

AD-773 769

WAMP: A USERS MANUAL FOR THE
WIRE ANTENNA MODELING PROGRAM

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Prepared for:

Coast Guard

10 December 1973

DISTRIBUTED BY:

NTIS

National Technical Information Service
U. S. DEPARTMENT OF COMMERCE
5285 Port Royal Road, Springfield Va. 22151

Unclassified

Security Classification

AD-773 769

DOCUMENT CONTROL DATA - R & D

Security classification of title, body of abstract and indexes annotation must be entered when the overall report is classified

1. FUNDING ACTIVITY (Corporate author) University of California Lawrence Livermore Laboratory	2a. REPORT SECURITY CLASSIFICATION Unclassified 2b. GROUP
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3. REPORT TITLE
WAMP: A Users Manual for the Wire Antenna Modeling Program

4. DESCRIPTIVE NOTES (Type of report and inclusive dates)
Final

5. AUTHOR(S) (First name, middle initial, last name)
Fred J. Deadrick & Edmund K. Miller

6. REPORT DATE December 10, 1973	7a. TOTAL NO. OF PAGES 106 / 10	7b. NO. OF ILLS. 10
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8a. CONTRACT OR GRANT NO. USCG MIPR #Z-70099-2-20568 b. PROJECT NO.	9. ORIGINATOR'S REPORT NUMBER UCID-30084 10. OTHER REPORT NUMS (Any other numbers that may be assigned this report)
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11. DISTRIBUTION STATEMENT
Unlimited

12. SUPPLEMENTARY NOTES	13. SPONSORING/MONITORING AGENCY NAME(S) U. S. Coast Guard
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14. ABSTRACT

Program WAMP is a Wire Antenna Modeling Program written in FORTRAN IV and applicable to arbitrary antenna and support structures. WAMP models an antenna as a series of interconnected straight wire segments, and solves the electromagnetic boundary value problems by numerically evaluating an electric field integral equation.

Antennas may be analyzed in free space, over a perfect ground, a radial ground screen or in the presence of any homogeneous media. Antenna input impedance, current distributions, near-electric fields and far-field radiation patterns are also calculated.

The users manual covers both the theory and numerical techniques employed in WAMP. The program's input variables are defined, and illustrative examples are used to demonstrate the program's capabilities.

14 KEY WORDS	LIST A		LIST B		LIST C	
	HOLE	WT	HOLE	WT	HOLE	WT
Antenna Computer Modeling Computer Program Thin-Wire Integral-Equation Program Users Manual Ground Interaction Ground Screen						

I. INTRODUCTON

WAMP (Wire Antenna Modeling Program) is a general purpose frequency domain antenna modeling computer program. Antennas of arbitrary form and orientation may be modeled and analyzed in both free-space and in the presence of a conducting half space.

The WAMP users manual is written to serve two purposes. The first, and prime purpose, is to provide the user with a basic working knowledge of the program's input requirements and capabilities. Examples will be employed to illustrate the use of the program. The second purpose of this manual is to present a brief description of the theoretical formulation behind the numerical techniques used in the WAMP code.

Section II, which immediately follows, presents a brief development of the thin-wire electric field integral equation (EFIE). This is followed in Section III by a description of the numerical techniques used to solve the EFIE for a segmented structure, and the methods employed to model an antenna in the presence of an imperfectly conducting half-space. (A listing of the program's subroutines and their function is provided in Appendix A).

Sections IV, V, AND VI are devoted to applications. Section IV describes in detail the program input parameters needed to model an antenna structure, Section V contains a series of illustrative examples designed to demonstrate the WAMP program, and Section VI concludes with a section on modeling guidelines and special calculations.

II. THEORETICAL DEVELOPMENT OF THE ELECTRIC FIELD INTEGRAL EQUATION

Discussion of the Boundary Value Problem. The solution to Maxwell's equation which satisfies specified behavior of the fields at given locations or boundaries, requires treatment of a boundary value problem. Classically, such problems have been successfully treated only for special boundary surfaces to which the method of separation of variables could be applied. This eliminated complex surface shapes from consideration and restricted the three-dimensional solutions to the sphere and various simple sphere modifications. This situation resulted principally from inability to calculate the eigenfunctions for the various special geometries, although considerable progress has been made in this area in recent years due to availability of numerical techniques and more sophisticated computers. An additional problem arises in these special coordinate systems since the boundary condition equations may be coupled, so that a solution of an infinite matrix is required, in principle at least, rather than the term-by-term solution which is encountered for the spherical case. At any rate, the shapes to which the classical analysis are restricted force the use of a different approach to problems which involve complex geometrical shapes.

Integral Equations Formulation. Because of the problems cited above, general three-dimensional boundary value problems require a basically numerical approach. This approach usually begins from an integral-equation viewpoint rather than the differential equation

approach implicit in the classical method. There is, however, no unique integral equation formulation for a given problem, so some leeway in treatment is afforded at the outset. For reasons to be clarified later, our attention is restricted here to the electric field integral equation (EFIE). As a starting point, the electric field \vec{E} due to a volume current distribution \vec{J} is written by means of Green's dyadic as

$$\vec{E}(\vec{r}_0) = \iiint_V i\omega\mu_0 \vec{J}(\vec{r}) \cdot \vec{G}(\vec{r}, \vec{r}_0) dV \quad (1)$$

where \vec{r}_0 and \vec{r} are the observation and source points, respectively, and the Green's dyadic is expressed in the usual notation as

$$\vec{G}(\vec{r}, \vec{r}_0) = -(1/4\pi) [\vec{I} + (1/k^2) \nabla\nabla] g$$

where

$$g = \exp(-ik|\vec{r}-\vec{r}_0|)/|\vec{r}-\vec{r}_0|$$

and \vec{I} is the unit second-rank tensor. The suppressed time variation is $\exp(i\omega t)$ with ω the radian frequency. The plane wave propagation constant is k , and is related to ϵ_0 and μ_0 , the permittivity and permeability of free space respectively, and ω by

$$k = \omega\sqrt{\mu_0\epsilon_0}$$

When the current distribution is limited to the surface of a perfectly conducting body, Equation (1) becomes

$$\vec{E}(\vec{r}_0) = \iint_S i\omega\mu_0 \vec{K}(\vec{r}) \cdot \vec{G}(\vec{r}, \vec{r}_0) dA \quad (2)$$

with \vec{K} the surface current density. If this surface current is induced by an incident electric field \vec{E}^I , then an integral equation for the unknown surface current \vec{K} can be obtained from Equation (2) and the boundary condition that

$$\hat{n}(\vec{r}_0) \times [\vec{E}^S(\vec{r}_0) + \vec{E}^I(\vec{r}_0)] = 0 \quad (3)$$

with $\hat{n}(\vec{r}_0)$ a surface normal at \vec{r}_0 and \vec{E}^S the scattered field of the individual current distribution. Equating \vec{E}^S of Equation (3) with \vec{E} of Equation (2) leads to

$$-\hat{n}(\vec{r}_0) \times \vec{E}^I(\vec{r}_0) = \hat{n}(\vec{r}_0) \times \int_S \int_S i\omega\mu_0 \vec{K}(\vec{r}) \cdot \vec{G}(\vec{r}, \vec{r}_0) dA \quad (4)$$

Thin Wire Approximation. Upon restricting our attention to circular cross-section bodies of small diameter compared with the wavelength, the azimuthal current may be neglected, and Equation (4) becomes

$$\hat{n}(\vec{r}_0) \times \vec{E}^I(\vec{r}_0) = \hat{n}(\vec{r}_0) \times \frac{1}{4\pi} \int_S \int_S \{i\omega\mu_0 K_s(\vec{r}) [\hat{s} + \frac{\hat{s} \cdot \nabla \nabla}{k^2}] g(\vec{r}, \vec{r}_0)\} dA \quad (5)$$

where \hat{s} is the unit tangent vector at \vec{r} pointing in the direction of the current. A scalar integral equation for the current is obtained by taking the dot product of Equation (5) with the unit tangent vector \hat{s}_0 at the observation point \vec{r}_0 as

$$\hat{s}_0 \cdot \vec{E}^I(\vec{r}_0) = \frac{1}{4\pi} \int_S \int_S i\omega\mu_0 K_s(\vec{r}_0) [\hat{s} \cdot \hat{s}_0 + (\hat{s} \cdot \nabla)^2 \frac{1}{k^2}] g(\vec{r}, \vec{r}_0) dA \quad (6)$$

If the assumption is now made that the surface current, K_s is independent of the azimuthal variable, Equation (6) can be written

$$\hat{s}_0 \cdot \vec{E}^I(\vec{r}_0) = \frac{1}{4\pi} \int_s \int_s a i\omega\mu_0 K_s(s) \int_0^{2\pi} [\hat{s} \cdot \hat{s}_0 + \frac{1}{k^2} \frac{\partial^2}{\partial s \partial s_0}] g(\vec{r}, \vec{r}_0) d\phi ds \quad (7)$$

where a is the wire radius and the s integration is over the entire length of wire.

A final approximation is that the current may be realistically represented as a filament of strength $I_s(s) = 2\pi a K_s(s)$ flowing on the wire axis while the field is evaluated on the wire surface, allowing Equation (7) to be written as

$$\hat{s}_0 \cdot \vec{E}^I(\vec{r}_0) = (i\omega\mu_0/4\pi) \int_L I(s) [\hat{s} \cdot \hat{s}_0 + \frac{1}{k^2} \frac{\partial^2}{\partial s \partial s_0}] g(\vec{r}, \vec{r}_0) ds \quad (8)$$

where $|\vec{r} - \vec{r}_0|$ is now measured from the wire axis, or source point, to the observation point on the wire surface, which can thus never be closer than the wire radius a . By considering the current as a tubular sheet on the wire axis while evaluating the electric field on the wire surface, one can resolve the ambiguity in the azimuth involved. The form of Equation (8) is not changed using this convention, but the interpretation of the tangential field evaluation is simplified when non-parallel, non-planar wires are considered.

III. METHOD OF NUMERICAL SOLUTION

A numerical solution to an integral equation may be undertaken by the method of moments. This is a well-founded mathematical technique briefly stated as a method for finding the unknown by forcing the integral equation which is solved by the method of moments described below.

Reduction to Linear System (Collocation) - Equation (8) may be written symbolically as:

$$L(f) = g \tag{9}$$

following Harrington's (1968) notation. The solution of Equation (8) (or of Equation 9) is obtained by the method of moments. An intuitive approach to solving Equation (9) for the unknown function f is to set f equal to a constant f_j within N subintervals of the domain of L , and to require Equation (9) to be satisfied at N points over the range of L , thus acquiring N equations in the f_j unknowns. This is a specialized application of the method of moments which is more generally written as follows. Let

$$f = \sum \alpha_n f_n$$

with the basis functions f_n defined in the domain of L so that Equation (9) may be written

$$\sum \alpha_n L(f_n) = g \tag{10}$$

Then, with the set of weighting functions w_m , defined in the range of L , the inner product is formed as

$$\sum \alpha_n \langle w_m, L(f_n) \rangle = \langle w_m, g \rangle \tag{11}$$

where $m = 1, 2, 3, \dots, N$, Equation (11) can be written in matrix form as

$$[G_{mn}][\alpha_n] = [s_m] \tag{12}$$

where

$$G_{mn} = \langle w_m, L(f_n) \rangle$$

and

$$s_m = \langle w_m, g \rangle$$

and the matrix G_{mn} is referred to in this case as the structure matrix. If the inverse of G_{mn} exists, then the α_n can be found and thus the function f which is the desired solution, for any specified source function s_m .

The proper choice of weighting functions and basis functions, as well as the subsectioning of the domain of L is not an obvious one. Although there is some leeway in the matter, careful consideration of the physics of the problem and the nature of the expected solution will show that some representations for the f_n will be more efficient than others in terms of computer time and accuracy. Constant, linear, quadratic, trigonometric and Fourier series have all been used for this role. The weighting functions have generally been more restricted in choice than f_n . The special case, $w_n = f_n$, is referred to as Galerkin's method (see, for example, Harrington, 1968). More often, the weights are δ -functions, a method referred to as collocation, so that the inner product (Eq. 11) merely becomes the sequence of values $L(f_n)_m$ and g_m . These are, respectively, the tangential electric fields due to current segment n at observation point m and the tangential incident electric field at observation point m . It is interesting to note that Galerkin's method is equivalent to the Rayleigh-Ritz variational method. Harrington (1968) thoroughly discusses the method moments.

Current Expansion - The WAMP thin-wire program employs the collocation method with constant, sine and cosine terms for the f_n segment or current basis function, i.e.,

$$I(s) = \sum_{i=1}^N U_i(s) [A_i + B_i \sin k(s-s_i) + C_i \cos k(s-s_i)] =$$

$$\sum_{i=1}^N U_i(s) I_i(s) \tag{13}$$

where $U_i(s)$ is 1 when s is on segment i and zero otherwise. Equation (13) is disadvantageous because three constants are required to specify the current on each segment, so that apparently $3N$ linear equations need be solved. This prompts the question, why use the sine and cosine terms for the current at all? There are a number of reasons for using this sinusoidal current expansion, but they are essentially summarized by the observation that a more physically realistic current solution is obtained with this expansion.

It is not necessary to employ the integral equation itself to find the extra unknowns introduced by the sinusoidal expansion. Two of the three constants for each segment may be obtained by requiring the current on adjacent segments to satisfy some specified mutual conditions. The extrapolated current from a given segment must match the center current values on two adjacent segments to satisfy the required condition for two-wire junctions in the thin-wire program. Junctions of three or more wires are handled in similar fashion (Maxum, et al, 1969).

The sinusoidal current expansion appears to make the system of equations resulting from collocation somewhat more involved, but the required computer time is not significantly increased when compared with the same number of current unknowns without using the sinusoidal expansion. Other current expansion functions - linear, quadratic, Fourier series - could be used in place of the constant-sine-cosine expression, but this particular expansion has a number of additional advantages over the other possibilities mentioned. For instance, a solution for the current to a specified accuracy for a half-wave dipole scatterer and antenna requires the fewest current segments using the sinusoidal expansion (Neureuther, et al, 1968). This advantage can be expected to carry over to more complex geometries. Second, the solution will more accurately exhibit the required dependence on wire radius (Andreason, 1968) because the constant current term produces infinite tangential electric field on the current axis, as opposed to the sine and cosine terms which do not.

Third, the parallel and perpendicular electric field components (due to the sine and cosine current terms) and the radial field components (due to constant current terms) may be analytically evaluated. This eliminates the necessity for extensive numerical integration to evaluate all the elements in the coefficient matrix G_{mn} . Only the tangential electric field excited by the constant current terms requires numerical integration and this is handled by applying a Romberg variable-interval width technique (Miller, 1970) to the difference integrand.

Calculation of the Structure Matrix - It is worthwhile to discuss here the form of the matrix elements which result from applying the method of collocation to the integral Equation (8). Each entry G_{ij} in the structure matrix represents the tangential electric field at observation point i on the structure produced by unit current flowing on segment j . The boundary condition on the tangential electric field is enforced at each observation point. The collocation method of solving the integral equation is thus basically one of calculating electric field components at specific points due to the induced current on the structure.

It was stated that the thin-wire approximation involves the explicit assumption that the effects of azimuthal currents can be neglected in comparison with those of axially directed currents and that, in addition,

the cylindrical tube of axial current has no azimuthal dependence. The former assumption allows us to consider only one current component rather than two, while the latter provides partial justification for reducing the surface integral to a line integral. It may be deduced from an examination of Equations (7) and (8), however, that even where K_s is independent of ϕ the kernel of the integration equation depends in general upon both ϕ and s . However, the integrand is independent of ϕ in the special case where the observation point is located on the axis of a linear tube of current, and the ϕ integration of Equation (7) may be replaced without approximation by the factor 2π .

We choose to locate the observation points where the tangential electric field is to be calculated on the axis rather than on the surface of each wire segment. The ϕ integration in Equation (7) is thus exact for the self-field as well as the mutual fields for all current segments having a common axis. In addition, the possible ambiguity involved in evaluating the incident field over a 2π variation in ϕ on the wire surface is resolved. As a final point, the observation point is always at least as far as the wire radius from the source point.

When the mutual fields of non-axially aligned current segments are required, the ϕ integration is not so simply performed. And if no approximation were used, the ϕ integration would require numerical evaluation. The most obvious approach is to then consider the tubular current source to approximate a linear filament on the wire axis, a procedure which again replaced the ϕ integration by a 2π factor. Unfortunately, this approximation eliminates the influence of the wire radius from all mutual field terms on the phase change and geometrical attenuation of the field caused by the separation of the source and observation points.

An alternative to the above method is replacement of the current tube by a current filament which is not located on the wire axis but is displaced in distance from it by the wire radius. The direction of displacement is perpendicular to the plane of the wire axis and the line joining the observation point and wire axis midpoint (the observation point for the self-term field). The geometry of this method is shown in Figure 1.

To summarize briefly, the surface integral is reduced to a line integral by neglecting azimuthal* currents and azimuthal variation of the axial current. Self-field terms are calculated with the observation point on the axis of a cylindrical current tube, while mutual field terms are calculated at the same observation point with the current represented as a filament displaced from the wire axis by the wire radius.

* Taken here to mean the direction measured along the intersection of the current tube surface with a plane perpendicular to the axis of the current tube.

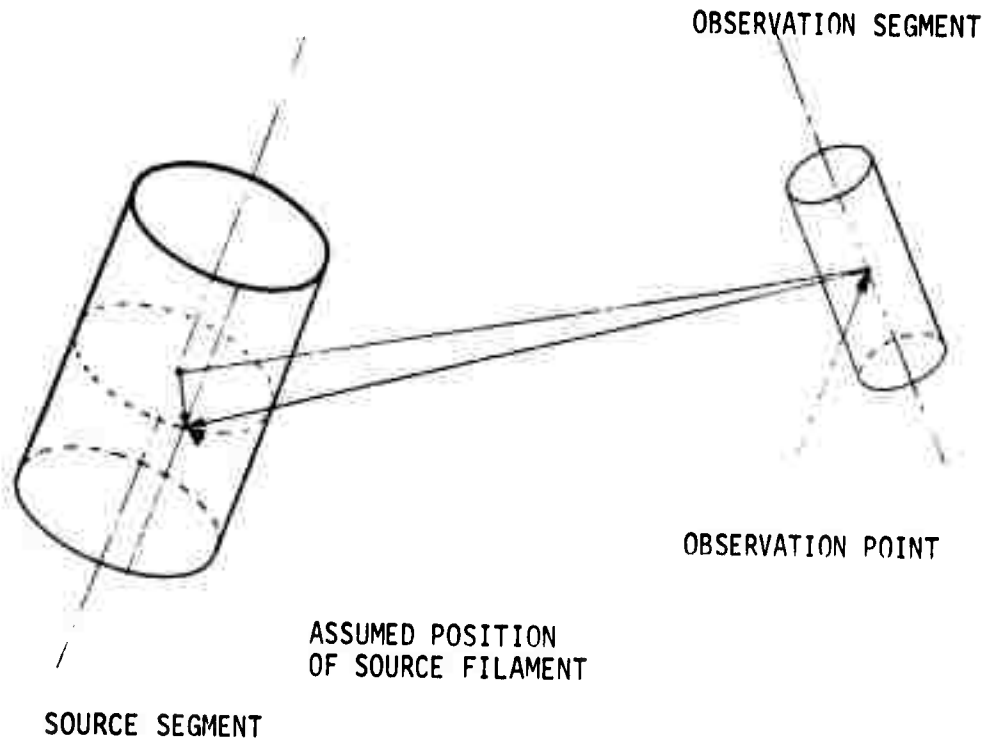


Figure 1a. Thin-wire Current Approximation

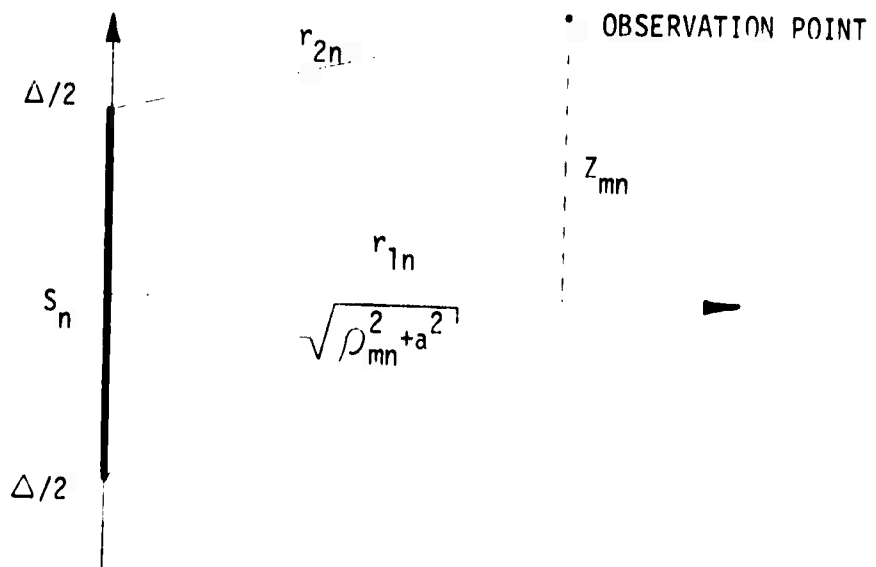


Figure 1b. Geometric Parameters for Field Evaluation.

We can now write the integral Equation (8) in the form

$$E_m = (i\omega\mu_0/4\pi) \sum_{n=1}^N \int_{\Delta S_n} \left[\hat{s}_m \cdot \hat{s}_n + \frac{1}{k^2} \frac{\partial^2}{\partial s \partial s_0} \right] g(r_m, r) I_n(s) ds \quad (14)$$

$m=1, 2, \dots, N$ observation points

where Δs_n denotes the length of source segment n ,

$$g(r_m, r) = \frac{\exp[-ik \sqrt{\rho^2 + (s_m - s)^2}]}{\sqrt{\rho^2 + (s_m - s)^2}}$$

and it should be noted that the integration over L has been reduced to a summation of N separate straight-wire segment integrals. It is convenient to rewrite Equation (14) in terms of cylindrical coordinates referred to the wire segment being integrated. Then we get

$$E_m = (i\omega\mu_0/4\pi) \sum_{n=1}^N \int_{\Delta z_n} \left[\hat{z}_n \cdot \hat{s}_m - \frac{1}{k^2} \frac{\partial^2}{\partial z_n \partial s_m} \right] g(r_{mn}, z_n) I_n(z_n) dz_n \quad (15)$$

where

$$g(r_{mn}, z_n) = \exp(-ikr_{mn})/r_{mn}$$

and

$$r_{mn} = \sqrt{(z_{mn} - z_m)^2 + \rho_{mn}^2 + a_n^2}$$

a_n is the radius of wire segment n , and ρ_{mn} and z_{mn} are the radial and z -coordinates of the observation point at the center of segment m referred to the midpoint of source segment n .

Ground Effects - The integral equations (14) and (15) apply only to wire structures located in free space, or more generally in a homogeneous medium having electrical constants μ and ϵ . Location of the structure near the interface between two electrically dissimilar media, however, leads to reflected fields which can modify the free space current distributions, and thus an additional term must be added to the integral equation for it to apply to the antenna problems of interest here.

Historically, the basic work on the solution of this problem was formulated in 1909 by A. Sommerfeld (1964). By deriving field expressions for vertical and horizontal electric and magnetic Hertzian dipoles in free space as influenced by the ground plane, Sommerfeld obtained the Green's functions which permit equations (14) or (15) to be rigorously extended to the interface problem.

Unfortunately, the solution of the Sommerfeld integrals requires a double numerical integration which can be very costly in terms of computer time. Miller, et al., (1972a, 1972b) describes an approximation that is used in the WAMP program. The approximation is to represent ground reflected fields via plane wave reflection coefficients. This procedure is basically quite simple, involving decomposition of the image fields into TM and TE modes relative to the vertical plane containing the image and observation points, after which the reflected fields are obtained by multiplying the image fields by the appropriate reflection coefficients. The advantage of this technique lies in the fact that it represents but a simple extension to the free space integral-equation treatment.

Impedance Loading - The discussion has been thus far limited to the case of a perfectly conducting element. The approach may also be generalized to allow for lumped loading of the structure by introducing a voltage drop term in the integral equation. If the impedance loading per unit length on segment m is z_m , then Equation (15) becomes:

$$E_m - I_m z_m = \text{same R.H.S. as Equation (15)}$$

Solution of the System of Equations - The solution of the integral equation is reduced by the method of collocation to the problem of solving a linear system of equations for the N sampled current values. The problem is far from being resolved at this point, however, since the linear system which is generated may contain a very large number of complex unknowns. A numerical solution of such a system would be impractical without the availability of a large-core, high-speed digital computer. An additional factor of importance in the linear system solution is the use of an efficient and accurate numerical technique; an especially significant aspect of the problem since the solution time increases as the cube of the number of unknowns (see, for example, Forsythe & Moler, 1967). The method used to solve the linear system of equations is discussed next.

The final step in solving the integral equation for the induced current is a matrix multiplication of the solution or inverse matrix times the source vector.

The induced current solution can be written in the form

$$[I] = - [G]^{-1} [E] \quad (16)$$

The values then obtained for the sampled current values at the centers of the N segments on the structure are used to obtain the current interpolation functions for each segment.

The solution technique employed in the WAMP program to solve the system of equations is the Gauss-Doolittle method (Ralston, 1965). The basic step in the Gauss-Doolittle method is factorization of the structure matrix [G] into the product of an upper triangular matrix [U], and a lower triangular matrix [L], i.e.

$$[G] = [L][U] \quad (17)$$

and thus,

$$[L][U][I] = -[E] \quad (18)$$

Let

$$[U][I] = [F] \quad (19)$$

so that

$$[L][F] = -[E] \quad (20)$$

Next, equation (20) is solved for the elements of [F] by forward substitution using the known elements of [E]. Equation (19) is then solved for the elements of [I] by backward substitution using the known elements of [F].

Matrix Symmetry - Many antenna structures exhibit geometric symmetries which may be exploited to a great advantage in terms of computer storage requirements and execution times. Structure symmetry can be used to increase the calculation efficiency by reducing the time required to fill the structure matrix G, and by decreasing the computations required to factor and solve the linear system of equations. The matrix fill time, for example, is reduced by a factor on the order of $1/n$, and the inversion time by $1/n^2$, for a structure with n-fold rotational symmetry.

The reduction in matrix fill time is easily understood. Consider a structure having 4-fold rotational symmetry, such as Figure 2 below,

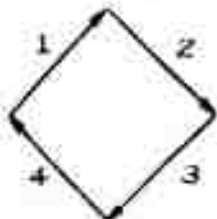


Figure 2. Structure with 4-fold Rotational Symmetry.

which will have a structure matrix [G] of the form:

$$[G] = \begin{bmatrix} g_{11} & g_{12} & g_{13} & g_{14} \\ g_{21} & g_{22} & g_{23} & g_{24} \\ g_{31} & g_{32} & g_{33} & g_{34} \\ g_{41} & g_{42} & g_{43} & g_{44} \end{bmatrix} \quad (21)$$

where

$$\begin{aligned} g_{11} &= g_{22} = g_{33} = g_{44} \\ g_{12} &= g_{23} = g_{34} = g_{41} \\ g_{13} &= g_{24} = g_{31} = g_{42} \\ g_{14} &= g_{21} = g_{32} = g_{43} \end{aligned} \quad (22)$$

Thus, rather than their being $N^2 = 16$ matrix elements to calculate, there are only N^2/n or 4 to obtain, and the reduction in the structure matrix fill time and storage is on the order of $1/n$.

This version of the WAMP code is set up to handle rotationally symmetric structures with elements located on the axis of rotational symmetry. An example of such a structure is shown in Figure 3.

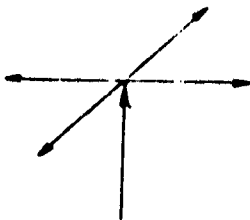


Figure 3. A structure with 4-fold rotational symmetry and center wire. (Turnstile Antenna)

The structure matrix is of the form given by (23):

$$[G] = \begin{bmatrix} A_1 & A_2 & A_3 & A_4 & | & B_1 \\ A_4 & A_1 & A_2 & A_3 & | & B_2 \\ A_3 & A_4 & A_1 & A_2 & | & B_3 \\ A_2 & A_3 & A_4 & A_1 & | & B_4 \\ \hline C_1 & C_2 & C_3 & C_4 & | & D_1 \end{bmatrix} \quad (23)$$

$$B_1 = B_2 = B_3 = B_4 \quad (24)$$

and

$$C_1 = C_2 = C_3 = C_4$$

which is stored in the computer as:

$$\begin{bmatrix} A_1 & A_2 & A_3 & A_4 & B_1 \\ C_1 & C_1 & C_1 & C_1 & D_1 \end{bmatrix} \quad (25)$$

The solution to a system of equations of this symmetric form may be accomplished by the following operations. First write the matrix equation, noting the partitioning:

$$\begin{bmatrix} A_{n \times n} & B_{n \times m} \\ \hline C_{m \times n} & D_{m \times m} \end{bmatrix} \begin{bmatrix} X_n \\ \hline Y_m \end{bmatrix} = \begin{bmatrix} U_n \\ \hline V_m \end{bmatrix} \quad (26)$$

where we have the relationships:

$$A_{m \times n} X_n + B_{n \times m} Y_m = U_n$$

and

$$C_{m \times n} X_n + D_{m \times m} Y_m = V_m \quad (27)$$

or

$$X_n = A_{m \times n}^{-1} U_n - A_{n \times n}^{-1} B_{n \times m} Y_m \quad (28)$$

and

$$(D_{m \times m} - C_{m \times n} A_{n \times n}^{-1} B_{n \times m}) Y_m = V_m - C_{m \times n} A_{m \times n}^{-1} U_n \quad (29)$$

To solve the equations (28) and (29) for the X_n and Y_m matrices, we perform the following sequence of operations:

1. Factor $A_{n \times n}$ in place.
2. Compute $E_{n \times m} = A_{n \times n}^{-1} B_{n \times m}$ by solving $A_{n \times n} E_{n \times m} = B_{n \times m}$ in place and storing the result in B.
3. Compute $F_{m \times m} = D_{m \times m} - C_{m \times n} E_{n \times m}$ and store the result in D.
4. Now factor $D_{m \times m}$ in place.
5. Then compute $W_n = A_{m \times n}^{-1} U_n$ in place.
6. Compute $Z_m = V_m - C_{m \times n} W_n$ and store in V_m .
7. Solve $F_{m \times m} Y_m = Z_m$ for Y_m and store in V_m .
8. Finally, compute $X_n = W_n - E_{n \times m} Y_m$ and store the result in U_n .

Solution Sequence - We may now list the sequence of operations performed in WAMP to obtain the solution for an antenna structure. The basic steps are listed below:

1. Initialize WAMP and read in the antenna input frequency and the media over which the antenna is located, i.e., free space, over a perfect ground or a finite homogeneous media.

2. Convert a physical description of the antenna structure given in terms of the cartesian end-point coordinates of wire elements into a series of interconnected short straight-line wire segments. The segments will be used to describe the structure's line integral path.
3. With the structure defined, the structure matrix is then computed by the methods outlined earlier in this section. The electric field integral equation is first evaluated for the antenna in free space, where the co-location technique is used to fill in all the entries of the structure matrix.

If it is selected to model the antenna over a homogeneous halfspace, the structure matrix entries are then recalculated and modified by the ground reflected terms. The reflected fields are found by first computing the perfect ground image fields, and then these fields are modified by the appropriate ground reflection coefficient which is a function of the ground media and the specular angle between the source and the observation points.

4. If segments on the structure are impedance loaded, e.g. an inductive load used for resonating an antenna, the loading may be simply included in the structure's impedance matrix at this point.
5. The structure matrix may then be factored as outlined in the previous section. Structure symmetries are exploited to minimize the amount of core storage required and the amount of computations needed.
6. Once the structure matrix is factored, the antenna source vector is setup by applying a tangential E-field to those segments excited.
7. The factored structure matrix and the source field vector are then solved to yield the unknown currents at the center of each of the N segments.
8. Once the segment currents are known, the antenna input impedance and admittance may be calculated.
9. The coefficients A, B and C of the current basis function, Equation (13) are next determined and their values printed out.
10. The two last steps in the program are the computations of the near and far electric fields. These computations are only made if selected as program options.

This completes the description of the WAMP code. The next section describes in some detail the input parameters required to use the program.

IV. PROGRAM INPUTS

WAMP is written to conform with ANSI X 3.0 FORTRAN, and allows for in core execution on a 147K CDC 3300 computer system. All program inputs are via punched card data decks, and all program outputs appear on an on-line printer output. A main executive program, and 20 subroutines comprise this version of the WAMP code.* Most of the program's inputs are requested by the MAIN program, however, several subroutines (DATAGN, CMSETUP and NEFLD) may also request additional data inputs. Figure 4 provides a basic flow diagram of the sequence of data inputs, which is followed by more complete details on the program's input variables.

DATA DECK STRUCTURE

The structure of the input data deck for WAMP is illustrated in the flow chart of Figure 4. Note that not all inputs are requested by the program, and they are dependent on the run options selected by the second input data card. Additional details on each data input parameter follows.

Run Comments - (Format [10A8]) The first card of a data deck is used to provide 80 characters of run identification. Any text which fits on one card may be used, and the message will appear on the first page of the program output.

Run Options (Format [16I5]) The option card allows the user to select several program options at execution time. The parameters are listed below:

- NPRINT --- Controls level of printed output. Input Range 0-2, with more detailed outputs are given for higher numbers. Typically NPRINT = 1.
- ILOAD --- If the structure has impedance loaded elements, make ILOAD = 1. The load values will be read in later.
- IPGND --- If you want to model the structure over a perfect ground plane, set IPGND = 1.
- IGSCRN --- If IGSCRN = 1, a radial wire ground screen is modeled. Screen parameters are read-in in subroutine CMSETUP later.

* A listing of WAMP is included in Appendix A. A brief description of each subroutine is also included.

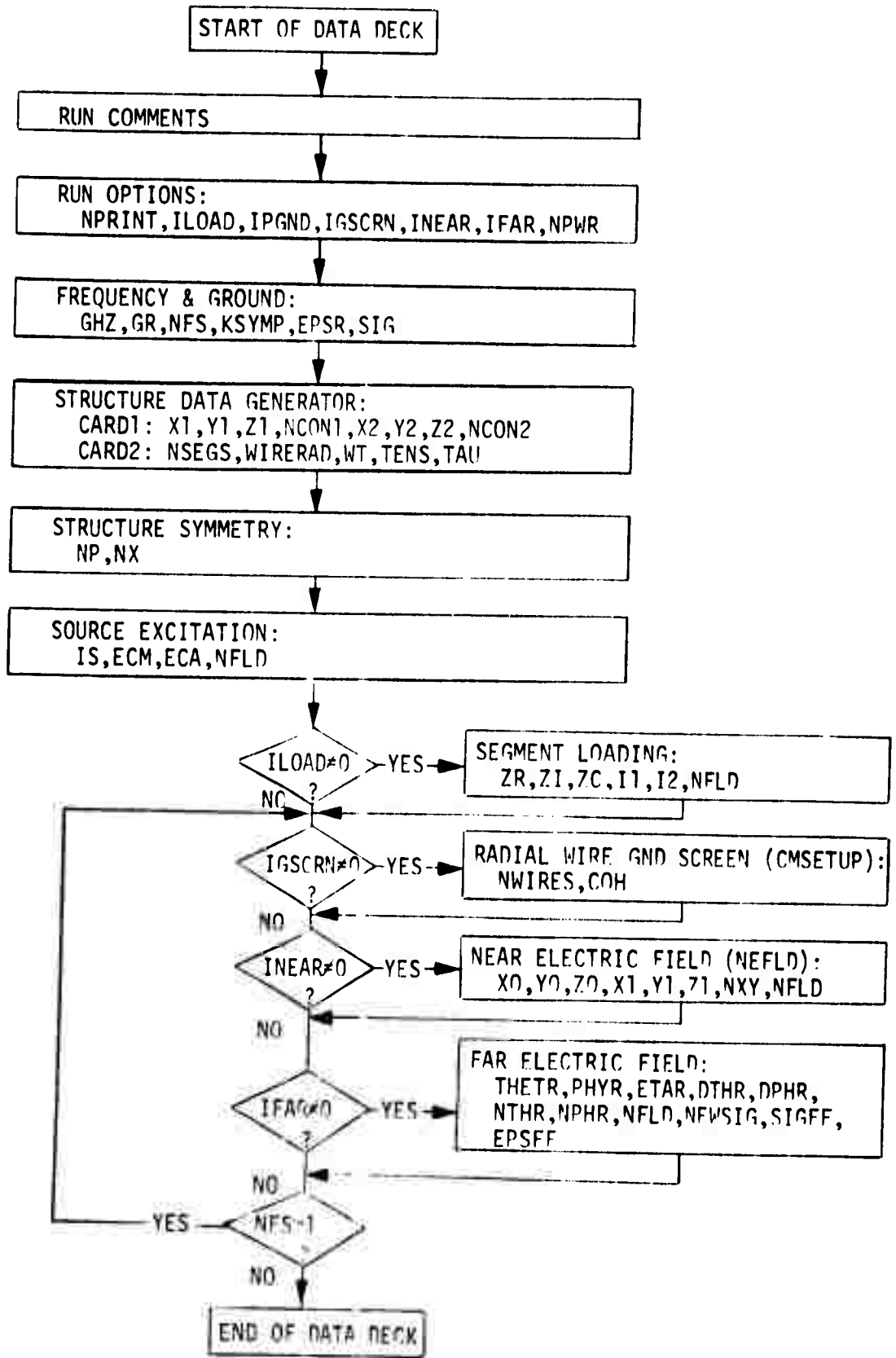


Figure 4. Flowchart of WAMP Input Data Deck Cards

- INEAR --- If INEAR = 1 the near field subroutine is called. The location of the field observation points to be evaluated are requested by the NEFLD subroutine.
- IFAR --- Far field calculations are performed if IFAR = 1. The far field observation angles are read in later.
- NPWR --- If NPWR = 1, the total input power to the antenna is normalized to 1 watt. All field values will also be normalized to this power.

Frequency and Ground (Format [2F10.5, 2I5, 2F10.5]) The frequency and ground card allows the user to specify the frequency at which the analysis is to be performed, a Δ frequency, and the number of Δ frequency steps to be evaluated. The user also has the option of performing the analysis in free-space or over a conducting half space. If the latter is selected, the ground media parameters must be supplied.

- GHZ --- Input frequency in Giga-Hertz.
- GR --- Δ frequency in GHz. If NFS is greater than 1, the frequency is changed by $\text{GHz} = \text{GHz} + \text{GR}$ each frequency step.
- NFS --- NFS is the number of frequency steps, and it must be greater than or equal to 1.
- KSYMP -- For analysis in free space, set KSYMP = 1. For an analysis over a halfspace, set KSYMP = 2. KSYMP must be either 1 or 2.
- EPSR -- If KSYMP = 2, then read in the value of the relative dielectric constant of the ground media.
- SIG --- If KSYMP = 2, then read in the value of the ground conductivity expressed in mhos/metre.

Structure Data Generator Inputs - After the frequency and ground cards are read, the main program calls the DATAGN subroutine. The purpose of this subroutine is to transform a physical model of an antenna structure into quantities which describe the structure to the WAMP code. Basically, all structures are modeled by straight-line wire elements. (A catenary element is a special feature allowed by this subroutine and is described in more detail in Section VI). The elements in turn are subdivided into a number of straight-line segments, and it is the segments which are used as the structure descriptors to the program.

Each wire segment is specified by its center-point coordinates, its orientation angles, its length, and its radius. In addition to the segment's physical parameters, electrical inter-connection data must also be provided. The two arrays, ICON1 and ICON2 of COMMON BLOCK/1/ are used to store the connection data relative to the negative and positive ends respectively of each segment.

All structure elements are specified to the program in terms of their cartesian end point coordinates, with dimensions given in metres. Elements are described to the DATAGN subroutine by specifying on one card the two end point coordinates, and interconnection data, and on a second card the number of segments to be used to describe the element, the element's wire radius, plus some additional details if a catenary is to be modeled, or if a variable length segment is to be used to model the element.

The connection data must conform to the following rules: Given a positive reference direction of the i -th segment defined by α_i and β_i and the arrow as illustrated in segment coordinate system of Figure 5, ICON1(i) must contain the index of the segment to which the negative end of the i -th wire is connected. A multiple connection is identified by assigning a unique negative number to the endpoint connection value of each segment connected to the junction, and an unconnected segment is assigned a value 0 at the unconnected end. One rule which must be observed is that if two segments are connected and the negative or positive ends coincide, as illustrated by Example 3 of Figure 6, this junction although not a multiple junction must also be assigned a unique negative number. Segments which are grounded must be given an ICON value equal to the segment's own index, see Example 4, Figure 6.

ICON2(i) array is similar and refers to the positive endpoint.

The input variables NCON1 and NCON2 allow the user to specify the ICON1 value (negative end) of the first segment of an element, and NCON2 allows the user to specify ICON2 value (positive end) of the last segment of the element. These rules are illustrated by a few examples as shown in Figure 6.

Thus the data generator input variables are described as follows:

DATAGN Card 1 (Format [3F10.5, I5, 3F10.5, I5]):

X1, Y1, Z1	---	Cartesian coordinates of the negative end point of the line element. Dimensions are in metres.
NCON1	---	Specifies the ICON1 value for the first segment of the element.
X2, Y2, Z2	---	Cartesian coordinates of the positive end point of the line element.
NCON2	---	Specifies the ICON2 value of the last segment of the element.

DATAGN Card 2 (Format [I5, 4F10.5]):

NSEGS* --- Number of segments in the line element.
WIRERAD --- Segment wire radius given in metres.
WT --- Wire weight given in pounds/metre. (Needed only if a catenary element is desired.)
TENS --- Wire tension in pounds, which is needed only for a catenary. If $TENS \leq 1.$, no catenary is used.
TAU --- Segment length expansion (contraction) factor.

Note: additional details on this factor are given later in Section VI and Appendix B.

Structure Symmetry (Format [I6I5]) After the antenna structure has been described, structure symmetries must be specified. The WAMP code is set up to exploit either no symmetry or up to 12 sectors of rotational symmetry. In addition, elements on the axis of rotational symmetry may also be used. The symmetry card requires the following parameters:

NP --- NP equals the number of segments in a rotationally symmetric sector. (Excluding segments on axis of symmetry.)
NX --- NX equals the number of segments on the axis of rotational symmetry.

Note that the program will work for the special case where $NP = N$ (no symmetry) and $NX = 0$. $NP + NX$ must be less than or equal to 22 for this version of WAMP.

If structure symmetry is to be exploited, a formalism exists which must be followed if a proper structure matrix is to be set up. Basically, the rules for setting up the proper structure symmetry are as follows:

1. All elements in a sector of rotational symmetry must be completely specified before going to the next sector.

* Note: If NSEGS is a positive number, the DATAGN subroutine jumps back to request an additional pair of line element cards to specify the next element of the structure. Input continues until a NSEGS (negative number of segments) is used to specify the last element of the structure. At this point, the program control returns to the main program.

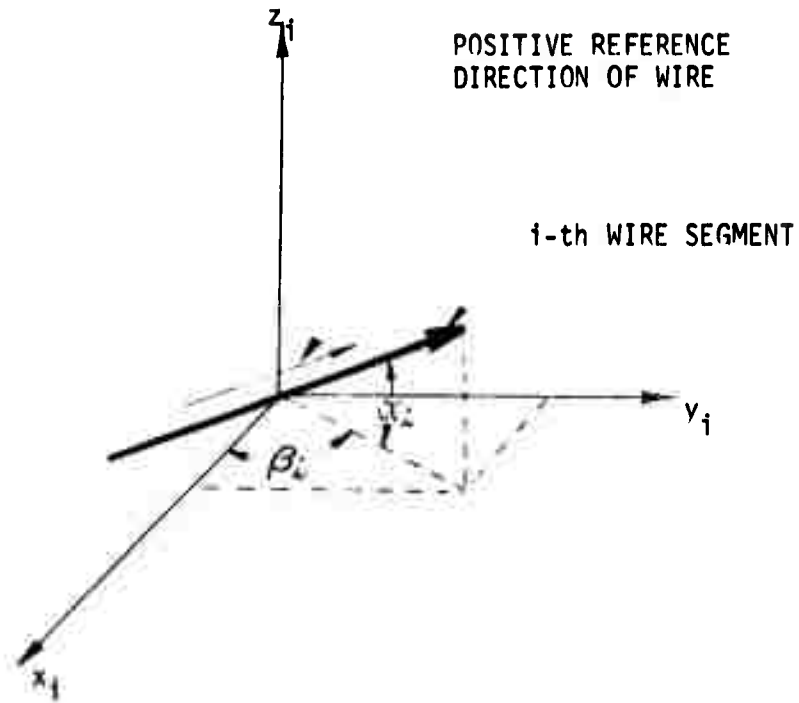


Figure 5a. Segment Coordinate System.

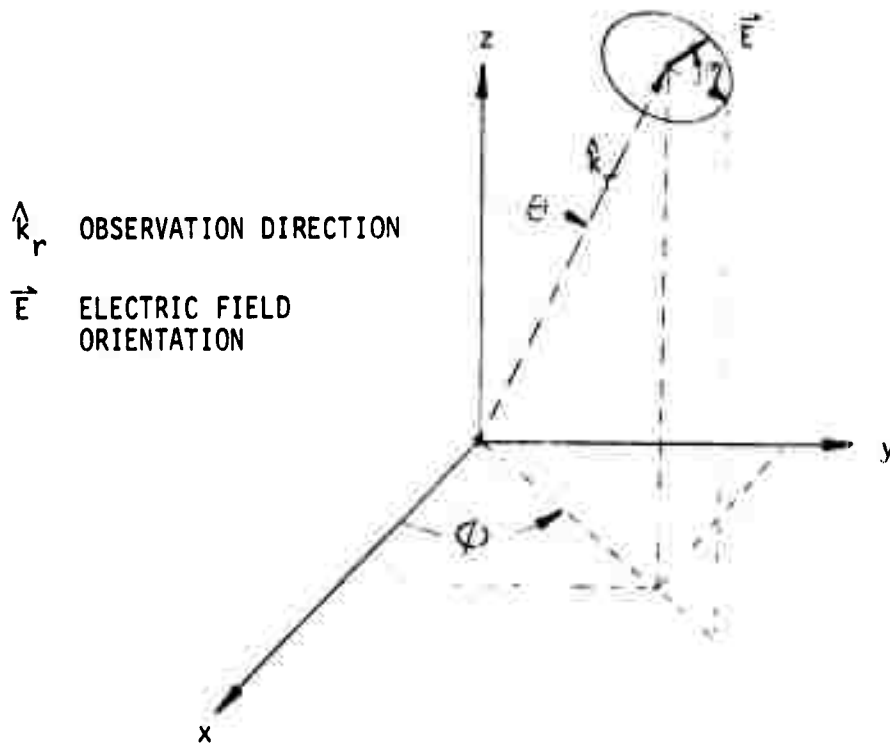


Figure 5b. Field Coordinate System

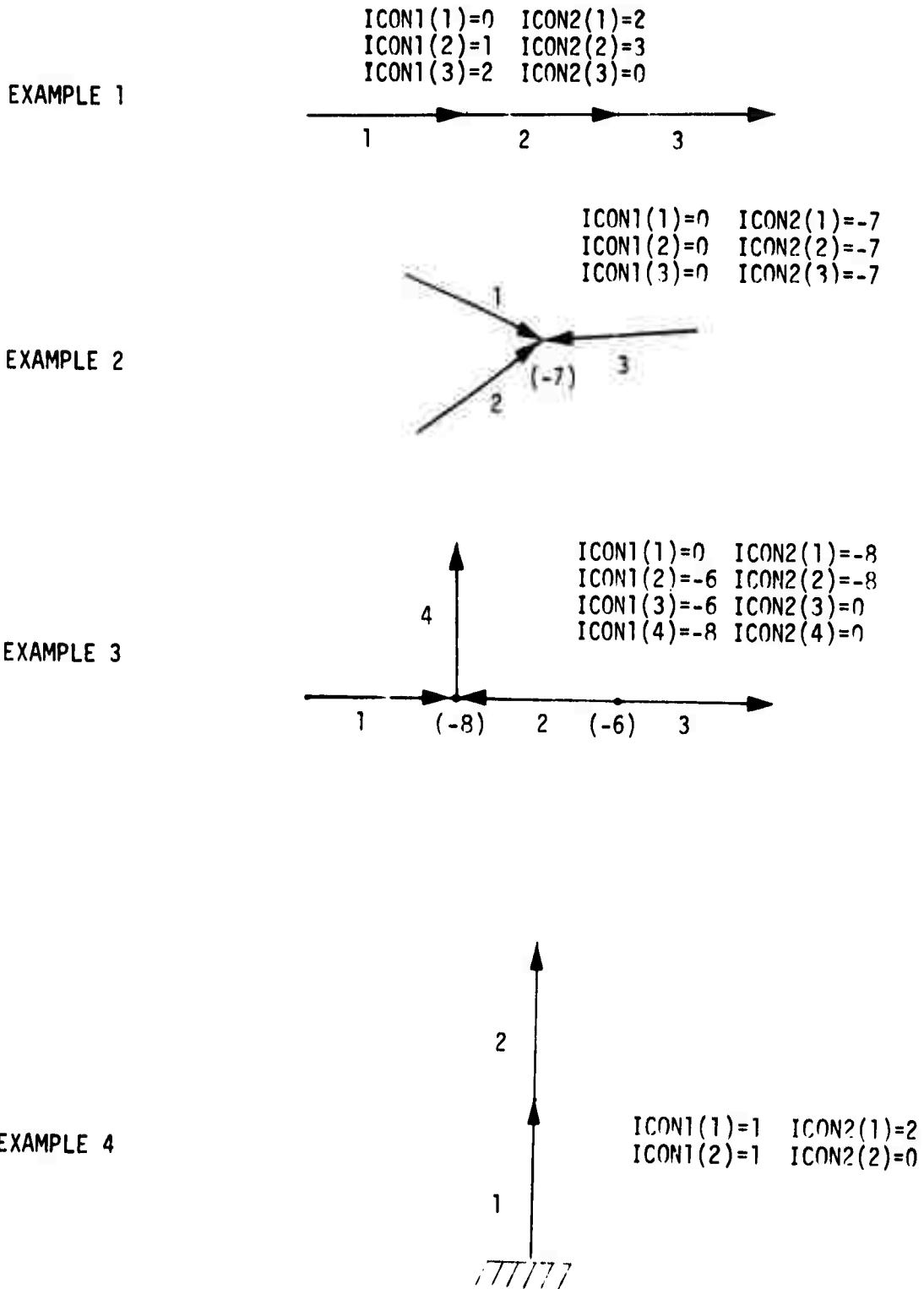


Figure 6. Examples of the ICON Connection Values for Various Types of Segment Junctions.

2. One may progress in either a clockwise or counter clockwise fashion around the axis of symmetry in describing the structure.
3. Elements located on the axis of rotational symmetry must be specified last to the data generator subroutine.

Source Excitation (Format [I5, 2F10.5, I5]) The source card allows the user to specify the segments which are driven. Any of the segments may be driven, and no symmetry of the excitation is assumed. In order to compute a correct input impedance, a 1.0 V at 0° phase angle source must be used.

- IS --- Segment number to which the source is applied.
- ECM --- Magnitude of driving source voltage in volts.
- ECA --- Phase of source in degrees.
- NFLD --- If NFLD = 1, an additional source card is read in. Input of source cards continues until NFLD = 0. This feature is used to specify multiple excitations.

This completes a normal data deck. Additional inputs may be required if certain options are selected on the run option card. These inputs are described below.

Segment Loading (Format [3E10.3, 3I5]): If ILOAD on the RUN OPTION card \neq 0, the segment loading option is selected. Resistive and reactive loading of segments is allowed. Symmetric loading of rotationally symmetric segments is assumed, and only one sector of symmetric loads need be specified. The inputs are listed below:

- ZR -- Resistance value in ohms distributed on each of the specified segments.
- ZI -- Inductance value in henrys distributed on each of the specified segments.
- ZC -- Series capacitance in Farads on each segment (Note: if ZC = 0. On input, no capacitive loading is included.)
- I1,I2 -- Specify the range of segments numbers which are loaded. All segments from I1 to I2 each receive the above load values.
- NFLD -- If NFLD \neq 0, an additional load card may be specified.

Radial Wire Ground Screen (Format [I5, E10.5]) If IGSCRN on the RUN OPTION card $\neq 0$, a radial wire ground screen model is placed in parallel with the normal ground media. The ground screen parameters listed below are read in by subroutine CMSETUP.

NWIRES -- Number of radial wires.

COH -- Wire radius in metres of the radial wires.

Near-Electric Field (Format [6F10.5, 2I5]) Subroutine NEFLD is used to compute the near E-field at specified points in the structure's coordinate system. NEFLD is called only if INEAR = 1 on OPTION CARD.

X \emptyset , Y \emptyset , Z \emptyset --- Are the initial field evaluation coordinates.
(metres)

X1, Y1, Z1 --- Are the final field evaluation coordinates. (metres)

NXY --- NXY + 1 field evaluation points are made along the straight line connecting point \emptyset with point 1.

NFLD --- If NFLD $\neq 0$ an additional near field evaluation path may be specified.

Far-Electric Field (Format [5F10.5, 4I5, 2F5.1]) A far field radiation pattern will be computed if IFAR on the OPTION CARD = 1. A provision is made to allow a far field calculation over a media which is different than the media over which the antenna is located. The farfield inputs, which refer to Figure 5, are listed below:

THETR --- Initial Theta angle in degrees.

PHYR --- Initial Phi angle in degrees.

ETAR --- Polarization angle Eta in degrees. (See Figure 5)

DTHR --- Delta Theta step size in degrees.

DPHR --- Delta Phi step size in degrees.

NTAR --- Number of Theta angle steps.

NPHR --- Number of Phi angle steps.

NFLD --- If NFLD $\neq 0$, another far field card may be read in. Up to five cards may be specified.

NEWSIG --- If NEWSIG = 1, new values of sigma and epsilon will be read in for the far field.

SIGFF --- Far field sigma value in mhos per metre.

EPSFF --- Relative dielectric constant for far field.

Note that if a new value of sigma and epsilon are not requested, the far field ground media is the same as the ground media of the antenna.

Multiple Data Decks - The above cards complete the data for one specific structure. Multiple data decks may be stacked one behind another to provide for multiple runs. The program control returns to read the next comment card, and if an end-of-file is encountered, the run is terminated.

V. SAMPLE PROBLEMS

Perhaps the easiest way to gain a working familiarity with the WAMP code is to use it to model some simple antenna structures which are familiar, and which we can compare the computed results with known data. This section illustrates the use of WAMP by a series of examples. The input data decks and the pertinent output will be shown.

Example 1 - Half Wavelength Horizontal Dipole - A half wavelength electric dipole is a good place to start. For this example, the antenna specifications are given as follows:

Frequency - 10 MHz

Length - $\lambda/2 = 15$ meters

Wire radius - 8.3 mm ($\Omega = 15$)

Height above ground - 10 meters

Ground media - $\epsilon_r = 25$, $\sigma = 10^{-2}$ mhos/m.

The first step is to describe the structure in terms of straight line elements whose end point coordinates are given in terms of the system's cartesian coordinates. For the dipole, this is simple, and is illustrated in Figure 8.

The input data deck may now be set up. (Refer to the previous section for a more detailed description of the input parameters)

Card 1: Comment Card

Halfwave Horizontal Dipole -- Example 1A

Card 2: Run Options

NPRINT = 2 Get a printout

ILOAD = 0 No loading

IPGND = 0 No perfect ground

IGSCRN = 0 No ground screen

INEAR = 0 No near field calculations

IFAR = 0 No far field calculations

NPWR = 0 Don't normalize input to 1 watt

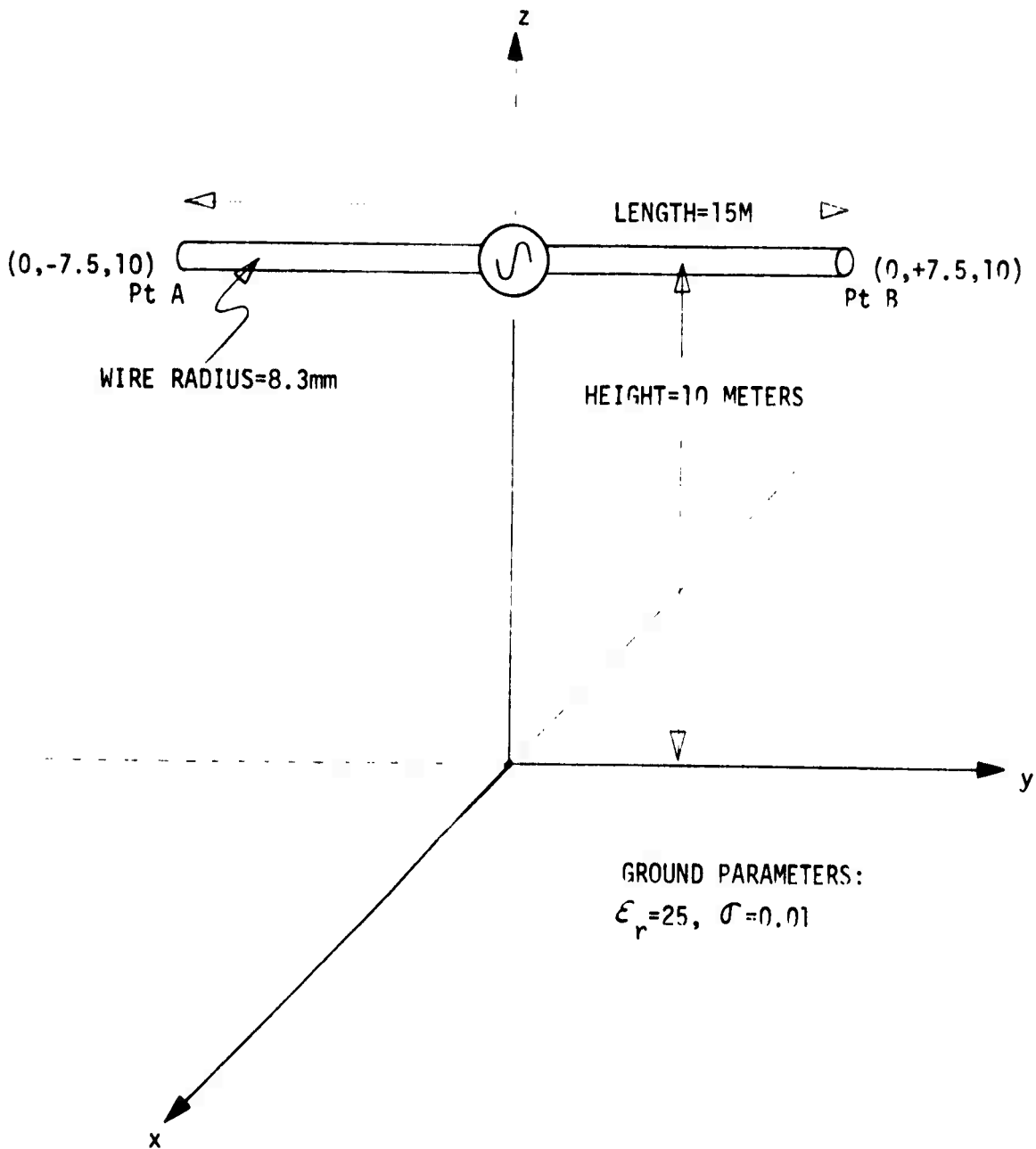


Figure 8. Half Wavelength Horizontal Dipole for Example 1.

We have chosen a very simple example, and none of the program's options have been selected.

Card 3: Frequency and Ground

GHZ = 0.01 Input frequency = 10 MHz
GR = 0. No frequency steps
NFS = 1 One frequency
*KSYMP = 1 First model antenna in free space
EPSR = 25. Ground dielectric constant
SIG = 0.01 Ground conductivity mhos/m

*Note that since the analysis is first being made in free space, ϵ_r and σ of the ground are not required at this point, but were included for later use.

Cards 4 and 5: Data Generator Inputs

CARD 4	X1 = 0	Coordinates
	Y1 = -7.5	of end point
	Z1 = 10.	A
	NCON1 = 0	End point A is an open circuit
	X2 = 0	Coordinates
	Y2 = 7.5	of end point
	Z2 = 10.	B
	NCON2 = 0	End point B is an open circuit
CARD 5	NSEGS = -7	Model with 7 segments (number is negative to end data generator inputs)
	WIRERAD = 0.0083	The radius of the dipole is 8.3 mm
	WT = 0.	No catenary
	TEN = 0.	
	TAU = 0.	No variable length segment, all =

The DATAGN will use this data to form a seven segment description of the dipole as shown in the program's output, Figure 10.

Card 6: Structure Symmetry

NP = 7 No symmetry will be used, NP = NSEGS = 7
NX = 0 No segments on axis of symmetry

Card 7: Source Excitation

IS = 4 Excite the center segment
ECM = 1. Use a 1.0 volt,
ECA = 0. 0.degree source
NFLD = 0 Only one source card needed.

This completes the data deck requirements. The next step is to submit the program for execution and obtain the computer printout. The punched data deck is shown in Figure 9, while the numerical results are shown below in Figure 10.

```
HALF WAVE HORIZONTAL DIPOLE -- EXAMPLE 1A
  2  0  0  0  0  0
    .01      0.  1  1      25.  .01
  -7  0.  .0083 -7.5  0.  10  0  n.  7.5  10.  0
    7  0
    4  1.  0.  0
```

WIRE ANTENNA MODELING PROGRAM

HALF WAVE HORIZONTAL DIPOLE -- EXAMPLE 1A

2 0 0 0 0 0 -0

FREQUENCY = 1.00000E-02
FREQUENCY INCREMENT = .0E+00
NO. FREQUENCY STEPS = 1
WAVELENGTH (METERS) = 2.99793E+01

ANTENNA IS MODELED IN FREE SPACE

I DATA GENERATOR INPUT DATA CARDS

.0 -7.50000 10.00000 0 0 7.50000 10.00000 0
-7 .00030 .0 .0 .0

NUMBER OF SEGMENTS = 7
NO. SEG. IN A SECTOR = 7
NO. SEG. ON AXIS OF ROTATION = 0

I STRUCTURE GEOMETRY (DIMENSIONS IN WAVELENGTHS)

COORDINATES OF SEG. CENTER			SEG. LENGTH	WIRE RADIUS	ORIENTATION ANGLES		CONNECTION DATA		
X	Y	Z			ALPHA	BETA	I-	I	I+
-.00000	-.21443	.33356	.07140	.00027606	.0	90.000	0	1	2
-.00000	-.14206	.33356	.07140	.00027606	.0	90.000	1	2	3
-.00000	-.07140	.33356	.07140	.00027606	.0	90.000	2	3	4
-.00000	-.00000	.33356	.07140	.00027606	.0	90.000	3	4	5
-.00000	.07140	.33356	.07140	.00027606	.0	90.000	4	5	6
-.00000	.14206	.33356	.07140	.00027606	.0	90.000	5	6	7
-.00000	.21443	.33356	.07140	.00027606	.0	90.000	6	7	0

TOTAL WIRE LENGTH = 5.00345230214E-01

ANTENNA SOURCE DISTRIBUTIONS

SEG. NO.	VOL MAG.	PHASE
4	1.00000	.0 0

I= 1
-4.934E+01 3.542E+04 -5.600E+01 -1.347E+04 -5.190E+01 9.222E+01 -4.600E+01 -1.011E+02 -4.017E+01 -4.397E+01
-3.207E+01 -1.002E+01 -2.232E+01 -2.003E+00

I= 2
-4.094E+01 -1.127E+04 -5.716E+01 1.900E+04 -5.527E+01 1.006E+04 -5.190E+01 -9.222E+01 4.600E+01 1.011E+02
-4.053E+01 -4.394E+01 -2.921E+01 -1.635E+01

I= 3
-4.504E+01 -9.044E+01 -5.595E+01 -1.006E+04 -5.640E+01 1.972E+04 -5.527E+01 -1.006E+04 -5.190E+01 -9.222E+01
-4.727E+01 -1.012E+02 -3.576E+01 -4.102E+01

I= 4
-4.145E+01 -9.641E+01 -5.250E+01 -9.239E+01 -5.527E+01 -1.006E+04 -5.640E+01 1.972E+04 -5.527E+01 -1.006E+04
-5.250E+01 -9.239E+01 -4.145E+01 -9.641E+01

I= 5
-3.570E+01 -4.102E+01 -4.727E+01 -1.012E+02 -5.190E+01 -9.222E+01 -5.527E+01 -1.006E+04 -5.640E+01 1.972E+04
-5.595E+01 -1.006E+04 -4.504E+01 -9.044E+01

I= 6
-2.921E+01 -1.635E+01 -4.053E+01 -4.394E+01 -4.600E+01 -1.011E+02 -5.190E+01 -9.222E+01 -5.527E+01 -1.006E+04
-5.716E+01 1.900E+04 -4.094E+01 -1.127E+04

I= 7
-2.232E+01 -2.003E+00 -3.207E+01 -1.002E+01 -4.017E+01 -4.397E+01 -4.600E+01 -1.011E+02 -5.190E+01 -9.222E+01
-5.600E+01 -1.347E+04 -4.934E+01 3.542E+04

1 SEGMENT EXCITATION (VOLTS/WAVELENGTH)
 SEG NUMBER REAL PART IMAGINARY PART

4 -1.399E+01 - .0E+00

1 SEG. CURRENT -					SEG. CURRENT -				
NO.	REAL	IMAGINARY	MAGNITUDE	PHASE	NO.	REAL	IMAGINARY	MAGNITUDE	PHASE
1	2.4803E-03	-1.5673E-03	2.93403710E-03	-32.289	5	9.0704E-03	-5.1156E-03	1.04134882E-02	-29.423
2	6.4306E-03	-3.8720E-03	7.50633360E-03	-31.053	6	6.4306E-03	-3.8720E-03	7.50633360E-03	-31.053
3	9.0704E-03	-5.1156E-03	1.04134882E-02	-29.423	7	2.4803E-03	-1.5673E-03	2.93403721E-03	-32.289
4	1.0002E-02	-5.0996E-03	1.12272401E-02	-27.014					
ADMIT = 1.0002E-02 -5.0996E-03					ZPED = 7.9351E+01 4.0456E+01				
	1.1227E-02	-27.014	8.9069E+01	27.014					

I	J	AR	AI	BR	BI	CR	CI
1	1	-3.5004E-03	3.5049E-03	1.0485E-02	-6.4669E-03	5.9807E-03	-5.0722E-03
2	1	-1.7703E-04	1.4781E-03	7.5893E-03	-4.0863E-03	6.6078E-03	-5.3502E-03
3	1	4.5920E-04	1.2354E-03	4.1133E-03	-1.4137E-03	8.6112E-03	-6.3510E-03
4	1	6.0482E-04	-5.2611E-03	3.9307E-11	-9.1031E-11	9.3974E-03	1.6156E-04
5	1	4.5920E-04	1.2354E-03	-4.1133E-03	1.4137E-03	8.6112E-03	-6.3510E-03
6	1	-1.7703E-04	1.4781E-03	-7.5893E-03	4.0863E-03	6.6078E-03	-5.3502E-03
7	1	-3.5004E-03	3.5049E-03	-1.0485E-02	6.4669E-03	5.9807E-03	-5.0722E-03

Example 1B. -- The results of example 1A were obtained for the dipole in free-space. It is a very simple matter to now model the antenna over an imperfectly conducting halfspace, as specified on Figure 8. Only the KSYMP variable on card 3 need be set to 2, as shown on the data card in Figure 12.

```
HALF WAVE HORIZONTAL DIPOLE -- EXAMPLE 1B
 2  0  0  0  0  0
    .01      0.  1  2  25.  0.  .01
-7  0.      -7.5 10.  0  0.  7.5  10.  0
 7  0      .0083  0.  0.  0.
 4  0      1.    0.  0
```

WIRE ANTENNA MODELING PROGRAM

HALF WAVE HORIZONTAL DIPOLE -- EXAMPLE 1B

2 0 0 0 0 0 -0
FREQUENCY = 1.0000E-02
FREQUENCY INCREMENT = .0E+00
NO. FREQUENCY STEPS = 1
WAVELENGTH (METERS) = 2.99793E+01

GROUND PLANE AT Z = 0.
DIELECTRIC CONSTANT = 2.5000E+01
CONDUCTIVITY = 1.0000E-02

I DATA GENERATOR INPUT DATA CARDS

.0 -7.50000 10.00000 0 .0 7.50000 10.00000 0
-7 .00030 .0 .0 .0

NUMBER OF SEGMENTS = 7
NO. SEG. IN A SECTOR = 7
NO. SEG. ON AXIS OF ROTATION = 0

I STRUCTURE GEOMETRY (DIMENSIONS IN WAVELENGTHS)

COORDINATES OF SEG. CENTER			SEG. LENGTH	WIRE RADIUS	ORIENTATION ANGLES		CONNECTION DATA		
X	Y	Z			ALPHA	BETA	I-	I	I+
-0.0000	-0.21443	.33356	.07140	.00027606	.0	90.000	0	1	2
-0.0000	-.14296	.33356	.07140	.00027606	.0	90.000	1	2	3
-0.0000	-.07140	.33356	.07140	.00027606	.0	90.000	2	3	4
-0.0000	-0.0000	.33356	.07140	.00027606	.0	90.000	3	4	5
-0.0000	.07140	.33356	.07140	.00027606	.0	90.000	4	5	6
-0.0000	.14296	.33356	.07140	.00027606	.0	90.000	5	6	7
-0.0000	.21443	.33356	.07140	.00027606	.0	90.000	6	7	0

TOTAL WIRE LENGTH = 5.00345238214E-01

ANTENNA SOURCE DISTRIBUTIONS

SEG. VOLTAGE
NO. MAG. PHASE
4 1.00000 .0 0

I= 1
-6.135E+01 3.541E+04 -6.977E+01 -1.340E+04 -6.403E+01 -9.330E+01 -5.063E+01 -1.011E+02 -5.065E+01 -4.270E+01
-4.104E+01 -1.417E+01 -2.073E+01 -5.232E-04

I= 2
-6.039E+01 -1.127E+04 -7.106E+01 1.979E+04 -6.077E+01 -1.006E+04 -6.403E+01 -9.330E+01 -5.063E+01 -1.011E+02
-5.111E+01 -4.270E+01 -3.712E+01 -1.435E+01

I= 3
-5.715E+01 -9.153E+01 -6.982E+01 -1.006E+04 -7.012E+01 1.972E+04 -6.077E+01 -1.006E+04 -6.403E+01 -9.330E+01
-5.923E+01 -1.012E+02 -4.504E+01 -4.007E+01

I= 4
-5.100E+01 -9.655E+01 -6.557E+01 -9.353E+01 -6.077E+01 -1.006E+04 -7.012E+01 1.972E+04 -6.077E+01 -1.006E+04
-6.557E+01 -9.353E+01 -5.100E+01 -9.655E+01

I= 5
-4.504E+01 -4.007E+01 -5.923E+01 -1.012E+02 -6.403E+01 -9.330E+01 -6.077E+01 -1.006E+04 -7.012E+01 1.972E+04
-6.982E+01 -1.006E+04 -5.715E+01 -9.153E+01

I= 6
-3.712E+01 -1.435E+01 -5.111E+01 -4.270E+01 -5.063E+01 -1.011E+02 -6.403E+01 -9.330E+01 -6.077E+01 -1.006E+04
-7.106E+01 1.979E+04 -6.039E+01 -1.127E+04

I= 7
-2.073E+01 -5.232E-04 -4.104E+01 -1.417E+01 -5.065E+01 -4.270E+01 -5.063E+01 -1.011E+02 -6.403E+01 -9.330E+01
-6.077E+01 -1.340E+04 -6.135E+01 3.541E+04

1 SEGMENT EXCITATION (VOLTS/WAVELENGTH)
 SEQ NUMBER REAL PART IMAGINARY PART

4 -1.388E+01 -.0E+00

ISEG.	CURRENT -				SEG.	CURRENT -			
NO.	REAL	IMAGINARY	MAGNITUDE	PHASE	NO.	REAL	IMAGINARY	MAGNITUDE	PHASE
1	2.1481E-03	-1.1743E-03	2.44809552E-03	-28.664	5	7.8629E-03	-3.6788E-03	8.68132877E-03	-25.078
2	5.5727E-03	-2.8536E-03	6.26084808E-03	-27.115	6	5.5727E-03	-2.8536E-03	6.26084813E-03	-27.115
3	7.8629E-03	-3.6788E-03	8.68132871E-03	-25.078	7	2.1481E-03	-1.1743E-03	2.44809555E-03	-28.664
4	8.6717E-03	-3.5162E-03	9.35749739E-03	-22.072					
ADMIT= 8.6717E-03 -3.5162E-03					ZPED= 9.9034E+01 4.0157E+01				
	9.3575E-03	-22.072	1.0687E+02	22.072					

1	1	AR	AI	BR	BI	CR	CI
1	1	-3.0077E-03	2.9457E-03	9.0655E-03	-4.8090E-03	5.1558E-03	-4.1200E-03
2	1	-1.4703E-04	1.4489E-03	6.5815E-03	-2.8853E-03	5.7197E-03	-4.3025E-03
3	1	3.8338E-04	1.3090E-03	3.5689E-03	-7.6312E-04	7.4698E-03	-4.9886E-03
4	1	5.1588E-04	-5.1638E-03	3.3652E-11	-8.4748E-11	8.1558E-03	1.6478E-03
5	1	3.9338E-04	1.3090E-03	-3.5689E-03	7.6312E-04	7.4698E-03	-4.9886E-03
6	1	-1.4703E-04	1.4489E-03	-6.5815E-03	2.8853E-03	5.7197E-03	-4.3025E-03
7	1	-3.0077E-03	2.9457E-03	-9.0655E-03	4.8090E-03	5.1558E-03	-4.1200E-03

Example 2 - Two 5/8 Wavelength Monopoles, 1/3 Wavelength Apart

As a second example, we will model two antennas, both fed, and compute the far field space radiation pattern. This example serves to illustrate the use of structure symmetry, the grounding of antenna elements and the use of the far-field option. The physical configuration is that shown in Figure 13, where we have made use of the structure's symmetry in the selection of the coordinate system's origin. The structure of the data deck is as follows:

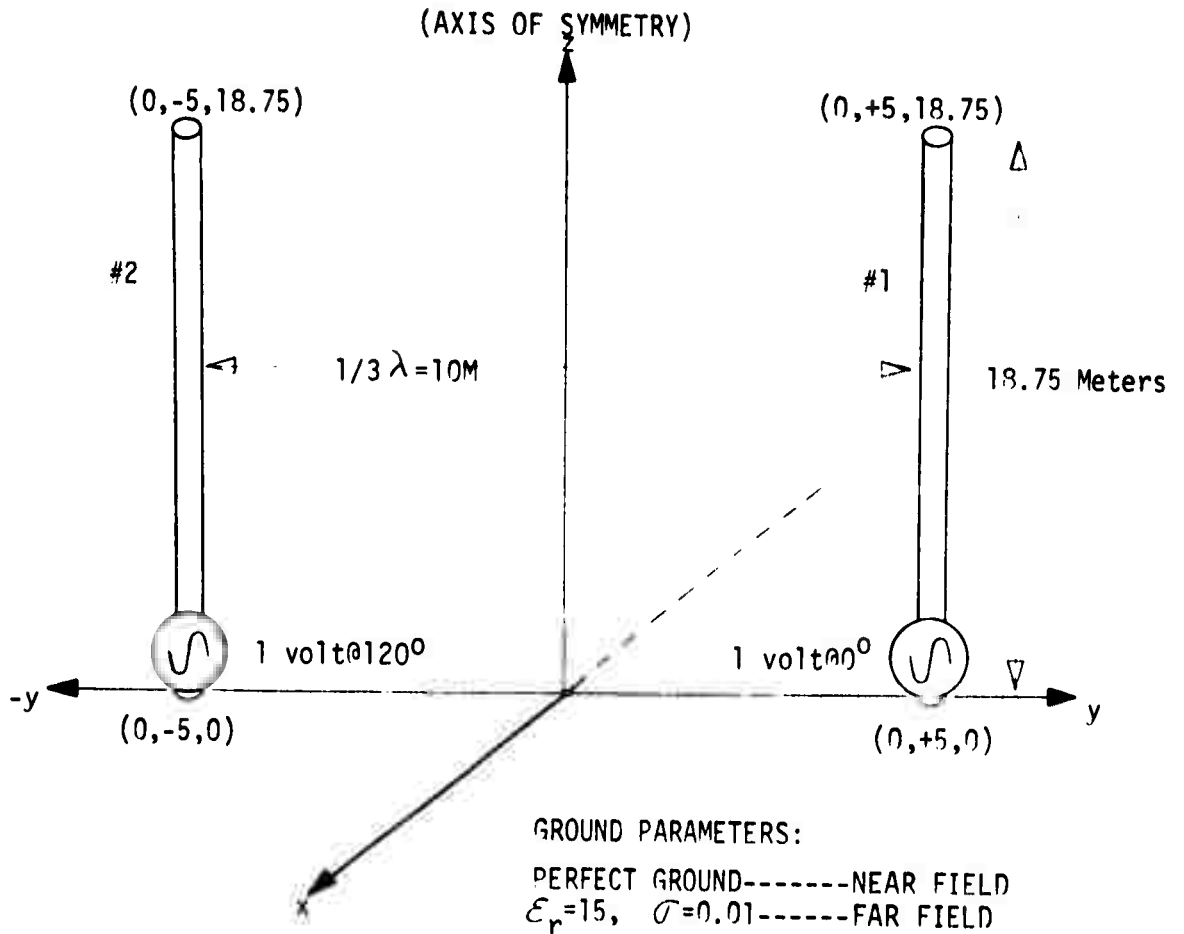
Card 1: Run Comments

Card 2: Run Options

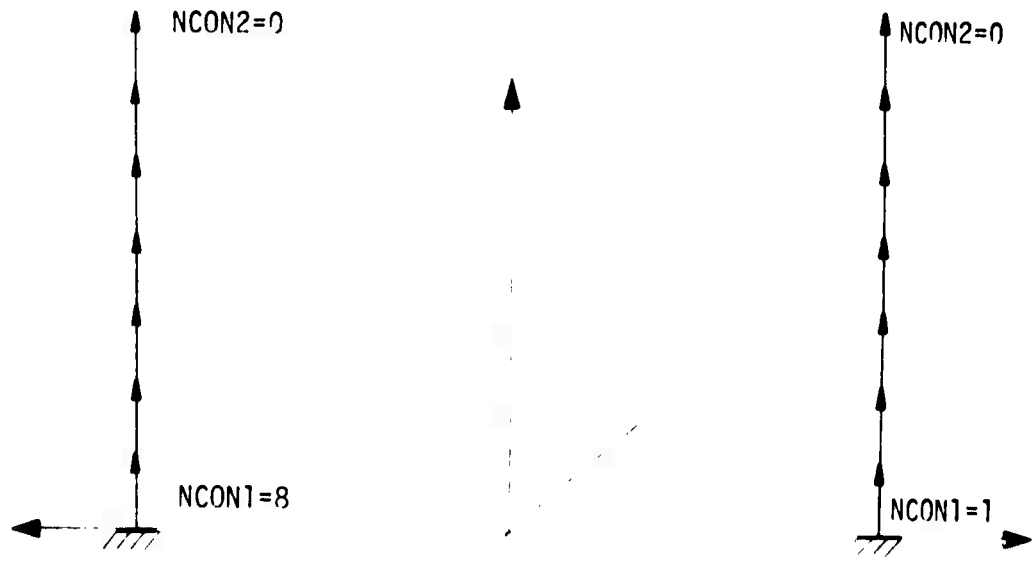
NPRINT	=	1	Nominal printout
ILOAD	=	0	No loading
IPGND	=	1	Use a perfect ground
IGSCRN	=	0	No ground screen
INEAR	=	0	No near fields
IFAR	=	1	Select far fields
NPWR	=	0	Don't normalize power

Card 3: Frequency and Grounds

GHZ	=	0.01	10 MHz input frequency
GR	=	0	No frequency steps
NFS	=	1	Only one frequency
KSYMP	=	2	Analyze over ground
EPSR	=	1.	Set EPSR = 1 as a dummy value even though
SIG	=	0.	ϵ_r and σ are not used.



a. Physical Model



b. Segmented Model

Figure 13 Two 5/8 Wavelength Monopoles--Both Fed.

Cards 4, 5, 6, and 7: Data Generator Inputs (see Figure 13)

	X1 = 0.	{	Coordinates of bottom
	Y1 = 5.		end of
	Z1 = 0.		Monopole 1
	NCON1 = 1		Segment 1 is grounded
Cards	X2 = 0.	{	Coordinates of top
4 & 5	Y2 = 5.		end of
Model	Z2 = 18.75		Monopole 1
Monopole #1	NCON2 = 0		Segment 7 is unconnected
	NSEGS = +7		7 segments used (A +7 means another set of data cards will be read)
	WIRERAD = 0.01		10 cm radius wire
	WT = 0.		No catenary
	TENS = 0.		No catenary
	TAU = 0.		Equal length segments
	X1 = 0.	{	Coordinates of bottom
	Y1 = -5.		of
	Z1 = 0.		Monopole #2
	NCON1 = 8		Segment #8 is grounded
Cards	X2 = 0.	{	Coordinates of top
6 & 7	Y2 = -5.		of
Model	Z2 = 18.75		Monopole #2
Monopole #2	NCON2 = 0		Segment #14 is unconnected
	NSEGS = -7		7 segments for monopole #2 (minus # ends input)
	WIRERAD = 0.01		10 cm radius wire
	WT = 0.		No catenary
	TENS = 0.		No catenary
	TAU = 0.		

Card 8: Structure Symmetry

NP = 7 7 segments per symmetric section - (2 sectors of symmetry)

NX = 0 No segments on axis of symmetry

Cards 9 and 10: Source Excitation

IS = 1	Excite segment #1
ECM = 1.	1 volt source
ECA = 0.	0 degree phase
NFLD = 1	Read in another source card
IS = 8	Also excite segment #8
ECM = 1.	1 volt source
ECA = +120.	+120° phase
NFLD = 0	End source input

Card 11: Far Field Input

THETR = 0. Initial θ angle

PHYR = -180. Initial ϕ angle

ETAR = 0. Polarization angle = 0°

DTHR = 5. 5 degree step in θ

DPHR = 10. 10 degree step in ϕ

NTHR = 19 19 steps in θ

NPHR = 19 19 steps in ϕ

NFLD = 0 No more farfield inputs

NEWSIG = 1 Compute far field over finite ground

SIGFF = 0.01 Conductivity of far field medium in mhos/metre

EPSFF = 15 Relative dielectric content of far field.

The completed data deck for example 2 is shown in Figure 14, and the computed results are shown in Figure 15. The far field radiation pattern is shown in Figure 16.

```

2 MONOPOLES 5/8 WAVELENGTH -- 1/3 WAVELENGTH APART -- EXAMPLE 2
  1  0  1  0  0  1  0
    01  0.  1  2  1  0.
  7  0.  5.  0.  1  0.  5.  18.75  0
-7  0.  -5.  0.  0.  0.  -5.  18.75  0
  7  0
  1  1.
  0  1  120.  0
    0.  -90.  0.  5.  10.  19  19  0  1  .01  15.

```

```

|*****
WIRE ANTENNA MODELING PROGRAM
*****

```

2 MONOPOLES 5/8 WAVELENGTH -- 1/3 WAVELENGTH APART -- EXAMPLE 2

```

  1  0  1  0  0  1  0
FREQUENCY      = 1.00000E-02
FREQUENCY INCREMENT = .0E+00
NO. FREQUENCY STEPS = 1
WAVELENGTH (METERS) = 2.99793E+01

```

A PERFECT GROUND PLANE AT Z=0.

1 DATA GENERATOR INPUT DATA CARDS

```

    .0  5.00000  .0  1  .0  5.00000  18.75000  0
  7  .01000  .0  .0  .0  .0  .0  .0
    .0 -5.00000  .0  0  .0 -5.00000  18.75000  0
-7  .01000  .0  .0  .0  .0  .0  .0

```

```

NUMBER OF SEGMENTS = 14
NO. SEG. IN A SECTOR = 7
NO. SEG. ON AXIS OF ROTATION = 0

```

1 STRUCTURE GEOMETRY (DIMENSIONS IN WAVELENGTHS)

COORDINATES OF SEG		CENTER	SEG	WIRE	ORIENTATION	ANGLES	CONNECTION DATA		
X	Y	Z	LENGTH	RADIUS	ALPHA	BETA	I	J	K
- .00000	16678	04467	08935	.00033356	90 000	0	1	1	2
- .00000	16678	13402	08935	00033356	90 000	0	1	2	3
- .00000	16678	22337	08935	00033356	90 000	0	2	3	4
- .00000	16678	31272	08935	00033356	90 000	0	3	4	5
- .00000	16678	40206	08935	00033356	90 000	0	4	5	6
- .00000	16678	49141	08935	00033356	90 000	0	5	6	7
- .00000	16678	58076	08935	00033356	90 000	0	6	7	0
- .00000	- 16678	04467	08935	00033356	90 000	0	8	8	9
- .00000	- 16678	13402	08935	00033356	90 000	0	8	9	10
- .00000	- 16678	22337	08935	00033356	90 000	0	9	10	11
- .00000	- 16678	31272	08935	00033356	90 000	0	10	11	12
- .00000	- 16678	40206	08935	00033356	90 000	0	11	12	13
- .00000	- 16678	49141	08935	00033356	90 000	0	12	13	14
- .00000	- 16678	58076	08935	00033356	90 000	0	13	14	0

TOTAL WIRE LENGTH = 1 25086309554E+00

ANTENNA SOURCE DISTRIBUTIONS

SEG. NO	MAG	PHASE
1	1.00000	.0 1
8	1.00000	120 00000 0

1 SEGMENT EXCITATION (VOLTS/WAVELENGTH)

SEG NUMBER	REAL PART	IMAGINARY PART
1	-1.119E+01	- 0E+00
8	5.596E+00	-9.693E+00

ISEG. NO	CURRENT- REAL	IMAGINARY	MAGNITUDE	PHASE	SEG. NO	CURRENT- REAL	IMAGINARY	MAGNITUDE	PHASE
1	9.4610E-04	2.4985E-03	2.67167139E-03	69.260	8	-1.9377E-03	7.7683E-04	2.08758554E-03	-158.154
2	7.4950E-04	3.2809E-04	8.18161744E-04	23.641	9	-1.0512E-04	2.4689E-04	2.68340310E-04	113.062
3	4.2282E-04	-1.8219E-03	1.87036486E-03	-76.935	10	1.6795E-03	1.2123E-03	2.07129504E-03	35.822
4	7.4905E-05	-3.3470E-03	3.34779155E-03	-88.718	11	2.9194E-03	1.8427E-03	3.45231094E-03	32.260
5	-1.8073E-04	-3.8063E-03	3.81059325E-03	-92.719	12	3.2594E-03	1.9521E-03	3.79329258E-03	30.918
6	-2.6253E-04	-3.0521E-03	3.06341972E-03	-94.916	13	2.5887E-03	1.5005E-03	2.99215487E-03	30.097
7	-1.4182E-04	-1.2557E-03	1.26363391E-03	-96.444	14	1.0588E-03	5.9774E-04	1.21587527E-03	29.447
8	-1.9377E-03	-7.7683E-04	2.08758554E-03	-158.154					

1	AR	AI	BR	BI	CR	CI
1	3.0563E-04	-4.5722E-03	-1.8465E-04	-2.0385E-03	6.4046E-04	7.0707E-03
2	3.2573E-04	3.9462E-04	-4.9147E-04	-4.0579E-03	4.2377E-04	-6.6532E-05
3	3.5366E-04	2.1421E-04	-6.3359E-04	-3.4517E-03	6.9161E-05	-2.0362E-03
4	3.7551E-04	1.2464E-04	-5.6688E-04	-1.8637E-03	-3.0060E-04	-3.4716E-03
5	3.8959E-04	1.4695E-04	-3.1692E-04	2.7688E-04	-5.6632E-04	-3.9533E-03
6	3.9719E-04	3.4351E-04	3.6550E-05	2.3956E-03	6.5972E-04	-3.3957E-03
7	5.2205E-04	1.4402E-03	4.1815E-04	4.1518E-03	-6.6387E-04	-2.6999E-03
8	4.0322E-03	2.5582E-03	1.7212E-03	9.6150E-04	5.9699E-03	-3.3350E-03
9	-2.8127E-04	5.6747E-05	3.3973E-03	1.8682E-03	1.5615E-04	1.9015E-04
10	-9.5079E-05	1.2132E-04	2.8406E-03	1.4988E-03	1.7746E-03	1.0909E-03
11	-1.1977E-04	1.4505E-04	1.4839E-03	6.9487E-04	2.9314E-03	1.6977E-03
12	-3.3312E-05	1.2455E-04	-3.1053E-04	3.2147E-04	3.2928E-03	1.8275E-03
13	-2.1048E-04	3.0969E-05	-2.0669E-03	1.2720E-03	2.7992E-03	1.4695E-03
14	-1.1484E-03	-4.7789E-04	-3.5103E-03	2.0085E-03	2.2072E-03	1.0756E-03

1 2 MONOPOLES 5/8 WAVELENGTH -- 1/3 WAVELENGTH APART -- EXAMPLE 2

DIELECTRIC CONSTANT = 1.9000E+01 AND CONDUCTIVITY = 1.0000E-02 FOR FAR FIELD CALCULATIONS

OBSERVATION ANGLES		ELECTRIC FIELD			MAGNITUDE	PHASE
THETA	PHI	R-COMPONENT	THETA-COMPONENT	PHI-COMPONENT		
0	-90.000	2.3936E-15	3.0859E-22	6.0467E-11	6.0467E-11	-116.2122
5.000	-90.000	2.0689E-15	2.0783E-02	3.6020E-11	2.0783E-02	-52.5110
10.000	-90.000	2.5039E-15	3.1645E-02	3.0825E-11	3.1645E-02	-55.5531
15.000	-90.000	1.3415E-15	3.2644E-02	2.5367E-11	3.2644E-02	-61.2927
20.000	-90.000	1.1698E-15	2.5259E-02	1.9265E-11	2.5259E-02	-74.3210
25.000	-90.000	1.6653E-16	1.5471E-02	1.2188E-11	1.5471E-02	-116.3613
30.000	-90.000	1.0146E-15	2.2725E-02	4.1005E-12	2.2725E-02	176.2860
35.000	-90.000	7.4683E-16	4.2057E-02	6.2834E-12	4.2057E-02	151.8711
40.000	-90.000	1.5464E-15	6.0345E-02	1.6897E-11	6.0345E-02	139.9473
45.000	-90.000	1.0053E-15	7.2481E-02	2.8708E-11	7.2481E-02	129.6603
50.000	-90.000	1.4433E-15	7.6358E-02	3.9249E-11	7.6358E-02	116.9156
55.000	-90.000	1.5424E-15	7.4257E-02	4.8868E-11	7.4257E-02	97.9356
60.000	-90.000	8.8991E-16	7.5224E-02	5.5771E-11	7.5224E-02	70.6651
65.000	-90.000	5.6610E-16	8.9595E-02	5.8709E-11	8.9595E-02	42.4816
70.000	-90.000	1.2947E-15	1.1407E-01	5.6703E-11	1.1407E-01	22.4997
75.000	-90.000	7.3277E-16	1.3455E-01	4.9276E-11	1.3455E-01	9.9769
80.000	-90.000	4.7073E-16	1.3517E-01	3.6620E-11	1.3517E-01	1.5081
85.000	-90.000	1.2261E-16	9.8724E-02	1.9664E-11	9.8724E-02	-5.4658
90.000	-90.000	8.3861E-27	8.4204E-12	1.1907E-21	8.4204E-12	166.4396
0	-80.000	2.3936E-15	1.0500E-11	5.9548E-11	6.0467E-11	63.7878
5.000	-80.000	3.2760E-15	2.0854E-02	3.5561E-11	2.0854E-02	-52.4842
10.000	-80.000	1.9565E-15	3.1931E-02	3.0548E-11	3.1931E-02	-55.4619
15.000	-80.000	7.4683E-16	3.3272E-02	2.5295E-11	3.3272E-02	-61.0111
20.000	-80.000	9.8787E-16	2.6262E-02	1.9430E-11	2.6262E-02	-73.2583
25.000	-80.000	7.1642E-16	1.6209E-02	1.2630E-11	1.6209E-02	-111.1011
30.000	-80.000	6.8448E-16	2.1416E-02	4.8026E-12	2.1416E-02	-179.3314
35.000	-80.000	8.6711E-16	3.9867E-02	5.2034E-12	3.9867E-02	153.7073
40.000	-80.000	1.2009E-15	5.7788E-02	1.5378E-11	5.7788E-02	141.1025
45.000	-80.000	8.0059E-16	6.9816E-02	2.6295E-11	6.9816E-02	130.6083
50.000	-80.000	1.6012E-15	7.3741E-02	3.6980E-11	7.3741E-02	117.8362
55.000	-80.000	8.8861E-16	7.1640E-02	4.6321E-11	7.1640E-02	98.8812
60.000	-80.000	1.4655E-15	7.2280E-02	5.3067E-11	7.2280E-02	71.4541
65.000	-80.000	1.1802E-15	8.6008E-02	5.6008E-11	8.6008E-02	42.8807
70.000	-80.000	8.8991E-16	1.0986E-01	5.4194E-11	1.0986E-01	22.6324
75.000	-80.000	4.9651E-16	1.2997E-01	4.7156E-11	1.2997E-01	10.0016
80.000	-80.000	2.2658E-16	1.3082E-01	3.5074E-11	1.3082E-01	1.4932
85.000	-80.000	1.7085E-16	9.5652E-02	1.8843E-11	9.5652E-02	-5.4950
90.000	-80.000	2.1867E-25	8.1613E-12	1.1408E-21	8.1613E-12	166.4035
0	-70.000	2.3936E-15	2.0681E-11	5.6820E-11	6.0467E-11	63.7878
5.000	-70.000	3.2451E-15	2.1066E-02	3.4181E-11	2.1066E-02	-52.4056
10.000	-70.000	2.4183E-15	3.2782E-02	2.9688E-11	3.2782E-02	-55.2004
15.000	-70.000	9.0366E-16	3.5136E-02	2.5021E-11	3.5136E-02	-60.2340
20.000	-70.000	1.0123E-15	2.9284E-02	1.9836E-11	2.9284E-02	-70.5355
25.000	-70.000	4.0341E-16	1.9060E-02	1.3826E-11	1.9060E-02	-98.3347
30.000	-70.000	9.3095E-17	1.8455E-02	6.8280E-12	1.8455E-02	-163.1064

1 2 MONOPOLES 5/8 WAVELENGTH -- 1 3 WAVELENGTH APART -- EXAMPLE 2

DIELECTRIC CONSTANT = 1 5000E+01 AND CONDUCTIVITY = 1 0000E-02 FOR FAR FIELD CALCULATIONS

OBSERVATION ANGLES		ELECTRIC FIELD					
THETA	PHI	R-COMPONENT	THETA COMPONENT	PHI-COMPONENT	MAGNITUDE	PHASE	
35 000	-70.000	6.7132E-16	3 3648E-02	2.4639E-12	3 3648E 02	160 5371	
40 000	-70.000	9 6946E-16	5 0326E-02	1 1104E 11	5 0326E 02	145 2187	
45 000	-70.000	1 4476E-15	6 1983E-02	2 0870E 11	6 1983E 02	133 9014	
50 000	-70.000	1 0533E-15	6 6037E 02	3 0518E 11	6 6037E 02	120 9982	
55 000	-70.000	9 8092E-16	6 3932E-02	3 9044E 11	6 3937E 02	102 1435	
60 000	-70.000	1 1271E-15	6 3554E 02	4 5321E 11	6 3554E 02	74 2274	
65 000	-70.000	1 0934E-15	7 5312E-02	4 8259E 11	7 5312E 02	44 2902	
70 000	-70.000	1 2225E-15	9 1192E 02	4 6976E 11	9 7192E 01	23 0948	
75 000	-70.000	5 2369E-16	1 1618E-01	4 1048E 11	1 1618E 01	10 0863	
80 000	-70.000	4 6848E-16	1 1773E 01	3 0616E 11	1 1773E 01	1 4422	
85 000	-70.000	5 8981E-17	8 6341E 02	1 6474E 11	8 6391E 02	5 5939	
90 000	-70.000	4 0915E-25	7 3790E 12	9 9829E 22	7 3790E 12	166 2969	
0	-60.000	2 3936E-15	3 0233E 11	5 2366E 11	6 0467E 11	63 7878	
5 000	-60.000	3 2141E-15	2 1419E 02	3 1876E 11	2 1419E 02	52 2806	
10 000	-60.000	2 0795E-15	3 4167E-02	2 8169E 11	3 4167E-02	-54 8014	
15 000	-60.000	1 0164E-15	3 8181E 02	2 4384E 11	3 8181E-02	-59 1284	
20 000	-60.000	7 8505E-16	3 4310E 02	2 0222E 11	3 4310E-02	-67 1319	
25 000	-60.000	7 3014E-16	2 5044E 02	1 5408E 11	2 5044E 02	-84 8314	
30 000	-60.000	6 2373E-16	1 7894E-02	9 7517E-12	1 7894E 02	-130 0149	
35 000	-60.000	2 1377E-16	2 5175E 02	3 3939E 12	2 5125E 02	178 4790	
40 000	-60.000	1 4433E-15	3 8843E-02	4 9542E-12	3 8883E 02	141 4751	
45 000	-60.000	1 3007E-15	4 9643E-02	1 2858E-11	4 9669E 02	155 2477	
50 000	-60.000	9 4206E-16	5 3869E-02	2 0881E 11	5 3869E-02	128 0777	
55 000	-60.000	8 2523E-16	5 1765E-02	2 8115E 11	5 1765E-02	109 5291	
60 000	-60.000	7 7728E-16	4 9552E-02	3 3625E-11	4 9552E 02	80 8418	
65 000	-60.000	4 9651E-16	5 7703E-02	3 6495E 11	5 7703E 02	47 7207	
70 000	-60.000	6 6613E-16	7 6134E-02	3 5989E 11	7 6134E-02	24 1837	
75 000	-60.000	5 7220E-16	9 3152E-02	3 1723E 11	9 3152E 02	10 2793	
80 000	-60.000	5 8039E-16	9 5811E-02	2 3795E-11	9 5811E 02	1 3286	
85 000	-60.000	2 0206E-16	7 0867E-02	1 2845E 11	7 0867E 02	-5 8117	
90 000	-60.000	4 2861E-25	6 0649E-12	7 7881E 22	6 0669E-12	166 0510	
0	-50.000	2 3936E-15	3 8867E-11	4 6320E-11	6 0467E-11	63 7878	
5 000	-50.000	2 4975E-15	2 1876E-02	2 8643E-11	2 1876E-02	-52 1173	
10 000	-50.000	2 8098E-15	3 6039E-02	2 5880E-11	3 6039E 02	-54 3088	
15 000	-50.000	1 2959E-15	4 2307E-02	2 3144E-11	4 2307E 02	-57 8821	
20 000	-50.000	1 6136E-15	4 1230E-02	2 0196E-11	4 1230E 02	-61 8513	
25 000	-50.000	5 0440E-16	3 4373E-02	1 6809E-11	3 4373E 02	-74 7121	
30 000	-50.000	8 9088E-16	2 5185E-02	1 2801E-11	2 5185E-02	-97 1350	
35 000	-50.000	7 7765E-16	2 1080E-02	8 0941E-12	2 1080E 02	-139 6731	
40 000	-50.000	1 3990E-15	2 7006E-02	2 9450E-12	2 7006E-02	-178 6291	
45 000	-50.000	7 2164E-16	3 5249E-02	3 8583E-12	3 5249E 02	159 8010	
50 000	-50.000	9 2389E-16	3 9326E-02	9 7135E-12	3 9326E 02	144 3826	
55 000	-50.000	8 9681E-16	3 7351E-02	1 5286E-11	3 7351E 02	126 8940	
60 000	-50.000	1 6739E-15	3 2342E-02	1 9768E 11	3 2342E 02	98 7211	
65 000	-50.000	4 7103E-16	3 4106E-02	2 2463E-11	3 4106E 02	58 0442	

1 2 MONOPOLES 5/8 WAVELENGTH -- 1/3 WAVELENGTH APART -- EXAMPLE 2

DIELECTRIC CONSTANT = 1.5000E+01 AND CONDUCTIVITY = 1.0000E-02 FOR FAR FIELD CALCULATIONS.

OBSERVATION ANGLES		ELECTRIC FIELD			MAGNITUDE	PHASE
THETA	PHI	R-COMPONENT	THETA-COMPONENT	PHI-COMPONENT		
70.000	-50.000	6.8943E-16	4.7025E-02	2.2806E-11	4.7025E-02	27.2373
75.000	-50.000	1.7772E-16	6.1103E-02	2.0487E-11	6.1103E-02	10.7769
80.000	-40.000	1.7145E-16	6.5209E-02	1.5552E-11	6.5209E-02	1.0504
85.000	-50.000	1.8697E-16	4.9150E-02	8.4513E-12	4.9150E-02	-6.3309
90.000	-50.000	5.5598E-25	4.2321E-11	5.1389E-22	4.2321E-12	165.4820
0	-40.000	2.3936E-15	4.6320E-11	3.8867E-11	6.0467E-11	63.7878
5.000	-40.000	3.1626E-15	2.2448E-02	2.4494E-11	2.2448E-02	-51.9253
10.000	-40.000	2.2549E-15	3.8334E-02	2.2702E-11	3.8334E-02	-53.7676
15.000	-40.000	1.6012E-15	4.7374E-02	2.1031E-11	4.7374E-02	-56.6440
20.000	-40.000	1.2947E-15	4.9815E-02	1.9306E-11	4.9815E-02	-61.0866
25.000	-40.000	1.7040E-15	4.6575E-02	1.7359E-11	4.6575E-02	-69.0248
30.000	-40.000	1.9142E-15	3.9469E-02	1.5038E-11	3.9469E-02	-79.4766
35.000	-40.000	9.9698E-16	3.1583E-02	1.2240E-11	3.1583E-02	-90.8210
40.000	-40.000	7.4683E-16	2.7154E-02	8.9511E-12	2.7154E-02	-127.6475
45.000	-40.000	1.2403E-15	2.7984E-02	5.2846E-12	2.7984E-02	-156.7538
50.000	-40.000	1.0602E-15	3.0171E-02	1.7577E-12	3.0171E-02	-177.4751
55.000	-40.000	1.3668E-16	2.9183E-02	2.9747E-12	2.9183E-02	-167.6040
60.000	-40.000	4.7752E-16	2.3228E-02	6.1398E-12	2.3228E-02	-152.5092
65.000	-40.000	3.6638E-16	1.3982E-02	8.4883E-12	1.3982E-02	-123.6922
70.000	-40.000	3.3537E-16	1.1787E-02	9.5618E-12	1.1787E-02	-52.3137
75.000	-40.000	1.2795E-16	2.0796E-02	9.1273E-12	2.0796E-02	13.5232
80.000	-40.000	4.8528E-16	2.6571E-02	7.1827E-12	2.6571E-02	-184.9
85.000	-40.000	1.3660E-16	2.1684E-02	3.9789E-12	2.1684E-02	-8.4214
90.000	-40.000	3.5223E-25	1.9095E-12	2.4333E-22	1.9095E-12	163.2208
0	-30.000	2.3936E-15	5.2366E-11	3.0233E-11	6.0467E-11	63.7878
5.000	-30.000	3.3343E-15	2.3107E-02	1.9465E-11	2.3107E-02	-51.7147
10.000	-30.000	3.1365E-15	4.0973E-02	1.8538E-11	4.0973E-02	-53.2161
15.000	-30.000	2.0488E-15	5.3198E-02	1.7786E-11	5.3198E-02	-55.5034
20.000	-30.000	1.4571E-15	5.9739E-02	1.7108E-11	5.9739E-02	-58.8407
25.000	-30.000	1.8611E-15	6.0983E-02	1.6386E-11	6.0983E-02	-63.6646
30.000	-30.000	2.3479E-15	5.7862E-02	1.5514E-11	5.7862E-02	-70.6805
35.000	-30.000	1.5414E-15	5.1989E-02	1.4394E-11	5.1989E-02	-80.9330
40.000	-30.000	1.0190E-15	4.5676E-02	1.2967E-11	4.5676E-02	-95.5104
45.000	-30.000	1.1116E-15	4.1421E-02	1.1225E-11	4.1421E-02	-114.0983
50.000	-30.000	1.8377E-15	4.0341E-02	9.2173E-12	4.0341E-02	-133.2809
55.000	-30.000	5.1179E-16	4.1018E-02	7.0609E-12	4.1018E-02	-148.9558
60.000	-30.000	7.3277E-16	4.0942E-02	4.9251E-12	4.0942E-02	-159.9453
65.000	-30.000	8.9088E-16	3.8470E-02	3.0067E-12	3.8470E-02	-167.0894
70.000	-30.000	2.8177E-16	3.3352E-02	1.4957E-12	3.3352E-02	-171.4498
75.000	-30.000	4.6476E-16	2.6198E-02	5.7367E-13	2.6198E-02	-173.8519
80.000	-30.000	2.2602E-16	1.8738E-02	3.6724E-13	1.8738E-02	-175.2032
85.000	-30.000	1.3986E-17	1.0717E-02	2.7577E-13	1.0717E-02	-177.2573
90.000	-30.000	1.7401E-25	8.3547E-13	1.8760E-23	8.3547E-13	-3.4473
0	-20.000	2.3936E-15	5.6820E-11	2.0681E-11	6.0467E-11	63.7878
5.000	-20.000	2.8231E-15	2.3822E-02	1.3823E-11	2.3822E-02	-51.4953

1 2 MONOPOLES 5/8 WAVELENGTH -- 1/3 WAVELENGTH APART -- EXAMPLE 2

DIELECTRIC CONSTANT = 1.5000E+01 AND CONDUCTIVITY = 1.0000E-02 FOR FAR FIELD CALCULATIONS

OBSERVATION ANGLES		ELECTRIC FIELD			MAGNITUDE		PHASE
THETA	PHI	R-COMPONENT	THETA-COMPONENT	PHI-COMPONENT			
10.000	-20.000	2.8475E-15	4.3864E-02	1.3335E-11	4.3864E-02		-52.6820
15.000	-20.000	1.7244E-15	5.9569E-02	1.3227E-11	5.9569E-02		-54.4999
20.000	-20.000	2.1102E-15	7.0610E-02	1.3248E-11	7.0610E-02		-57.1037
25.000	-20.000	1.7902E-15	7.6886E-02	1.3339E-11	7.6886E-02		-60.7443
30.000	-20.000	1.9270E-15	7.8678E-02	1.3432E-11	7.8678E-02		-65.8097
35.000	-20.000	1.7772E-15	7.6519E-02	1.3455E-11	7.6519E-02		-72.8762
40.000	-20.000	2.0015E-15	7.2064E-02	1.3336E-11	7.2064E-02		-82.7018
45.000	-20.000	1.1226E-15	6.7281E-02	1.3013E-11	6.7281E-02		95.9377
50.000	-20.000	7.4683E-16	6.4736E-02	1.2442E-11	6.4736E-02		-112.2253
55.000	-20.000	9.5505E-16	6.6064E-02	1.1598E-11	6.6064E-02		-129.3306
60.000	-20.000	8.9509E-16	7.0665E-02	1.0484E-11	7.0665E-02		-144.3904
65.000	-20.000	6.7760E-16	7.6081E-02	9.1237E-12	7.6081E-02		-156.1380
70.000	-20.000	1.0130E-15	7.9383E-02	7.5554E-12	7.9383E-02		-164.9302
75.000	-20.000	1.0873E-16	7.7797E-02	5.8244E-12	7.7797E-02		-171.6780
80.000	-20.000	2.4286E-16	6.8290E-02	3.106E-12	6.8240E-02		-177.3438
85.000	-20.000	1.3895E-16	4.6040E-02	2.0237E-12	4.6040E-02		-176.9568
90.000	-20.000	4.5525E-25	3.8249E-12	1.2040E-22	3.8249E-12		-10.7511
.0	-10.000	2.3936E-15	5.9549E-11	1.0500E-11	6.0467E-11		63.7878
5.000	-10.000	2.9528E-15	2.4600E-02	7.0820E-12	2.4600E-02		-51.2756
10.000	-10.000	3.4294E-15	4.6908E-02	7.1181E-12	4.6908E-02		-52.1864
15.000	-10.000	2.0015E-15	6.6254E-02	7.2754E-12	6.6254E-02		-53.6411
20.000	-10.000	2.3714E-15	8.1999E-02	7.5387E-12	8.1999E-02		-55.7551
25.000	-10.000	2.5895E-15	9.3562E-02	7.8888E-12	9.3562E-02		-58.7164
30.000	-10.000	2.5631E-15	1.0052E-01	8.2952E-12	1.0052E-01		-62.8243
35.000	-10.000	2.7465E-15	1.0282E-01	8.7179E-12	1.0282E-01		-68.5455
40.000	-10.000	2.5662E-15	1.0107E-01	9.1051E-12	1.0107E-01		-76.5738
45.000	-10.000	2.4582E-15	9.6999E-02	9.3982E-12	9.6999E-02		-87.7852
50.000	-10.000	1.8073E-15	9.3733E-02	9.5321E-12	9.3733E-02		-102.7434
55.000	-10.000	2.8021E-15	9.5089E-02	9.4414E-12	9.5089E-02		-120.4518
60.000	-10.000	1.6505E-15	1.0300E-01	9.0673E-12	1.0300E-01		-137.8700
65.000	-10.000	1.3506E-15	1.1527E-01	8.3642E-12	1.1527E-01		-152.2656
70.000	-10.000	2.0260E-15	1.2660E-01	7.3075E-12	1.2660E-01		-163.1043
75.000	-10.000	8.9207E-16	1.3039E-01	5.8976E-12	1.3039E-01		-171.2092
80.000	-10.000	5.5720E-16	1.1909E-01	4.1656E-12	1.1909E-01		-177.7025
85.000	-10.000	1.4120E-16	8.2354E-02	2.1710E-12	8.2354E-02		-176.1562
90.000	-10.000	1.0041E-24	6.8996E-12	1.3017E-22	6.8996E-12		-11.7009
.0	.0	2.3936E-15	6.0467E-11	0E+00	6.0467E-11		63.7878
5.000	.0	3.1287E-15	2.5387E-02	0E+00	2.5387E-02		-51.0629
10.000	.0	3.2880E-15	5.0005E-02	0E+00	5.0005E-02		-51.7374
15.000	.0	3.1426E-15	7.3016E-02	0E+00	7.3016E-02		-52.9186
20.000	.0	2.7104E-15	9.3466E-02	0E+00	9.3466E-02		-54.7044
25.000	.0	2.5317E-15	1.1030E-01	0E+00	1.1030E-01		-57.2587
30.000	.0	2.1557E-15	1.2247E-01	0E+00	1.2247E-01		-60.8481
35.000	.0	3.5958E-15	1.2916E-01	0E+00	1.2916E-01		-65.9000
40.000	.0	3.7221E-15	1.3029E-01	0E+00	1.3029E-01		-73.0832

1 2 MONOPOLES 5/8 WAVELENGTH -- 1/3 WAVELENGTH APART -- EXAMPLE 2

DIELECTRIC CONSTANT = 1.5000E+01 AND CONDUCTIVITY = 1.0000E-02 FOR FAR FIELD CALCULATIONS.

OBSERVATION ANGLES		ELECTRIC FIELD				MAGNITUDE	PHASE
THETA	PHI	R-COMPONENT	THETA-COMPONENT	PHI-COMPONENT			
45.000	.0	3.5804E-15	1.2715E-01	.0E+00	1.2715E-01	-83.3473	
50.000	.0	2.5177E-15	1.2324E-01	.0E+00	1.2324E-01	-97.6244	
55.000	.0	1.6910E-15	1.2420E-01	.0E+00	1.2420E-01	-115.5956	
60.000	.0	2.7822E-15	1.3454E-01	.0E+00	1.3454E-01	-134.3442	
65.000	.0	1.4043E-15	1.5287E-01	.0E+00	1.5287E-01	-150.2797	
70.000	.0	4.9651E-16	1.7171E-01	.0E+00	1.7171E-01	-162.2340	
75.000	.0	2.2861E-15	1.8065E-01	.0E+00	1.8065E-01	-171.0011	
80.000	.0	7.7743E-16	1.6772E-01	.0E+00	1.6772E-01	-177.8531	
85.000	.0	3.6638E-16	1.1714E-01	.0E+00	1.1714E-01	-175.8319	
90.000	.0	1.2941E-24	9.8491E-12	.0E+00	9.8491E-12	-12.0854	
0	10.000	2.3936E-15	5.9548E-11	1.0500E-11	6.0467E-11	63.7878	
5.000	10.000	3.1632E-15	2.6167E-02	7.4178E-12	2.6167E-02	-50.8629	
10.000	10.000	3.7364E-15	5.3055E-02	7.8142E-12	5.3055E-02	-51.3403	
15.000	10.000	2.2542E-15	7.9626E-02	8.3808E-12	7.9626E-02	-52.3175	
20.000	10.000	3.4884E-15	1.0460E-01	9.1258E-12	1.0460E-01	-53.8807	
25.000	10.000	2.7756E-15	1.2644E-01	1.0047E-11	1.2644E-01	-56.1801	
30.000	10.000	2.9790E-15	1.4350E-01	1.1124E-11	1.4350E-01	-59.4639	
35.000	10.000	5.6173E-15	1.5428E-01	1.2308E-11	1.5428E-01	-64.1373	
40.000	10.000	3.3009E-15	1.5804E-01	1.3522E-11	1.5804E-01	-70.8516	
45.000	10.000	1.7772E-15	1.5566E-01	1.4652E-11	1.5566E-01	-80.5836	
50.000	10.000	1.9860E-15	1.5097E-01	1.5557E-11	1.5097E-01	-94.4488	
55.000	10.000	3.3009E-15	1.5112E-01	1.6075E-11	1.5112E-01	-112.5425	
60.000	10.000	3.9534E-15	1.6307E-01	1.6037E-11	1.6307E-01	-132.1183	
65.000	10.000	2.2204E-15	1.8644E-01	1.5294E-11	1.8644E-01	-149.0520	
70.000	10.000	3.1402E-15	2.1183E-01	1.3740E-11	2.1183E-01	-161.7139	
75.000	10.000	1.3609E-15	2.2536E-01	1.1338E-11	2.2536E-01	-170.8807	
80.000	10.000	9.7799E-16	2.1097E-01	8.1375E-12	2.1097E-01	-177.9380	
85.000	10.000	1.9107E-16	1.4810E-01	4.2828E-12	1.4810E-01	-175.6514	
90.000	10.000	1.8515E-24	1.2469E-11	2.5758E-22	1.2469E-11	-12.2885	
0	20.000	2.3936E-15	5.6820E-11	2.0681E-11	6.0467E-11	63.7878	
5.000	20.000	3.1846E-15	2.6917E-02	1.4926E-11	2.6917E-02	-50.6800	
10.000	20.000	3.6701E-15	5.5964E-02	1.6034E-11	5.5964E-02	-50.9963	
15.000	20.000	2.0837E-15	8.5874E-02	1.7511E-11	8.5874E-02	-51.8218	
20.000	20.000	3.5108E-15	1.1502E-01	1.9393E-11	1.1502E-01	-53.2317	
25.000	20.000	3.3984E-15	1.4141E-01	2.1689E-11	1.4141E-01	-55.3844	
30.000	20.000	3.4684E-15	1.6282E-01	2.4362E-11	1.6282E-01	-58.4544	
35.000	20.000	6.2686E-15	1.7714E-01	2.7312E-11	1.7714E-01	-62.8910	
40.000	20.000	5.2779E-15	1.8304E-01	3.0356E-11	1.8304E-01	-69.3114	
45.000	20.000	2.4349E-15	1.8110E-01	3.3231E-11	1.8110E-01	-78.7023	
50.000	20.000	2.6658E-15	1.7537E-01	3.5597E-11	1.7537E-01	-92.2851	
55.000	20.000	2.3804E-15	1.7431E-01	3.7055E-11	1.7431E-01	-110.4331	
60.000	20.000	3.8721E-15	1.8705E-01	3.7197E-11	1.8705E-01	-130.5657	
65.000	20.000	2.2204E-15	2.1422E-01	3.5651E-11	2.1422E-01	-148.2822	
70.000	20.000	1.8971E-15	2.4485E-01	3.2155E-11	2.4485E-01	-161.3882	
75.000	20.000	1.7841E-15	2.6208E-01	2.6613E-11	2.6208E-01	-170.8802	

1 2 MONOPOLES 5/8 WAVELENGTH -- 1/3 WAVELENGTH APART -- EXAMPLE 2

DILECTRIC CONSTANT = 1.5000E+01 AND CONDUCTIVITY = 1.0000E-02 FOR FAR FIELD CALCULATIONS.

OBSERVATION ANGLES		ELECTRIC FIELD				
THETA	PHI	R-COMPONENT	THETA-COMPONENT	PHI-COMPONENT	MAGNITUDE	PHASE
00.000	20.000	6.8189E-16	2.4647E-01	1.9141E-11	2.4647E-01	-177.9941
05.000	20.000	3.3887E-16	1.7348E-01	1.0086E-11	1.7348E-01	175.5332
10.000	20.000	1.6898E-24	1.4623E-11	6.0672E-22	1.4623E-11	-12.4275
15.000	30.000	2.3936E-15	5.2366E-11	3.0233E-11	6.0467E-11	63.7878
20.000	30.000	3.9629E-15	2.7616E-02	2.2248E-11	2.7616E-02	-50.5174
25.000	30.000	4.3002E-15	5.8647E-02	2.4302E-11	5.8647E-02	-50.7043
30.000	30.000	3.5423E-15	9.1579E-02	2.6928E-11	9.1579E-02	-51.4175
35.000	30.000	3.0627E-15	1.2443E-01	3.0202E-11	1.2443E-01	-52.7198
40.000	30.000	3.3009E-15	1.5476E-01	3.4152E-11	1.5476E-01	-54.7390
45.000	30.000	5.1789E-15	1.7983E-01	3.8726E-11	1.7983E-01	-57.6884
50.000	30.000	5.4616E-15	1.9700E-01	4.3764E-11	1.9700E-01	-61.9751
55.000	30.000	4.1895E-15	2.0446E-01	4.8967E-11	2.0446E-01	-68.1941
60.000	30.000	7.1063E-15	2.0255E-01	5.3895E-11	2.0255E-01	-77.3454
65.000	30.000	2.6853E-15	1.9545E-01	5.7973E-11	1.9545E-01	-90.7172
70.000	30.000	3.6621E-15	1.9294E-01	6.0545E-11	1.9294E-01	108.8824
75.000	30.000	2.6738E-15	2.0570E-01	6.0924E-11	2.0570E-01	-129.4103
80.000	30.000	1.8971E-15	2.3536E-01	5.8496E-11	2.3536E-01	-147.5703
85.000	30.000	1.8971E-15	2.6975E-01	5.2825E-11	2.6975E-01	-161.0994
90.000	30.000	1.7764E-15	2.8965E-01	4.3757E-11	2.8965E-01	-170.7414
00.000	40.000	9.0632E-16	2.7305E-01	3.1489E-11	2.7305E-01	-178.0348
05.000	40.000	7.9201E-16	1.9245E-01	1.6598E-11	1.9245E-01	175.4479
10.000	40.000	2.1765E-24	1.6226E-11	9.9857E-22	1.6226E-11	-12.5237
15.000	40.000	2.3936E-15	4.6320E-11	3.8867E-11	6.0467E-11	63.7878
20.000	40.000	3.9931E-15	2.8243E-02	2.9094E-11	2.8243E-02	-50.3774
25.000	40.000	3.6557E-15	6.1031E-02	3.2222E-11	6.1031E-02	-50.4623
30.000	40.000	3.6338E-15	9.6594E-02	3.6109E-11	9.6594E-02	-51.0925
35.000	40.000	3.7813E-15	1.3260E-01	4.0871E-11	1.3260E-01	-52.3185
40.000	40.000	3.9284E-15	1.6619E-01	4.6560E-11	1.6619E-01	-54.2579
45.000	40.000	4.8698E-15	1.9418E-01	5.3109E-11	1.9418E-01	-57.1254
50.000	40.000	5.5644E-15	2.1346E-01	6.0290E-11	2.1346E-01	-61.2879
55.000	40.000	4.5722E-15	2.2140E-01	6.7674E-11	2.2140E-01	-67.3609
60.000	40.000	4.6364E-15	2.1964E-01	7.4646E-11	2.1964E-01	-76.3338
65.000	40.000	4.0288E-15	2.1118E-01	8.0400E-11	2.1118E-01	-89.5399
70.000	40.000	4.5084E-15	2.0684E-01	8.4016E-11	2.0684E-01	-107.7006
75.000	40.000	2.5121E-15	2.1898E-01	8.4550E-11	2.1898E-01	-128.5177
80.000	40.000	2.8436E-15	2.4994E-01	8.1161E-11	2.4994E-01	-147.0805
85.000	40.000	2.2644E-15	2.8667E-01	7.3264E-11	2.8667E-01	-160.8980
90.000	40.000	2.2267E-15	3.0823E-01	6.0658E-11	3.0823E-01	-170.6962
00.000	50.000	4.5776E-16	2.9085E-01	4.3634E-11	2.9085E-01	-178.0660
05.000	50.000	5.5788E-16	2.0512E-01	2.2993E-11	2.0512E-01	175.3826
10.000	50.000	2.5701E-24	1.7298E-11	1.3831E-21	1.7298E-11	-12.5981
15.000	50.000	2.3936E-15	3.8867E-11	4.6320E-11	6.0467E-11	63.7878
20.000	50.000	3.3420E-15	2.8780E-02	3.5174E-11	3.3420E-02	-50.2614
25.000	50.000	4.7080E-15	6.3058E-02	3.9388E-11	6.3058E-02	-50.2680
30.000	50.000	3.3839E-15	1.0081E-01	4.4512E-11	1.0081E-01	-50.8377

1 2 MONOPOLES 5/8 WAVELENGTH -- 1/3 WAVELENGTH APART -- EXAMPLE 2

DIELECTRIC CONSTANT = 1.5000E+01 AND CONDUCTIVITY = 1.0000E-02 FOR FAR FIELD CALCULATIONS.

OBSERVATION ANGLES		ELECTRIC FIELD			MAGNITUDE	PHASE
THETA	PHI	R-COMPONENT	THETA-COMPONENT	PHI-COMPONENT		
20.000	50.000	3.8050E-15	1.3938E-01	5.0710E-11	1.3938E-01	-52.0091
25.000	50.000	3.4542E-15	1.7554E-01	5.8045E-11	1.7554E-01	-53.8918
30.000	50.000	5.1789E-15	2.0571E-01	6.6429E-11	2.0571E-01	-56.8829
35.000	50.000	5.6741E-15	2.2644E-01	7.5564E-11	2.2644E-01	-60.7717
40.000	50.000	5.7732E-15	2.3533E-01	8.4911E-11	2.3533E-01	-66.7350
45.000	50.000	5.4672E-15	2.3247E-01	9.3674E-11	2.3247E-01	-75.5727
50.000	50.000	3.5537E-15	2.2252E-01	1.0085E-10	2.2252E-01	-88.6464
55.000	50.000	5.1263E-15	2.1641E-01	1.0528E-10	2.1641E-01	-106.7905
60.000	50.000	4.9302E-15	2.2755E-01	1.0582E-10	2.2755E-01	-127.8208
65.000	50.000	2.5895E-15	2.5887E-01	1.0143E-10	2.5887E-01	-146.6958
70.000	50.000	1.7342E-15	2.9674E-01	9.1438E-11	2.9674E-01	-160.7397
75.000	50.000	1.8784E-15	3.1910E-01	7.5614E-11	3.1910E-01	-170.6607
80.000	50.000	8.8821E-16	3.0116E-01	5.4341E-11	3.0116E-01	-178.0905
85.000	50.000	6.7422E-16	2.1240E-01	2.8618E-11	2.1240E-01	-175.3314
90.000	50.000	2.0417E-24	1.7812E-11	1.7213E-21	1.7812E-11	-12.6522
0	60.000	2.3936E-15	3.0233E-11	5.2366E-11	6.0467E-11	63.7878
5.000	60.000	4.4171E-15	2.9213E-02	4.0222E-11	2.9213E-02	-50.1704
10.000	60.000	3.9923E-15	6.4678E-02	4.5418E-11	6.4678E-02	-50.1193
15.000	60.000	3.8867E-15	1.0415E-01	5.1639E-11	1.0415E-01	-50.6460
20.000	60.000	3.2853E-15	1.4468E-01	5.9081E-11	1.4468E-01	-51.7791
25.000	60.000	5.6741E-15	1.8274E-01	6.7819E-11	1.8274E-01	-53.6217
30.000	60.000	6.0606E-15	2.1443E-01	7.7744E-11	2.1443E-01	-56.3751
35.000	60.000	3.4684E-15	2.3607E-01	8.8494E-11	2.3607E-01	-60.3930
40.000	60.000	4.8698E-15	2.4506E-01	9.9418E-11	2.4506E-01	-66.2764
45.000	60.000	4.1792E-15	2.4149E-01	1.0958E-10	2.4149E-01	-75.0108
50.000	60.000	6.2247E-15	2.3016E-01	1.1780E-10	2.3016E-01	-87.9810
55.000	60.000	4.0214E-15	2.2244E-01	1.2277E-10	2.2244E-01	-106.1037
60.000	60.000	1.7342E-15	2.3243E-01	1.2316E-10	2.3243E-01	-127.2877
65.000	60.000	2.3915E-15	2.6351E-01	1.1784E-10	2.6351E-01	-146.3986
70.000	60.000	2.8087E-15	3.0170E-01	1.0604E-10	3.0170E-01	-160.6177
75.000	60.000	4.0156E-15	3.2429E-01	8.7557E-11	3.2429E-01	-170.6334
80.000	60.000	1.5545E-15	3.0598E-01	6.2851E-11	3.0598E-01	-178.1093
85.000	60.000	5.5511E-16	2.1573E-01	3.3076E-11	2.1573E-01	-175.2920
90.000	60.000	1.3705E-24	1.8191E-11	1.9885E-21	1.8191E-11	-12.7060
0	70.000	2.3936E-15	2.0681E-11	5.6820E-11	6.0467E-11	63.7878
5.000	70.000	4.5041E-15	2.9531E-02	4.4006E-11	2.9531E-02	-50.1051
10.000	70.000	4.6242E-15	6.5859E-02	4.9981E-11	6.5859E-02	-50.0144
15.000	70.000	3.2972E-15	1.0656E-01	5.7058E-11	1.0656E-01	-50.5124
20.000	70.000	3.4542E-15	1.4847E-01	6.5455E-11	1.4847E-01	-51.6200
25.000	70.000	3.6146E-15	1.8782E-01	7.5255E-11	1.8782E-01	-53.4356
30.000	70.000	4.2422E-15	2.2050E-01	8.6326E-11	2.2050E-01	-56.1567
35.000	70.000	5.5644E-15	2.4262E-01	9.8251E-11	2.4262E-01	-60.1324
40.000	70.000	4.9851E-15	2.5153E-01	1.1030E-10	2.5153E-01	-65.9597
45.000	70.000	6.2804E-15	2.4730E-01	1.2143E-10	2.4730E-01	-74.6215
50.000	70.000	3.7880E-15	2.3488E-01	1.3034E-10	2.3488E-01	-87.5163

1 2 MONOPOLES 5/8 WAVELENGTH -- 1/3 WAVELENGTH APART -- EXAMPLE 2

DIELECTRIC CONSTANT = 1 5000E+01 AND CONDUCTIVITY = 1 0000E-02 FOR FAR FIELD CALCULATIONS.

OBSERVATION ANGLES		ELECTRIC FIELD			MAGNITUDE	PHASE
THETA	PHI	R-COMPONENT	THETA-COMPONENT	PHI-COMPONENT		
55.000	70.000	3.7560E-15	2.2580E-01	1.3559E-10	2.2580E-01	-105.6185
60.000	70.000	4.3113E-15	2.3482E-01	1.3577E-10	2.3482E-01	-126.9069
65.000	70.000	2.2204E-15	2.6542E-01	1.2966E-10	2.6542E-01	-146.1865
70.000	70.000	2.9790E-15	3.0349E-01	1.1648E-10	3.0349E-01	-160.5297
75.000	70.000	2.2542E-15	3.2599E-01	9.6042E-11	3.2599E-01	-170.6136
80.000	70.000	6.6905E-16	3.0741E-01	6.8860E-11	3.0741E-01	-178.1229
85.000	70.000	4.4674E-16	2.1669E-01	3.6218E-11	2.1669E-01	-175.2634
90.000	70.000	9.6187E-25	1.8271E-11	2.1774E-21	1.8271E-11	-12.7325
0	90.000	2.3936E-15	1.0500E-11	5.9548E-11	6.0467E-11	63.7878
5.000	80.000	4.5912E-15	2.9725E-02	4.6350E-11	2.9725E-02	-50.0658
10.000	80.000	4.3356E-15	6.6577E-02	5.2825E-11	6.6577E-02	-49.9520
15.000	80.000	3.1421E-15	1.0802E-01	6.0444E-11	1.0802E-01	-50.4334
20.000	80.000	4.4367E-15	1.5074E-01	6.9440E-11	1.5074E-01	-51.5285
25.000	80.000	5.6741E-15	1.9084E-01	7.9897E-11	1.9084E-01	-53.3265
30.000	80.000	4.0943E-15	2.2405E-01	9.1667E-11	2.2405E-01	-56.0286
35.000	80.000	5.3660E-15	2.4640E-01	1.0430E-10	2.4640E-01	-59.9795
40.000	80.000	6.3584E-15	2.5520E-01	1.1701E-10	2.5520E-01	-65.7733
45.000	80.000	2.5895E-15	2.5050E-01	1.2869E-10	2.5050E-01	-74.3916
50.000	80.000	3.1735E-15	2.3734E-01	1.3797E-10	2.3734E-01	-87.2402
55.000	80.000	1.7798E-15	2.2751E-01	1.4334E-10	2.2751E-01	-105.3279
60.000	80.000	2.2315E-15	2.3577E-01	1.4334E-10	2.3577E-01	-126.6769
65.000	80.000	2.3604E-15	2.6595E-01	1.3671E-10	2.6595E-01	-146.0572
70.000	80.000	1.6012E-15	3.0380E-01	1.2267E-10	3.0380E-01	-160.4761
75.000	80.000	1.7841E-15	3.2613E-01	1.0104E-10	3.2613E-01	-170.6816
80.000	80.000	1.3184E-16	3.0742E-01	7.2398E-11	3.0742E-01	-178.1313
85.000	80.000	6.7132E-16	2.1664E-01	3.8057E-11	2.1664E-01	-175.2460
90.000	80.000	5.0965E-25	1.8267E-11	2.2872E-21	1.8267E-11	-12.7633
0	90.000	2.3936E-15	3.0859E-22	6.0467E-11	6.0467E-11	-116.2122
5.000	90.000	4.0943E-15	2.9790E-02	4.7144E-11	2.9790E-02	-50.0526
10.000	90.000	5.1182E-15	6.6817E-02	5.3791E-11	6.6817E-02	-49.9312
15.000	90.000	4.2855E-15	1.0850E-01	6.1595E-11	1.0850E-01	-50.4073
20.000	90.000	4.0762E-15	1.5150E-01	7.0795E-11	1.5150E-01	-51.4926
25.000	90.000	4.2422E-15	1.9183E-01	8.1474E-11	1.9183E-01	-53.2905
30.000	90.000	5.7732E-15	2.2522E-01	9.3479E-11	2.2522E-01	-55.9863
35.000	90.000	6.6798E-15	2.4764E-01	1.0635E-10	2.4764E-01	-59.9291
40.000	90.000	7.7557E-15	2.5638E-01	1.1827E-10	2.5638E-01	-65.7118
45.000	90.000	3.6621E-15	2.5152E-01	1.3113E-10	2.5152E-01	-74.3155
50.000	90.000	2.6820E-15	2.3811E-01	1.4053E-10	2.3811E-01	-87.1485
55.000	90.000	5.0096E-15	2.2799E-01	1.4593E-10	2.2799E-01	-105.2309
60.000	90.000	1.6012E-15	2.3600E-01	1.4585E-10	2.3600E-01	-126.5998
65.000	90.000	2.3604E-15	2.6602E-01	1.3904E-10	2.6602E-01	-146.0137
70.000	90.000	3.2024E-15	3.0377E-01	1.2471E-10	3.0377E-01	-160.4581
75.000	90.000	1.8453E-15	3.2603E-01	1.0269E-10	3.2603E-01	-170.5975
80.000	90.000	8.9207E-16	3.0727E-01	7.3556E-11	3.0727E-01	-178.1341
85.000	90.000	6.6844E-16	2.1651E-01	3.8659E-11	2.1651E-01	-175.2402

1 2 MONOPOLES 3/8 WAVELENGTH -- 1/3 WAVELENGTH APART -- EXAMPLE 2

DIELECTRIC CONSTANT = 1.5000E+01 AND CONDUCTIVITY = 1.0000E-02 FOR FAR FIELD CALCULATIONS.

OBSERVATION ANGLES		ELECTRIC FIELD					
THETA	PHI	R-COMPONENT	THETA-COMPONENT	PHI-COMPONENT	MAGNITUDE	PHASE	
90.000	90.000	3.5213E-27	1.0255E-11	2.3230E-21	1.0255E-11	-12.7637	

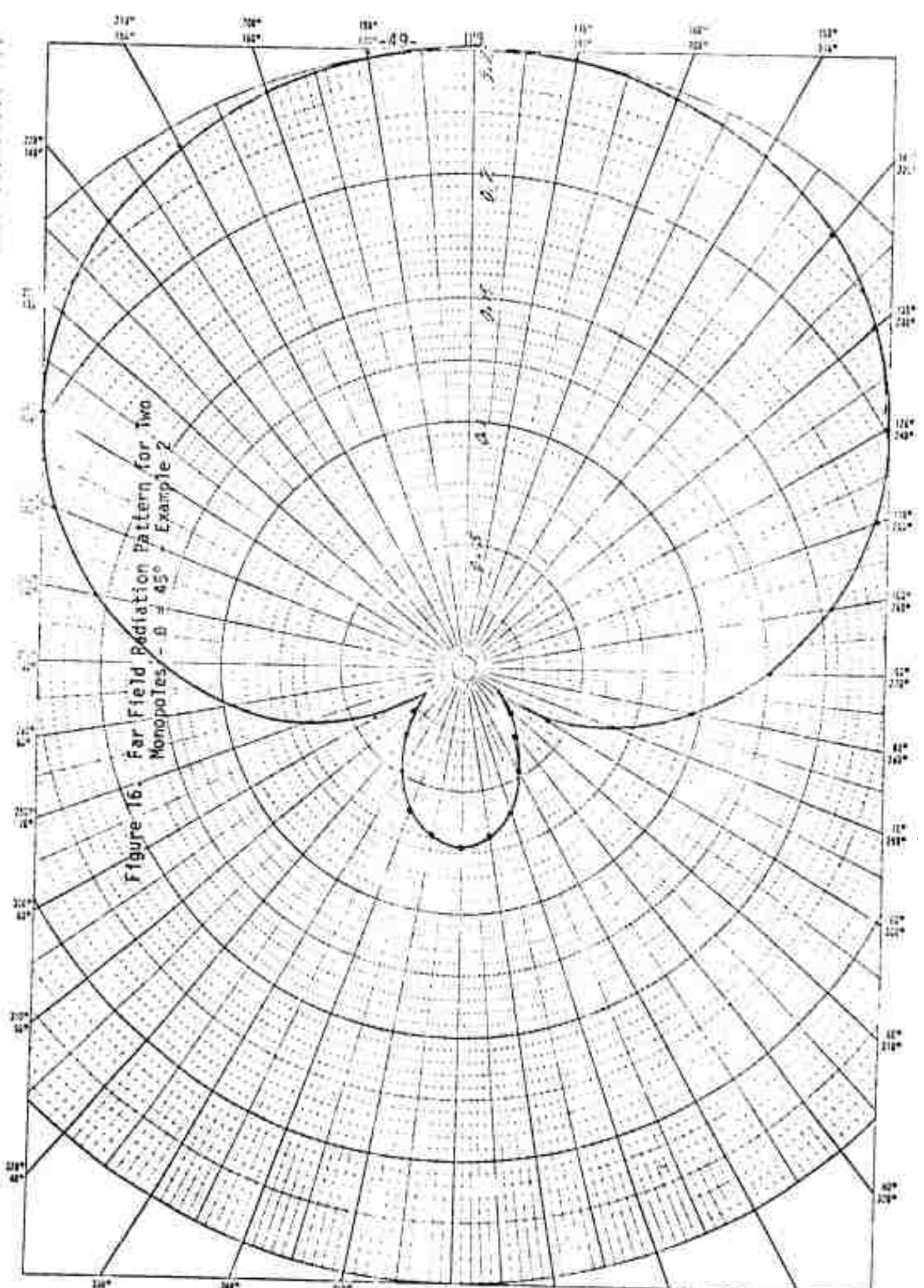


Figure 16. Far Field Radiation Pattern for Two Monopoles - $b = 45^\circ$ Example 2

Example 3 - Top Loaded Monopole - As a third example we will model the antenna shown in Figure 17. This antenna, used for LORAN-C, has a high degree of rotational symmetry. It also has an element on the axis of rotational symmetry and a multiple junction of elements. We will model this antenna over a radial wire ground screen, and compute the electric field from one of the top-loaded radials to ground for a normalized input power of 1 watt.

The input data deck for this antenna shown in Figure 18 is similar to those of the previous examples, however, the following points should be noted.

Card 2: We select a ground screen, near field and normalization of input power.

Card 3: The input frequency of this antenna is 100 kHz, and the ground media parameters are needed.

Card 4 — 30 are used to describe the antenna. Note that the NCON1 value of all the top-load radials is given a value of -1000 to designate that it is a multiple junction of wires and that the NCON2 value of each top load radial is 0 since they are unconnected. A catenary model is used to specify the physical droop of the wires, and thus a wire weight of 1.586 pounds/metre and a wire tension of 1000 pounds is specified. The last element specified is the tower which is on the axis of symmetry and note that its base is grounded, so that NCON1 of the tower element is 97 and NCON2 of the tower is -1000 since it connects with the top-load radials.

Card 31: Specifies the structure symmetry. Note that 24 sectors of rotational symmetry actually exist on this structure, but since the program limits the user to 12 sectors, two top load radials are included in each of the 12 allowed sectors. Four segments are used for each radial load, and four segments are used for the tower, so NP = 8 and NX = 4. The total number of segments used equals 100, which is the limit of this version of WAMP.

Card 32: Specifies that the base segment, Number 97, of the monopole is driven, and card 33 specifies that 180 radials of #8 AWG wire are used for the ground screen.

Card 34: is the last card, and it specifies that we want to compute the near field from the ground up to the vicinity of one of the top-load radials. 41 points will be evaluated along the path between the two points, and the electric field tangent to the path will be integrated to give the potential between the two end points.

The complete data deck is shown in Figure 18 and the computed results are shown in Figure 19.

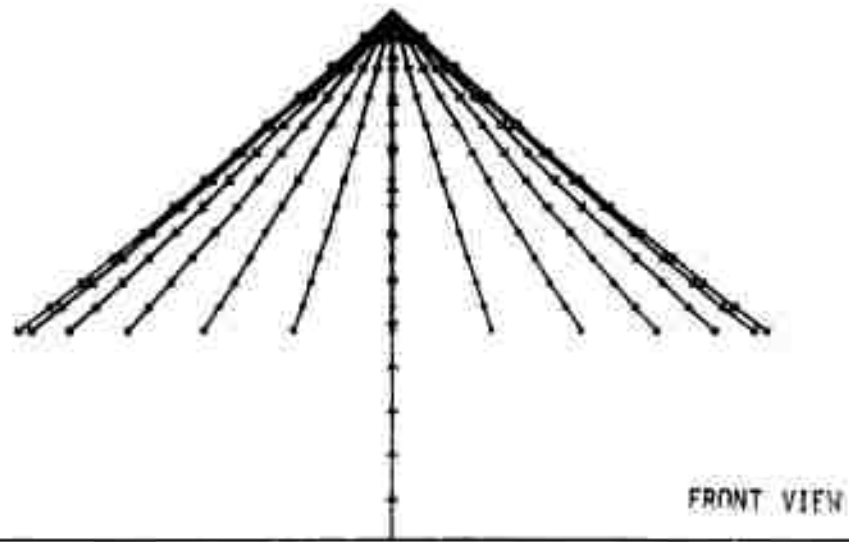
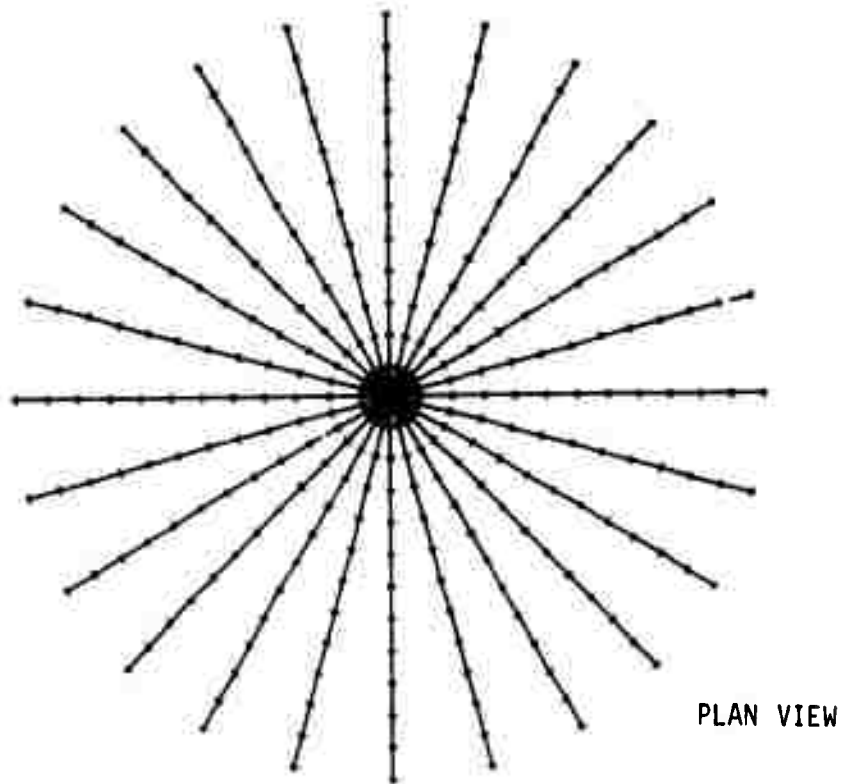


Figure 17. Computer Drawn Model of U. S. Coast Guard Top Loaded Monopole
(Note: More segments are shown than actually used in example.)

USCG TOP LOADED MONOPOLE ANTENNA -- EXAMPLE 3

	1	0	0	1	1	0	1					
	.00010	0.	1	2	15.		.01					
	.0	.0	192.94	-1000	140.21		.0	75.90		0		
4	.00914	1.586	1000.0		0.							
	.0	.0	192.94	-1000	135.43587	36.29681	75.90			0		
4	.00914	1.586	1000.0		0.							
	.0	.0	192.94	-1000	121.42163	70.10009	75.90			0		
4	.00914	1.536	1000.0		0.							
	.0	.0	192.94	-1000	99.14901	99.14923	75.90			0		
4	.00914	1.586	1000.0		0.							
	.0	.0	192.94	-1000	70.07992	121.42179	75.90			0		
4	.00914	1.586	1000.0		0.							
	.0	.0	192.94	-1000	36.29651	135.43599	75.90			0		
4	.00914	1.586	1000.0		0.							
	.0	.0	192.94	-1000	-0.00031	140.21	75.90			0		
4	.00914	1.586	1000.0		0.							
	.0	.0	192.94	-1000	-36.29711	135.43579	75.90			0		
4	.00914	1.586	1000.0		0.							
	.0	.0	192.94	-1000	-70.10036	121.42148	75.90			0		
4	.00914	1.586	1000.0		0.							
	.0	.0	192.94	-1000	-99.14945	99.14879	75.90			0		
4	.00914	1.586	1000.0		0.							
	.0	.0	192.94	-1000	-121.42194	70.079955	75.90			0		
4	.00914	1.586	1000.0		0.							
	.0	.0	192.94	-1000	-135.43603	36.29621	75.90			0		
4	.00914	1.586	1000.0		0.							
	.0	.0	192.94	-1000	-140.21	-0.00062	75.90			0		
4	.00914	1.586	1000.0		0.							
	.0	.0	192.94	-1000	-135.43571	-36.29741	75.90			0		
4	.00914	1.586	1000.0		0.							
	.0	.0	192.94	-1000	-121.42132	-70.10053	75.90			0		
4	.00914	1.586	1000.0		0.							
	.0	.0	192.94	-1000	-99.14657	-99.14967	75.90			0		
4	.00914	1.586	1000.0		0.							
	.0	.0	192.94	-1000	-70.07392	-121.4221	75.90			0		
4	.00914	1.586	1000.0		0.							
	.0	.0	192.94	-1000	-36.29591	-135.43611	75.90			0		
4	.00914	1.586	1000.0		0.							
	.0	.0	192.94	-1000	.00093	-140.21	75.90			0		
4	.00914	1.586	1000.0		0.							
	.0	.0	192.94	-1000	36.29771	-135.43563	75.90			0		
4	.00914	1.586	1000.0		0.							
	.0	.0	192.94	-1000	70.10089	-121.42117	75.90			0		
4	.00914	1.586	1000.0		0.							
	.0	.0	192.94	-1000	99.14969	-99.14835	75.90			0		
4	.00914	1.586	1000.0		0.							
	.0	.0	192.94	-1000	121.42225	-70.07990	75.90			0		
4	.00914	1.586	1000.0		0.							
	.0	.0	192.94	-1000	135.43619	-36.29551	75.90			0		
4	.00914	1.586	1000.0		0.							
-4	.3048	0.	0.	97	0.	0.	192.94-1000					
8												
97	1.	0.	0									
180	.00163											
	140.3	0.	0.	140.3	0.	75.85	40	0				

Figure 18. Data Deck for Example 3.

WIRE ANTENNA MODELING PROGRAM

USCG TOP LOADED MONOPOLE ANTENNA -- EXAMPLE 3

1 0 0 1 1 0 1

FREQUENCY = 1 0000E+04
FREQUENCY INCREMENT = 0E+00
NO. FREQUENCY STEPS = 1
WAVELENGTH (METERS) = 2 99793E+03

GROUND PLANE AT Z = 0
DIELECTRIC CONSTANT = 1 5000E+01
CONDUCTIVITY = 1 0000E-02

1 DATA GENERATOR INPUT DATA CARDS

	.0	.0	192.94000-1000	140	21000	0	75.90000	0	
4	.00914	1.506001000	00000	0	0	0	0	0	
	.0	.0	192.94000-1000	135	43587	36	29681	75.90000	
4	.00914	1.506001000	00000	0	0	0	0	0	
	.0	.0	192.94000-1000	121	42163	70	10009	75.90000	
4	.00914	1.506001000	00000	0	0	0	0	0	
	.0	.0	192.94000-1000	99	14901	99	14923	75.90000	
4	.00914	1.506001000	00000	0	0	0	0	0	
	.0	.0	192.94000-1000	70	07998	121	42179	75.90000	
4	.00914	1.506001000	00000	0	0	0	0	0	
	.0	.0	192.94000-1000	36	29651	135	43595	75.90000	
4	.00914	1.506001000	00000	0	0	0	0	0	
	.0	.0	192.94000-1000	-00031	140	21000	75	90000	
4	.00914	1.506001000	00000	0	0	0	0	0	
	.0	.0	192.94000-1000	-36	29711	135	43579	75	90000
4	.00914	1.506001000	00000	0	0	0	0	0	
	.0	.0	192.94000-1000	-70	10036	121	42148	75	90000
4	.00914	1.506001000	00000	0	0	0	0	0	
	.0	.0	192.94000-1000	-99	14945	99	14979	75	90000
4	.00914	1.506001000	00000	0	0	0	0	0	
	.0	.0	192.94000-1000	-121	42194	70	07996	75	90000
4	.00914	1.506001000	00000	0	0	0	0	0	
	.0	.0	192.94000-1000	-135	43603	36	29621	75	90000
4	.00914	1.506001000	00000	0	0	0	0	0	
	.0	.0	192.94000-1000	-140	21000	00062	75	90000	
4	.00914	1.506001000	00000	0	0	0	0	0	
	.0	.0	192.94000-1000	-135	43571	36	29741	75	90000
4	.00914	1.506001000	00000	0	0	0	0	0	
	.0	.0	192.94000-1000	-121	42132	70	10063	75	90000
4	.00914	1.506001000	00000	0	0	0	0	0	
	.0	.0	192.94000-1000	-99	14857	99	14967	75	90000
4	.00914	1.506001000	00000	0	0	0	0	0	
	.0	.0	192.94000-1000	-70	07992	121	42210	75	90000
4	.00914	1.506001000	00000	0	0	0	0	0	
	.0	.0	192.94000-1000	-36	29591	135	43611	75	90000
4	.00914	1.506001000	00000	0	0	0	0	0	
	.0	.0	192.94000-1000	00093	140	21000	75	90000	
4	.00914	1.506001000	00000	0	0	0	0	0	
	.0	.0	192.94000-1000	36	29771	135	43563	75	90000
4	.00914	1.506001000	00000	0	0	0	0	0	
	.0	.0	192.94000-1000	70	10089	121	42117	75	90000
4	.00914	1.506001000	00000	0	0	0	0	0	
	.0	.0	192.94000-1000	99	14989	99	14835	75	90000
4	.00914	1.506001000	00000	0	0	0	0	0	
	.0	.0	192.94000-1000	121	42225	70	07990	75	90000
4	.00914	1.506001000	00000	0	0	0	0	0	
	.0	.0	192.94000-1000	135	43619	36	29561	75	90000
4	.00914	1.506001000	00000	0	0	0	0	0	
	.0	.0	0 97	0	0	0	0	192.94000-1000	
-4	.30480	0	0	0	0	0	0	0	

NUMBER OF SEGMENTS = 100
NO. SEG. IN A SECTOR = 0
NO. SEG. ON AXIS OF ROTATION = 4

I STRUCTURE GEOMETRY (DIMENSIONS IN WAVELENGTHS)

COORDINATES OF SEQ. CENTER			SEQ.	WIRE LENGTH	WIRE RADIUS	ORIENTATION ANGLES		CONNECTION DATA		
X	Y	Z				ALPHA	BETA	I-	I	I+
.00505	.0	.05099	.01507	.0000305	-42.556	0-1000	1	2		
.01794	.0	.04050	.01544	.0000305	-40.779	0	1	2	3	
.02923	.0	.03002	.01502	.0000305	-38.902	0	2	3	4	
.04092	.0	.02971	.01462	.0000305	-36.920	0	3	4	0	
.00505	.00151	.05099	.01507	.0000305	-42.555	15.003-1000	5	6		
.01694	.00494	.04050	.01544	.0000305	-40.778	15.003	5	6	7	
.02824	.00757	.03002	.01502	.0000305	-38.901	15.003	6	7	0	
.03953	.01059	.02971	.01463	.0000305	-36.918	15.003	7	0	0	
.00506	.00292	.05099	.01507	.0000305	-42.557	29.999-1000	9	10		
.01519	.00877	.04050	.01544	.0000305	-40.780	29.999	9	10	11	
.02531	.01461	.03002	.01502	.0000305	-38.903	29.999	10	11	12	
.03544	.02046	.02971	.01462	.0000305	-36.921	29.999	11	12	0	
.00413	.00413	.05099	.01507	.0000305	-42.555	45.000-1000	13	14		
.01240	.01240	.04050	.01544	.0000305	-40.778	45.000	13	14	15	
.02067	.02067	.03002	.01502	.0000305	-38.900	45.000	14	15	16	
.02894	.02894	.02971	.01463	.0000305	-36.918	45.000	15	16	0	
.00292	.00506	.05099	.01507	.0000305	-42.559	60.000-1000	17	18		
.00877	.01519	.04050	.01544	.0000305	-40.782	60.000	17	18	19	
.01461	.02531	.03002	.01502	.0000305	-38.905	60.000	18	19	20	
.02045	.03544	.02971	.01462	.0000305	-36.923	60.000	19	20	0	
.00151	.00565	.05099	.01507	.0000305	-42.555	74.997-1000	21	22		
.00494	.01694	.04050	.01544	.0000305	-40.778	74.997	21	22	23	
.00757	.02824	.03002	.01502	.0000305	-38.901	74.997	22	23	24	
.01059	.03953	.02971	.01463	.0000305	-36.918	74.997	23	24	0	
-.00000	.00505	.05099	.01507	.0000305	-42.556	90.000-1000	25	26		
-.00000	.01794	.04050	.01544	.0000305	-40.779	90.000	25	26	27	
-.00000	.02923	.03002	.01502	.0000305	-38.902	90.000	26	27	28	
-.00000	.04092	.02971	.01462	.0000305	-36.920	90.000	27	28	0	
-.00151	.00565	.05099	.01507	.0000305	-42.555	105.003-1000	29	30		
-.00494	.01694	.04050	.01544	.0000305	-40.778	105.003	29	30	31	
-.00757	.02824	.03002	.01502	.0000305	-38.901	105.003	30	31	32	
-.01059	.03953	.02971	.01463	.0000305	-36.918	105.003	31	32	0	
-.00292	.00506	.05099	.01507	.0000305	-42.557	119.999-1000	33	34		
-.00877	.01519	.04050	.01544	.0000305	-40.780	119.999	33	34	35	
-.01461	.02531	.03002	.01502	.0000305	-38.903	119.999	34	35	36	
-.02046	.03544	.02971	.01462	.0000305	-36.921	119.999	35	36	0	
-.00413	.00413	.05099	.01507	.0000305	-42.555	135.000-1000	37	38		
-.01240	.01240	.04050	.01544	.0000305	-40.778	135.000	37	38	39	
-.02067	.02067	.03002	.01502	.0000305	-38.900	135.000	38	39	40	
-.02894	.02894	.02971	.01463	.0000305	-36.918	135.000	39	40	0	
-.00506	.00292	.05099	.01507	.0000305	-42.559	150.000-1000	41	42		
-.01519	.00877	.04050	.01544	.0000305	-40.782	150.000	41	42	43	
-.02531	.01461	.03002	.01502	.0000305	-38.905	150.000	42	43	44	
-.03544	.02045	.02971	.01462	.0000305	-36.923	150.000	43	44	0	
-.00505	.00151	.05099	.01507	.0000305	-42.555	164.990-1000	45	46		

1 STRUCTURE GEOMETRY (DIMENSIONS IN WAVELENGTHS)

COORDINATES OF SEG. CENTER			SEG. LENGTH	WIRE RADIUS	ORIENTATION ANGLES		CONNECTION DATA		
X	Y	Z			ALPHA	BETA	I-	I	I+
-01694	00454	04858	01544	00000305	-40.778	164.998	45	46	47
-02824	00757	03882	01502	00000305	-38.901	164.998	46	47	48
-03953	01059	02971	01463	00000305	-36.918	164.998	47	48	0
-00585	-00000	05899	01587	00000305	-42.556	-180.000	-1000	49	50
-01754	-00000	04858	01544	00000305	40.779	-180.000	49	50	51
-02923	-00000	03882	01502	00000305	-38.902	-180.000	50	51	52
-04092	-00000	02971	01462	00000305	36.920	180.000	51	52	0
-00565	-00151	05899	01587	00000305	-42.555	164.997	-1000	53	54
-01694	-00454	04858	01544	00000305	40.778	164.997	53	54	55
-02824	-00757	03882	01502	00000305	38.901	164.997	54	55	56
-03953	-01059	02971	01463	00000305	36.918	164.997	55	56	0
-00506	-00292	05899	01587	00000305	42.557	150.001	-1000	57	58
-01519	-00877	04858	01544	00000305	40.780	150.001	57	58	59
-02531	-01461	03882	01502	00000305	38.903	150.001	58	59	60
-03544	-02446	02971	01462	00000305	36.921	150.001	59	60	0
-00413	-00413	05899	01587	00000305	42.555	135.000	1000	61	62
-01240	-01240	04858	01544	00000305	40.778	135.000	61	62	63
-02067	-02067	03882	01502	00000305	38.900	135.000	62	63	64
-02894	-02894	02971	01463	00000305	36.918	135.000	63	64	0
-00292	-00506	05899	01587	00000305	-42.559	119.992	-1000	65	66
-00877	-01519	04858	01544	00000305	40.782	119.992	65	66	67
-01461	-02531	03882	01502	00000305	38.905	119.992	66	67	68
-02045	-03544	02971	01462	00000305	36.923	119.992	67	68	0
-00151	-00565	05899	01587	00000305	-42.555	105.002	-1000	69	70
-00454	-01694	04858	01544	00000305	40.778	105.002	69	70	71
-00757	-02824	03882	01502	00000305	38.901	105.002	70	71	72
-01059	-03953	02971	01463	00000305	-36.918	-105.002	71	72	0
00000	-00585	05899	01587	00000305	-42.556	-90.000	-1000	73	74
00000	-01754	04858	01544	00000305	40.779	-90.000	73	74	75
00000	-02923	03882	01502	00000305	-38.902	-90.000	74	75	76
00000	-04092	02971	01462	00000305	36.920	-90.000	75	76	0
00151	-00565	05899	01587	00000305	-42.555	-74.997	-1000	77	78
00454	-01694	04858	01544	00000305	40.778	-74.997	77	78	79
00757	-02824	03882	01502	00000305	-38.901	-74.997	78	79	80
01059	-03953	02971	01463	00000305	-36.918	-74.997	79	80	0
00292	-00506	05899	01587	00000305	42.557	60.001	1000	81	82
00877	-01519	04858	01544	00000305	40.780	60.001	81	82	83
01461	-02531	03882	01502	00000305	38.903	60.001	82	83	84
02046	-03544	02971	01462	00000305	36.921	60.001	83	84	0
00413	-00413	05899	01587	00000305	-42.555	-45.000	-1000	85	86
01240	-01240	04858	01544	00000305	40.778	-45.000	85	86	87
02067	-02067	03882	01502	00000305	-38.900	-45.000	86	87	88
02894	-02894	02971	01463	00000305	-36.918	-45.000	87	88	0
00506	-00292	05899	01587	00000305	-42.559	-29.992	-1000	89	90
01519	-00877	04858	01544	00000305	40.782	-29.992	89	90	91

1 STRUCTURE GEOMETRY (DIMENSIONS IN WAVELENGTHS)

COORDINATES OF SEG. CENTER			SEG. LENGTH	WIRE RADIUS	ORIENTATION ANGLES		CONNECTION DATA		
X	Y	Z			ALPHA	BETA	I-	I	I+
.02531	-.01461	.03002	.01502	.00000305	-30.905	-29.902	90	91	92
.03544	-.02045	.02971	.01462	.00000305	-36.923	-29.902	91	92	0
.00565	-.00151	.05099	.01507	.00000305	-42.555	-15.002	1000	93	94
.01694	-.00454	.04050	.01544	.00000305	-40.778	-15.002	93	94	95
.02024	-.00757	.03002	.01502	.00000305	-30.901	-15.002	94	95	96
.03053	-.01059	.02971	.01463	.00000305	-36.918	-15.002	95	96	0
-.00000	.0	.00004	.01609	.00010167	90.000	.0	97	97	98
-.00000	.0	.02413	.01609	.00010167	90.000	.0	97	98	99
-.00000	.0	.04022	.01609	.00010167	90.000	.0	98	99	100
-.00000	.0	.05631	.01609	.00010167	90.000	.0	99	100	1000

TOTAL WIRE LENGTH = 1.52746229115E+00

ANTENNA SOURCE DISTRIBUTIONS

SEG. NO.	MAG.	VOLTAGE PHASE
97	1.00000	.0 0

A RADIAL GROUND SCREEN OF 100 RADIALS WITH A WIRE RADIUS OF 5.4371E-07 WAVELENGTH WAS USED.

1 SEGMENT EXCITATION (VOLTS/WAVELENGTH)

SEG NUMBER	REAL PART	IMAGINARY PART
97	-6.215E+01	-.0E+00

ISEG NO.	CURRENT -			
	REAL	IMAGINARY	MAGNITUDE	PHASE
1	7.6227E-04	3.5172E-03	3.59002600E-03	77.772
2	6.3462E-04	2.9566E-03	3.02391799E-03	77.885
3	4.4515E-04	2.0931E-03	2.13909524E-03	77.993
4	1.7951E-04	0.5059E-04	0.69322332E-04	78.003
5	7.6233E-04	3.5174E-03	3.59009304E-03	77.772
6	6.3467E-04	2.9560E-03	3.02412074E-03	77.885
7	4.4518E-04	2.0932E-03	2.14003639E-03	77.993
8	1.7953E-04	0.5064E-04	0.69302102E-04	78.003
9	7.6227E-04	3.5172E-03	3.59002600E-03	77.772
10	6.3462E-04	2.9566E-03	3.02391799E-03	77.885
11	4.4515E-04	2.0931E-03	2.13909524E-03	77.993
12	1.7951E-04	0.5059E-04	0.69322332E-04	78.003
13	7.6233E-04	3.5174E-03	3.59009304E-03	77.772
14	6.3467E-04	2.9560E-03	3.02412074E-03	77.885
15	4.4518E-04	2.0932E-03	2.14003639E-03	77.993
16	1.7953E-04	0.5064E-04	0.69302102E-04	78.003
17	7.6227E-04	3.5172E-03	3.59002600E-03	77.772
18	6.3462E-04	2.9566E-03	3.02391799E-03	77.885
19	4.4515E-04	2.0931E-03	2.13909524E-03	77.993
20	1.7951E-04	0.5059E-04	0.69322332E-04	78.003
21	7.6233E-04	3.5174E-03	3.59009304E-03	77.772
22	6.3467E-04	2.9560E-03	3.02412074E-03	77.885
23	4.4518E-04	2.0932E-03	2.14003639E-03	77.993
24	1.7953E-04	0.5064E-04	0.69302102E-04	78.003
25	7.6227E-04	3.5172E-03	3.59002600E-03	77.772
26	6.3462E-04	2.9566E-03	3.02391799E-03	77.885
27	4.4515E-04	2.0931E-03	2.13909524E-03	77.993
28	1.7951E-04	0.5059E-04	0.69322332E-04	78.003
29	7.6233E-04	3.5174E-03	3.59009304E-03	77.772
30	6.3467E-04	2.9560E-03	3.02412074E-03	77.885
31	4.4518E-04	2.0932E-03	2.14003639E-03	77.993
32	1.7953E-04	0.5064E-04	0.69302102E-04	78.003
33	7.6227E-04	3.5172E-03	3.59002600E-03	77.772
34	6.3462E-04	2.9566E-03	3.02391799E-03	77.885
35	4.4515E-04	2.0931E-03	2.13909524E-03	77.993
36	1.7951E-04	0.5059E-04	0.69322332E-04	78.003
37	7.6233E-04	3.5174E-03	3.59009304E-03	77.772
38	6.3467E-04	2.9560E-03	3.02412074E-03	77.885
39	4.4518E-04	2.0932E-03	2.14003639E-03	77.993
40	1.7953E-04	0.5064E-04	0.69302102E-04	78.003
41	7.6227E-04	3.5172E-03	3.59002600E-03	77.772
42	6.3462E-04	2.9566E-03	3.02391799E-03	77.885
43	4.4515E-04	2.0931E-03	2.13909524E-03	77.993
44	1.7951E-04	0.5059E-04	0.69322332E-04	78.003
45	7.6233E-04	3.5174E-03	3.59009304E-03	77.772

SEG NO.	CURRENT -			
	REAL	IMAGINARY	MAGNITUDE	PHASE
51	4.4515E-04	2.0931E-03	2.13909524E-03	77.993
52	1.7951E-04	0.5059E-04	0.69322332E-04	78.003
53	7.6233E-04	3.5174E-03	3.59009304E-03	77.772
54	6.3467E-04	2.9560E-03	3.02412074E-03	77.885
55	4.4518E-04	2.0932E-03	2.14003639E-03	77.993
56	1.7953E-04	0.5064E-04	0.69302102E-04	78.003
57	7.6227E-04	3.5172E-03	3.59002600E-03	77.772
58	6.3462E-04	2.9566E-03	3.02391799E-03	77.885
59	4.4515E-04	2.0931E-03	2.13909524E-03	77.993
60	1.7951E-04	0.5059E-04	0.69322332E-04	78.003
61	7.6233E-04	3.5174E-03	3.59009304E-03	77.772
62	6.3467E-04	2.9560E-03	3.02412074E-03	77.885
63	4.4518E-04	2.0932E-03	2.14003639E-03	77.993
64	1.7953E-04	0.5064E-04	0.69302102E-04	78.003
65	7.6227E-04	3.5172E-03	3.59002600E-03	77.772
66	6.3462E-04	2.9566E-03	3.02391799E-03	77.885
67	4.4515E-04	2.0931E-03	2.13909524E-03	77.993
68	1.7951E-04	0.5059E-04	0.69322332E-04	78.003
69	7.6233E-04	3.5174E-03	3.59009304E-03	77.772
70	6.3467E-04	2.9560E-03	3.02412074E-03	77.885
71	4.4518E-04	2.0932E-03	2.14003639E-03	77.993
72	1.7953E-04	0.5064E-04	0.69302102E-04	78.003
73	7.6227E-04	3.5172E-03	3.59002600E-03	77.772
74	6.3462E-04	2.9566E-03	3.02391799E-03	77.885
75	4.4515E-04	2.0931E-03	2.13909524E-03	77.993
76	1.7951E-04	0.5059E-04	0.69322332E-04	78.003
77	7.6233E-04	3.5174E-03	3.59009304E-03	77.772
78	6.3467E-04	2.9560E-03	3.02412074E-03	77.885
79	4.4518E-04	2.0932E-03	2.14003639E-03	77.993
80	1.7953E-04	0.5064E-04	0.69302102E-04	78.003
81	7.6227E-04	3.5172E-03	3.59002600E-03	77.772
82	6.3462E-04	2.9566E-03	3.02391799E-03	77.885
83	4.4515E-04	2.0931E-03	2.13909524E-03	77.993
84	1.7951E-04	0.5059E-04	0.69322332E-04	78.003
85	7.6233E-04	3.5174E-03	3.59009304E-03	77.772
86	6.3467E-04	2.9560E-03	3.02412074E-03	77.885
87	4.4518E-04	2.0932E-03	2.14003639E-03	77.993
88	1.7953E-04	0.5064E-04	0.69302102E-04	78.003
89	7.6227E-04	3.5172E-03	3.59002600E-03	77.772
90	6.3462E-04	2.9566E-03	3.02391799E-03	77.885
91	4.4515E-04	2.0931E-03	2.13909524E-03	77.993
92	1.7951E-04	0.5059E-04	0.69322332E-04	78.003
93	7.6233E-04	3.5174E-03	3.59009304E-03	77.772
94	6.3467E-04	2.9560E-03	3.02412074E-03	77.885
95	4.4518E-04	2.0932E-03	2.14003639E-03	77.993

ISEG NO.	CURRENT -			
	REAL	IMAGINARY	MAGNITUDE	PHASE
46	6.3467E-04	2.9560E-03	3.02412074E-03	77.885
47	4.4518E-04	2.0932E-03	2.14003639E-03	77.993
48	1.7953E-04	0.5064E-04	0.69302102E-04	78.003
49	7.6227E-04	3.5172E-03	3.59002600E-03	77.772
50	6.3462E-04	2.9566E-03	3.02391799E-03	77.885
51	4.4515E-04	2.0931E-03	2.13909524E-03	77.993

SEG NO.	CURRENT -			
	REAL	IMAGINARY	MAGNITUDE	PHASE
96	1.7953E-04	0.5064E-04	0.69302102E-04	78.003
97	1.0351E-02	0.4925E-02	0.68047318E-02	77.806
98	1.0364E-02	0.4704E-02	0.67496522E-02	77.779
99	1.0393E-02	0.4032E-02	0.68025469E-02	77.767
100	1.0422E-02	0.4954E-02	0.69280175E-02	77.785

ADMIT* 1.0351E-02 0.4925E-02 ZPED* 2.4310E+00 -1.1250E+01
0.6805E-02 77.806 1.1510E+01 -77.806

	AR	AI	BR	BI	CR	CI
1	5.2200E-04	4.6061E-04	-1.2070E-03	-5.5573E-03	2.3939E-04	3.0566E-03
2	-6.3994E-03	-3.1312E-02	-1.6459E-03	-7.3947E-03	7.0340E-03	3.4269E-02
3	-8.7955E-03	-4.3652E-02	-2.4253E-03	-1.1227E-02	9.2407E-03	4.5745E-02
4	-1.4947E-02	-7.3319E-02	-3.5609E-03	-1.6015E-02	1.5126E-02	7.4169E-02
5	5.2101E-04	4.5601E-04	-1.2079E-03	-5.5575E-03	2.4052E-04	3.0614E-03
6	-6.3994E-03	-3.1312E-02	-1.6460E-03	-7.3952E-03	7.0340E-03	3.4269E-02
7	-8.7955E-03	-4.3651E-02	-2.4254E-03	-1.1227E-02	9.2406E-03	4.5744E-02
8	-1.4940E-02	-7.3321E-02	-3.5611E-03	-1.6016E-02	1.5127E-02	7.4172E-02
9	5.2270E-04	4.5978E-04	-1.2078E-03	-5.5574E-03	2.3957E-04	3.0574E-03
10	-6.3997E-03	-3.1313E-02	-1.6459E-03	-7.3949E-03	7.0343E-03	3.4270E-02
11	-8.7960E-03	-4.3654E-02	-2.4253E-03	-1.1227E-02	9.2411E-03	4.5747E-02
12	-1.4940E-02	-7.3322E-02	-3.5610E-03	-1.6016E-02	1.5127E-02	7.4172E-02
13	5.2100E-04	4.5640E-04	-1.2079E-03	-5.5575E-03	2.4043E-04	3.0610E-03
14	-6.3992E-03	-3.1311E-02	-1.6460E-03	-7.3951E-03	7.0339E-03	3.4269E-02
15	-8.7953E-03	-4.3650E-02	-2.4254E-03	-1.1227E-02	9.2404E-03	4.5743E-02
16	-1.4947E-02	-7.3320E-02	-3.5611E-03	-1.6016E-02	1.5127E-02	7.4170E-02
17	5.2240E-04	4.5934E-04	-1.2079E-03	-5.5576E-03	2.3907E-04	3.0560E-03
18	-6.4003E-03	-3.1310E-02	-1.6460E-03	-7.3952E-03	7.0349E-03	3.4273E-02
19	-8.7967E-03	-4.3657E-02	-2.4254E-03	-1.1227E-02	9.2410E-03	4.5750E-02
20	-1.4940E-02	-7.3320E-02	-3.5612E-03	-1.6016E-02	1.5120E-02	7.4170E-02
21	5.2101E-04	4.5601E-04	-1.2079E-03	-5.5575E-03	2.4052E-04	3.0614E-03
22	-6.3994E-03	-3.1312E-02	-1.6460E-03	-7.3952E-03	7.0340E-03	3.4269E-02
23	-8.7955E-03	-4.3651E-02	-2.4254E-03	-1.1227E-02	9.2406E-03	4.5744E-02
24	-1.4940E-02	-7.3321E-02	-3.5611E-03	-1.6016E-02	1.5127E-02	7.4172E-02
25	5.2200E-04	4.6061E-04	-1.2078E-03	-5.5573E-03	2.3939E-04	3.0566E-03
26	-6.3994E-03	-3.1312E-02	-1.6459E-03	-7.3947E-03	7.0340E-03	3.4269E-02
27	-8.7955E-03	-4.3652E-02	-2.4253E-03	-1.1227E-02	9.2407E-03	4.5745E-02
28	-1.4947E-02	-7.3319E-02	-3.5609E-03	-1.6015E-02	1.5126E-02	7.4169E-02
29	5.2101E-04	4.5601E-04	-1.2079E-03	-5.5575E-03	2.4052E-04	3.0614E-03
30	-6.3994E-03	-3.1312E-02	-1.6460E-03	-7.3952E-03	7.0340E-03	3.4269E-02
31	-8.7955E-03	-4.3651E-02	-2.4254E-03	-1.1227E-02	9.2406E-03	4.5744E-02
32	-1.4940E-02	-7.3321E-02	-3.5611E-03	-1.6016E-02	1.5127E-02	7.4172E-02
33	5.2270E-04	4.5978E-04	-1.2078E-03	-5.5574E-03	2.3957E-04	3.0574E-03
34	-6.3997E-03	-3.1313E-02	-1.6459E-03	-7.3949E-03	7.0343E-03	3.4270E-02
35	-8.7960E-03	-4.3654E-02	-2.4253E-03	-1.1227E-02	9.2411E-03	4.5747E-02
36	-1.4940E-02	-7.3322E-02	-3.5610E-03	-1.6016E-02	1.5127E-02	7.4172E-02
37	5.2100E-04	4.5640E-04	-1.2079E-03	-5.5575E-03	2.4043E-04	3.0610E-03
38	-6.3992E-03	-3.1311E-02	-1.6460E-03	-7.3951E-03	7.0339E-03	3.4269E-02
39	-8.7953E-03	-4.3650E-02	-2.4254E-03	-1.1227E-02	9.2404E-03	4.5743E-02
40	-1.4947E-02	-7.3320E-02	-3.5611E-03	-1.6016E-02	1.5127E-02	7.4170E-02
41	5.2240E-04	4.5934E-04	-1.2079E-03	-5.5576E-03	2.3907E-04	3.0560E-03
42	-6.4003E-03	-3.1310E-02	-1.6460E-03	-7.3952E-03	7.0349E-03	3.4273E-02
43	-8.7967E-03	-4.3657E-02	-2.4254E-03	-1.1227E-02	9.2410E-03	4.5750E-02
44	-1.4940E-02	-7.3320E-02	-3.5612E-03	-1.6016E-02	1.5120E-02	7.4170E-02
45	5.2101E-04	4.5601E-04	-1.2079E-03	-5.5575E-03	2.4052E-04	3.0614E-03
46	-6.3994E-03	-3.1312E-02	-1.6460E-03	-7.3952E-03	7.0340E-03	3.4269E-02
47	-8.7955E-03	-4.3651E-02	-2.4254E-03	-1.1227E-02	9.2406E-03	4.5744E-02
48	-1.4940E-02	-7.3321E-02	-3.5611E-03	-1.6016E-02	1.5127E-02	7.4172E-02
49	5.2200E-04	4.6061E-04	-1.2078E-03	-5.5573E-03	2.3939E-04	3.0566E-03
50	-6.3994E-03	-3.1312E-02	-1.6459E-03	-7.3947E-03	7.0340E-03	3.4269E-02
51	-8.7955E-03	-4.3652E-02	-2.4253E-03	-1.1227E-02	9.2407E-03	4.5745E-02
52	-1.4947E-02	-7.3319E-02	-3.5609E-03	-1.6015E-02	1.5126E-02	7.4169E-02
53	5.2101E-04	4.5601E-04	-1.2079E-03	-5.5575E-03	2.4052E-04	3.0614E-03
54	-6.3994E-03	-3.1312E-02	-1.6460E-03	-7.3952E-03	7.0340E-03	3.4269E-02
55	-8.7955E-03	-4.3651E-02	-2.4254E-03	-1.1227E-02	9.2406E-03	4.5744E-02
56	-1.4940E-02	-7.3321E-02	-3.5611E-03	-1.6016E-02	1.5127E-02	7.4172E-02
57	5.2270E-04	4.5978E-04	-1.2078E-03	-5.5574E-03	2.3957E-04	3.0574E-03
58	-6.3997E-03	-3.1313E-02	-1.6459E-03	-7.3949E-03	7.0343E-03	3.4270E-02
59	-8.7960E-03	-4.3654E-02	-2.4253E-03	-1.1227E-02	9.2411E-03	4.5747E-02
60	-1.4940E-02	-7.3322E-02	-3.5610E-03	-1.6016E-02	1.5127E-02	7.4172E-02
61	5.2100E-04	4.5640E-04	-1.2079E-03	-5.5575E-03	2.4043E-04	3.0610E-03
62	-6.3992E-03	-3.1311E-02	-1.6460E-03	-7.3951E-03	7.0339E-03	3.4269E-02
63	-8.7953E-03	-4.3650E-02	-2.4254E-03	-1.1227E-02	9.2404E-03	4.5743E-02

64	-1.4947E-02	-7.3320E-02	-3.5611E-03	-1.6816E-02	1.5127E-02	7.4170E-02
65	5.2241E-04	4.5838E-04	-1.2879E-03	-5.5576E-03	2.3886E-04	3.0588E-03
66	-6.4003E-03	-3.1316E-02	-1.6460E-03	-7.3952E-03	7.0349E-03	3.4273E-02
67	-8.7887E-03	-4.3657E-02	-2.4254E-03	-1.1227E-02	9.2418E-03	4.5750E-02
68	-1.4949E-02	-7.3327E-02	-3.5612E-03	-1.6816E-02	1.5128E-02	7.4178E-02
69	5.2181E-04	4.5601E-04	-1.2879E-03	-5.5575E-03	2.4052E-04	3.0614E-03
70	-6.3994E-03	-3.1312E-02	-1.6460E-03	-7.3952E-03	7.0340E-03	3.4269E-02
71	-8.7895E-03	-4.3651E-02	-2.4254E-03	-1.1227E-02	9.2406E-03	4.5744E-02
72	-1.4948E-02	-7.3321E-02	-3.5611E-03	-1.6816E-02	1.5127E-02	7.4172E-02
73	5.2288E-04	4.6061E-04	-1.2878E-03	-5.5573E-03	2.3939E-04	3.0566E-03
74	-6.3994E-03	-3.1312E-02	-1.6459E-03	-7.3947E-03	7.0340E-03	3.4268E-02
75	-8.7896E-03	-4.3652E-02	-2.4253E-03	-1.1227E-02	9.2407E-03	4.5745E-02
76	-1.4947E-02	-7.3318E-02	-3.5609E-03	-1.6815E-02	1.5126E-02	7.4168E-02
77	5.2181E-04	4.5601E-04	-1.2878E-03	-5.5575E-03	2.4052E-04	3.0614E-03
78	-6.3994E-03	-3.1312E-02	-1.6460E-03	-7.3952E-03	7.0340E-03	3.4269E-02
79	-8.7895E-03	-4.3651E-02	-2.4254E-03	-1.1227E-02	9.2406E-03	4.5744E-02
80	-1.4948E-02	-7.3321E-02	-3.5611E-03	-1.6816E-02	1.5127E-02	7.4172E-02
81	5.2270E-04	4.5978E-04	-1.2878E-03	-5.5574E-03	2.3957E-04	3.0574E-03
82	-6.3997E-03	-3.1313E-02	-1.6459E-03	-7.3949E-03	7.0343E-03	3.4270E-02
83	-8.7888E-03	-4.3654E-02	-2.4253E-03	-1.1227E-02	9.2411E-03	4.5747E-02
84	-1.4948E-02	-7.3322E-02	-3.5610E-03	-1.6816E-02	1.5127E-02	7.4172E-02
85	5.2189E-04	4.5840E-04	-1.2879E-03	-5.5575E-03	2.4043E-04	3.0610E-03
86	-6.3992E-03	-3.1311E-02	-1.6460E-03	-7.3951E-03	7.0339E-03	3.4268E-02
87	-8.7893E-03	-4.3650E-02	-2.4254E-03	-1.1227E-02	9.2404E-03	4.5743E-02
88	-1.4947E-02	-7.3320E-02	-3.5611E-03	-1.6816E-02	1.5127E-02	7.4170E-02
89	5.2241E-04	4.5840E-04	-1.2879E-03	-5.5576E-03	2.3886E-04	3.0588E-03
90	-6.4003E-03	-3.1316E-02	-1.6460E-03	-7.3952E-03	7.0349E-03	3.4273E-02
91	-8.7887E-03	-4.3657E-02	-2.4254E-03	-1.1227E-02	9.2418E-03	4.5750E-02
92	-1.4949E-02	-7.3327E-02	-3.5612E-03	-1.6816E-02	1.5128E-02	7.4178E-02
93	5.2181E-04	4.5601E-04	-1.2879E-03	-5.5575E-03	2.4052E-04	3.0614E-03
94	-6.3994E-03	-3.1312E-02	-1.6460E-03	-7.3952E-03	7.0340E-03	3.4269E-02
95	-8.7895E-03	-4.3651E-02	-2.4254E-03	-1.1227E-02	9.2406E-03	4.5744E-02
96	-1.4948E-02	-7.3321E-02	-3.5611E-03	-1.6816E-02	1.5127E-02	7.4172E-02
97	1.9952E-02	7.1130E-02	6.0744E-05	-6.9786E-04	-1.2007E-03	1.3795E-02
98	2.0033E-02	1.0326E-01	2.0595E-04	-4.6118E-04	-1.6696E-03	-1.8473E-02
99	1.8382E-02	9.2104E-02	2.9034E-04	8.4128E-04	1.3485E-06	-7.2728E-03
100	2.9727E-03	1.9711E-02	-4.9131E-04	-2.0914E-03	1.5449E-02	6.5243E-02

.....
VOLTAGE TO DRIVE ANTENNA AT 1-WATT = 10.43954
ALL FIELDS ARE NORMALIZED TO THIS INPUT POWER
.....

1 E-FIELD FROM 140.3000 .0 .0 METERS TO 140.3000 .0 75.0500 METERS

POSITION ON PATH (METERS)	E-TANGENT (V/M)	TOTAL E-FIELD (V/M)
.0E+00	5.17051E-01	5.20094E-01
1.09625E+00	5.20804E-01	5.23061E-01
3.79250E+00	5.24430E-01	5.26110E-01
5.60875E+00	5.27986E-01	5.29327E-01
7.50500E+00	5.31594E-01	5.32740E-01
9.48125E+00	5.35146E-01	5.36400E-01
1.13775E+01	5.38810E-01	5.40344E-01
1.32737E+01	5.42576E-01	5.44606E-01
1.51700E+01	5.46474E-01	5.49214E-01
1.70662E+01	5.50520E-01	5.54200E-01
1.89625E+01	5.54765E-01	5.59591E-01
2.08587E+01	5.59209E-01	5.65418E-01
2.27550E+01	5.63884E-01	5.71710E-01
2.46512E+01	5.68817E-01	5.78500E-01
2.65475E+01	5.74035E-01	5.85824E-01
2.84437E+01	5.79568E-01	5.93721E-01
3.03400E+01	5.85449E-01	6.02234E-01
3.22362E+01	5.91715E-01	6.11413E-01
3.41325E+01	5.98411E-01	6.21318E-01
3.60287E+01	6.05586E-01	6.32014E-01
3.79250E+01	6.13302E-01	6.43583E-01
3.98212E+01	6.21634E-01	6.56124E-01
4.17175E+01	6.30674E-01	6.69753E-01
4.36137E+01	6.40538E-01	6.84621E-01
4.55100E+01	6.51377E-01	7.00915E-01
4.74062E+01	6.63388E-01	7.18876E-01
4.93025E+01	6.76632E-01	7.38820E-01
5.11987E+01	6.92068E-01	7.61171E-01
5.30950E+01	7.09592E-01	7.86505E-01
5.49912E+01	7.30105E-01	8.15627E-01
5.68875E+01	7.54607E-01	8.49663E-01
5.87837E+01	7.84644E-01	8.90360E-01
6.06800E+01	8.22537E-01	9.40301E-01
6.25762E+01	8.72026E-01	1.00364E+00
6.44725E+01	9.39431E-01	1.08740E+00
6.63687E+01	1.03620E+00	1.20430E+00
6.82650E+01	1.18535E+00	1.38004E+00
7.01612E+01	1.44027E+00	1.67458E+00
7.20575E+01	1.95912E+00	2.26672E+00
7.39537E+01	3.50338E+00	4.02825E+00
7.58500E+01	3.44369E+01	6.31509E+01

THE INTEGRAL OF THE E-FLO TANGENT TO THE PATH IS 9.10214E+01 VOLTS. 41 POINTS USED TO EVALUATE INTEGRAL.

VI. MODELING GUIDELINES AND SPECIAL CALCULATIONS

Experience with the WAMP antenna modeling program has proven the old adage that "Garbage in equals garbage out" is applicable. If care in setting up the numerical model is not taken, completely erroneously results may be realized. In this section, I will try to cover a few important modeling rules which can help to achieve good numerical results. It is always important to carefully question numerical results for reasonability, and if possible, compare with experimental data.

Segmentation - Often the key to a good numerical model lies in the segmentation of the physical structure. Miller, et al. (1971) established some fundamental segmentation guidelines for a variety of structures. Typically, at least six segments per wavelength must be used for reasonable accuracy. At the other extreme, one must not over segment a structure such that "pancake" segments are formed --- this violates the thin-wire approximations.

Multiple junctions and segment length discontinuities, particularly in regions near sources, can lead to troubles. A good rule of thumb is to make all segments at a multiple wire junction of equal length. (Often times this is difficult to achieve when a limited number of segments are available due to computer core-size limitations and execution times.)

The data generator in the WAMP code has a provision for modeling elements with variable length segments. The factor TAU, read in by the DATAGN allows for an exponential increase or decrease in segment length on an element. For an element of N segments, the segment lengths vary by the relation (30):

$$L_i = L_0(1 + \tau)^{(i-1)} \quad (30)$$

If we know the initial length we need, L_0 , the total length of the element, L , and the number of segments available to model the element, N , Table I of Appendix B can be used to find the proper value of TAU.

Miller and Deadrick (1973) have studied in greater detail the consequences of segment length discontinuities and the possible remedies available to achieve good numerical impedance data. These techniques, i.e. accurate near field integrations, however, are expensive in terms of computer times, and should be avoided whenever possible.

Near Field Anomalies - the near electric field subroutine in the WAMP code has been thoroughly checked and found to give good, consistent, numerical results. If one evaluates the near field at the surface of the wire segments, however, large field perturbations may be found near segment ends. These field perturbations are due to segment

current discontinuities at segment junctions; a result of the current interpolation method used WAMP. Care should be exercised in interpreting near field results in these regions, particularly if you are trying to evaluate the voltage across a gap region used to model an insulator.

Horizontal Elements Near the Interface - The reflection coefficient approach used to model structures in the vicinity of an imperfectly conducting halfspace has been found (Miller, et al. (1972)) to give stable numerical results for horizontal structures whose height above the interface is greater than 0.1 wavelength. Below this height, the results may become invalid to the point of producing negative input impedances. Long horizontal elements of several wavelengths near the interface can exhibit growing currents. One approximation which has proven to "cure" some of the above limitations has been to fix the angle of incidence of the reflected wave to $\pi/2$. This may be accomplished by setting the variable CTH = 1.0 in the subroutine CMSETUP.

Radial Wire Ground Screen Model - The model used to simulate a radial wire ground screen is relatively simple in form in this program. (A radially varying screen impedance is modeled in parallel with the normal ground plane wave impedance.) The results with this model are quite good, except for vertical structures located at the center of the radial system. In this case the screen appears to the program as a perfect ground, and other techniques, i.e., application of the compensation theorem, (Maley, et al, 1963) must be employed.

Special Calculations - The WAMP code may be used to compute many antenna parameters which often times are difficult to experimentally measure. Listed below are some of the procedures required to compute some of these special quantities.

Bandwidth-Efficiency - The bandwidth efficiency product is of interest to the designers of pulsed Loran systems. Equation (31) defines this parameter

$$\eta_{BW} = \frac{2R_r}{\frac{dx}{df} + \left| \frac{x}{f} \right|} \quad (31)$$

where R_r is the antenna radiation resistance and X and $\frac{dx}{df}$ are the reactance and rate of change of reactance at a frequency f . A calculation of this quantity is easily performed by determining the input resistance and reactance for the antenna over a perfect ground (no losses in ground), and then computing the input reactance for two different frequencies to compute $\Delta X/\Delta f$.

Insulator Modeling - Often times it is necessary to access the potential across support insulators to estimate breakdown problems. Several ways have been found to model insulators. One can impedance load a segment with a very high value of resistance, solve for the segment current and compute an IR drop across the segment. Another technique is to model the insulator as a physical gap and use the near field routine to integrate the fields across the gap. This technique has been found to give somewhat high results, while if you take the E-field at the center of the gap and multiply it by the gap length, you can get a lower bound estimate. These techniques allow an estimation of the insulator voltage drops. One can either reference the drops to a 1 volt input, or normalize the antenna power to 1 watt. See Miller and Deadrick (1973) for more details on insulator modeling.

Corona Discharge Assessment - By using the near-field subroutine to follow a path along the surface of the antenna wire, one may examine the potential for corona discharge. The coordinates of the field evaluation path should be displaced a wire radius away, and sufficient points should be evaluated to resolve the segment end near field discontinuities mentioned above. Again, the fields may be referenced to a 1 volt input source or normalized to a 1 watt input.

Catenary Model - A special feature of the DATAGN subroutine in this program is the inclusion of a catenary model to account for the drop in long length wires used in some of the large LORAN antennas. The catenary curve is modeled by a series of straight line segments which approximate the catenary curve of the form shown in Figure 20.

Number of Segments Limitation

This version of WAMP is designed to operate specifically on the CDC 3300 hardware configuration for the U. S. Coast Guard headquarters in Washington, D. C., and as such, limits have been placed on the number of segments which may be used to model an antenna structure. A total of 100 segments may be used, and in one sector of symmetry, only 22 segments may be used. This means that if no structure symmetry is employed, only 22 segments are allowed. The number of segments in one sector of symmetry plus the number of segments on the axis of symmetry must be less than or equal to 22. The program allows up to 12 sectors of rotational symmetry maximum. To illustrate this point, see example 3.

In order to adapt this program to larger machines, one must increase the size of the arrays and also check to see that the error checking limits in the MAIN program are modified accordingly. The structure array; CM (N,N) is a complex element array, and as such uses two floating point variables for each entry. It is the primary user of core in this program. The EINC array and the P array of common block /2/ are the excitation vector to the system, and an array of pivots respectively used by the factor and solve routines.

EQUATION OF CATENARY:

$$Z = Z_2 - (X_2 - X) \left[\frac{W}{2T} (X - X_1) + \frac{(Z_2 - Z_1)}{(X_2 - X_1)} \right]$$

W=WIRE WEIGHT/M
T=WIRE TENSION

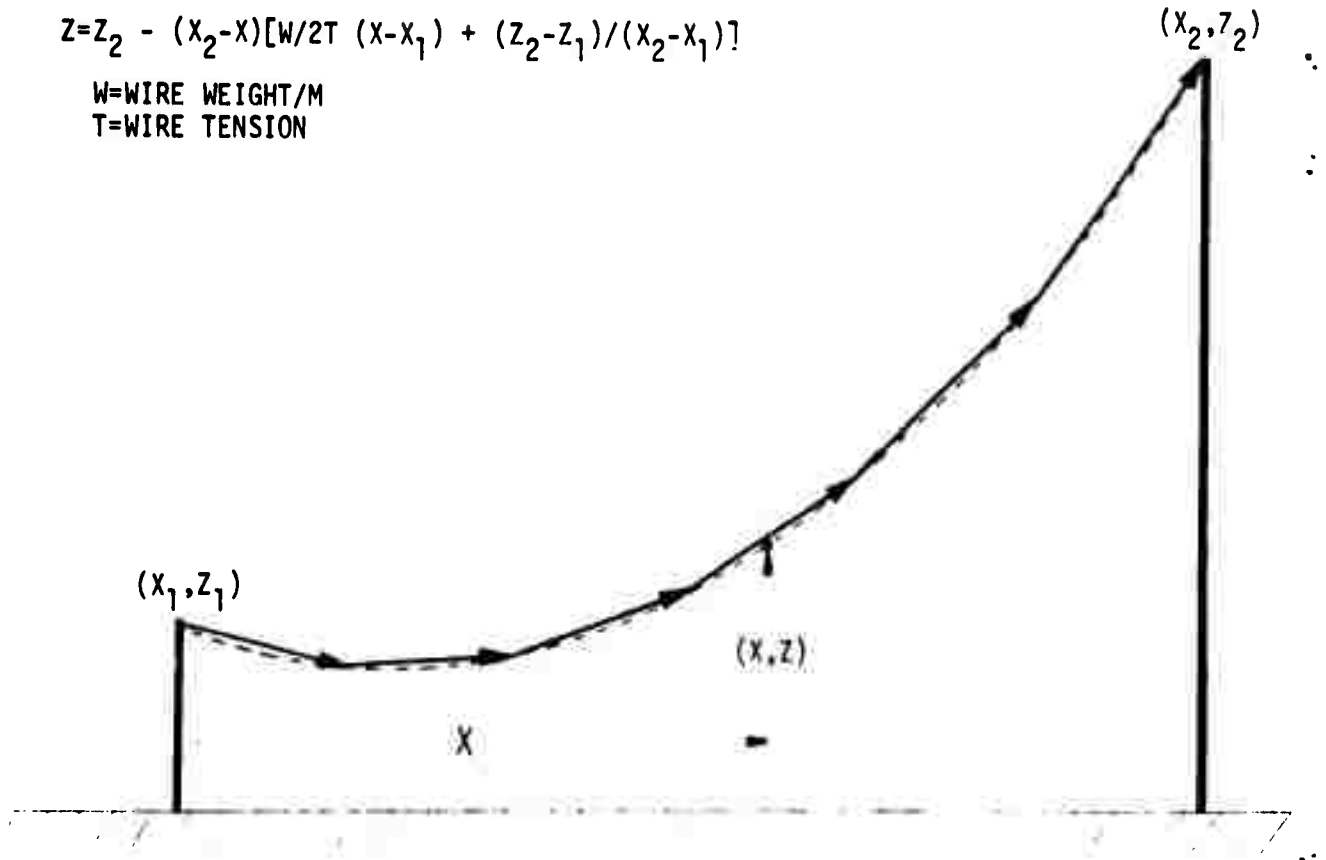


Figure 20. Catenary Element Modeled by Subroutine DATAGN.

Common block /1/ contains the physical description of the structure in terms of wavelength dimensions and direction angles, plus the electrical interconnection data.

Common block /3/ contains the direction cosines for each of the N segments.

Common block /4/ is used to hold data on multiple junctions set up by subroutine TRIO and JUNC. Presently, the program limits multiple junctions to 25 segments at a point, however, this is easily expanded.

Common block /SCRATM/ is used as a temporary scratch location by the factor and solve routines, and should be set by the size of the CM matrix.

Common block /SMAT/ is a square scratch matrix used by the symmetric factor and solve routines. Its dimensions are of the order of the number of symmetric sectors allowed, i.e., 12 x 12 in this version.

Dimensioned variables, CURR, CURI, ZLR, ZLI, ZLC, AIR, AII, BIR, BII, CIR, CII in the main program and NEFLD subroutine must also be expanded to the appropriate size of the maximum structure allowed.

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```
3  WRITE (61,71) A 65
    WRITE (61,72) A 66
    WRITE (61,73) A 67
    WRITE (61,74) (CON(I),I=1,10) A 68
    WRITE (61,69) NPRINT,ILOAD,IPOND,IOSCRN,INEAR,IFAR,NPWR A 69
    WRITE (61,75) GMZ,OR,NFS,COLAM A 70
    IF (IPOND.EQ.1) WRITE (61,103) A 71
    IF (KSYMP.EQ.2.AND.IPOND.EQ.0) WRITE (61,76) EPSR,SIG A 72
    IF (KSYMP.EQ.1) WRITE (61,66) A 73
C A 74
C BEFORE READING IN MORE DATA CARDS IN THE MAIN PROGRAM, CALL THE A 75
C DATA GENERATOR TO FILL UP THE GEOMETRY ARRAYS -- THE DATAON A 76
C WILL REQUIRE THE NEXT-N-DATA CARDS--SEE THE APPROPRIATE DATA ON A 77
C SUBROUTINE FOR ADDITIONAL DETAILS. A 78
C A 79
C CALL DATAON A 80
C A 81
C *****STRUCTURE SYMMETRY***** A 82
C A 83
C NEXT READ IN NP AND NX TO SET UP SYMMETRY CALCULATIONS A 84
C NP=NUMBER OF SEGMENTS IN A ROTATIONALLY SYMMETRIC SECTION A 85
C NX=NUMBER OF SEGMENTS ON THE AXIS OF ROTATION A 86
C NOTE THAT THE PROGRAM WILL WORK IF NP=N AND NX=0 A 87
C A 88
C A 89
    READ (60,69) NP,NX A 90
    WRITE (61,77) N,NP,NX A 91
    NSIZE=NP*NY A 92
    NCOLSY=N NX A 93
C A 94
C IF NSIZE IS GREATER THAN 22, TOO MANY SEGMENTS PER SECTOR ARE USED A 95
C AND IERR=1. A 96
C A 97
    IF (NSIZE.LE.22) GO TO 4 A 98
    IERR=1 A 99
    GO TO 64 A 100
4  IP=NPAGE A 101
    SLEN=0. A 102
    DO 8 I=1,N A 103
    IF (NPRINT*1) 7,5,5 A 104
5  AP=ALP(I)*TD A 105
    BT=BET(I)*TD A 106
    IP=IP+1 A 107
    IF (IP.LE.NPAGE) GO TO 6 A 108
    WRITE (61,78) A 109
    IP=1 A 110
6  WRITE (61,79) X(I),Y(I),Z(I),S1(I),B1(I),AP,BT,(CON1(I),I,CON2(I)) A 111
7  ALP1=ALP(I) A 112
    BET1=BET(I) A 113
    CALP=COS(ALP1) A 114
    SALP(I)=SIN(ALP1) A 115
    CAB(I)=CALP*COS(BET1) A 116
    SAB(I)=CALP*SIN(BET1) A 117
    SLEN=SLEN+S1(I) A 118
    IF (S1(I).GT.0.0) GO TO 8 A 119
    WRITE (61,80) I A 120
    STOP A 121
8  CONTINUE A 122
    WRITE (61,81) SLEN A 123
    ISEG(I3)=0 A 124
    DO 9 K=1,12 A 125
    ISEG(K)=0 A 126
    ENCR(K)=0.0 A 127
9  ENCI(K)=0.0 A 128
```

```
IN=0
WRITE (81,82)
C
C *****SOURCE EXCITATION*****
C READ N+5TH DATA CARD FOR THE SOURCE TO DRIVE ANTENNA
C IS = EXCITED SEGMENT NUMBER
C ECM = MAGNITUDE OF SEGMENT EXCITATION--VOLTS
C ECA = PHASE OF EXCITATION IN DEGREES
C NFLD = NFLD=1 IF MORE SEGMENTS ARE TO BE EXCITED, =0 TO END
C USE 1.0 VOLT AT 0 DEG PHAZE FOR CORRECT INPUT IMPEDANCE CALC
C 1 TO 12 SEGMENTS MAY BE SIMULTANEOUSLY EXCITED.
C
10 READ (80,83) IS,ECM,ECA,NFLD
WRITE (81,83) IS,ECM,ECA,NFLD
IF (1.LE.IS.AND.IS.LE.N) GO TO 11
C
C IERR=2 IF SOURCE SEGMENT IS LT 1 OR GREATER THAN NUM OF SEGS USED
C
C IERR=2
GO TO 84
11 K=ISEG(13)+1
IF (K.LE.12) GO TO 12
C
C IERR=3 IF TOO MANY SEGMENTS ARE SELECTED FOR SOURCES
C
C IERR=3
GO TO 84
12 ISEG(13)=K
ISEG(K)=IS
ECA=ECA+TA
ENCR(K)=ECM*COS(ECA)
ENCI(K)=ECM*SIN(ECA)
IF (NFLD.NE.0) GO TO 10
IF (ILOAD) 13,18,13
13 DO 14 I=1,NP
ZLC(I)=0.0
ZLR(I)=0.0
14 ZL(I)=0.0
C
C *****SEGMENT LOADING*****
C IF ILOAD = 1, THEN READ IN SEGMENT LOAD PARAMETERS
C ZR = RESISTANCE IN OHMS ON EACH OF THE SPECIFIED SEGMENTS
C ZI = INDUCTANCE IN HENRYS ON EACH OF THE SPECIFIED SEGMENTS
C ZC = CAPACITANCE IN FARADS ON EACH OF THE SPECIFIED SEGMENTS
C I1 = LOADS ARE CONNECTED FROM
C I2 = SEGMENTS I1 TO I2 INCLUSIVE
C NFLD = 1 FOR MORE LOAD CARDS, =0 FOR END OF LOAD INPUT DATA
C
C NOTE IF SYMMETRY IS EMPLOYED AND A ROTATIONALLY SYMMETRIC SEGMENT
C IS LOADED, THEN ALL LIKE SYMMETRIC SEGMENTS WILL ALSO BE LOADED.
C
15 READ (80,85) ZR,ZI,ZC,I1,I2,NFLD
IF (I2.EQ.0) I2=I1
WRITE (81,85) I1,I2,ZR,ZI,ZC
IF (I2.LT.I1) GO TO 17
IF (I1.LE.NCOLSYM) I1=MOD(I1,NP)
IF (I2.LE.NCOLSYM) I2=MOD(I2,NP)
IF (I1.GT.NCOLSYM) I1=MOD(I1,NCOLSYM)+NP
IF (I2.GT.NCOLSYM) I2=MOD(I2,NCOLSYM)+NP
DO 16 I=I1,I2
ZLC(I)=ZLC(I)+ZC
ZLR(I)=ZLR(I)+ZI
16 ZL(I)=ZL(I)+ZI
17 CONTINUE
```

A 129
A 130
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A 191
A 192

```
IF (INFLD.NE.0) GO TO 15 A 193
C A 194
C *****BEGIN FREQUENCY DO LOOP***** A 195
C A 196
18 DO 83 MKS=1,NFS A 197
FR=(GHZ*OR)/GHZ A 198
IF (MKS.EQ.1) FR=1. A 199
GHZ=GHZ*FR A 200
COLAM=0.299793/GHZ A 201
C A 202
C ZRAT1 = THE RATIO OF THE HALF-SPACE TO FREE-SPACE PLANE WAVE IMPED A 203
C A 204
ZRAT1=CSORT(1./((EPSR-FJ)*SIG*COLAM*59.921)) A 205
IF (MKS.EQ.1) GO TO 23 A 206
IF (INPRINT) 20,19,19 A 207
19 WRITE (61,87) GHZ A 208
20 DO 22 I=1,N A 209
X(I)=X(I)*FR A 210
Y(I)=Y(I)*FR A 211
Z(I)=Z(I)*FR A 212
S(I)=S(I)*FR A 213
B(I)=B(I)*FR A 214
IF (INPRINT) 22,21,21 A 215
21 WRITE (61,87) X(I),Y(I),Z(I),S(I),B(I) A 216
22 CONTINUE A 217
23 CONTINUE A 218
C A 219
C CHSETUP IS USED TO SET UP THE COMPLEX IMPEDANCE MATRIX A 220
C A 221
CALL CHSETUP (ZRAT1,KSYMP,IPGMD,IGSCRN) A 222
C A 223
C NOW ADD IN THE IMPEDANCE LOADING ON THE SELF TERMS. A 224
C A 225
IF (LOAD.EQ.0) GO TO 25 A 226
DO 24 I=1,NSIZE A 227
J=I A 228
IF (I.GT.NP) J=NCOLSYM+I-NP A 229
IF (ZLC(I) GT.0) CH(I,J)=CH(I,J)+FJ/ITP*GHZ*1E+9*ZLC(I)*S(I) A 230
24 CH(I,J)=CH(I,J)-ZLR(I)/S(I)-FJ*TP*GHZ*1E+9*ZC(I)/S(I) A 231
25 IF (INPRINT-1) 26,26,26 A 232
26 CONTINUE A 233
DO 27 I=1,NSIZE A 234
27 WRITE (61,88) I,(CME(KAY,I),J),J,AY+1,21,J+1,N A 235
C A 236
C *****SOLUTION OF THE MATRIX EQUATION***** A 237
C A 238
28 CONTINUE A 239
C A 240
C FACTOR THE IMPEDANCE MATRIX A 241
C A 242
C A 243
C NOP=NUMBER OF SYNETRIC SECTIONS--MUST BE LE 12 A 244
C A 245
NOP=(N-NX)/NP A 246
CALL FACTRCS (NP,NOP,NX,CM,P,NROW,NCOL,1) A 247
C A 248
C SET UP THE EXCITATION SOURCE VECTOR ANY OF THE N SEGMENTS MAY A 249
C BE EXCITED. A 250
C A 251
DO 29 I=1,N A 252
EINC(I)=CMPLX(0,0) A 253
ISEGL=ISEG(I) A 254
DO 30 I=1,ISEGL A 255
```

```

IS=ISEG(1)
30 EINC(15)=-CMPLX(ENCR(1),ENCI(1))/S(15)
IF (NPRINT.LT.0) GO TO 32
WRITE (61,90)
DO 31 IP=1,N
X1=REAL(EINC(IP))
X2=AIMAG(EINC(IP))
IF (X1.NE.0 .OR. X2.NE.0) WRITE (61,90) IP,X1,X2
31 CONTINUE
32 CONTINUE
C
C SOLVE THE SYSTEM OF EQUATIONS FOR SEGMENT CURRENTS---THE EINC
C ARRAY IS THE EXCITING SOURCE MATRIX TO THE SOLVE SUBROUTINE
C AND THE CURRENTS ARE RETURNED FROM SOLVE IN THIS ARRAY.
C CUR1 = INVICH1 * EINC1
C CALL SOLVECS (NP,NOP,NX,CM,P,EINC,NROW,NCOL)
C
C THE SEGMENT CURRENTS ARE RETURNED THROUGH THE EINC ARRAY
C
DO 33 I=1,N
CURR(I)=REAL(EINC(I))
33 CURR(I)=AIMAG(EINC(I))
NHALF=(N+1)/2
IP=NRPAGE
DO 34 J=1,N
J=J+NHALF
IP=IP+1
CHAG=SQRT(CURR(I)*CURR(I)+CUR(J)*CUR(J))
PH=TD*AATAN2(CURR(I),CURR(J))
IF (J.GT.N) GO TO 35
CHAGP=SQRT(CURR(J)*CURR(J)+CUR(I)*CUR(I))
PHP=TD*AATAN2(CURR(J),CURR(I))
IF (IP.LE.NRPAGE) GO TO 34
WRITE (61,91)
IP=1
34 WRITE (61,92) I,CURR(I),CUR(J),CHAG,PH,J,CURR(J),CUR(I),CHAGP,PH
IP
35 WRITE (61,92) I,CURR(I),CUR(I),CHAG,PH
C
C IF MORE THAN 1 SEGMENT IS EXCITED, AN INPUT IMPEDANCE IS NOT CALC
C
C TO OBTAIN THE CORRECT INPUT IMPEDANCE THE ANTENNA SEGMENTS
C MUST BE EXCITED WITH A 1 J VOLT AT 0 DEG PHASE SOURCE.
C
IF (ISEGL.GT.1) GO TO 36
ADMIT=CURR(15)+FJ*CUR(15)
ADMAG=CABS(ADMIT)
ADFAZ=AATAN2(CUR(15),CURR(15))*TD
ZPED=1./ADMIT
ZMAG=1./ADMAG
ZFAZ=-ADFAZ
ADMITR=REAL(ADMIT)
ADMITI=AIMAG(ADMIT)
ZPEDR=REAL(ZPED)
ZPEDI=AIMAG(ZPED)
WRITE (61,93) ADMITR,ADMITI,ZPEDR,ZPEDI,ADMAG,ADFAZ,ZMAG,ZFAZ
36 CONTINUE
IF (NPRINT.GT.0) WRITE (61,94)
C
C EXPAND THE SOLVED CURRENTS AT SEGMENT CENTERS INTO A CONSTANT PLUS
C A SINE AND COSINE TERM: I(S) = A + B*SIN(K*S) + C*COS(K*S)
C
DO 54 I=1,N

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A 256
A 257
A 258
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A 319

```

CALL TRIO (I,JCO1,JCO2,DIL,DIK)
S=SI(I)
CL=TP*DIL
CK=TP*DIK
SINL=SIN(CL)
COSL=COS(CL)
SINK=SIN(CK)
COSK=COS(CK)
SILK=SIN(CL*CK)
CELLO=SINL*SINK-SILK
IF (JCO1) 37,42,43
37 CRL0=0.0
CIL0=0.0
IF (NCIX.LT.1) GO TO 39
DO 38 K=1,NCIX
JIXK=JIX(K)
CRL0=CRL0+CURR(JIXK)
38 CIL0=CIL0+CUR(JIXK)
39 CONTINUE
IF (NCOX.LT.1) GO TO 41
DO 40 K=1,NCOX
JOKK=JOK(K)
CRL0=CRL0+CURR(JOKK)
40 CIL0=CIL0+CUR(JOKK)
41 CONTINUE
GO TO 44
42 CRL0=0.0
CIL0=0.0
GO TO 44
43 CRL0=CURR(JCO1)
CIL0=CUR(JCO1)
44 CRL=CURR(I)
CILL=CUR(I)
IF (JCO2) 45,50,51
45 CRLY=0.0
CILY=0.0
IF (NCOZ.LT.1) GO TO 47
DO 46 K=1,NCOZ
JOKK=JOK(K)
CRLY=CRLY+CURR(JOKK)
46 CILY=CILY+CUR(JOKK)
47 CONTINUE
IF (NCIZ.LT.1) GO TO 49
DO 48 K=1,NCIZ
JIZK=JIZ(K)
CRLY=CRLY+CURR(JIZK)
48 CILY=CILY+CUR(JIZK)
49 CONTINUE
GO TO 52
50 CRLY=0.0
CILY=0.0
GO TO 52
51 CRLY=CURR(JCO2)
CILY=CUR(JCO2)
52 AIR(I)=(CRL0*SINK-CRL*SILK+CRLY*SINL)/CELLO
AII(I)=(CIL0*SINK-CILL*SILK+CILY*SINL)/CELLO
BIR(I)=(CRL0*(COSK-I.0)+CRL*(COSL-COSK)+CRLY*(I-COSL))/CELLO
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A 382
A 383
53 WRITE (61,95) I,AIR(I),AII(I),BIR(I),BII(I),CIR(I),CII(I)
54 CONTINUE
C .....

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C
C IF (NPNR=1) NORMALIZE THE ANTENNA'S INPUT POWER TO A REFERENCE A 384
C 1-WATT INPUT. A 385
C A 386
C A 387
C . . . . . A 388
C IF (NPNR.EQ.0) GO TO 57 A 389
PWR5/M=0. A 390
DO 55 I=1,ISEGL A 391
IEXC(I)=ISEG(I) A 392
PWR=0.5*REAL((ENCR(I)*FJ*ENC(I))*CURR(IEXC(I)-FJ*CUR(IEXC(I))) A 393
PWRSUM=PWRSUM+PWR A 394
55 CONTINUE A 395
MATTI=SQRT(1./PWRSUM) A 396
WRITE (61,96) MATTI A 397
DO 56 I=1,N A 398
C A 399
C NORMALIZE THE ANTENNA CURRENTS TO AN EQUIVALENT 1 WATT DRIVE FOR A 400
C FIELD CALCULATIONS. A 401
C A 402
AIR(I)=AIR(I)*MATTI A 403
AII(I)=AII(I)*MATTI A 404
BIR(I)=BIR(I)*MATTI A 405
BII(I)=BII(I)*MATTI A 406
CIR(I)=CIR(I)*MATTI A 407
CII(I)=CII(I)*MATTI A 408
56 CONTINUE A 409
57 CONTINUE A 410
C A 411
C *****NEAR FIELD CALCULATIONS***** A 412
C A 413
IF (INEAR.EQ.1) CALL NEFLD (AIR,AII,BIR,BII,CIR,CII,ZRATI,KSYPH) A 414
C A 415
C A 416
C *****FAR FIELD CALCULATIONS***** A 417
C A 418
K=0 A 419
IF (IFAR.EQ.0) GO TO 63 A 420
58 K=K+1 A 421
C A 422
C FAR FIELD INPUT SELECTIONS--UP TO 5 CARDS MAY BE USED A 423
C THETR = INITIAL THETA COORDINATE--DEGREES A 424
C PHYR = INITIAL PHI COORDINATE--DEGREES A 425
C ETAR = ETA ANGLE--DEGREES A 426
C DTHR = DELTA THETA STEP--DEG A 427
C DPHR = DELTA PHI STEP--DEG A 428
C NTHR = NUMBER OF THETA STEPS A 429
C NPHR = NUMBER OF PHI STEPS A 430
C NFLD = +0 IF NO MORE FAR-FIELD CARDS, +1 FOR MORE INPUT A 431
C NEWSIG = +1 IF FAR FIELD CALC OVER DIFFERENT MEDIA, +0 IF NOT A 432
C SIOFF = CONDUCTIVITY OF FAR-FIELD MEDIA--MHOS/METER A 433
C EPSFF = RELATIVE DIELECTRIC CONSTANT OF FAR-FIELD MEDIA A 434
C A 435
READ (80,84) THETR(K),PHYR(K),ETAR(K),DTHR(K),DPHR(K),NTHR(K),NPHR A 436
I(K),NFLD,NEWSIG,SIOFF,EPSFF A 437
IF (NFLD.NE.0) GO TO 58 A 438
KPR=K A 439
IP=NRPAGE A 440
DO 82 KR=1,KPR A 441
DPHRK=DPHR(KR) A 442
NPHRK=NPHR(KR) A 443
THRO=THETR(KR) A 444
DTHRK=DTHR(KR) A 445
NTHRK=NTHR(KR) A 446
SETA=SIN(ETAR(KR)*TA) A 447

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CETA=COS(ETAR(KR)*TA)
PHRK=PHYR(KR)-DPHRC
C
C LOOP THROUGH THE PHI ANGLES
C
DO 62 KP=1,NPHRK
PHRK=PHRK+DPHRC
PHRKA=PHRK*TA
SPHI=SIN(PHRKA)
CPHI=COS(PHRKA)
THRK=THRO-DTHRK
PKX=-SPHI
PKY=CPHI
C
C LOOP THROUGH THE THETA ANGLES
C
DO 62 KT=1,NTHRK
THRK=THRK+DTHRK
THRKA=THRK*TA
STHET=SIN(THRKA)
ROX=STHET*CPHI
ROY=STHET*SPHI
ROZ=COS(THRKA)
EIXR=CETA*ROZ*CPHI-SETA*SPHI
EIYR=CETA*ROZ*SPHI+SETA*CPHI
EIZR=-CETA*STHET
IP=IP+1
CIX=CMPLX(0.,0.)
CIY=CMPLX(0.,0.)
CIZ=CMPLX(0.,0.)
C
C IF YOU WANT TO CALCULATE THE FAR FIELD RADIATION PATTERN OVER A
C MEDIA WHICH IS DIFFERENT THAN THE NEAR FIELD, SET NEWSIG=1 AND
C READ IN THE NEW GROUND PARAMETERS ON THE FAR-FIELD CARD.
C
IF (NEWSIG.EQ.1) ZRATI=CSQRT(1/(EPSFF-FJ*SIGFF*COLAM*59.92))
ZRSIN=CSQRT(1.-ZRATI*ZRATI)*STHET*STHET
RRV=(ROZ-ZRATI*ZRSIN)/(ROZ+ZRATI*ZRSIN)
RRH=-(ZRATI*ROZ-ZRSIN)/(ZRATI*ROZ+ZRSIN)
RRD=RRH-RRV
DO 60 I=1,N
CABI=CAB(I)
SABI=SAB(I)
SALPI=SALP(I)
RFL=-1.
DO 60 K=1,KSYMP
RFL=-RFL
ARG=X(I)*ROX+Y(I)*ROY+Z(I)*ROZ*RFL
CARG=COS(IP*ARG)
SARG=SIN(IP*ARG)
EXA=CMPLX(CSARG,SARG)
DODEL=ROX*CABI*ROY*SABI*ROZ*SALPI*RFL
XODEL=CABI-ROX*DODEL
YODEL=SABI-ROY*DODEL
ZODEL=SALPI*RFL-ROZ*DODEL
OMEGA=-DODEL
EL=PI*S(I)
SILL=OMEGA*EL
TOP=EL+SILL
BOT=EL-SILL
A=(2.0-OMEGA*OMEGA*EL*EL/3.0)*EL
IF (ABS(OMEGA) GE.1.E-7) A=2.*SIN(SILL)/OMEGA
TOO=1.0-TOP*TOP/6.0
IF (ABS(TOP) GE.1.E-9) TOO=SIN(TOP)/TOP
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BOO=1.0-BOT*BOT/6.0
IF (ABS(BOT).GE.1.E-9) BOO=SIN(BOT)/BOT
B=EL*(BOO-TOO)
C=EL*(BOO+TOO)
RR=A*AIR(1)+B*BIR(1)+C*CIR(1)
RI=A*AI(1)+B*BIR(1)+C*CI(1)
RRX=RR*XOEL
RRY=RR*YOEL
RRZ=RR*ZOEL
RIX=RI*XOEL
RIY=RI*YOEL
RIZ=RI*ZOEL
ERX=CMPLX(RRX,RIX)
ERY=CMPLX(RRY,RIY)
ERZ=CMPLX(RRZ,RIZ)
IF (K.NE.2) GO TO 59
EPY=PKX*ERX+PKY*ERY
EPX=PKX*EPY
EPY=PKY*EPY
ERX=- (RRY*ERX+RRD*EPX)
ERY=- (RRY*ERY+RRD*EPY)
ERZ=-RRV*ERZ
59 CIX=CIX+ERX*EXA
CIX=CIX+ERX*EXA
60 CIZ=CIZ+ERZ*EXA
ERX=CONST*CIX
ERY=CONST*CIX
ERZ=CONST*CIZ
EPC=ERX*EIXR+ERY*EIXR+ERZ*EIZR
ER=RE+(ERC)
E1=AIMAG(ERC)
ERAD=CABS(ERX*STHET*CPHI+ERY*STHET*SPHI+ERZ*ROZ)
ETHETA=CABS(ERX*ROZ*CPHI+ERY*ROZ*SPHI-ERZ*STHET)
EPMI=CABS(-ERX*SPHI+ERY*CPHI)
PHAZE=TD*ATAN2(E1,ER)
ERMAG=SQRT(ERAD**2+ETHETA**2+EPMI**2)
IF (IP.LE.NRPAGE) GO TO 61
WRITE (61,97) (CON(I),I=1,10)
IF (NEWSIG.EQ.1) WRITE (61,98) EPSFF,SIGFF
WRITE (61,99)
IP=1
61 WRITE (61,100) THRK,PHRK,ERAD,ETHETA,EPMI,ERMAG,PHAZE
62 CONTINUE
63 CONTINUE
GO TO 1
64 WRITE (61,101) IERR
65 CONTINUE
STOP
C
C
C
66 FORMAT (//33# ANTENNA IS MODELED IN FREE SPACE//)
67 FORMAT (21#1 FREQUENCY IN GHZ = .F12.8,/,6X,4HX(1),6X,4HY(1),6X,4
1#Z(1),5X,5#S1(1),5X,5#B1(1))
68 FORMAT (10#B)
69 FORMAT (1#B1)
70 FORMAT (2#10.5,215.2#10.5)
71 FORMAT (///,30#1)
72 FORMAT (/,30# HIRE ANTENNA MODELING PROGRAM,/)
73 FORMAT (30# )
74 FORMAT (///,1X,10#B/)
75 FORMAT (/1X,9#FREQUENCY12X,1#E13.5/1X,2#FREQUENCY INCREMENT =E1
13.5/1X,2#NO. FREQUENCY STEPS =/4/1X,2#WAVELENGTH (METERS) =E13
2.5//)
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76  FORMAT (1X,22HGROUND PLANE AT Z = 0./1X,22HDIELECTRIC CONSTANT =E A 576
    113.5/1X,12HCONDUCTIVITY9X,1H=E13.5/) A 577
77  FORMAT (7/1X,22HNUMBER OF SEGMENTS =14/1X,22HNO. SEG. IN A SECTO A 578
    1R =14/1X,31HNO. SEG. ON AXIS OF ROTATION = .14) A 579
78  FORMAT (49H) STRUCTURE GEOMETRY (DIMENSIONS IN WAVELENGTHS)/2X,6 A 580
    13HCOORDINATES OF SEG. CENTER SEG. WIRE ORIENTATION AN2 A 581
    22HOL'S CONNECTION DATA /6X33HX Y Z LENGTHH A 582
    38H RADIUS ALPHA BETA I- I I+ ) A 583
79  FORMAT (4F10.5,F10.7,2F10.3,3I5) A 584
80  FORMAT (30H NEGATIVE SEGMENT LENGTH. I=15) A 585
81  FORMAT (1/23H TOTAL WIRE LENGTH =E18 11) A 586
82  FORMAT (1/29H ANTENNA SOURCE DISTRIBUTIONS/24H SEG. VOLTAG A 587
    1E /26H NO. MAG. PHASE) A 588
83  FORMAT (115,2F10.5,15) A 589
84  FORMAT (5F10.5,4I5,2F5.1) A 590
85  FORMAT (3E10.3,3I5) A 591
86  FORMAT (9H SEGMENTS,14.5H THRU,14.12H LOADED WITH,E10 3,16H OHMS R A 592
    1E1STANCE,E10.3,23H HENRIES INDUCTANCE AND,E10 3,20H FARADS CAPACI A 593
    2TANCE.) A 594
87  FORMAT (6F10.5) A 595
88  FORMAT (/1X,3H I=13/1X,10E11 3) A 596
89  FORMAT (/1X,15,5X,E11.3,3X,E11.3) A 597
90  FORMAT (144H) SEGMENT EXCITATION (VOLTS/WAVELENGTH) /41H SEG NU A 598
    1MBER REAL PART IMAGINARY PART ) A 599
91  FORMAT (5H)SEG.3X,8HCURRENT=-,48X,4HSEG 4X,9HCURRENT -/1X,3HNO 5X,4 A 600
    1HREAL 63H IMAGINARY MAGNITUDE PHASE NO. A 601
    2 RE3MAL IMAGINARY MAGNITUDE PHASE ) A 602
92  FORMAT (1X,14,E13.4,E12.4,E16.8,F9.3,9X,14,E13.4,E12.4,E16.8,F9.3) A 603
93  FORMAT (14X,6HADMIT=2E12.4,10X,5HZPED=2E12.4/6XE10.4,2XF10.3,18X,E1 A 604
    10.4,2X,F10.3) A 605
94  FORMAT (14H) 16X,2HAR10X,2HA11X,2HBR10X,2HB11X,2HCR10X,2HC1) A 606
95  FORMAT (1X,14,3(1X,2E12.4)) A 607
96  FORMAT (39H) * * * * * //,40H VOLTAG A 608
    1E TO DRIVE ANTENNA AT 1-WATT = ,F12 5,/,50H ALL FIELDS ARE NO A 609
    2NORMALIZED TO THIS INPUT POWER//41H* * * * * A 610
    3 * * * ) A 611
97  FORMAT (2H) ,10AB/) A 612
98  FORMAT (22HDIELECTRIC CONSTANT = ,E12 4,20H AND CONDUCTIVITY = ,E1 A 613
    12.4,29H FOR FAR FIELD CALCULATIONS. //) A 614
99  FORMAT (141H OBSERVATION ANGLES ELECTRIC FIELD /4X,5H)THETA,5X A 615
    1,3HPHI,13X,11HR-COMPONENT,7X,15H)THETA-COMPONENT,6X,13HPHI-COMPONEN A 616
    2T10X,9HMAGNITUDE,10X,6H PHASE) A 617
100 FORMAT (2(1X,F8.3),12X,4(E12.4,8X),F10 4) A 618
101 FORMAT (17H STOP ERROR NO.15) A 619
102 FORMAT (32H A PERFECT GROUND PLANE AT Z 0 ) A 620
    END A 621

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SUBROUTINE CMSETUP (ZRATI,KSYP,IPGND,IGSCRN)
C
C SUBROUTINE CMSETUP IS USED TO SETUP THE COMPLEX IMPEDANCE MATRIX
C CH.
C
COMMON /1/ N,NP,X(100),Y(100),Z(100),S(100),B(100),ALP(100),BET(
100),ICON1(100),ICON2(100),COLAM,NX
INTEGER P
COMPLEX ZRATI,REFS,REFPS,ZRSIN
COMPLEX ZRATIS,ZSCRN,ZFACT
COMPLEX FJ,CM,EINC
COMMON /2/ CM(22,100),EINC(100),P(100)
COMMON /3/ CAB(100),SAB(100),SALP(100)
COMMON /4/ NCOX,JOX(25),NCIX,JIX(25),NCOZ,JOZ(25),NCIZ,JIZ(25)
COMMON /REFL/ RHOX,RHOY,RHOZ,CABJ,SABJ,SALPR,PX,PY,REFS,REFPS
DIMENSION ETR(3), ETI(3)
FJ=CMPLX(0.,1.)
PI=3.14159265
SIGN=-1.
NR=NP+NX
NSYM=N-NR
ZRATIS=ZRATI
IF (IGSCRN EQ.0) GO TO 1
C
C IF IGSCRN INPUT IN THE MAIN PROGRAM = 1, A RADIAL WIRE GROUND
C SCREEN IS SELECTED. YOU WILL BE REQUIRED TO READ IN THE NUMBER OF
C OF WIRES---NWIRES AND THE RADIUS OF THE RADIAL WIRES FOR EACH PASS
C THROUGH THE FREQUENCY DO LOOP.
C
READ (60,19) NWIRES,COM
C
C IF A RADIAL GROUND IS SELECTED, COMPUTE SOME PARAMETERS FROM THE
C NUMBER OF RADIAL WIRES---NWIRES, AND FROM THE RADIUS---COM (METERS)
C
COM=COM/COLAM
FLWIRE=NWIRES
SNFACT=FLWIRE*COM
M=2.*PI*0.299793*1.E+9/COLAM
UO=4.*PI*1.E-7
ZFACT=FJ*UO*M*COLAM/FLWIRE
ETAD=120.*PI
WRITE (61,20) NWIRES,COM
1 CONTINUE
DO 2 I=1,NR
DO 2 J=1,N
C(1,J)=CMPLX(0.,0.)
2
C
C J--SOURCE LOOP INDEX
C
DO 10 J=1,N
CALL TR10 (J,JCO1,JCO2,DIL,DIK)
S=S1(J)
B=B1(J)
XJ=X(J)
YJ=Y(J)
ZJ=Z(J)
CABJ=CAB(J)
SABJ=SAB(J)
SALPJ=SALP(J)
DO 10 I=1,NR
C
C I--OBSERVATION LOOP INDEX
C
IX=I
C 1
C 2
C 3
C 4
C 5
C 6
C 7
C 8
C 9
C 10
C 11
C 12
C 13
C 14
C 15
C 16
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IF (1.GT.NP) IX=1+NSYM C 65
XIJ=X(I,X)-XJ C 66
YIJ=Y(I,X)-YJ C 67
IJ=IX-J C 68
CABI=CAB(I,X) C 69
SABI=SAB(I,X) C 70
SALPI=SALP(I,X) C 71
RFL=-1. C 72
DO 10 IP=1,KSYP C 73
C C 74
C KSYMP LOOP--WHEN IP=1 DO FREE SPACE, IP=2 DO GROUND IMAGE CALC C 75
C RFL=-RFL C 76
C C 77
C ZP = DISTANCE FROM SOURCE SEGMENT TO OBSERVATION POIN MEAS C 78
C ALONG THE AXIS OF THE SOURCE SEGMENT AND A LINE PERPENDICULAR C 79
C TO THE AXIS AND THE OBSERVATION POINT C 80
C C 81
C RS = SQUARE OF THE DISTANCE BETWEEN THE SOURCE AND THE OBSERVATION C 82
C POINT. C 83
C C 84
C RH = PERPENDICULAR DISTANCE BETWEEN OBSERVATION POINT AND AXIS OF C 85
C SEGMENT. C 86
C C 87
C ZIJ=Z(I,X)-RFL*ZJ C 88
C Q1=CABI*CABJ+SABI*SABJ+SALPI*SALPJ+RFL C 89
C Q2=XIJ*CABI+YIJ*SABI+ZIJ*SALPI C 90
C ZP=XIJ*CABJ+YIJ*SABJ+ZIJ*SALPJ+RFL C 91
C RS=XIJ*XIJ+YIJ*YIJ+ZIJ*ZIJ C 92
C RM2=RS-ZP*ZP C 93
C IF (RM2.LT.1.E-20) GO TO 3 C 94
C RH=SQRT(RM2) C 95
C QP2=(Q2-ZP*Q1)/RH C 96
C GO TO 4 C 97
3 QP2=0. C 98
C RH=0. C 99
C 4 CONTINUE C 100
C C 101
C SKIP OVER THE GROUND IMAGE STUFF IF DOING A FREE SPACE CALC. C 102
C C 103
C IF (IP.NE.2) GO TO 10 C 104
C SALPR=SALPJ+RFL C 105
C RHOX=XIJ-CABJ*ZP C 106
C RHOY=YIJ-SABJ*ZP C 107
C RHOZ=ZIJ-SALPJ*ZP+RFL C 108
C RMAG=SQRT(RHOX*RHOX+RHOY*RHOY+RHOZ*RHOZ) C 109
C IF (RMAG.GT.1.E-6) GO TO 5 C 110
C RHOX=0. C 111
C RHOY=0. C 112
C RHOZ=0. C 113
C GO TO 6 C 114
5 RHOX=RHOX/RMAG C 115
C RHOY=RHOY/RMAG C 116
C RHOZ=RHOZ/RMAG C 117
C 6 RMAG=SQRT(YIJ*YIJ+XIJ*XIJ) C 118
C C 119
C MODIFY THE GROUND IMPEDANCE -- ZRAT1 -- BY THE RADIAL GROUND C 120
C SCREEN IMPEDANCE IN PARALLEL WITH THE GROUND IMPEDANCE. C 121
C C 122
C IF (IOSCRN.EQ.0) GO TO 7 C 123
C XSPEC=(X(I,X)*ZJ+Z(I,X)*XJ)/(Z(I,X)+ZJ) C 124
C YSPEC=(Y(I,X)*ZJ+Z(I,X)*YJ)/(Z(I,X)+ZJ) C 125
C RHOSPC=SQRT(XSPEC*XSPEC+YSPEC*YSPEC+SNFACT*SNFACT) C 126
C ZSCRN=ZFACT*RHOSPC/ALOG(RHOSPC/SNFACT) C 127
C C 128
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      ZRATI=(ZRATIS+ZSCRNI)/(ETA0+ZRATIS+ZSCRNI)
7  CONTINUE
   IF (RMAG.GT.1.E-6) GO TO 8
   PX=0.
   PY=0.
   CTH=1.
   ZRSIN=CMPLX(1.,0.)
   GO TO 9
8  PX=YIJ/RMAG
   PY=-XIJ/RMAG
   CTH=ZIJ/SQRT(RS)
   ZRSIN=CSORT(1.-ZRATI*ZRATI*(1.-CTH*CTH))
9  REFS=(CTH-ZRATI*ZRSIN)/(CTH+ZRATI*ZRSIN)
   REFPS=-(ZRATI*CTH-ZRSIN)/(ZRATI*CTH+ZRSIN)
   REFPS=REFPS-REFS
C
C  IF (IPOND = 1, A PERFECT GROUND IS MODELED BY FIXING THE REFLECTION
C  COEFFICIENTS AS FOLLOWS
C
   IF (IPOND.EQ.0) GO TO 10
   ZRSIN=CMPLX(1.,0.)
   REFS=CMPLX(1.,0.)
   REFPS=CMPLX(0.,0.)
10 CONTINUE
C
C  INTEGRATE THE E-FIELD AT OBS POINT DUE TO THE SOURCE SEGMENT
C
   CALL INTG (B,S,RH,ZP,O1,QP2,ETR,ETI,DIL,DIK,IJ,IP)
   IJ=1
   IF (IP.NE.2) GO TO 12
   DO 11 IC=1,3
   ETR(IC)=SIGN*ETR(IC)
11  ETI(IC)=SIGN*ETI(IC)
12  IF (JCO) 13,15,14
13  CALL JPELS (ETR(1),ETI(1),NCIX,JIX,NCOX,JOX,1)
   GO TO 15
14  CH(I,JCO1)=CH(I,JCO1)+ETR(1)+FJ*ETI(1)
15  CH(I,J)=CH(I,J)+ETR(2)+FJ*ETI(2)
   IF (JCO2) 16,10,17
16  CALL JPELS (ETR(3),ETI(3),MCOZ,JOZ,MC1Z,J1Z,1)
   GO TO 10
17  CH(I,JCO2)=CH(I,JCO2)+ETR(3)+FJ*ETI(3)
18 CONTINUE
   ZRATI=ZRATIS
   RETURN
C
C
C
19 FORMAT (15,E10.5)
20 FORMAT (////27H A RADIAL GROUND SCREEN OF .15,31H RADIALS WITH A WI
   RE RADIUS OF .E12.4,21H WAVELENGTH WAS USED ///)
   END
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C 129
C 130
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C 171
C 172
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C 174
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C 176
C 177
C 178
C 179
C 180-


```

SUBROUTINE EFLD (B,S,RH,ZP,IJ,EZRS,EZIS,ERRS,ERIS,EZRC,EZIC,ERRC,E E 1
IRIC,EZRK,EZIK,ERRK,ERIK) E 2
C E 3
SUBROUTINE EFLD COMPUTES THE AXIAL AND RADIAL ELECTRIC FIELDS REF E 4
TO THE SOURCE SEGMENT AT A SPECIFIED OBSERVATION POINT. E 5
C E 6
INPUTS: E 7
C B = SOURCE RADIUS E 8
C S = SOURCE LENGTH E 9
C RH = PERPENDICULAR DISTANCE BETWEEN OBS PT AND SOURCE SEG AXIS E 10
C ZP = AXIAL DISTANCE BETWEEN SOURCE SEG AND RH MEASURED ALONG E 11
C SOURCE AXIS. E 12
C IJ = OBSERVATION SEG NUMBER - SOURCE SEG NUMBER E 13
C OUTPUTS: E 14
C EZRS = AXIAL FIELD DUE TO SIN TERM--REAL PART E 15
C EZIS = IMAGINARY PART E 16
C ERRS = RADIAL FIELD DUE TO SIN TERM--REAL PART E 17
C ERIS = IMAGINARY PART E 18
C E 19
C EZRC = AXIAL FIELD DUE TO COS TERM--REAL PART E 20
C EZIC = IMAGINARY PART E 21
C ERRC = RADIAL FIELD DUE TO COS TERM--REAL PART E 22
C ERIC = IMAGINARY PART E 23
C E 24
C EZRK = AXIAL FIELD DUE TO CONSTANT TERM--REAL PART E 25
C EZIK = IMAGINARY PART E 26
C ERRK = RADIAL FIELD DUE TO CONSTANT TERM--REAL PART E 27
C ERK = IMAGINARY PART E 28
C E 29
COMMON /TH1/ ZPK,RKB2,IJX E 30
DATA ZZ,TP,TP2/100,363635,6,203105308,39,47841764/ E 31
IJX=IJ E 32
RHK=RH*TP E 33
ZPK=ZP*TP E 34
BK=B*TP E 35
RKB2=RHK*RHK+BK*BK E 36
RKB=SQRT(RKB2) E 37
COINC=RHK/RKB E 38
SKT=TP*S*0.5 E 39
ZD2=ZPK-SKT E 40
ZD1=ZPK+SKT E 41
R2KS=RKB2*ZD2*ZD2 E 42
R2K=SQRT(R2KS) E 43
R1KS=RKB2*ZD1*ZD1 E 44
R1K=SQRT(R1KS) E 45
SR2=SIN(R2K)/R2K*ZZ E 46
CR2=COS(R2K)/R2K*ZZ E 47
SR1=SIN(R1K)/R1K*ZZ E 48
CR1=COS(R1K)/R1K*ZZ E 49
SR2R=SR2/R2K E 50
SR2RR=SR2/R2KS E 51
CR2R=CR2/R2K E 52
CR2RR=CR2/R2KS E 53
SR1R=SR1/R1K E 54
SR1RR=SR1/R1KS E 55
CR1R=CR1/R1K E 56
CR1RR=CR1/R1KS E 57
CST=COS(SKT) E 58
SST=SIN(SKT) E 59
T1=(CR2R-SR2RR)*ZD2 E 60
T2=(CR1R-SR1RR)*ZD1 E 61
T3=(SR2R+CR2RR)*ZD2 E 62
T4=(SR1R+CR1RR)*ZD1 E 63
E 64

```

```
T1S=T1*SST
T2S=-T2*SST
T3S=T3*SST
T4S=-T4*SST
EZRS=(SR2-SR1)*CST*T1S-T2S
EZIS=(CR2-CR1)*CST-T3S+T4S
ERRS=-((SR2*ZD2-SR1*ZD1)*CST+(SR2*SR1)*SST*T1S+ZD2-T2S*ZD1)/RKB*CO
IINC
ERIS=-((CR2*ZD2-CR1*ZD1)*CST+(CR2*CR1)*SST-T3S*ZD2+T4S*ZD1)/RKB*CO
IINC
T1S=T1*CST
T2S=T2*CST
T3S=T3*CST
T4S=T4*CST
EZRC=(-(SR2*SR1)*SST*T1S-T2S)
EZIC=-((CR2*CR1)*SST-T3S+T4S)
ERRC=-((SR2*ZD2-SR1*ZD1)*SST+(SR2-SR1)*CST*T1S+ZD2-T2S*ZD1)/RKB*C
I0INC
ERIC=-((CR2*ZD2-CR1*ZD1)*SST+(CR2-CR1)*CST-T3S*ZD2+T4S*ZD1)/RKB*C
I0INC
ERRK=RKB*(CA2R-SA2RR-CR1R-SR1RR)*COINC
ERIK=-RKB*(SA2R+CR2RR-SR1R-CR1RR)*COINC
C
C
C
C
ONLY THE AXIAL FIELD DUE TO THE CONSTANT CURRENT TERM MUST BE
INTEGRATED NUMERICALLY
CALL INTX (-SKT,SKT,RKBP,ZPK,BK,IJ,CINT,SINT)
EZRK=-ZZ*SINT*T1-T2
EZIK=-ZZ*CINT-T3+T4
RETURN
END
E 85
E 86
E 87
E 88
E 89
E 90
E 91
E 92
E 93
E 94
E 95-
```



```

SUBROUTINE FACTOR (N,A,P,NDIM)
C
C SUBROUTINE TO FACTOR A MATRIX INTO A UNIT LOWER TRIANGULAR MATRIX
C UPPER TRIANGULAR MATRIX USING THE GAUSS-DOOLITTLE ALGORITHM PRESEN
C PAGES 411-416 OF A. RALSTON--A FIRST COURSE IN NUMERICAL ANALYSIS.
C BELOW REFER TO COMMENTS IN RALSTONS TEXT.
C
C COMPLEX A,D,DETER
C INTEGER R,P,RM1,RP1,PJ,PR
C DIMENSION A(NDIM,NDIM), P(NDIM)
C COMMON /SCRATH/ D(100)
C IFLG=0
C DO 9 R=1,N
C
C STEP 1
C
C DO 1 K=1,N
C D(K)=A(K,R)
C CONTINUE
C
C STEPS 2 AND 3
C
C RM1=R-1
C IF (RM1.LT.1) GO TO 4
C DO 3 J=1,RM1
C PJ=P(J)
C A(J,R)=D(PJ)
C D(PJ)=D(J)
C JP1=J+1
C DO 2 I=JP1,N
C D(I)=D(I)-A(I,J)*A(J,R)
C CONTINUE
C
C CONTINUE
C
C CONTINUE
C
C STEP 4
C
C DMAX=D(R)*CONJG(D(R))
C P(R)=R
C RP1=R+1
C IF (RP1.GT.N) GO TO 6
C DO 5 I=RP1,N
C ELMAG=D(I)*CONJG(D(I))
C IF (ELMAG.LT.DMAX) GO TO 5
C DMAX=ELMAG
C P(R)=I
C CONTINUE
C
C CONTINUE
C
C IF (DMAX.LT.1.E-10) IFLG=1
C PR=P(R)
C A(R,R)=D(PR)
C D(PR)=D(R)
C
C STEP 5
C
C IF (RP1.GT.N) GO TO 8
C DO 7 I=RP1,N
C A(I,R)=D(I)/A(R,R)
C CONTINUE
C
C CONTINUE
C
C IF (IFLG.EQ.0) GO TO 9
C WRITE (6,10) R,DMAX
C IFLG=0
C CONTINUE
C
F 1
F 2
F 3
F 4
F 5
F 6
F 7
F 8
F 9
F 10
F 11
F 12
F 13
F 14
F 15
F 16
F 17
F 18
F 19
F 20
F 21
F 22
F 23
F 24
F 25
F 26
F 27
F 28
F 29
F 30
F 31
F 32
F 33
F 34
F 35
F 36
F 37
F 38
F 39
F 40
F 41
F 42
F 43
F 44
F 45
F 46
F 47
F 48
F 49
F 50
F 51
F 52
F 53
F 54
F 55
F 56
F 57
F 58
F 59
F 60
F 61
F 62
F 63
F 64
```

```
      RETURN  
C  
C  
C  
18  FORMAT (7H PIVOT(,13,24)=,E16.8)  
      END
```

```
F 85  
F 86  
F 87  
F 88  
F 89  
F 90-
```

```
C SUBROUTINE FACTORS (N,NOP,A,P,NROW,NCOL) G 1
C SUBROUTINE FACTORS IS USED TO SET-UP THE FACTORIZATION OF THE G 2
C SYMMETRIC PART OF THE IMPEDANCE MATRIX G 3
C G 4
C G 5
C COMPLEX A,D,DETER,S G 6
C INTEGER P G 7
C COMMON /SMAT/ S(12,12) G 8
C DIMENSION A(NROW,NCOL), P(NCOL) G 9
C COMMON /SCRATH/ D(100) G 10
C IF (NOP.EQ.1) GO TO 6 G 11
C PHAZ=6.2831853072/NOP G 12
C DO 1 I=2,NOP G 13
C DO 1 J=1,NOP G 14
C ARG=PHAZ*(I-1)*(J-1) G 15
C XXX=COS(ARG) G 16
C YYY=SIN(ARG) G 17
C S(I,J)=CHPLX(XXX,YYY) G 18
1 S(J,I)=S(I,J) G 19
C DO 5 I=1,N G 20
C DO 5 J=1,N G 21
C DO 2 K=1,NOP G 22
C KA=J+(K-1)*N G 23
C D(K)=A(I,KA) G 24
2 CONTINUE G 25
C DETER=D(1) G 26
C DO 3 KK=2,NOP G 27
3 DETER=DETER*D(KK) G 28
C A(I,J)=DETER G 29
C DO 5 K=2,NOP G 30
C KA=J+(K-1)*N G 31
C DETER=D(1) G 32
C DO 4 KK=2,NOP G 33
4 DETER=DETER*D(KK)*S(K,KK) G 34
5 A(I,KA)=DETER G 35
6 DO 7 KK=1,NOP G 36
C KA=(KK-1)*N+1 G 37
C CALL FACTOR (N,A(I,KA),P(KA),NROW) G 38
7 CONTINUE G 39
C RETURN G 40
C END G 41
```



```

SUBROUTINE OF (ZK,CO,S1)
C
C SUBROUTINE OF PROVIDES THE FUNCTION TO BE NUMERICALLY INTEGRATED
C BY INTX. THE FORM IS:  $EXP(I \cdot K \cdot R) / K \cdot R$ . THE REAL PART IS RETURNED
C THROUGH CO AND THE IMAGINARY PART OF THE INTEGRAND IS S1.
C
COMMON /TH1/ ZPK,RKB2,IJ
ZDK=ZK-ZPK
RK=SQRT(RKB2+ZDK*ZDK)
S1=SIN(RK)/RK
IF (IJ) 1,2,1
1 CO=COS(RK)/RK
RETURN
C
C WHEN I=J, SUBTRACT OUT A SINGULAR (1/RK) POINT--IT WILL BE
C INCLUDED AT A LATER POINT IN SUBROUTINE INTX
C
2 CO=(COS(RK)-1.)/RK
RETURN
END

```

```

SUBROUTINE ON (EZR,EZI,ERR,ERI)
C
C SUBROUTINE ON MODIFIES THE PERFECT IMAGE FIELDS BY THE
C APPROPRIATE REFLECTION COEFFICIENTS EVALUATED AT THE SPECULAR
C POINTS.
C
COMMON /REFL/ RHOX,RHOY,RHOZ,CABJ,SABJ,SALPR,PX,PY,REFS,REFPS
EZ=CMPLX(EZR,EZI)
ER=CMPLX(ERR,ERI)
ERX=RHOX*ER+CABJ*EZ
ERY=RHOY*ER+SABJ*EZ
ERZ=RHOZ*ER+SALPR*EZ
EPX=PX*ERX+PY*ERY
EPY=PY*ERY
EPX=PX*EPY
EPY=PY*EPY
ERX=REFS*ERX+REFPS*EPX
ERY=REFS*ERY+REFPS*EPY
ERZ=REFS*ERZ
EZ=ERX*CABJ+ERY*SABJ+ERZ*SALPR
ER=ERX*RHOX+ERY*RHOY+ERZ*RHOZ
EZR=REAL(EZ)
EZI=AIMAG(EZ)
ERR=REAL(ER)
ERI=AIMAG(ER)
RETURN
END
J 1
J 2
J 3
J 4
J 5
J 6
J 7
J 8
J 9
J 10
J 11
J 12
J 13
J 14
J 15
J 16
J 17
J 18
J 19
J 20
J 21
J 22
J 23
J 24
J 25
J 26
J 27-
```

```

SUBROUTINE INTG (B,S,RH,ZP,Q1,QP2,ETR,ETI,DIL,DIK,IJ,IP)      K  1
C                                                              K  2
C SUBROUTINE INTG IS USED IN COMPUTING THE ENTRIES FOR THE CH MATRIX K  3
C IT COMPUTES THE E-FIELD AT AN OBSERVATION SEGMENT DUE TO A UNIT K  4
C CURRENT ON THE SOURCE SEGMENT. THE SUBROUTINE RETURNS THE K  5
C INTERPOLATED TANGENTIAL FIELDS FOR THE SEGMENTS AS FOLLOWS: K  6
C ETR(1) AND ETI(1) ARE THE FIELDS FOR SEGMENTS CONNECTED TO THE K  7
C NEGATIVE END OF THE SOURCE SEGMENT. K  8
C ETR(2) AND ETI(2) ARE THE FIELDS FOR THE OBSERVATION POINT SEG K  9
C ETR(3) AND ETI(3) ARE THE FIELDS FOR SEGMENTS CONNECTED TO THE K 10
C POSITIVE END OF THE SOURCE SEGMENT. K 11
C                                                              K 12
C DIMENSION ETR(3), ETI(3) K 13
C DATA TP/S.203105308/ K 14
C                                                              K 15
C COMPUTE THE E-FIELDS REFERENCED TO THE SOURCE SEGMENT. K 16
C                                                              K 17
C CALL EFLD (B,S,RH,ZP,IJ,EZRS,EZIS,ERRS,ERIS,EZRC,EZIC,ERRC,ERIC,EZ K 18
C IRK,EZIK,ERRK,ERIK) K 19
C IF (IP.NE.2) GO TO 1 K 20
C                                                              K 21
C IF COMPUTATION IS PERFORMED FOR THE IMAGE FIELDS, MODIFY THE K 22
C PERFECT GROUND IMAGE FIELDS BY THE APPROPRIATE REFLECTION COEFF K 23
C                                                              K 24
C CALL GN (EZRS,EZIS,ERRS,ERIS) K 25
C CALL GN (EZRC,EZIC,ERRC,ERIC) K 26
C CALL GN (EZRK,EZIK,ERRK,ERIK) K 27
C                                                              K 28
C TAKE A DOT PRODUCT OF THE SOURCE FIELDS TO COMPUTE THE TANGENTIAL K 29
C FIELDS AT THE OBSERVATION POINT. K 30
C                                                              K 31
C ETRS=EZRS*Q1+ERRS*QP2 K 32
C ETIS=EZIS*Q1+ERIS*QP2 K 33
C ETRC=EZRC*Q1+ERRC*QP2 K 34
C ETIC=EZIC*Q1+ERIC*QP2 K 35
C ETRK=EZRK*Q1+ERRK*QP2 K 36
C ETIK=EZIK*Q1+ERIK*QP2 K 37
C CL=TP*DIL K 38
C CK=TP*DIK K 39
C SINL=SIN(CL) K 40
C COSL=COS(CL) K 41
C SINCK=SIN(CK) K 42
C COSCK=COS(CK) K 43
C SILK=SIN(CL*CK) K 44
C CONS=SINL*SINCK-SILK K 45
C ETR(1)=(SINCK*ETRK+(COSCK-1.)*ETRS-SINCK*ETRC)/CONS K 46
C ETI(1)=(SINCK*ETIK+(COSCK-1.)*ETIS-SINCK*ETIC)/CONS K 47
C ETR(2)=(1-SILK*ETRK+(COSL-COSK)*ETRS+(SINL+SINCK)*ETRC)/CONS K 48
C ETI(2)=(1-SILK*ETIK+(COSL-COSK)*ETIS+(SINL+SINCK)*ETIC)/CONS K 49
C ETR(3)=(SINL*ETRK+(1.-COSL)*ETRS-SINL*ETRC)/CONS K 50
C ETI(3)=(SINL*ETIK+(1.-COSL)*ETIS-SINL*ETIC)/CONS K 51
C RETURN K 52
C END K 53

```

```

SUBROUTINE INTX (EL1,EL2,RKB2,ZPK,B,IJ,SG1,SG1)          L  1
C                                                       L  2
C   INTX IS AN ADAPTIVE RHOMBERG INTEGRATION SCHEME     L  3
C                                                       L  4
C                                                       L  5
C   REFERENCE: JOURNAL OF COMPUTATIONAL PHYSICS 5, PP 265-279 L  6
C   1970-- A VARIABLE INTERVAL WIDTH QUADRATURE TECHNIQUE BASED ON L  7
C   RHOMBERG'S METHOD , E. K. MILLER, ET AL             L  8
C                                                       L  9
C   DATA NX,NH,NTS,RX/1.65536,4,1.E-4/              L 10
C   Z=EL1                                              L 11
C   ZE=EL2                                            L 12
C   IF (IJ.EQ 0) ZE=0.                                L 13
C   S=ZE-Z                                            L 14
C   EP=10*NM                                          L 15
C   EP=S/EP                                           L 16
C   ZEND=ZE-EP                                        L 17
C   SQR=0.0                                          L 18
C   SG1=0.0                                          L 19
C   NS=NX                                             L 20
C   NT=0                                             L 21
C                                                       L 22
C   OF IS THE FUNCTION TO BE NUMERICALLY INTEGRATED L 23
C                                                       L 24
C   CALL OF (Z,G1R,G1I)                               L 25
C   DZ=S/NS                                           L 26
C   DZOT=DZ*0.5                                       L 27
C   ZP=Z+DZ                                           L 28
C   IF (ZP-ZE) 3,3,2                                   L 29
C   DZ=ZE-Z                                           L 30
C   IF (ABS(DZ)-EP) 17,17,3                             L 31
C   DZOT=DZ/2.                                         L 32
C   ZP=Z+DZOT                                         L 33
C   CALL OF (ZP,G3R,G3I)                               L 34
C   ZP=Z+DZ                                           L 35
C   CALL OF (ZP,G5R,G5I)                               L 36
C   T00R=(G1R+G5R)*DZOT                               L 37
C   T00I=(G1I+G5I)*DZOT                               L 38
C   T01R=(T00R+DZ*G3R)*0.5                            L 39
C   T01I=(T00I+DZ*G3I)*0.5                            L 40
C   T10R=(4.0*T01R-T00R)/3.0                          L 41
C   T10I=(4.0*T01I-T00I)/3.0                          L 42
C   TE1R=TEST(T01R,T10R)                              L 43
C   TE1I=TEST(T01I,T10I)                              L 44
C   IF (TE1I-RX) 5,5,6                                 L 45
C   IF (TE1R-RX) 8,8,6                                 L 46
C   ZP=Z+DZ*0.25                                       L 47
C   CALL OF (ZP,G2R,G2I)                               L 48
C   ZP=Z+DZ*0.75                                       L 49
C   CALL OF (ZP,G4R,G4I)                               L 50
C   T02R=(T01R+DZOT*(G2R+G4R))*0.5                    L 51
C   T02I=(T01I+DZOT*(G2I+G4I))*0.5                    L 52
C   T11R=(4.0*T02R-T01R)/3.0                          L 53
C   T11I=(4.0*T02I-T01I)/3.0                          L 54
C   T20R=(16.0*T11R-T10R)/15.0                       L 55
C   T20I=(16.0*T11I-T10I)/15.0                       L 56
C   TE2R=TEST(T11R,T20R)                              L 57
C   TE2I=TEST(T11I,T20I)                              L 58
C   IF (TE2I-RX) 7,7,14                               L 59
C   IF (TE2R-RX) 9,9,14                               L 60
C   SQR=SQR+T10I                                       L 61
C   SG1=SG1+T10I                                       L 62
C   NT=NT+2                                             L 63
C   GO TO 10                                           L 64

```



```
9   SQR=SQR+T20R          L 85
    SG1=SG1+T20I          L 86
    NT=NT+1               L 87
10  Z=Z+DZ               L 88
    IF (Z-ZEND) 11,17,17 L 89
11  G1R=GSR              L 90
    G1I=GSI              L 91
    IF (NT-NTS) 1,12,12  L 92
12  IF (NS-NX) 1,1,13    L 93
13  NS=NS/2              L 94
    NT=1                 L 95
    GO TO 1              L 96
14  NT=0                 L 97
    IF (NS-NM) 16,15,15  L 98
15  WRITE (61,20) Z      L 99
    GO TO 9              L 100
16  NS=NS*2             L 101
    DZ=S/NS              L 102
    DZOT=DZ*0.5         L 103
    GSR=GSR              L 104
    GSI=GSI              L 105
    G3R=G2R              L 106
    G3I=G2I              L 107
    GO TO 4              L 108
17  CONTINUE            L 109
C                                     L 110
C   IF I=J AN ANALYTIC DIFFERENCE TERM IN THE INTEGRAND L 111
C   IS NOW INCLUDED IN THE CONTRIBUTION L 112
C                                     L 113
C   IF (IJ) 19,18,15    L 114
18  SGR=2.*(SGR+ALOG((SQRT(B*B+S*S)+S)/B)) L 115
    SG1=2.*SG1          L 116
19  CONTINUE            L 117
    RETURN              L 118
C                                     L 119
C                                     L 120
C                                     L 121
20  FORMAT (24H STEP SIZE LIMITED AT Z=F10.5) L 122
    END                  L 123
```

```

SUBROUTINE JNELS (ETR,ETI,NCP,JP,NCH,JM,I)
C
C JNELS HANDLES THE STUFFING OF THE COMPLEX IMPEDANCE MATRIX--CM.
C THE SUBROUTINE IS CALLED ONLY WHEN AN ICON VALUE IS NEGATIVE--
C INDICATING EITHER A MULTIPLE JUNCTION OR A CHANGE IN REF POLARITY.
C
C INPUTS:
C I = OBSERVATION POINT SEGMENT
C ETR = TANGENTIAL ELECTRIC FIELD (REAL) AT SEGMENT I
C ETI = TANGENTIAL ELECTRIC FIELD (IMAG) AT SEGMENT I
C NCP = NUMBER OF SEGMENTS CONNECTED TO POSITIVE END OF JTH SEG
C JP = ARRAY OF SEGMENT NUMBERS CONNECTED TO POS END OF JTH SEG
C NCH = NUMBER OF SEGMENTS CONNECTED TO NEGATIVE END OF JTH SEG
C JM = ARRAY OF SEGMENT NUMBERS CONNECTED TO NEG END OF JTH SEG
C
C INTEGER P
C COMPLEX CM,FJ,EINC
COMMON /2/ CM(22,100),EINC(100),P(100)
DIMENSION JP(25), JM(25)
FJ=CHPLX(0.,1.)
IF (NCP.LT.1) GO TO 2
DO 1 J=1,NCP
JPJ=JP(J)
1 CM(I,JPJ)=CM(I,JPJ)+ETR+FJ*ETI
2 CONTINUE
IF (NCH.LT.1) GO TO 4
DO 3 J=1,NCH
JMJ=JM(J)
3 CM(I,JMJ)=CM(I,JMJ)-ETR-FJ*ETI
4 CONTINUE
RETURN
END
M 1
M 2
M 3
M 4
M 5
M 6
M 7
M 8
M 9
M 10
M 11
M 12
M 13
M 14
M 15
M 16
M 17
M 18
M 19
M 20
M 21
M 22
M 23
M 24
M 25
M 26
M 27
M 28
M 29
M 30
M 31
M 32-

```

```

SUBROUTINE JUNC (J,JNO,NC1,NSEG1,NC2,NSEG2,D)           N  1
C                                                       N  2
C SUBROUTINE JUNC IS USED TO CHECK SEGMENT ENDS FOR MULTIPLE N  3
C JUNCTIONS. THIS SUBR IS ONLY CALLED IF AN ICON VALUE IS NEGATIVE. N  4
C INPUTS:                                               N  5
C   J           = SEGMENT NUMBER TO BE TESTED          N  6
C   JNO        = ICON VALUE OF J-TH SEGMENT TO BE CHECKED N  7
C OUTPUTS:                                             N  8
C   NC1        = NUMBER OF SEGMENTS WHOSE NEG END IS CONNECTED TO JNO N  9
C   NC2        = NUMBER OF SEGMENTS WHOSE POS END IS CONNECTED TO JNO N 10
C   NSEG1     = ARRAY OF SEG NUMBERS WHOSE NEG END IS CONNECTED TO JNO N 11
C   NSEG2     = ARRAY OF SEG NUMBERS WHOSE POS END IS CONNECTED TO JNO N 12
C   D          = AVG LENGTH OF J-TH SEG AND AVG OF ALL OTHER CONN4 SEGS. N 13
C                                                       N 14
COMMON /1/ N,NP,X(100),Y(100),Z(100),S1(100),B1(100),ALP(100),BET( N 15
1100),ICON1(100),ICON2(100),COLAM,NX N 16
DIMENSION NSEG1(25), NSEG2(25) N 17
NC1=0 N 18
NC2=0 N 19
SNC=0.0 N 20
C                                                       N 21
C CHECK FOR NEG ENDS CONNECTED TO JNO N 22
C                                                       N 23
DO 4 I=1,N N 24
IF (ICON1(I)-JNO) 2,1,2 N 25
IF (I.EQ.J) GO TO 2 N 26
NC1=NC1+1 N 27
IF (NC1.GT.25) GO TO 5 N 28
NSEG1(NC1)=I N 29
SNC=SNC+S1(I) N 30
C                                                       N 31
C CHECK FOR POS ENDS CONNECTED TO JNO N 32
C                                                       N 33
2 IF (ICON2(I)-JNO) 4,3,4 N 34
IF (I.EQ.J) GO TO 4 N 35
NC2=NC2+1 N 36
IF (NC2.GT.25) GO TO 5 N 37
NSEG2(NC2)=I N 38
SNC=SNC+S1(I) N 39
4 CONTINUE N 40
FC=NC1+NC2 N 41
C                                                       N 42
C COMPUTE AN AVERAGE SEGMENT LENGTH FOR THE MULTIPLE JUNCTION N 43
C                                                       N 44
D=(S1(J)+SNC/FC)/2.0 N 45
RETURN N 46
5 WRITE (61,6) JNO N 47
STOP N 48
C                                                       N 49
C                                                       N 50
C                                                       N 51
6 FORMAT (41H ERROR - TOO MANY CONNECTIONS TO JUNCTION)4) N 52
END N 53-

```

```

SUBROUTINE LINE (I,NS,TAU,WT,TENS,WRAD,X1,Y1,Z1,X2,Y2,Z2)      0 1
COMMON /1/ N,NP,X(100),Y(100),Z(100),S1(100),B1(100),ALP(100),BET( 0 2
100),ICON1(100),ICON2(100),COLAM,NX                          0 3
C                                                                0 4
C THIS SUBROUTINE IS USED TO CALCULATE THE GEOMETRIC COORDINATES OF 0 5
C EACH MAJOR ANTENNA ARM. THE DATA GENERATED BY THIS SUBROUTINE IS 0 6
C X,Y,Z COORDINATE OF THE CENTER OF A SEGMENT PLUS THE ALPHA AND BET 0 7
C ORIENTATION ANGLES OF EACH POINT. INTERCONNECTION DATA IS ALSO GE 0 8
C EACH OF THE SEGMENTS. BY SPECIFYING THE PROPER PARAMETERS IN THE 0 9
C CALL ONE HAS THE CHOICE OF USING A TAPERED SEGMENT LENGTH WITH A 0 10
C CATENARY FORM OR A LINEAR FORM.                               0 11
C                                                                0 12
C IF TENS = 0. THEN DONT MODEL A CATENARY                       0 13
C IF (TENS.LE 1.) TENS=1.E100                                   0 14
C XINC=X2-X1                                                    0 15
C YINC=Y2-Y1                                                    0 16
C ZINC=Z2-Z1                                                    0 17
C RHO=SQRT(XINC**2+YINC**2+ZINC**2)                              0 18
C RHOXY=SQRT(XINC**2+YINC**2)                                    0 19
C BETA=AATAN2(YINC,XINC)                                         0 20
C EXPSUM=0.                                                      0 21
C NEXP=-1                                                        0 22
C                                                                0 23
C CALCULATE SEGMENT LENGTH SLO IF TAPERED SEG IS USED SPECIFY TA; 0 24
C                                                                0 25
C DO 1 LS=1,NS                                                  0 26
C EXPSUM=EXPSUM+(1+TAU)**(LS-1)                                  0 27
C SLO=RHO/EXPSUM                                                0 28
C                                                                0 29
C CALC AN APPROX VALUE FOR THE ALPHA ANGLE USING ST LINE SEG ALFA 0 30
C WILL BE USED TO DETERMINE THE INCREMENTAL X AND Y STEP      0 31
C                                                                0 32
C ALFA=AATAN2(ZINC,RHOXY)                                        0 33
C CA=COS(ALFA)                                                  0 34
C SA=SIN(ALFA)                                                  0 35
C CAB=CA*COS(BETA)                                             0 36
C SAB=CA*SIN(BETA)                                             0 37
C                                                                0 38
C SET UP SEGMENT PARAMETERS IF WIRE WEIGHT IS SPECIFIED, CATENARY 0 39
C BE CALCULATED FOR VERTICAL ELEMENTS NO CATENARY WILL BE USED 0 40
C                                                                0 41
C NEND=NS+1-1                                                  0 42
C XX1=X1                                                         0 43
C YY1=Y1                                                         0 44
C ZZ1=Z1                                                         0 45
C CAT=WT*COLAM/2*TENS)                                         0 46
C DO 4 M=1,NEND                                                 0 47
C NEXP=NEXP+1                                                  0 48
C SL=SLO*(1+TAU)**NEXP)                                        0 49
C SLX=SL*CAB                                                    0 50
C SLY=SL*SAB                                                    0 51
C XX2=XX1+SLX                                                  0 52
C YY2=YY1+SLY                                                  0 53
C XPRIME=SQRT((XX2-XX1)**2+(YY2-YY1)**2)                       0 54
C IF (ABS(ALFA).LE 1.5) GO TO 2                                  0 55
C ZZ2=ZZ1+SL*SA                                                0 56
C GO TO 3                                                        0 57
2 ZZ2=ZZ2-(RHOXY-XPRIME)*(CAT/XPRIME+ZINC/RHOXY)              0 58
3 XY=SQRT(SLX**2+SLY**2)                                        0 59
C ALPHA=AATAN2(ZZ2-ZZ1),XY)                                    0 60
C X(M)=(XX1+XX2)/2                                             0 61
C Y(M)=(YY1+YY2)/2                                             0 62
C Z(M)=(ZZ1+ZZ2)/2                                             0 63
C S1(M)=SQRT(XY**2+(ZZ2-ZZ1)**2)                                0 64

```

01(M)=M*AD	0 65
ALP(M)=ALPHA	0 66
BET(M)=BETA	0 67
ICON1(M)=M-1	0 68
ICON2(M)=M+1	0 69
XX1=XX2	0 70
YY1=YY2	0 71
ZZ1=ZZ2	0 72
CONTINUE	0 73
RETURN	0 74
END	0 75-

```

SUBROUTINE NEFLD (AIR,A11,BIR,B11,CIR,C11,ZRATI,KSYP)      P 1
C                                                         P 2
C SUBROUTINE NEFLD IS USED TO CALCULATE THE NEAR ELECTRIC FIELD P 3
C AT A SELECTION OF OBSERVATION POINTS. THE INPUTS ARE THE INITIAL P 4
C POINT: X0,Y0,AND Z0 AND THE FINAL POINT X1,Y1 AND Z1. NXY+1 POINTS P 5
C ARE EVALUATED, AND AN INTEGRAL OF THE TANGENTIAL E-FIELD IS EVALUA P 6
C TO GIVE A VOLTAGE DROP ALONG THE PATH. INPUT POSITIONS ARE GIVEN P 7
C IN METERS AND THE FIELD VALUES ARE RETURNED IN VOLTS/METER. P 8
C                                                         P 9
COMMON /1/ N,NP,X(100),Y(100),Z(100),S1(100),B1(100),ALP(100),BET( P 10
100),ICON1(100),ICON2(100),COLAM,NX P 11
COMPLEX ZRATI,REFS,REFPS,ZRSIN P 12
COMMON /3/ CAB(100),SAB(100),SALP(100) P 13
COMMON /REFL/ RHOX,RHOY,RHOZ,CABJ,SABJ,SALPR,PX,PY,REFS,REFPS P 14
DIMENSION AIR(100), A11(100), BIR(100), B11(100), CIR(100), C11(10 P 15
10) P 16
COMPLEX FJ,EZP,ERHO,EX,EY,EZ,EP,SUM,ET P 17
FJ=CMPLX(0.,1.) P 18
PI=3.141592654 P 19
TP=2.*PI P 20
TA=PI/100. P 21
FACTOR=1./COLAM P 22
C                                                         P 23
C READ IN INITIAL AND FINAL POINT COORDINATES---DIM ARE IN METERS P 24
C                                                         P 25
I READ (60,15) X0,Y0,Z0,X1,Y1,Z1,NXY,NFLD P 26
WRITE (61,16) X0,Y0,Z0,X1,Y1,Z1 P 27
WRITE (61,17) P 28
C                                                         P 29
C CALCULATE DIRECTION COSINES FOR OBSERVATION VECTOR P 30
C                                                         P 31
RHOXY=SQRT((X1-X0)**2+(Y1-Y0)**2) P 32
RHOXZ=SQRT(RHOXY**2+(Z1-Z0)**2) P 33
BETA=AATAN2((Y1-Y0),(X1-X0)) P 34
ALPHA=AATAN2((Z1-Z0),RHOXY) P 35
COSALPO=COS(ALPHA) P 36
SALPO=SIN(ALPHA) P 37
CABO=COSALPO*COS(BETA) P 38
SABO=COSALPO*SIN(BETA) P 39
C                                                         P 40
C CONVERT DIMENSIONS FROM METERS TO WAVELENGTHS FOR THE PROGRAM P 41
C                                                         P 42
DX=FACTOR*(X1-X0)/NXY P 43
DY=FACTOR*(Y1-Y0)/NXY P 44
DZ=FACTOR*(Z1-Z0)/NXY P 45
XOB=X0+FACTOR P 46
YOB=Y0+FACTOR P 47
ZOB=Z0+FACTOR P 48
DRHOXZ=RHOXZ/NXY P 49
PATH=0. P 50
SUM=CMPLX(0.,0.) P 51
C                                                         P 52
C MAIN LOOP TO CALC NEAR FIELDS ALONG SPECIFIED PATH P 53
C                                                         P 54
NXY=NXY+1 P 55
DO 14 I=1,NXY P 56
EX=CMPLX(0.,0.) P 57
EY=CMPLX(0.,0.) P 58
EZ=CMPLX(0.,0.) P 59
DO 11 J=1,N P 60
S=S1(J) P 61
B=B1(J) P 62
XJ=X(J) P 63
YJ=Y(J) P 64

```

```
ZJ=Z(J) P 65
CABJ=CAB(J) P 66
SABJ=SAB(J) P 67
SALPJ=SALP(J) P 68
XIJ=XOB-XJ P 69
YIJ=YOB-YJ P 70
RFL=-1. P 71
DO 11 IP=1,KSYMP P 72
RFL=-RFL P 73
ZIJ=ZOB-ZJ*RFL P 74
ZP=XIJ*CABJ+YIJ*SABJ+ZIJ*SALPJ*RFL P 75
RS=XIJ*XIJ+YIJ*YIJ+ZIJ*ZIJ P 76
RM2=RS-ZP P 77
IF (RM2.LT.1.E-20) GO TO 2 P 78
RM=SQRT(RM2) P 79
GO TO 3 P 80
2 RM=0. P 81
3 CONTINUE P 82
SALPR=SALPJ*RFL P 83
RHOX=XIJ-CABJ*ZP P 84
RHOY=YIJ-SABJ*ZP P 85
RHOZ=ZIJ-SALPJ*ZP*RFL P 86
RMAG=SQRT(RHOX*RHOX+RHOY*RHOY+RHOZ*RHOZ) P 87
IF (RMAG.GT.1.E-6) GO TO 4 P 88
RHOX=0. P 89
RHOY=0. P 90
RHOZ=0. P 91
GO TO 5 P 92
4 RHOX=RHOX/RMAG P 93
RHOY=RHOY/RMAG P 94
RHOZ=RHOZ/RMAG P 95
5 RMAG=SQRT(YIJ*YIJ+XIJJ*XIJJ) P 96
IF (IP.NE.2) GO TO 8 P 97
IF (RMAG.GT.1.E-6) GO TO 6 P 98
PX=0. P 99
PY=0. P 100
CTH=1. P 101
ZRSIN=CMPLX(1.,0.) P 102
GO TO 7 P 103
6 PX=YIJ/RMAG P 104
PY=-XIJJ/RMAG P 105
CTH=ZIJ/SQRT(RS) P 106
ZRSIN=CSQRT(1.-ZRATI*ZRATI*(1.-CTH*CTH)) P 107
REFS=(CTH-ZRATI*ZRSIN)/(CTH+ZRATI*ZRSIN) P 108
REFPS=-(ZRATI*CTH-ZRSIN)/(ZRATI*CTH+ZRSIN) P 109
REFPS=REFPS-REFS P 110
8 CONTINUE P 111
CALL EFLD (B,S,RM,ZP,1,EZRS,EZIS,ERRS,ERIS,EZRC,EZIC,ERRC,ERIC,EZR P 112
IK,EZIK,ERRK,ERIK) P 113
IF (IP.NE.2) GO TO 9 P 114
CALL GN (EZRS,EZIS,ERRS,ERIS) P 115
CALL GN (EZRC,EZIC,ERRC,ERIC) P 116
CALL GN (EZRK,EZIK,ERRK,ERIK) P 117
9 EZP=EZRK*AIR(J)-EZIK*AI(IJ)+EZRS*BIR(J)-EZIS*BII(J)+EZRC*CI(RJ)-EZ P 118
IC*CI(IJ)+EJ*(EZRK*AI(IJ)+EZIK*AIR(J)+EZRS*BII(J)+EZIS*BIR(J)+EZRC P 119
2*CI(IJ)+EZIC*CI(RJ)) P 120
ERHO=ERRK*AIR(J)-ERIK*AI(IJ)+ERRS*BIR(J)-ERIS*BII(J)+ERRC*CI(RJ)-E P 121
IRIC*CI(IJ)+EJ*(ERRK*AI(IJ)+ERIK*AIR(J)+ERRS*BII(J)+ERIS*BIR(J)+ERR P 122
2*CI(IJ)+ERIC*CI(RJ)) P 123
IF (IP.NE.2) GO TO 10 P 124
EZP=-EZP P 125
ERHO=-ERHO P 126
10 EX=EX+EZP*CABJ+ERHO*RHOX P 127
EY=EY+EZP*SABJ+ERHO*RHOY P 128
```

```

EZ=EZ+EZP*SALPJ+RFL*ERHO+RHOZ
11 CONTINUE
C
C EX,EY AND EZ ARE THE COMPLEX E-FIELDS IN THE CARTESIAN
C COORDINATE DIRECTIONS AT THE OBSERVATION POINT.
C
ETOTAL=SQRT(CABS(EX)**2+CABS(EY)**2+CABS(EZ)**2)/COLAM
ET=(EX*CABO+EY*SABO+EZ*SALPO)/COLAM
ETANG=CABS(ET)
IF (I.EQ.1) GO TO 12
IF (I.EQ.NXY) GO TO 12
SUM=SUM+ET
GO TO 13
12 SUM=SUM+ET/2.
13 CONTINUE
XOB=XOB+DX
YOB=YOB+DY
ZOB=ZOB+DZ
WRITE (61,10) PATH,ETANG,ETOTAL
PATH=PATH+DRHOXYZ
14 CONTINUE
C
C THE TANGENTIAL E-FIELD IS INTEGRATED VIA THE TRAPEZOIDAL
C RULE TO COMPUTE THE VOLTAGE DROP ALONG THE SPECIFIED PATH
C
VDROP=CABS(SUM)*DRHOXYZ
WRITE (61,19) VDROP,NXY
IF (INFLD.NE.0) GO TO 1
RETURN
C
15 FORMAT (6F10.4,2I5)
16 FORMAT (16H) E-FIELD FROM .3F10.4,11H METERS TO .3F10.4,7H METERS
1)
17 FORMAT (//,46H POSITION ON PATH (METERS) E-TANGENT (V/M),24H
1 TOTAL E-FIELD (V/M),//)
18 FORMAT (3E20.5)
19 FORMAT (//,50H THE INTEGRAL OF THE E-FLD TANGENT TO THE PATH IS .E
115.5,7H VOLTS..15,34H POINTS USED TO EVALUATE INTEGRAL.,//)
END
P 129
P 130
P 131
P 132
P 133
P 134
P 135
P 136
P 137
P 138
P 139
P 140
P 141
P 142
P 143
P 144
P 145
P 146
P 147
P 148
P 149
P 150
P 151
P 152
P 153
P 154
P 155
P 156
P 157
P 158
P 159
P 160
P 161
P 162
P 163
P 164
P 165
P 166
P 167
P 168
```



```

SUBROUTINE SOLVE (N,A,P,B,NDIM)
C
C SUBROUTINE TO SOLVE THE MATRIX EQUATION LU*X=B WHERE L IS A UNIT L
C TRIANGULAR MATRIX AND U IS AN UPPER TRIANGULAR MATRIX BOTH OF WHICH
C IN A. THE RHS VECTOR B IS INPUT AND THE SOLUTION IS RETURNED THRO
C
C COMPLEX A,B,Y,SUM
C INTEGER P,PI
C DIMENSION A(NDIM,NDIM), P(NDIM), B(NDIM)
C COMMON /SCRATCH/ Y(100)
C
C FORWARD SUBSTITUTION
C
C DO 3 I=1,N
C PI=P(I)
C Y(I)=B(PI)
C B(PI)=B(I)
C IP1=I+1
C IF (IP1.GT.N) GO TO 2
C DO 1 J=IP1,N
C B(J)=B(J)-A(I,J)*Y(I)
1 CONTINUE
2 CONTINUE
3 CONTINUE
C
C BACKWARD SUBSTITUTION
C
C DO 6 K=1,N
C I=N-K+1
C SUM=CHPLX(0.,0.)
C IP1=I+1
C IF (IP1.GT.N) GO TO 5
C DO 4 J=IP1,N
C SUM=SUM+A(I,J)*B(J)
4 CONTINUE
5 CONTINUE
C B(I)=(Y(I)-SUM)/A(I,I)
6 CONTINUE
C RETURN
C END
O 1
O 2
O 3
O 4
O 5
O 6
O 7
O 8
O 9
O 10
O 11
O 12
O 13
O 14
O 15
O 16
O 17
O 18
O 19
O 20
O 21
O 22
O 23
O 24
O 25
O 26
O 27
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O 31
O 32
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O 34
O 35
O 36
O 37
O 38
O 39
O 40-
```


	SUBROUTINE SOLVES (N,NOP,A,P,B,NROW,NCOL)	S	1
C		S	2
C	SUBROUTINE SOLVES SOLVES ROTATIONALLY SYMMETRIC MATRICIES	S	3
	COMPLEX A,B,Y,SUM,S	S	4
	INTEGER P	S	5
	COMMON /SMAT/ S(12,12)	S	6
	DIMENSION A(NROW,NCOL), P(NCOL), B(NCOL)	S	7
	COMMON /SCRATH/ Y(100)	S	8
	IF (NOP.EQ.1) GO TO 5	S	9
	FNORM=1./NOP	S	10
	DO 4 I=1,N	S	11
	DO 1 K=1,NOP	S	12
	IA=(K-1)*N	S	13
1	Y(K)=B(IA)	S	14
	SUM=Y(I)	S	15
	DO 2 K=2,NOP	S	16
2	SUM=SUM+Y(K)	S	17
	B(I)=SUM*FNORM	S	18
	DO 4 K=2,NOP	S	19
	IA=(K-1)*N	S	20
	SUM=Y(I)	S	21
	DO 3 J=2,NOP	S	22
3	SUM=SUM+Y(J)*CONJG(S(K,J))	S	23
4	B(IA)=SUM*FNORM	S	24
5	DO 6 KK=1,NOP	S	25
	IA=(KK-1)*N+1	S	26
	CALL SOLVE (N,A(I,IA),P(IA),B(IA),NROW)	S	27
6	CONTINUE	S	28
	IF (NOP.EQ.1) RETURN	S	29
	DO 10 I=1,N	S	30
	DO 7 K=1,NOP	S	31
	IA=(K-1)*N	S	32
7	Y(K)=B(IA)	S	33
	SUM=Y(I)	S	34
	DO 8 K=2,NOP	S	35
8	SUM=SUM+Y(K)	S	36
	B(I)=SUM	S	37
	DO 10 K=2,NOP	S	38
	IA=(K-1)*N	S	39
	SUM=Y(I)	S	40
	DO 9 J=2,NOP	S	41
9	SUM=SUM+Y(J)*S(K,J)	S	42
10	B(IA)=SUM	S	43
	RETURN	S	44
	END	S	45
		S	46-


```

SUBROUTINE TRIO (J,JCO1,JCO2,DIL,DIK)
C
C SUBROUTINE TRIO IS USED TO DETERMINE THE TYPE OF JUNCTION USED AT
C THE SEGMENT ENDS. THREE TYPES OF SEGMENT END JUNCTIONS ARE ALLOW
C --IF ICON IS 0, SEGMENT END IS OPEN--IF ICON EQUALS SEGMENT NUMBER
C THEN SEG END IS GROUNDED--IF ICON IS NEG THEN A CHECK IS MADE FOR
C MULTIPLE JUNCTIONS. TRIO RETURNS AN EQUIVALENT DISTANCE (DIL,DIK)
C WHICH IS USED FOR INTERPOLATING CURRENTS FROM 1 SEG TO THE NEXT.
C
C INPUTS:
C J = SEGMENT NUMBER TO BE CHECKED
C
C OUTPUTS:
C JCO1 = ICON VALUE OF J-TH SEG NEG END
C JCO2 = ICON VALUE OF J-TH SEG POS END
C DIL = AVG DISTANCE FOR CURRENT INTERPOLATION ON SEG NEG END
C DIK = AVG DISTANCE FOR CURRENT INTERPOLATION ON SEG POS END
C
C
C COMMON /1/ N,NP,X(100),Y(100),Z(100),S1(100),B1(100),ALP(100),BET(
100),ICON1(100),ICON2(100),COLAM,NX
C COMMON /4/ NCOX,JOX(25),NCIX,JIX(25),NCOZ,JOZ(25),NCIZ,JIZ(25)
C S=S1(J)
C JCO1=ICON1(J)
C JCO2=ICON2(J)
C
C *****CHECK SEGMENT NEGATIVE END*****
C
C IF (JCO1) 1,2,3
C
C MULTIPLE JUNCTION
C
1 CALL JUNC (J,JCO1,NCOX,JOX,NCIX,JIX,DIL)
GO TO 4
C
C OPEN CIRCUIT--FREE END -- INTERPOLATE ONLY TO END OF SEGMENT
C
2 DIL=S/2.0
GO TO 4
C
C NORMAL JUNCTION CONNECTION--SIMPLE JUNCTION OR GROUNDED SEGMENT
C
3 DIL=(S1(JCO1)+S)/2.0
C
C *****CHECK SEGMENT POSITIVE END*****
C
4 IF (JCO2) 5,6,7
C
C MULTIPLE JUNCTION
C
5 CALL JUNC (J,JCO2,NCOZ,JOZ,NCIZ,JIZ,DIK)
GO TO 8
C
C OPEN CIRCUIT--FREE END
C
6 DIK=S/2.0
GO TO 8
C
C NORMAL JUNCTION CONNECTION--SIMPLE JUNCTION OR GROUNDED SEGMENT
C
7 DIK=(S1(JCO2)+S)/2.0
8 CONTINUE
RETURN
END

```

APPENDIX B TABLE I SEGMENT EXPANSION COEFFICIENTS (TAU)

L/L0	NUMBER OF SEGMENTS--N																		
	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20
2	0	.39	.46	.49	.50	.50	.50	.50	.50	.50	.50	.50	.50	.50	.50	.50	.50	.50	.50
3	1.00	0	.19	.26	.30	.31	.32	.33	.33	.33	.34	.34	.34	.34	.34	.34	.34	.34	.34
4	2.00	.30	0	.12	.17	.20	.22	.23	.24	.24	.25	.25	.25	.25	.25	.25	.25	.25	.25
5	3.00	.56	.15	0	.08	.12	.15	.16	.17	.18	.19	.19	.19	.20	.20	.20	.20	.20	.20
6	4.00	.79	.28	.09	0	.06	.09	.11	.13	.14	.14	.15	.15	.16	.16	.16	.16	.17	.17
7	5.00	1.00	.39	.17	.06	0	.04	.07	.09	.10	.11	.12	.12	.13	.13	.14	.14	.14	.14
8	6.00	1.19	.49	.24	.11	.04	0	.03	.06	.07	.08	.09	.10	.10	.11	.11	.11	.12	.12
9	7.00	1.37	.56	.30	.16	.08	.03	0	.03	.05	.06	.07	.08	.08	.09	.09	.10	.10	.10
10	8.00	1.54	.66	.35	.20	.12	.06	.03	0	.02	.04	.05	.06	.07	.07	.08	.08	.09	.09
11	9.00	1.70	.74	.40	.24	.15	.09	.05	.02	0	.02	.03	.04	.05	.06	.06	.07	.07	.07
12	10.00	1.85	.81	.45	.28	.18	.11	.07	.04	.02	0	.02	.03	.04	.05	.06	.06	.07	.07
13	11.00	2.00	.88	.49	.31	.20	.14	.09	.06	.03	.01	0	.02	.03	.04	.05	.06	.06	.06
14	12.00	2.14	.94	.53	.34	.23	.16	.11	.07	.05	.03	.01	0	.02	.03	.04	.05	.05	.05
15	13.00	2.27	1.00	.57	.37	.25	.17	.12	.08	.06	.04	.02	0	.01	.02	.03	.04	.04	.04
16	14.00	2.41	1.06	.61	.39	.27	.19	.14	.10	.07	.05	.03	.01	0	.01	.02	.03	.03	.04
17	15.00	2.53	1.11	.64	.42	.29	.21	.15	.11	.08	.06	.04	.03	.02	0	.01	.02	.02	.03
18	16.00	2.65	1.16	.67	.44	.31	.22	.17	.13	.09	.07	.05	.04	.03	.02	0	.01	.02	.02
19	17.00	2.77	1.22	.70	.46	.33	.24	.18	.14	.10	.08	.06	.05	.03	.02	.01	0	.01	.02
20	18.00	2.89	1.26	.73	.48	.34	.25	.19	.15	.11	.09	.07	.06	.05	.03	.02	.01	0	.01
21	19.00	3.00	1.31	.76	.50	.36	.27	.20	.16	.12	.10	.08	.07	.05	.04	.03	.02	.01	0
22	20.00	3.11	1.36	.79	.52	.37	.28	.21	.17	.13	.10	.08	.07	.05	.04	.03	.02	.01	.01
23	21.00	3.22	1.40	.82	.54	.39	.29	.22	.18	.14	.11	.09	.07	.06	.05	.04	.03	.02	.01
24	22.00	3.32	1.44	.84	.56	.40	.30	.23	.19	.15	.12	.10	.08	.06	.05	.04	.03	.02	.01
25	23.00	3.42	1.48	.87	.58	.42	.31	.24	.19	.15	.12	.10	.08	.06	.05	.04	.03	.02	.01
26	24.00	3.52	1.53	.89	.60	.43	.32	.25	.20	.16	.13	.10	.08	.07	.06	.05	.04	.03	.02
27	25.00	3.62	1.56	.91	.61	.44	.33	.26	.21	.17	.14	.11	.09	.07	.06	.05	.04	.03	.02
28	26.00	3.72	1.60	.94	.63	.45	.34	.27	.22	.18	.14	.11	.10	.08	.07	.06	.05	.04	.03
29	27.00	3.82	1.64	.96	.64	.47	.35	.28	.22	.18	.15	.12	.10	.08	.07	.06	.05	.04	.03
30	28.00	3.91	1.68	.98	.66	.48	.36	.29	.23	.19	.15	.13	.10	.09	.07	.06	.05	.04	.04
31	29.00	4.00	1.71	1.00	.67	.49	.37	.29	.24	.19	.16	.13	.11	.09	.08	.07	.06	.05	.04
32	30.00	4.09	1.75	1.02	.69	.50	.38	.30	.24	.20	.17	.14	.12	.10	.09	.07	.06	.05	.05
33	31.00	4.18	1.78	1.04	.70	.51	.39	.31	.25	.20	.17	.14	.12	.10	.09	.07	.06	.05	.05
34	32.00	4.27	1.81	1.06	.71	.52	.40	.31	.25	.21	.18	.15	.12	.10	.09	.08	.07	.06	.05
35	33.00	4.35	1.85	1.08	.73	.53	.40	.32	.26	.22	.18	.15	.13	.11	.09	.08	.07	.06	.05
36	34.00	4.44	1.88	1.10	.74	.54	.41	.33	.27	.22	.18	.16	.13	.11	.10	.08	.07	.06	.06
37	35.00	4.52	1.91	1.11	.75	.55	.42	.33	.27	.23	.19	.16	.14	.12	.10	.09	.08	.07	.06
38	36.00	4.60	1.94	1.13	.76	.56	.43	.34	.28	.23	.19	.16	.14	.12	.10	.09	.08	.07	.06
39	37.00	4.68	1.97	1.15	.77	.57	.43	.35	.28	.23	.20	.17	.14	.12	.11	.09	.08	.07	.06
40	38.00	4.76	2.00	1.17	.79	.57	.44	.35	.29	.24	.20	.17	.15	.13	.11	.10	.09	.08	.07
41	39.00	4.84	2.03	1.18	.80	.58	.45	.36	.29	.24	.21	.18	.15	.13	.12	.10	.09	.08	.07
42	40.00	4.92	2.06	1.20	.81	.59	.46	.36	.30	.25	.21	.18	.15	.13	.12	.10	.09	.08	.07
43	41.00	5.00	2.09	1.21	.82	.60	.46	.37	.30	.25	.21	.18	.15	.13	.12	.10	.09	.08	.07
44	42.00	5.08	2.11	1.23	.83	.61	.47	.37	.31	.26	.22	.18	.16	.14	.12	.11	.09	.08	.07
45	43.00	5.15	2.14	1.24	.84	.62	.47	.38	.31	.26	.22	.19	.16	.14	.12	.11	.10	.09	.08
46	44.00	5.23	2.17	1.26	.85	.62	.48	.38	.32	.26	.22	.19	.16	.14	.13	.11	.10	.09	.08
47	45.00	5.30	2.19	1.27	.86	.63	.49	.39	.32	.27	.23	.20	.17	.15	.13	.12	.10	.09	.08
48	46.00	5.37	2.22	1.29	.87	.64	.49	.39	.32	.27	.23	.20	.17	.15	.13	.12	.11	.09	.08
49	47.00	5.45	2.25	1.30	.88	.65	.50	.40	.33	.28	.23	.20	.18	.15	.14	.12	.11	.10	.09
50	48.00	5.52	2.27	1.32	.89	.65	.50	.40	.33	.28	.24	.20	.18	.16	.14	.12	.11	.10	.09