons. The target should appear on the screen and bounce around the screen until hit.



The PCB layout is shown full size on the left. Above is a picture showing the games rifle.

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KEY DATA

Volts DC	200mV, 2V, 20V, 200V, 1000V Input impedance 4.2M Ω Maximum sensitivity 100 μ V
Volts AC	200mV, 2V, 20V, 200V, 500V Input impedance 4.2M Ω Maximum sensitivity 100 μ V
Current AC/DC	200 μ A, 2mA, 20mA, 200mA, 2A
Resistance	200 Ω , 2K Ω , 20K Ω , 200K Ω , 2M Ω

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DESIGNING & USING ACTIVE FILTERS

PART 1

A SHORT SERIES BY TIM ORR WHICH WILL ENABLE THE HOME CONSTRUCTOR TO UTILISE CIRCUITS OF HIGH COMPLEXITY AS EASILY AS PLUGGING IN A RESISTOR!

THERE IS NO DOUBT that active filters are very useful devices. Also, there is no shortage of literature on the subject. This would seem to suggest that designing active filters is a fairly straightforward business. Well, it is and it isn't. It is if you read *this* article. It isn't if you read the aforementioned literature. Most of the books on this subject have filled our heads with terms such as poles and zeros, Laplace transforms, transfer functions, etc, which haven't actually helped us to design anything!

Some basic theory

It is advisable quickly to run through some basic terms and expressions. Firstly, consider a simple low pass filter, Fig. 1a. The frequency response (Fig. 1b) is

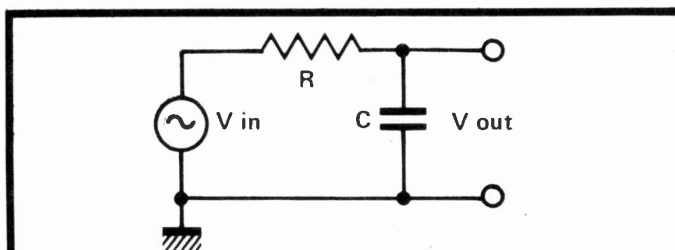


Fig. 1a. Simple low pass filter.

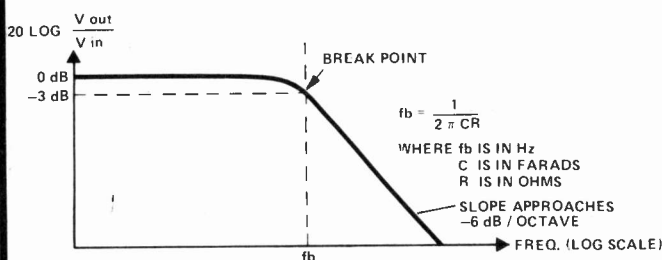


Fig. 1b. Frequency response of above.

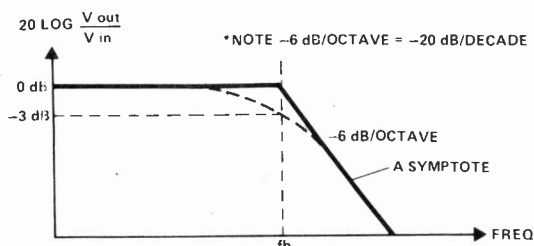


Fig. 1c. Approximation to response.

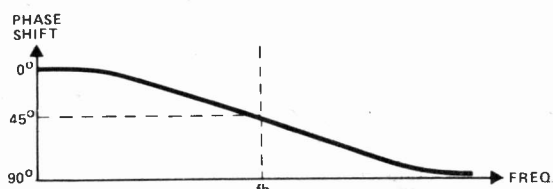


Fig. 1d. Phase shift v Frequency plot.

nearly flat until the break point, denoted by f_b . After this point the response rolls off at 6dB/octave, that is signals above this frequency are increasingly attenuated. The break point is defined as being the frequency where the resistance equals the capacitive reactance. When this occurs, the output is attenuated to 0.707(-3dB) of the input. Although the resistance equals the capacitive reactance, the output is *not* half of the input. (This is because it is the vector sum of the two and hence equals 0.707 of the input!)

As the frequency response is a rather complex curve it is very useful to use a straight line approximation to it. These lines are called asymptotes (Fig. 1c). Note that the frequency response graph uses the convention of logarithmic scales, octaves or decades along the frequency axis, and dBs along the vertical axis representing output voltage divided by input voltage.

Phase shift with respect to frequency is also often plotted as in Fig. 1d. These two (the phase and frequency response plots) are known as Bode diagrams and are generally considered the most useful way of representing a filter's performance.

You will note that for the lowpass filter of Fig. 1a, the phase shift starts at 0° , is 45° at f_b and then approaches 90° as the frequency approaches infinity. This is not an active filter, it is composed entirely of passive components which means that its output cannot be effectively loaded without changing its performance.

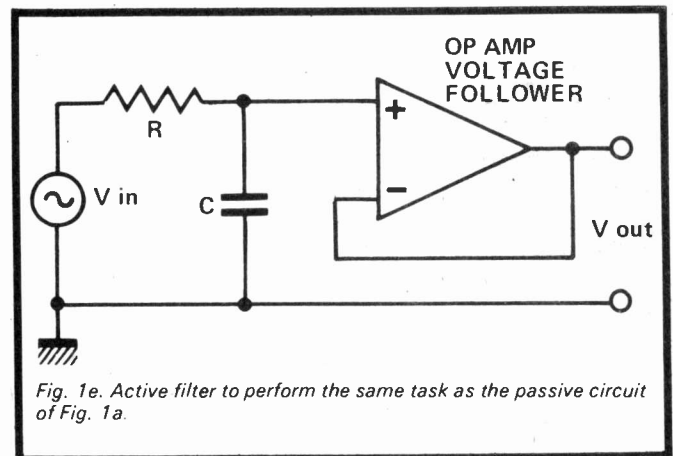


Fig. 1e. Active filter to perform the same task as the passive circuit of Fig. 1a.

Fig. 1e shows the same filter but in its active form, the op amp being used as a voltage follower serving only to isolate the filter's output. This type of filter is known as a First Order filter — a measure of the roll off slope.

When a more rapid slope is required, a higher order filter structure (one with more reactive elements) must be used. This is dealt with later.

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Summary of low pass filter of Fig. 1.

Filter type	Low pass
Filter order	First order
Roll off slope	-6dB/octave or -20dB/decade (the same)
Breakpoint fb	$f_b = 1 / 2\pi CR$ Hz
Phase shift at fb	45°

TABLE 1.

Passing highs

Next, let us consider the simple high pass filter of Fig. 2a. It is the complement of the low pass filter, the elements having been interchanged. Therefore it is not

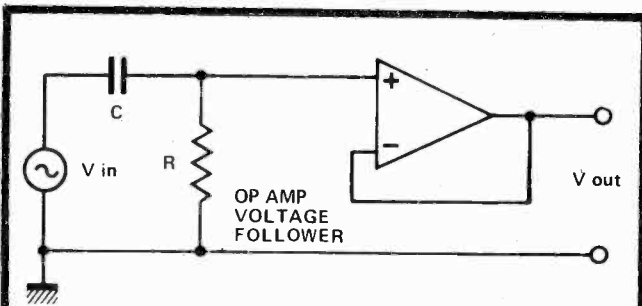


Fig. 2a. Simple high-pass active filter.

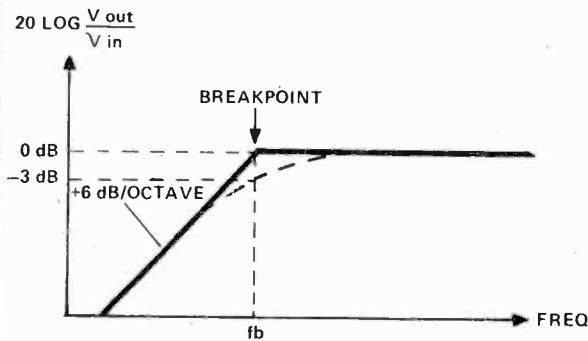


Fig. 2b. Frequency response of the high-pass filter.

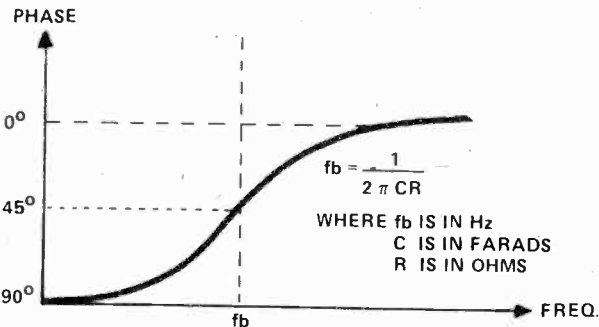


Fig. 2c. Phase v frequency plot of the same filter.

difficult to accept the complementary phase and frequency response curves of Fig. 2b. Note that the break point is the same and so is the roll off slope.

Summary of the high pass filter of Fig. 2.

Filter type	High pass
Filter order	First order
Roll off slope	+6dB/octave or +20dB/decade
Break point fb	$f_b = 1 / 2\pi CR$ Hz
Phase shift at fb	45°

TABLE 2

Passing bands

The next type to be considered is a simple band pass filter shown in Fig. 3a. Although it uses an inductor it is only to illustrate the bandpass theory. Later on in this series, inductors will be replaced by their active equivalents.

The frequency response (Fig. 3b) shows that this circuit is symmetrical, having roll off slopes of 6dB/octave on either side of its RESONANT peak. This

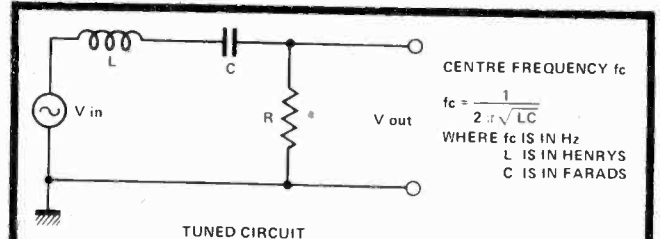


Fig. 3a. Simple band-pass filter.

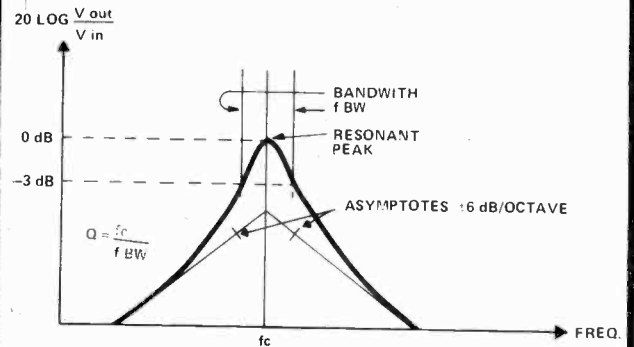


Fig. 3b. Band-pass frequency response.

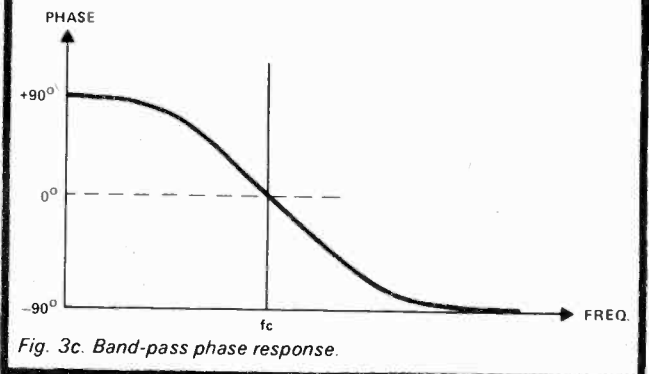


Fig. 3c. Band-pass phase response.

filter is known as a second order filter, because it has two reactive sections, the L and the C. The C is responsible for the +6dB/octave portion of the slope, the L for the -6dB/octave portion. But where these two slopes should meet, the response of the filter peaks and the slopes become much larger (Reson-

ance). The sharpness of this peak is described as the Quality of the filter, the Q factor. Resonance occurs at a frequency known as the Centre frequency denoted by f_c .

The bandpass filter is so called because it only passes signals within a certain bandwidth, which is defined as being the frequency range contained between the two points that are 3dB below the resonant peak. There is a fixed relationship between Centre frequency (f_c), bandwidth (fbw) and Q factor, given by $Q=f_c/\text{fbw}$.

The centre frequency is given by $f_c \approx 1/2\pi\sqrt{LC}$ Hz. This is only approximate, as it assumes that the value of R is relatively low. As R decreases, the Q factor increases. Thus R has the effect of damping the resonances, and so as it approaches zero ohms, Q approaches infinity.

The phase shift is shown in Fig. 3c. As this filter is a second order structure, then the total phase movement will be twice that for a first order structure, i.e. 180° . Fig. 3d shows the phase and frequency responses for different values of Q. Note that a high Q has a very rapid rate of change of phase, a low Q has only a slow rate of change.

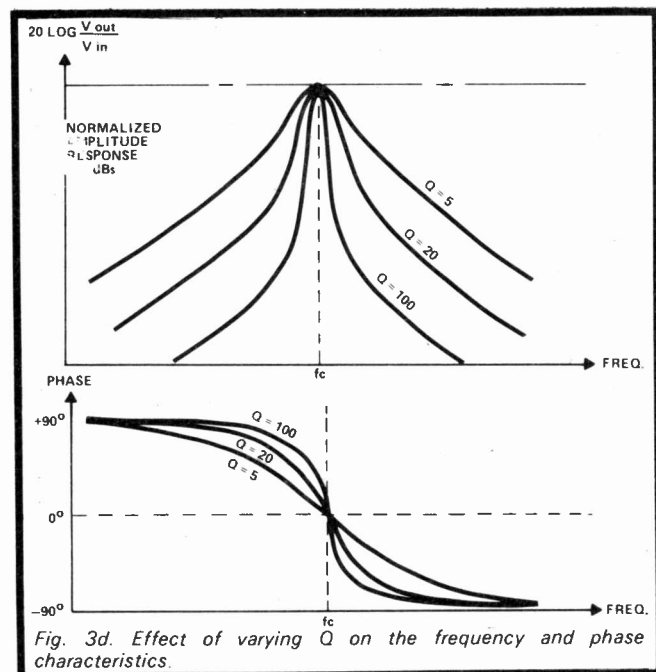


Fig. 3d. Effect of varying Q on the frequency and phase characteristics.

Time please

Bandpass filters also have a time response, as opposed to their frequency response. When an impulse is applied to a bandpass filter it rings (Fig. 3e). The filter oscillates at the centre frequency, f_c , the amplitude of the oscillations decaying exponentially in time. The ringing time, T_r , is the time taken for the oscillations to decay to 37% of their initial value. Ringing time is related to the Q and f_c by the following equation:

$$T_r = Q/2\pi f_c$$

When a high Q filter has been constructed, it may prove difficult to measure its Q factor accurately due to the narrowness of its bandwidth. However, if the filter is made to ring, a reasonably accurate measurement of the Q can be obtained by measuring T_r and f_c .

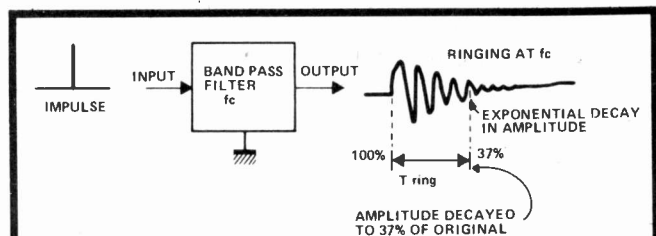


Fig. 3e. Ringing in a band-pass filter.

Filter type	Band pass
Filter order	Second order
Roll off slopes	+ and -6dB/octave greater near to resonance
Centre frequency f_c	$f_c \sim 1/2\pi\sqrt{LC}$
Phase shift at f_c	0
Q factor	f_c/fbw where fbw is the 3dB bandwidth
3dB bandwidth fbw	f_c/Q
Ringing time, T_r	$Q/2\pi f_c$

TABLE 3. Summary of band-pass filter.

Failed band

Another common filter structure is the band reject or notch filter. There are many ways of realising this filter, one of which is shown in Fig. 4. The input signal is subtracted from the bandpass output. By adjusting R_a with respect to R, complete cancellation can be obtained at f_c .

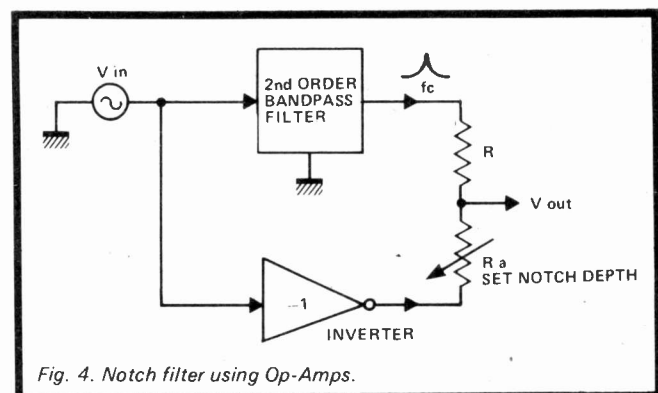


Fig. 4. Notch filter using Op-Amps.

Thus the centre frequency of the bandpass filter is the centre frequency of the notch, whose depth can be varied by altering R_a .

Very deep notches are possible, 50dB is easily obtained. As the Q of the bandpass filter is increased, so is the Q of the notch filter. However, R_a will have to be reset for each value of Q.

Filter Order

Consider the ideal low pass filter shown in Fig. 5a. Its response is flat right up until the break frequency f_b . Frequencies above f_b are attenuated to nothing! You won't be surprised to learn that filters like this don't exist. However, it is a common requirement to produce filters with very steep roll off slopes and this is achieved by designing filters with lots of sections, to increase the

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filter order. Each reactive element in the filter increases the filter order by one, therefore a low pass active filter with three capacitors is known as a third order filter and will have an ultimate roll off of three times 6dB/octave, which is 18dB/octave.

However, designing a third order lowpass filter is not just a simple case of sticking three first order RC circuits in a line. What you get when you do this is a very soggy curve indeed! The filter should be flat in the pass band, then it should turn over and rapidly assume its ultimate roll off slope. Examples of this type of Maximally flat filter are shown in Figs 5b and c. The effect of order number upon a bandpass filter is shown in Fig. 5d.

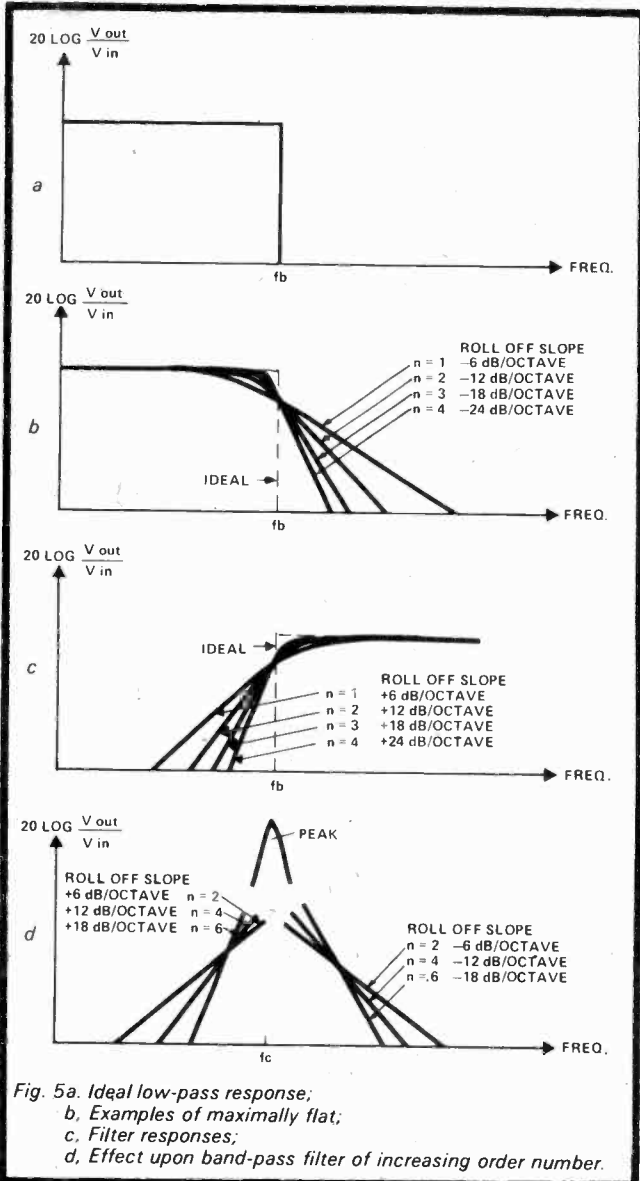


Fig. 5a. Ideal low-pass response;
b. Examples of maximally flat;
c. Filter responses;
d. Effect upon band-pass filter of increasing order number.

Later on in this series the circuit diagrams and design charts are given for various filter types and order numbers. It would seem that to get a filter to approach its ideal response, all that is needed is to increase the order number. This is in fact true, but there are certain tolerance problems. (When 8th order filters are designed, component tolerances of about 1% are required!)

Which filter shape?

The type of filter that is chosen to do a particular job will depend on what parameters are thought to be important. There are three basic characteristics to be considered (lowpass and highpass filters only).

1. Good transient response.
2. Maximum flatness of the filter within its passband.
3. Maximum rolloff slope outside the passband.

The type of filter used should be chosen to fit the job that they are being designed for. The filters have been categorised into three basic types for the purpose of simplicity.

Filter number 1 is known as a Bessel filter. Its phase changes almost linearly with frequency. It is useful for systems where a good transient response is required, such as joining the dots up on the output of a digital to analogue converter. It has a very poor initial roll off slope.

Filter number 2 is known as a Butterworth filter. It has the flattest pass band possible. Its other two parameters are a compromise. That is it has a reasonable overshoot and a fairly fast initial roll off.

Filter number 3 is known as a Chebyshev filter. It has some ripple in its pass band, although this is small, and a very fast initial roll off, and a poor transient response.

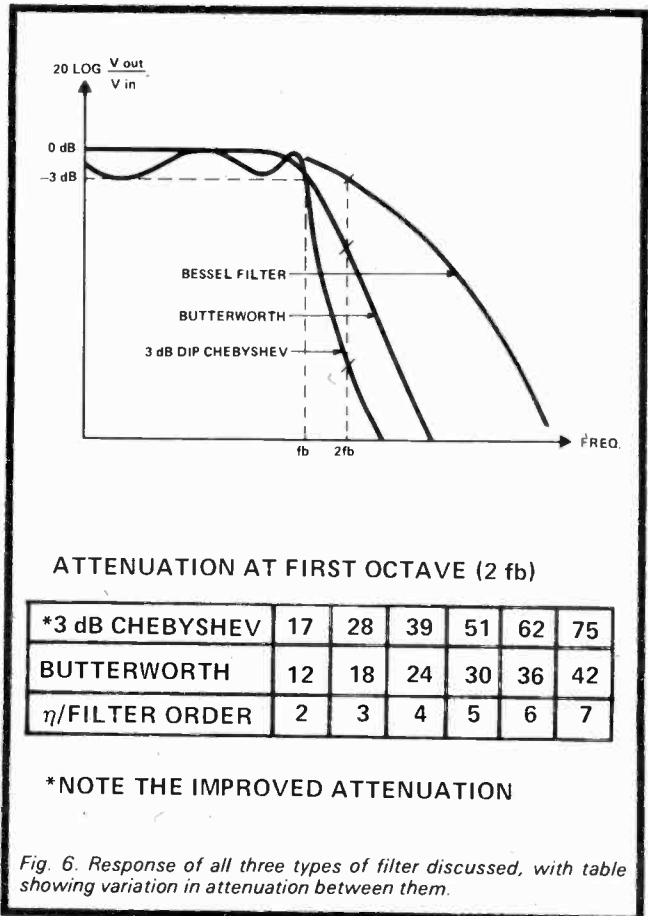


Fig. 6. Response of all three types of filter discussed, with table showing variation in attenuation between them.

Next Month: Full design charts and circuits for three types of Active Filter.

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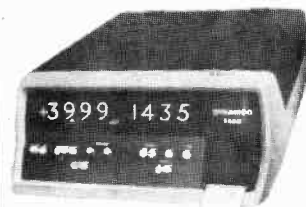
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QOV03/10 (ex-equipment), 75p or 2 for £1.20.

2C39A ceramic (ex-equipment), £2.00 each.

DET-22 (ex-equipment), 2 for £1.00.

6BH6 (ex-equipment), 2 for 50p.

All the above valves are untested, except for heaters, and no guarantee of percentage of emission is given. Sorry, no returns.

MULLARD 85A2 85V STABILISER VALVES (brand new), 70p each of 2 for £1.20.

TRANSISTORS

PNP Audio Type T05 Transistors, 12 for 25p.

BFY51 Transistors, 4 for 60p.

BYX38/300 Stud Rectifiers, 300V at 2.5A, 4 for 60p.

8CY72 Transistors, 4 for 50p.

BSX20 VHF osc./mult.), 3 for 50p.

BC108 (metal can.), 4 for 50p.

PBC108 (plastic BC108), 4 for 50p.

BF152 (UHF amp/mixer), 3 for 50p.

2N3819 Fet., 3 for 60p.

BC148 NPN SILICON, 4 for 50p.

BC158 PNP SILICON, 4 for 50p.

BA131 Signal Diodes, 10 for 35p.

BA121 Varicap Diodes, 4 for 50p.

ALL BELOW — ADD 8% VAT

DIECAST BOXES. We still stock these but, owing to frequent price rises from our suppliers, and costly postal charges, it has been found impossible to publish up-to-date prices on these items. Please ring or write (with SAE) for latest mail-order prices.

AEI CS108/R MICROWAVE DIODES, up to X-Band, max. noise figure 8.5dB at 9.375GHz, 80p each.

14-DIL REED RELAYS, our selection only £1.00 pack. Designed to work directly from TTL Logic. Single Pole Changeover. Contact ratings, 28V, 1/4A, 3V, £1.75 each.

SPIRAL Tools for Electronics enthusiasts. SAE for list.

4MHz XTAL PACKS (10 assorted xtals between 4MHz and 5MHz), our selection only £1.00 pack.

ALU-SOL ALUMINIUM SOLDER (made by Multi-core), solders aluminium to itself or copper, brass, steel, nickel or tinplate. 16SWG with Multicore flux with instructions, approx 1m coil, 40p pack. Large reel £2.75.

SOLDER SUCKERS (Plunger Type). Standard Model. £4.50.

Skirted Model. £4.95.

Spare Nozzles. £60p each.

MULTICORE SOLDER

Size 5 Savbit, 18SWG, in alloy dispenser, 32p.

Size CISAV18 Savbit, 18 SWG, 56p.

£3kg. (1.1lb.) 60/40 20 SWG on plastic reel, £27.00.

WELLER TCP2 and PU2D PSU Temperature controlled soldering iron, with matching Power Supply Unit, containing sponge and spring stand. £27.00.

SPARE TIPS (for TCP1/2). Three types available.

Type CC7 (Standard). TYPE K7 (long fine tip).

TYPE P7 (very fine tip). £1.00 each.

WELLER W60D Mains operated temperature control soldering iron. £13.80.

SPARE TIPS (for W60D). Two types available. TYPE CC7 (W60D) Standard, TYPE AA7 (W60D). Finer tip. £1.15 each.

A RANGE OF DRAPER TOOLS FOR THE ELECTRONICS ENTHUSIAST. (GOOD QUALITY, FULLY GUARANTEED).

MAINS TESTER SCREWDRIVERS, 100 to 500V. Standard size 50p. Large 70p.

RADIO PLIERS, 5 1/4" £1.60 6 3/4" £1.80.

DIAGONAL SIDE CUTTERS, 6 1/4" £1.90.

SMALL SIDE CUTTERS, L.J2 Standard £3.70.

LJ7 (with wire holding device), £4.10.

FLAT NOSE TONG CUTTERS, L.J1 £4.10.

COMBINATION PLIERS, 6 1/4" £1.75, 7" £1.80.

ALL BELOW — ADD 8% VAT

TIN SNIPS, 7" £2.25.

SLIM OPEN ENDED SPANNER SETS, 0+3, 1+2, 2+4, 3+5, 4+6, 6+BBA SIZES, £1.15.

MIDGET OPEN ENDED SPANNER SETS

0+1, 2+2, 3+5, 4+6, 6+BBA SIZES, £2.85 set of 5.

4+4, 5, 5+5, 5+6, 6+7, 8+9, 10+11 MM SIZES, £3.50 set of 6.

MINIATURE WATCHMAKERS SCREWDRIVER SETS, £1.50 set of 5.

MINIATURE FILE SETS, set of 6 £1.90. Set of 10 £3.25 (Round, flat, etc.).

TAP AND DIE SETS (18 piece) contain 1 each of 0, 2, 4, 6, 8, BA SIZES IN Dies, Plug taps, Taper Taps + American Type tap wrench, T-type tap wrench, Die Holder, £11.60.

TUBULAR HACKSAW FRAMES (with Blade), £2.75.

HEAVY DUTY RELAYS, 24V DC operated (will work on 18V), 3 heavy duty make contacts (around 10A rating) + 4 change-over contacts + 1 break contact. New, complete with mounting bracket (ideal for switching HT on Linears). Many uses for this high quality unit. £1.50 each.

VARICAP TUNERS, Mullard type, ELC1043/05. Brand new. £4.40.

T V Sockets (metal type), 4 for 50p.

T V Sockets (metal type), 4 for 50p.

T.V. Line connectors (back-to-back sct.), 4 for 50p.

3-pin Din Plugs, 4 for 50p.

Din 3-pin Line Sockets, 15p each.

Din Speaker Scts, 2-pin, 4 for 30p.

TWIN IF CANS, approx. 1" x 1/2" x 1 1/2" high, around 3.5 to 5MHz, 2 separate transformers in one can, internally screened, 5 for 50p.

Dubilier Electrolytics, 50µF 450V, 2 for 50p.

Dubilier Electrolytics, 100µF 275V, 2 for 50p.

Plessey Electrolytics, 470µF 63V, 3 for 50p.

TCC Electrolytics, 1000µF 30V, 3 for 50p.

Dubilier Electrolytics, 5000µF 35V, 50p each.

Oublier Electrolytics, 5000µF 50V, 60p each.

ITT Electrolytics, 6800µF 25V, high grade, screw terminals, with mounting clips, 50p each.

A RANGE OF CAPACITORS AVAILABLE AT BARGAIN PRICES. SAE FOR LIST.

videocraft

Half price Teletext

You can now buy Texas **Tifax** module Teletext decoder complete with matching cable connected keyboard, power supply, interface board and complete instructions for installation in most common television receivers for only £180 + VAT and £2.50 postage, packing and insurance.

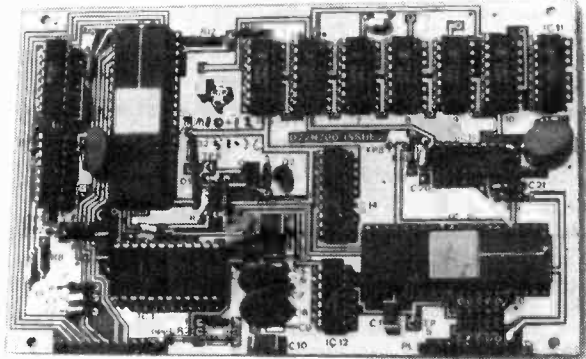
Since the interface is connected directly to the television's video output circuitry, picture quality is excellent with pure colours — much more so than is possible from decoders which feed the aerial socket.

Due to the compact nature of the **Tifax** module, installation within most receiver cabinets is no problem. Facilities include seven colours, upper and lower case alphanumeric, graphics, time coded display, and newflash and subtitle inserted in TV picture.

To enable us to supply the correct interface board and instructions, we must know your television set make and model and, if possible, chassis type.

Additionally, for those uncertain about installing a decoder in their own television set, a colour television receiver complete with a fully operational Teletext decoder is being offered for under £500 — that's less than half the cost of existing receivers. Please send an SAE for full details and prices.

Videocraft, Assets House, Elverton Street, London SW1P 2QR
Phone: 01-828 2731. Telex: 896953



Tifax modules are available from our own stock

To Videocraft, Assets House, Elverton Street, London SW1P 2QR

Please supply _____ Texas **Tifax** modules, power supply, interface boards and wired keyboard at £196.90 each including VAT and postage, packing and insurance. I enclose my remittance for £ _____

Television make _____ Model No. _____

Chassis type (if known) _____

Name _____

Address _____

Registered in England No. 1297569
Registered Office, 27 Dover St., London W1

ET7

SHORT CIRCUITS

KEEP THIEVES AWAY — WITH OUR INVALUABLE

ALARM ALARM

ONE PROBLEM WITH BURGLAR alarms is that they don't 'go off' until the burglar has broken in, but here is a project which can be installed in a car to warn thieves that a burglar alarm is operating. It should warn a thief to go and find a car which is not owned by an ETI reader! Even if there is actually no burglar alarm, the 'alarm alarm' can still be used. It's what the car thief believes that counts — and he's not going to investigate to see whether there really is an alarm.

The unit is simply a box containing two lamps which flash slowly on and off, together, and shine through a Perspex panel to illuminate the words ALARM ACTIVE. It uses a 555 timer IC, which is used as an astable multivibrator.

As the circuitry is isolated from the box this alarm can be used with any car having a 12 volt battery - whether the positive or negative terminal is connected to the chassis. Take care to see that the unit is correctly connected.

Installation

The unit can be permanently mounted in a car near one corner of the windscreen and the wiring neatly run to a switch below the dashboard. Alternatively it may simply be placed in position when required, and plugged in to the cigarette lighter socket. To work effectively it should be prominent day or night.

Construction

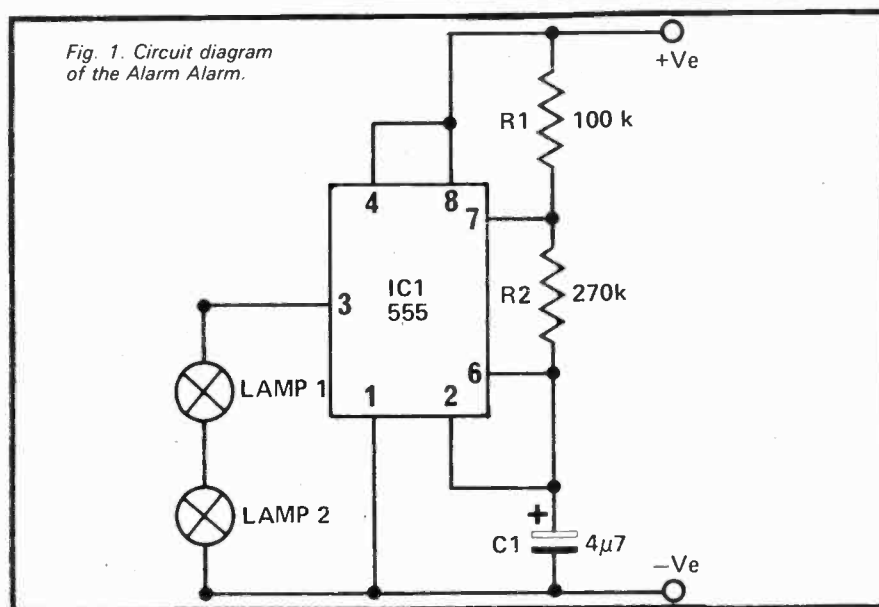
We mounted the components on an 'L' shaped bracket which is ideal for fitting to the dashboard. Lamps 1 and 2 are push fitted into two rubber grommets mounted on an aluminium bracket, and arranged to illuminate the perspex panel as shown. We used Letraset for the panel lettering.

The components are assembled onto the small PCB according to the component overlay, taking care that the 555 and C1 are correctly orientated.

We fitted an On/Off switch but if the car actually has a burglar alarm,



Fig. 1. Circuit diagram of the Alarm Alarm.



HOW IT WORKS

The 555 IC is used as an astable (i.e. not stable) multivibrator. As soon as it is connected to the supply it starts to oscillate (slowly in this case) and the output voltage at pin 3 changes regularly and suddenly from high to low and low to high as the capacitor is charged and discharged.

The charge time (during which the output is high and the lamps are on) is given by the formula:

$T_c = 0.69 (R_1 + R_2) \times C$ and is in seconds when R1 and R2 are in megohms and C is in microfarads. So

$$T_c = 0.69 (0.1 + 0.27) \times 4.7 = 1.2 \text{ seconds.}$$

The discharge time (during which the output is low and the lamps are off) is given by the formula:

$$T_d = 0.69 \times R_2 \times C = 0.69 \times 0.27 \times 4.7 = 0.88 \text{ seconds.}$$

Total time of one oscillation = $T_c + T_d = 2.08$ seconds. So, we have a flasher which is on for about 1 second in 2. The exact timing depends on the actual capacitance of the capacitor C, and this may differ from its rated value by as much as -20% and +50%.

The rate of flashing may be changed by changing the values of R1 and R2. Higher values cause slower flashing.

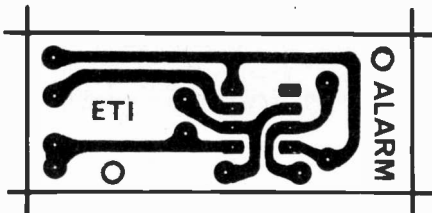
Short Circuits

then this device should be connected so that it is activated as the burglar alarm is energised.

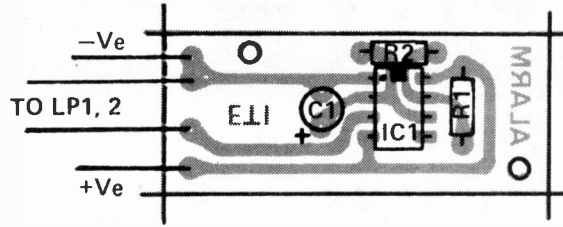
The parts list specifies two 6 volt lamps of 60 mA rating which are connected in series. The current consumption is so low that the unit could be left operating for many hours without any danger of running down a car battery.

The IC is actually capable of switching up to 200 mA through pin 3, so there is no reason why two or even three slave units (with lamps only) should not be run in parallel with the lamps in the master unit. This could provide warnings at all vulnerable points in a car.

This same device can be used in windows of homes as a discouragement to house burglars. In this case it could be operated from a simple power supply running from the mains.



Foil pattern shown full size.



Component overlay of Alarm Alarm. Because of the small size, miniature components should be used.

PARTS LIST

RESISTORS (all 1/4 W 5%)

R1 100 k
R2 270 k

CAPACITOR

C1 4u7 16 V tantalum

SEMICONDUCTOR

IC1 555 timer

LAMPS

LP1,2 6 V .06 A MES type

SWITCH

See text

MISCELLANEOUS

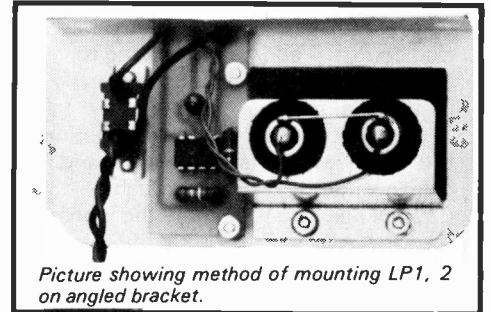
Nuts, bolts, spacers, etc.
Aluminium for front panel and bracket
Grommets, flex, red translucent perspex
PCB as pattern.

GETTING HOLD OF COMPONENTS

All the components used in this project should be available from most component suppliers.

COST OF CONSTRUCTION

The total cost of this project should be about £2.00.



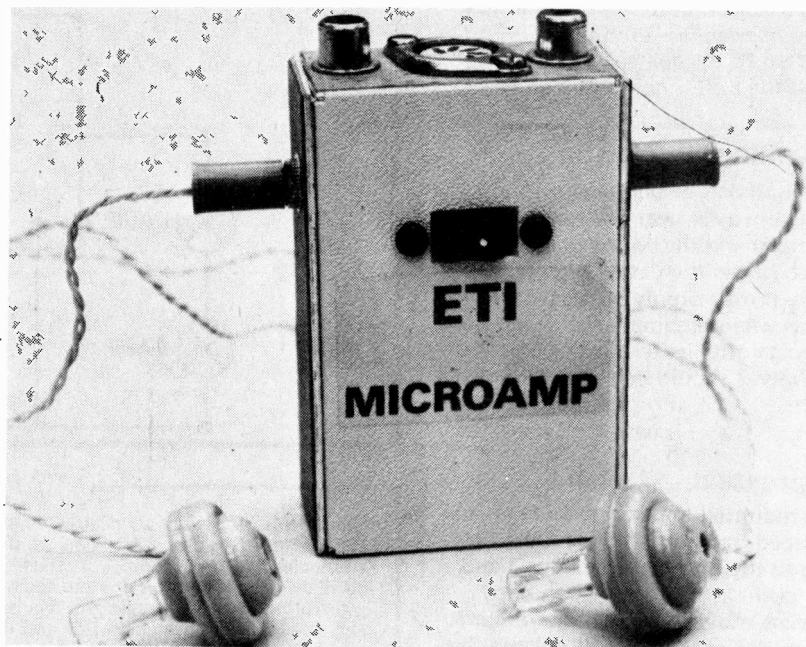
Picture showing method of mounting LP1, 2 on angled bracket.

microAMP

THERE IS OFTEN A NEED for a piece of equipment which can give a reliable answer as to another unit's state of being. In audio, for instance, a repaired amplifier might need to be tried without risking a pair of expensive monitor loudspeakers, or even headphones (which are worth a few bob themselves these days!).

Our micro-amp is designed to be a portable stereo test amp, capable of betraying any faults or distortions inherent in the suspect unit. The transducers utilised are low-cost crystal earpieces, for which the design has been optimised. Although there are only a handful of components in the design, the amp gives exceptionally good sound quality suitable, say, in checking whether that cassette deck in 'Rip-Off Hi-Fi' has 1% or 100% distortion.

Quality is ultimately limited by the earpieces, but they are capable of doing better than the two-transistor 'Super-Squark' portable radios to which they are more usually mated.



View of completed Micro amp

In and Out and In . . .

In the prototype, sockets were provided for a 'tape input' type of signal, i.e. from a cassette recorder at the DIN socket pins 3 and 5. If a signal is to be input from a tuner or amplifier, either use the phono sockets

or pins 1 and 4 so that you keep things standard.

Input level is ideally around 100 mV; if vastly different to this, R1 can be juggled in value to compensate. Increase if the level is higher.

PARTS LIST

RESISTORS (all 1/4 W 5%)

R1,4 680 k
R2,5 2M2
R3,6 39 k

CAPACITORS

C1,3,4,6 4u7 tantalum
C2,5 22 p polystyrene

SEMICONDUCTORS

Q1,2 BC109 C - see text

SWITCH

SW1 On-off rocker, or slide type

SOCKETS

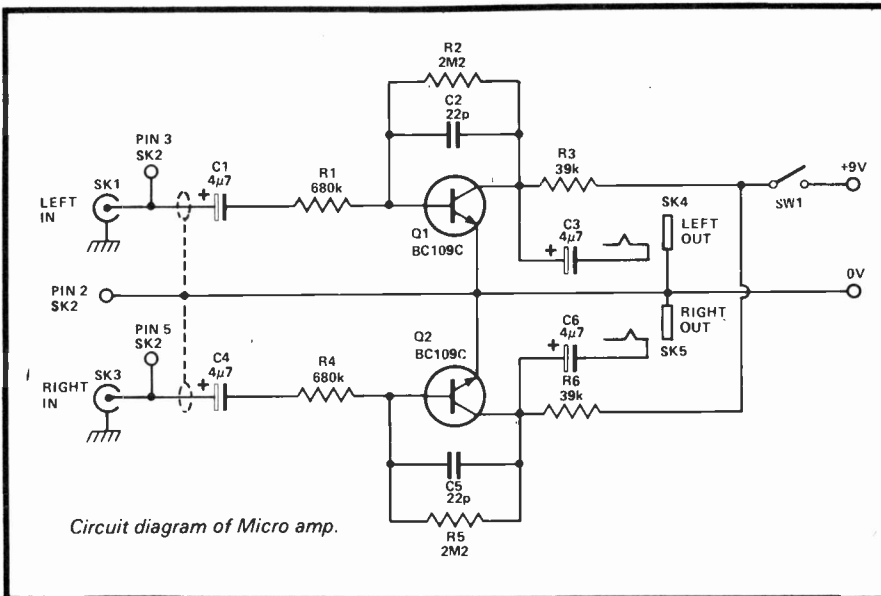
SK1,3 Chassis phono sockets
(Doram: 478 093 red, or:
477 848 black)
SK2 Chassis 5 pin DIN 180° socket
SK4,5 3.5 mm chassis jack socket

CASE

Norman type AB12 or similar (3 x 2 x 1")

MISCELLANEOUS

PP3 battery, clip to suit,
Miniature screened wire flex
Nuts, bolts, spacers etc.
PCB as pattern
2 off crystal earpieces with 3.5 mm jack
plugs



HOW IT WORKS

Q1 and Q2 are base biased single stage amplifiers. The feedback capacitors C2 and C5 are there to provide high frequency correction, and experimentation with the value will change the resultant sound quite noticeably.

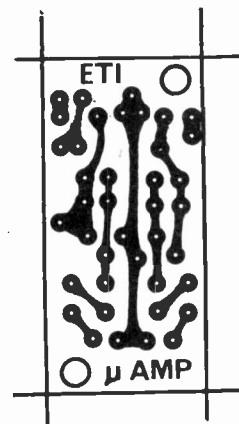
C1 and C4 decouple the input from preceding circuits, and the resistors R1 and R4 will set the level seen by the amplifier, and hence by the earpieces. No volume controls are provided, as none proved to be necessary with the prototype. C3 and C6 serve to decouple output from dc. Crystal earpieces only are recommended.

GETTING HOLD OF COMPONENTS

Most components should be readily available from a number of suppliers. We have listed Doram part numbers for the phono sockets and Norman's reference for the box we used in our microamp.

COST OF CONSTRUCTION

The total cost of this project should be about £3.50 inc VAT.



Foil pattern of PCB is shown full size.

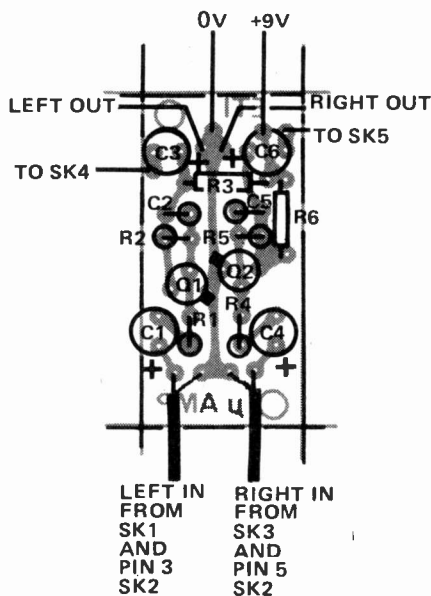
As seems to be usual for us nowadays, space within the box is very restricted, but it will go into the case if you take some care over the layout. Perhaps our photographs will help.

Power and Construction

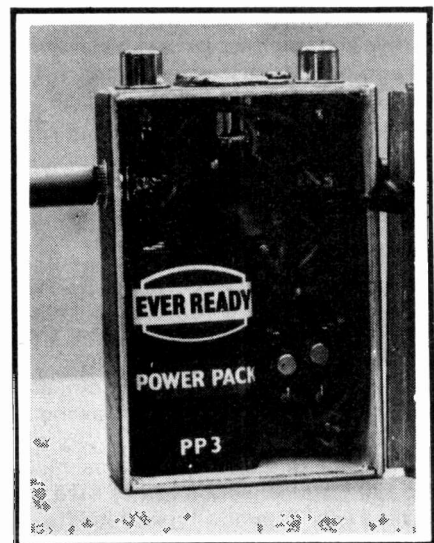
A PP3 is all that will fit into our box and is all that is needed. Current drain is around 300 uA (hence the name!) and so even this will have a life-span approaching that which it would have enjoyed had you left it sat sitting merrily on a shelf.

The PCB is smaller than most, so take care when soldering it up: too long with the iron in one place, and the track will become emotionally attached to the bit, and not wish to leave it!

BC109Cs must be used to give a high enough output from the specified input. Surplus transistors will obviously work, but don't blame us if the sound is bad!



Component overlay of Micro amp.



SHORT CIRCUITS

CONVERT ANY 1mA METER MOVEMENT INTO AN ACCURATE TACHO FOR YOUR CAR WITH OUR

TACHOMETER

This design uses a single integrated circuit to provide an easily calibrated unit that will provide rpm indications of a wide range of engine speeds. It is suitable for 4 or 6 cylinder engines.

UNTIL TEN OR SO YEARS AGO, car tacho's were cumbersome mechanical devices usually driven via a flexible cable from gearing attached to the shaft of the vehicle's dynamo — or sometimes via the distributor shaft.

The advent of transistor technology changed all this and since then almost all car tacho's are electronically operated.

The basic principle is much the same for all electronic tacho's an electrical signal taken from the low tension side of the distributor is converted into a voltage proportional to engine rpm and this voltage is displayed on a meter calibrated accordingly.

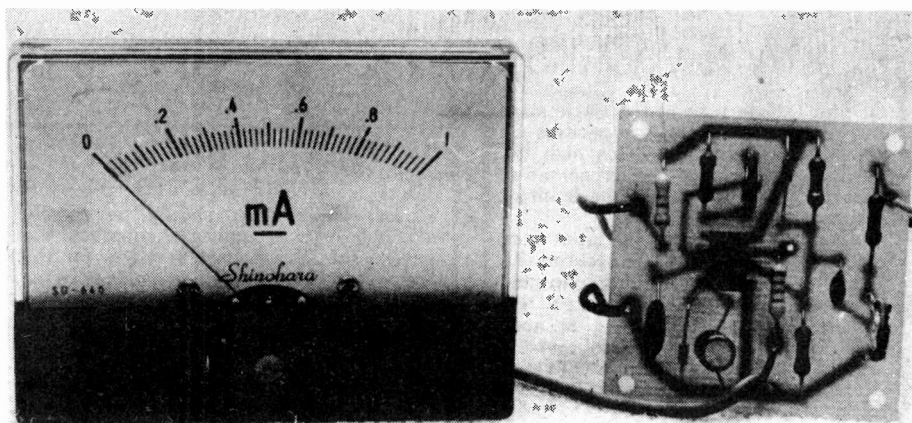
Most car tacho's are complex and expensive devices — but here's one with a difference! It is simple yet extremely effective. Its simplicity is due to our using one single integrated circuit rather than the more conventional multiplicity of individual transistors.

The unit will operate on both positive and negative earth vehicles and will also operate successfully and without modification with most types of electronic ignition systems as well as the more common electro-mechanical systems.

Construction

As there are so few components, construction is very simple and straightforward. Do make sure though that the 555 IC is soldered in the right way round — ditto the two diodes. Compare your work against our layout drawing as a final check.

Any type of meter that has one milliamp full scale deflection can be used. This is a very common type of instrument and you should be able to obtain



Photograph showing a view of the completed pcb connected to a 1mA meter with 120° movement. The compact layout of our single integrated circuit design is apparent from this picture.

one new or secondhand with no difficulty. Ideally you should choose one that has 180° or 280° movement but these tend to be rather expensive. The meter size should be chosen to suit your proposed housing.

When the meter has been assembled connect it to the vehicle's battery and connect the input to the contact breaker side of the coil.

Calibration

We can think of three ways to calibrate this unit. The easiest method is to borrow an already calibrated tacho which can be temporarily connected to your car. RV2 may then be adjusted until the two readings agree over a range of engine speeds.

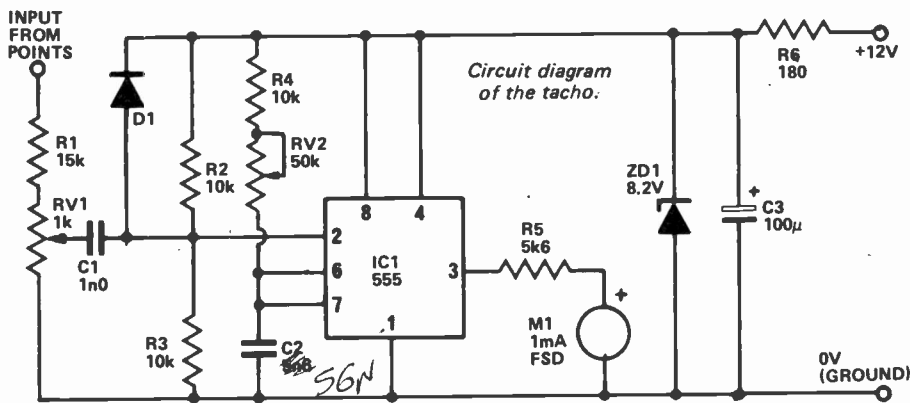
The unit may also be calibrated by the use of a signal generator by recognising that with a four cylinder engine there are two sparks per engine

revolution. To calibrate the unit, take the output from a signal generator to the tacho (via an amplifier if necessary) and adjust RV2 until the reading on M1 satisfies the relationship: $f = 2M$ where f = frequency in Hz, and M = meter reading in r.p.m.

Our final method is to calculate the vehicle speed per 1000 r.p.m. in top gear and adjust RV2 accordingly. Needless to say, this is a two person operation.

If the adjustment of RV2 is found to be too coarse its value may be reduced to 25k or lower. If this is done it will be necessary to increase the value of R4 accordingly.

Before making final calibration, adjust RV1 to eliminate any false triggering - check at all engine speeds. This unit may be used with either +Ve or -Ve earth vehicles - simply connect the battery leads as shown.



HOW IT WORKS

The 555 timer IC is used as a monostable which, in effect, converts the signal pulse from the breaker points to a single positive pulse the width of which is determined by the value of $R4 + RV2$ and $C2$. The mathematical formula is $T = 1.1 \times R \times C$ where $R = R4 + RV2$ (the section of $RV2$ in use) and $C = 5.6 \times 10^{-9}$ (Farads), and T = pulse length in seconds.

Resistors $R2$ and $R3$ set a voltage of about 4 volts at pin 2 of $IC1$. The IC is triggered if this voltage is reduced to less than approx. 2.7 volts (1/3 of supply voltage) and this occurs due to the voltage swing when the breaker points open.

An adjustment potentiometer $RV1$ enables the input level to be set to avoid false triggering.

Zener diode $ZD1$ and the 180 ohm resistor stabilise the unit against voltage variations.

PARTS LIST

RESISTORS (all 1/2 W 5%)

R1 15k
R2-4 10k *3x9*
R5 5k6
R6 180 R

CAPACITORS

C1 1n0 polyester
C2 ~~5.6n~~ polyester *5.6n*
C3 100 u 10 V electrolytic

SEMICONDUCTORS

D1 1N914
ZD1 8V2 300 mW
IC1 NE555

POTENTIOMETERS

RV1 1k
RV2 50k *20V*

MISCELLANEOUS

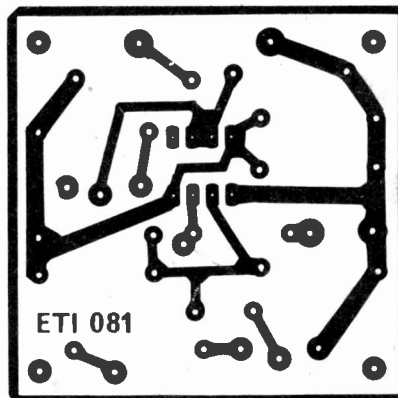
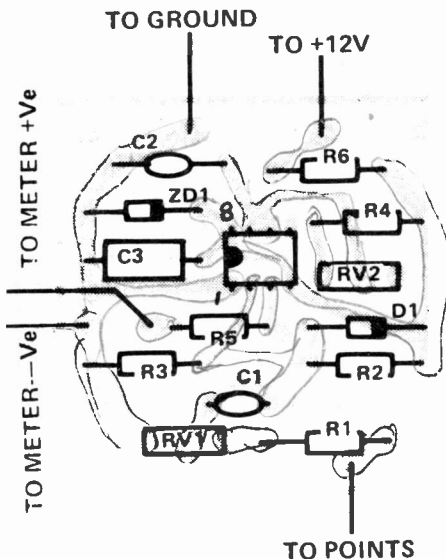
PC board ETI 081
Meter 1 mA FSD

GETTING HOLD OF COMPONENTS

There should be nothing here to trouble the constructor — the 555 is advertised by a number of firms in this issue and the other semi-conductors and passive components should be readily available.

COST OF CONSTRUCTION

Cost of components for this project should be about £2.00. This includes the PCB but is exclusive of meter $M1$.



The foil pattern of the printed circuit is shown full size above and to the right is the component overlay for the PCB.

BUILD THE TREASURE TRACER MK III



METAL
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AS SEEN
ON BBC-1
& BBC-2
TV

- Genuine 5 silicon transistor circuit, does not need a transistor radio to operate.
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VALVE SOUND on the rebound?

GEORGE CHKIA NTZ
AND
RICHARD ELEN

TWO PROFESSIONAL SOUND ENGINEERS HAVE A NEW AND HIGHLY CONTROVERSIAL THEORY TO ADD TO THE CONTINUING DEBATE AROUND THE SUPPOSED SUPERIORITY OF VALVE AMPLIFIERS

IMAGINE YOU ARE SITTING in front of a pair of monitor speakers, listening to some master tapes. The first track you hear is a very good, high quality mix, such as one could find on a fair proportion of modern albums. But the second one is something else.

It too is of very high quality, but in addition, it seems to possess an undefinable 'something'. Could one call it 'clarity'? 'Rightness'? It's hard to put into words, but there is something about that mix which makes it stand out from the rest — an unknown factor which is nonetheless sufficiently noticeable for you to be able to tell the difference every time. Yet you know that the two recordings were made in the same studio, on the same desk, the same tape machines — every link in the chain from microphone to master tape was duplicated. But with one subtle change. The first mix was monitored on a pair of loudspeakers driven by a modern, high-power transistor amplifier with a rated output of 300 W.

The second, on the other hand, was reproduced using a pair of old, glowing valve amps running a mere 50 W — and unaccountably sounding just as loud.

Why the difference?

Valves versus Transistors

It is important to state at the outset that this is intended to be a *qualitative* analysis of points that we have found to be significant over the past few years during our work as

sound recording engineers. Some of the points we are raising are nowadays being discussed, but not all. One reason for this is that several of them are not easily measured — rather, they are experienced.

Until a couple of years ago, the standard answer you *would* get from any audio engineer worth his salt when asked the question, "*Why do valve amps sound better than transistors?*" would have been, "*Well, valves produce more even-harmonic distortion, and you like the distortion, which is more 'musical' than the distortion produced by a transistor amp. And after all, transistors are more distortion-free than valves ever were.*" He would not give the same answer today, unless he spent all his time listening to test instruments instead of real signals.

A big question that needs to be discussed is, "*Is distortion just a variable you can measure with test equipment or is high fidelity, in the words of a well-respected manufacturer, 'the closest approach to the original sound'?*" What does it mean to have 0.0001% distortion when it *still* doesn't sound like the original (and when the speakers are adding a further 2%)?

Distorted assumption

The tacit assumption in the design of so-called 'valve sound' musical instrument amplifiers is: "*Musicians like the sound of a valve amp. Valve amps may have more distortion. Therefore musicians like the distortion.*"

Manufacturers thus have produced special boxes which may introduce up to 25% second harmonic distortion and come up with a 'valve-sound' which some musicians like, but many describe as 'sounding more like a fuzz-box than an AC30.'

This seems to suggest that maybe we aren't just talking about distortion. And when we come on to studio monitoring amplifiers, we find that there is little significant difference in distortion figures between solid state and thermionic equipment. It must be remembered, however, that we are *not* suggesting that *all* valve amps are 'nice', or that *all* transistor amps are 'nasty'. The major problems inherent in transistor amplifier design are the direct result of developments made on valve circuits. Also, these problems, as will be seen, do not necessarily apply to low-power transistor amplifiers, as usually encountered in Hi-Fi systems, for example. The problems seem to arise in the method and amount of negative feedback application.

Feedback and its Misuse

There are two major ways of applying negative feedback to an amplifier: in the form of an overall feedback loop from one end of the amp to the other, or by the application of an individual feedback path per stage. Valve amplifiers like the Leak 'triple-loop feedback' series primarily used the latter method with little additional overall feedback (typically about 10dB). Certain other valve amps relied mainly on a high

VALVE SOUND

degree of overall feedback, producing a sound much closer to that of transistor amps. The main reason for the choice of overall feedback was probably economic: a glance at the handful of components underneath a certain well-respected Hi-Fi amplifier will indicate how cheap they must have been to produce. But it appears that it is the amps which used a feedback loop on each individual stage that produce the oft-mentioned 'valve sound'.

OK, you may well ask, what's so bad about high levels of overall feedback? Well, there are three main reasons, one well-known, the others totally unheard of.

T.I.D.

Almost everyone has heard of manufacturers recent efforts to eliminate the newly-discovered menace of Transient Intermodulation Distortion, but in case you missed the blurb, it may be broadly described as the type of distortion produced when the rising edge of a waveform applied to the amplifier input is rising faster than the total transit time (i.e. the time taken for a signal to pass through the amplifier). This may result in a spike of pure, 100% distortion, whose output power may well be limited only by the peak handling capacity of the PSU.

There are two ways of dealing with this. Either the initial stages of the amplifier must be capable of taking the maximum level of applied signal with no feedback applied, or the rising edges must be 'slowed down' to avoid this effect. This latter solution is the one generally used today, and is equivalent to placing a low-pass filter on the input to limit

the frequency response to something below the level of 'flat from DC to Light'.

TID was not so prevalent in certain valve amplifiers because of a) locally applied feedback on the stage itself, thus reducing the transit time (and perhaps electrons crawl more slowly across semiconductors than they travel *in vacuo*), and b) because the linear portion of the valve transfer curve is longer than the corresponding area of transistor characteristics.

The Ricochet Effect

A point often not considered in amplifier design is that as negative feedback across the amp is increased, the 'back-to-front-impedance' (ie the impedance from the output looking back up the feedback loop towards the input) is reduced. Some high-power transistor amps have loops with an impedance of only a few ohms. This is because transistors are generally low impedance devices.

But how does this affect the sound produced? Let us remember that amplifiers are most often used to drive loudspeakers which are placed in fortunately-not-normally-anechoic rooms. A loudspeaker can also be considered as a microphone, and as such it is evident that whilst sound is being projected into the room, it is also being picked up (after multiple bouncing around) by reflection and reconverted into voltages, which will appear at the amplifier input, if the back-to-front impedance is sufficiently low. Another way of looking at this would be to say that the loudspeaker is loaded by standing — and other pressure-waves in a non-linear manner with respect to frequency. The total effect is that the same source-/amplifier/speaker system will sound different in various locations,

whilst a system without this effect will vary only slightly according to room acoustics. We noticed this whilst working in a studio fitted with Tannoy Monitor Reds in Lockwood cabinets, (over)driven by well-known 300 W transistor power amps. The act of panning a signal from left to right changed its sound dramatically, far more than one would expect in an acoustically treated studio environment. Replacing the studio amps with our venerable bottles, however, removed the effect almost entirely, most other factors (including subjective SPL) remaining equal. The transistor amps, of course, had a high level of overall feedback, unlike the valve system.

Frequency Response

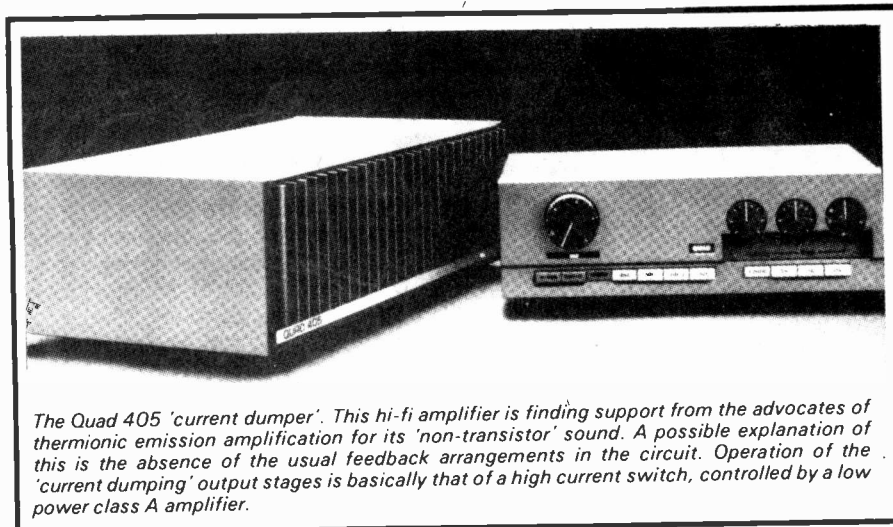
One problem with evaluating the Ricochet Effect and, to some extent, TID, is that they are very hard to measure objectively. You should be able to pick them out by listening critically, but using normal test procedures it will often be found that there is little discernible difference between 'nice' and 'nasty' sounding amps on the test bench. However, one point of comparison has emerged. It was found that a number of 'nasty' sounding amplifiers had a very limited frequency response with the feedback loop removed (one specimen managed 3 dB points at 250 Hz and 8 kHz approx., with 40 Hz and 15 kHz over 10 dB down!), whereas our 'nicest' amplifier was found to be nearly flat under similar conditions (3 dB points at 40Hz and 15 KHz approx.). However, this point required further investigation, and removing a feedback loop is not always as easy as one would wish!

As far as the ricochet effect is concerned, we have so far found no other satisfactory explanation for the phenomenon (try it and see!).

Conclusions

The result of a few years of subjective listening tests has indicated very firmly to us that the 'second harmonic distortion' explanation for the 'valve sound' IS LARGELY A FALLACY. And after all, we were always taught at college that push-pull output stages *cancelled out* even harmonic distortion. Where is it all coming from?

We feel that the major reason for *certain* valve amps sounding better than *certain* transistor amps is largely the result of the use of smaller amounts of overall feedback in the former, resulting in less TID and a reduction of what we have termed the hypothetical 'Ricochet Effect'.



The Quad 405 'current dumper'. This hi-fi amplifier is finding support from the advocates of thermionic emission amplification for its 'non-transistor' sound. A possible explanation of this is the absence of the usual feedback arrangements in the circuit. Operation of the 'current dumping' output stages is basically that of a high current switch, controlled by a low power class A amplifier.

Some transistor amps — mainly low power types — sound particularly good because they have better linearity, lower transit times, etc. and hence less 'Ricochet and 'less objectionable' TID (the 100% momentary distortion of TID is far more painful if the amplifier is capable of delivering 500 W than if it can only manage 2!)

And more conclusions

We have found that a well-designed, high quality valve amplifier can provide adequate power reserve. (remember that a Tannoy Red, for instance, only takes 35 W, not 300!) and distortion figures and is capable of reproducing the textural subtlety of the original performance with ease, whatever the apparent performance on non-typical test tones may indicate. After all, it is usually music we are listening to, with all its transients and non-cyclic waveforms, rather than what to us are rather boring sine waves. This aspect of 'high-fidelity' — the ability of the system to reproduce the subtle nuances of the original instruments — is something that few transistor amplifiers (with the possible exception of amps like the Quad 405, with its totally different feedback technique, the H/H monitoring amplifier with its special attention to TID, and the Vertical FET amplifier (Yamaha) which seems to exhibit all the 'niceness' of a valve amplifier), have shown themselves able to do.

Of course, a few people will contest our results, which are admittedly only preliminary: we say to them, listen for yourself and see. It is likely that more distortion will be caused by a closed mind than will be exhibited by the (valve) amplifier under examination!

*ABOUT THE AUTHORS:

George Chkiantz is a freelance sound recording engineer of long experience with many of the world's top recording groups, including Family, King Crimson, Led Zeppelin (including 'Whole Lotta Love'), Hawkwind and many others. He also received a Gold Disc for the Rolling Stones album 'It's Only Rock 'N' Roll.

Richard Elen is Studio Manager of KPM Sound Studios, Denmark Street, a part of EMI Music Publishing. He is quite clever as well.

Both are moderately sensible. ●

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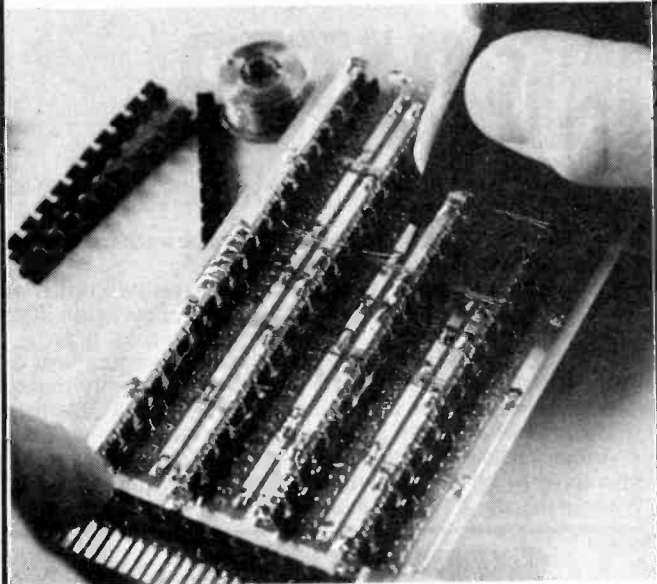
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Did you know that overwatering is worse than underwatering and it causes the edges of the leaves to go yellow? (too little water does exactly the same, just to cause confusion).

In these days of nitrogen, phosphates, paraquat, etc, sticking your finger into the earth of a pot plant is not only messy — it's also unscientific. One of August's Short Circuits describes a unit that shows on a row of LEDs if water is needed.

And it works! ETI's pot plants are all in fine shape now, in fact you have to be careful when walking past our outsized Venus fly-trap in case it takes a snap at you!



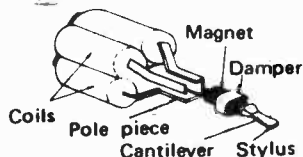
SWEEP GENERATOR

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3-TRANSISTOR RADIO

We don't do so many radio circuits in ETI but we've made an exception next month and will be describing a 3-transistor radio covering the medium waves and driving a loudspeaker. A nice little circuit which can be built up in a couple of hours and ideal for kids of all ages. (The only difference between men and boys is the price of their toys!)

PICKUP PRINCIPLES



THINGS have moved a long way since the steel needles were changed in your wind-up gramophone — OK you know that, but do you know how far things have come?

Next month ETI takes a look at pickup principles being used in modern audio systems.

ETI'S COVER PRICE

SORRY FOLKS, but the August issue will cost you 40p, an increase of 5p. As most of you will know, at ETI we have for a long time lagged our price behind those of our competitors; this has been possible due to our rapid circulation rise whilst most of the competition is slumping. We're still growing — faster than ever in fact — but the hairy Vikings, who make the paper that we're printed on, reckon that our pounds aren't worth what they used to be and want more of them. We can't blame them really but paper is the only thing a magazine doesn't save on with increased circulation and the extra is costing a bomb. We hope you'll still buy us.

FREE INSIDE THE AUGUST ISSUE: MARSHALLS CATALOGUE

The August issue's a real bumper — 116 pages. That's because Marshall's have chosen ETI to include a whopping great 32-page catalogue, bound into every copy of ETI.

Normally you have to pay for catalogues and this is the first time that we know of that any of the hobby-electronics magazines has been able to give one away.

The articles described here are in an advanced state of preparation but circumstances may necessitate changes in the issue that appears.

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MICROPROFILE

Gary Evans, ETI's resident microprocessor man, powers up microfile to report on a new personal computing system and Europe's first conference for home computer builders.

THE 'HOMEBREW' COMPUTER market in this country seems about to 'take off'. With sales of manufacturer's development kits on the increase, and signs that many firms are beginning to move into the hardware/software supply field, it seems that this sector of the hobby electronics market will be one of the major growth areas over the next few years.

This month we report on two recent events which reflect this growing interest in the home computing field. The interest ranges from the large manufacturers ready to sell complete systems, to the smaller supplier specialising in the provision of hardware and software for the D.I.Y. market.

Household Pet

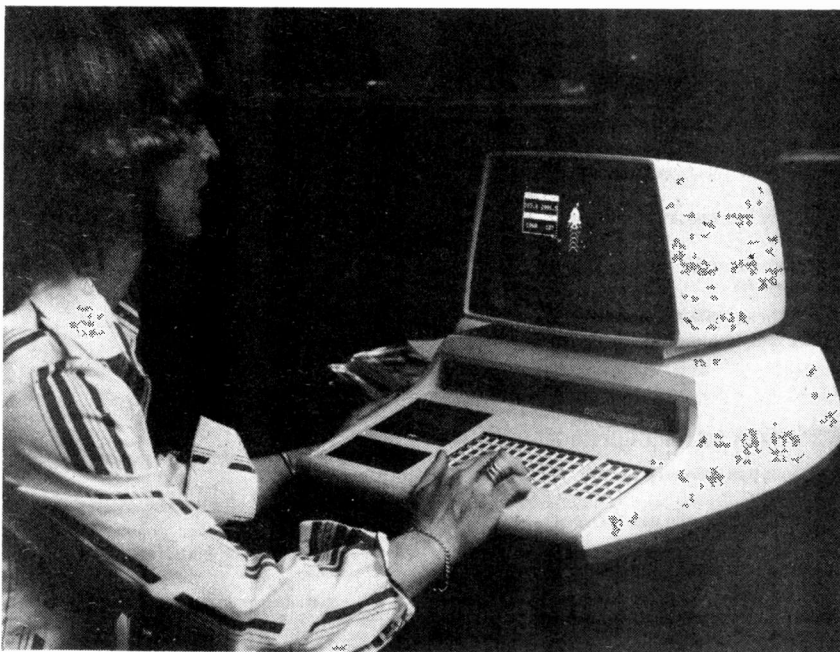
First off we look at CBM's announcement that they intend to launch their PET Model 2001 towards the end of this year. PET (Personal Electronics Transaction) is a compact and very versatile home computer system, complete with display screen, cassette interface and a calculator style keyboard.

The basic unit's display screen is a nine inch CRT capable of a 1000 character display arranged as 40 columns x 25 lines. The display has a standard 64 character ASCII set as well as 64 graphics related characters.

Input to PET is via a calculator style keyboard which allows the basic 64 characters to be entered without shift, while the shift control allows the graphics characters to be entered. Various screen editing and control functions are also available from the keyboard.

Model 2001 provides a cassette interface which uses standard audio cassette tapes to provide program and data storage for the system. The cassette storage system has a number of facilities which allow easy editing and management of user programs.

PET is based on the MOS Technology's (not MOSTEK) MOS 6502 microprocessor. This micro is popular in the States but has not been seen very much over here, although with the recent announcement that Rockwell are to second source this chip, the situation might change.



Our astronaut is seen landing her LEM with the help of PET's moon landing program. A wide variety of software is available for PET, covering applications in the home, for example storing recipes, to programs for small industrial users. The 9in. CRT with graphic handling capability, the calculator style keyboard, and cassette interface can be seen in our picture.

The system memory is provided by 12 K of ROM and 4 K of RAM (expandable to 32 K). The ROM is divided into two blocks. The first block of 4 K provides the operating system. This will allow Machine language accessibility and is capable of supporting multiple languages.

The resident high level language is BASIC, and is provided by an interpreter occupying the remaining 8 K of system ROM. CBM claim that this BASIC interpreter is 20% faster than other 8 K BASICs. The interpreter provides facilities for handling strings, integer and multiple dimension arrays, as well as providing a random number generator.

CBM hope to provide a large range of cassette based software for the system. The programs will cover such areas as video games, list management and inventory control. As well as CBM's software, there is a vast amount of material already available in BASIC language.

There are plans to produce peripherals that will extend the scope of PET's activities; floppy disks, printers and a MODEM are at present under development.

PET will even tell you when it is sick. A 'self-diagnostic' trouble light will tell the operator if any of the boards are not working properly.

The unit should be in production during the summer with a target production figure of 1 000 (yes one thousand) per day. The price in this country is expected to be - wait for it - £600.

At this price, not much more than a teletext decoder or a large colour-TV set, CBM see a vast potential in the home and small industrial markets. They are reducing their commitments to electronic calculators and plan to devote much of their energies to the manufacture and supply of PET.

About the only thing this PET will not do is come when you call it, and CBM may well be working on that.

MICROFILE

Doing It Your Way

Of more interest to the person who wants to build his own system was the first European D.I.Y. Computing Conference held in London during May. The event took place at the IEE and was organised by Online Conferences. More than 500 delegates attended, while many had to be turned away.

The conference featured speakers having a wide variety of experience with microprocessors, both in the professional and amateur fields.

The morning began with an introductory Teach-In which took delegates through the basic terminology and concepts that were to be developed during the conference. As the day progressed a wide variety of topics covering both the hardware and software aspects of home computing were discussed.

In addition to the formal conference activity, constructors had the chance to see a wide range of displays at a Constructor's Forum. This was held in the rooms surrounding the main hall.

Among the firms with their products on display were SINTEL, who were showing a new Z80 based system which they hope to market in kit form later this year. Amongst other interesting features this system has a very versatile VDU.

System 68 was seen in action on the Bywood Electronics stand, while nearby Penny and Giles were showing their new VDU kit. Unlike most VDU kits this comes complete with CRT, keyboard and attractive case.

At the conference Computer Workshop showed for the first time their four-terminal, multi-user system, developed in conjunction with Technical Systems Consultants, running a time-sharing BASIC software package.

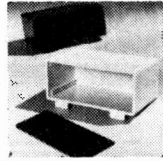
The system is aimed at educational establishments, and Computer Workshop think that it will fill a need that they have at a price they can afford.

The apparatus is to be manufactured in the U.K. with production scheduled to commence shortly.

This Time Next Year

The conference proved a great success and Online hope to make this an annual event. We just hope that any future conference can be held at a larger venue as, judging by the response to this first event, the interest in home computing is already large and growing every day.

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4017	1.12	4041	0.93	4068	0.24
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74C30	0.26	74C151	2.63	74C901	0.74
74C32	0.26	74C154	3.93	74C902	0.74
74C42	1.20	74C157	2.36	74C903	0.74
74C48	0.26	74C160	1.49	74C904	0.74
74C73	2.34	74C161	1.49	74C905	0.74
74C74	0.63	74C162	1.49	74C906	0.74
74C76	0.74	74C163	1.49	74C907	0.74
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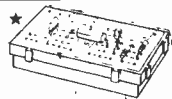
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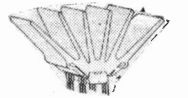
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ETI DATA SHEET

ICL 8038 WAVEFORM GENERATOR / V.C.O.

INTERSIL

The 8038 has been around for about 5 years — which is a long time in electronics. In fact it has reached the position of becoming an 'Industry Standard' on a par with the 741. An inherently versatile device it has its drawbacks like most chips — but overall has a lot going for it, Intersil even produced a very honest application bulletin (AO13) called 'Everything you always wanted to know about the 8038', which explained how to get the best out of this device and admitted its defects — an uncommon event with most manufacturers! Some of the data from AO13 has been included in this data sheet, but for fuller information ask Rapid Recall for AO13, AO12 and the latest information sheet (strangely referenceless, but brown in colour!)

Description

The 8038 Waveform Generator is a monolithic integrated circuit, capable of producing sine, square, triangular, sawtooth and pulse waveforms of high accuracy. The frequency (or repetition rate) can be selected externally over a range of less than 1/1000 Hz to more than 1 MHz and is highly stable over a wide temperature and supply voltage range. Frequency modulation and sweeping can be accomplished with an external voltage and the frequency can be programmed digitally through the use of either resistors or capacitors. The Waveform Generator utilizes advanced monolithic technology, such as thin film resistors and Schottky-barrier diodes.

Theory of operation

A block-diagram of the waveform generator is shown in Figure 1. An external capacitor C is charged and discharged by two current sources. Current source #2 is switched on and off by a flip-flop, while current source #1 is on continuously. Assuming that the flip-flop is in a state such that current source #2 is off, then the capacitor is charged with a current I. Thus the voltage across the capacitor rises linearly with time. When this voltage reaches the level of comparator #1 (set at 2/3 of the supply voltage), the flip-flop is triggered, changes states, and releases current source #2. This current source normally carries a current 2I, thus the capacitor is discharged with a net-current I and the voltage across it drops linearly with time. When it has reached the level of comparator #2 (set at 1/3 of the supply voltage), the flip-flop is triggered into its original state and the cycle starts anew.

Four waveforms are readily obtainable from this basic generator circuit. With the current sources set at I and 2I respectively, the charge and discharge times are equal.

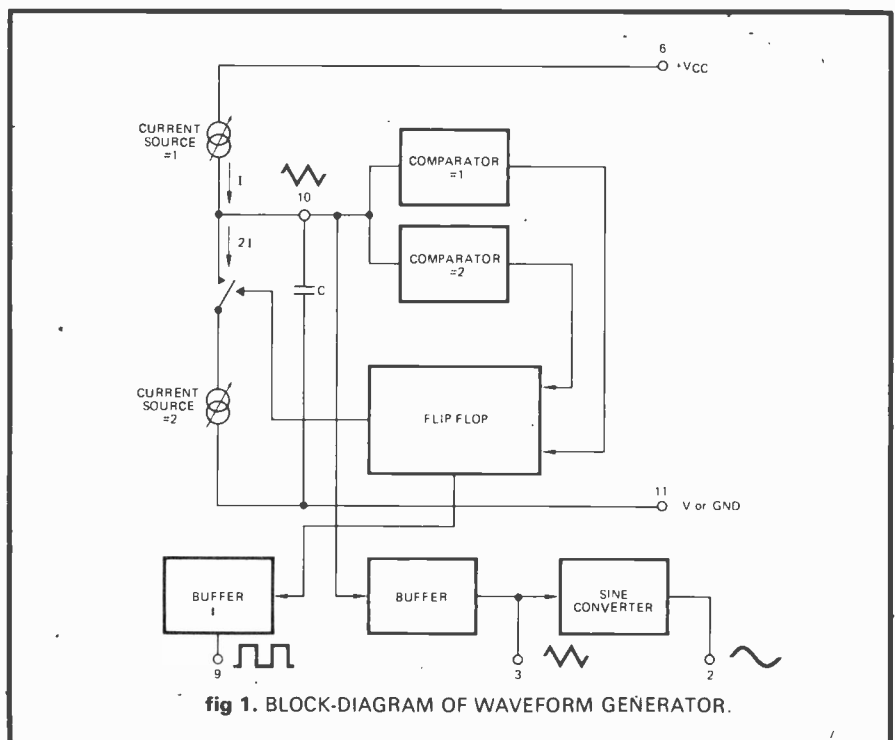
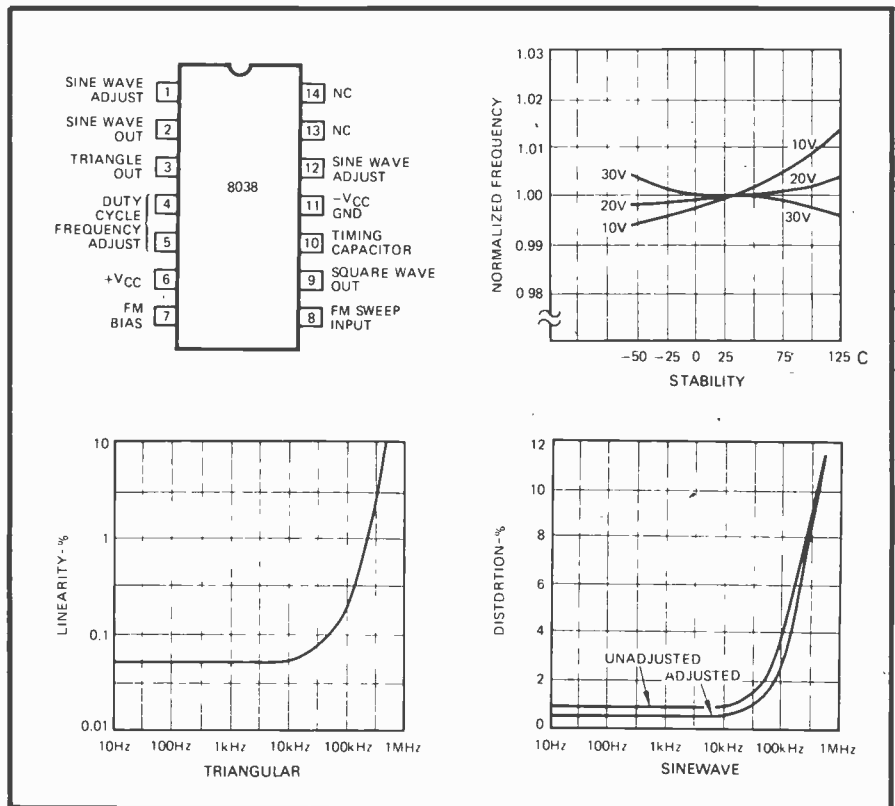


fig 1. BLOCK-DIAGRAM OF WAVEFORM GENERATOR.

Thus a triangle waveform is created across the capacitor and the flip-flop produces a square-wave. Both waveforms are fed to buffer stages and are available at pins 3 and 9.

The levels of the current sources can, however, be selected over a wide range with two external resistors. Therefore, with the two currents set at values different from 1 and 21, an asymmetrical sawtooth appears at terminal 3 and pulses with a duty cycle from less than 1% to greater than 99% are available at terminal 9.

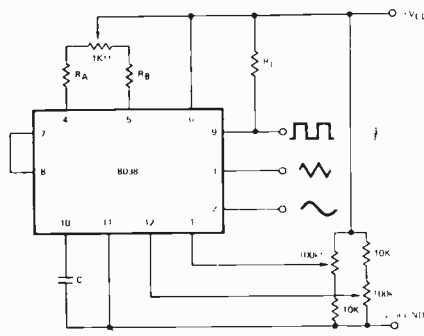
The sine-wave is created by feeding the triangle-wave into a non-linear network (sine-converter). This network provides a decreasing shunt-impedance as the potential of the triangle moves toward the two extremes.

Power Supply

The waveform generator can be operated either from a single power supply (10 to 30 Volts) or a dual power supply (± 5 to ± 15 Volts). With a single power supply the average levels of the triangle and sine-wave are at exactly one-half of the supply voltage, while the square wave alternates between +V and ground. A split power supply has the advantage that all waveforms move symmetrically about ground.

Also notice that the square wave output is not committed. The load resistor can be connected to a different power supply, as long as the applied voltage remains within the breakdown capability of the waveform generator (30 V). In this way, for example, the square-wave output can be made TTL compatible (load resistor connected to +5 Volts) while the waveform generator itself is powered from a much higher voltage.

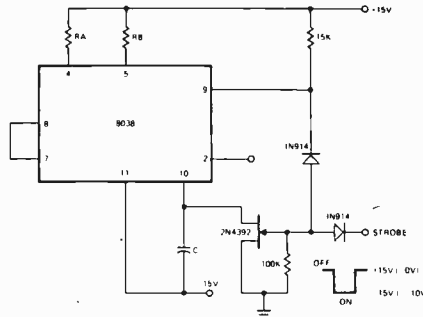
Purity



The symmetry of all waveforms can be adjusted with the external timing resistors. To minimize sine-wave distortion the resistors between pins 11 and 12 are best made variable ones. With this arrangement distortion of less than 1% is achievable. To reduce this even further, two potentiometers can be connected as shown. This configuration allows a reduction of sinewave distortion close to 0.5%.

Both the sine-wave and triangular outputs, are only useful up to about 20kHz if a reasonably pure signal is required. A perusal of the graphs will show why.

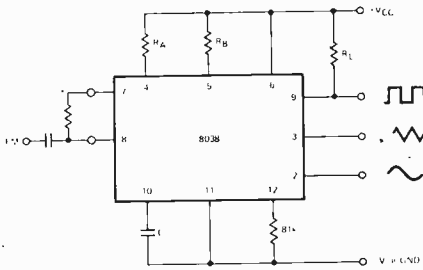
Strobe



With a dual supply voltage (e.g., $\pm 15V$) the external capacitor (pin 10) can be shorted to ground so that the sine wave and triangle wave always begin at a zero crossing point. Random switching has a 50/50 chance of starting on a positive or negative slope. A simple AND gate using pin 9 will allow the strobe to act only on one slope or the other.

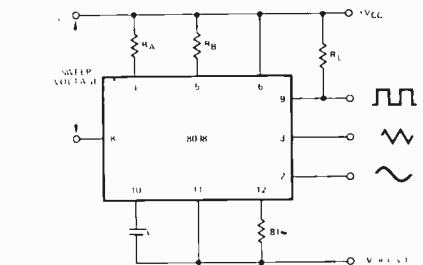
Using only a single supply, the capacitor (pin 10) can be switched either to $V+$ or ground to force the comparator to set in either the charge or discharge mode. The disadvantage of this technique is that the beginning cycle of the next burst will be 30% longer than the normal cycle.

F.M. and Sweeping



The frequency of the waveform generator is a direct function of the DC voltage at terminal 8 (measured from +VCC). Thus by altering this voltage, frequency modulation is achieved.

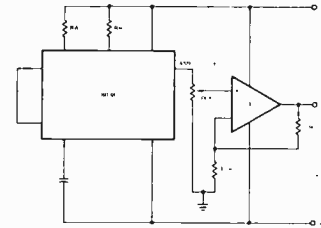
For small deviations (i.e. $\pm 10\%$) the modulating signal can be applied directly to pin 8, merely providing dc decoupling with a capacitor. An external resistor between pins 7 and 8 is not necessary, but it can be used to increase input impedance. Without it (i.e. terminals 7 and 8 connected together), the input impedance is 8k, with it, this impedance increases to $(R+8k)$.



For larger FM deviations or for frequency sweeping, the modulating signal is applied between the positive supply voltage and pin 8. In this way the entire bias for the current sources is created by the modulating signal and a very large (e.g. 1000:1) sweep range is created ($f = 0$ at $V_{sweep} = 0$). Care must be taken, however, to regulate the supply voltage; in this configuration the charge

current is no longer a function of the supply voltage (yet the trigger thresholds still are) and thus the frequency becomes dependent on the supply voltage. The potential on pin 8 may be swept from V_{CC} to about $2/3 V_{CC}$.

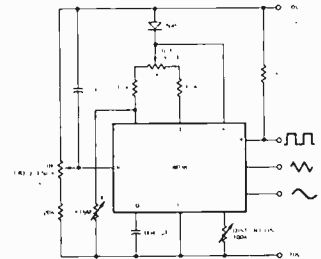
Buffering



The sine wave output has a relatively high output impedance (1K Typ). The circuit provides buffering, gain and amplitude adjustment. A simple op amp follower could also be used.

If the available outputs are all fed through a buffer, extra resistors can be inserted in series with the signal before a switch. Values of 47k (square wave), 15k (triangular) and 10k (sine wave) will ensure equal amplitude signals.

Audio Oscillator



To obtain a 1000:1 Sweep Range on the 8038 the voltage across external resistors RA and RB must decrease to nearly zero. This requires that the highest voltage on control Pin 8 exceed the voltage at the top of RA and RB by a few hundred millivolts.

The Circuit achieves this by using a diode to lower the effective supply voltage on the 8038. The large resistor on pin 5 helps reduce duty cycle variations with sweep. The range of this circuit is 20Hz to 20 kHz, output buffer can be added to make a general purpose bench unit.

Points to Note!

The 8038 runs hot to touch, this is normal, and is due to the resistive nature of the sinewave shaping network.

The optimum supply voltage, for minimum temperature drift is 20V, this can be seen in the stability graph.

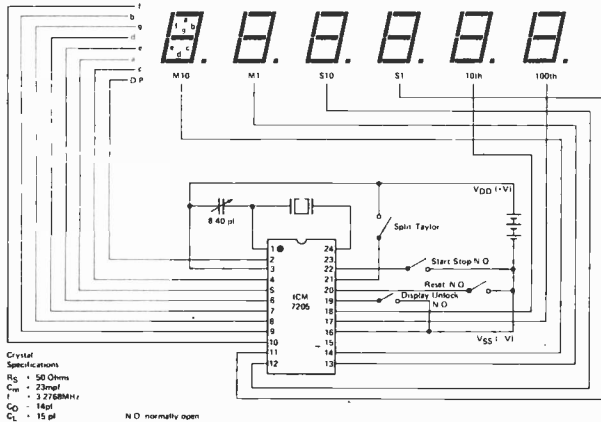
The 8038 is available from, **Rapid Recall Ltd., 9 Betterton Street, Drury Lane, London WC2H 9BS.** Price for 1 off is **£4.08** inclusive.

The Intersil ICM 7205 is a relatively new device, main points of interest are: on chip display drivers, fully protected against static - no special handling precautions required, average current of only 10 mA when in operation (including display!)

The ICM 7205 is a fully integrated CMOS six digit stopwatch circuit. The circuit interfaces directly with a six digit/segment common cathode LED display. The low battery indicator can be connected to the decimal point anode or to a separate LED lamp. The only components required for a complete stopwatch besides the display are: three SPST switches, a 3.2768 MHz crystal, a trimming capacitor, three AA batteries, and an on-off switch. For a two function stopwatch one additional switch would be required.

The circuit divides the oscillator frequency by 2¹⁵ to obtain 100 Hz which is fed to the fractional seconds, seconds and minutes counters. An intermediate frequency is used to obtain the 1/6 duty cycle 1.07 kHz multiplex waveforms. The blanking logic provides leading zero blanking for seconds and minutes independently of the clock. The ICM7205 is packaged in a 24 lead plastic DIP.

Stopwatch Circuit



Crystal Specifications
 R_s : 50 Ohms
 C₀ : 23pF
 f : 3.2768MHz
 C₁ : 16pF
 C₂ : 15 pF
 N/O normally open

Switch Characteristics

The ICM7205 is designed for use with SPST switches throughout. On the display unlock and reset inputs the characteristics of the switches are unimportant, since the circuit responds to a logic level held for any length of time, however short. Switch bounce on these inputs does not need to be specified. The Start/Stop input, however, responds to an edge and it requires a switch with less than 15 ms of switch bounce. The bounce protection circuitry has been specifically designed to let the circuit respond to the first edge of the signal, so as to preserve the full accuracy of the system.

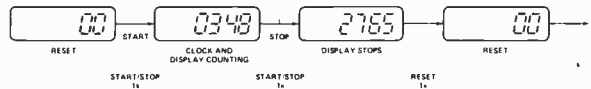
Low Battery Indicator

The on-chip low battery indicator is intended for use with a small LED lamp or with the decimal points on a standard LED display. The output is the drain of a P-channel transistor of approximately half the size of one of the segment drivers. The LBI circuitry is designed always to provide a voltage difference between the LBI trigger voltage and the minimum operating voltage, i.e. the lower the LBI trigger voltage the lower the the minimum operating voltage. In this way a stopwatch using three AA batteries will provide at least 15 minutes of accurate timekeeping after the LBI comes on.

Functional Operation

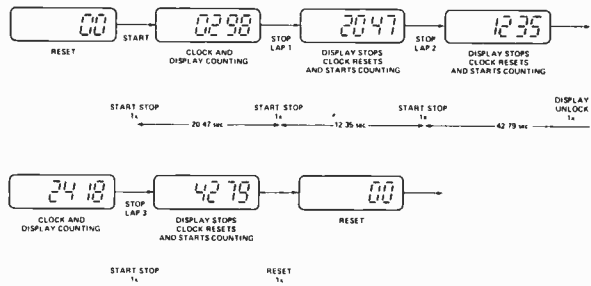
Turning on the stopwatch will bring up the reset state where the fractional seconds are on displaying 00 and the other digits are blanked. This display always indicates that the stopwatch is ready to go.

Start/Stop



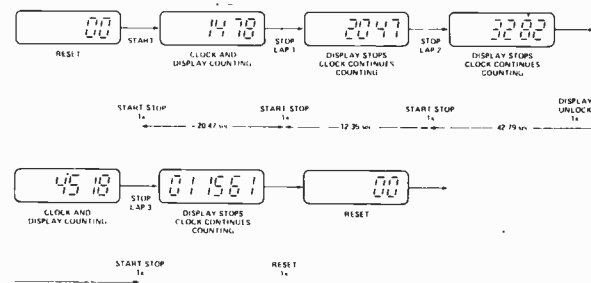
The Start/Stop modes can be used for a single event timing with the Split/Taylor input in either state. The illustration indicates the operations and the results. To time another event the reset switch must be used prior to the start of the event. Seconds will be displayed after one second, minutes after one minute. The range of the stopwatch is 59 minutes 59.99 seconds. If an event exceeds one hour, the number of hours must be remembered by the user. Leading zeros are not blanked after one hour.

Taylor



When the Split/Taylor input is left open circuit or is connected to Vss, the stopwatch can be used in the Taylor or sequential mode. As depicted graphically above, each split time is measured from zero in the Taylor mode, i.e. after stopping the watch, the counters reset to zero momentarily and start counting the next interval. The time displayed is the time elapsed since the last activation of Start/Stop. The display is stationary after the first interval unless the display unlock is used to show the running clock. Reset can be used at any time.

Split



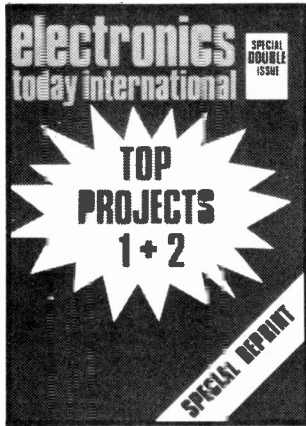
When the Split/Taylor input is connected to VDD the stopwatch is in the Split mode. The Split mode differs from the Taylor in that the lap times are cumulative in the Split mode. The counters do not reset or stop after the first start until reset is activated. Any time displayed is the cumulative time elapsed since the first start after reset. Display unlock can be used to let the display 'catch up' with the clock. Reset can be used at any time.

Points to Note!

Absolute maximum supply voltage is 5V5. Never short outputs to earth or low impedance power supply as this will destroy the device.

The ICM 7205 is available from: Rapid Recall Ltd., 9, Betterton Street, Drury Lane, London WC2H 9B5. Price for one off is £12.60 inclusive.

SPECIALS FROM ETI

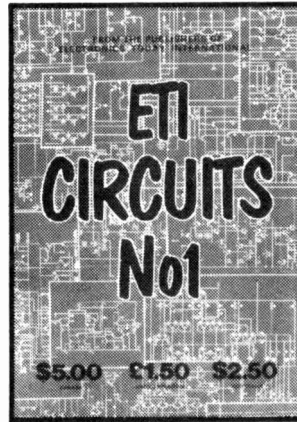


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Projects include: Master Mixer, 100W Guitar Amp, Low Power Laser, Printimer, Transistor Tester, Mixer Preamp, Logic Probe, Simple Amp, Ni-Cad Battery Charger, Loudhailer, Scope Calibrator, Electronic Ignition, Automatic Car Theft Alarm, Turn Indicator Cancellor, Brake Light Warning, LM380 Circuits, Temperature Alarm, Aerial Matcher, UHF TV Preamp, Metal Locator, Four Input Mixer, Super Stereo, IC Power Supply, Rumble Filter, IC Tester, Ignition Timing Light, 50W Stereo Amp PLUS MANY MORE.

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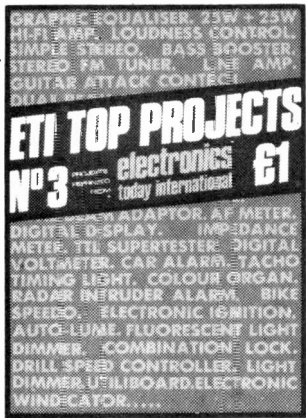


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Contains nearly 250 circuits, largely taken from the best of our Tech-Tips. Great care has been taken to index each circuit for rapid selection. An additional section at the back gives plenty of reference data including transistor specs and equivalents.

Sales of this publication have been excellent — hardly surprising when the circuits cost less than 1p each!

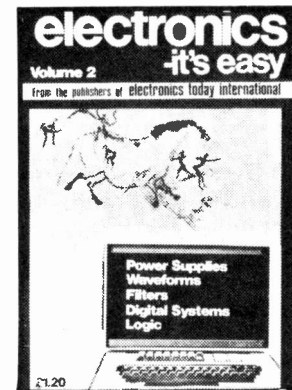
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TOP PROJECTS No. 3

Originally published in March 1976 Top Projects No. 3 contains 27 constructional projects including Graphic Equaliser International 25W Stereo Amp Simple Stereo, New Sound for your Guitar Bass, Booster, Line Amplifier Loudness Control Electronic Ignition, Tacho Timing Light Car Alarm, Dual-Beam Adaptor, AF Meter Impedance Meter Digital Display Digital Voltmeter TTL Supertester Fluorescent Light Dimmer Radar Intruder Alarm Light Dimmer, FM Tuner, Colour Organ Drill Speed Controller plus many more.

£1.00 + 20p P&P

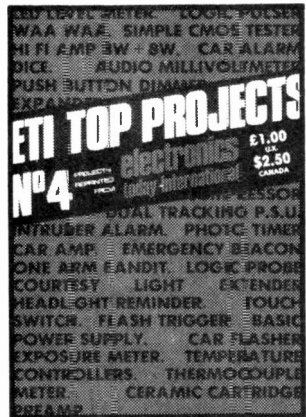


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Volumes 2 and 3 are available now (No. 3 completes the series and includes those parts still to be carried in the magazine) but Volume 1 sold out of its original print run and is currently being reprinted. Orders can be accepted now for immediate despatch upon delivery to us.

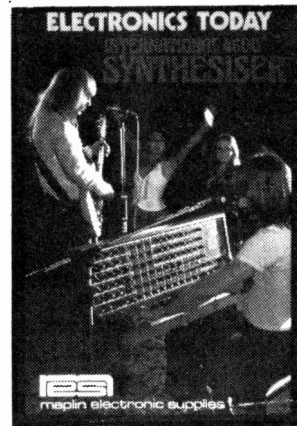
£1.20 + 20p P&P



TOP PROJECTS No. 4

Published October, 1976. This includes Sweet-Sixteen Stereo Amp, Waa-Waa, Audio Level Meter, Expander-Compressor, Car Anti-Theft Alarm, Headlight Reminder, Dual-Tracking Power Supply, Audio Millivoltmeter, Thermocouple Meter, Intruder Alarm, Touch Switch, Push-Button Dimmer, Exposure Meter, Photo Timer, Electronic Dice, High Power Beacon, Temperature Controller, Electronic One-Armed Bandit, plus many more.

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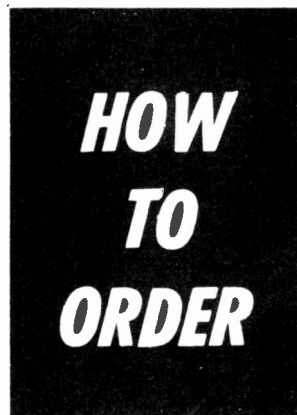
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Please mark the back of your cheque or PO with your name and address.



Described by John Miller-Kirkpatrick

LAST MONTH WE LEFT YOU with board A completed and checked by temporarily connecting the address generator outputs to the data inputs, all of these temporary connections are now removed and a ribbon cable connected to the points marked. The reason for using ribbon cable is for neatness and sanity, for connection of about 9 in. of 20-way cable is needed.

Assembly Board B

As with PCB A the first thing to do is insert all of the through hole links, solder and test for continuity. Sockets can then be installed in all IC locations and, if required, in the keyboard input connector. The 31-way connector should be installed at this stage and mounted on to the PCB with nuts and bolts through the holes provided.

The ribbon cable from PCB A is wired to connect the addresses from board A to B, the data from B to A and a power supply from B to A. A second cable carries the video signal, the video invert signal and ground to the front panel. With all of the soldering completed the ICs can be inserted and power applied to PCB B (and thus also to A).

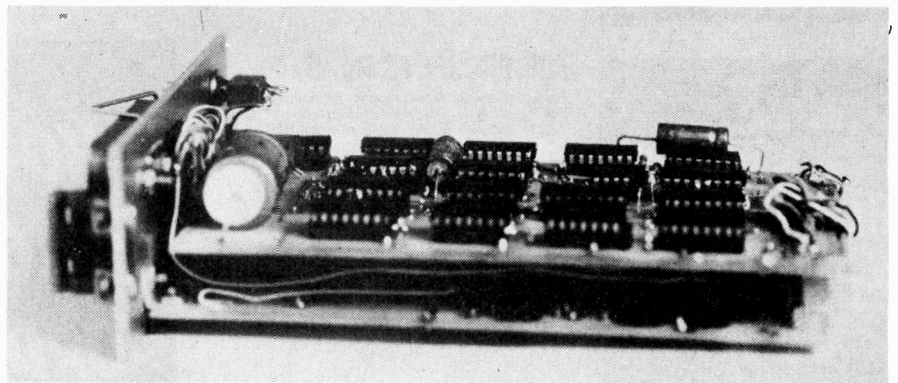
How it works and why it doesn't

When the power is applied the sync oscillator may be out of alignment and need adjustment, but leave the unit for about

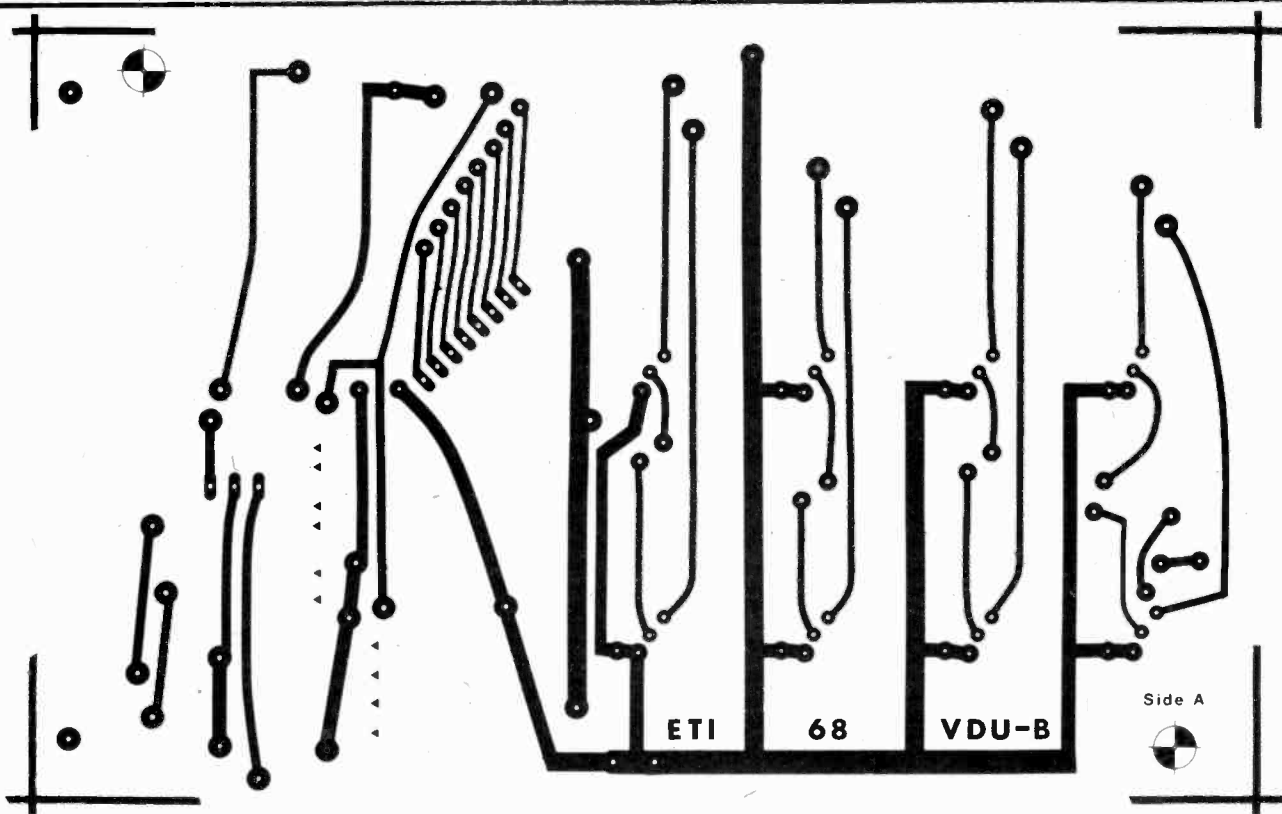
30 seconds before making any adjustment, as the oscillator seems to need a warm-up time. With the sync locked the display will probably show the same character in all locations on the screen. If you think about this it seems very unlikely that all of the locations of an 8Kbit RAM will turn on in the same state, the reason is quite simply that the screen is only getting one byte of data from the RAM. Although the address counters from board A are physically connected to the RAM they are directed via the 74C157s, and the select lines of these chips are currently open circuit. A logical '1' at pin one of the 74C157s or at pin 15 of the 31-way connector will cause the selectors to use the PCB A addresses to access the RAM. With an MPU connected to the VDU this is the state that the selectors are normally in. If the MPU wishes to

access the RAM it will signal that it wishes to do so by taking the VDU select line to logical '0'.

As each RAM location is addressed by board A so the data for that location becomes available on the data bus. One problem here is that the time difference between asking for and getting the data is typically 650nS, this is approximately the same time as it takes to display a character width. Thus a sequence from the first character is that address 000000000 is presented to the RAM when the count enable line goes high, as the counters have been reset for some time the RAM has had time to retrieve the data and present it to the character generator, thus if the first characters were ABCDE then an A will be displayed first. At the end of the first character time the address 000000001 is given to the RAM but during the 650nS before



View of the A and B boards ready for insertion into mainframe.



VDU board B component side foil pattern shown full size

the data for 'B' is available at the character generator the generator has data for an 'A' and thus the display now shows 'AA'. At some time near the end of the second character time the data for 'B' will become available at the character generator, but as the RAMs can react at different times different bits become true at different times, thus making the second half of the second character change to a 'B'. Similarly the third character will become a B/C, the fourth a C/D and so on up to the end of a character row. At the end of the line the display enable turns off the character display, but the characters keep going until the count enable halts the count. The data for the last character is not available until just before the display is disabled, but it is available for nearly 650nS after the display has been disabled.

Problems

There are thus two problems, the first to latch the data bits at a specific time during each character to stop the ragged change from one displayed character to another, the second problem is then to lose the double first character and retrieve the last.

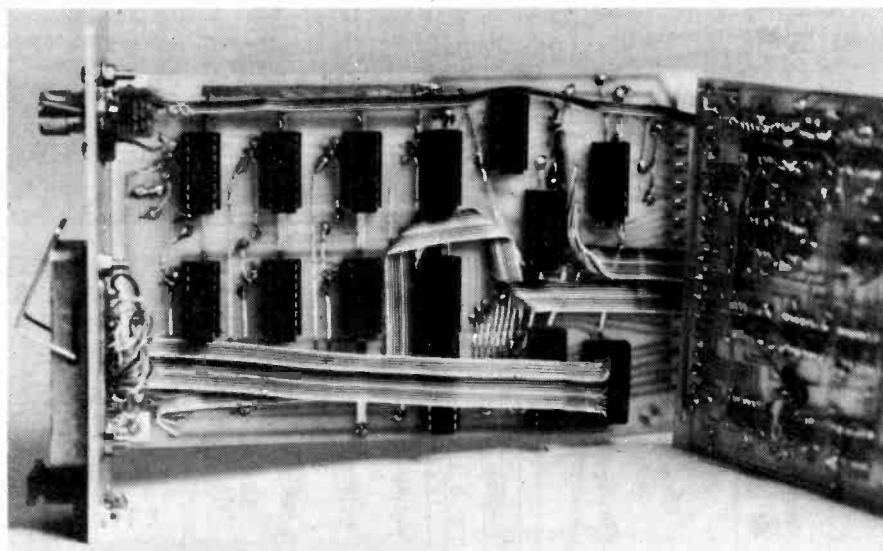
The DM8678 has an internal latch and can thus handle the latching for six data bits (if an 8679

is used in tandem then seven bits will be latched) leaving the extra data bits (if used) to be latched with the 7475. This latching was not shown in the circuit for PCB A but no additional ICs are required. Pin 4 of the DM8678 is connected to the inverse of the load signal on pin 10, ideally the store signal should occur one bit time before the load, but in practise it was found that the inverse of the load could be used. This is available at the output of IC11c.

To solve the second problem it would be useful if the whole display

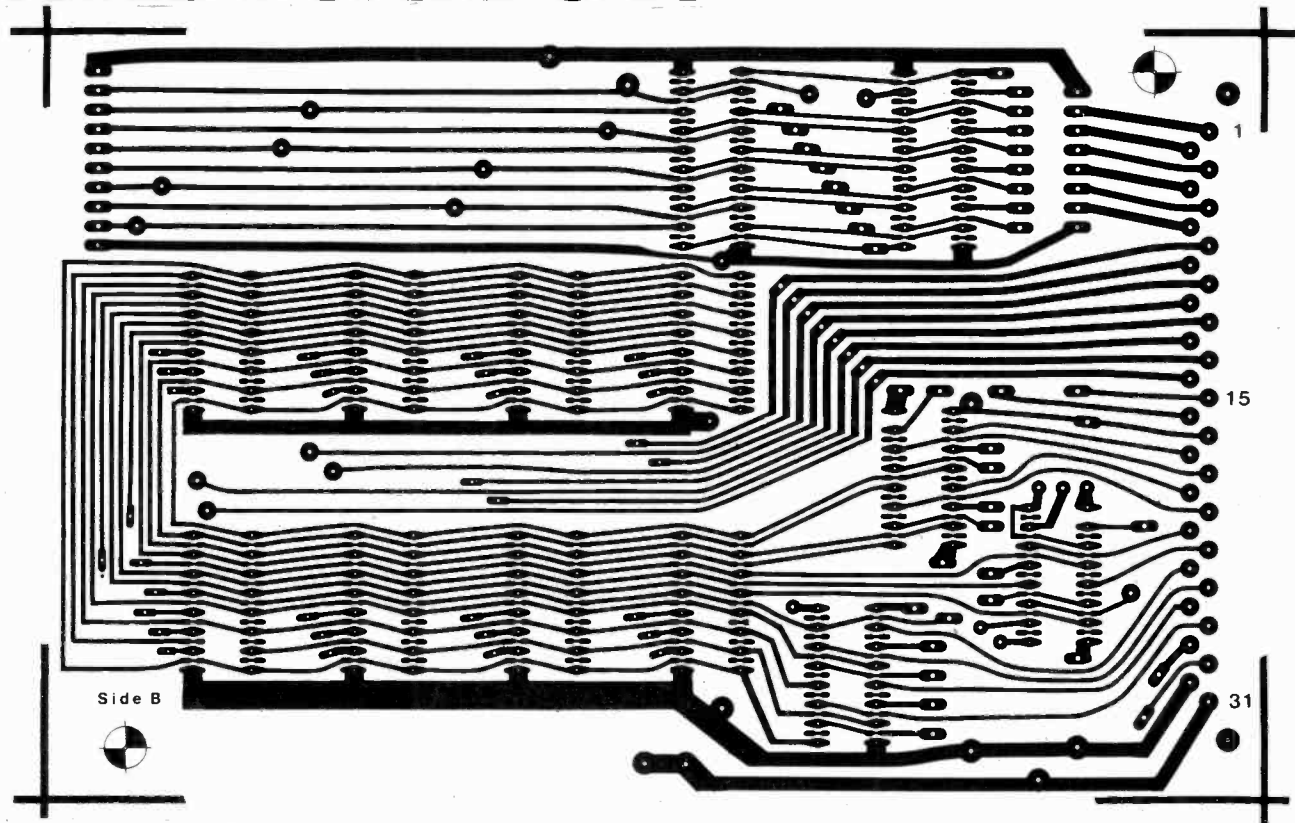
enable area could be moved one position to the right. A simple way to do this is to break the signal from IC7 to IC12 and process this via the 7475 latch. As this latch is operated at the same time as the DM8678 Load and Store it means that the display enable signal is delayed by about one character time. This means that the enable is now turned on at the start of the second character and off at the start of character time after the last character in the row.

If at this time the display is not nearly perfect then it is best to



The two boards opened out, to give general idea of cable layout.

VIDEO DISPLAY UNIT



VDU board B underside foil pattern shown full size

PARTS LIST

INTEGRATED CIRCUITS

IC 18-20 74C157
 IC 21-28 MM2102-2
 IC 29-28 DM81L595

MISCELLANEOUS

SW1 SPDT (316-709)*
 D1 1N4148*

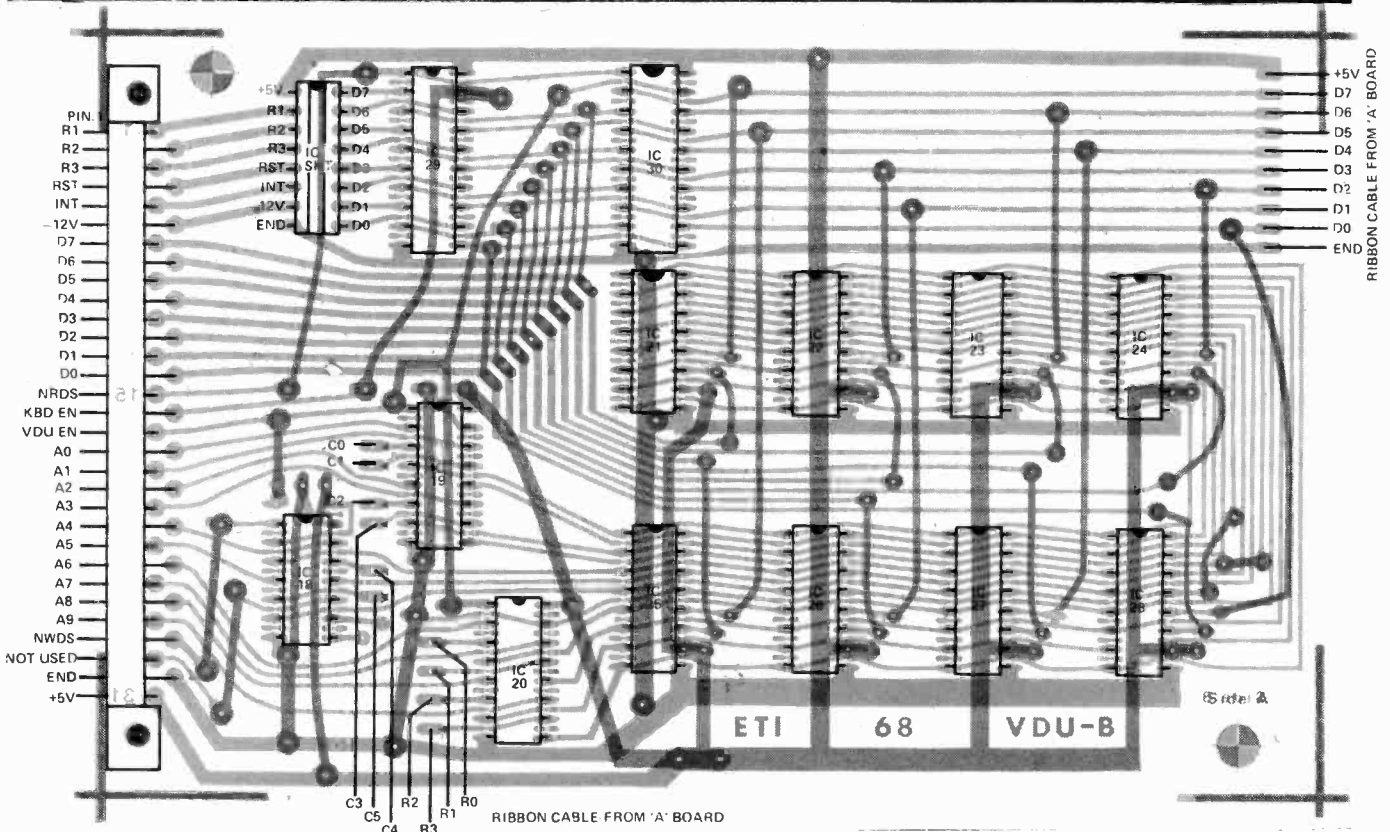
HARDWARE

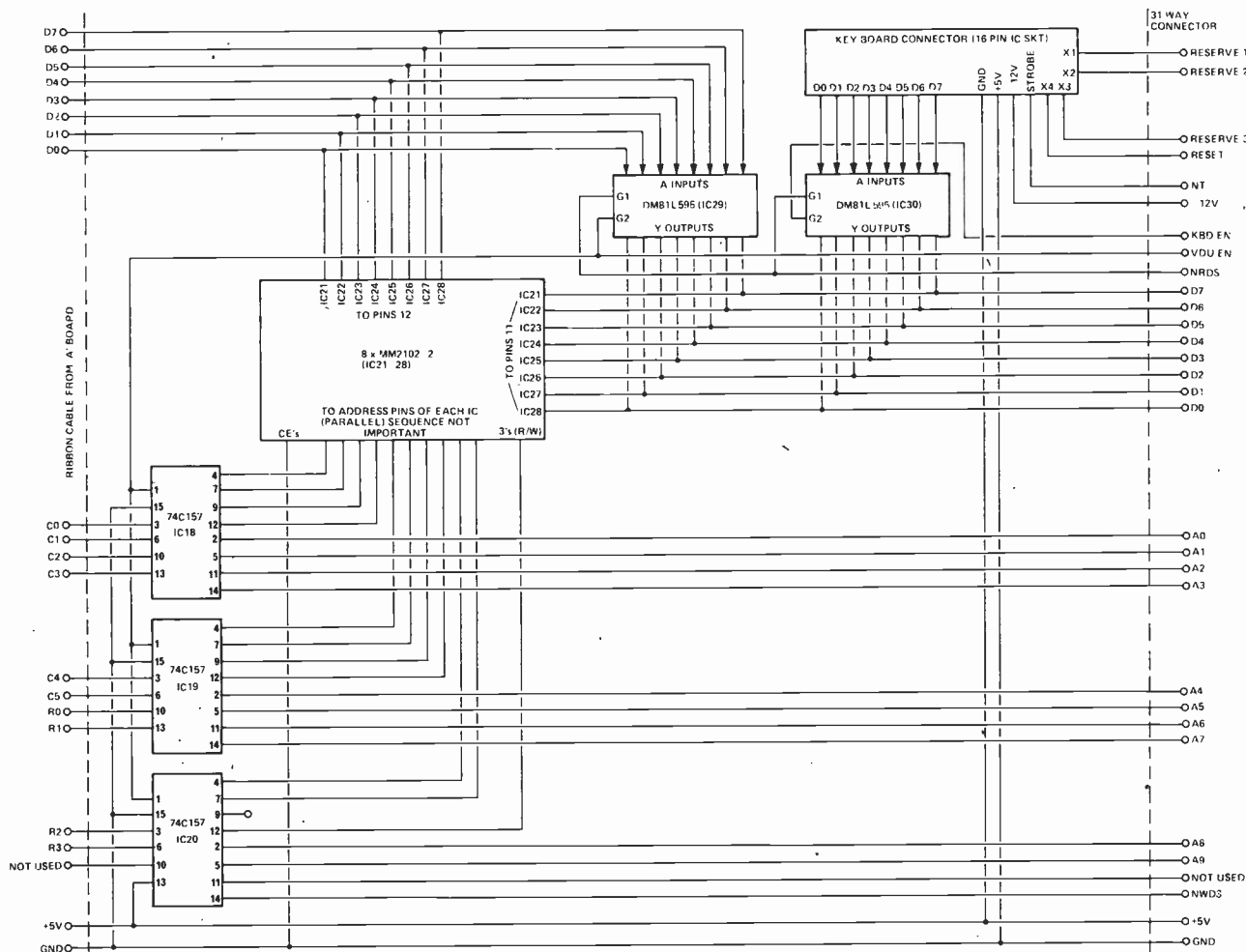
16-pin dil plug (401-841)
 IC socket strips (401-902) + holders (401-920)
 ETI 68VDU board B (Crofton)
 Terminal pins (433-860)

*Omitted from last month's parts list.

Doram (RS) numbers in brackets.

A complete kit of all parts (including pcbs) is available from **Bywood Electronics**.





Full circuit diagram of the B board, in conjunction with last month's A board forms complete VDU module.

assume race conditions in the ICs to be a problem. CMOS will not work much above 1MHz at 5v so check that there are no 74C series where there should be 74 series and vice-versa. It is also advisable to check for solder runs as we found a couple causing problems in the prototype. Problems with this type of equipment can be very difficult to cure without a very fast scope or a lot of thinking, thinking is usually cheaper than scopes so use the circuit to narrow down faults and they soon become apparent.

Keyboard interface

In addition to the VDU interface this PCB also allows an ASCII keyboard to be connected into the system, as VDUs and keyboards go together in most applications it makes sense to have them on one interface board. If you are not using the System 68 case you may prefer to connect

the keyboard into the PCB using a standard 16-pin socket and pin header/plug. With the System 68 case we use a 2 in. front panel module to carry the Video output socket, video invert switch and keyboard connector. We have used professional 25-way plugs and sockets wherever required as these tend to be an industry standard on this type of equipment. As the keyboard can be disconnected at the front panel the front panel connector can be wired into the 16-pin socket holes.

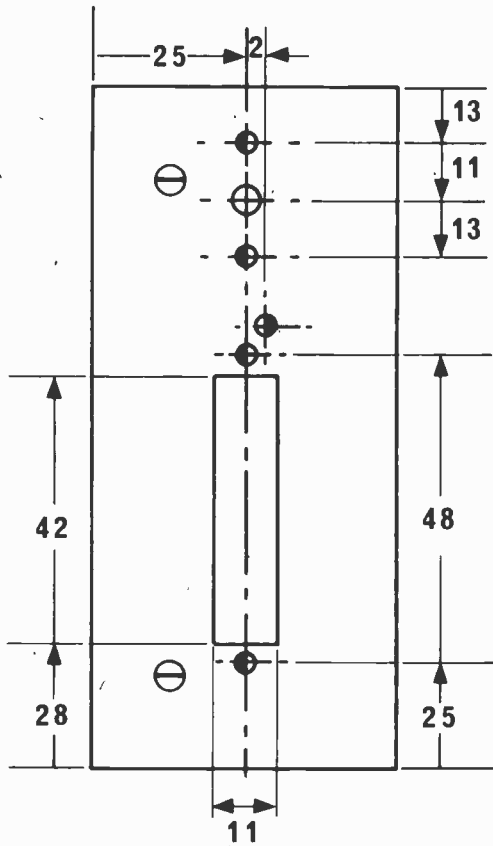
Whichever way you do it this connector carries 16 signals to/from the keyboard. We have allowed eight bit ASCII whereas a lot of keyboards have only six or seven bit output, the unused inputs can be connected to ground or to the keyboard strobe signal. Most keyboards feed on a diet of +5v, GND and -12v and so we

have provided for these lines to be keyboard. The other line we have to have is a strobe signal which should be negative going, ie, it is normally logic '1' unless a key is pressed in which case it goes to logic '0' until the key is released. We have four of our 16 leads unconnected, these are reserved for additional I/O signals to/from the keyboard. Such signals could be switches to the interrupt and/or reset pins of the MPU or could be status LEDs mounted on the keyboard.

Checking out

Checking out is very tedious without an MPU because you have to simulate the control signals and that involves a lot of logic state changing and testing. It's worth doing at this stage because you then know that the VDU element works. A check sequence is given.

VIDEO DISPLAY UNIT



Alternative positioning of sockets on front panel.

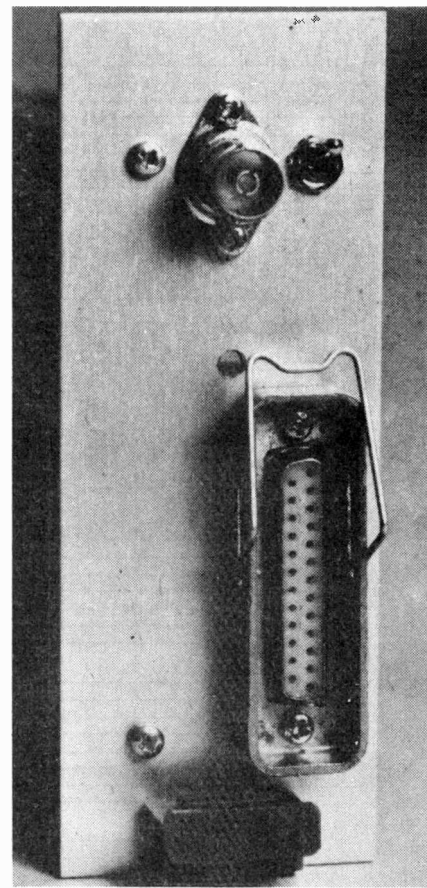
Folding up

Now is the time to fold up the two boards and check that they will fit in the case. Note that only PCB B is screwed to the front panel, A is held parallel to B by the card guides in the case, they should fit 1/2 in. apart and thus fit into adjacent card guides. If you have fitted your keyboard connector slightly off centre then you may find that the boards fit better 1 in. part, either is acceptable. Slide the unit into the case in the correct location and mark the position of the 31-way connector, slide the unit out, fit the 31-way socket, slide the unit home and feel that satisfying 'click' as it locks into place.

Next month

How to use the VDU with an MPU, data and address buffers and things.

- ⊙ DRILL 1 HOLE 2BA
- ⊕ DRILL 1 HOLE 4BA
- ⊙⊕ DRILL 4 HOLES 6BA



System 68 VDU Checkout

This checkout routine involves a couple of subroutines called TEST BUS; this involves checking each bit of the appropriate data bus, unless you wish to do the FULLCHECK. It is only necessary to convince yourself that no data bit is affecting another (solder bridge) and that all are connected at the appropriate points (open circuit, broken track).

	KBD	EN	VDU	EN	NWDS	NRDS	
START	1	1	1	1	1	1	TEST BUS AT VDU = BUS AT RAM
WRITE TO VDU	1	0	0	0	1	1	TEST BUS AT RAM = BUS AT INPUT
READ FROM VDU	0	1	1	0	0	0	Ditto
READ FROM KBD	0	1	1	1	0	0	TEST BUS KBD = BUS AT INPUT
FULLCHECK	1	0	0	0	1	1	ADDR AT INPUT = 00000000
LOOP 1							DATA AT INPUT = 10101010
READ	1	0	1	0	0	0	DATA AT RAM = 10101010
WRITE	1	0	0	1	0	1	DATA AT INPUT = 01010101
READ 2	1	0	1	0	0	0	DATA AT RAM = 01010101

GOTO LOOP 1 after incrementing INPUT ADDR by 1

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RESISTOR CODES PART 11

An explanation of the colour codings on all types of resistors

THE VALUE AND TOLERANCE, and other pertinent characteristics, of resistors may be marked on the body of the component in one of three ways. Viz:

(1) By marking directly on the body.
 (2) By using a standard colour code — coloured bands or dots, etc, read in sequence.

(3) By using an appropriate typographic code, consisting of letters and numerals arranged according to a convention.

Which method is used depends on the type and physical size of the component to a large extent and also according to the manufacturer's preference. The larger components, such as power resistors (particularly wire-wound types), usually have the value, tolerance and wattage rating marked directly on the body. Most common low power resistors, from 0.05 W to 2 W, use the standard resistor colour code. Some manufacturers use a typographic code on their resistors, physical size allowing (usually radial-lead types having wattage ratings between 0.25 W and 10 W). The special resistors (PTC, NTC thermistors and Varistors) also may be marked with a colour code or typographic code to indicate their value and characteristics.

The Standard Colour Code and Markings

The common axial-lead, composition and film-type resistors are marked with a series of coloured bands, as shown in Figure 1, which are read according to the standard colour code table in Table 1. The standard E24 (5%), E12 (10%) and E6 (20%) series components are marked with either three or four bands. Components below 10 ohms in the E6 series may have only two bands indicating the value. Resistor values in the E48 (2%) and E96 (1%) series are marked with five bands.

The bands are located on the component towards one end. If the resistor is oriented with that end towards the left, the bands are read from left to right as shown. The extreme left (or first) band colour indicates the value of the first digit of the component value; the next, or second, band indicates the

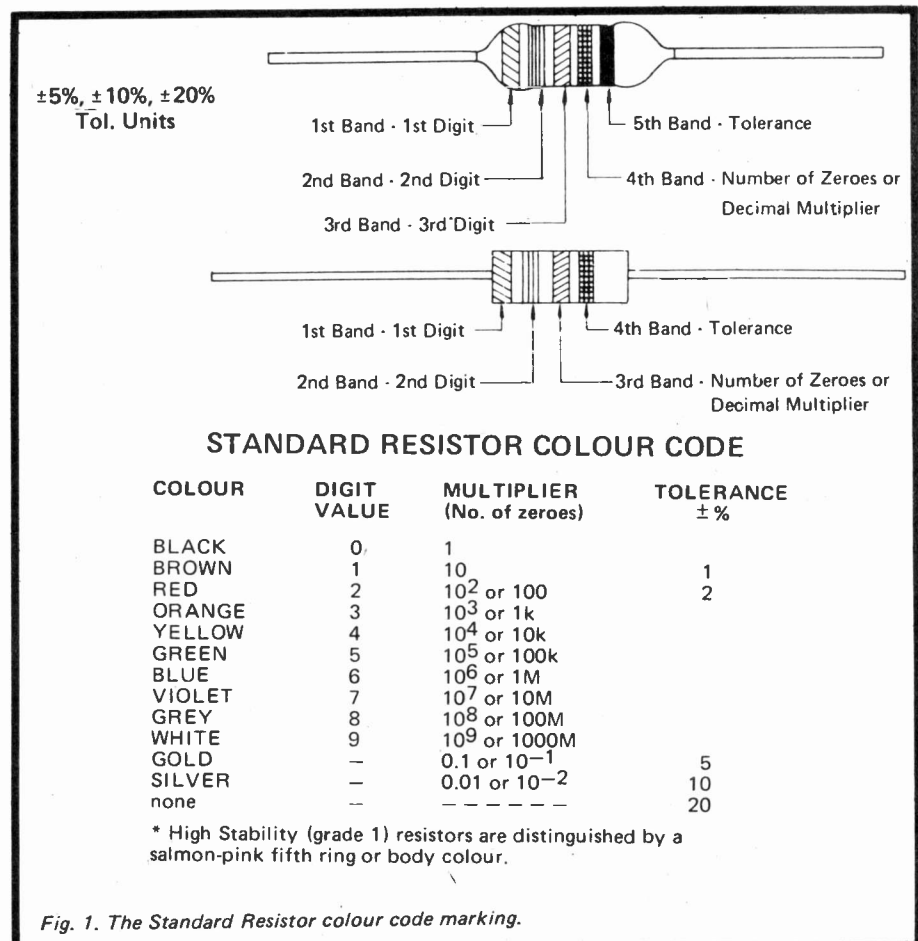
second digit of the value and so on. If the bands are not clearly oriented towards one end of the resistor it is best sorted out by trying to locate the tolerance band first. As the most commonly used resistors these days are either E12 or E24 series, the tolerance band is either silver or gold respectively. If still in doubt — resort to an ohmmeter.

The body colour of modern resistors is also used to indicate the resistor type. Carbon film resistors have a very light tan body, and carbon composition resistors have a medium tan body — somewhat darker than the carbon film body colour. Metal film resistors have a brown body colour — quite distinguishable from composition resistors and metal-glazed film resistors have a light blue body colour.

High stability resistors (E48, E96, E192 series) are distinguished by salmon-pink 5th band or body colour. For those who have difficulty remembering the resistor colour code, Table 2 lists the most commonly used values in the E12 series, between 4R7 and 2M2.

Old-Style Resistors

Prior to the standardisation of the banded system of resistor marking, resistors were colour coded with their value and tolerance by either one of two systems. These were the "Body-End-Dot" and the "Body-End-Band" systems, which are illustrated in Figure 2 (a) and (b) respectively. The body colour represents the first digit of the resistor value, the end colour the second digit, the dot or band colour, the multiplier. The tolerance was indicated



RESISTOR CODES

by a coloured spot which partially covered the end of the resistor opposite the 'end' colour or a band much narrower than the 'end' colour. In the body-end-dot system, the dot was generally located midway along the body. In the body-end-band system the band was generally located closer to the 'end' colour. Omission of the tolerance colour indicated a tolerance of $\pm 20\%$.

Some other manufacturers indicate the component value and tolerance by a series of dots or small bands which do not completely encircle the resistor body. This system of marking is commonly used on radial-lead and upright mounting styles of resistor from some manufacturers (particularly Erie Co., and some Japanese firms); these are illustrated in Fig. 2c. With the upright mounting style of resistor, the colour code is located towards the upper end of the body. The colour closest to the upper end indicates the first digit of the value; the next colour down, the second digit and so on.

Typographic Codes and Markings

Resistors may be marked with a combination of letters and figures to indicate the value, and tolerance. Alternatively a combination of direct marking and typographic code may be employed.

The typographic codes used are illustrated in Figure 4. A series of three letters, R, k, M, are used to indicate multipliers of x1, x1000 and x1000 000. The significant figures of the value are indicated directly with figures, the position of the multiplier indicating the decimal point. For example:—

4R7	= 4.7 ohms
330R	= 330 ohms
5k6	= 5.6k (5600 ohms)
68K	= 68k (68,000 ohms)
1M8	= 1.8M (1.8 megohms)
22M	= 22M (22 megohms)

The tolerance is indicated by one of five letters (see Figure 4) which immediately follow the value code on com-

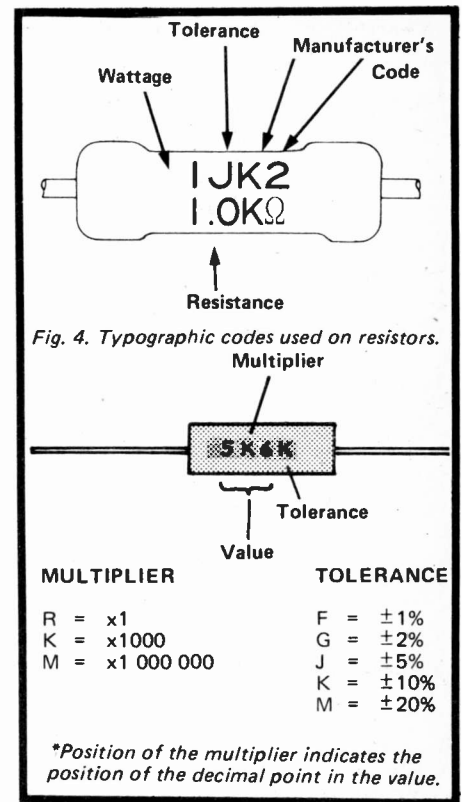
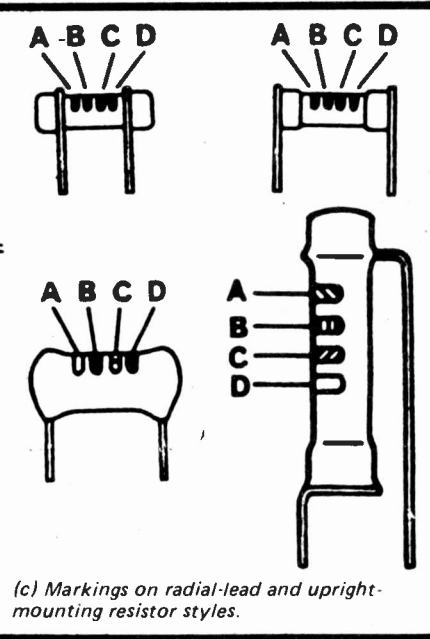
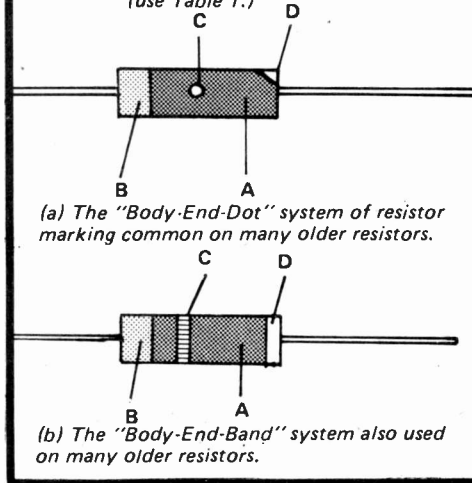


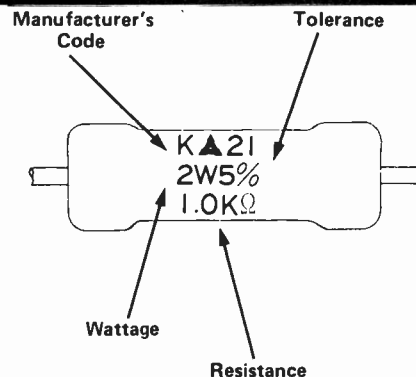
Fig. 2. Other styles of resistor colour code markings.

- A: 1st Digit
- B: 2nd Digit
- C: Multiplier
- D: Tolerance (use Table 1.)



Direct Marking

This style of marking a resistor is commonly used on power resistors (usually from 2 W), wirewound and precision resistors. It usually includes a manufacturer's code indicating the type of resistor perhaps including a date code indicating when the component was manufactured. Figure 3 illustrates a 1 k, $\pm 5\%$, 2 W resistor.



ponents which are marked completely with a typographic code. Some examples of the complete code are as follows:

2k2F	= 2.2k, $\pm 1\%$
120kG	= 120k, $\pm 2\%$
2M2J	= 2.2M, $\pm 5\%$
150RK	= 150 ohm, $\pm 10\%$
6R8M	= 6.8 ohm, $\pm 20\%$

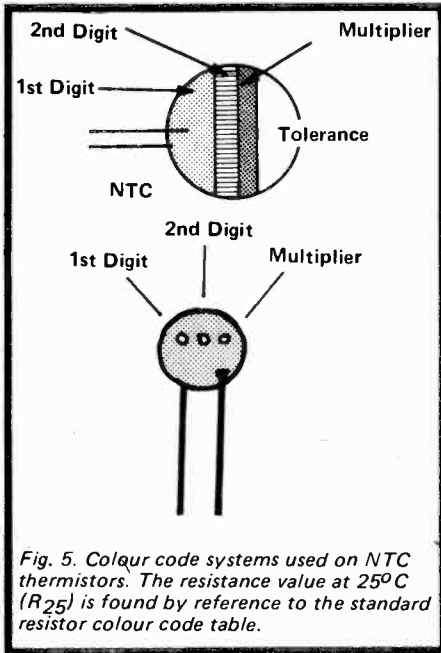
THERMISTOR MARKING CODES

Thermistors may be marked with a colour code or a typographic code, or may have no markings at all! The manner in which they are marked depends largely on their construction and the preference of the manufacturer. NTC thermistors may be marked with either a colour code or typographic code (or not at all) but PTC thermistors are marked with a typographic code only — when they are marked!

Whatever marking is employed, the resistance value at 25°C (R_{25}), and its tolerance at that temperature (if included) are generally the basic characteristics indicated. Other parameters (such as the B value) may be indicated when a typographic code is employed. The manufacturer's data should be consulted for the complete thermistor characteristics.

Colour Coded NTC Thermistors

Two basic methods of colour coding NTC thermistors are used, illustrated in Figure 5. The value of R_{25} is found by reference to the standard resistor colour



code table. The tolerance is sometimes omitted. The marking method illustrated on the left in Figure 5 distinguishes NTC thermistors from varistors (see Figure 8).

Typographic Coded NTC Thermistors

The typographic code occasionally employed on NTC thermistors is illustrated in Figure 6. This code is from the American EIA system of component designation. The tolerance range of NTC thermistors extends from $\pm 5\%$ to $\pm 40\%$ and two extra letters are added to the standard typographic tolerance code. The temperature constant B, is also indicated with the typographic code and reference to the manufacturer's data for the basic parameters is not necessary. However, if the dissipation, wattage rating, etc. are needed then the manufacturer's data will need to be consulted.

The typographic code consists of a prefix which may be 'ERT' to indicate and NTC thermistor or simply NTC. The value and characteristics may follow immediately or a manufacturer's code may precede it (usually indicating component type). However, the characteristics are always the last group.

TOLERANCE CONSTANT(B), $^{\circ}\text{K}$ (at 25°C)	
J = $\pm 5\%$	A = up to 1000
K = $\pm 10\%$	B = 1000-1500
L = $\pm 15\%$	C = 1501-2000
M = $\pm 29\%$	E = 2501-3000
R = $\pm 40\%$	F = 3001-3500
	G = 3501-4000
	H = 4001-4500
	I = 4501-5000
	J = 5001-5500
	K = 5501-6000
	L = over 6001

PTC Thermistor Marking Codes

The typographic code that may be used on PTC thermistors is from the EIA system code, illustrated in Figure 7. The prefix ERP indicates that the component is a PTC thermistor. The suffix is divided into three portions. The first consists of a letter and a numeral indicating the prime characteristic of the component. If it is an A-type PTC thermistor the temperature coefficient is indicated, as shown in the accompanying table. If it is a B-type, which changes resistance abruptly at a specified temperature (the 'switching' temperature), then the switching temperature is indicated as shown in the Table.

The tolerance and the resistance at 25°C (R_{25}) follow, and are read off in the same way as for NTC thermistors — see Figure 6.

PTC thermistors are often not marked, but their packaging may contain the above typographic code along with a manufacturer's component code.

Varistor Marking Codes

Both colour and typographic codes are used to mark varistors. As they are voltage dependent devices, the voltage value and its tolerance are given. The colour code that is used on ZNR and SiC varistors is illustrated in Figure 8. The value and tolerance is found from the standard colour code table (see section on Component Marking Codes). The tolerance is the first band on these components when held with the colour bands at the left as illustrated. Just to confuse matters, some manufacturers use the 1st, 2nd and 3rd digit bands to indicate the last three digits of their type number!

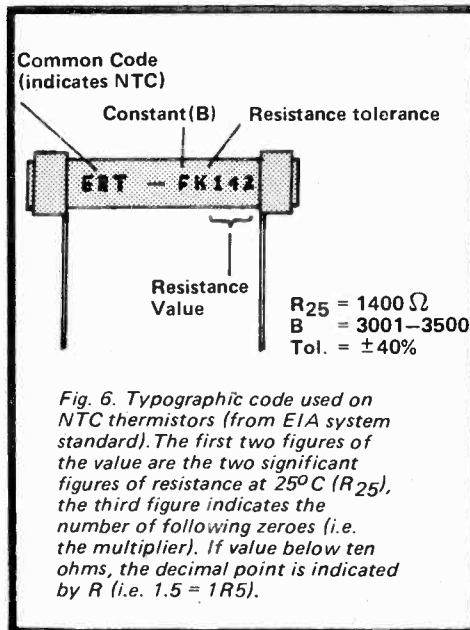


Fig. 6. Typographic code used on NTC thermistors (from EIA system standard). The first two figures of the value are the two significant figures of resistance at 25°C (R_{25}), the third figure indicates the number of following zeroes (i.e. the multiplier). If value below ten ohms, the decimal point is indicated by R (i.e. $1.5 = 1R5$).

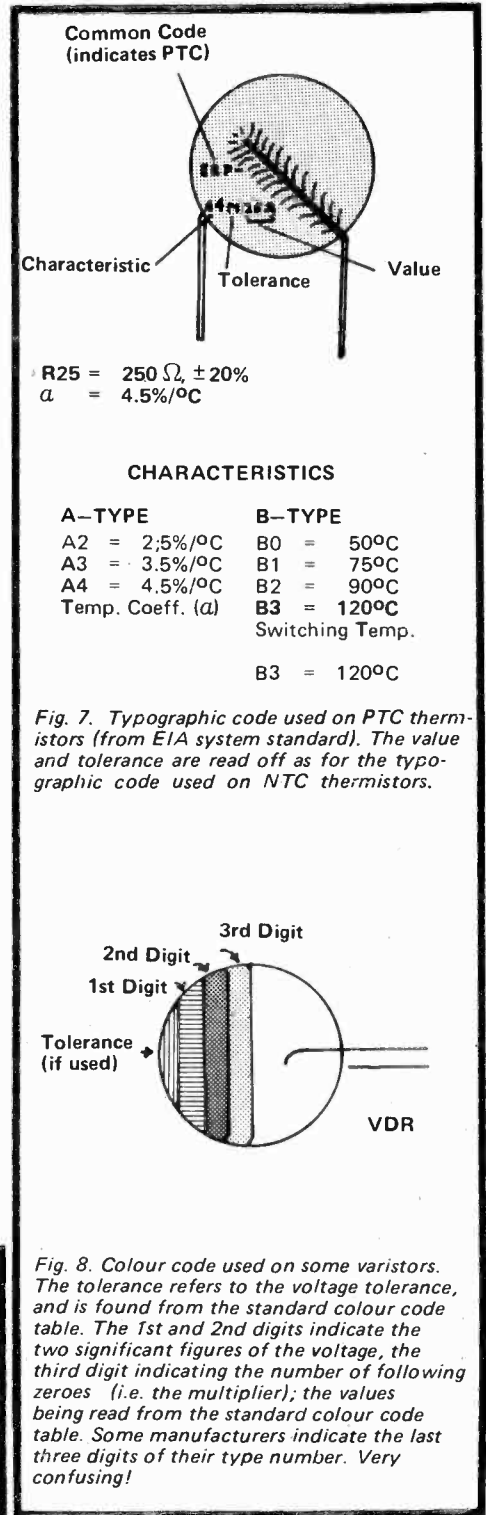


Fig. 7. Typographic code used on PTC thermistors (from EIA system standard). The value and tolerance are read off as for the typographic code used on NTC thermistors.

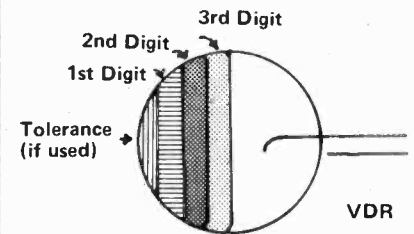


Fig. 8. Colour code used on some varistors. The tolerance refers to the voltage tolerance, and is found from the standard colour code table. The 1st and 2nd digits indicate the two significant figures of the voltage, the third digit indicating the number of following zeroes (i.e. the multiplier); the values being read from the standard colour code table. Some manufacturers indicate the last three digits of their type number. Very confusing!

Ceramic Diode (Variatite) Varistors

These devices have an asymmetric voltage characteristic and it is the value of the forward voltage that is of interest. They are generally made to a specified forward voltage and a colour code is used to indicate the value as illustrated in Figure 9. A single colour spot is used, and it is applied to the cathode side of the device.

RESISTOR CODES

ZNR Varistor Typographic Code

The typographic code used on ZNR varistors is usually arranged in one of two ways, as indicated in Figure 10. The disc-shaped varistors are generally marked in the manner illustrated, the ZNR marking directly indicating the type of component. This is followed by a single letter indicating the voltage tolerance followed by the voltage value. A 220 V, $\pm 15\%$ varistor is illustrated.

The cylindrical body style of varistor is generally marked according to the EIA system standard, as illustrated on the right in Figure 10. This code gives a more complete specification of the component's characteristics. The wattage rating and shape may sometimes be omitted. Reading this sort of code on any component can be confusing — it is best to first identify the component by the prefix and then read the code groups commencing from the right. The voltage value is always indicated last — but watch it again . . . the manufacturer may attach a suffix for his own purpose! It is usually a single letter and thus the voltage value group is easily recognised.

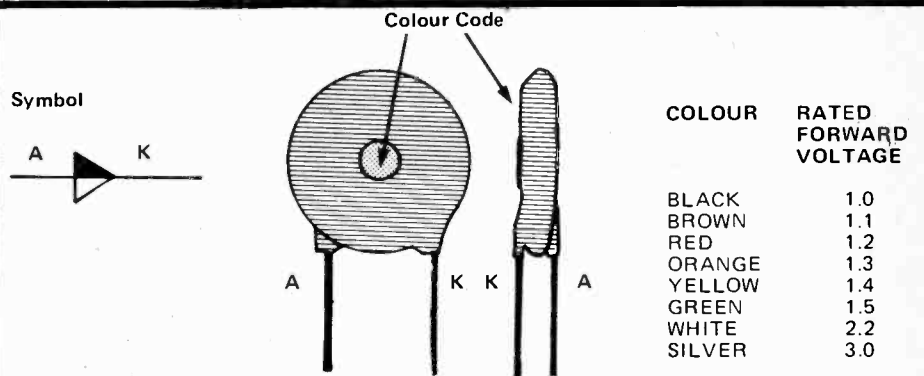
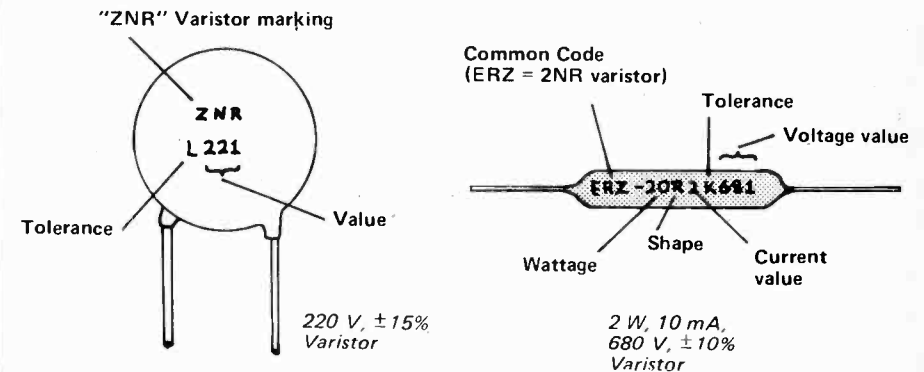


Fig. 9. Colour code used on Ceramic Diode (variatite) varistors. These have an asymmetric voltage characteristic and the colour code, indicating the rated forward voltage, is marked on the cathode.

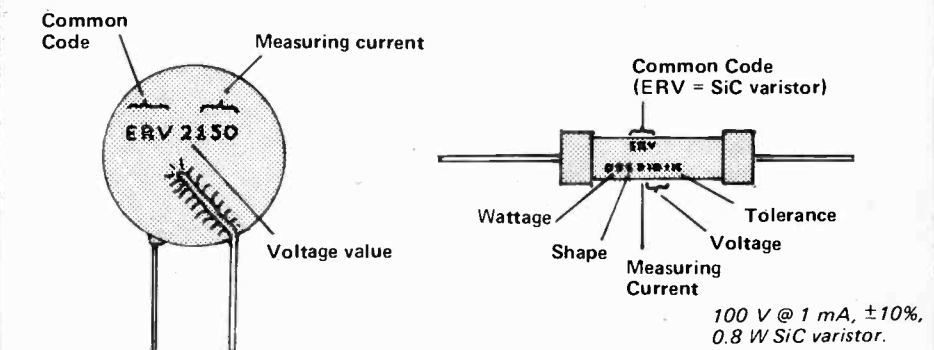


TOLERANCE	WATTAGE	CURRENT VALUE	VOLTAGE SUPPLY
J = $\pm 5\%$	03 = 0.3 watts	2 = 10 mA	First two digits are the two significant figures, the third digit being the number of following zeroes. Decimal point is indicated by R.
K = $\pm 10\%$	08 = 0.8 watts	3 = 1 mA	
L = $\pm 15\%$	15 = 1.5 watts	4 = 0.1 mA	
M = $\pm 20\%$	20 = 2.0 watts	9 = 0.5 mA	
S = $\pm 3\%$	60 = 6.0 watts		
	80 = 8.0 watts		
	1A = 10 watts		
	1B = 15 watts		
	1C = 20 watts		

Fig. 10. Typographic code combinations used on common (ZNR) varistors. The more complete form is shown on the right. It may be abbreviated however as indicated on the left. The current value is sometimes included as well, the wattage rating is usually only included where the more complete form of the EIA code is used.

Silicon Carbide (SiC) Varistor Marking

These varistors are also generally marked using the EIA system code, in a similar manner to ZNR varistors. The two basic marking styles are illustrated in Figure 11. The common code signifying a SiC varistor, ERV, prefix is invariably marked on both disc and cylindrical-shaped components, the disc-shaped varistors generally having an abbreviated code indicating only the voltage value and measuring current. The cylindrical-shaped varistors have the more complete code marked on the component body, as illustrated on the right in Figure 10. The wattage rating, measuring current, voltage value and voltage tolerance are the characteristics indicated. Note that the wattage rating code differs from that for ZNR varistors in that only a single figure is used to indicate components having a wattage rating of 1 W and 2W respectively.



WATTAGE	MEASURING CURRENT	VOLTAGE VALUE	TOLERANCE
01 = 0.1 watts	1 = 100 mA DC	First two digits are the two significant figures, the third digit being the number of following zeroes. Decimal is indicated by R.	J = $\pm 5\%$
02 = 0.2 watts	2 = 10 mA DC		K = $\pm 10\%$
03 = 0.3 watts	3 = 1 mA DC		L = $\pm 15\%$
08 = 0.8 watts			M = $\pm 20\%$
10 = 1.0 watts			
15 = 1.5 watts			
20 = 2.0 watts			

Fig. 11. Typographic code combinations used on Silicon Carbide (SiC) varistors. The complete form of the code is illustrated on the right. It is also used in an abbreviated form, as illustrated on the left, only the voltage value and current being indicated, although the tolerance is sometimes also included.

E12 SERIES RESISTOR COLOUR CODE

OHMS	BAND 1	BAND 2	BAND 3	OHMS	BAND 1	BAND 2	BAND 3
4.7	yellow	violet	none	3k3	orange	orange	red
5.6	green	blue	none	3k9	orange	white	red
6.8	blue	grey	none	4k7	yellow	violet	red
8.2	grey	red	none	5k6	green	blue	red
10	brown	black	black	6k8	blue	grey	red
12	brown	red	black	8k2	grey	red	red
15	brown	green	black	10k	brown	black	orange
18	brown	grey	black	12k	brown	red	orange
22	red	red	black	15k	brown	green	orange
27	red	violet	black	18k	brown	grey	orange
33	orange	orange	black	22k	red	red	orange
39	orange	white	black	27k	red	violet	orange
47	yellow	violet	black	33k	orange	orange	orange
56	green	blue	black	39k	orange	white	orange
68	blue	grey	black	47k	yellow	violet	orange
82	grey	red	black	56k	green	blue	orange
100	brown	black	brown	68k	blue	grey	orange
120	brown	red	brown	82k	grey	red	orange
150	brown	green	brown	100k	brown	black	yellow
180	brown	grey	brown	120k	brown	red	yellow
220	red	red	brown	150k	brown	green	yellow
270	red	violet	brown	180k	brown	grey	yellow
330	orange	orange	brown	220k	red	red	yellow
390	orange	white	brown	270k	red	violet	yellow
470	yellow	violet	brown	330k	orange	orange	yellow
560	green	blue	brown	390k	orange	white	yellow
680	blue	grey	brown	470k	yellow	violet	yellow
820	grey	red	brown	560k	green	blue	yellow
1k	brown	black	red	680k	blue	grey	yellow
1k2	brown	red	red	820k	grey	red	yellow
1k5	brown	green	red	1M	brown	black	green
1k8	brown	grey	red	1M2	brown	red	green
2k2	red	red	red	1M5	brown	green	green
2k7	red	violet	red	1M8	brown	grey	green
				2M2	red	red	green

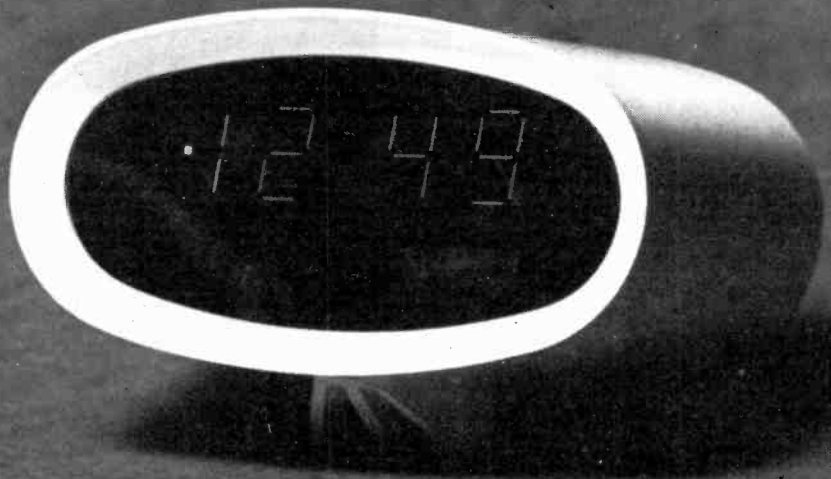
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ELECTRONICS

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PART 41

Oscilloscopes; The refinements

MANY MEASUREMENTS IN electronics can be handled by the relatively unsophisticated oscilloscopes described in the last part of this series. More capability can be provided at greater cost and this can be valuable if the user understands how to make the most of it. This part describes refinements that will be encountered in more advanced oscilloscopes.

IMAGE STORAGE

Screen persistence: Repetitive signals, such as a sinewave signal, can be made to repeat on the screen overlapping the previous trace produced. If the time-base frequency is sufficiently high — from thirty or forty hertz upward — the screen provides an apparently stationary signal of constant and adequate intensity. This is primarily because the eye cannot detect individual scans (as in motion pictures and television) and secondly because the phosphor, at frequencies above a few hundred hertz, is re-energized before its light emission due to the previous scan, has decayed away.

Phosphors with large time-constants are available (such as P2, which takes one second to reduce to 10% of original brightness and P7 which takes three seconds) and oscilloscopes have been manufactured which use these to enable signals of less than one hertz to be studied. This feature, however, largely restricts the use of the instrument to low frequency work because medium and high-frequency signals that are not well synchronised will produce separate traces which remain and add up with time to produce an unclear picture. This method of studying slow-transient phenomena has not been developed to any great degree because of this and other factors (such as poor resistance to burning). In addition the retained-image times are still inadequate for many applications.

CAMERAS

Storage requirements fall into two classes — those where the transient is unique and therefore needs to be recorded only long enough to allow the trace to be studied and those where a permanent record is needed.

The oscilloscope fulfils both these

needs. Until the advent of the Polaroid-Land process this involved a time-consuming development process before the operator was certain of having even recorded the trace. Most oscilloscope makers now offer specially built trace-recording cameras that fasten onto the large bezel surrounding the screen.

Such cameras use a Polaroid-Land film pack of some kind and often incorporate a 35 mm roll film facility also. A Dumont unit is shown in Fig.1. The user sets the CRO controls until satisfied that the trace will be as needed. This is done using the viewing aperture which reflects the screen image to the observer via a mirror. It is essential that the camera has the correct focal distance set for the CRO concerned, so in general cameras relate to specific units. Some models incorporate adjustable object-image ratios; a few are fixed ratio. With experience it is even possible to capture multiple trace events (by multiple exposure) for comparison purposes.

A considerable amount of film and patience can be consumed trying to record once-only events. Cameras can be quite expensive — several hundred pounds — but they do provide a

permanent record for reports which no other storage system can provide, and the price of a camera is not as great as the extra cost of the variable — persistence storage units to be discussed later.

STORAGE OSCILLOSCOPE

Most of the objections of the above storage methods, with the exception of permanent photographic reproduction, are overcome by using an advanced form of the basic CRO tube. It is called a variable persistence storage tube and is a development of early 1950's storage tubes in which the waveform could only be held at a constant intensity (without the feature of gradual fade out). In fact variable persistence is a feature of tube operating circuitry not the tube itself.

The construction of a typical storage tube is given in Fig.2. The phosphor viewing screen (having 0.1 s persistence time from P31 material) and the writing electron gun shown are similar to those used in the simple cathode ray tube. Additional components are the flooding electron gun system, a storage mesh which is coated

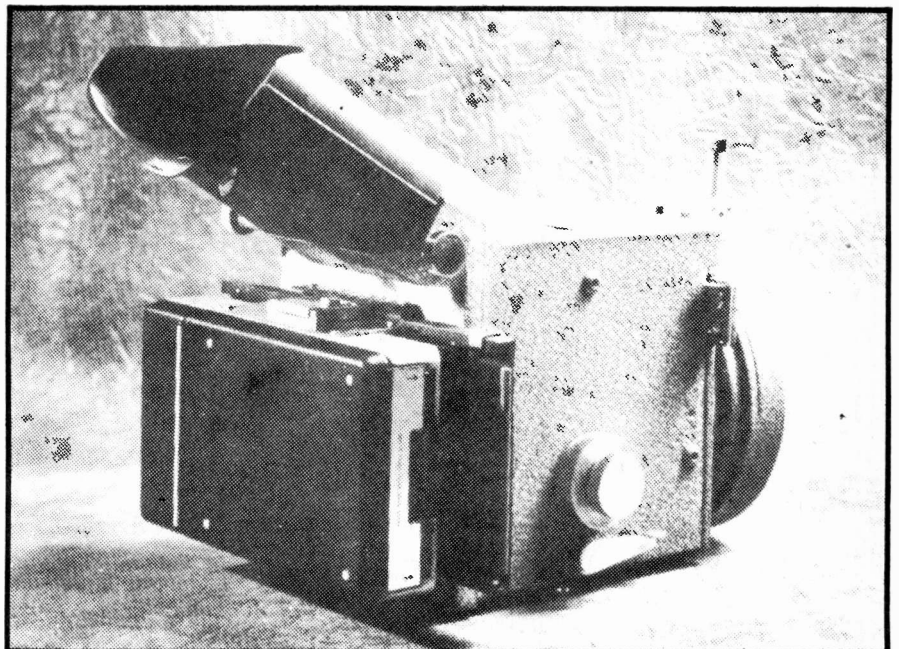
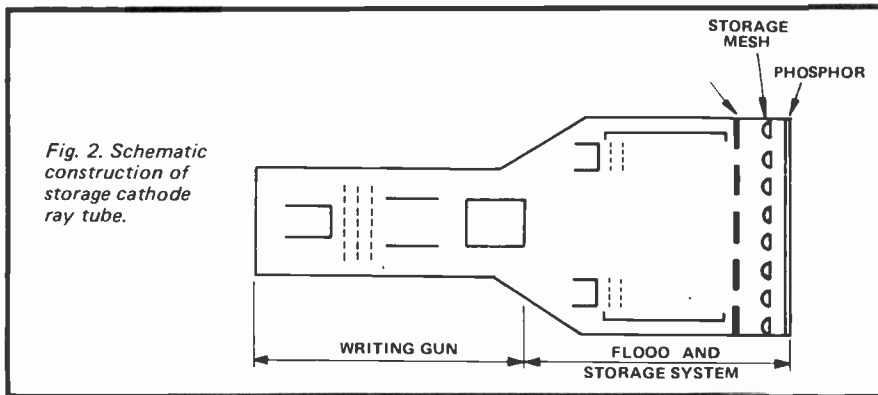


Fig. 1. Recording camera using Polaroid film pack.

ELECTRONICS—it's easy!



with a non-conducting, highly-resistive material such as magnesium fluoride, and a collector mesh which is held at a positive potential.

To store a trace the writing gun is scanned over the storage surface. Where the beam strikes the storage mesh electrons are knocked loose leaving a positive-charge pattern. The high-resistivity of the surface prevents the charges moving toward a neutral state: the scan is thus stored — and can be held for at least an hour (one maker offers four hours) in a reduced intensity mode.

To make the trace visible, low velocity electrons are sprayed by the flood guns onto the entire mesh surface. These electrons are allowed to pass through to the phosphor in proportion to the amount of positive charge at each aperture of the storage mesh. The positive field pulls many electrons through causing them to pass on to hit the phosphor.

The collector mesh is provided to help accelerate the flood electrons; to repel the positive ions generated by the flood guns; (which would otherwise write the whole screen bright)

and to absorb the emitted secondary-emission electrons produced whilst writing is in operation. It is not possible to store the trace in the view mode for as long as in the store mode: one to ten minutes of viewing time are typical for various makers' designs.

Erase is done by applying a large positive voltage to the storage mesh which charges capacitively to the same value. The mesh voltage is then brought back to a small positive value whereupon the flood guns reduce the voltage to zero. A small sudden negative excursion is finally applied to the mesh making it ready to write. (This procedure is automatically initiated at the single action of a switch.)

Variable persistence is incorporated by changing the time taken to erase the picture. In the Hewlett-Packard unit, shown in Fig.3, this is achieved by using a variable-width pulse generator that applies erase voltage pulses to the storage mesh. The positive-ions created by the flood-guns limit this mode to a maximum of 10 minutes persistence.

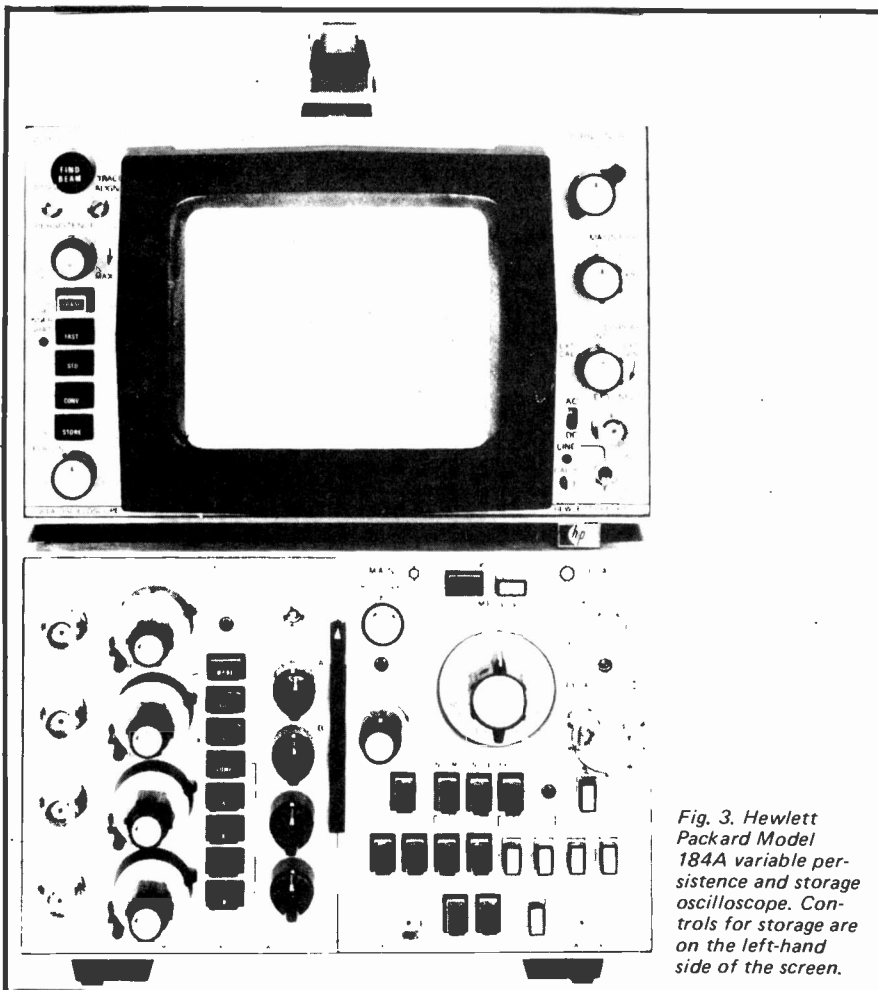
Storage oscilloscopes can be used as conventional units by applying about 30 volts to the storage and collector meshes. Long persistence has many virtues — it enables successive traces resulting from adjustments to a system response to be overlaid together for comparison purposes. It also allows us to see very low-frequency scans, and to plot scans of spectrum analysers. Long persistence also finds use in time-domain reflectometry where the time between send and receive pulse needs measuring.

By stacking sweeps on top of each other a long persistence time can be used to integrate or average a set of traces. Variable-persistence storage oscilloscopes are extremely versatile but the high price restricts their use to large laboratory groups.

STORAGE USING DIGITAL MEMORY

Figure 4 shows a unit marketed around 1972. The transient recorder unit accepts the analogue signal, converts it to a digital equivalent with respect to time and stores the values in digital registers. Readout can be obtained by using digital-to-analogue conversion of the stored increments which are scanned sequentially, the resultant analogue voltage being fed to an oscilloscope or chart recorder. Digital print-out is taken direct from the scanned store locations.

This method is less common than the storage oscilloscope alternative but the ever-reducing cost of digital methods may put this technique into a competitive price region.



Another method of capturing difficult to see, once-only transient signals, and very slowly changing waveforms is to record the level of the signal, increment by increment, as the signal occurs, using a digital memory. The concept is simple and the method offers certain advantages. These include ability to speed up or slow down the timescale of the original event, ease of providing a permanent numerical printout and the facility to process the signal before display.

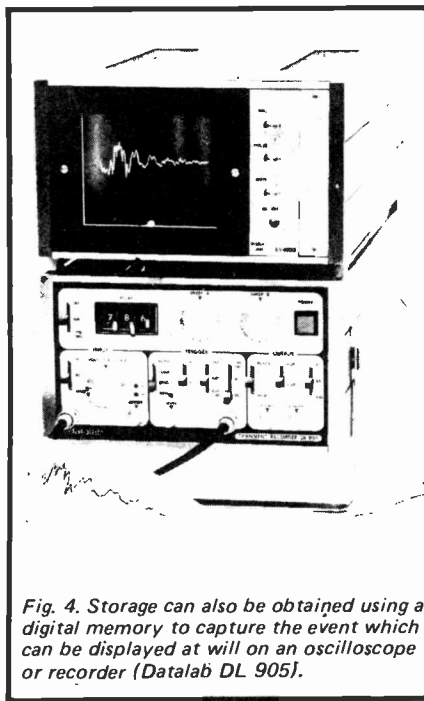


Fig. 4. Storage can also be obtained using a digital memory to capture the event which can be displayed as will on an oscilloscope or recorder (Datalab DL 905).

SAMPLING OSCILLOSCOPES

How to capture a very fast repetitive event, say near to the GHz region where scan times of 0.1 ns/division are needed, is a problem because the electron beam cannot transfer enough energy into the phosphor to obtain a useable trace brilliance. Further it

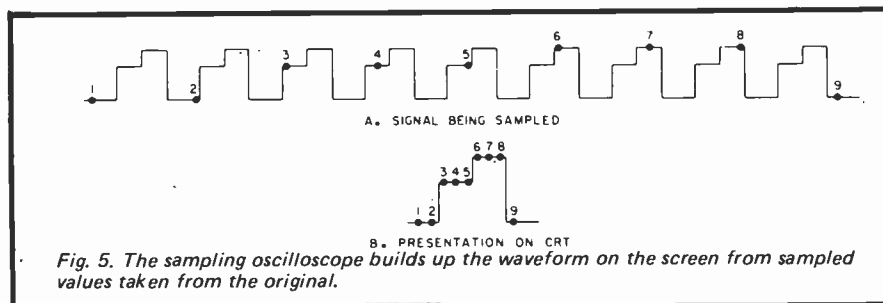


Fig. 5. The sampling oscilloscope builds up the waveform on the screen from sampled values taken from the original.

becomes increasingly difficult to deflect the beam at such speeds. The sampling oscilloscope offers a solution to these problems.

The sampling oscilloscope makes use of the stroboscope concept to look at a waveform, which must therefore

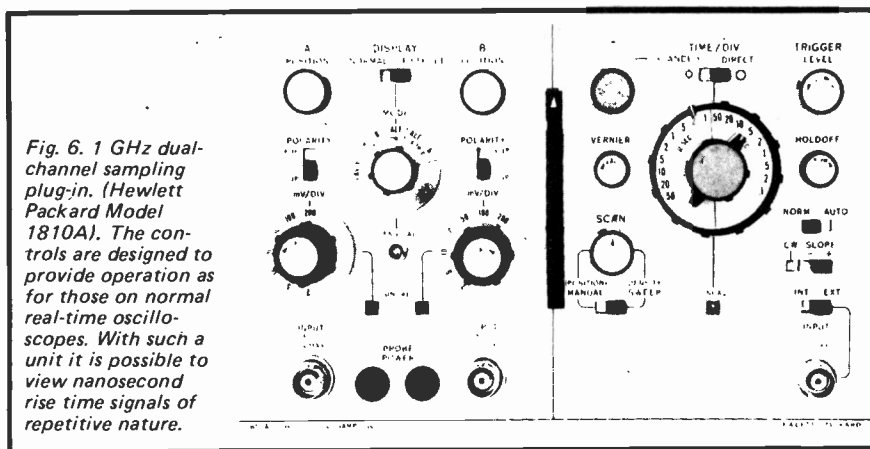
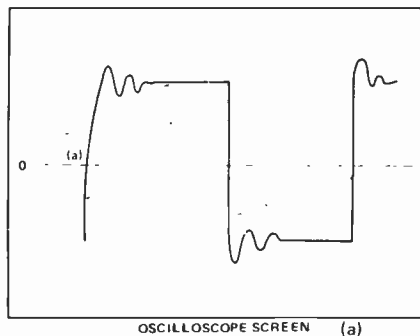
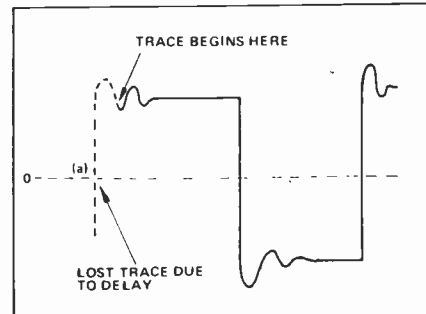


Fig. 6. 1 GHz dual-channel sampling plug-in. (Hewlett Packard Model 1810A). The controls are designed to provide operation as for those on normal real-time oscilloscopes. With such a unit it is possible to view nanosecond rise time signals of repetitive nature.



OSCILLOSCOPE SCREEN (a)



(b)

Fig. 7. Inherent trigger delay, if not compensated for, will lose the leading edge of a waveform.

be repetitive (as shown in Fig.5). The beam is set to illuminate the screen at point 1 in the diagram, waiting there until the next cycle where it moves to point 2 — and so on. The trace therefore gradually works its way through the complete cyclic waveform and because the scan speed is slower than with a conventional sweep system the cathode-ray tube system can operate with a lower bandwidth than the signal. The waveform produced is an average of many so the display is not only sharper but more uniform. (This may be a disadvantage in some applications for the sampling unit is effectively smoothing the unknown true original signal). Sample and hold

DELAY FACILITIES

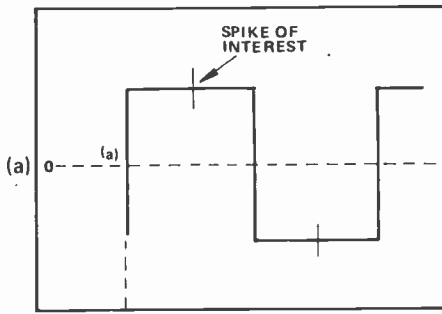
Often one needs to study a certain part of a repetitive waveform — the very beginning, for instance. An example is the ringing of a non-ideal square wave shown in Fig.7a. The trace is triggered, to begin the sweep, by a fast-going edge. Due to circuit response-times, the trace does not begin to sweep at exactly that time but begins a little later. The result is loss of the leading edge region of the wave as shown in Fig.7b. The following waveform may provide the information sought but attempts to widen the waveform in the horizontal direction lead to the second front disappearing. The simplest solution to this problem is to incorporate an appropriate fixed delay into the triggering circuits and this is often provided within the circuits. A slightly better method is to provide an adjustable delay control on the trigger panel.

A more difficult problem is capturing a point on the signal train that is remote from the triggering transient. Consider the signal shown in Fig.8(a), where the problem is to investigate the spike transient on the pedestals of the square wave. Triggering is best achieved by using the edge (a). But this means that scale expansion puts the spike off scale when the horizontal expansion scale is great enough to provide information about the spike structure.

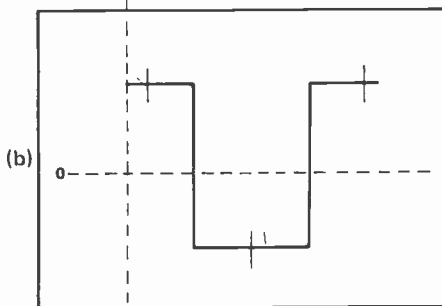
methods were discussed in the previous part discussing D-A and A-D conversion.

In practice a sampling oscilloscope is a normal high quality scope which can accept a sampling plug-in: Fig. 6 is the panel of a dual sampling unit.

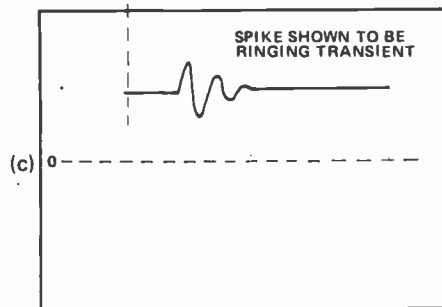
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ZERO TIME DATUM FOR TRACE



OSCILLOSCOPE SCREEN



(c)

Fig. 8. Use of introduced delay in triggering to enable an event away from trigger transient to be investigated.
 (a) Original spike on pedestal of square wave.
 (b) Delay introduced to bring spike back to time origin.
 (c) Scale expanded to reveal true nature of spike.

Variable delayed sweep is the answer. The trigger circuit is set by the (a) edge but trace scan does not begin until after a period, as in 8(b). Thus the trace captures the spike at the left-hand side of the screen and scale expansion will now be possible as in 8(c).

To make this workable in practice the operator must know just where triggering occurs for there may be

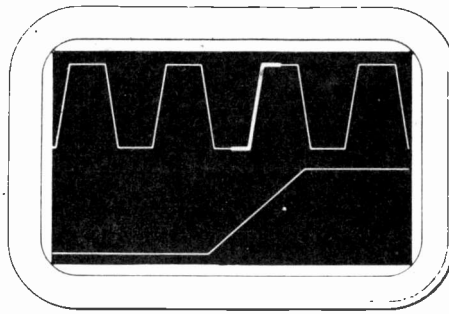
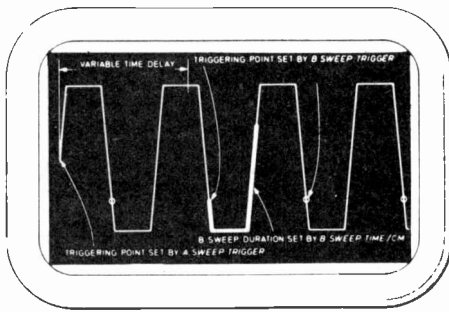
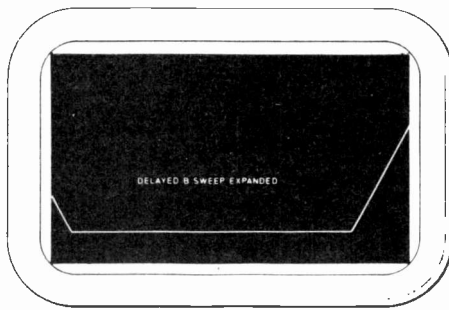


Fig. 9. Trace brightening is used to show which part of the waveform is to be expanded. In this display the expanded portion is also displayed on the second trace of the CRO.



(a)



(b)

Fig. 10. Use of dual delayed triggering point.
 (a) original (b) expanded.

several somewhat similar events along the trace. It is vital to know which one is being viewed. A refinement provided in variable delay circuits is to brighten the original display from the point where triggering will begin. Taking the idea one step further leads to a second delay that effectively decides where the trace stops. Fig. 9 shows the waveform brightened to show the portion that will be expanded and the second trace of the dual-beam unit is used to show the expanded part. Another useful feature is to be able to use a trigger point not on the origin of the first trace set up — as in Fig. 10. Here a marker dot is provided to help the operator.

PROBES

Passive probes for voltage measurement: In Nos. from May 77 the importance of providing the right matching conditions between two electronic systems was stressed. This is also important when connecting an oscilloscope to a circuit, for each putput and input has certain resistive and reactive conditions which must be properly combined to get realistic signal transfer.

The oscilloscope can be represented as an ideal termination shunted by a large R and an adequately small C value — or at least they appear this way at first sight. Fig. 11 is the most common approximate equivalent circuit, (others used include 50 ohms with negligible reactance in certain applications). Referring to the chart in Fig. 12, it can be seen that with 20pF at 10 MHz the circuit being measured must have the equivalent output resistance of no more than 8 ohms!

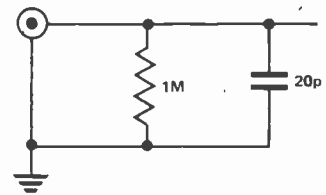


Fig. 11. Most oscilloscopes have this input equivalent circuit. Although the values seem insignificant, at high frequencies they become dominant requiring the use of special probes.

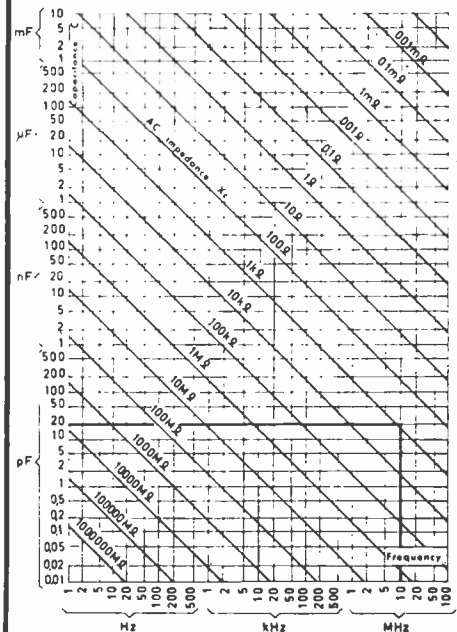


Fig. 12. Chart for obtaining reactance of capacitors at various frequencies of operation.

For high frequencies, those above 100 kHz say, we therefore need a better connection method. To further compound the problem the oscilloscope input leads can easily increase the equivalent C value to 100 pF — leads for 1:1 connection must therefore be carefully designed to ensure known loading conditions which can be allowed for in signal measurement corrections. It is very bad practice to use any piece of coaxial cable and connector for frequencies beyond 100 kHz.

The first improvement is to use a probe which has 10:1 attenuation built in, for these are designed to have a lower effective cable capacitance — see Fig. 13(a). Still better is a special correction arrangement that balances the shunt against series capacitance to provide a wider bandwidth — see Fig. 13(b). By the use of inductive tuning a further improvement in bandwidth can be obtained — Fig. 13(c). Probes with division ratio of 100:1 also are manufactured — these can provide equivalent termination conditions of 5K/0.7 pF, 10M/1.8 pF, 1 M/1 pF. The reason for different pair combinations arises from the need to alter the trade-offs between rise time and signal loss in high-frequency and very fast transient measurements.

There is no easy answer to the question of which attenuator probe to use. These guides are the start. For amplitude measurements select a minimum-impedance source point to measure from. The best probe to use here is one with the highest impedance at the frequency of interest. Capacitance is less important here than resistance for it alters edge shapes, not amplitude.

For fast risetime measurements again select a low impedance source point and use a probe with lowest effective capacitance — signal attenuation is less important than transient edge shape changes.

ACTIVE PROBES FOR VOLTAGE MEASUREMENT

The above probes make use of passive matching arrangements. But for the extremes of frequency and/or risetime measurements the values of components required in passive probes become impractical. However active amplifiers interposed between the circuit and the oscilloscope can be used to improve performance by increasing input resistance and lowering capacitance (short loads). FET probes are marketed to meet this.

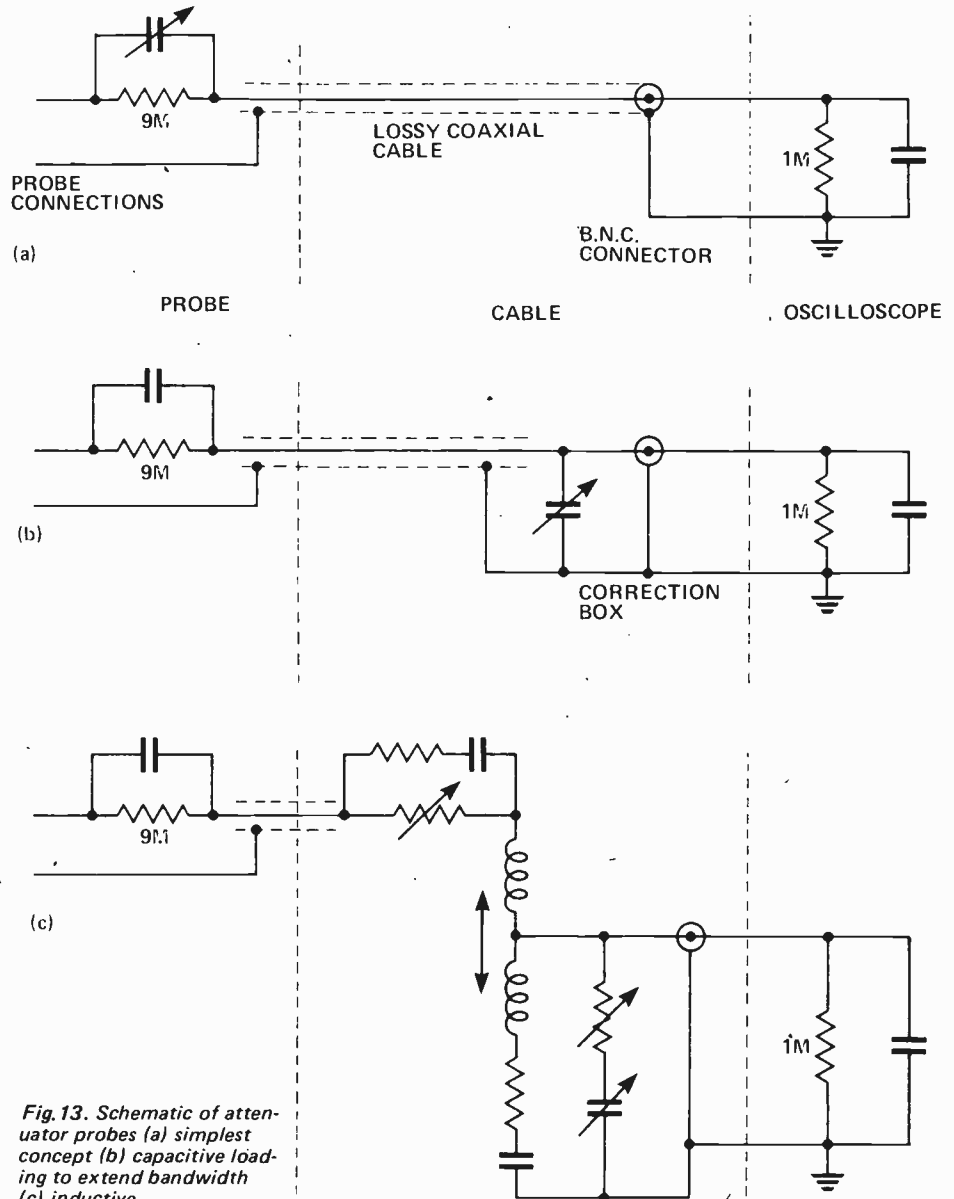


Fig. 13. Schematic of attenuator probes (a) simplest concept (b) capacitive loading to extend bandwidth (c) inductive.

OTHER PROBES

Voltage measurements are by far the most frequent measurements made but in some instances it may not be possible to determine voltages, and current measurement is used instead. An example is the current flowing in a direct-coupled Darlington pair configuration where no significant resistance exists over which a voltage can be

developed. DC current probes (see Fig. 14) clip over the wire in question coupling the dc magnetic field created by the current flowing in the wire into a Hall effect transducer which generates a voltage equivalent to the current flowing. These will also measure ac currents. The maker specifies the conversion constant — typically 1 mV/mA. AC only, current

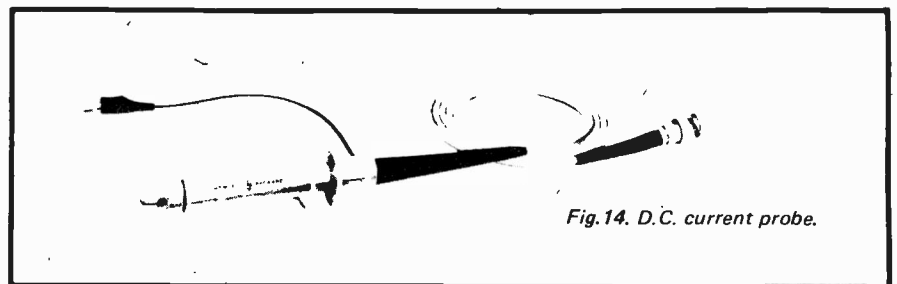


Fig. 14. D.C. current probe.

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probes are also made using a current-transformer principle.

Probes for use in digital circuits are also available. These may incorporate a logic gate that combines the outputs from up to 6 circuit points as shown in Fig. 15. Power for the gate is obtained from the circuit under test.

SPECIAL PLUG-INS

The oscilloscope, due to its extensive flexibility, can form a major part of many test systems, thereby reducing the overall price of advanced measurement systems where a suitable CRO already is available. Special plug-ins are offered (to suit certain main-

cost is concerned) is the CRT itself for frames) that will convert an oscilloscope into a spectrum analyser or into a semiconductor characteristic-curve tracer. Another plug-in is offered that converts the CRO into a four-trace unit.

A basic need in manual measurement is the provision of output form that best suits the operator. In many tasks a visual output in the form of a picture or graph is better than having to view many traces of a time sequence taken over the whole system. Already we have logic analysers which it is just about the last remnant of thermionic device technology remain-

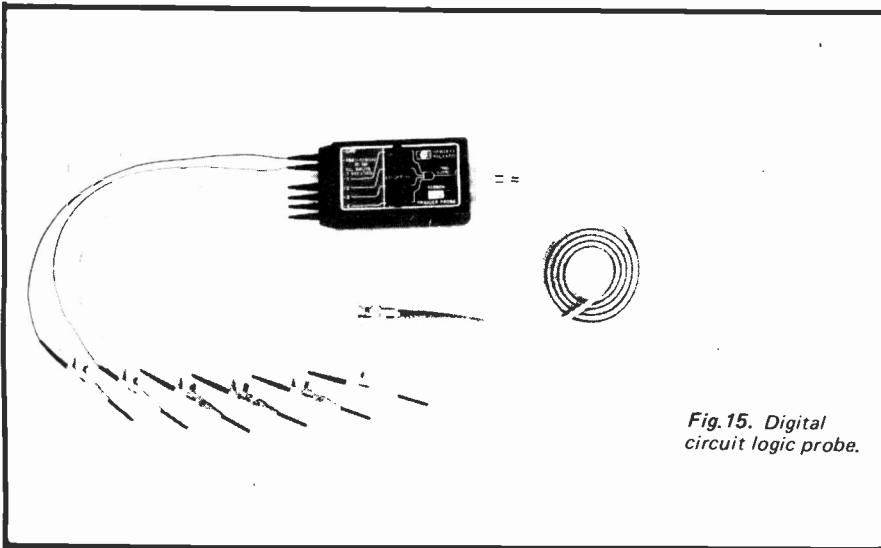


Fig.15. Digital circuit logic probe.

display space-plane information on the CRO screen, multi-meter CRO units that write digital values on the screen and units that provide axes information on screen graphs. With the reducing cost of advanced processing it will not be long before the micro-processor and memory (already in use in very sophisticated units) are introduced into quite moderately priced oscilloscopes for converting the information taken from the circuit into better forms of display. Display monitors are already available with many display forms. The next stage must be the marrying of the basic CRO unit to such capability via a wider range of sophisticated plug-ins. The colour oscilloscope will also soon be with us extending the information rate at which the operator can be informed about a system via a CRO.

The only weak link in present systems (as far as robustness, life and

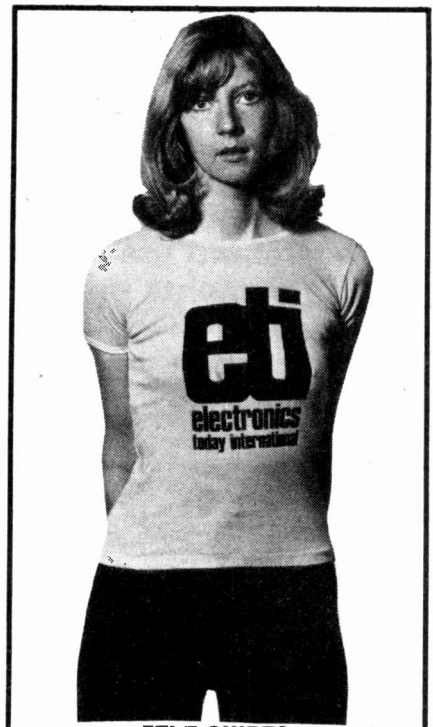
ing in general use. This too will soon be replaced by a solid-state equivalent. Perhaps this will take the form of a matrix of three-colour, LEDs in a flat display — making maximum use of the low-cost production advantages of LSI techniques.

Continued next month . . .

FURTHER READING

Due to the versatility of the oscilloscope most books on electronic instrumentation include basic descriptions of how oscilloscopes work and how to perform basic measurements with them. Many books are devoted entirely to the oscilloscope.

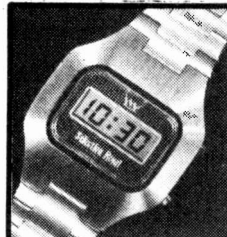
General considerations are discussed in "Test and measuring instruments — 1974 Catalogue", (Philips). Tektronix, Hewlett-Packard, Dumont and Marconi outlets also provide basic articles on the selection and use of oscilloscopes. ●



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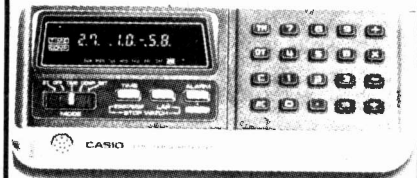
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ELECTRONICS TOMORROW

by John Miller-Kirkpatrick

Standardisation et al.

I WAS READING IN an American magazine how programs written for SWTP6800 don't work on an Altair 6800,8080 programs not running in a different mainframe and other such non-standardisation. What that means if you're not quite up to MPU jargon is that a program written in machine coding for one manufacturer's MPU will not work in another manufacturer's version of the same MPU kit. Exchanging programs must involve careful checking of the code to ensure that any subroutines required are available and in the correct locations. As different manufacturers of microcomputer systems use different decoding techniques for device addresses they have a different MONITOR or DEBUG program.

The monitor program is a program which is operated upon after Power-up or Reset has been signalled to the MPU. The Monitor sets up any variables or pointer registers and then branches to an I/O routine for a command from the operator. This command is a programming code which instructs the Monitor to perform a set of tasks which will result in the required operation. Examples of such commands are Print Memory, Alter Memory, Excute Program, Store and Read.

The monitor program must reside in a fixed location in memory (HICORE for 6800,8080; LOCORE for SC/MP, etc), and is usually sold in the form of PROM. On some machines the basic monitor program is simply a BOOTSTRAP, this is a program entered via data and address switches. The program is usually only 40 or so bytes and it causes data to be read into RAM and then that program data to be executed. The program data read is in the usual form of monitor described earlier.

Because the Monitor program requires to branch to I/O routines these sub-routines must also lie in a fixed location in memory and thus they are usually stored in the same PROM. These subroutines are useful and convenient, useful because they are probably the first MPU programs you will get to know well and convenient because any programs that you write can use the same sub-routines for I/O. Herein lies the biggest problem you

are likely to come across in MPU's. Any programs you write may not work on another MPU system based on the same chip. Conversely any programs you require from a library must have been written with your system in mind. Hardware standardisation is much simpler, you either go for one of the big systems with 100 or more parallel connectors or a minibus with just the bare essentials paralleled. Standardisation of hardware is being discussed variously at present, whether it be bus structure or pin and signal compatability between MPUs. Interface and communication standardisation is supported by 20mA loops, CUTS, ASCII, etc.

Software standardisation seems to be based solely in the idea of using the 50 or so characters on a keyboard in whatever sequence seems right at the time. I have been guilty of doing this myself but I have now seen the light and feel that I should pass on the message to all concerned.

Instead of basing everything on a fixed location monitor use a fixed location STACK or Working Storage RAM. You still need your basic monitor or a switch entry system at reset but once set up the system no longer needs that program. Any parameters such as data or addresses to be passed to/from subroutines or I/O devices are stored in this WS-RAM. The Monitor stores, for instance, the address of the I/O routines in the PROM at fixed locations in the WS-RAM, in order to branch to a routine it must first load the data from the WS-RAM into a branching register and then jump to that address.

Any programs you write can use the same I/O routines by simply following the address load and branch sequence. Any subroutines you write use the same technique of somehow storing their location at a fixed location in WS-RAM, if the subroutine happens to be an I/O routine then its address could replace the address of the PROM I/O routine in WS-RAM. Any reference by a program to the PROM I/O routine will now use your routine instead, if something goes wrong in your routine hit Reset and you are back with the standard I/O.

Perhaps you are an old hand at MPU's or computers and you think in terms of a programming language such as BASIC, FORTRAN, COBOL or PLI. At least one English language compiler is available for most MPUs and some have several optional compilers. Some compilers end up with a machine code program while others work as interpreters of the stored program text. Back in the bad old days of IBM 360s, 1400s, System 4s, etc manufactured by large international companies 1400 COBOL didn't work on 360s, IBM FORTRAN had to be modified for ICL or whoever. The mainframe boys made a bad start in software standardisation so please lets have some sense with MPU software.

Time on my hands

After my appeal for information on DCF77 I have received several letters with copies of articles on DCF77 time code receivers, all in German. My thanks to all of those who sent in this data and to those who correctly assumed my knowledge of German and offered translations.

One point about DCF77 which I feel that I should point out to any prospective user is that there is an annual licence fee of 2.00DM but I presume that this is only payable for reception in Germany. ●

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TTL KEYS

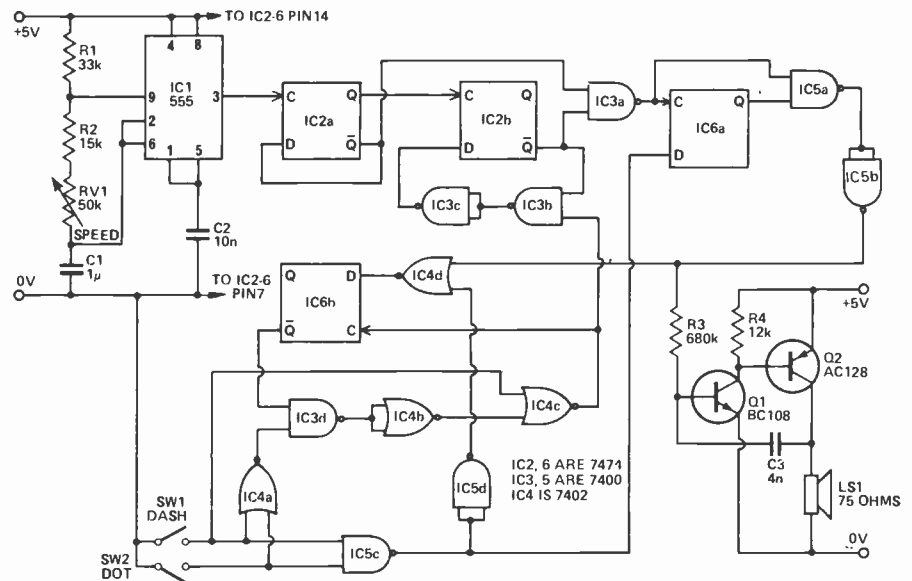
This device can be used to send perfectly spaced Morse at very high speeds - up to twice as fast as with an ordinary Morse key. It uses six integrated circuits, and also requires two special switches, SW1 and SW2, which are described later.

To describe the operation of the circuit fully would take up over a page of ETI, and so a simplified explanation is given here. IC1 is a 555 timer connected as an astable multivibrator, whose frequency is varied by RV1. The output is fed to IC2a, a D flip-flop, which divides the input frequency by 2, producing a square-wave with a 1:1 mark-space ratio (dots).

If SW1 and SW2 are both open, the D inputs of IC2b and IC6a are both at logic 0, so that the dots from IC2a are inverted by IC3a, but blocked by IC5a. IC5b output is a 0, and so the audio oscillator made up of Q1 and Q2 and the associated components is disabled and no tone is fed to the speaker.

If SW2 is closed, IC6a's D input becomes logic 1. However IC6a's output can only change state on the rising edge of a clock pulse (i.e. the beginning of a dot). Hence if a dot has already started when SW2 is closed, it will not get through to the speaker, but the next dot will, because it will make IC6a's Q output to go to 1. Hence the dots now get through to the oscillator and successively enable and disable it, causing dots to be heard coming from the speaker. When SW2 is opened, if a dot is in progress it will continue until it has finished, and then at the beginning of the next dot, IC5a output will go low and no more dots will be heard. There is a short delay between the beginning of the dot and the Q output going low, which does cause a short 'blip' at IC5b output, but the blip is too fast to be heard.

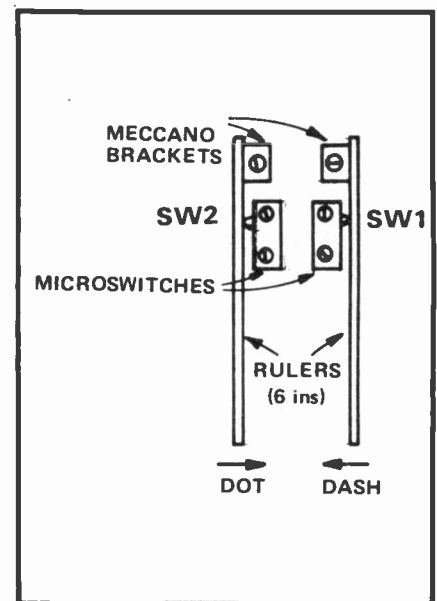
If SW2 is closed, but SW1 open, IC4c output goes to 1 and IC2b's \bar{Q} output is effectively shorted to its D input. This causes IC2b to divide the string of dots from IC2a by two. The outputs of IC2a and IC2b are combined by IC3a to produce a waveform with a 3:1 mark-space ratio (dashes). These are passed on to the audio oscillator just as before. The dashes, like the dots, are self-completing. Notice that IC4c output determines whether dots or dashes are produced.



While SW1 or SW2 is closed, IC6b's D input is fed from IC4c output and clock pulses come from IC5b. If SW1 and SW2 are both operated together, IC4a allows the output of IC6b to pass to IC4c, and IC4d inverts IC4c output again, so that IC6b \bar{Q} output is shorted to its D input. Thus IC6b changes state every time a dot or dash begins at the output, and causes alternate dots and dashes to be produced. This is useful when sending a letter like C (dash-dot-dash-dot), as the switches SW1 and SW2 each need to be closed and opened just once.

It was found after the unit had been built, that it was difficult to send a letter like A (dot-dash) at high speed because SW1 had to be closed a fraction of a second after SW2, which was difficult to achieve at the first attempt. Hence IC5d and IC4d were added. When both switches are released, IC6b input becomes 0. A clock pulse is then applied to IC6b by the 'blip' described earlier. This makes the output go high, and if now SW1 and SW2 are closed simultaneously, the first thing to be heard in the speaker will be a dot.

SW1 and SW2 are push-button microswitches, and these are operated by means of a lever arrangement as shown in the diagram. Plastic rulers were used on the unit built because they are flexible.



The component values shown around IC1 give a speed range of 11-30 words per minute. The upper limit can be raised by decreasing R2. I have so far reached a speed of 20 wpm on the unit, after only a week or so of using it. As it stands it is a Morse practise unit, but if IC5b output is taken to a transistor driving a relay, the relay contacts could be used in place of an ordinary Morse key in a C.W. transmitter.

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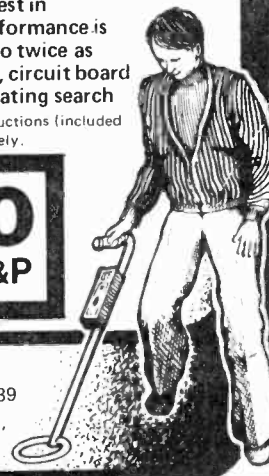
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NON ALARM £12.50 VAT + £1.00
ALARM £15.28 VAT £1.22
 Including P&P

"DELTA"

Teak or Perspex Case, 6" x 2 1/2" x 3"

FEATURES:

4 Red 1/2-inch high LEDs. 12 hour display with AM/PM indication. Power failure indicated by flashing display.

Precise accuracy from mains frequency.

Beautiful Burma teak case, or stylish Perspex (state first and second colour choice: White, Red, Blue, Green, Black, Mauve)

NON ALARM: Complete Kit, including Teak case £13.50 incl.

Module Kit, excluding case £9.50 incl.

ALARM EXTRAS: Pulsed alarm tone. Tilt operated 'Snooze' period Automatic Brightness control. Simple setting.

ALARM: Complete kit, including Teak case £16.50 incl.

Module Kit including case £13.00 incl.

PERSPEX CASES: 50p less than Teak.

READY BUILT: Extra £1.00 on complete clocks. Extra 50p on Modules.

TIMER FACILITY: Count in seconds up to 9m. 59 sec. Extra 50p.



'ALPHA' £12.00 incl.

Ready built. Non alarm, Perspex only. 4 digits, phosphorescent green, 1/2" high, 12 or 24 hour display.

Discount on orders over five clocks or kits. Overseas. Add Air Mail or Surface Postage.



Send payment with order. SAE for details.

Baron

(E1), 6 Gower Road
 Royston, Hertfordshire
 Phone Royston (0763) 43695

Are you receiving us? NEW Doram DIGITAL FREQUENCY METER YOURS FOR ONLY £54.50+ £4.36 VAT



The DFM you must have at a price that could save you up to £25. It's easy to see how accurate this new Doram meter is thanks to 4 digit display with extra shift left to check 5 digits for even greater accuracy!

You can see the well-designed shatterproof case with its built-in handle/display tilt-foot looks so much more expensive than many a kit product.

But Doram kits always give you the best for less. This complete kit with case, printed circuit boards, all components and instructions is yours for only £54.50 + £4.36 VAT.

The meter covers 20Hz to 50MHz with prescaler (additional purchase) provision allowing you to measure up to 500MHz. Input impedance and sensitivity are very high with read-out accurate to ± 0.01%!

So the message comes through loud and clear—don't miss this latest Doram digital bargain.

DEPEND ON DORAM

Thousands of kit makers do for top value and modern technology.

Doram Electronics Ltd, PO Box TR8, Wellington Road Industrial Estate, Wellington Bridge, Leeds, West Yorkshire LS12 2UF.

Registered in London No. 1155856 Directors R. A. Marler, D. I. Turner, F. Chable.

KEY DATA

Frequency range	20Hz to 50MHz in 3 steps
Sensitivity	20mV r.m.s.
Input Impedance	1MΩ in parallel with 30pF
Input Coupling	AC
Maximum Input	10V r.m.s.
Frequency Standard	1MHz 0.01% calibration tolerance
Display	4 Digit LED display with shift left for 5 digit readout
Accuracy	± 1 digit (0.01%)
Supply Voltage	220-240V ac with 12V dc 100mA output for prescaler supply

Yes, I want to save up to £25

Please send me (subject to availability) complete kit(s) for DORAM DIGITAL FREQUENCY METER at (£54.50 + £4.36 VAT inc p & p) total **£58.86**. I understand I can return unused kit(s) within 7 days and claim my money back if kit(s) returned in the form received and ready for re-sale.

Please send me complete DORAM KITS CATALOGUE(s) showing an additional 25 kits at 25p each I enclose cheque/PO value

NAME (BLOCK CAPITALS)

ADDRESS

TOWN

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POST TO DORAM ELECTRONICS LTD., DEPT. EF2, PO BOX TR8, LEEDS WEST YORKSHIRE LS1 2UF

SINTEL BARGAINS OFFER OPEN UNTIL 31st JULY, 1977

SINTEL SPECIAL OFFER — CONDITIONS OPEN UNTIL 31st JULY, 1977 INCLUSIVE — WHILE SPECIAL OFFER STOCKS LAST

Orders must bear a postmark dated before 31st July, 1977 inclusive.
Only written orders, either C.W.O. or Access/Barclaycard will be accepted.
SPECIAL OFFER prices are before VAT and p&p charges.
This offer only applies to orders enclosing our cut-out SPECIAL OFFER COUPON in the bottom right hand corner of this advertisement.
Orders will be met while SPECIAL OFFER stocks last.
Offer applies to UK only.
After 31st July prices will be as our current Price List.

ACK

~~£27.80~~
£24.80



BARGAIN KIT PRICES

£3 OFF ACK AND CCK



CCK
~~£42.00~~
£39.00

ATTRACTIVE SIX DIGIT ALARM CLOCK KIT

A complete kit including slim white case h. 40mm w. 205mm d. 140mm with deep red panel. Features: sleep alarm, snooze, automatic intensity control and high brightness display driving Twelve or twenty-four hour format (easy to add a switch to switch between them). Touch switch snooze.

Order as **ACK** ~~£27.80~~ **£24.80**
This alarm clock is also available with CRYSTAL CONTROL and BATTERY BACKUP. If mains power is disconnected (deliberately or accidentally) the clock will keep perfect time. Accuracy to within a few seconds a month. While on backup, the displays are off to conserve battery but the alarm remains fully operational.
Order as **ACK+BBK+XTK** ~~£34.33~~ **£31.33**

CRYSTAL CONTROLLED 6 DIGIT CAR CLOCK WITH INDEPENDENT JOURNEY-TIMER
Shows time or elapsed time in hrs. mins. secs. • Runs off car 12v supply • Nine push buttons for Start-Stop-Reset selecting display, to show time or elapsed time • All controls functional irrespective of display mode • Kit complete with case
Order as **CCK** ~~£42.00~~ **£39.00**

50Hz CRYSTAL TIMEBASE KIT

Use to improve accuracy of your digital clock • As a 50Hz source in a clock with battery backup • Accurate to within a few seconds a month • Size h. 14mm w. 64mm d. 49mm
Order as **XTK** ~~£4.28~~ **£4.28**



£2 OFF THESE KITS

THE SINTEL CAR CLOCK KIT

Four 0.5" red digits • Neat white case • Crystal control • Battery backup • Suitable for all 12v negative earth cars • Size h. 40mm w. 143mm d. 85mm Complete less battery.
Order as **AUT-CK** ~~£17.85~~ **£15.85**



GCK CLOCK KIT
Four bright green 0.5" digit mantelpiece or office clock • White case • Size h. 40mm w. 154mm d. 85mm. Complete less mains cable, plug and battery
Order as **GCK** ~~£14.40~~ **£12.40**
Also available with crystal control and battery back-up.
Order as **GCK+GBBK+XTK** ~~£17.65~~ **£17.65**

SIX JUMBO DISPLAYS

FOR ONLY **£6**



Common Cathode
0.5in. Character Height
Order as **FND500x6**
~~£7.80~~ **£6.00**

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CD4000 0.17	CD4025 0.23	CD4048 0.58	CD4077 0.45	CD4514 2.84
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CD4002 0.17	CD4027 0.58	CD4050 0.58	CD4081 0.23	CD4516 1.40
CD4006 1.20	CD4028 0.92	CD4051 0.94	CD4082 0.23	CD4518 1.25
CD4007 0.18	CD4029 1.18	CD4052 0.94	CD4085 0.74	CD4520 1.19
CD4008 1.00	CD4030 0.58	CD4053 0.94	CD4086 0.74	CD4527 1.64
CD4009 0.58	CD4031 2.30	CD4054 1.20	CD4089 1.60	CD4532 1.39
CD4010 0.58	CD4032 1.02	CD4055 1.36	CD4093 0.92	CD4555 0.90
CD4011 0.20	CD4033 1.44	CD4056 1.36	CD4094 1.94	CD4556 0.90
CD4012 0.23	CD4034 1.97	CD4059 4.93	CD4095 1.08	MC14528 1.22
CD4013 0.58	CD4035 1.22	CD4060 1.15	CD4096 1.08	MC14553 4.68
CD4014 1.04	CD4036 3.29	CD4063 1.13	CD4097 3.85	1M6508 8.05
CD4015 1.04	CD4037 0.98	CD4066 0.63	CD4098 1.13	
CD4016 0.58	CD4038 1.10	CD4067 3.85	CD4099 1.90	
CD4017 1.04	CD4039 3.20	CD4068 0.23	CD4502 1.24	
CD4018 1.03	CD4040 1.11	CD4069 0.23	CD4510 1.41	
CD4019 0.58	CD4041 0.86	CD4070 0.51	CD4511 1.72	
CD4020 1.28	CD4042 0.86	CD4071 0.23		
CD4021 1.04	CD4043 1.01	CD4072 0.23		
CD4022 0.94	CD4044 0.96	CD4073 0.23		
CD4023 0.23	CD4045 1.45	CD4075 0.23		
CD4024 0.80	CD4046 1.37	CD4076 1.34		

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KITS FOR TTL LATCHED COUNTER MODULES at SPECIAL OFFER PRICES
Each kit includes a set of displays (0.5" Red LED type TIL321/FND507) and two PCBs, one each per digit of 7447, 7475 and 7490, plus brackets, resistors, capacitors, a single in-line plug and socket nuts, screws, washers, instructions, etc. Sockets are NOT included and we recommend that you order Soldercon Pins separately (112 for a 2 digit module, 224 for a 4 digit module and 336 for a 6 digit module).
Each module consists of one vertical PCB holding the displays, fixed by two brackets to a horizontal PCB, which pbs the remaining components.



Order as **721-412** ... ~~£10.52~~ **£8.52**

Order as **657-412** ... ~~£17.98~~ **£15.98**

Order as **526-412** ... ~~£25.66~~ **£23.66**

KIT FOR SINGLE PCB CMOS COUNTER MODULE WITH LATCH
Order as **142-269** ~~£8.80~~ **£6.50**

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Each kit consists of the appropriate number of 0.5" red LED displays (either common anode TIL321/FND507s or common cathode TIL322/FND500s) and a display holding PCB. OPTIONS: PCBs wired for multiplexing or non-multiplexing, clock format or counter format.

TYPE	COMMON ANODE Part No.	Price	COMMON CATHODE Part No.	Price
2 digit Counter	574-822	£3.37	446-822	£2.97
4 digit Counter	777-822	£6.63	128-822	£5.83
6 digit Counter	684-822	£9.89	271-822	£8.69
Multiplexed				
4 digit Clock	801-822	£6.66	262-822	£5.86
6 digit Clock	417-822	£10.15	452-822	£8.95
8 digit Counter	119-822	£13.09	515-822	£11.49

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VALID UNTIL 31st JULY 1977 INCL.

tech-tips

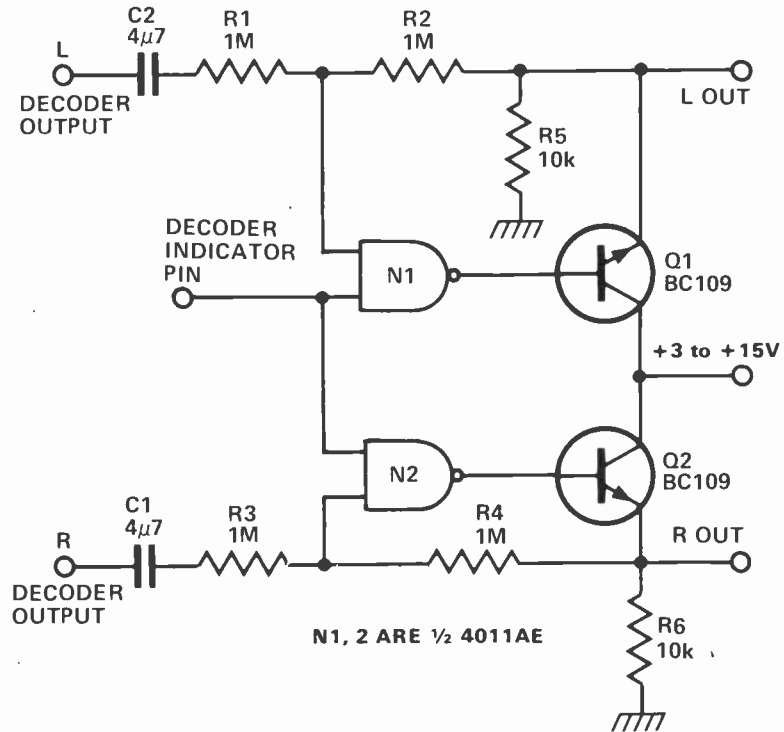
SIMPLE STEREO SWITCH

A device to switch the audio from a stereo tuner only when a stereo signal is being received.

Two CMOS NAND gates and two transistors are employed. One of the inputs from each gate is connected together and to the indicator output of the decoder IC.

The other gate inputs are connected to the emitter's of Q1 and Q2 respectively, by means of the feedback resistors R2 and R4. On reception of a stereo signal the indicator output of the decoder goes high and the feedback resistors bias the gates into the linear region passing the signal. On reception of a mono signal or interstation noise, no signals pass through the gates, the circuit providing a 'mute' function.

N.B. Some CMOS will not function in the linear mode. CMOS that *will not* work include R.S., Doram, Signetics. CMOS that *will* work include N.S., R.C.A., Motorola. All unused inputs should be connected to ground.

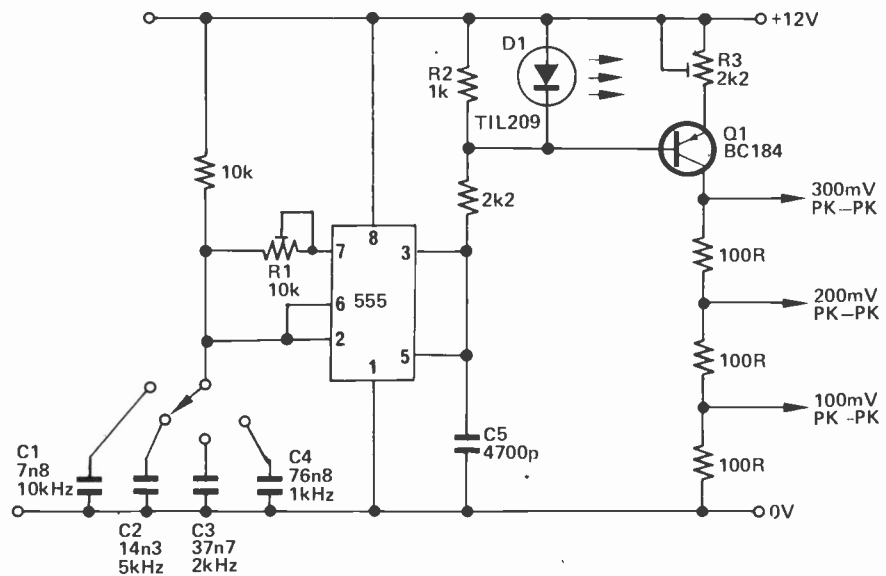


VOLTAGE AND FREQUENCY CALIBRATOR

This circuit provides simultaneous voltage and frequency calibrations by generation of a precision squarewave.

The 555 timer IC is used in a slightly unusual configuration, having the advantage that an exact 50:50 mark/space ratio may be attained by trimming R1. The frequency of oscillation may be set between 10 kHz and 1 kHz by switching timing capacitors C1-4. C5 decouples the internal reference potential-divider of the 555 from supply-transients.

The squarewave output from pin 3 of the IC, while stable in frequency, is not stable in peak-to-peak voltage as this depends on the supply voltage. This is used to switch on and off a temperature-compensated constant-current source Q1. R2 ensures that the current-source turns off completely when pin 3 goes high. The current-source output, trimmed by R3 to be exactly 1 mA, drives a resistor ladder network so that a series of precise squarewave voltages are generated. The advantage of current drive rather



than voltage drive for this sort of network is that calibration is much easier. A simple ladder network is shown by way of example, and more complex ones may simply be constructed to give a wider variety of output voltages.

The non-standard component values used were obtained by paralleling standard values. For the timing capacitors several in parallel had to be used, and only the resultant value is shown on the diagram.

TTLs by TEXAS			OP. AMPS			TV GAME I.C.			TRANSISTORS			RECTIFIER			BRIDGE RECTIFIERS		
7400	16p	7486	36p	74191	160p	301A	8 pin DIL	40p	BF194	13p	TIP41A	70p	2N4403	34p	BY100	31p	1A 50V 25p
7401	18p	7489	340p	74192	130p	324	14 pin DIL	120p	BF195	11p	TIP41C	81p	2N4427	97p	BY126	12p	1A 50V 27p
7402	18p	7490	430p	74193	160p	339	14 pin DIL	200p	BF197	19p	TIP42A	78p	2N5089	34p	BY127	12p	1A 100V 27p
7403	18p	7491	81p	74194	130p	709 B / 14	pin DIL	40p	BF200	40p	TIP2955	85p	2N5401	62p	1N4001	6p	1A 400V 31p
7404	25p	7492	55p	74195	96p	3140	8 pin DIL	25p	BF257	34p	TIP3055	70p	2N6107	70p	1N4002	6p	1A 800V 37p
7405	25p	7493	43p	74197	120p	747	14 pin DIL	70p	BF258	39p	TIS93	30p	2N6247	175p	1N4003	7p	2A 50V 37p
7406	45p	7494	96p	74198	220p	748 B / 14	pin DIL	40p	BF259	48p	ZTX108	11p	Comp to		1N4005	7p	2A 100V 44p
7407	45p	7495	75p	74199	220p	776	TO99	200p	AC188	20p	ZX300	16p	2N3055		1N4007	8p	2A 200V 52p
7408	25p	7496	90p	C-MOS ICs		1458	8 pin DIL	70p	AC188K	25p	ZTX500	19p	2N6254	140p	1N5401	15p	2A 400V 56p
7409	27p	7497	340p	4000	21p	3130	8 pin DIL	108p	AD149	70p	BF41	34p	2N6292	70p	1N5404	20p	3A 200V 70p
7410	18p	74100	116p	4001	21p	3140	8 pin DIL	108p	AD161	45p	ZN698	45p	40361	43p	1N5407	25p	3A 600V 75p
7411	26p	74104	60p	4002	21p	3900	14 pin DIL	108p	AF115	22p	BF888	37p	2N706	22p			4A 100V 84p
7412	27p	74105	60p	4006	120p			70p	AF116	22p	BFX30	36p	2N708	22p			4A 400V 90p
7413	38p	74107	36p	4007	21p			70p	AF117	22p	BFX84	30p	2N918	43p			6A 50V 96p
7414	96p	74109	96p	4009	67p			70p	AF139	43p	BFX85	30p	2N930	19p			6A 100V 96p
7416	34p	74110	55p	4011	21p			70p	AF239	48p	BFX86	30p	2N1131	20p			6A 200V 108p
7417	40p	74116	216p	4012	19p			70p	BC107/B	10p	BFX87	30p	40594	90p			6A 400V 120p
7420	18p	74118	90p	4013	55p			70p	BC108/B	10p	BFX88	30p	2N1304	70p			
7421	43p	74120	130p	4015	90p			70p	BC109/C	11p	BFY50	18p	2N1305	70p			
7422	24p	74121	32p	4016	54p			70p	BC147	9p	BFY51	16p	2N1306	70p			
7423	40p	74122	52p	4017	110p			70p	BC148	9p	BFY52	18p	2N1613	27p			
7425	33p	74123	73p	4018	21p			70p	BC149	10p	BRV39	45p	2N1711	27p			
7427	40p	74126	75p	4020	140p			70p	BC157	11p	BSX19	20p	2N1893	22p			
7428	39p	74132	75p	4022	18p			70p	BC158	13p	BSX20	20p	2N2219	25p			
7430	18p	74136	81p	4023	15p			70p	BC159	13p	BU105	175p	2N2222	25p			
7432	37p	74141	80p	4024	100p			70p	BC189C	15p	BU108	312p	2N2369	15p			
7437	37p	74142	300p	4025	18p			70p	BC171	12p	MJE340	70p	2N2484	32p			
7438	37p	74145	90p	4026	200p			70p	BC172	12p	MJE2955	130p	2N2904/A	25p			
7440	18p	74148	175p	4027	81p			70p	BC173	13p	MJE2955	130p	2N2905/A	25p			
7441	85p	74152	90p	4028	152p			70p	BC177	20p	MJE3055	80p	2N2906	25p			
7442	75p	74151	77p	4029	130p			70p	BC178	17p	MPSA06	40p	2N2026RB	9p			
7443	116p	74153	92p	4030	59p			70p	BC179	20p	MPSA12	62p	2N2926G	11p			
7444	116p	74154	164p	4042	150p			70p	BC182	12p	MPSA56	40p	2N3053	20p			
7445	108p	74155	96p	4043	218p			70p	BC183	12p	MPSU05	72p	2N3054	65p			
7446	108p	74156	96p	4046	150p			70p	BC184	14p	MPSU06	78p	2N3055	65p			
7447	90p	74157	97p	4047	110p			70p	BC187	32p	MPSU55	90p	2N3442	151p			
7448	85p	74160	120p	4049	68p			70p	BC212	14p	MPSU56	98p	2N3702	14p			
7451	20p	74161	120p	4050	50p			70p	BC213	12p	OC2B	90p	2N3703	14p			
7453	20p	74162	120p	4054	130p			70p	BC214	17p	OC35	90p	2N3704	14p			
7454	20p	74163	120p	4055	140p			70p	BC478	32p	OC71	25p	2N3705	14p			
7460	20p	74164	130p	4056	145p			70p	BC547	12p	TIP29A	50p	2N3706	14p			
7470	32p	74166	136p	4060	130p			70p	BC557	12p	TIP29C	62p	2N3708	14p			
7472	32p	74167	370p	4069	30p			70p	BCY70	22p	TIP30A	60p	2N3709	14p			
7473	36p	74174	131p	4071	29p			70p	BCY71	24p	TIP30C	72p	2N3707	14p			
7474	36p	74175	92p	4072	22p			70p	BD124	140p	TIP31A	56p	2N3773	270p			
7475	48p	74176	131p	4081	21p			70p	BD131	60p	TIP31C	68p	2N3866	92p			
7476	37p	74177	120p	4082	29p			70p	BD132	63p	TIP32A	63p	2N3904	27p			
7480	54p	74180	120p	4510	142p			70p	BD135	54p	TIP32C	85p	2N3905	25p			
7481	108p	74181	322p	4511	200p			70p	BD136	55p	TIP33A	97p	2N3906	22p			
7482	85p	74182	89p	4516	140p			70p	BD139	54p	TIP33C	120p	2N4058	19p			
7483	99p	74185	146p	4518	140p			70p	BD140	60p	TIP34A	124p	2N4060	19p			
7484	103p	74190	155p	4528	130p			70p	BDY56	225p	TIP34C	160p	2N4123	22p			
VOLTAGE REGULATORS			Fixed Plastic 3 Terminals			Fixed Plastic 3 Terminals			Fixed Plastic 3 Terminals			Fixed Plastic 3 Terminals			Fixed Plastic 3 Terminals		
1 Amp +ve			-ve			-ve			-ve			-ve			-ve		
5V	7805	130p	7905	215p	70-92	60p	78L05	60p	75491	84p							
12V	7812	130p	7912	215p	78L05	60p	78L08	60p	75492	104p							
15V	7812	150p	7915	215p	78L12	60p	78L12	60p	9368	200p							
18V	7818	150p	7918	215p	78L15	60p	78L15	60p	TIL209 Red	14p							
24V	7824	150p	7924	215p	78L15	60p	78L15	60p	TIL211 Green	36p							
LM309K	5V	1 Amp	TO3	150p	79L05	80p	79L05	80p	ORP12	75p							
LM309H	5V	100mA	TO5	97p	79L08	80p	79L08	80p	ORP61	90p							
TBA625B	12V	0.5A	TO5	106p	79L12	80p	79L12	80p	ORP61	90p							
LM323K	5V	3A	TO3	650p	79L15	80p	79L15	80p	Red	16p							
Variable:	723	14pin DIL	45p						Green	20p							
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8 pin	12p	16 pin	14p	28 pin	60p	3015F	200p	DL707	160p	DL707	160p	DL707	160p	DL707	160p	DL707	160p
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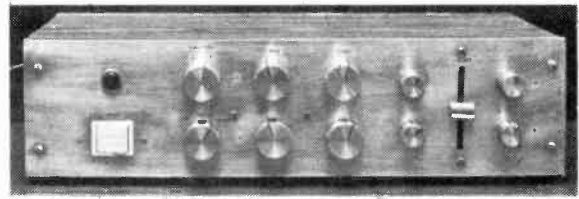
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JUNE ISSUE INCLUDES

RADIO & ELECTRONICS CONSTRUCTOR

IMPORTANT PROJECT... SIMPLE QUADRAPHONIC AMPLIFIER - Part 1

A 3-part article, this quadraphonic amplifier incorporates comprehensive tone and balance controls and takes advantage of the Motorola SQ decoder, i.e. type MC1312.



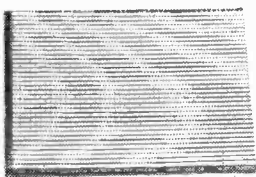
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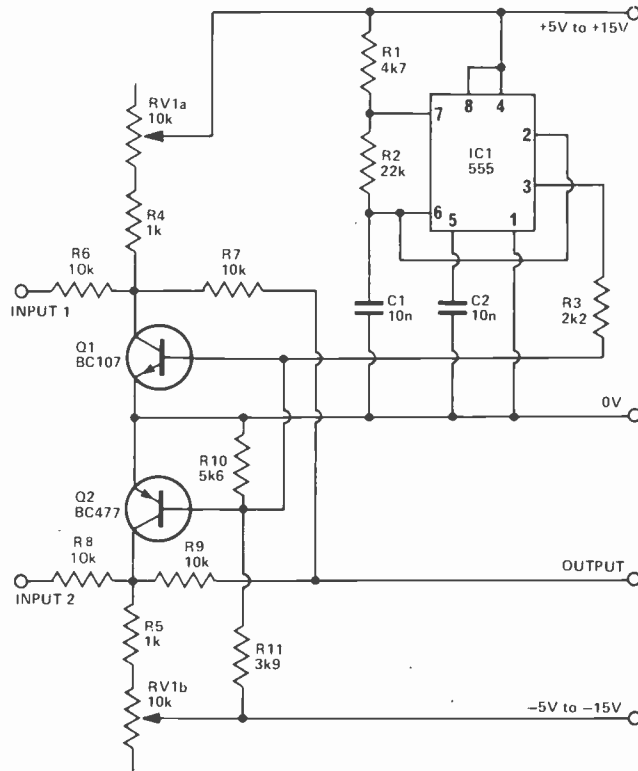
tech-tips

BEAM SPLITTER FOR OSCILLOSCOPE

The basis of the beam splitter is a 555 timer connected as an astable multi-vibrator, components R1, R2 and C1 being selected to give approximately equal high/low pulses of about 3 kHz.

Resistor R3 couples the output of the oscillator to the npn/pnp pair Q1 and Q2. When the output of the oscillator is low, resistors R10 and R11 allow Q2 to be on so that any signal applied to input 2 is effectively short-circuited via resistor R8 to the common line of the power supply. At the same time, the npn transistor Q1 is off, so that any signal at input 1, plus a positive voltage provided by RV1a and R4, appears at the output via R7.

Conversely, when the output of the oscillator is high, Q1 is biased on whilst Q2 is off. A signal at input 2 plus a negative voltage via RV1b and R5 appears at the output via R9. Thus signals at the two inputs are alternately displayed on the oscilloscope with a clear separation between them. The separation is controlled by the tandem potentiometer RV1a/b which also varies the amplitude of the traces.



A SIMPLE V.C.O.

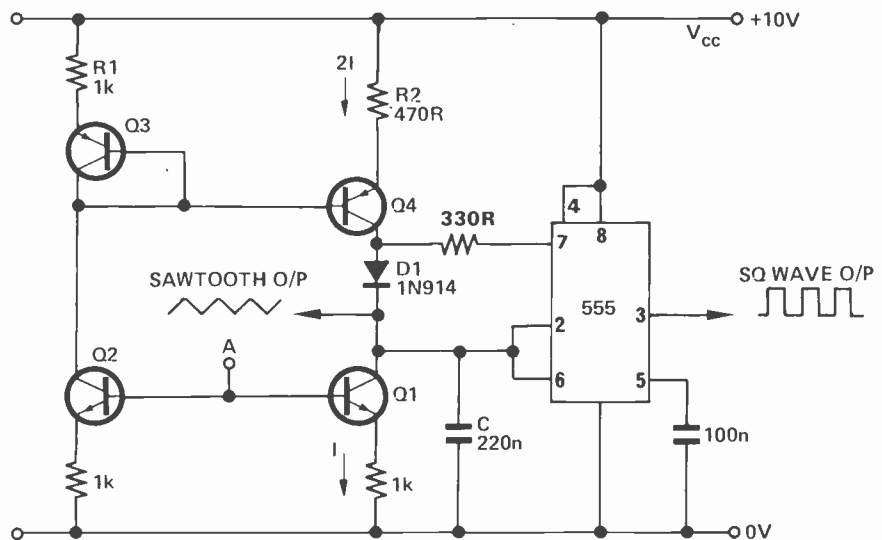
This circuit generates sawtooth and triangle waveforms at a frequency set by an external control voltage.

Current source Q1 draws a current I from timing capacitor C. Simultaneously current source Q2 draws the same current from current mirror Q3, Q4; this is set up (by R1 and R2) to deliver (from the collector of Q4) twice the current leaving Q2.

Hence C receives a current 2I from the top rail, at the same time delivering I to the bottom rail, the net effect being that the capacitor is charged by a constant current I, its voltage rising linearly until the 555's upper trigger point (at 2/3Vcc) is reached.

The output (pin 3) then goes low, as does the open-collector discharge output at pin 7. The latter shunts the output of the current mirror to earth, D1 becoming reverse-biased and isolating C.

Now only current source Q1 is connected to the timing capacitor which is now linearly discharged by current I. In this way C is alternately charged and discharged. When the voltage on C falls to the 555's lower trigger point at 1/3Vcc, the output and discharge pins go high, and the



cycle recommences; the repetition frequency is determined by the magnitude of I, which is set by the voltage applied at the input point A.

With the component values shown, the frequency range is from approx.

2.5 kHz to less than 10 Hz, as the control voltage varies from +10 V to zero; the frequency is directly proportional to the control voltage. Other ranges may be obtained by altering the value of C.

15 — 240 Watts!

HY5 Preamplifier

The HY5 is a mono hybrid amplifier ideally suited for all applications. All common input functions (mag Cartridge, tuner, etc.) are catered for internally, the desired function is achieved either by a multi-way switch or direct connection to the appropriate pins. The internal volume and tone circuits merely require connecting to external potentiometers (not included). The HY5 is compatible with all I.L.P. power amplifiers and power supplies. To ease construction and mounting a P.C. connector is supplied with each pre-amplifier.

FEATURES: Complete pre-amplifier in single pack — Multi-function equalization — Low noise — Low distortion — High overload — two simply combined for stereo

APPLICATIONS: Hi-Fi — Mixers — Disco — Guitar and Organ — Public address

SPECIFICATIONS:

INPUTS: Magnetic Pick-up 3mV, Ceramic Pick-up 30mV, Tuner 100mV, Microphone 10mV, Auxiliary 3-100mV, input impedance 47k Ω at 1kHz

OUTPUTS: Tape 100mV, Main output 500mV R.M.S.

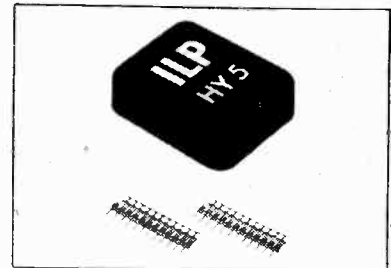
ACTIVE TONE CONTROLS: Treble \pm 12dB at 10kHz, Bass \pm at 100Hz

DISTORTION: 0.1% at 1kHz, Signal/Noise Ratio 68dB

OVERLOAD: 38dB on Magnetic Pick-up, **SUPPLY VOLTAGE:** \pm 16.50V

Price £5.22 + 65p VAT P&P free

HY5 mounting board B1 48p + 6p VAT P&P free.



HY30 15 Watts into 8 Ω

The HY30 is an exciting New kit from I.L.P. it features a virtually indestructible I.C. with short circuit and thermal protection. The kit consists of I.C., heatsink, P.C. board, 4 resistors, 6 capacitors, mounting kit together with easy to follow construction and operating instructions. This amplifier is ideally suited to the beginner in audio who wishes to use the most up-to-date technology available.

FEATURES: Complete kit — Low Distortion — Short, Open and Thermal Protection — Easy to Build

APPLICATIONS: Updating audio equipment — Guitar practice amplifier — Test amplifier — Audio oscillator

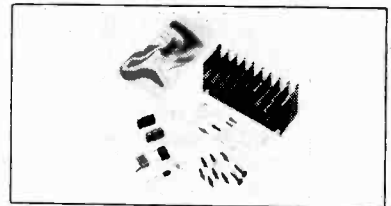
SPECIFICATIONS:

OUTPUT POWER: 15W R.M.S. into 8 Ω , **DISTORTION:** 0.1% at 15W

INPUT SENSITIVITY: 500mV, **FREQUENCY RESPONSE:** 10Hz-16kHz — 3dB

SUPPLY VOLTAGE: \pm 18V

Price £5.22 + 65p VAT P&P free.



HY50 25 Watts into 8 Ω

The HY50 leads I.L.P.'s total integration approach to power amplifier design. The amplifier features an integral heatsink together with the simplicity of no external components. During the past three years the amplifier has been refined to the extent that it must be one of the most reliable and robust-High Fidelity modules in the World.

FEATURES: Low Distortion — Integral Heatsink — Only five connections — 7 Amp output transistors — No external components

APPLICATIONS: Medium Power Hi-Fi systems — Low power disco — Guitar amplifier

SPECIFICATIONS: **INPUT SENSITIVITY:** 500mV

OUTPUT POWER: 25W RMS into 8 Ω , **LOAD IMPEDANCE:** 4-16 Ω , **DISTORTION:** 0.04% at 25W at 1kHz

SIGNAL/NOISE RATIO: 75dB, **FREQUENCY RESPONSE:** 10Hz-45kHz — 3dB

SUPPLY VOLTAGE: \pm 25V, **SIZE:** 105.50 x 25mm

Price £6.82 + 85p VAT P&P free



HY120 60 Watts into 8 Ω

The HY120 is the baby of I.L.P.'s new high power range, designed to meet the most exacting requirements including load line and thermal protection, this amplifier sets a new standard in modular design.

FEATURES: Very low distortion — Integral Heatsink — Load line protection — Thermal protection — Five connections — No external components

APPLICATIONS: Hi-Fi — High quality disco — Public address — Monitor amplifier — Guitar and organ

SPECIFICATIONS:

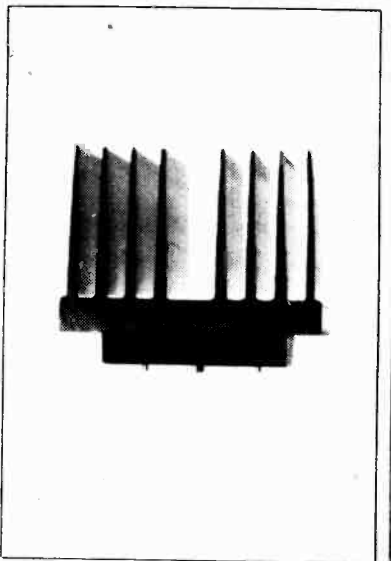
INPUT SENSITIVITY: 500mV

OUTPUT POWER: 60W RMS into 8 Ω , **LOAD IMPEDANCE:** 4-16 Ω , **DISTORTION:** 0.04% at 60W at 1 kHz

SIGNAL/NOISE RATIO: 90dB, **FREQUENCY RESPONSE:** 10Hz-45kHz — 3dB, **SUPPLY VOLTAGE:** \pm 35V

Size: 114 x 50 x 85mm.

Price £15.84 + £1.27 VAT P&P free.



HY200 120 Watts into 8 Ω

The HY200, now improved to give an output of 120 Watts, has been designed to stand the most rugged conditions, such as disco or group while still retaining true Hi-Fi performance.

FEATURES: Thermal shutdown — Very low distortion — Load line protection — Integral Heatsink — No external components

APPLICATIONS: Hi-Fi — Disco — Monitor — Power Slave — Industrial — Public address

SPECIFICATIONS:

INPUT SENSITIVITY: 500mV

OUTPUT POWER: 120W RMS into 8 Ω , **LOAD IMPEDANCE:** 4-16 Ω , **DISTORTION:** 0.05% at 100W at 1kHz

SIGNAL/NOISE RATIO: 96dB, **FREQUENCY RESPONSE:** 10Hz-45kHz — 3dB, **SUPPLY VOLTAGE:** \pm 45V

SIZE: 114 x 100 x 85mm.

Price £23.32 + £1.87 VAT P&P free.

HY400 240 Watts into 4 Ω

The HY400 is I.L.P.'s "Big Daddy" of the range producing 240W into 4 Ω ! It has been designed for high power disco or public address applications. If the amplifier is to be used at continuous high power levels a cooling fan is recommended. The amplifier includes all the qualities of the rest of the family to lead the market as a true high power hi-fidelity power module.

FEATURES: Thermal shutdown — Very low distortion — Load line protection — No external components

APPLICATIONS: Public address — Disco — Power slave — Industrial

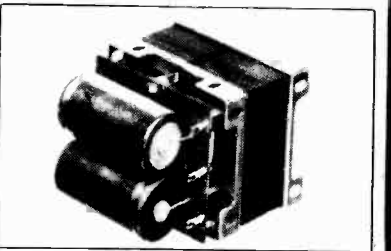
SPECIFICATIONS:

OUTPUT POWER: 240W RMS into 4 Ω , **LOAD IMPEDANCE:** 4-16 Ω , **DISTORTION:** 0.1% at 240W at 1 kHz

SIGNAL/NOISE RATIO: 94dB, **FREQUENCY RESPONSE:** 10Hz-45kHz — 3dB, **SUPPLY VOLTAGE:** \pm 45V

INPUT SENSITIVITY: 500mV, **SIZE:** 114 x 100 x 85mm

Price £32.17 + £2.57 VAT P&P free.



POWER SUPPLIES

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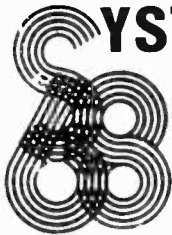
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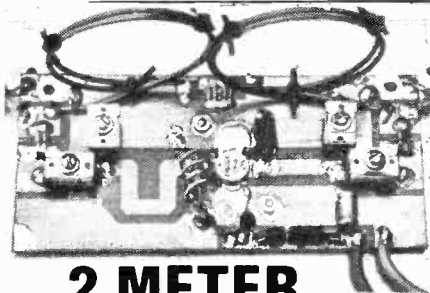
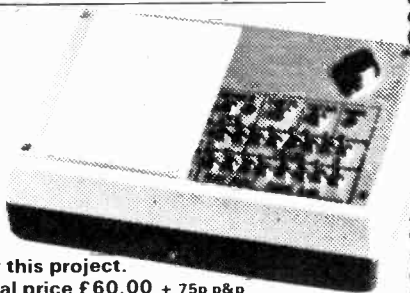
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2 METER POWER AMP
as described
SEPTEMBER ISSUE
(page 19)

2 METER POWER AMPLIFIER

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Detecknow- ledgey

Detecknowledgey? No, we haven't forgotten how to spell - a number of you thought we had during our 'Tecknowledgey' ad series last year - it is simply an expression of Ambit's specialist topic for Summer 1977: Metal Detectors. In other words, we want you to know that we know about them, not just selling them, but how they work, and why. It is surprising how few people can explain the theory and application of first principles to metal locators. Things like why a bunch of keys can sometimes fail to register, whilst a 2p piece indicates at 8-10" or why a lump of iron reacts as if it were non-ferrous. See our new publication on the subject, a concise reference work that sets out to explain from first principles the various phenomena you will encounter. £1.00 inc PP.

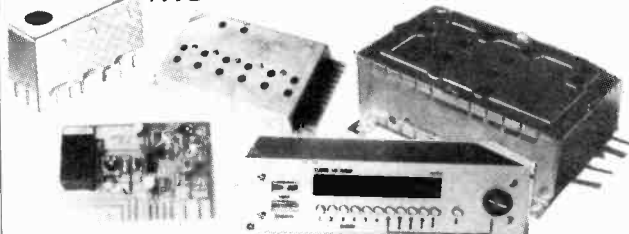
PW SEEKIT COMPLETE KIT

Speaking of technology - that is, the application of scientific discoveries to practical situations and every day equipment - the PW Seekit can claim to have used a few innovative features only achievable with devices that we have been largely instrumental in promoting, so it is not surprising we are able to offer complete kits, and all parts for this exciting project. The complete Seekit-kit costs £36.00 including PP AND VAT. This kit contains all electronic and plastic components (undrilled), bar things like glue, foil etc., which you are more than likely to have around the house. An SAE will bring you our new leaflet (A5 size please) with a more detailed list. Please note we can only consider service/testing on complete kits purchased from US.

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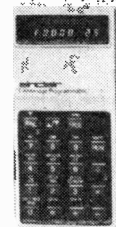
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We cannot solve the problems faced by individual readers building our projects unless they are concerning interpretation of our articles. When we know of any error we print a correction as soon as possible at the end of News Digest. Any useful addenda to a project will be similarly dealt with. We cannot advise readers on modifications to our projects.

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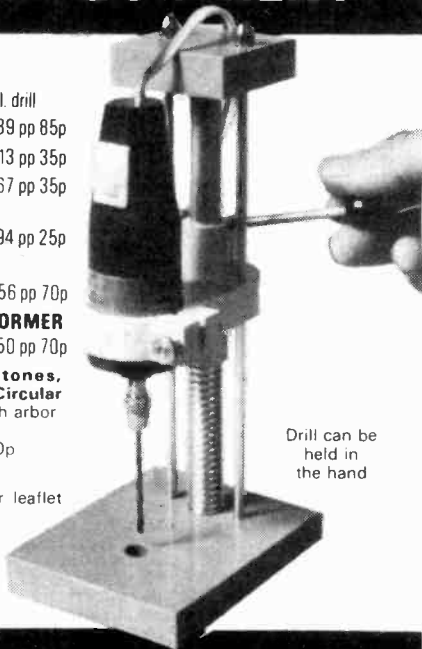
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INTROKIT £66.33, KBDKIT £66.50

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		Floppy disks SA400 Minifloppy	£495.00
		Floppy disks SA800 Standard	£625.00

BOOKS AND DATA

SC/MP Technical Description	£1.95
SC/MP Programmers Guide	£6.30
SC/MP Applications Guide	£2.95
SCRUMPI Data Sheets	£0.75
F8 Data (Xerox)	£0.75

All prices exclude VAT at 8%

SYSTEM



BYWOOD ELECTRONICS is supporting ETI System 68 with a range of MPU kits, VDU kit, Case kit and many add-ons. An expanding range of compatible products building up to a complete home or small business computer system from about £300 upwards

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17-0268C	31-way Socket	7	1.00
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	Complete kit		39.92

6800 CPU CARD

To be released when published

SC/MP CPU CARD

SC/MP MPU, 256 bytes RAM, 512 bytes PROM (VDUBUG), address decoding chips, IC sockets, PCB, etc. Interfaces to System 68 VDU and its case/connector system. VDUBUG program uses ASCII keyboard input to allow SC/MP programs to be written, displayed and executed.

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CLOCK CHIPS & KITS

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MM5312	7seg + BCD 4 DIGIT ONLY	5.65	
MM5313	7 seg + BCD	6.50	
MM5314	7seg + BASIC CLOCK	4.88	7.00
MM5315	7 seg + BCD REST ZERO	£6.50	
MM5316	Non-mpx ALARM	10.17	
MM5318	7seg + BCD External digit select	4.50	8.00
MM5371	ALARM 50Hz	10.15	
MM5377	CAR clock Crystal control LCD	8.40	
MM5378	CAR clock Crystal control LED	7.55	14.00
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AY5 1230	7seg ON and OFF ALARM	5.25	TBA

All above clock kits include clock PC board, clock chip, socket and CA3081 driver IC. MH15378 also includes crystal and trimmers. When ordering kit, please use prefix MHI, e.g. MHI 5309.

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DL747 750, 746 0 6"	2.45	DL747E 1.50

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MHI727 / 6	12.00	MHI727E / 6	6.00
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Soldercon strip skts, 50 pins	0.30

CLOCK MODULES

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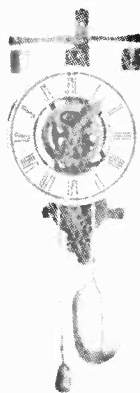
Module only	7.97
MTX1001 transformer	0.90
Kit including case	13.00

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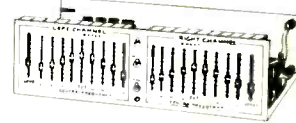


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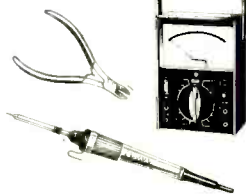
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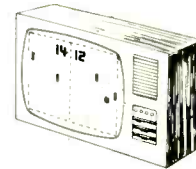
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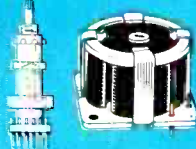


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