

PRACTICAL

ELECTRONICS

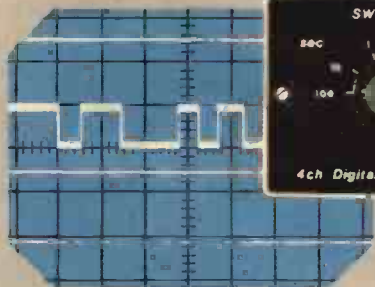
FEBRUARY 1980

55p

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- High quality thru plated P.C.B. with all I.C.'s mounted on sockets
- Professional 52 Key keyboard in 3 colours — software polled meaning that all debouncing and key decoding done in software

COMMANDS
CONT LIST NEW NULL RUN
STATEMENTS
CLEAR DATA DEF DIM END FOR
GOTO GOSUB IF GOTO IF THEN INPUT LET
NEXT ON GOTO ON GOSUB POKE PRINT READ
REM RESTORE RETURN STOP

EXPRESSIONS
OPERATORS
+ * / ↑ NOT AND OR > < <> >=< RANGE 10⁻³² to 10⁺³²

VARIABLES
A,B,C Z and two letter variables
The above can all be subscripted when used in an array. String variables use above names plus \$ e.g AS



*8K Microsoft Basic means conversion to and from Pet, Apple and Sorcerer easy. Many compatible programs already in print.
SPECIAL CHARACTERS
@ Erases line being typed, then provides carriage return, line feed.
Erases last character typed.
CR Carriage Return — must be at the end of each line.
: Separates statements on a line.
CONTROL/C Execution or printing of a list is interrupted at the end of a line.
"BREAK IN LINE XXXX" Is printed, indicating line number of next statement to be executed or printed.
CONTROL/O No outputs occur until return made to command mode. If an Input statement is encountered, either another CONTROL/O is typed, or an error occurs.
? Equivalent to PRINT

Simple Soldering due to clear and concise instructions compiled by Dr. A.A. Berk, BSc.PhD

NO EXTRAS NEEDED JUST HIT 'RETURN' AND GO.

Build, understand, and program your own computer for only a small outlay.

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FUNCTIONS
ABS(X) ATN(X) COS(X) EXP(X)
LOG(X) PEEK(I) POS(I) RND(X)
SPC(I) SOR(X) TAB(I) TAN(X)
FRE(X) INT(X)
SGN(X) SIN(X)
USR(I)
STRING FUNCTIONS
ASC(X\$) CHR\$(I) FRE(X\$) LEFT\$(X\$,I)
RIGHT\$(X\$,I) STR\$(X)
LEN(X\$) MID\$(X\$,I,J)
VAL(X\$)

COLOUR ADD-ON CARD AVAILABLE SOON

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VOLUME 16 No. 2 FEBRUARY 1980

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OUR MARCH ISSUE WILL BE ON SALE FRIDAY, 8 FEBRUARY 1980
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(Continued from opposite side)

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BOXES 3x2x1" 55 2 1/2 x 5 1/2 75 4x4x1 1/2 75 4x2 1/2 x 1 1/2 75 4x5 1/2 x 1 1/2 100 4x4 1/2 x 2 1/2 75 5x4x2" 98 6x4x2" 110 7x5x2 1/2 185 8x6x3" 185 10x7x3" 210 10x4 1/2 x 3" 198 12x5x3" 215 12x8x3" 265	ZENERS Rng. 2V7-39V 400mW 8p Rng. 3V3-33V 1.3W 15p	VEROBOARD 0-1 Pitch (copper clad) 0-15 (plain) 0-1 31p 0-15 24p	VARIACAPS MVAM115140 BA102 25 BB104 40 BB105B 40 BB106 40 3x1 1/2" 218p 4x1 1/2" 280p	Noise Diode Z5J 160	TRIACS 3A/100V 48 3A/400V 50 8A/100V 54 8A/400V 64 8A/500V 85 8A/800V 108 12A/100V 80 12A/400V 70 12A/800V 130 16A/100V 95 16A/400V 105 25A/400V 160 25A/800V 295 TZ800D 120	FERRIC CHLORIDE 1 lb bag Anhydrous 95p + 35p P&P	EURO BREADBOARD £5.30	DALO ETCH RESIST Pen + Spare tip 95p	BIMBOARDS £8.58	COPPER CLAD BOARDS Fibre Single-Double- Glass sided sided SRBP 6" x 6" 75p 90p 80p 6" x 12" 130p 175p	SOLDERING PINS 100 pins 50p: 500 pins 200p	OPTO ELECTRONICS LED Plus, Clip TIL209 Red 125" 13 TIL211 Gm 125" 18 TIL212 Yellow 22 TIL32 Infra Red 58 TIL2200.2" Red 15 0.2" Yellow, green and Amber 18 Rectangular LEDs, Red, Green and Yellow 36 OCPT1 120 ORP12 63 ZNS577 45	OPTO Isolators IL74 48 TIL111/2 85 TIL117 110	7 Segment Displays LS400 255 TIL307 675 TIL312 & 313.3" 105 TIL321 .5" C.An 115 TIL322 .5" C.th 115 DL704 .3" C.Ct.h 99 DL707 .3" C.Anod 99 DL747 .6" An 180 FND357 120 MAN3640 165 XAN351 .3" Green 180	Liquid Crystal Display 3 1/2 digit 875p; 4 digit 975p	VOLTAGE REGULATORS 1A TO3 +ve -ve 5V 7805 145p 7905 220p 12V 7812 145p 7912 220p 15V 7815 145p 18V 7818 145p 1A TO220 Plastic Casing 5V 7805 65p 7905 75p 12V 7812 65p 7912 75p 15V 7815 65p 7915 75p 18V 7818 65p 7918 75p 24V 7824 65p 100mA TO92 Plastic Casing 5V 78L05 30p 79L05 65p 6V 78L62 30p 8V 78L82 30p 12V 78L12 30p 79L12 65p 15V 78L15 30p 79L15 65p CA3085 95 LM323K 625 LM300H 170 LM325N 240 LM305H 140 LM326N 240 LM309K 135 LM327N 270 LM317K 350 LM723 39 78H05 + 5V/5A 595p; 78H6 + 5 to +24V 650p	SWITCHES SLIDE 250V 1A DPDT 14 1A DPDT C/OFF 15 4 pole 2-way 13 4 pole 2-way 24 PUSH BUTTON Spring loaded Latching SPST on off 60 SPDT C/Over 65 DPDT 6 Tag 85 MINIATURE Non Locking Push to make 215 Push Break 25 ROCKER: 5A 250V SPST 23 ROCKER: (white) 5A 250V SP change-over centre off 40 ROCKER: Lights red when on, Chrome Bezel, 3A 250V, DPST 85p ROTARY: "Make-A-Switch" Make your own multiway Switch, Adjustable Stop Shafting Assembly, Accommodates up to 6 Wafers 75 Mains Switch DPST to fit 34 Break Before Make Wafers, 1 pole/12 way, 2p/6 way, 3p/4 way, 4p/3 way, 6p/2 way 47 Spacer and Screen 5 ROTARY: (Adjustable Stop) 1 pole/2 to 12 way, 2p/2 to 6 way, 3 pole/2 to 4 way, 4 pole/2 to 3 way 41 ROTARY: Mains 250V AC, 4 Amp 45	TOGGLE 2A 250V SPST 28 DPDT 34 4 pole on off 54 SUB-MIN TOGGLE SP changeover 59 SPST on off 54 SPST biased 95 DPDT 6 tags 70 DPDT C/OFF 79 DPDT Biased 115 3 pole c/over 150	'D' CONNECTORS (Cannon type) (Special introductory prices) Plug Sockets Covers (Plastic) 9 way 70p 90p 15 way 95p 125p 114p 25 way 140p 198p 128p 37 way 210p 295p 135p	EDGE CONNECTORS double type DIL switches 2x10 way 85p 2x15 way 99p 2x18 way 115p 120p (SPST) 2x22 way 130p 135p 4 way 2x25 way 149p 160p 2x30 way 170p 120p 2x36 way 194p 2x40 way 210p 160p 8 way 2x43 way 232p 180p	CRYSTALS 6.5536MHz 200p 100kHz 385p 455kHz 385p 1MHz 323p 1.008MHz 323p 1.80MHz 385p 1.832MHz 362p 1.6MHz 395p 3.2768MHz 323p 4MHz 290p 4.032MHz 323p 4.433619M 135p 5.0MHz 355p 6.5536MHz 323p 9.375MHz 323p 10MHz 323p 10.7MHz 323p 12MHz 392p 14.318118M 300p 18MHz 323p 18.432M 33p 20MHz 362p 27.648MHz 350p 48MHz 323p 100MHz 300p	Power Supply Kit (5V/3A) incl. RF Modulator £21.50 Readybuilt and tested £28.50 Extra 4k of RAM £35.00 UHF Modulator 250p Wide Bandwidth Modulator Special for Computers 470p Power Supplies, Readybuilt: ±8V ±0.5A £14.95 ±12V ±0.25A £14.95 ±15V ±0.2A £14.95 KEYPADS, 4 x 4 matrix, Reed switch assembly 350p Full ASCII Keyboard 350p Model 756 £45.50 10" MONITOR, suitable for alphanumeric or analogue signals. 220V mains or 12V DC. Compatible to most Computer System. Value for money £85.00	TRANSFORMERS (mains Prim. 220-240V) 6-0.6V 100mA; 9-0.9V 75mA; 12-0.12V 75mA 95p 8VA type: 6V-5A 6V-5A; 9V-4A 9V-4A; 12V-3A 12V-3A; 15V-2.5A 15V-2.5A 195p 12VA: 4.5-1.3A 4.5V-1.3A; 6V-1.2A 6V-1.2A 12V-5A 12V-5A 220p (20p p&p) 24VA: 6V-1.5A 6V-1.5A; 9V-1.2A 9V-1.2A; 12V-1A 12V-1A; 15-8A 15-8A; 20V-6A 20V-6A 290p (45p p&p) 50VA: 6V-4A 6V-4A; 9V-2.5A 9V-2.5A; 12V-2A 12V-2A; 15V-1.5A 15V-1.5A; 20V-1.2A 20V-1.2A; 25V-1A 25V-1A; 30V-.8A 30V-.8A 350p (50p p&p) 100VA: 12V-4A 12V-4A; 15V-3A 15V-3A; 20V-2.5A 20V-2.5A; 30V-1.5A 30V-1.5A; 40V-1.25A 40V-1.25A; 50V-1A 50V-1A 650p (60p p&p) (N.8. P&P charge to be added above our normal postal charge.)
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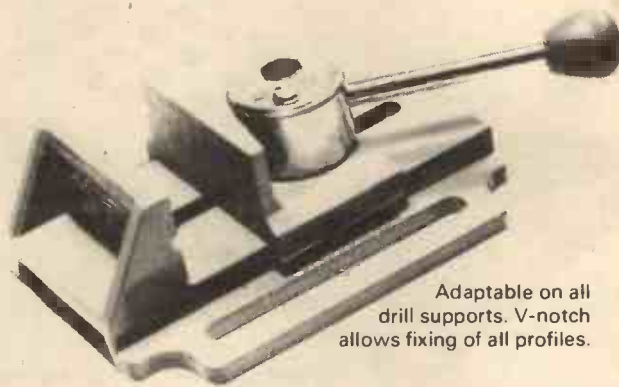
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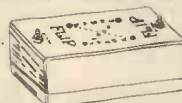
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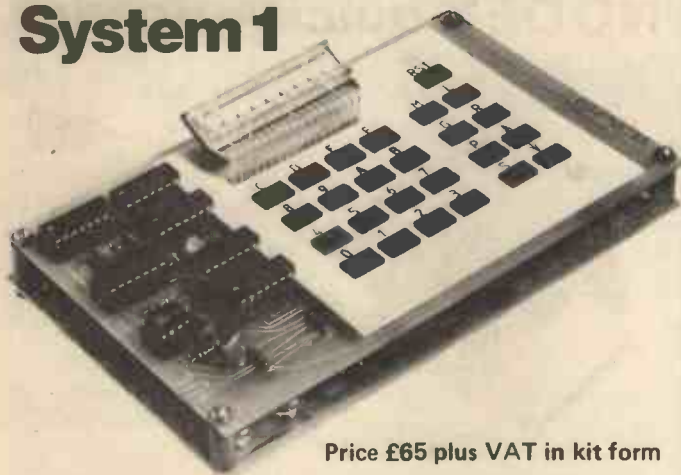
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Acorn Microcomputer System 1



Price £65 plus VAT in kit form

This compact stand-alone microcomputer is based on standard Eurocard modules, and employs the highly popular 6502 MPU (as used in APPLE, PET, KIM, etc). Throughout, the design philosophy has been to provide full expandability, versatility and economy.

Specification

The Acorn consists of two single Eurocards.

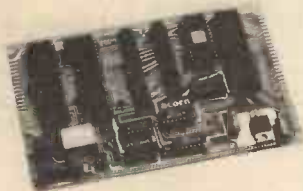
1. MPU card
 - 6502 microprocessor
 - 512 x 8 ACORN monitor
 - 1 K x 8 RAM
 - 16-way I/O with 128 bytes of RAM
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 - Stepping up through memory
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 - Restore from break
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 - Store on tape
 - Go (recalls last address used)
 - Reset
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- System program
 - Set of sub-routines for use in programming
 - Powerful de-bugging facility displays all internal registers
 - Tape load and store routines

Applications

- As a self teaching tool for beginners to computing.
- As a low cost 6502 development system for industry.
- As a basis for a powerful microcomputer in its expanded form.
- As a control system for electronics engineers.
- As a data acquisition system for laboratories.

START WITH SYSTEM 1 AND CONTINUE AS AND WHEN YOU LIKE



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£35 plus VAT (min config.)

the CPU card of System 1, it allows for up to 4½ k EPROM, 1¼ k RAM and 32 I/O lines. It has on board 5 V regulator and optional crystal control. Custom programs may be developed on System 1 and the card makes an ideal dedicated hardware module.



Acorn Memory 8 k
£95 plus VAT (kit form)

A fully buffered memory card allowing up to 8 k RAM plus 8 k EPROM on one eurocard, in an Acorn system both BASIC and DOS may be contained in this module. Static RAM (2114) is used and the card may be wired into other systems.



Acorn VDU
£88 plus VAT (kit form)

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Acorn Software
in ROM

- Acorn BASIC — a very fast integer BASIC in 4 k
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KITS FOR SYNTHESISERS, SOUND EFFECTS



COMPONENTS SETS include all necessary resistors, capacitors, semiconductors, potentiometers and transformers. Hardware such as cases, sockets, knobs, keyboards, etc. are not included but most of these may be bought separately. Fuller details of kits, PCBs and parts are shown in our lists.

LAYOUT DIAGRAMS are supplied free with all PCBs unless "as published".

PHONOSONICS

MAIL ORDER SUPPLIERS OF QUALITY PRINTED CIRCUIT BOARDS, KITS AND COMPONENTS TO A WORLD-WIDE MARKET

P.E. 128-NOTE PROGRAMMABLE SEQUENCER

Enables a voltage controlled synthesiser to automatically play pre-programmed tunes of up to 32 pitches and 128 notes long. Programs are keyboard initiated and note length and rhythmic pattern are externally variable.

Set of basic component kits	KIT 76-5	£28.92
Set of PCBs & layout charts	KIT 76-6	£5.66
Set of text photocopies		£1.36

P.E. 16-NOTE PROGRAMMABLE SEQUENCER

Sequences of up to 16 notes may be programmed by the use of external panel controls and fed into most voltage controlled synthesisers.

Set of basic component kits	KIT 86-3	£22.90
Set of PCBs	KIT 86-4	£5.09
Set text photocopies		£1.84

P.E. STRING ENSEMBLE

A multivoiced string instrument synthesiser.

Set of basic component kits	KIT 77-6	£88.70
Set of PCBs & layout charts	KIT 77-7	£24.19

P.E. JOANNA PLUS ORGAN VOICING

A modified version of the P.E. 5-octave piano that retains all the original facilities and also includes switchable organ voicing circuitry.

Set of basic component kits	KIT 71-5	£90.36
Set of PCBs & layout charts	KIT 71-6	£29.51
"Sound Design" booklet		£1.00

ELEKTOR ELECTRONIC PIANO

A touch-sensitive multiple-voicing piano using the latest integrated circuit techniques for the keying and envelope shaping, and virtually eliminating "bee-hive" noise hitherto inherent in previous electronic pianos.

5-octave set of basic components	KIT 80-6	£110.39
5-octave set of PCBs (as published)	KIT 80-7	£26.02
Additional 3-octave extension basic parts	KIT 80-5	£45.05
Additional 3-octave set of PCBs (as published)	KIT 80-8	£9.55
Set of text photocopies		£1.81

P.E. MINISONIC MK2 SYNTHESISER

A portable mains operated miniature sound synthesiser with keyboard circuits. Although having slightly fewer facilities than the large Formant and P.E. synthesisers the functions offered by this design give it great scope and versatility.

Set of basic component kits (excl. KBD R's & tuning pots - see list for options available)	KIT 38-23	£87.05
Set of PCBs (incl. layout charts)	KIT 38-24	£9.87
"Sound Design" booklet		£1.00

P.E. SYNTHESISER

The well acclaimed and highly versatile large scale mains operated synthesiser. Other circuits in our lists may be used with it to good advantage.

Main Unit basic component kits	KIT 23-27	£88.95
Main Unit set of PCBs & layout charts	KIT 23-28	£14.48
Keyboard Unit basic component kits	KIT 23-29	£52.06
Keyboard Unit set of PCBs & layout charts	KIT 23-30	£8.41
Main Unit set of text photocopies		£5.91
Keyboard Unit set of text photocopies		£2.30

ELEKTOR FORMANT SYNTHESISER

A very sophisticated synthesiser for the advanced constructor who puts performance before price.

Set of basic component kits	KIT 66-12	£193.88
Set of PCBs (as published)	KIT 66-13	£53.92
Set of text photocopies		£7.83

P.E. GUITAR EFFECTS PEDAL

Modulates the attack, decay and filter characteristics of a signal from most audio sources, producing 8 different switchable effects that can be further modified by manual controls.

Basic parts with foot switches	KIT 42-1	£8.45
Basic parts with panel switches	KIT 42-2	£5.56
PCB & layout chart	PCB 42A	£1.67
Text photocopy		28p

ELEKTOR DIGITAL REVERB UNIT

A very advanced unit using sophisticated i.c. techniques instead of mechanical spring lines. The basic delay range of 24 to 90mS can be extended up to 450mS using the extension unit. Further delays can be obtained using more extensions.

Main unit basic component kit	KIT 78-1	£49.99
Main unit PCB (as published)	PCB 9913	£3.69
Extension unit basic component kit	KIT 78-2	£47.69
Extension unit PCB (as published)	PCB 78B	£1.16
Text photocopy		

ELEKTOR ANALOGUE REVERB UNIT

Using i.c.s instead of spring-lines the main unit has a maximum delay of up to 100mS, and the additional set extends this up to 200mS. May be used in either mono or stereo mode.

Main unit basic component set	KIT 83-1	£29.49
Additional Delay basic components	KIT 83-2	£20.07
PCB (as publ.) to hold both kits	PCB 9973	£4.31
Text photocopy		

P.E. GUITAR MULTIPROCESSOR

An extremely versatile sound processing unit capable of producing, for example, flanging, vibrato, reverb, fuzz and tremolo as well as other fascinating sounds. May be used with most electronic instruments.

Set of basic component kits	KIT 85-3	£43.76
Set of PCBs & layout charts	KIT 85-4	£10.62
Set of text photocopies		£2.52

P.E. PHASER

An automatically controlled 6-stage phasing unit with integral oscillator.

Set of basic components, incl. PCB & chart	KIT 88-1	£10.14
Text photocopy		68p

ELEKTOR PHASING & VIBRATO UNIT

Includes manual and automatic control over the rate of phasing & vibrato, and has been slightly modified to also include a 2-input mixer stage.

Set of basic components	KIT 70-1	£19.11
PCB & layout chart	PCB 70A	£2.56
Text photocopy		87p

P.E. PHASING UNIT

A simple but effective manually controlled phasing unit.

Set of basic components, incl. PCB & chart	KIT 25-1	£3.62
Text photocopy		28p

PHASING CONTROL UNIT

For use with Phasing Kit 25 to automatically control rate of phasing.

Set of basic components, incl. PCB & chart	KIT 36-1	£5.21
Text photocopy		10p

P.E. SWITCHED TONE TREBLE BOOST

Provides switched selection of 4 preset tonal responses.

Set of basic components, PCB & chart	KIT 89-1	£3.82
Text photocopy		78p

P.E. TREBLE BOOST UNIT

A simple treble boost unit with manual control of depth.

Set of basic components, PCB & chart	KIT 53-1	£2.76
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ELEKTOR RESONANCE FILTER

Allows a synthesiser to produce a more realistic simulation of natural musical instruments.

Set of basic components	KIT 82-1	£16.61
PCB (as published)	PCB 9951	£3.29
Text photocopy		

P.E. GUITAR OVERDRIVE

Sophisticated versatile fuzz unit including variable controls affecting the fuzz quality whilst retaining the attack and decay, and also providing filtering. Can be used with other electronic instruments.

Set of basic components	KIT 56-1	£7.57
PCB & layout chart	PCB 56A	£1.78
Text photocopy		68p

P.E. SMOOTH FUZZ

Set of basic components, PCB & chart	KIT 91-1	£5.01
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TREMOLLO UNIT

A slightly modified version of the simple P.E. unit.

Set of basic components, PCB & chart	KIT 54-1	£3.23
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GUITAR FREQUENCY DOUBLER

A slightly modified and extended version of the P.E. unit.

Set of basic components, PCB & chart	KIT 74-1	£4.97
Text photocopy		39p

P.E. GUITAR SUSTAIN

Maintains the natural attack whilst extending note duration.

Basic components, foot switches, PCB & chart	KIT 75-1	£5.64
Basic components, panel switches, PCB & chart	KIT 75-2	£4.08
Text photocopy		38p

P.E. WAH-WAH UNIT

Can be controlled manually or by integral automatic control.

Set of basic components, PCB & chart	KIT 51-1	£3.99
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P.E. AUTO-WAH UNIT

Automatically Wah or Swell sounds with each note played.

Basic components, foot switches, PCB & chart	KIT 58-1	£8.43
Basic components, panel switches, PCB & chart	KIT 58-2	£5.31
Text photocopy		58p

ELEKTOR WAVEFORM CONVERTOR

Converts a saw-tooth waveform into sinewave, mark-space saw-tooth, regular triangle, or square-wave with variable mark-space ratio.

Basic components, PCB & chart, but excl. sw's.	KIT 67-1	£9.24
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P.E. VOLTAGE CONTROLLED FILTER

Extracted from P.E. Minisonic project.

Set of basic components, PCB & chart	KIT 65-1	£7.88
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P.E. RING MODULATOR

Extracted from P.E. Minisonic project.

Set of basic components, PCB & chart	KIT 59-1	£6.06
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ELEKTOR RING MODULATOR

Compatible with the Formant & most other synthesisers.

Set of basic components	KIT 87-1	£4.86
PCB (as published)	PCB 79040	£1.74
Text photocopy		38p

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TERMS: Goods in current adverts & lists over £50 goods value (excl. P.P. & VAT). Correctly coded, C.W.O., U.K. orders only. This voucher must accompany order. Valid until end of month on cover of P.E. Does not apply to credit card orders.

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U.K. orders: Keyboards add £2.30 each. Other goods: Under £5 add 25p, under £20 add 50p, over £20 add 75p. Recommended insurance against postal mishaps: add 50p for cover up to £50, £1 for £100 cover, etc., pro-rata. Insurance must be added for credit card orders. N.B. Eire, C.I., B.F.P.O. and other countries are subject to higher export postage rates.

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AND OTHER PROJECTS

PHOTOGRAPHS in this advertisement show two of our units containing some of the P.E. projects built from our kits and PCBs. The cases were built by ourselves and are not for sale, though a small selection of other cases is available.

LIST—Send stamped addressed envelope with all U.K. requests for free list giving fuller details of PCBs, kits and other components.

OVERSEAS enquiries for list Europe—send 20p; other countries—send 50p.



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KIMBER-ALLEN KEYBOARDS as required for many published projects. The manufacturers claim that these are the finest moulded plastic keyboards available. All octaves are C to C, the keys are plastic, spring-loaded, fitted with actuators, and mounted on a robust aluminium frame.

3 Octave (37 notes) **£26.60**
4 Octave (49 notes) **£32.25**
5 Octave (61 notes) **£39.75**

CONTACT ASSEMBLIES (gold-clad wire) — 1 required for each KBD note:

Type GJ — SPCO 25½p ea. Type GA — 1 pr of contacts, normally open 24p ea. Type GB — 2 pr N/O 28½p ea. Type GC — 3 pr N/O 37½p ea. Type GE — 4 pr N/O 48½p ea. Type GH — 5 pr N/O 58½p ea. Type 4PS — 3 pr N/O plus SPCO 87p ea.

P.E. NOISE GENERATOR

Extracted from the P.E. Minisonic.

Set of basic components, PCB & chart **KIT 60-1 £4.00**

WIND & RAIN EFFECTS UNIT

A slightly modified version of the original P.E. unit.

Set of basic components, PCB & chart **KIT 28-1 £4.88**
Text photocopy **28p**

P.E. ENVELOPE SHAPER WITHOUT VCA

Provides full manual control over attack, decay, sustain and release functions, and is for use with an existing VCA.

Set of basic components, PCB & chart **KIT 44-1 £5.24**
Text photocopy **49p**

P.E. ENVELOPE SHAPER WITH VCA

Has an integral Voltage Controlled Amplifier, and has full manual control over the A,D,S,R functions.

Set of basic components, PCB & chart **KIT 50-1 £7.34**
Text photocopy **58p**

P.E. GENERATOR

An ADSR envelope shaper without VCA, and additional providing Repeat-triggering enabling a synthesiser to be programmed for mandolin or banjo effects.

Set of basic components, PCB & layout chart **KIT 63-1 £5.13**
PCB 63A £2.00
Text photocopy **58p**

P.E. EXTERNAL-INPUT SYNTHESISER-INTERFACE

Allows external inputs such as guitars, microphones etc., to be processed by synthesiser circuits.

Set of basic components, PCB & chart **KIT 81-1 £3.23**

P.E. TUNING FORK

Produces 84 switch-selected frequency-accurate tones with an LED monitor clearly displaying beat-note adjustments.

Set of basic components, PCB & chart **KIT 46-1 £16.42**
Power Supply components, PCB & chart **KIT 46-2 £6.90**
Text photocopy **97p**

P.E. TUNING INDICATOR

A simple 4-octave frequency comparator for use with synthesisers and other instruments where the full versatility of KIT 46 is not required.

Set of basic components, PCB & chart, but excl. sw. **KIT 69-1 £8.19**
Text photocopy **58p**

P.E. DYNAMIC RANGE LIMITER

Preset to automatically control sound output levels.

Set of basic components, PCB & chart **KIT 62-1 £5.03**

P.E. CONSTANT DISPLAY FREQUENCY COUNTER

A 5-digit counter for 1Hz to 55KHz with 1Hz sampling rate. Readout does not count visibly or flicker due to blanking.

Set of basic components **KIT 79-1 £26.46**
PCB (as published) **PCB 79A £3.33**
Text photocopy **78p**

INTEGRATED CIRCUITS

301	48p
318	220p
320-15	195p
323	562p
324	87p
341-15	87p
356	101p
709	48p
723	51p
726	1005p
741	24p
748	57p
4001	15½p
4007	17½p
4011	17½p
4013	34p
4016	33p
4017	54p
4024	46½p
4046	96p
4049	37½p
4066	40½p
4069	18p
4081	16½p
4136	126p
AM2833	396p
AY-1-0212	617p
AY-1-1320	636p
AY-1-6721/6	188p
CA3046	71p
CA3080	63p
CA3084	209p
FX209	729p
M252	680p
MC3340	150p
MCM6810	670p
RC4195	117p
SAD1024	1762p
SG3402	262p
TDA1022	582p
TLO74	120p
XR2207	420p
ZN425	375p
7400	20p
7402	20p
7404	18p
7413	31p
7420	15p
7447	72p
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Auditorium	8, 16	15	60	Hi-Fi	£30
Group 35	8, 16	12	40	PA	£12
Group 45	4, 8, 16	12	45	PA	£15
Group 50	4, 8, 16	12	60	PA	£20
Group 50	8, 16	15	75	PA	£30
Group 75	4, 8, 16	12	75	PA	£24
Group 100	8, 16	12	100	PA	£29
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EMI 15 x 8 1/2 in. 3-way Loudspeaker System 5in. Bass, 5in. Middle, 3in. Tweeter; 3-way Cross-over & Ready Cut Baffle. Full assembly instructions supplied. Response 60 to 20000 c.p.s. 12 watt RMS 8 ohms £10.95 per kit. Two kits £20. Suitable Bookshelf Cabinet £9.50 each. Post £1.60.



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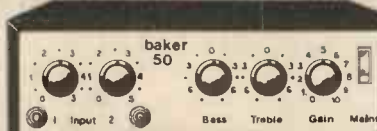
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Ideal for Groups, Disco, P.A. and Musical instruments. 4 inputs speech and music 4 way mixing. Output 4/8/16 ohm, a.c. Mains 240V. Separate treble and bass controls. 100 volt line model £14 extra. **£85** Post £1.60



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Two inputs with volume controls. Master treble bass and volume controls. Suitable for all loudspeakers. £65. Post £1.60.

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Complete kit of parts with R.C.S. printed circuit. Three 1000W channels. Will operate from 200mV signal source. CABINET extra £4. **KIT = £18.00**

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This kit is suitable for record players, tape play back, guitars, electronic instruments or small PA systems. Two versions are available. The mono kit uses 13 semiconductor. The stereo kit uses 22 semiconductors. Both kits have printed front panel and volume, bass and treble controls. Spec. 10W output into 8 ohms 7W into 15ohms. Response 20c.s to 30Kc.s. Size 9 1/2 x 3 x 2in. A/C mains operated.

Mono kit **£12.50** Stereo kit **£20** Post 45p
Easy to build. Full instructions supplied

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ELECTRICAL DATA VERTICAL AXIS (Y) Deflection Sensitivity - 100mV/division Bandwidth (between 3 dB points) - DC - 5MHz. Input Attenuator - (calibrated) - 9 step 0.1, 0.2, 0.5, 1, 2, 5, 10, 20, 50/div Input impedance - 1 Meg/40 pF inshunt Input Voltage - Max - 600V P.P. **HORIZONTAL AXIS (X)** Deflection Sensitivity - 0.400mV/division Bandwidth (between 3 dB points) - 1Hz-350KHz Gain Control - Continuous when time bases in EXT position Input Impedance - 1 Meg Input Voltage - Max - 500V P.P. **TIME BASE** Sweep Range (calibrated) - 100msec/div to 1 1/2 sec/div in 5 steps FINE Control - Variable between steps - includes time-base calibration position Blanking - Internal - on all ranges

SYNCHRONISATION Selection - Internal external Synchronisation Level - Continues from positive to negative **POWER SUPPLY** Input voltage - 115/200V AC - 10% at 50/60Hz Power Dissipation - 18W **CRT DATA** - 4in - flat face single beam - Maximum high voltage - 1.5kV - Fitted with 8x10 division blue filter graticule **PHYSICAL DATA** Dimensions - 15cm(h)x20.5cm(w)x28cm(d) Weight - 4.3Kg (approx.) Stand - 2 position flat and inclined Case - Steel, epoxy enamelled Front panel - Aluminium enamelled epoxy printing

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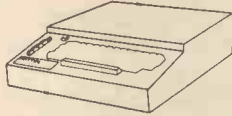
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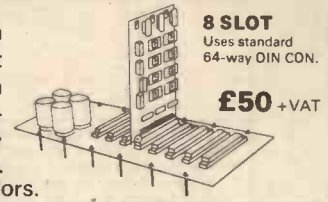
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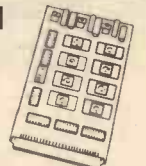
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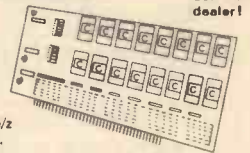
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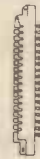
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SN74LS14N	89	SN74LS61N	35	SN74LS146N	174	SN74LS193N	175	SN74LS260N	175
SN74LS15N	25	SN74LS62N	115	SN74LS151N	85	SN74LS194N	185	SN74LS263N	180
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Guaranteed same day despatch. Very slim, only 6mm thick.



M1

SOLAR QUARTZ LCD 5 Function

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M2

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SLIM
CHRONO

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M3

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M4

MULTI ALARM 6 Digits 10 Functions

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- Basic alarm.
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- Timer alarm with dual.
- Time and 10 country zone.
- Back-light.
- 8mm thick.

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M5

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M6

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6 digits, 5 flags, 22 functions. Solar panel with battery back-up. 6 basic functions. Stop-watch to 12 hours 59.9 secs., in 1/10 sec. steps. Split and lap timing modes. Dual time zones. Alarm, 9mm thick. Back-light. Fully adjustable bracelet.

£27.95



M7

ALARM CHRONO with 9 world time zones

- 6 digits, 5 flags.
- 6 basic functions.
- 8 further time zones.
- Count-down alarm.
- Stop-watch to 12 hours 59.9 secs. in 1/10 sec. steps.
- Split and timing modes.
- Alarm.
- 9 mm thick.
- Back-light.
- Fully adjustable bracelet.

£29.65



M8

SOLAR QUARTZ LCD Chronograph

Powered from solar panel with battery back-up. 6 digit, 11 functions. Hours, mins., secs., day, date, day of week, 1/100th, 1/10th, secs., 10X secs., mins. Split and lap modes. Back-light, auto calendar. Only 8mm thick. Stainless steel bracelet and back. Adjustable bracelet. Metac Price

£13.65

Guaranteed same day despatch.



M9

QUARTZ LCD Ladies Day Watch

Only 26 x 20mm and 6mm thick. Hours, minutes, seconds, day, date, backlight and auto calendar. Elegant metal bracelet in silver or gold fully adjustable to suit very slim wrists. State colour preference.

£9.95

Guaranteed same day despatch.



M15

QUARTZ LCD Ladies Fashion Watch

Elegant bracelet in bronze/gold finish or silver colour. Hours, mins, secs, day, date, backlight and auto calendar. Adjustable for the slimmest of wrists. State colour preference.

£14.95

Guaranteed same day despatch



M17

QUARTZ LCD Ladies Cocktail Watch

Highly functional watch which also suits those special occasions. Beautifully designed with a very thin bracelet which retains strength as well as elegance. Hours, mins, secs, day, date, backlight and auto calendar. Bracelet fully adjustable to suit slim wrists. State gold or silver finish.

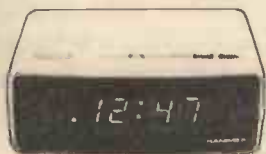
£19.95

Guaranteed same day despatch



M18

HANIMEX Electronic LED Alarm Clock



Features and Specification:
Hour/minute display. Large LED display with p.m. and alarm on indicator. 24 Hours alarm with on/off control. Display flashing for power loss indication. Repeatable 9-minute snooze. Display bright/dim modes control. Size: 5.15" x 3.93" x 2.36" (131mm x 11mm x 60mm). Weight: 1.43 lbs (0.65 kg). AC power 220V.

£10.20 Thousands sold!

Mains operated.

Guaranteed same day despatch.

M13

EXECUTIVE ALARM WATCH

6 Functions plus Alarm: Conference signal, 5 minute snooze alarm. Conference signal sounds 4 secs., before main alarm to give advance warning and an option to cancel. Snooze sounds 5 mins., after main alarm and is always preceded by the conference signal.

£14.95

M60



MACY QUARTZ ANALOGUE

Automatic Calendar Day and Date infinite battery. This mans watch has elegance as well as the robust appearance provided by a watch with traditional features. Accuracy is provided by a quartz crystal powered by a long life miniature battery.

£24.95

M21



HOW TO ORDER

Payment can be made by sending cheque, postal order, Barclay, Access or American Express card numbers. Write your name, address and order details clearly, enclose 40 pence per single item for post and packing or the amount stated in the advert. All products carry 1 year written guarantee and full money-back 10 day reassurance. Battery fitting and electronic calibration service is available to customers at any Metac shop. All prices include VAT currently at 15%.

Metac Wholesale:

Trade enquiries - send for a complete list of prices for all the goods advertised plus many more not shown also minimum order details. Telephone orders: Credit card customers can telephone orders direct to Daventry (03272) 76545 or Edgware Rd. 01-723 4753 24 hours a day.



Service Enquiries 03272-77659
CALLERS WELCOME Shops open 9-30am-6.00

Metac price breakthrough for an Alarm Chronograph with Dual Time only

£18.95

OUTSTANDING FEATURES

- **DUAL TIME.** Local time always visible and you can set and recall any other time zone (such as GMT). Also has a light for night viewing.
 - **CALENDAR FUNCTIONS** Include the date and day in each time zone.
 - **CHRONOGRAPH/STOPWATCH** displays up to 12 hours, 59 minutes, and 59.9 seconds.
 - On command, stopwatch display freezes to show intermediate (split/lap) time while stopwatch continues to run. Can also switch to and from timekeeping and stopwatch modes without affecting either's operation.
 - **ALARM** can be set to anytime within a 24 hour period. At the designated time, a pleasant, but effective buzzer sounds to remind or awaken you!
- Guaranteed same day despatch. **M16**

Metac

ELECTRONICS
& TIME CENTRES

North & Midlands
67 High Street, DAVENTRY
Northamptonshire
Telephone: 03272 76545

South of England
327 Edgware Road
LONDON W.2
Telephone: (01) 723 4753

MICRODIGITAL 1980

Apple II plus Nascom 2



Apple II Plus will change the way you think about computers. That's because it is specifically designed to handle the day to day activities of education, business, financial planning, scientific calculation and entertainment.

APPLESOFT
A fast, extended 10K BASIC with 9-digit precision and graphics extensions.
HIGH RESOLUTION GRAPHICS
On a matrix of 280 x 192 individually addressable points.
AUTO-START ROM
With power on boot of applications programs, reset protection and improved screen editing.
INTERNAL MEMORY EXPANSION TO 64K BYTES
For bit system performance at a low cost.
EIGHT EXPANSION SLOTS
To let the system grow with your needs

	Nett	V.A.T.	Total
Apple II Plus,			
16K RAM	750.00	112.50	862.50

APPLE PASCAL
Apple Pascal is the new extension to microcomputer power.
Pascal Incorporating UCSD PASCAL TM, offers extended features in a complete interactive package employing today's most sophisticated structured programming language. It provides advanced capabilities that boost performance and cut development time for large business, scientific and educational programs.
This software package provides the most powerful set of tools yet available for the microcomputer programmer.

	Nett	V.A.T.	Total
Apple Pascal			
System	296.00	44.40	340.40

FLOPPY DISCS
Gives your system immediate access to large quantities of data. The subsystem consists of an intelligent interface card, a powerful Disk Operating System and one or two mini-floppy drives.

	Nett	V.A.T.	Total
Floppy disk			
Subsystem	398.00	59.70	457.70
Second disk drive			
and connecting			
cable	355.00	53.25	408.25

Parallel Printer Interface Card
Allows you to connect almost any popular printer to your apple. A BASIC program can produce hard-copy output as easily as it prints to the TV monitor screen.
Command interpretation and printer control details are handled by the firmware built into the card, to eliminate user programming requirements.

	Nett	V.A.T.	Total
Parallel Printer			
Interface Card	110.00	16.50	126.50

Communications Interface Card
Allows your Apple to "talk" (through a modem) with other computers and terminals over ordinary telephone and load programs over the phone, send messages to remote terminals or access your office computer from the comfort of your home

	Nett	V.A.T.	Total
Communications			
Interface Card	110.00	16.50	126.50

High Speed Serial Interface Card
Allows Apple to exchange data with printers, plotters and computers in serial format at up to 19.2K Baud.

	Nett	V.A.T.	Total
High Speed Serial			
Interface Card	110.00	16.50	126.50

Apple Computer
Sales and Service

Microprocessor
280A 8 bit CPU. This will run at 4 MHz but is selectable between 1/2/4/ MHz.
Hardware
12" x 8" Card
All bus lines are to the Nasbus specifications
All bus lines are full buffered
Memory
On-board, addressable memory:
2K Monitor - Nasbus 1
1K Video RAM (MK4118)
1K Work space/User RAM (MK4118)
8K Microsoft Basic (MK3600 ROM)
8K Static RAM/2708 EPROM



Keyboard
New expanded 57 Key Licon solid state keyboard especially built for Nascom. Uses standard Nascom, monitor controlled, decoding.
T.V.
The T.V. Peak to peak video signal can drive a monitor directly and is also fed to the on-board modulator to drive the domestic T.V.
I/O
On-board UART (Int. 6402) which provides serial handling for the on-board cassette interface or the RS 232/20mA teletype interface. The cassette interface is Kansas City standard at either 1200 or 300 baud. This is a link operation on the Nascom-2
P/O
There is also a totally uncommitted P/O (MK3881) giving 16, programmable, I/O lines.

Character Generator
The 1K video RAM drives a 2K ROM character generator providing the standard ASCII Character set with some additions, 128 characters in all. There is a second 2K ROM socket for an on-board graphics package which is software selectable.

	Nett	V.A.T.	Total
Nascom-2 In kit			
form	295.00	44.25	339.25
Power Supply ..	24.50	3.68	28.18
Graphics ROM ..	15.00	2.25	17.25

REED RELAY BOARD
Put your Nascom to work with the new Microdigital Relay Board.
16 Reed Relays, totally isolated 200 mA, 50 V.D.C. 5W max each. Operate and release time 1 ms (including bounce).
Single sided, glass fibre board, with gold plated edge connectors and silk screened component layout.
Plugs directly into Nasbus, does not interfere with normal Nascom operation, all interrupt and D.M.A. Daisy Chain Links carried on. Draws only 250 mA from each of the + and +12V Rails.
All components supplied, all IC's socketed, easy to build, and easy to program in Basic or Machine Code.
Occupies 2 consecutive ports, link selectable - several boards can be used on one Nascom system.
Output is via front edge connector on 0.1" centres. Uses standard edge connectors for connection to controlled devices.
Complete manual with sample software.

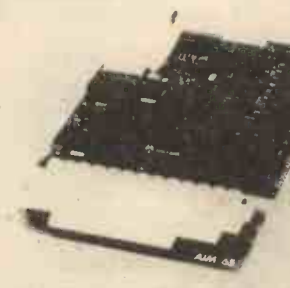
Applications
Light displays
Industrial process control
Model Railway Control
Pre programmed music generation
Robots, Central Heating Systems
Stepping Motors

	Nett	V.A.T.	Total
P.C.B. (+ Manual)	15.00	2.25	17.25
Kit	49.95	7.49	57.44
Assembled	60.00	9.00	69.00

Rockwell Aim 65

Thermal Printer
Wide 20 column printout.
Versatile 5 x 7 dot matrix format
Complete 64 - character ASCII alphanumeric format
Fast 120 lines per minute
Quiet thermal operation
Full Size Alphanumeric Keyboard
Standard 54 key, terminal-style layout
26 alphabetic characters
10 number characters
22 special characters
9 control functions
3 User-defined functions

True Alphanumeric Display
20 characters wide
16-segment characters
High contrast monolithic characters
Complete 64 - character ASCII alphanumeric format
Read/Write Memory, using R2114 Static RAM devices. Available in 1K byte and 4K byte version.
8K Monitor Program Memory, using R2332 Static ROM devices. Has sockets to accept additional 2332 ROM or 2532 PROM devices, to expand on-board Program memory up to 20K bytes.
8K532 RAM Input/Output Timer (RIOT) combination device. Multipurpose circuit for AIM 65 Monitor functions.
Two R6522 Versatile Interface Adaptor (VIA) devices, which support AIM65 and user functions.



	Nett	V.A.T.	Total
AIM-65 with 1K	249.50	37.43	286.93
AIM-65 with 4K	315.00	47.25	362.25
8K BASIC ROM	70.00	10.50	80.50
4K ASSEMBLER			
ROM	59.50	8.93	68.43
CASE (includes			
P.S.U.)	78.00	11.70	89.70
P.S.U.	41.83	6.27	48.10
MOTHERBOARD	136.50	20.48	156.98

Sharp

SHARP MZ-80K
Z-80 based CPU
4K Byte monitor in ROM
Internal memory capacity from 4 to 48K RAM
14K Extended BASIC.
10 in video display, 40 chars of 24 lines
80 x 50 bit mapped graphics.
Extensive character set with upper, lower case, graphics etc
Full 79 Key Keyboard.
Built in music synthesizer with 3 octaves.
Fast reliable cassette unit with tape counter 1200 bps
Wide variety of system software on cassette
50 pin bus connector for system expansion



A complete personal computer system for the microcomputer user, at an economic price. The Sharp comes complete with all necessary peripherals, sample software and excellent documentation - giving the user a personal system of unmatched flexibility and ease of use. At the heart of the machine is the Z-80 CPU - widely accepted as the most powerful 8-bit CPU on the market. A 4K byte system monitor controls system operation. From 4 to 48K of RAM can be resident on board, enough room for the most demanding applications

An extensive graphics character set, plus 3 octave sound generator and fast cassette unit hi-resolution video monitor complement these basic facilities. It has the ease of use and compactness of 'black box' computer combined with extensive peripherals and facilities for expansion.
Sharp Basic occupies 14K of RAM and offers extended features above those of normal microcomputer implementations.

	Nett	V.A.T.	Total
Model			
6K	520.00	78.00	598.00
10K	540.00	81.00	621.00
18K	620.00	93.00	713.00
22K	640.00	96.00	736.00
34K	740.00	111.00	851.00

Internal RAM 16K Bytes
External ROM (Plug in software modules) 30K Bytes

Keyboard:
Staggered QWERTY Layout, full travel with overlay for second functions.
Sound:
5 Octaves, 3 simultaneous tones plus noise generator
Colours: 16
Graphics resolution: 256/192

Input/Output:
Composite video and audio output for monitor. Interface for 2 audio cassette recorders 44-pin peripheral connector with system memory and address signals available. Mini-earphone jack. Hand controller interface.
Built in software:
14K Byte T.I. BASIC, equation calculator and control software.
Size: 25.9/38.1/7.1 cm.
Display:
Uses colour monitor, 24 lines of 32 characters

	Nett	V.A.T.	Total
Console			
CPU: 9900 family, 16 bit microprocessor, plus 256 byte scratchpad RAM.	569.57	85.43	655.00
Modules	15	40.00	
Joysticks		25.00	
Speech synthesizer ..		45.00	
* Please note these are estimated prices only.			

Texas

The remarkable TI-99/4 Home Computer. Superior colour, music, sound and graphics - and a powerful extended BASIC all built in. Plus a unique, new Solid State Speech Synthesizer and TI's special Solid State Software.

The TI-99/4 was designed to be the first true home computer - skilled computer users and beginners alike will be able to put it to effective use right away. You can begin using the TI Home Computer minutes after unpacking it; simply snap in a Solid State Software Module, touch a few keys and step-by-step instructions appear on the screen.

Powerful TI-Basic: Accuracy and power for demanding technical applications, yet easy to use for the beginner. 13-digit, floating point Basic, with special features and extensions for colour, sound and graphics.
16-colour graphics capability - Easy to use, high resolution graphics with special features that let you define your own characters, create animated displays, charts, graphics, and more, with a resolution of 256 x 192 individually addressable points.
Music and sound effects: provides outstanding audio capability. Build three-note chords and adjust frequency, duration and volume quickly and simply.

Console:
CPU: 9900 family, 16 bit microprocessor, plus 256 byte scratchpad RAM.
Memory:
Total combined memory capacity 72K Bytes
Internal ROM 26K Bytes



Bigger and better than ever!

Commodore Pet



A complete Computer for the price of a good typewriter! With a library of over 200 programs in business, science, education and entertainment.

Pet can store and retrieve data which conveniently occupies large storage capacity, and solve numerical problems traditionally tedious and time consuming.

Ease of Operation
The Commodore PET comes complete with a built-in T.V. screen, and keyboard as well as its full computer circuitry. It is plugged into any 13 amp and no special computer knowledge is needed for running standard programs. Personal programs can readily be written in the BASIC computer language of PET which is easily learned.

An Expandable System
Further expansion is a prime design concept enabling PET to be made the heart of a much larger system incorporating printers, floppy discs etc., as and when required.

Computers
PET 2001-8 — PET with integral cassette and calculator type keyboard, 8K bytes memory.
Nett 550.00 V.A.T. 82.50 Total 632.50

PET 2001-16N — PET with 16K bytes memory and large keyboard. External cassette optional.
Nett 675.00 V.A.T. 101.25 Total 776.25

PET 2001-32N — PET with 32K bytes memory and large keyboard. External cassette optional.
Nett 795.00 V.A.T. 119.25 Total 914.25

Computink Disk Units
400K Random for 8K Pet 795.00 119.25 914.25
400K Random for New Pet 2 840.00 126.00 966.00
800K Random for New Pet 2 995.00 149.25 1,144.25

Acorn

This compact stand-alone micro-computer is based on Eurocard modules, and employs the highly popular 6502 MPU. Take a look at the full specifications, and see how Acorn meets your requirements.

The Acorn consists of two single Eurocards:
1. MPU card; 6502 microprocessor, 512 x 8 ACORN Monitor, 1K x 8 RAM; 15-way I/O with 128 bytes of RAM; 1 MHz crystal; 5V reg. sockets for 2K EPROM and second RAM I/O chip.
2. Keyboard card; 25 click-keys (16 hex, 9 control); 8 digit, 7 segment display. CUTS standard crystal controlled tape interface circuitry.

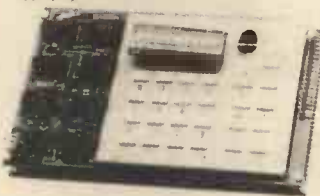
Acorn Operating Manual
With Acorn, you'll receive an operating manual that covers computing in full, from first principles of binary arithmetic, to efficient hex programming with the 6502 instruction set. The manual also includes a listing of the monitor programs and the instruction set, and other useful tabulations; plus sample programs.

Nett V.A.T. Total
Kit 65.00 9.75 74.75
Ready Built 75.00 11.25 86.25

A high quality fibre glass, through hole plated, PCB with solder resist and component identification, this eurocard has provision for 8K of RAM (2114) and 8K of EPROM (2732).

Nett V.A.T. Total
8K RAM (Kit) 95.00 14.25 109.25

ACORN VDU
The Acorn VDU Board connects to the Acorn Computer Bus and contains memory mapped character storage RAM which is transparently written to or read from, by the C.P.U.



An MC 6845 programmable controller IC. Provides all the synchronisation signals to drive a 625 line 50 fields per second V.D.U. together with read addresses for the character R.A.M. Characters are then fed to an SAA 5050 character generator IC which produces the necessary dot patterns to create the characters to refresh the V.D.U.

The SAA 5050 produces Teletext standard characters and has Red, Green and Blue drive outputs giving coloured characters or graphics.

Nett V.A.T. Total
V D U Card (Kit) 88.00 13.20 101.20

MICRODIGITAL

25 Brunswick Street, Liverpool L2 0PJ Mail orders to: MICRODIGITAL LIMITED,
Tel: 051-236 0707 (24 Hour Mail Order) FREEPOST (No Stamp Required)
051-227 2535/6/7/8 (All other Depts.) Liverpool L2 2AB.

Microdigital Software Announce

5 packages which are:
General Ledger £295.00
Purchase Ledger £295.00
Sales Ledger £295.00
Stock Control £200.00
Payroll £360.00

These packages are now available, demonstrations on request
Ring 051-227 2535 and ask for Graham Jones (Software Manager).
These are fully tested systems which run on the Apple/ITT 2020 with one or two disc drives.
The Stock Control package can handle up to 1250 stock items and uses two disc drives.

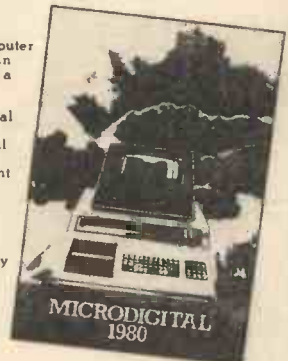
The Mighty Microdigital

Microdigital are one of the largest and longest established Microcomputer firms in Europe. We sell a wide range of systems, backed up by support services that are second to none.
Our present retail outlet is at 25 Brunswick Street, Liverpool. Our well informed staff are happy to demonstrate equipment, provide technical help, or just chat.
Microdigital (mail order) is the fastest, friendliest and most efficient mail order service available. All orders (or an acknowledgement if the goods are temporarily out of stock) are despatched by return post. Telephone orders are welcome (24 hours a day) and we even have a "Freepost" service so you don't have to remember the stamp!
Microdigital Manufacturing is our hardware department. We carry out repairs and servicing in-house rather than depending on the manufacturer. In addition we design and manufacture our own peripheral boards for the systems we support. Custom design services are also available.

Microdigital (Software) is responsible for the development of commercial, high quality, computer programs. We can advise on the suitability of an existing package, modify the package, or write a completely new system to the customers specification.
Microdigital (Hire) provides a service for potential customers — the capabilities of a particular machine can be evaluated without a substantial capital investment.
All in all we try and provide the most competent service in the Microcomputer industry.

The Microcomputer shop providing a complete service from a single chip to a data processing installation.
Opening hours: 9 - 5.30 Mondays to Saturday
Friendly, expert staff always on hand.

Our new, glossy, 16 page brochure is now the talk of the industry! — Send for your free copy today.



At last! Britain's very own monthly journal for Microsoftware ...

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Programming practices and techniques — *Dr. M. Beer*.
I'm Pilot, fly me — *D. Straker*. Acorn Mastermind
— *L. Hardwick*. Apple pips -
C. Phillips.
PETS corner — *J. Stout*.



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SJ30	10 C30	15 min per side	£2.00
SJ55	10 C46	23 min per side (LP)	£2.50
SJ31	10 C90	45 min per side	£3.50
SJ32	10 C120	40 min per side	£4.50

ALL REDUCED! CAPACITOR PAKS

16201	18 electrolytics	4.7uf-10uf
16202	18 electrolytics	10uf-100uf
16203	18 electrolytics	100uf-680uf
ALL 3 at SPECIAL PRICE of £1.30		
16160	24 ceramic caps	22pf-82pf
16161	24 ceramic caps	100pf-390pf
16162	24 ceramic caps	470pf-3300pf
16163	24 ceramic caps	4700pf-0.047pf.
ALL 4 at SPECIAL PRICE of £1.80		

RESISTOR PAKS

16213	60 1/2w resistors	100ohm-820ohm
16214	60 1/2w resistors	1K-8.2K
16215	60 1/2w resistors	10K-82K
16216	60 1/2w resistors	100K-820K
ALL 4 at SPECIAL PRICE of £1.80		
16217	40 1/2w resistors	100ohm-820ohm
16218	40 1/2w resistors	1K-8.2K
16219	40 1/2w resistors	10K-82K
16220	40 1/2w resistors	100K-820K
ALL 4 at SPECIAL PRICE of £1.80		

IC SOCKET PAKS

		F.E.T.'s	
SJ36	14	8 pin	2N3819 £0.17
SJ37	12	14 pin	2N5458 £0.18
SJ38	11	16 pin	2N4220 £0.28
SJ39	8	18 pin	2N4860 £0.25
SJ40	7	20 pin	
SJ41	6	22 pin	
SJ42	5	24 pin	
SJ43	4	28 pin	
SJ44	3	40 pin	2N6027 £0.26
ALL at ONLY £1.00 EACH			

VOLTAGE REGULATORS

Positive	Case TO220	Negative	
uA7805	£0.65	uA7905	£0.70
uA7812	£0.65	uA7912	£0.70
uA7815	£0.65	uA7915	£0.70
uA7818	£0.65	uA7918	£0.70
uA7824	£0.65	uA7924	£0.70
UA723 14 pin DIL	£0.35		
LM309K TO3	£1.10		

OPTOELECTRONICS DISPLAYS

1510	707 LED Display	Price each	£0.70
1511	747 LED Display	Price each	£1.50
1512	727 LED Display	Price each (dual)	£1.55

L.E.D.'s

		Price each
SJ78	.125 LED Diffused RED	£0.08
SJ79	.2 LED Diffused RED	£0.08
S120	.125 LED Bright RED	£0.09
S121	.2 LED Bright RED	£0.09
S102	.125 LED Diffused GREEN	£0.11
S105	.2 LED Diffused GREEN	£0.11
S103	.125 LED Diffused YELLOW	£0.11
S106	.2 LED Diffused YELLOW	£0.11
SJ80	.2 LED Bright YELLOW	£0.14
SJ82	.2 LED Clear Illuminating RED	£0.10
SJ83	.125 LED Clear Illuminating RED	£0.10

2nd QUALITY LED PAKS

1507	10 assorted colours & size	£0.85
S122	10 .125 RED	£0.50
S123	10 .2 RED	£0.50

LED CLIPS

1508/125	.125	5 for £0.10
1508/2	.2	5 for £0.12

SJ81	1 Infra RED emitter - Fairchild FP100	£0.25
SJ98	5 Photo Detector MEL11 + Data	£1.00
ORP12	NORP12 Cad Cell	£0.45
SJ99	4 ITT 5870 SX Nitride Tubes	£1.00

SJ29	Texas NPN silicon transistors 2S503=BC108 TO-18 metal can - perfect & coded	
50 off £2.50 - 100 off £4.00 - 1000 off £35.00		

SPECIAL OFFER

SJ100	12v Electric Drill 7,500 RPM for all your PCB drilling complete with 2 drills - 1 & .75.	
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SUPER DUPER COMPONENT BOX

Min. 3 lbs in weight consisting of a fantastic assortment of Electronic Components - Pots, Resistors, Condensers, Switches, Relays, Board-Semiconductors, wire, hardware, etc., etc., etc.
This is a large box and is sent separate to your order
£140 £2.50 including p&p

CALCULATOR CHIP

GOM2-C500	24 pin MOS	£0.50
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IC INSERTION/EXTRACTION TOOL		
2015	DEPT. PE2	£0.35 each

TRANSISTORS

Type	Price*	Type	Price	Type	Price
AC107	£0.20	BC441	£0.25	TP30A	£0.30
AC126	£0.14	BC460	£0.28	TP30B	£0.32
AC127	£0.16	BC481	£0.28	TP30C	£0.34
AC128	£0.15	BC477	£0.15	TP31	£0.30
AC128K	£0.24	BC478	£0.15	TP31A	£0.30
AC176	£0.16	BC479	£0.15	TP31B	£0.32
AC171K	£0.24	BC547	£0.08	TP31C	£0.34
AC187	£0.16	BC548	£0.08	TP32	£0.30
AC187K	£0.26	BC549	£0.08	TP32A	£0.30
AC188	£0.16	BC557	£0.10	TP32C	£0.32
AC188K	£0.26	BC558	£0.09	TP32E	£0.34
AD161/162MP	£0.65/pr	BC559	£0.10	TP41	£0.34
AO140	£0.80	BCY70	£0.13	TP41A	£0.34
AO149	£0.53	BCY71	£0.13	TP41B	£0.36
AF239	£0.35	BCY72	£0.13	TP41C	£0.38
BC107	£0.06	BD115	£0.45	TP42	£0.34
BC107A	£0.06	BD131	£0.30	TP42A	£0.34
BC107B	£0.07	BD132	£0.30	TP42B	£0.36
BC107C	£0.09	BD135	£0.28	TP42C	£0.38
BC108	£0.06	BD136	£0.28	TP2955	£0.50
BC108A	£0.06	BD239A/		TP3056	£0.48
BC108B	£0.07	BD240A/MP		ZTX107	£0.08
BC108C	£0.09		£0.80/pr	ZTX108	£0.08
BC109	£0.06	BF115	£0.20	ZTX109	£0.09
BC109B	£0.07	BF157	£0.20	ZTX300	£0.10
BC109C	£0.09	BF173	£0.20	ZTX301	£0.10
BC113	£0.10	BF195	£0.09	ZTX302	£0.12
BC114	£0.12	BF196	£0.09	ZTX500	£0.10
BC116	£0.16	BF197	£0.10	ZTX501	£0.10
BC118	£0.10	BF257	£0.22	ZTX502	£0.12
BC140	£0.20	BF258	£0.22	2N696	£0.10
BC141	£0.20	BF259	£0.24	2N697	£0.10
BC142	£0.18	BF339	£0.20	2N706	£0.09
BC147	£0.07	BF440	£0.20	2N708A	£0.10
BC148	£0.07	BF797	£0.22	2N708	£0.10
BC149	£0.07	BF800	£0.22	2N1302	£0.15
BC157	£0.09	BFT84	£0.20	2N1303	£0.15
BC158	£0.09	BFX29	£0.20	2N1613	£0.18
BC159	£0.09	BFX84	£0.20	2N1711	£0.18
BC169C	£0.09	BFY50	£0.15	2N1893	£0.25
BC170	£0.06	BFY51	£0.15	2N2218	£0.18
BC171	£0.07	BFY52	£0.15	2N2218A	£0.20
BC172	£0.07	BIP19/20MP		2N2219	£0.18
BC173	£0.08		£0.70/pr	2N2219A	£0.20
BC177	£0.13	MJE340	£0.60	2N2221	£0.18
BC178	£0.13	MJE2955	£0.75	2N2221A	£0.20
BC179	£0.13	MJE3055	£0.50	2N2222	£0.18
BC182	£0.07	MPSA05	£0.15	2N2222A	£0.18
BC182L	£0.07	MPSA06	£0.15	2N2369	£0.12
BC183	£0.07	MPSA55	£0.15	2N2904	£0.16
BC183L	£0.07	MPSA56	£0.15	2N2904A	£0.17
BC184	£0.07	OC25	£0.50	2N2905	£0.16
BC184L	£0.07	OC26	£0.45	2N2905A	£0.18
BC207	£0.08	OC28	£0.60	2N2906	£0.18
BC208	£0.08	OC29	£0.55	2N2907	£0.15
BC209	£0.09	OC35	£0.55	2N2907A	£0.16
BC212	£0.07	OC36	£0.60	2N2926G	£0.08
BC212L	£0.07	OC42	£0.18	2N3063	£0.15
BC213	£0.07	OC44	£0.20	2N3054	£0.30
BC213L	£0.07	OC45	£0.18	2N3055	£0.35
BC214	£0.07	OC71	£0.12	2N3056	£0.35
BC214L	£0.07	OC72	£0.16	2N3702	£0.07
BC251	£0.10	OC75	£0.18	2N3703	£0.07
BC261	£0.14	OC81	£0.20	2N3704	£0.06
BC327	£0.12	TP29	£0.30	2N3705	£0.08
BC328	£0.12	TP29A	£0.30	2N3706	£0.07
BC337	£0.12	TP29B	£0.32	2N3771	£1.00
BC338	£0.12	TP29C	£0.34	2N3772	£1.10
BC440	£0.25	TP30	£0.30	2N3773	£1.50

DIODES

Type	Price	Type	Price	Type	Price
AA119	£0.06	OA70	£0.06	IN4004	£0.06
BA100	£0.08	OA79	£0.08	IN4005	£0.07
BA148	£0.13	OA81	£0.08	IN4006	£0.08
BA173	£0.13	OA90	£0.08	IN4007	£0.09
BAX13	£0.05	OA91	£0.08	IN5400	£0.12
BAX16	£0.06	OA95	£0.08	IN5401	£0.13
OA200	£0.06	IN34	£0.06	IN5402	£0.14
OA202	£0.07	IN60	£0.07	IN5404	£0.15
BY100	£0.18	IN414B	£0.05	IN5406	£0.19
BY126	£0.12	IN4001	£0.04	IN5407	£0.23
BY127	£0.14	IN4002	£0.04	IN5408	£0.28
OA47	£0.06	IN4003	£0.05	IS44	£0.03

LINEAR

Type	Price	Type	Price	Type	Price
CA270	£0.95	SL414A	£1.75	TB810	£0.85
CA3089	£1.70	SN76013N	£1.65	TB820	£0.85
CA3090	£3.00	SN76023N	£1.60	uA703	£0.05
LM380	£0.80	SN76115	£1.60	uA709C	£0.25
LM381	£1.35	TAA550	£0.30	uA710	£0.25
LM3900	£0.50	TAA621A	£1.80	uA711	£0.26
MC1310P	£0.85	TBA120B	£0.60	741P	£0.16
NE555	£0.18	TBA641A	£1.10	TAA661	£1.25
NE556	£0.55	TBA800	£0.75	TAA661B	£1.25

O/N/O	Quantity	£	p
SJ1	200 Resistors mixed values	0.50	
SJ2	200 Carbon resistors 1/4 watt preformed	0.50	
SJ3	100 1/2 watt miniature resistors mixed values	0.50	
SJ4	60 1/2 watt resistors mixed values	0.50	
SJ5	50 1-2 watt resistors mixed pot values	0.50	
SJ6	50 Precision resistors 1-2" tol. mixed	0.50	
SJ7	30 5-10 watt wirewound resistors mixed	0.80	
SJ11	150 Capacitors mixed types & values	0.80	
SJ12	60 Electrolytics all sorts mixed	0.50	
SJ13	50 Polyester/polystyrene capacitors mixed	0.50	
SJ14	50 C280 type capacitors mixed	1.00	
SJ15	40 High Quality electrolytics 100-470mf	1.00	
SJ16	40 Low Vts Electrolytics mixed up to 10v	0.50	
SJ17	20 Electrolytic transistor types mixed	0.50	
SJ18	20 Tantulum bead capacitors mixed	0.50	
SJ20	2 large croc clips 25A rated	0.30	
SJ21	Large 7 1/2" Mains Neon Tester screwdriver	0.85	
SJ22	Small pocket size Mains Neon Tester	0.55	
SJ23	Siemens 220v AC Relay DPDT contacts 10amp rating - housed in plastic case	1.00	
SJ24	Black PVC tape (1) 15mm x 25m - strong tape for electrical & household use	1.50	0.35 per roll
SJ25	100 Silicon NPN transistors all perfect & coded - mixed types with data & equivalent sheet	2.50	
SJ26	100 Silicon PNP trans		

BREADBOARD

AT THE time of writing the Breadboard exhibition is pulling in the crowds in London and the editorial staff of PE are at full stretch manning the stand, visiting exhibitors and producing this issue—we hope it doesn't show! Breadboard looks like becoming the premier event for the electronics hobbyist. In only its second year it has attracted large crowds, is very well presented and has an excellent atmosphere—even though it is sometimes difficult to talk to people over the raucous noise being generated by various synthesisers, organs, discos etc. (organisers please note!).

An exhibition such as this gives us a chance to meet casual and dedicated readers. We take the opportunity to discuss various projects and enquire into your views on PE. Thankfully most are full of praise and often some new project ideas get thrown up for our consideration. Some, of course, are not so happy—unfortunately we have failed to please them. It would appear that in these cases we are not giving enough space to their particular subject or, more commonly, we are giving

too much space to another facet of our hobby in which they are not interested.

The problem is, having talked to a number of readers, often with widely varying views, just what do we do with all the information. It is all too easy to twist it around so that it fits in with our own views, or to balance out two divergent attitudes and take the middle path; which may in fact please neither party.

For instance, two comments which came up were (a) we carry too many microprocessor oriented articles and (b) we give too much space to music and audio projects. Now we could please both parties by cutting back on these subjects, but wait, we know these are very popular areas by the number of kits and components sold—2000 Compukits (at £250 each) at the last count and excellent quantities of mixers, guitar sound multiprocessors etc. So we would probably offend many readers by cutting back on either subject. We do try to keep the balance right, to be aware of new trends and to provide the right information at the right time but it's not easy to please everyone.

PROJECTS

One thing that fascinates us is that virtually all the comments are concerned with projects. Although we try to publish about five projects, or parts thereof, each month these normally make up less than half of our total number of editorial pages—presumably everyone is over the moon with the rest of the contents?

Another interesting angle is uncovered when one enquires as to how many projects most readers construct during a year—the average is probably about one project per reader per year. So why are some readers so much in favour of reading about their pet subject but so much against learning, or even just reading, about ideas and circuitry from another discipline? Any comments?

Finally, our thanks to all those who visited our stand at Breadboard. To those who live too far away or who just could not come along—it's a great event—if you get the chance we would be pleased to see you next time.

We hope to publish a review of the exhibition next month; our apologies to those who have seen it all first hand.

Mike Kenward

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Technical Queries

We are unable to offer any advice on the use or purchase of commercial equipment or the incorporation or modification of designs published in Practical Electronics.

All letters requiring a reply should be accompanied by a stamped, self addressed envelope and each letter should relate to one published project only.

Components are usually available from advertisers; where we anticipate supply difficulties a source will be suggested.

Back Numbers

Copies of most of our recent issues are available from: Post Sales Department (Practical Electronics), IPC Magazines Ltd., Lavington House, 25 Lavington Street, London SE1 0PF, at 75p each including Inland/Overseas p&p.

Binders

Binders for PE are available from the same address as back numbers at £3.75 each to UK or overseas addresses, including

postage and packing, and VAT where appropriate. Orders should state the year and volume required.

Subscriptions

Copies of PE are available by post, inland or overseas, for £10.60 per 12 issues, from: Practical Electronics, Subscription Department, Oakfield House, Perrywood Road, Haywards Heath, West Sussex RH16 3DH. Cheques and postal orders should be made payable to IPC Magazines Limited.

Market Place

Items mentioned are usually available from electronic equipment and component retailers advertising in this magazine. However, where a full address is given, enquiries and orders should then be made direct to the firm concerned. All quoted prices are those at the time of going to press.

by
David Shortland

AUDIO VISUAL

Two new AV machines have just been launched on the market. The 340 model is a cassette recorder with four identical audio channels, each with full metering facilities and Dolby Noise Reduction. Each channel has two inputs, microphone and line, a fixed level



line output, and a controlled output which can give in excess of +8dBm.

Independent record selection is provided for each pair of tracks to facilitate stereo recording (tracks 1 and 2) or all four tracks simultaneously. If required, tracks 1 and 2 can be recorded first then tracks 3 and 4 added whilst replaying tracks 1 and 2. Alternatively, tracks 1 and 2 can be recorded after tracks 3 and 4.

The 330 model has two audio channels with a separate sync track for audio visual applications.

The two audio channels have full metering facilities and Dolby Noise Reduction and each audio channel has the same specification as the 340. They sync channel also provides full metering facilities and selection for the use of normal and special tapes.

When the 330 is used with external sync pulse coding/decoding equipment, the sync channel has a single line input and a line output. It also has an internal sync pulse generator triggered by a button on the front panel to provide a recorded pulse for direct connection to a projector.

For further details including price and delivery contact: Neal-Ferrograph, Simonside Works, South Shields, Tyne & Wear NE34 9NX.

FLUKE DMM

Microprocessor techniques have allowed Fluke to incorporate some very useful features in their latest low cost 4½ digit 8050A DMM. This highly accurate bench/portable model with 39 measurement ranges and nine functions also provides unique dB computing and offset modes in addition to a high performance true r.m.s. capability.

In the dB mode, the 8050A DMM allows the user to call up any of 16 reference impedance levels from 8 to 1,200 ohms and to display the readings directly in dBs.

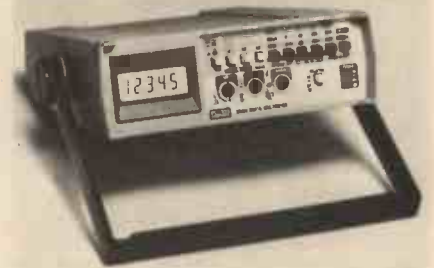
Additionally, a reference/offset mode allows any input signals to be stored either as a reference value for relative dB readings or as an offset against any reading. In offset mode, the user can zero-out any lead resistances for really high resolution impedance measurements or set up a reference offset and display

only the variance from that reading.

These absolute and relative dB modes with offset greatly simplify measurements in audio, amplifiers and telecommunications circuits as well as in production testing where only the variance from the stored value may be required. The offset facility is available on all functions such as a.c./d.c. volts or amps, resistance or conductance.

The high resolution 4½ digit l.c.d. display is matched by a basis d.c. accuracy of 0.03 per cent specified over a full year. A.c. or d.c. measurements can be made down to 10µV, 10 nA or 10 milliohms.

In addition to its volts, ohms and amps ranges, the 8050A also has two conductance ranges for high impedance measurements to



100,000 Megohms, as well as low power ranges for in-circuit measuring.

A wide range of accessories such as high voltage probes, current transformers, shunts, temperature and r.f. probes, remote hold probe, battery pack, and safety leads make the 8050A a complete measurement system for the bench or field.

The 8050A is priced at £199 ex. VAT and carriage.

Fluke International Corporation, Colonial Way, Watford, Herts WD2 4TT.

BOOTS VIDEO

All Boots audio departments are now selling the Ferguson VHS Videostar recorder (Model 3V00) which has a maximum recording time of 3 hours.

The 3V00 has a remote hand-held "pause unit" which can be used to edit unwanted material during recording or playback. Boots



will also carry a wide range of blank video tapes not only for the Videostar and other VHS recorders but also for machines using VCR and Beta format tapes.

Also available is a full colour catalogue of 45 pre-recorded programmes and feature films.

The price of the 3V00 is £595.00 with blank cassettes from £6.95 to £16.95 and pre-recorded films from £17.50.

ELECTRONIC NOTEBOOK

Like the first model introduced last year, the Toshiba LC-1038MN calculator offers alpha-numeric facilities which has now been expanded to a full ten digit alphabetic and/or numeric entry and display, plus sign. There is an independent 30-register memory bank, each register able to retain data consisting of 10-alphabetic characters and 10-numerals. The addition of a 26-location world-time display, plus clock/calendar and alarm. There is instant retrieval of the world-time display; the calendar indicates month, date and day of



week; and the alarm can be set up to one year in advance and to prompt an alpha display.

The calculator is ideal for the storage of information in constant use, but which is subject to frequent change such as currency exchange rates; share buying and selling prices; bank and audit card balances; train and plane departure and arrival times; stock levels; metric conversion factors; travellers cheque numbers.

The LC-1038MN operates on batteries lasting around 9,000 hours; is sized 70 x 10 x 137mm and weighs 80g (2.8oz). It is priced around £75, plus VAT.

SAFETY TESTER

Clare Instruments Limited have developed a portable safety test unit for checking double insulated and earthed electrical equipment.



The VI52 will flash test double insulated appliances at 4kV. Another noteworthy feature, is the high current used for earth bond testing (25A at 8V). Too many testers, at present available, offer only a simple continuity check for this important test.

The unit operates from a mains supply and the appliance under test is plugged into the 13 amp socket on the front panel of the tester. When the test button is pressed, neon lamps indicate that the series of tests is completed, with green lights for passes and red lights for failures.

The tester which is housed in a hardwood case is priced at £144.50 ex. VAT and p&p.

Clare Instruments Limited, Clare Works, Woods Way, Goring-by-Sea, Worthing.

NEGATIVE ION GENERATORS

Most of the air around us is electrically neutral with only a few of the air molecules gaining or losing electrons and therefore acquiring a positive or negative charge.

The effect of these charged molecules, which are called air ions, on individual health has been researched over many years and it has become clear that it is the balance of negative and positive ions rather than high ion concentration which is important.

Research has also shown that sufferers from asthma, hay fever and bronchitis can benefit from having the air they breathe charged with negative ions. There is also considerable evidence that ion depletion, such as



occurs in offices or crowded rooms, produces discomfort, drowsiness, fatigue and loss of mental and physical efficiency.

The concentration of negative ions in the air can be increased using a negative ion generator. Two such generators have just been introduced by J. P. Bell Ltd., the Kobelair Model 20 is a desk unit suitable for individual usage whereas the Model 40 is suitable for rooms up to 40 cubic metres (1500 cubic feet) with a maximum effectiveness over a range of 0.5 to 2.5 metres (2 to 8 feet).

The model 20 is priced at £57.50 and the Model 40 at £98.50 excluding VAT and p&p.

J. P. Bell (Machinery) Ltd., Jubilee works, London Road, Woolmer Green, Knebworth, Herts.

SUPERDECK FOR MPUs

The V&T Superdeck is a high speed cassette unit which will, under CPU control, find any file stored on a C60 cassette (one megabit of unformatted memory on each side) in under one minute using fast forward and fast reverse speeds (up to 50 i.p.s.).

Other features of the Superdeck include: 5000 baud CUTS format as standard (which can be switched to other lower speeds, error detection and correction (it will also write protected areas of tape containing hard errors), it will write into the first empty file, on tape or into any file number or name specified (unless that particular file is write protected in which case an error message is generated).

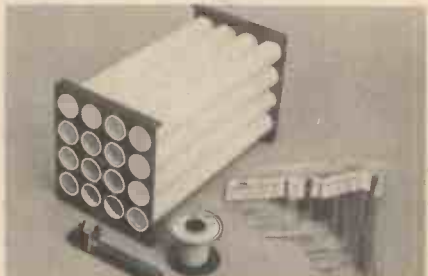
At the present time only a Z80 operating system is available.

The unit which connects to any 8bit I/O port and UART includes a 240V supply unit.

The price of the Superdeck is £110 plus VAT and p&p. V&T Electronics, 82 Chester Road, London N19.

WIRE BIN

The WB-16 wire bin has been designed for both storage and wiring table use, keeping assorted wires separated in their correct size groups and preventing wasteful tangling.



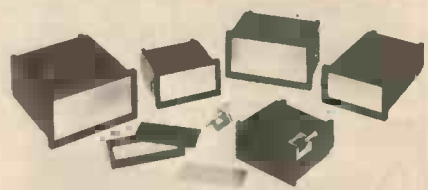
The sixteen wire storage tubes have adjustable depth stops to take wire lengths from 25mm and 350mm. The price of the WB-16 is £33.81 excluding VAT and p&p.

OK Machine & Tool (UK) Ltd, Dulton Lane, Eastleigh, Hants. SO5 4AA

PANEL MOUNTED CASES

A new range of panel mounted cases which are ideal for housing projects such as clocks, meters, counters etc., have just been introduced by Perance Ltd.

The black ABS cases are supplied with two front panels, one is an anti-reflective filter for displays and the other is anodised aluminium for mounting controls. Other features include



p.c.b. guides and slots, matt black clip-on bezels and optional mounting clamps for wood or metal panels.

There are four sizes of cases available, 96 x 48 mm and 72 x 36 mm front panels in two lengths 120 mm and 75 mm.

These cases could also make excellent housings for our Car Instrument Devices if the i.e.d. mountings were altered.

Distributors for the cases include Watford Electronics, Home Radio and Bi-Pak all of whom advertise in the magazine.

EPI

The 1978 edition of the Electronic Projects Index is now available. This index covers all the constructional projects published in sixteen magazines including PE, radio, television, hi-fi and computing. The projects are arranged under thirty six headings arranged alphabetically from aerials to Zener testers.

Other information given with each article includes a component guide and method of construction used, i.e. p.c.b., Veroboard etc.

The index is priced at £1.30 including p&p, and is available from M.L. Scaife, Central Library, Northumberland Square, North Shields, Tyne & Wear, NE30 1QU.

Market Place

ELECTRONICS 79

The Electronics 79 Show was held this year at Olympia and many of the leading UK component manufacturers and distributors were exhibiting there together with equipment, instrumentation and packaging companies.

VIEWDATA

The latest Viewdata adaptor from Labgear, the 7050 Viewdapta, was on display. This is a compact unit for receiving Viewdata which can be used with an existing TV set without modification to the receiver.

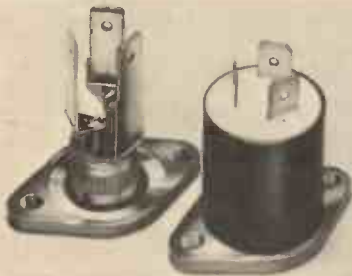


The 7050 consists of a desk top keyboard and a wall mounted Viewdata Processing Unit which are connected together via a multiway cable. A built in loudspeaker gives an audible indication of engaged, dialling and other tones obtained.

TRIACS AND THYRISTORS

Motorola were showing their range of 15 to 40A triacs which they have now fitted into a TO-3 base type package with push-on terminals.

These 200 to 800V triacs are primarily designed for full-wave control to a.c. loads and are electrically isolated from the mounting base with a high isolation voltage of 2500V. Applications include appliance controls, power supplies, solid-state relays, heating and motor controls.



Motorola were also displaying their 8A MCR 72 thyristor which needs just 30µA to trigger on and can easily be directly coupled with an MPU or other driver i.c. for power control applications.

CIRCUIT TESTER

Among the many instruments on display Vero Systems had their "Soundout" touch sensitive cable continuity tester on display which uses the body as a conductor leaving both hands free of probes.

The unit which is battery operated (9V) has a dual input impedance (15MΩ and 2.5kΩ) and is available in two versions with the MK II model having the additional feature of an earphone for noisy environments or where more than one unit is in use.



Both units have a volume control and are supplied with an interchangeable crocodile lead. A wrist strap and probe lead are available as accessories. The price of the MK I is £20.00 and the MK II is £24.00.

ZIP SOCKETS

With the high cost of LSI circuits many constructors are now using zero insertion pressure sockets to prevent pin damage to components whilst they are being inserted.

The BFI range of ZIP sockets included several new types never exhibited before. The socket range is one of the most comprehensive currently available and may be used for almost any device, including multi-pin integrated circuits axial lead components, power transistors and non-standard hybrid circuits.



Of particular interest is a new range of sockets for use with LSI devices in test equipment or development "breadboards". The sockets will accept 28 (0.4" pitch), 42 and 64 pin devices with no insertion pressure, thus eliminating the lead damage and distortion which can occur when leads are forced into spring contact sockets.

Each socket has a lever at one end which is connected to an internal cam. The device is simply dropped into the socket and the lever flicked up to positively clamp the leads inside the socket. This protects the leads from

damage and ensures a good electrical connection. When the lever is released the device may be removed without force.

PLASMA PANELS

Included in the Thompson-CSF display was their TH 7604 plasma-panel module which is intended for low-capacity alphanumeric-display applications.

This module which includes all the necessary panel-drive electronics has a power consumption of less than 30W. Being a.c. driven, the panel itself features inherent storage which means that the high-intensity display is also free of flicker,

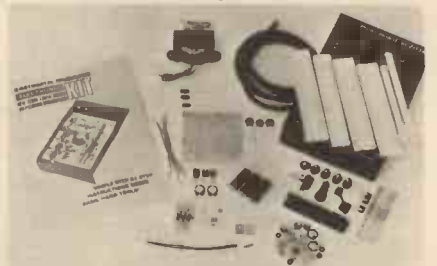


The module's overall dimensions are 295 x 125 x 57 mm. Its useful panel area (219 x 52mm) permits displaying up to 6 lines of 40 characters (5 by 7), with the possibility of adding underlining or a mobile cursor.

BREADBOARD KIT

The latest kit from the CSC is the proto-board PB-203AK. This kit, featured in a special pre-release offer in PE (Nov. 79), contains all the components needed to make a solderless breadboard unit with three regulated d.c. power supplies.

The kit comes complete with all the elec-



tronic components, case and breadboard modules, as well as nuts, bolts, connecting wire and solder. The assembly instructions have been written without any assumptions about the constructor's past experience.

The finished Proto-Board incorporates three large breadboards plus four long busbars and one shorter one, giving a constructional area sufficient for 24 integrated circuits in 14-pin packages. In addition, terminal posts allow connection to earth and to the +5V, 1A and ±15V, 0.5A power supplies. The power supplies are independent and fully regulated, and the ±15V supplies can be internally adjusted over the range 7-18V. The three power-supply rails allow the board to be used with most types of circuitry, including TTL and CMOS logic.

The PB-203AK is supplied with an earthed metal case measuring 248 x 168 x 83mm, and is still available for a short period at a special offer price of £55.00 excluding VAT and p&p.

DYNAMIC

NOISE

LIMITER

R.A. Penfold

ALTHOUGH compact cassettes have certain advantages over other forms of recording medium, they have the major disadvantage of a comparatively poor signal-to-noise ratio unless they are used in conjunction with some form of noise reduction system. Many pre-recorded cassettes are encoded using the Dolby system and are capable of excellent results, but there are still numerous pre-recorded cassettes currently available that are non-Dolby as are many cassettes purchased some time ago.

The dynamic noise limiter (DNL) which forms the subject of this article was designed for use with a high quality cassette deck to enable an improved signal-to-noise ratio to be obtained when playing non-Dolby cassettes. Of course, the unit can also be employed with a cassette deck which

does not incorporate any form of noise reduction circuitry when playing any cassette. It can also be used in addition to some other form of noise reduction system to further increase the dynamic range. This can effect a very worthwhile improvement when the original recording is fairly old, and a significant amount of tape hiss has been recorded onto the cassette. It is even possible to use the unit to improve noisy f.m. radio reception, or any other programme source that is affected by low level high frequency noise.

DNL PRINCIPLE

Tape noise consists mainly of high frequencies, or to be more accurate, it is this high frequency content that tends to be most noticeable and objectionable. Turning back the treble control while playing a cassette will show this quite clearly by apparently greatly reducing the noise. It is by reducing the upper frequency response during playback that a DNL effects a reduction in tape noise, but it only applies the full amount of treble cut at low signal levels. It is then that the noise is most noticeable.

At high signal levels the treble cut is removed to some degree, and at very high signal levels it is totally eliminated. The increased tape noise will not be audible as it will be masked by the main signal.

In this way a DNL provides a very significant reduction in background noise, but there is a minimum loss of treble response.

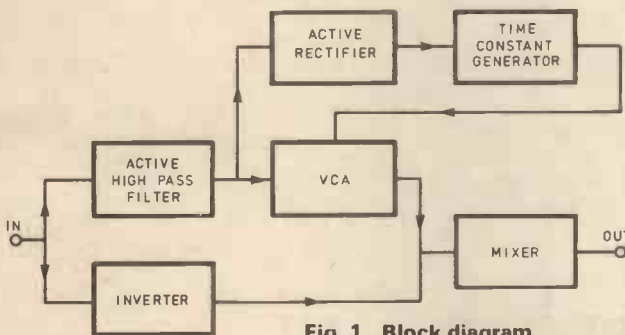


Fig. 1. Block diagram

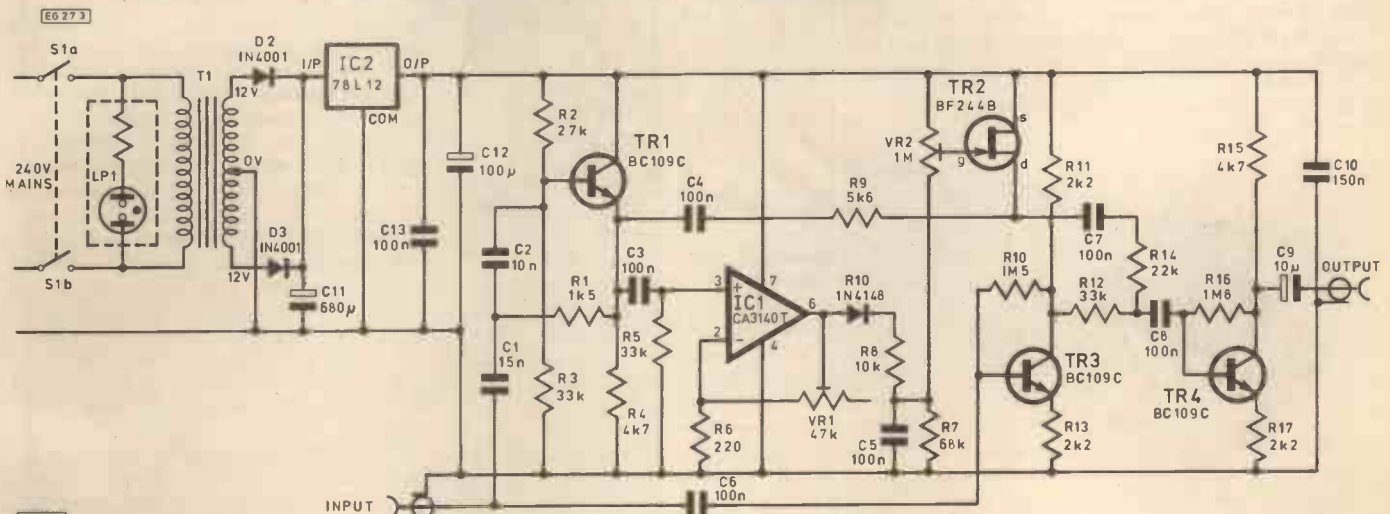


Fig. 2. Complete circuit diagram of the DNL

The DNL described in this article operates in the manner shown in the block diagram of Fig. 1. The input signal is split into two parts which are then mixed in a simple passive mixer circuit. One part of the signal is simply inverted before being fed to the mixer, and the other part is fed to the mixer via a high pass filter and a voltage controlled attenuator. The high pass filter rolls off signals below about 4kHz.

The two signals at the mixer are out of phase and therefore tend to cancel one another out. However, as only high frequencies are present at one input, it is only these high frequencies that are attenuated to a significant extent. Under quiescent conditions the VCA is adjusted to balance the two inputs and so optimise the high frequency cut. The output of the mixer is coupled to a low gain amplifier which compensates for circuit losses and provides the unit with almost exactly unity voltage gain.

Some of the output from the high pass filter is used to generate a control voltage for the VCA. An active rectifier and smoothing network are used to provide this voltage. Under low signal conditions only a very small voltage will be generated, and this will not greatly affect the circuit. Higher signal conditions will produce a large enough voltage to significantly attenuate the signal through the VCA. The higher the signal level, the greater the attenuation.

If the output from the VCA is reduced, the high frequency cancelling effect on the signal from the inverter will also be reduced, and so the higher the input signal level, the less the amount of treble cut that the circuit applies. Thus the desired circuit action is provided.

The reason the active rectifier stage is fed from the high pass filter rather than the main input is that high frequency signals mask the tape noise far better than low or middle frequencies do. The unit is therefore designed to respond more readily to high frequency input signals.

CIRCUIT DESCRIPTION

The circuit diagram for one channel of the DNL is shown in Fig. 2. TR1 is used as a conventional high pass filter of the type often encountered in rumble filters, but the component values have been modified to provide a much higher cut off frequency of course. TR1 is connected in the emitter follower mode and so the filter has approximately unity voltage gain at pass frequencies.

The VCA is formed by R9 and TR2, the latter being a JUFET which is used here as a voltage controlled resistance. TR3 is used as the inverting amplifier and this is a common emitter stage. As the emitter resistor (R13) is not bypassed and is equal in value to the collector load resistor (R11), this stage has almost unity voltage gain. R12 and R14 form a simple passive mixer circuit and TR4 is used as the output amplifier. TR4 only needs a voltage gain of a little



COMPONENTS ...

Resistors

*R1	1k5
*R2	27k
*R3	33k
*R4	4k7
*R5	33k
*R6	220
*R7	68k
*R8	10k
*R9	5k6
*R10	1M5
*R11	2k2
*R12	33k
*R13	2k2
*R14	22k
*R15	4k7
*R16	1M8
*R17	2k2

All resistors $\frac{1}{4}$ W 5% (10% over 1M)

Potentiometers

*VR1	47k hor preset
*VR2	1M hor preset

Capacitors

*C1	15n
*C2	10n
*C3	100n
*C4	100n
*C5	100n
*C6	100n
*C7	100n
*C8	100n
*C9	10 μ 16V elect
*C10	150n
C11	680 μ 25V elect
C12	100 μ 16V elect
C13	100n

All capacitors C280 type except where otherwise stated

Semiconductors

*IC1	CA3140T
IC2	78L12
*TR1	BC109C
*TR2	BF244B
*TR3	BC109C
*TR4	BC109C
*D1	1N4148
D2	1N4001
D3	1N4001

Switches

S1	Rocker switch d.p.s.t.
S2	Rocker switch d.p.d.t.

Transformer

T1	Mains primary, 12-0-12V 50mA secondary, miniature type with flying leads (M.E.S.)
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Miscellaneous

BEC case type GB 1a or similar
Four phono sockets
Miniature panel neon
p.c.b.

*Indicates that two devices are required for stereo operation

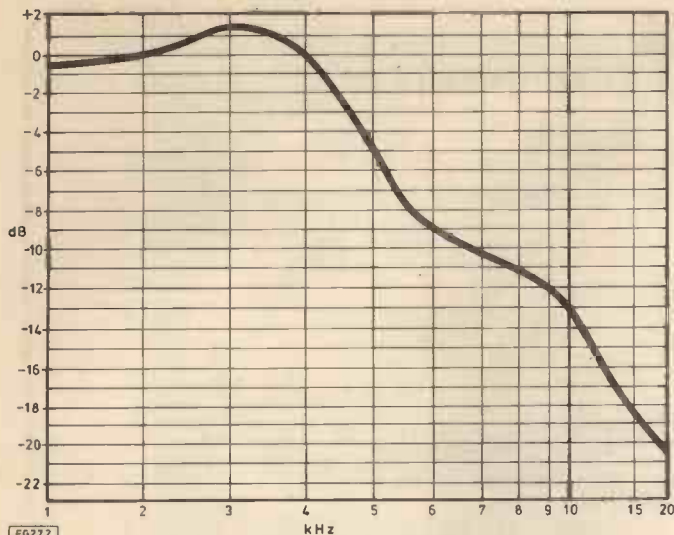


Fig. 3. Frequency response (low level inputs)

more than two in order to make good the losses in the circuit, and so again an unbypassed emitter resistor is used to introduce negative feedback and set the voltage gain of the stage at the required level.

IC1 is used as a form of active rectifier, and its voltage gain is controlled by feedback potentiometer VR1. This enables the input signal level required in order to remove the treble cut to be varied over a considerable range so that the unit is suitable for use with any normal cassette deck.

VR2 enables the bias on TR2 gate to be adjusted and this control is set to slightly reverse bias TR2 so that there is only a small amount of attenuation through the VCA under quiescent conditions, and the two inputs to the mixer are precisely balanced. Only positive going output signals are produced from the active rectifier in the presence of a suitable input, and these have the effect of raising the voltage at the lower end of VR2 and removing the reverse bias from TR2. This increases the attenuation through the VCA and removes the treble cut.

C5 integrates the output pulses from the rectifier in order to prevent significant distortion, but the attack and decay times are both short so that the circuit quickly responds to changes in dynamic level. As is normal for this type of device, the circuit has hysteresis. This simply means that the attack time is faster than the decay time, which helps the unit to act as fast as possible without generating significant distortion.

The hysteresis is produced by D1 and R8. C5 is charged through the relatively low impedance of R8, but it cannot discharge through the same path since D1 prevents this. Instead it must discharge into the relatively high impedance of R7. This is the only reason for including D1 in the circuit since the CA3140 used in the IC1 position is only operated from a single supply rail, and will provide a rectifier action without D1.

With a high level input the frequency response of the unit is virtually flat over the audio spectrum, but on low level in-

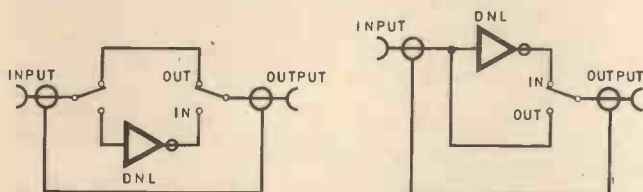


Fig. 4 a & b. Switching alternatives

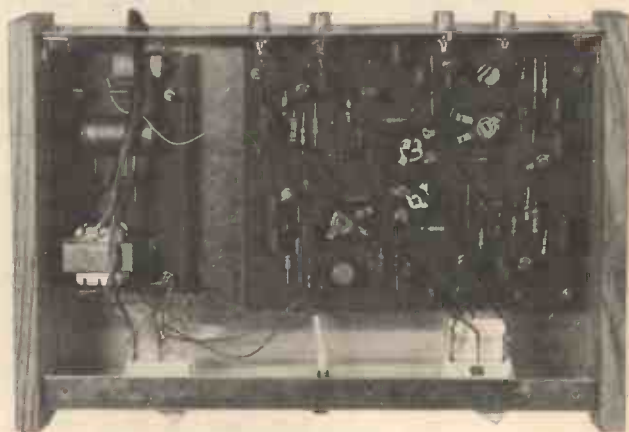
puts the frequency response of Fig. 3 is obtained on the prototype. The attenuation rate is about 12dB per octave from 3.5kHz to 8kHz. There is a slight peak in the response at approximately 3kHz and the roll off rate above 8kHz is only very gradual, but neither of these points are of any real consequence in practice.

P.S.U. AND SWITCHING

The filter is powered from a simple stabilised mains p.s.u. This is quite straight forward and consists of a push pull rectifier and smoothing network feeding a monolithic regulator i.c. A well smoothed and regulated output is provided, and the regulator i.c. has output current limiting and thermal overload protection circuitry.

In most cases it will be desirable to have some means of switching the unit out of circuit so that the cassette deck can be used normally. Some amplifiers and receivers have built in switching in the form of a tape monitor facility, or something of this nature, which could be used to accomplish this. An alternative is to use the simple in/out switching arrangement used on the prototype and shown in the circuit of Fig. 4a. This leaves the filter input permanently connected to the cassette deck output, but this does not seem to affect performance even if the cassette deck is used while the filter is switched off.

If preferred, the system shown in Fig. 4b can be used. Here the filter is switched completely out of circuit when the



Internal view

switch is in the "out" position, but a four pole switch is required. It would be possible to gang the on/off and in/out switches, but if this is done care must be taken to ensure that mains hum is not picked up in the in/out switch wiring. Also, care must be taken to ensure that the mains wiring cannot accidentally come into contact with the input or output wiring!

CONSTRUCTION

The prototype unit was housed in a BEC cabinet but any case of mainly metal construction and having similar dimensions (230 x 150 x 50mm) should also be suitable. The filters and power supply are constructed on two separate p.c.b.s and these are shown in Figs. 5 to 8 respectively. The negative supply rail connection between the two panels is carried via the mounting bolts and the metal case. The earth connections between the filter p.c.b. and the input and output sockets is obtained in the same manner. The filter and p.s.u. boards should be mounted as far apart as possible.

ADJUSTMENT AND USE

In use the filter is either connected between the cassette deck output and the "tape" input of the amplifier, or it is

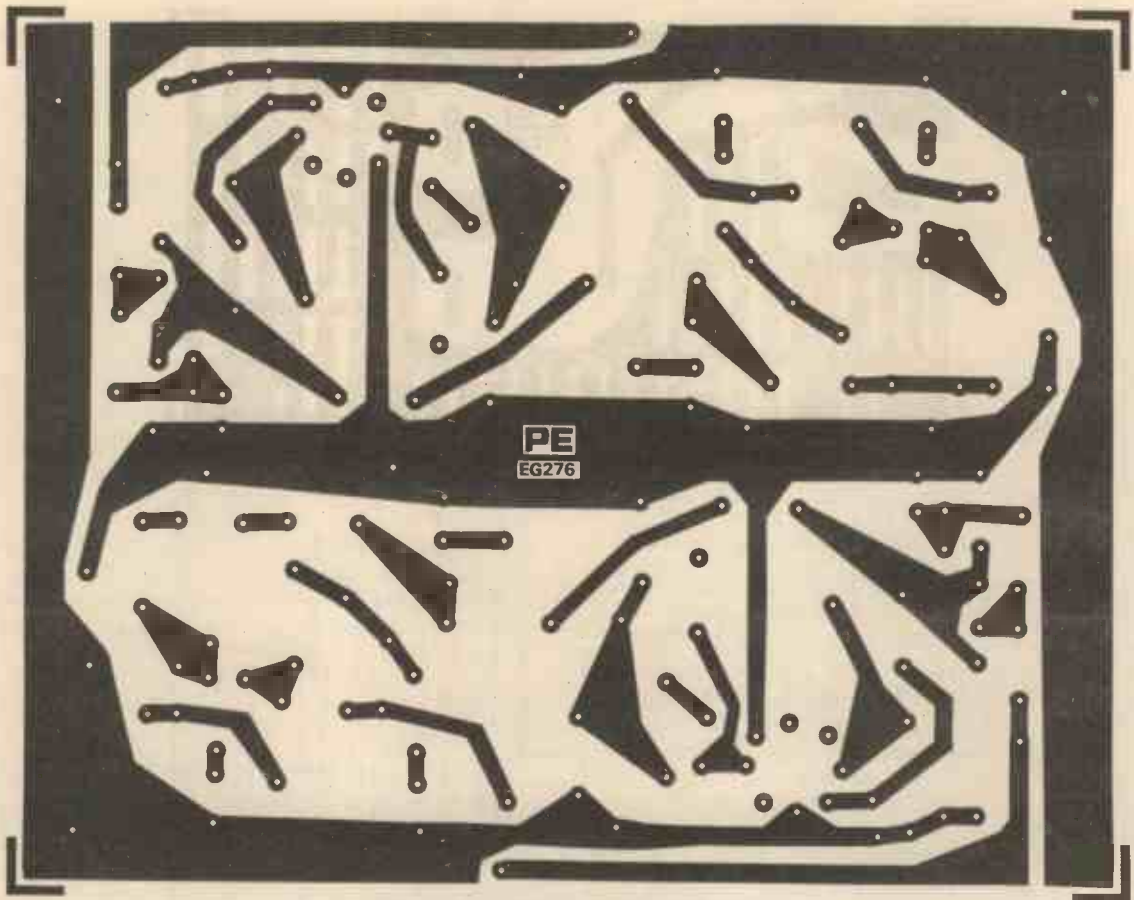
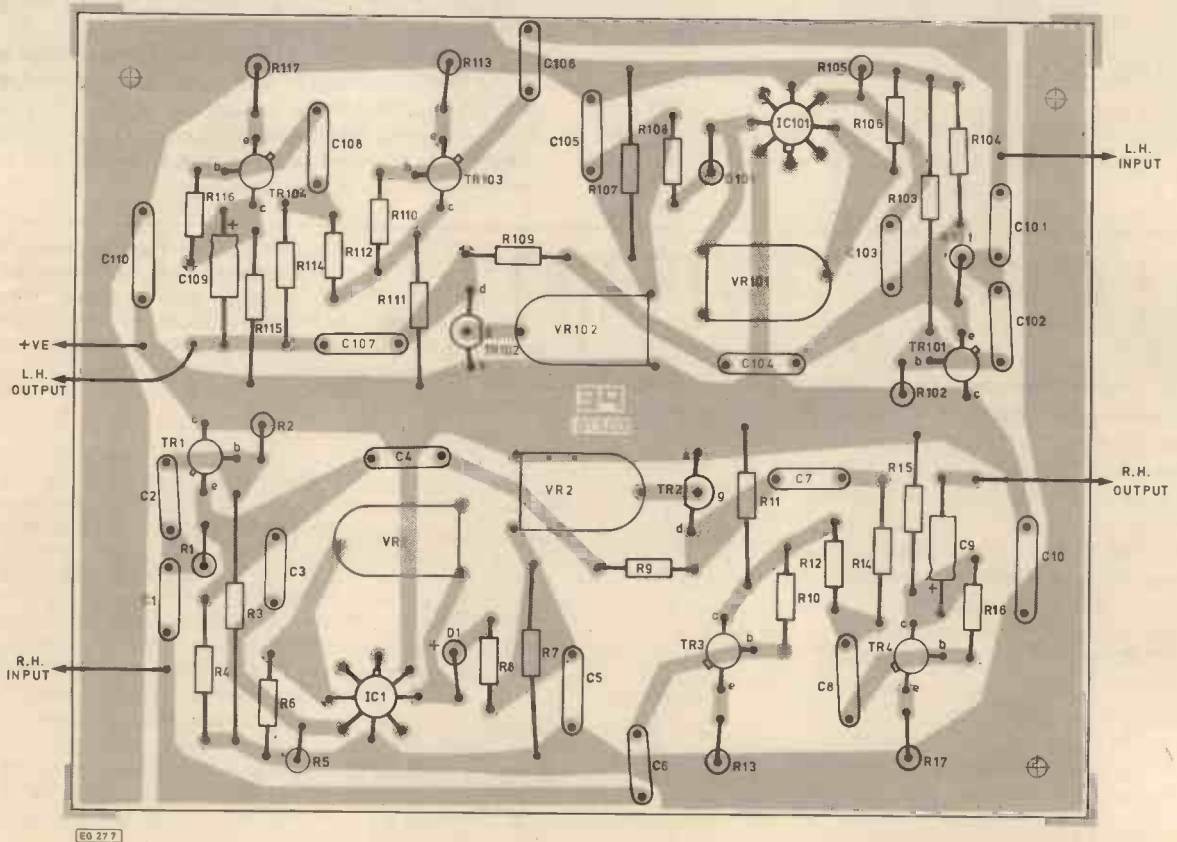


Fig. 5. P.c.b. design for the main board

Fig. 6. Component layout



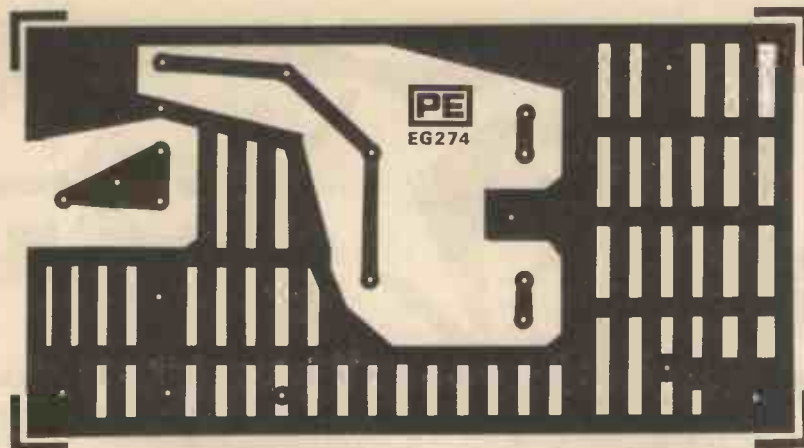


Fig. 7. P.c.b. design for the p.s.u.

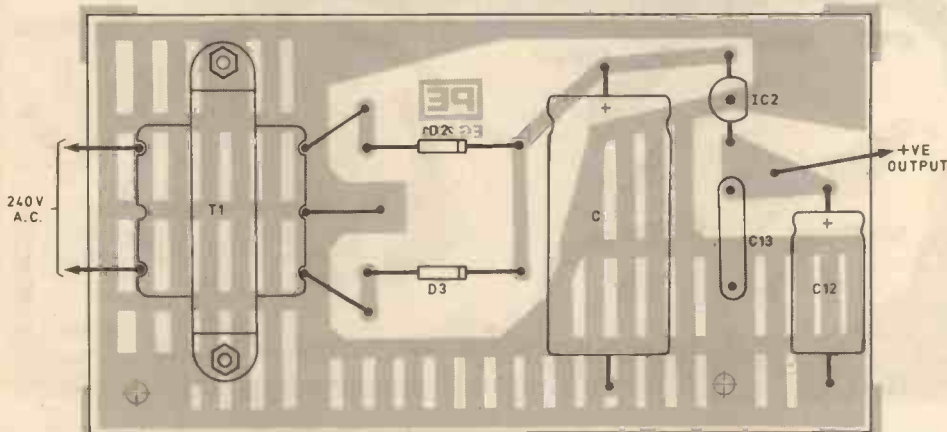


Fig. 8. Component layout

connected into the tape monitor or some similar facility. In either case both the input and output leads should be proper screened types.

Initially VR2 is adjusted in an almost fully clockwise direction and a blank cassette is played through the system to provide a source of tape noise. If VR2 is now adjusted in an anti-clockwise direction a point should be reached where there is a slight but noticeable null in the noise, with further adjustment causing a rise in the pitch and intensity of the

noise. The correct setting for VR2 is at this null point and both channels are adjusted in this way.

The best setting for VR1 can be found by trial and error. The further it is adjusted in a clockwise direction, the lower the input signal level at which the treble cut starts to be lifted. It should therefore be adjusted as far in a clockwise direction as possible without the lifting of the treble cut becoming audible on low level signals. ★

MICRO PROMPT.

The hardware and software exchange point for PE computer projects

If you have ever built a computer, or computer peripheral using plans from the pages of PE, then it is likely that you would have appreciated some "follow up" data on such matters as interfacing and software. Keep an eye on this new column then, if you have constructed Champ, the UK101 or a peripheral device such as our printer or VDU; for it is from this column that we hope hints and snippets of information, short programs, applications and discoveries will spring forth eternal... with your help of course!

Sir—Having recently built the Compukit 101 computer, I must say that I am

delighted with it, although it has its limitations (one small annoying one being the lack of PRINT USING).

I find that I am learning BASIC by using the computer (hit and miss), and am following your article on learning BASIC.

However, as far as my programming is concerned, one or two things have so far eluded me, and I hope that you or your staff can help.

(a) What are the POKE commands to switch to printer, and to VDU, when writing programs in BASIC.

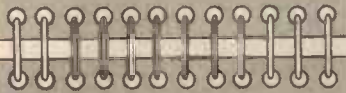
(b) Having written and loaded a program which has variables giving running totals, how can I load and save these updated programs? (I have a feeling that machine code, POKE, or something of the like is involved but what?) (I have already worked

out that to run these programs, RUN sets the variables to 0, and I have to use GOTO 1 to run them without clearing the variables. The books on BASIC I have do not mention these things, which for a real beginner are a nightmare).

Iain Corrance
Glasgow

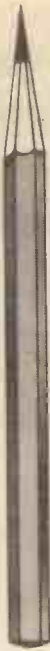
Ideally, to carry data over, and update it at will, a cassette file should be generated so that the operating program is recorded on one tape and a data file on another. Alas, the 101 does not have cassette file firmware, but we have been told it can be done using machine code routines which of course can be accessed from BASIC via USR(I). An alternative, if somewhat clumsy, is to list as DATA statements those variables you wish to maintain, before re-recording the program.

We shall endeavour to answer all questions in due course. Next month we will be publishing a game for the 101 called "Torpedo Run", and give details of an independent UK101 users club.



INDUSTRY NOTEBOOK

By Nexus



God and Virtue

How do the Japanese do it? Have they some secret formula denied to the rest of the industrial world? There have been many attempts at analysis.

In two recent commentaries one suggests that the secret is that to a Japanese company and its employees the customer is regarded as God. Actually, not an attitude to be despised although strange for a country whose religious beliefs are principally in Buddhism which has no God and Shinto in which the object of veneration in the shrine may be anything from a stone to a household utensil in which resides the spirit.

The other preaches a virtuous circle of good industrial relations allowing management to get on full time with managing and innovation, coupled with lifetime jobs generating confidence in accepting new technology by employees instead of, as in the UK, resisting it.

In all the years of *Industry Notebook* I never imagined I should use such a heading as that above! Nonetheless, such abstract concepts are very real to the Japanese character with their singing of company songs before taking to their work benches, and their dedication to a cause as exemplified by the terrifying suicide attacks by Kamikaze pilots in World War 2.

Sting Ray

Marconi Space and Defence Systems have won their biggest ever contract in the Sting Ray project. More than £200 million in the first instance and probably worth £800 million over its lifetime of 20 years. Among the other heady statistics, 50,000 man years of work spread over a dozen cities (if you include the subcontractors) make good reading.

Sting Ray is a lightweight homing torpedo with an advanced on-board computer, a multi-mode, multi-beam sonar which detects and tracks the target while resisting false echoes and enemy countermeasures, and a quiet yet high speed propulsion system. Deliveries to the Royal Navy should start in the early 1980s and there are expectations of it being adopted by other NATO navies.

Energy

The energy crisis persists but it is now almost a way of life, the previously unthinkable £1 per gallon being now a matter of history. The Americans have a conservation policy well reflected in energy-saving campaigns in manufacturing plants strongly re-inforced by publicity in house journals. But are they taking it seriously? I had my doubts when I read in one house journal from an electronics company that temperature in the offices and factory was being maintained at a steady 78°F. Surely a misprint, but then I read that the company dress code had been relaxed to permit jackets off and ties loosened. Slowly the penny dropped. This was the Texas summer, not the UK winter. Their problem is cooling as well as heating. As much as 20

per cent is being saved by some companies by simple expedients and even further gains by the addition of MPU-controlled fine tuning of air conditioning.

Paradox

The EMI Scanner, according to the medical profession, was the greatest advance in medical diagnostics since the Röntgen's discovery of X-rays in 1895. Its creator, Godfrey N. Hounsfield, had engineering ambitions when he was a five year old tot, yet he was a late developer who didn't complete his formal engineering education until he was 30. But his most useful period of training, on his own admission, was self-tuition between the ages of 13 and 19—an encouragement to every hobbyist!

Few people could ever have received so many awards and honours for a single invention, culminating in a Nobel Prize and even more recently the 1979 Aachen and Munich Prize for Technology and Applied National Sciences. And he deserved every one of them as well as the gratitude of the sick.

What a fantastic success story for the man and, initially, for the product which up to the mid 70s was a jack-pot profit maker for EMI. Then came the big cut backs in procurement in the USA, then the biggest market, and an increasing flood of competitive equipment. As the honours poured in for Godfrey Hounsfield the debts poured in for EMI.

Bad organisation, bad management, bad luck? It is difficult to analyse, no single factor being responsible. Hounsfield's great technological achievement still stands, whatever the fortunes of his employer. All the same, one imagines he must be looking over the past five years with very mixed feelings.

A similar story, perhaps less dramatic is the scanning electron microscope, a world-beater for Cambridge Instruments, a company now in the protective embrace of the National Enterprise Board but still losing millions on a good product. And older readers may remember the British company Perdio who bravely produced the first transistor portable radio in the world before going broke.

Does it pay?

Does it pay to be first in the field? There is much to be said for the "me-too" philosophy which, put crudely, means letting someone else bear the major development costs and carry the burden of breaking open the market, then moving in at the right moment saying "me-too". I recall Motorola, in its early and middle period in semiconductors, regarding Fairchild as the innovator, TI as the market developer and themselves as the "me-too" company with special skills in manufacturing coming in later with fast delivery and keen prices.

On the bid front Decca looks even more vulnerable. Thorn, at the time of writing, was still battling for EMI and GEC for Avery, their bid having been upped to a £98 million package.

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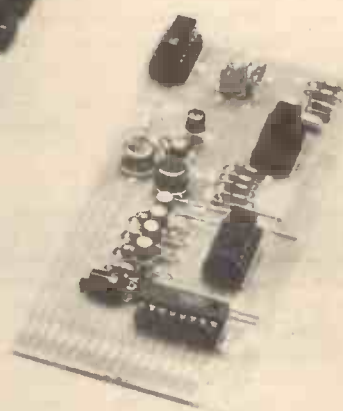
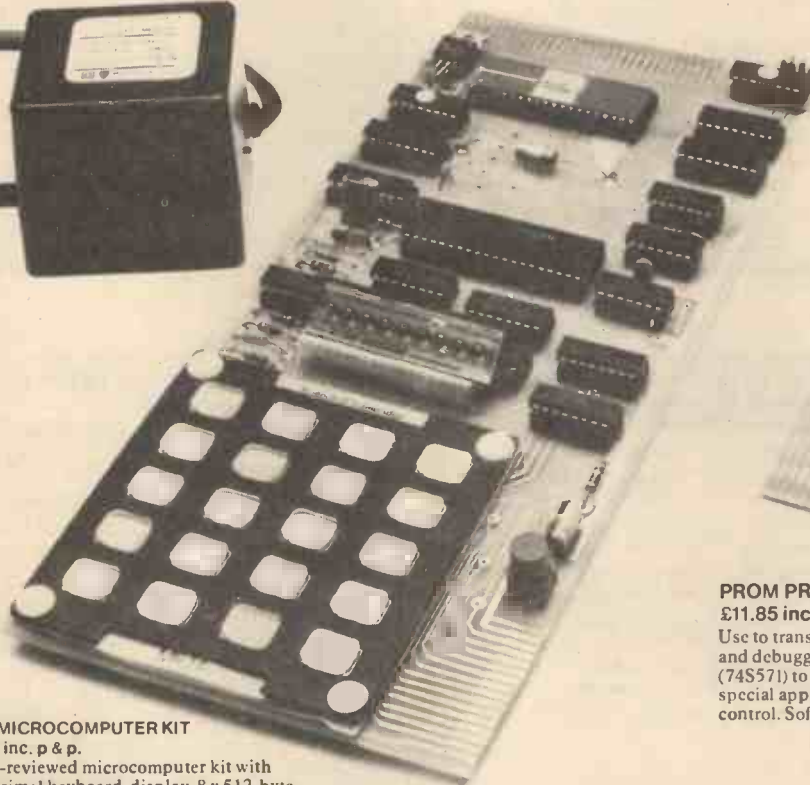
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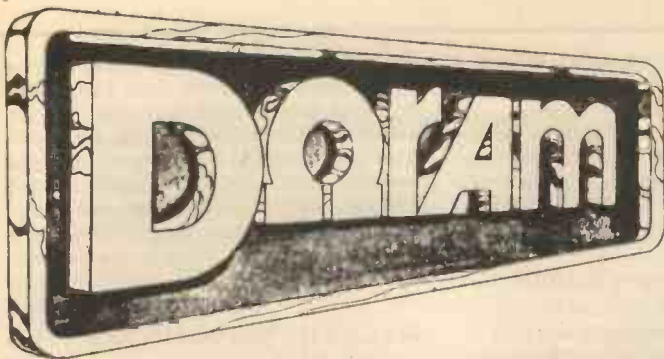
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SPACEWATCH

FRANK W. HYDE

DEATH OF A SATELLITE

One of the most remarkable satellites launched in recent years is the United Kingdom, ARIEL V. It has enjoyed 5 years of great activity. When it was originally launched it was not expected that its operable life would be much more than a year because the propane fuel thrusters would run out of gas. This would deprive the control centre of means to "point" the satellite at suitable targets. The amount of data accumulated for the estimated period alone would have been a major contribution. In fact the satellite has been operational far beyond expectations. This has been due to the expertise developed by the team at the control centre, the Appleton Laboratory near Slough at Ditton Park.

Among other things this team developed a technique which enabled them to economise the gas use thereby adding years to the useful operation. This in itself was a magnificent contribution. Even when the gas was finally exhausted, use was made of the Earth's magnetic field to maintain some control. In all this the team have been very modest. This official statement of the record of this mission is extremely restrained:

"The success of Ariel V is twofold: Outstanding scientific observations have been made by Ariel V and have established the UK group at the forefront of X-ray Astronomy.

"The technical expertise and knowhow in the fields of satellite and data management which have been built up in the Control Centre at the Appleton Laboratory are second to none and invaluable for subsequent projects, and could be an important factor in the UK involvement in future international space programmes."

Ariel V was launched on the 15th of October 1974 to make observations of cosmic X-ray sources. With 5 years of observations completed it is clearly the most successful UK satellite to date, both operationally and scien-

tifically. The spacecraft, designed and constructed for SRC by Marconi Space and Defence Systems, contains six scientific experiments provided by the Mullard Space Science Laboratory, of University College London, The University of Leicester, Imperial College, London and the Goddard Space Flight Centre of the US.

TECHNICAL ACHIEVEMENTS

The establishment of the Ariel V Control Centre at Slough in 1974 was the beginning of a major initiative in the area of satellite control and space data support as a complement to the established space project management. It is being further improved for support of the Infra-Red Astronomical Satellite (IRAS).

The staff responsible for the design of the Ariel V covered the following expertise. Electronics, real time software, data communications, celestial mechanics, operational research and large scale data processing. The electronics were needed for the building of the special computer interfaces and data display equipment; the real time software knowledge was needed for the satellite control and data reception; data communications skill was required for the setting up of links between Control Centre real time computers, the Laboratory mainframe computer, back-up computers and computers at four University sites. The celestial mechanics knowledge was required for the design of orbit prediction programmes, attitude computation and observation programmes; O.R. techniques were used to optimise the gas control, finally data base on a large scale was needed for preprocessing of the satellite data.

The Control Centre had two objectives:

1. To control the spacecraft and its experiments and, by monitoring the critical satellite sub-systems, keep the satellite in a safe and healthy state.
2. To interact rapidly with the astronomers.

FLEXIBILITY

Both these objects were achieved successfully. The second objective was however the most successful for it had a major impact on the programme. The key was the observing programme which was so flexible that astronomers were able to interact with the Control Centre and alter plans within an hour or so. The system operates as follows. The prime Ground Station for Ariel V is the NASA Ground station at Quito in South America. The satellite passes over Quito every 90 minutes. During a satellite pass commands are sent to the satellite and data received from it on-line in the Control Centre. As soon as the raw telemetry data are received they are checked and re-formatted. Orbit and Attitude computations are carried out and data tapes created for each experimenter. These pre-processed data are then transmitted by Post Office lines to Birmingham, Imperial College, Leicester and the Mullard Space Science Laboratory. Data will be received within 45 minutes of the satellite passing over Quito. Thus it is feasible for the experimenter to examine his on-line data and request a change in the observing programme *in time for the next pass*. This important feature was a basic requirement for the Control Centre because of

the nature of the experiments. Frequently one of the two scanning experiments would detect an X-ray source which was new. The versatility of the control would enable the satellite to be manoeuvred so that the four pointing experiments could examine this source in detail. The value of this facility is very great for the reason that many sources are transient and in a few hours could fall in brilliance below detectable level.

MAGNETIC MANOEUVRES

In the early days the attitude of the satellite was controlled by the propane gas jets at the base of the satellite. The gas was expected to last a year, in the event it was prolonged by nearly three years by the ingenuity of the control team. The gas finally ran out in June 1977. Since then attitude has been controlled by using a magnetic dipole cancellation system. This was of course not installed for that purpose but to prevent magnetic drift. In a sense it could be said that it was used in reverse. Since the satellite has an equatorial orbit it passes through the field lines of the earth's magnetic field. By passing a small current round the loop the earth's field was used to control the attitude of the satellite. The torques made available thus, were able to achieve manoeuvre rates of $10^\circ/\text{day}$ in right ascension and $0.5^\circ/\text{day}$ in declination.

The death of Ariel V is expected between the middle of January and February 1980.

THE EXPERIMENTS

Project Scientist: Professor A. P. Willmore, Department of Space Science, University of Birmingham.

A. University College London's Mullard Space Science Laboratory, Holmbury St. Mary, Dorking.

Measurement of X-ray source positions and sky survey in the energy range 0.3 to 30keV. This pointing experiment uses a rotation modulation collimator with proportional counters and channel electron multipliers as X-ray detectors, together with a collimated photomultiplier for optical star detection. **Principal Investigator:** Professor R. L. F. Boyd.

B. Physics Dept., University of Leicester, University Road, Leicester LE1 7RH.

Sky survey in the energy range 1.5 to 20keV. This scanning experiment, which views from the side of the spacecraft, uses large proportional counters. **Principal Investigator:** Professor K. A. Pounds.

C. UCL Mullard Space Science Laboratory.

Study of the spectra of individual sources in the 2 to 30keV range. This is a pointing experiment using a proportional counter. **Principal Investigator:** Professor R. L. F. Boyd.

D. Physics Dept., University of Leicester.

Measurement of the polarisation of X-rays from 1.5 to 8keV. This pointing experiment will determine polarisation or line emission by means of a Bragg crystal spectrometer using moveable plane crystals. **Principal Investigator:** Professor K. A. Pounds.

F. Physics Dept., Imperial College, London SW7.

The study of sources of high energy X-rays up to 2.0MeV. This pointing experiment uses an active collimator, a caesium iodide scintillator and photomultiplier detectors. It is designed to investigate the energy spectra and time variations of known sources and to measure the spectrum of the diffused cosmic X-ray background. **Principal Investigator:** Professor H. Elliot.

G. Goddard Space Flight Centre, Greenbelt, Maryland, USA.

An all-sky monitor in the energy range 3 to 6keV. This survey experiment uses two pin-hole sensors to detect transient effects in the X-ray sky enabling the other experiments to make the earliest possible measurements on these important phenomena. **Principal Investigator:** Dr. S. S. Holt.

ARIEL 6

Ariel 6 which followed Ariel 5 has also a lifetime of about one year. It has not so far achieved full design operation because the control system has been subject to interference from both Russian and American radar systems. The effects are varied but one

troublesome one is that the satellite switching is corrupted. However valuable data is still being acquired.

THE MAGNETIC FIELD SATELLITE (MAGSAT)

MAGSAT is for the purpose of studying the movements of the Earth's crust and the location of mineral deposits. It is a polar orbiting satellite, and is designed to measure the Earth's near surface magnetic fields and indirectly crustal features related to earthquakes. During the life-span of this satellite which is 120 days it will measure the strength and trace the direction of both the global magnetic field and surface magnetic fields. The surface magnetic fields are caused by the electric currents effect of magnetic storms and by certain elements in the Earth's crust. For example one large magnetic surface anomaly is due to an iron deposit in Africa. This was detected by a satellite magnetometer. After making this survey it will be possible to make available maps of the Earth's surface showing the location of magnetic irregularities and a global magnetic field model.

Having completed the surface irregularities it will be possible to gain a better understanding of the evolution of the crust of the Earth and the various geological processes

which have led to the formation of ore and petroleum deposits. This satellite possibility has only recently been understood when satellites, carrying magnetometers for the measurement of the global fields, mainly due to the Earth's core and working in low orbits showed anomalies which were localised.

It has been known for many years that some rock formations are better conductors of electricity than others and that these played a part in indicating where mineral and petroleum deposits could be found. What was a surprise was that satellite location was possible. The data produced by six satellites carrying magnetometers has led to this special satellite experiment. It is hoped also to shed more light on the reversing of the Earth's magnetic field over cycles of time. It is known now that the field suddenly collapses and slowly rises but in the opposite mode where north becomes south. The building up may take a longer time perhaps a thousand years. There are many consequences of such a happening. It could be responsible for sudden destruction of species of animals or other life.

The clues to drift of continents may help to anticipate earthquake events when the magnetic and gravitational anomalies are better understood. This project will involve most countries in the world over the next two years.

Countdown

Organisers: Please send details of exhibitions, club open days and other events to Mike Abbott at least six weeks in advance. Inclusion will be subject to space etc.

BEX Feb. 20-21. Pavilion Bournemouth. **K**
IEA/Electrex Feb. 25-29. National Exhibition Centre, Birmingham. **I**
Viewdata March 26-28. Wembley Conference Centre, London. **O**
Computer-Aided Design (conference & exhibition) March 31-April 2. Metropole, Brighton. Details: CAD 80/0483-31261
Seminex April 14-18. Dept. Physics, Imperial College, London. **HI**
Communications 80 April 14-18. National Exhibition Centre. **I**
Electronic Test & Measuring Information April 22-24. Wythenshaw Forum, Manchester. **T**
International Conference On The Electronic Office April 22-25. London Penta Hotel. Organised principally by the Institute of Electronics & Radio Engineers. 99 Gower St., London WC1E 6AZ
North Midlands Mobile Rally April 27. Drayton Manor Park, Tamworth, Staffs. Details: Norman Gutteridge, 68 Max Rd., Quinton, Birmingham.
All-Electronics Show April 29-May 1. Grosvenor House, London. **E**
The Mersey Micro Show April 30-May 2. Adelphi Hotel, Liverpool. **O**
Compec Europe May 6-8. Centre International Rogier, Brussels. **L**
Great British Electronics Bazaar June 20-22. Alexandra Palace. **E**
Intel Fair June 24. Wembley Conference Centre, London. **U**
Tempecon July 1-3. Wembley Conference Centre. Exhibition devoted to temperature control & measurement. **T**
Transducer July 1-3. Wembley Conference Centre. **T**
Microsoft (symposium) July 7-10. University of Sussex. **S1**
The 1980 Microcomputer Show July 10-12. Royal Lancaster Hotel, London. **O**
Avionics (symposium) Sept. University of Surrey. **S1**
Harrogate International Festival of Sound Aug. 16-19 (18 & 19 trade). The Exhibition Centre + hotels. **X**

- E** Evan Steadman, 34-36 High st., Saffron Walden, Essex. ☎ 0799 22612
- HI** Seminex Ltd., 79 High st., Tunbridge Wells, Kent. TNI 1XZ. ☎ 0892 39664/5
- I** Industrial Trade Fairs, Radcliffe Ho., Blenheim Court, Solihull, W. Midlands B91 2BG. ☎ 021-705 6707
- K** Douglas Temple Studios, 1046 Old Christchurch Rd., Bournemouth, Dorset BH1 1LR. ☎ 020 20533
- L** Iliffe Promotion, Dorset Ho., Stamford St., London SE1 9LU. ☎ 01-261 8437/8
- O** Online Conferences, Cleveland Rd., Uxbridge, Middx. UB8 2DD. ☎ 0895 39262
- T** Trident International Exhibition, Abbey Mead Ho., 23a Plymouth Rd., Tavistock. Devon PL19 8AU. ☎ 0822 4671
- U** Brian Crank Associates, 58 London Rd., Southborough, Kent. ☎ 0892-31812 38414
- X** Exhibition & Conference Services, Clarendon Ho., Victoria Ave., Harrogate, Yorks. ☎ 0423-62677
- S1** Society of Electronic & Radio Technicians, 57-61 Newington Causeway, London SE1 6BL. ☎ 01-403 2351

News Briefs

SEASIDE COMPUTER CLUB

A COMPUTER club in the Bournemouth area has successfully taken off, the first meeting of which was held at the Poole Arts Centre on October 28th last year.

The response was very good and the club looked all set to form a committee. A second meeting was organised for November 30th at Kinson Community centre, where it was expected that a representative from Tandy would demonstrate some equipment.

Details from Robin Pink, 10 Harbour View Road, Poole, Dorset.

Semiconductor UPDATE...

FEATURING

ZN419CE

87C48

MCM68732

R. W. Coles

MODEL CIRCUIT

If you are interested in the radio control of model planes, ships, cars or robots, then you will be very pleased with the Ferranti ZN419CE device because it could save you money on servos. For the uninitiated, servos are used in radio control models to position control surfaces and other mechanisms. Their great advantage over the now rather outdated "on-off" escapement actuator is that they provide a proportional control action which is not jerky.

The R/C modeller controls his creation by adjusting a joystick potentiometer which causes the transmitter to send out a pulse width modulated signal where the width of pulses is proportional to the deflection of the control. If more than one control function is required, then the associated pulse width signals are time multiplexed and sent to the model in a repeating sequence.

At the receiver the separate channel signals are recovered by demultiplexing, and it is the function of the servo to respond to the changing width of individual channel pulses by producing at its output an angular change proportional to the joystick deflection at the transmitter. The servo output is linked to control surfaces or engine controls, with the result that the model responds in a smooth, proportional manner to every new control setting selected by the modeller.

The clever bit, as if you hadn't guessed, is turning those variable width pulse signals into a proportional mechanical movement, and that is where the Ferranti ZN419CE comes in. The motive power for the servo is usually provided by a small d.c. motor linked via gearing to the output shaft or lever, and to a variable potentiometer which rotates in synchronism with the servo output. The pot is there to provide information on the present servo output setting, and the pulse signal is there to tell the servo system what new setting is required. Inside the ZN419CE there is a monostable which has to be connected to the servo-pot so that the monostable pulse width is proportional to the pot position. A pulse comparator in the chip compares the monostable pulse width with the received pulse width for the channel, and the result is an error pulse which is stretched and used to turn on one of a pair of external *p.n.p.* drive transistors which control the motor current.

The motor turns to drive the servo pot (and of course the servo output shaft) until the monostable pulse width is equal to the

received pulse width, whereupon the motor stops.

A great deal of "electrickery" is necessary to produce a practical servo decoder and driver, but fortunately you can get most of it in the 14 pin package of the ZN419CE.

Complete servo modules are already available for R/C modellers of course, but by using the ZN419CE in a home-built servo it should be possible to save money. It would also be possible to use this device in "special" servos, just what you need, perhaps, for that towering microprocessor based robot you've been working on in the garage!

CMOS SINGLE CHIPPER

Intel were one of the first in the field with a true single-chip microprocessor CPU, RAM and ROM all in a single 40 pin package. Their original 8048 and 8748 devices have now been joined by a whole family of similar devices such as the 8021, 8022, 8035, 8039, 8041 and 8741 featuring various combinations of memory and input/output lines.

Most of these devices are of little interest to hobbyists because they utilise mask-programmed ROM to hold their programs, but those with a "7" as their second digit are useful, even for one-off projects, because they employ program memory of the erasable and reprogrammable (EPROM) variety.

The 8748 for example, although intended as a prototyping aid for the pin compatible 8048, can be very useful where the small size of a single chip system is desirable. It contains within its 40 pin package a digital processor system complete with 1K of EPROM, 64 bytes of RAM a programmable counter/timer and no less than 27 lines of I/O.

Hats off to Intel then for a useful workhorse, but it is not the 8748 I want to tell you about because it isn't really new. The 87C48, however, is very new and it comes from Intersil, the CMOS wizards of Cupertino, California. It seems that Intersil have an agreement to second source the highly successful Intel device, but although identical to the 8748 in most respects, the 87C48 adds that magic ingredient, CMOS technology, to make a good thing even better. It runs from a 5 volt supply, just like the NMOS 8048 but the Intersil device consumes a maximum of 50mW against well over 300 mW for the Intel part at the same 6Mhz clock frequency.

Rumour has it that the 87C48 will actually cost less than the 8748 when it starts to appear on stockists shelves early in 1980, a fact which could make it very attractive for battery powered hobby projects. One problem though, the 87C48 uses a different EPROM technology to that of its Intel cousin and cannot therefore be programmed by using a standard 2716 programmer and a pin-out adapter. Putting together a special programmer circuit may prove to be well worth the effort, however!

HALF A LOAF

Talking of EPROMs, you may already know about the current world shortage of devices such as the 2716, an effect which has apparently been caused by the popularity of 5 volt EPROM technology in general, combined with a long delay before the appearance of alternative sources to challenge Intel.

Some second sources have appeared however, notably Texas Instruments, closely followed by a collection of Japanese manufacturers, but now Motorola have made a surprise entry into the fray by introducing not just another second source for the 2716, but a brand new device with four times the capacity, the MCM68764. This monster 8K x 8 EPROM has leapfrogged the biggest Intel device, the 4K x 8 2732, and will surely carve out a useful niche for its canny manufacturer—you could get a fair sized BASIC interpreter into a single MCM68764!

Making a device with no less than 65,536 separate bits to go wrong is quite a challenge but Motorola obviously hope to achieve useful yields. Those devices that do get rejected may not be wasted however, because Motorola have also announced the MCM68732 which is, surprise surprise, half of an MCM 68764!

They don't saw an MCM 68764 down the middle, they just use a reject device where one complete half of the memory is still fully serviceable. The new '32 comes in two versions, -1 or -0, where the suffix indicates whether the most significant address bit (pin 21) should be tied permanently high or low. This technique is becoming more widespread as memories grow in size, and Intel themselves already sell "partial" versions of some of their memory parts, particularly 16K dynamic RAMs.

Half a loaf is better than none seems to be a fitting motto for tomorrow's memory makers!

SOLDERING IRON CONTROLLER

M.S. DHINGRA



IN the workshop probably the least cared for tool is the soldering iron, excessive temperature of which can ruin present day CMOS devices and f.e.t.s. This article describes a highly sensitive and responsive unit for controlling the tip temperature of soldering irons. The unit can also be used for other appliances like a colour developer bath etc., where a very accurate temperature within the range of $\pm 0.5^\circ\text{C}$ is required to be maintained, by a single adjustment.

In the ordinary solid state heat control systems, power to the element is usually delivered through a triac. At the preset temperature level the triac is switched off. Power to the element is switched on with the differential drop of the controller which is in the order of $\pm 2^\circ$ to $\pm 5^\circ\text{C}$. Such 'on-off' control produces large overshoots and undershoots and therefore the heat regulation is coarse.

For more accurate control, power is delivered to the element continuously. Only the duty cycle of the power is varied to achieve the maintaining level. The variation of duty cycle or the pulse width, controls the temperature. Such a method is termed as proportional heat control system as it draws the energy in proportion to the requirement. The temperature control characteristics of the above two systems are shown in Fig. 1.

SCHEMATIC

A block schematic of the proportional heat control unit is shown in Fig. 2. A1 is an integrator, which integrates any error signal developed due to the difference in sensor diode drop and set reference. The resulting integrated voltage change on biasing resistance R regulates the charging current to capacitor C of the square wave generator A2. The charging and discharging of C determines the output pulse width of A2. The regulated pulse width in turn controls the duty cycle of the firing of the triac correcting the sensor temperature and returning the sensor voltage to that of reference. As the sensor temperature is gradually corrected, the pulse width of A2 also changes due to decreasing error. At equilibrium, the pulse width stays at the maintaining level, when the error signal is zero. Maintenance of the constant temperature is possible only due to the holding feature of integrator A1. When the integrator input error signal is at zero level, the integrator output holds at existing level, this in turn holds the controller output pulse width. The resultant input and control curves are shown in Fig. 3.

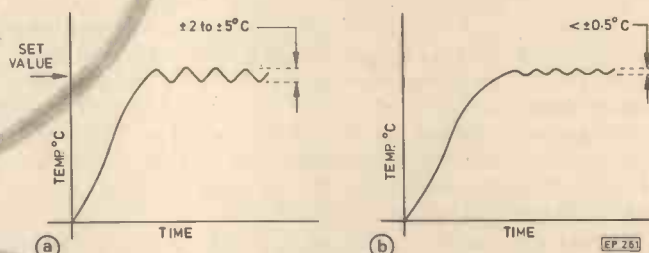


Fig. 1. Control characteristics of (a) on/off controller (b) proportional controller

CIRCUIT

The circuit diagram of the system is shown in Fig. 4. General purpose 741 operational amplifiers are used for the integrator as well as multivibrator.

The split supply to the system is ± 5.6 volt derived from a centre tapped transformer T1 and regulated by Zener diodes D3 and D4. Capacitors C1 to C4 serve for storage and to supply ripple free current.

Two 6 volt batteries can also be used. Silicon diode (D5's) negative temperature coefficient characteristics has been used for the sensor. For this purpose the author used a diode type 1N914, however silicon *n.p.n.* transistors like BC107 etc., with the base and collector tied, can also be used with the same accuracy.

Current to the sensor diode is limited within 1 to 5 mA by resistance R3.



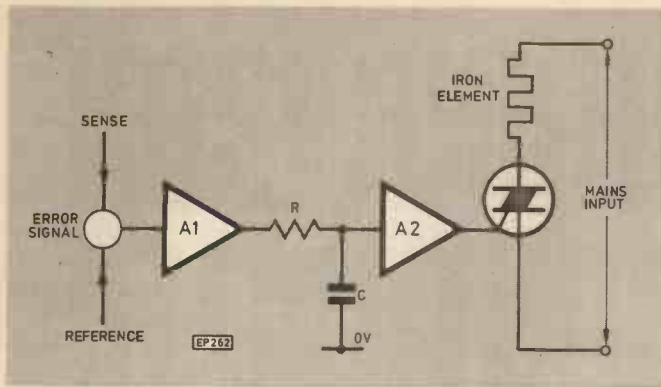


Fig. 2. Block diagram

The sensors output is fed to the inverting input at pin 2 of integrator IC1, through a resistor R8. The reference signal is generated by a 2.2 volt Zener diode D6 and this reference potential is divided by the chain VR1, VR2, R6, and R5. While VR1 is the calibration preset, VR2 is the control potentiometer.

The potential at the wiper of VR2 is fed to the non-inverting input at pin 3 of integrator IC1 through a resistor R7. C5 is the integrating capacitor. Its value in combination with R8 defines the rate of charge.

Integrator output at pin 6 is fed to C6 of the multivibrator formed around IC2, through resistor R9. The potential here regulates the charging current of C6 and therefore the output pulse width at pin 6 of the multivibrator IC2.

The frequency of the multivibrator is defined by the value of R12, which is about 70Hz with the values used. The ratio of R11 to R10 defines the maximum mark-to-space ratio, that is the maximum conduction period for the triac during one cycle of the multivibrator.

The pulse width controlled output at pin 6 of IC2 swings between the positive and negative supply lines and rectified by diode D7 to give positive pulses for triac firing. These are fed to the gate of the triac through current limiting resistor R13. The triac remains on during the entire positive gate pulse period providing a supply to the soldering iron.

ASSEMBLY

The p.c.b. of the control unit is shown in Fig. 5 and the component placement in Fig. 6. Fix the components in place with due care for polarities. The assembled p.c.b., supply transformer, triac and the three pin socket outlet should be mounted in a suitable metal box.

Mount the triac on a heatsink, well insulated from the

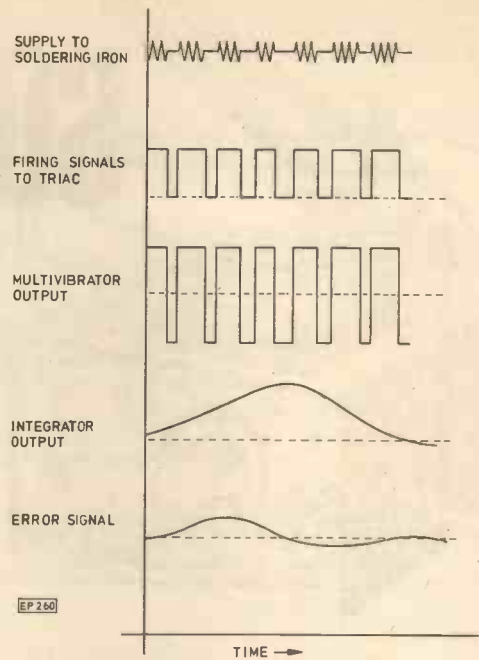


Fig. 3. Control graphs

metal cover. Fix the control pot VR1 on the front of the cover, with a knob graduated evenly to read 200° to 300°C for its full variation.

Clamp the sensor diode onto the metal body of the soldering iron, near the insulated handle grip, well away from the tip. This is because silicon diodes operation is limited to 150°C. Further, for a 100°C variation of tip temperature from 300° to 200°C, a variation of 10°C from 145° to 135°C is present at the location indicated, making it the best place for monitoring the tip temperature. Connect the sensor leads with the control unit by means of a screened shielded cable, caring for the proper polarity of the diode.

CALIBRATION

Before energising the unit, check once again that all connections are made properly. Now switch on the supply to the unit, without a soldering iron at the socket outlet and measure across Zener diodes D3, D4 and D6 that the voltages are +5.6, -5.6 and +2.2 volts with respect to common point. Also check the functioning of free-running multivibrator by a multimeter set to the range 10 volts d.c.

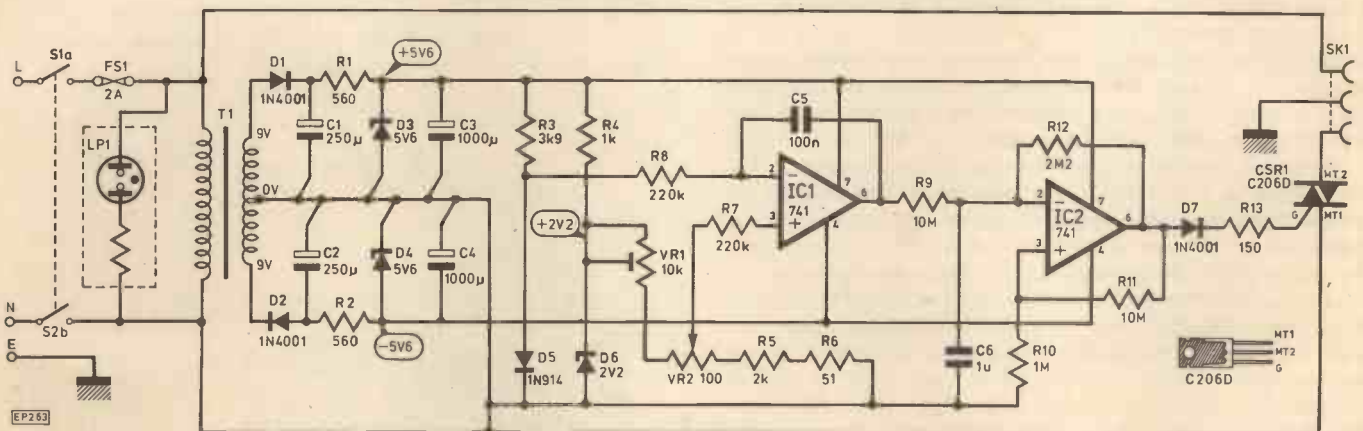
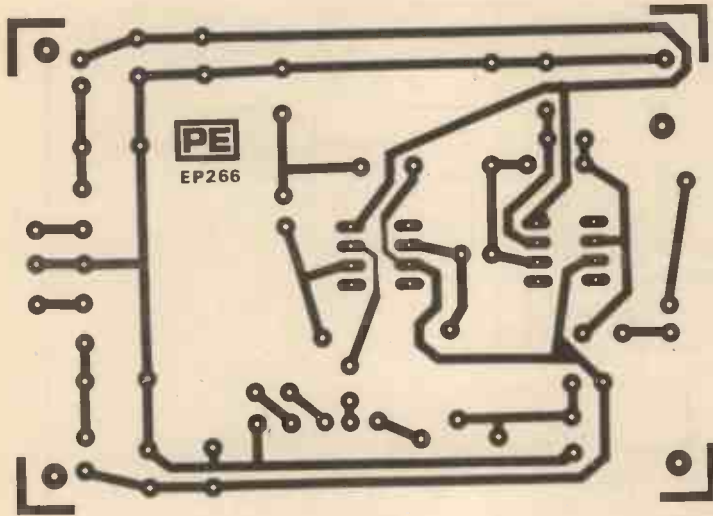


Fig. 4. Circuit diagram



COMPONENTS

Resistors

R1, R2	560 1W
R3	3k9
R4	1k
R5	2k 2%
R6	51 2%
R7, R8	220k
R9, R11	10M
R10	1M
R12	2M2
R13	150

Capacitors

C1, C2	250µ elect
C3, C4	1,000µ elect
C5	100n polyester
C6	1µ polyester

Semiconductors

D1, D2, D7	1N4001
D3, D4	BZX85-5-6 1.3W Zener
D6	BZY88-5-6 400mW Zener
IC1-IC2	741
D5	1N914
CSR1	C206D 400V/3A triac

Potentiometers

VR1	10k
VR2	100 linear

Transformer

T1	230V pri.—9-0-9V, 400mA sec
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Miscellaneous

S1	Double pole mains on/off, FS1-2A fuse, LP1-mains neon
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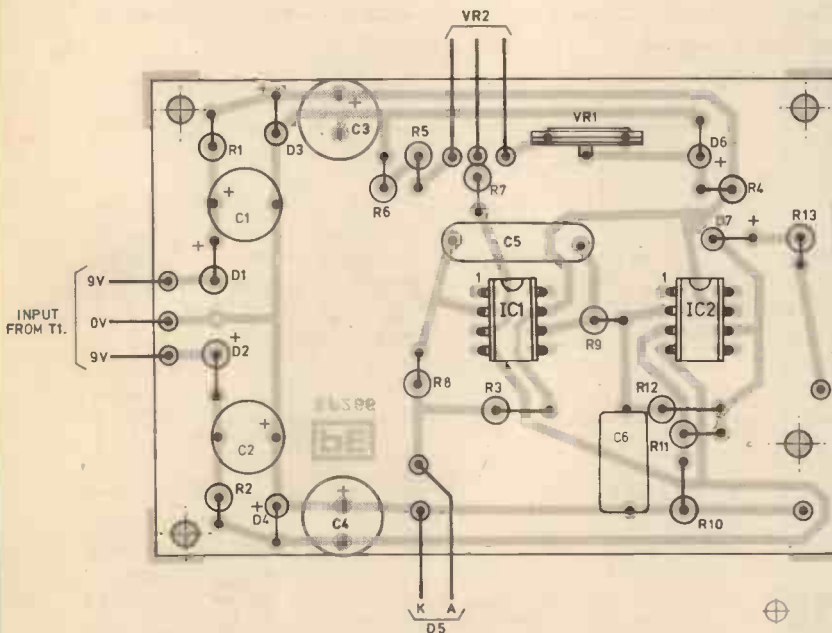


Fig. 5. P.c.b. layout of controller

Fig. 6. Component layout and external wiring

between pin 6 of IC2 and neutral, oscillations should be observed.

Now, energise the soldering iron from a normal supply source, not through the unit, and let it heat for about thirty minutes, to get the iron's temperature and in turn the sensor's output stabilised. Keep pot VR1 set to maximum resistance and turn the control potentiometer anticlockwise, i.e. at 200°C mark. Connect a multimeter set to the 10 volt d.c. range between pin 6 of IC1 and common point.

The voltage indicated will be zero. Now slowly reduce the resistance of VR1 until the meter needle remains at about 5V. Slow needle movement is due to integrator action.

Now turn VR2 to its other extreme (300°C), and the meter needle will swing back to zero value.

With careful setting of VR1, arrange that the transition of the meter deflection from 0 to 5 volt commences at exactly mid position of the control pot VR2, i.e. at 250°C mark. You

will be able to achieve this state in two to three steps. Now plug in the soldering iron at the socket outlet of the control unit and temperature control is yours.



NEXT MONTH



CAR RADIO...

A five push button long and medium waveband car radio for under £11. This design employs an excellent tuner head with a ceramic i.f. module and 6W i.c. output stage to give exceptional performance for price. Our article will give full constructional details.

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SPEAKER SYSTEMS** for Discos and Rock Bands.



PRACTICAL ELECTRONICS

OUR MARCH ISSUE WILL BE ON SALE FRIDAY 8 FEBRUARY 1980

THE ULA

UNCOMMITTED LOGIC ARRAY

E. FRY B.Sc.

CUSTOM designed integration may be every designers dream, but it is the financial planners nightmare, for with every i.c. needing its own individual mask the cost is enormous and must be borne by the buyer. Therefore, custom designed integration is usually confined to applications involving large volume production, and is impractical for smaller low cost projects.

Universal logic elements endeavour to maintain the advantage of custom circuits while reducing some of their disadvantage. They are produced as standard devices designed to fit user application with little modification required, this consisting of custom design for the metallisation mask used on a standard logic array. Such universal logic elements appear in the ROM, PLA (programmable logic array) and ULA (uncommitted logic array) and it is this which will be looked at.

THE ULA DESCRIBED

As the name implies, the ULA consists of a single chip containing an array of uncommitted gates or cells in which the silicon wafers are processed up to the stage just before the final metallisation and then stored. Each cell is separate and all that is required is a mask for an interconnection pattern, this being designed for the users specific circuit requirements. As the individual component in each cell can be readily interconnected, a combination of linear and digital circuits can be produced.

The success of the ULA is due to Ferranti who developed the CDI (collector diffusion isolation) process. A feature of CDI which gives it its facility is that the bulk semiconductor material used within the i.c. can be used for supply and ground return currents. This process greatly assists the manufacture of ULAs as all power rails are removed from the top interconnection plane, enabling the interconnection pattern to be devoted to purely programming the cells.

SCHEMATIC

A schematic diagram of the CDI process is shown in Fig. 1.

The low resistivity of the collector isolation diffusion allows supply connections to be made without metal, and earth connections can be made direct as the epitaxial p layer is on a p type substrate.

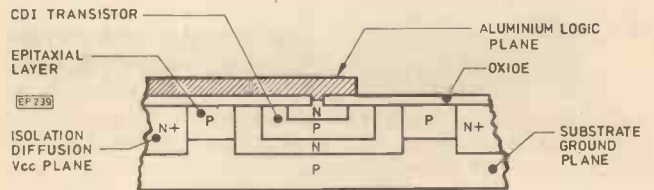


Fig. 1. The CDI process

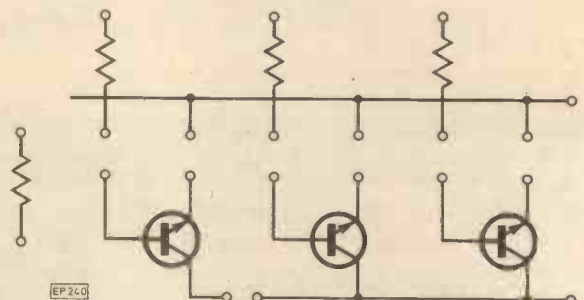


Fig. 2. Basic uncommitted cell

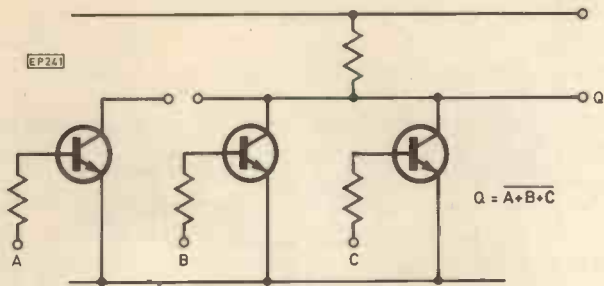


Fig. 3. Gate cell

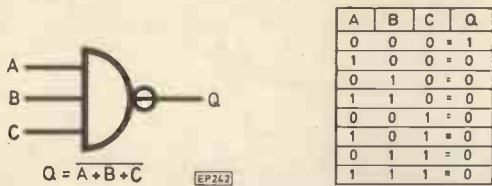


Fig. 4. Logic symbol and truth table

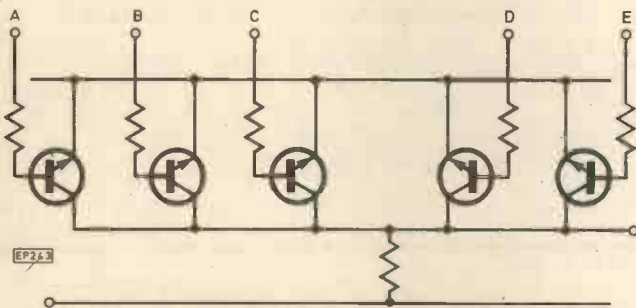


Fig. 5. Collector ORing to increase fan-in

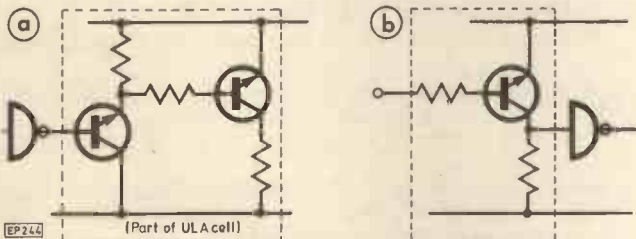


Fig. 6(a). Input interfacing (b) Output interfacing

The basic uncommitted cell (Fig. 2) consists of three transistors and four resistors, thus providing the usual RTL logic. Though each individual component can be connected as required, the most common cell connection to be found in the array is shown in Fig. 3. This is called the basic gate cell. It is a positive NOR gate, giving great versatility as a basic building block. Fig. 4 shows the truth for this gate. How these gates can be used as complete circuit functions will be shown later.

INCREASING FAN IN

An example of inter-cell connecting can be seen when designing for a higher fan-in, as in Fig. 5, where components from two cells are used to increase the fan-in to five. This technique termed 'collector ORing' can be used to increase the fan-in up to a maximum of ten. The remaining transistor and two resistors can be used in some later stage.

Owing to the cell being uncommitted, interfacing can be achieved directly for most types of logic and circuitry. In Fig. 6 we see how interfacing is done by using components from one cell. As the maximum input current for the interface gate is in the order of 40µA, it allows several gates to be driven

from any of the typical logic families. For example TTL will drive up to ten ULA interface gates, CMOS, ten also. DTL up to six, etc. Output interfacing is capable of driving much lower numbers though, such as TTL—three gates. This can be increased by paralleling the interface gates, thus three ULA gates will drive nine TTL gates.

ULA is also capable of driving directly i.e.d.s, and discrete transistors, etc, and in such a mode of operation a total of 10mA drive current is available.

APPLICATIONS

The basic uncommitted cell can be connected to provide a large variety of linear and digital circuits. A few of the many ways in which the basic ULA can be used as building blocks are shown in Figs. 7 to 10.

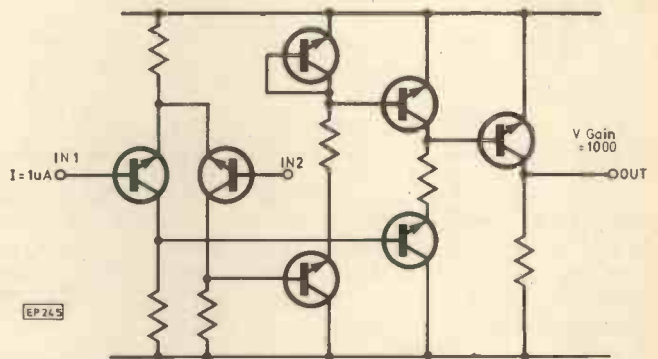


Fig. 7. Operational amplifier, using 13 components from three cells

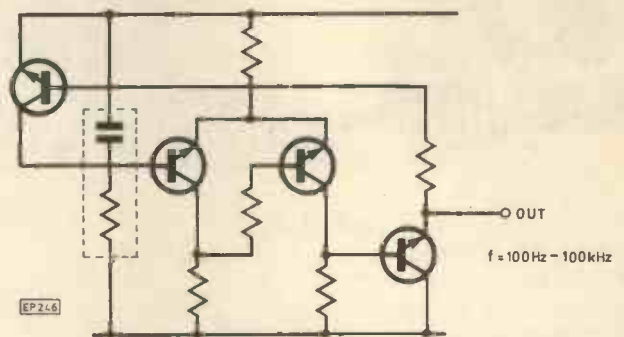


Fig. 8. Basic oscillator, using ten components from two cells and external timing R/C network

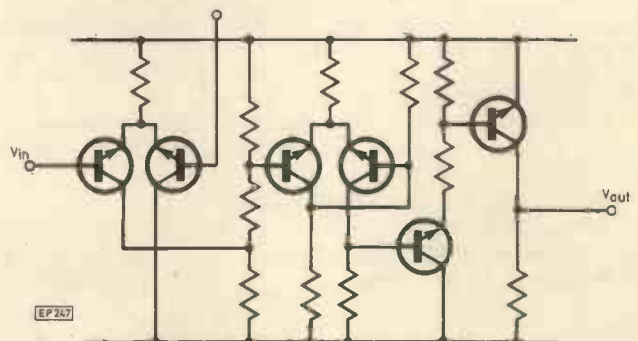


Fig. 9. Analogue voltage comparator using 17 components from four cells

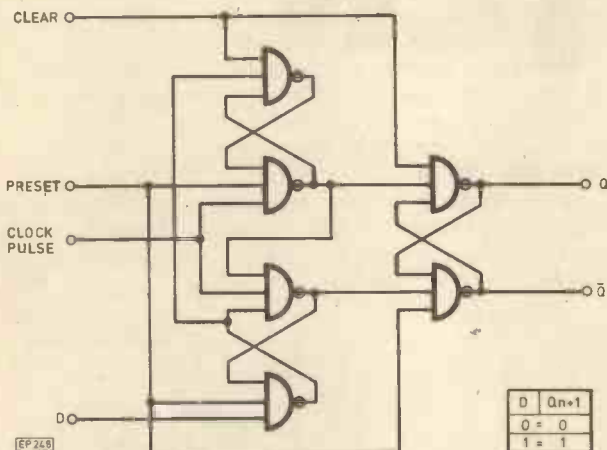


Fig. 10. D type flip-flop using all components from 6 cells

An example of how an entire array may be committed to perform a set function is shown in Fig. 11. An automatic control for an item of photographic equipment, containing all the necessary logic in one 28 pin i.c. package. It utilises approx. 140 of the cells, and replaces an equivalent 20 packages of MSI.

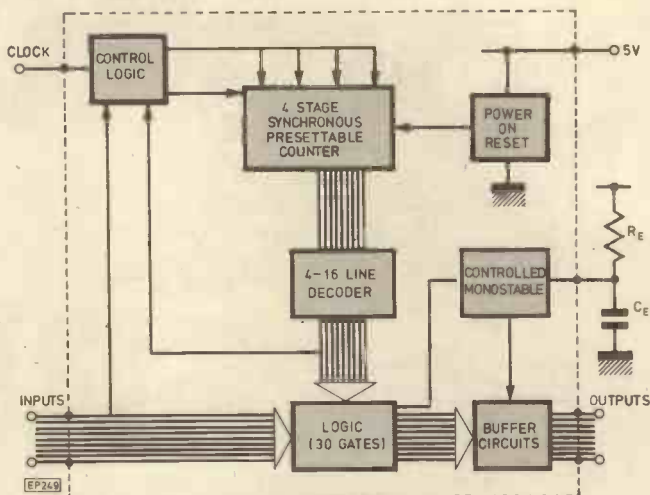


Fig. 11. A photographic control system using ULA

To summarise then—ULA, because of low cost and quick availability is ideal for smaller projects, and as an evaluation vehicle. It saves space and assembly costs. Both reliability and performance are greatly improved; basically everything that custom designed integration offers, but cheaper and faster.



BOOK REVIEWS

TELECOMMUNICATIONS SYSTEMS FOR TECHNICIANS 1

by Walker & Danielson

Published by Newnes-Butterworths
103 pages. Price £2.95

PRIMARILY intended for TEC (Technician Education Council) level students taking courses in electronics, telecommunications and marine radio, this volume should prove popular to a wide readership as it covers a wide range of subjects at elementary level; information transmission, radio, radar, radio navigation, telephony, telegraphy, routing and data communication.

The text is well illustrated by over 180 diagrams and where applicable, BS symbols are maintained throughout.

To check recall simple revision exercises terminate each chapter.

SEMICONDUCTOR TECHNOLOGY

Edited by G. W. A. Dummer

Published by Pergamon Press
203 pages. Price £20

THIS is a collection of papers read at the Seminex technical seminars held at Imperial College, London, during the period 10-14 April, 1978. Topics cover a variety of developments in semiconductor technology.

TELEVISION PRINCIPLES AND PRACTICE

by Zarach & Morris

Published by Macmillan Press

294 pages. Price £12.50 hardback, £5.95 paperback

ANOTHER volume that caters for the various television options within the TEC framework as well as City and Guilds courses.

Throughout, circuits relate to modern TV receivers which include transistor, i.c.s., thyristors, etc.

The first three chapters cover the science of colours and the formation of monochrome and colour signals. Then follows a review in block form of the receiver which forms the basis of circuit analyses for future chapters.

Throughout the book emphasis is placed on the practical aspects of TV servicing and as such should prove invaluable to anyone professionally involved whether experienced or just beginning a career. This applies equally to the enthusiastic amateur.

AUDIO SYSTEM DESIGN FOR SCHOOLS AND COLLEGES

by R. H. Welch, B.Sc.

Published by NCST Trent Polytechnic
195 pages. Price £2.75

IN contrast to many books on this subject which set out the relative merits of design bricks in the audio chain tailored to a specification with the inevitable juxtaposing of commercially available systems, this book introduces the basic ideas necessary for the enquiring reader to design and build his own.

Aimed primarily at sixth formers and college students it forms an excellent practical guide and reference for the construction of turntable, pick-up arm, amplifier and loudspeaker cabinets etc., in project work allied to formal Engineering courses.

Albeit motivated academically it is pitched at a level which should prove enlightening to anyone brave enough to go it alone on a d.i.y. system. Take heart, lasers and digital encoding aren't mentioned.

COMPUTER CASE OFFER!

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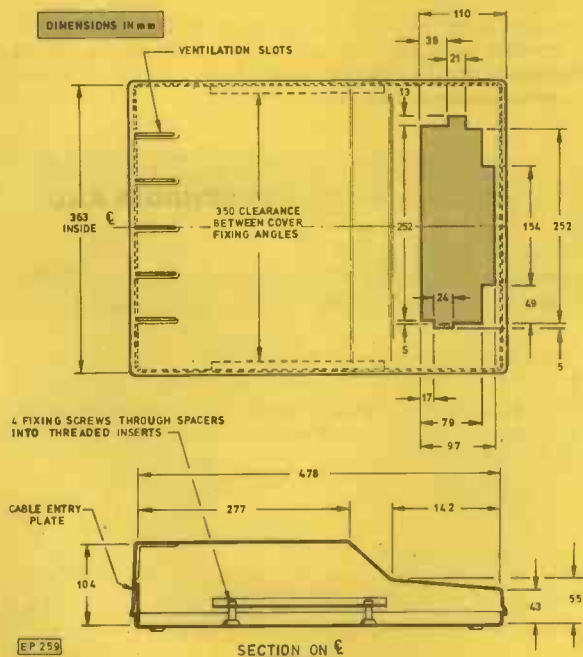
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This beautiful orange and black finish plastic case is available for Superboard II, Compukit UK101 or, with an uncut keyboard panel, for mounting many other hobby computers. It is supplied with a mounting wedge to give a suitable keyboard angle and fixing screws for Superboard or Compukit. The case is strong enough to support a small portable TV or video monitor and has ventilation slots and a cable access panel at the back. It **does not** carry the "PE Compukit" badge shown in the photograph.

The dimensions of the case (with Superboard keyboard cut out) are shown below—case material is approximately 2mm thick with 4mm radius corners. We recommend that the power regulator fitted to Compukit boards is mounted on a heatsink and fixed to the outside back of the case.

The front cover illustration shows part of our own office system employing this case. PE has been able to arrange this special price so don't miss out as the offer closes Friday 29th February 1980.



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PATENTS REVIEW...

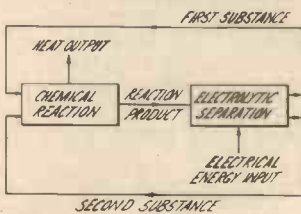
BATTERY ALTERNATIVE

The Secretary of State for Industry has patented (British patent No 1 552 436 which dates from 1976 and is thus issued under the old laws), an interesting idea for storing large quantities of electrical energy by chemical means other than a vast and expensive battery of electrolytic cells. The inventor, Albert Montgomery, suggests that the system could be used either to even out the peak and low load demands on an electric power station or to drive a vehicle.

The invention relies on a reversible exothermic chemical reaction, that is to say a reaction which produces heat from the mixture of two materials which are subsequently recovered and regenerated by electrolysis. This process is, under some conditions, more efficient at the bulk storage of energy than a conventional battery.

Fig. 1 shows the flow diagram for the cycle of the combination of the two materials

FIG. 1.



to produce a heat output from exothermic chemical reaction and a resultant electrolytic separation of the reaction product resulting from this. Recombination to produce the original materials for recycling is effected by the electrical energy input.

Fig. 2 shows the basic "circuit". Two stainless steel vessels, 1, 2 contain sodium and sulphur respectively. These raw materials are fed through valves to reaction vessel 5 which is of heat and corrosion-resistant material. Copper conduit 6 for a fluid to be heated (for instance water or gas) is coiled round the vessel 5.

The sodium and sulphur react together to form sodium polysulphide with the release of considerable quantities of heat energy. This energy is taken up by the fluid in the coil 5 and led off to a load.

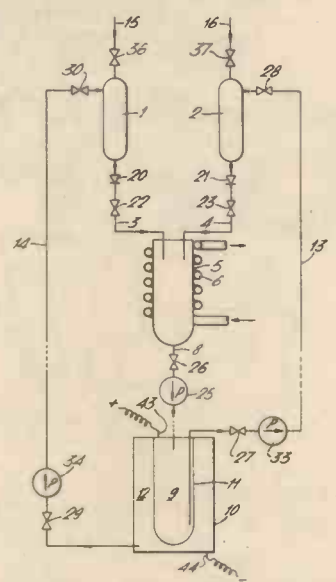
The load is a heating stage for an electric turbine or vapour engine for a vehicle, such as a Stirling cycle heat engine.

The spent polysulphide is fed to recovery cell 10 which has a blind-end tube of solid beta-alumina electrolyte. This tube defines a cathode, 9 inside an anode 12. A 3 volt d.c. supply is fed through the electrodes to the polysulphide, and sodium ions are conducted through solid electrolyte 11 to the anode while the sulphur remains in the cathode 9.

Regenerated sodium and sulphur are then recycled to the containers 1 and 2. Thus the patented energy storage system has a capacity limited only by the size of the storage containers.

Supplies of polysulphide can be converted to sulphur and sodium throughout the entire period when there is excess power available, for instance during low-load night conditions on the national grid, and supplies of sodium and sulphur recombined to provide heat and generate power during the

FIG. 2



entire period when there is a heavy load on the grid. In this way large quantities of electrical energy can be stored at far more economical price than with conventional batteries and without recourse to exotic alternative energy storage techniques, such as pumping water up a hill, lifting weights up a slope or compressing gas into vast reinforced cylinders or underground caves.

The fact that the British government has patented the invention suggests that its practical use is under serious consideration. The idea could perhaps also be applicable to domestic central heating. The chemicals would be separated on night storage rates and combined during the day to produce hot water for central heating.

PE/LEKTROKIT PRIZES

The following prizes were awarded as a result of the PE/Lektrokit competition run in our September 1979 issue.

1st Prize a Lektrokit Powerace 102, a jumper wire kit and 16 pin Test Clip goes to *Mr. D. J. Speakman* of Braintree for his dual timebase submission.

Runners up prizes have been awarded to *Mr. T. Johansson* of Sweden for a l.e.d. logarithmic level meter; *Mr. I. M. Crann* of Brecon for a novel tone generator for electronic organs and to *Mr. T. Davies* of Swansea for a combined voltmeter/logic scope.

Due to the very limited number of entries, first, third, fourth and seventh prizes only have been awarded.

POINTS ARISING

ULTRASONIC CLEANER (January 1980)

The value of C2 should be 150pF and not 15μ.

ULTRASONIC BURGLAR ALARM (December 1979)

The 18V battery voltage should be wired through the keyswitch instead of the neutral lead.

DIGITAL TEMPERATURE CONTROLLER (October 1979)

The wiper and top end of VR1 should go to +9V and not to IC1/14 (p.c.b. is correct).

ACOUSTICALLY COUPLED TELEPHONE MODEM

Kenneth Amor

THIS article outlines the methods used to send and receive digital information, and shows in detail an inexpensive method of accomplishing this using conventional telephone facilities. The Acoustic Modem allows data and software exchange almost instantly, over any distance.

Data to be transmitted over varying distances has been standardised in its format, known as the American Standard Code for Information Interchange, and is abbreviated to ASCII. The ASCII Code format is shown in Fig. 1.

ASCII CODE

Every Alphanumeric character to be transmitted differs in bit pattern, and even the number of bits varies. It is possible to identify the whole of the ASCII character set using only seven bits, plus the start and stop bits. Fig. 2 shows this set of characters with their respective logic levels.

A code can be obtained for every Alphanumeric character, and many other characters to be found on certain keyboards.

THE MODEM

The acoustic modem is able to deal directly with the waveform shown in Fig. 1, because it is in serial form. A typical period for one bit is ten milliseconds, thus allowing one character to be transmitted in approximately 100 milliseconds. The bit period may be varied depending upon whether electromechanical or all electronic data transmission/receiving equipment is being used.

Having applied data to the Modem we now wish to transmit same over many miles to a fellow computer user, and it is a question of sending and receiving this digital information over long distances without tampering with post office equipment. The answer is, of course, the acoustically coupled telephone modem. By inserting the handset of a domestic telephone into this device, data may be transmitted and received at will, albeit somewhat slowly.

The principle of operation employs frequency modulation techniques involving the phase locked loop principle. A full explanation of how the complete system works is given in conjunction with the circuit diagrams of Figs. 5, 6 and 7.

Constructional dimensions are given for both the standard telephone handset and the trimphone version. Of course, two such acoustic modems will be required to set up a data link, but it is assumed that each enthusiast would construct just one modem.

THE FREQUENCY SHIFT KEYED TRANSMITTER

The voltage controlled oscillator section of the phase locked loop (IC2) is utilised for the 1000Hz tone. The capacitor C1 and resistors VR2, VR3 are responsible for setting this initial frequency. This 1000Hz tone would be present for the logic level "0" on the input of the transmitter. This situation is so when TR1 is turned off. If now a logic level "1" is applied to the data input, IC1, using double negation, follows the input voltage level and therefore turns on TR1, effectively connecting R3, VR2 in Parallel with R4, and VR3. The new resistance value produced by this resistor combination produces a new VCO frequency of 1200Hz which is the frequency representing logic level "1". Pin 4 on IC2 (4046) contains the frequency shifting signal that is applied as base current to TR2 via R6. TR2 drives the miniature loudspeaker producing the two audible tones which represents the corresponding logic levels of input data. When the modem is in the "Send" mode the 1000Hz tone can clearly be heard from within the modem.

SETTING UP THE FSK TRANSMITTER

- Switch the Modem to SEND.
- Apply a short circuit to the data input.
- Connect a frequency meter to TP.
- Adjust VR3 until 1000Hz is shown on frequency meter.
- Now remove the short circuit from the input and apply +5V with respect to the common connection.
- Adjust VR2 until the new frequency becomes 1200Hz i.e. 200Hz change between "0" and "1".
- An audible indication will verify that all is well as these two frequencies are trimmed.

When this has been carried out it is worth applying a low frequency square wave to the data input of the transmitter to ensure that the frequency shift operation follows the logic levels being applied. A circuit suitable to carry out this test is shown in Fig. 4.

THE RECEIVER SECTION

Let us assume we are receiving the transmitted signal from a distant modem. The modem at our end picks up from the telephone handset the faint signals and applies them to an audio amplifier (IC3). A gain of approximately 100 was found to be sufficient to provide a working signal for the phase locked loop (IC4). C2 couples the FSK (Frequency-Shift Keying) signal to pin 2 of IC4. Diodes D5, D6, across R16 serve to provide some degree of limiting, as well as protection to IC4's input. Limiting by D5, D6 prevents amplitude modulation interference. The timing components VR1, C3, determine the frequency that the phase locked loop will free run at. This frequency should be adjusted to be 1100Hz, i.e. between the two FSK frequencies to be received. When a logic "0" is received (1000Hz) at pin 2 of IC4, the phase locked loop will lock onto this causing the voltage controlled oscillator of the PLL to suddenly shift from 1100Hz to 1000Hz. As this happens, the output signal from the phase comparator (pin 7 of IC4) becomes negative with respect to the reference potential at pin 6 of IC4. This potential difference will force the d.c. comparator IC5 to swing negative and this output signal will be caught at approximately -0.5V by D7 when the logic "0" is received. If now we receive a logic "1" signal (1200Hz) at pin 2 of IC4, the PLL will lock with its VCO frequency at 1200Hz, resulting in the phase comparator output becoming greater than the d.c. reference on pin 6 of IC4. Thus the voltage comparator IC5

Fig. 3. Block schematic of the Modem link-up. The system is half-duplex. Serial pulse repetition rates should be no higher than the VCO Tx frequency. Input levels are TTL. Applications might include micro' link-ups at 110 Baud, VDU communication using a 20mA loop, facsimile transfer etc. Trials with the Modem included a link-up with a teletype from Bebington (Merseyside) to Luton

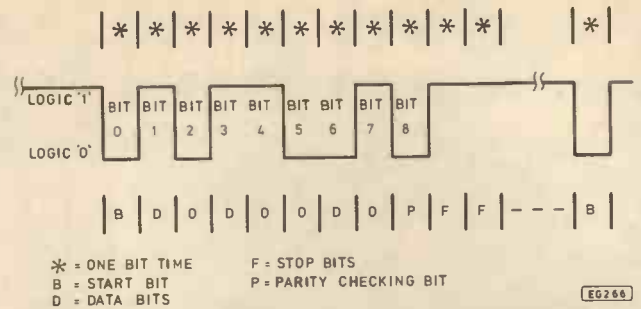
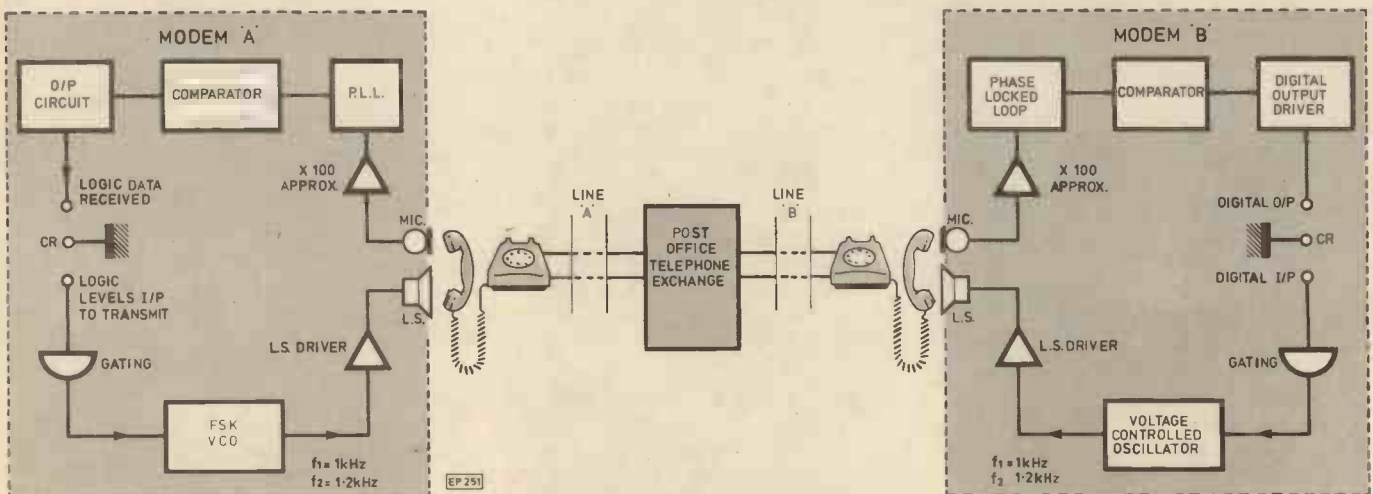


Fig. 1. ASCII code format, in this case representing the character "M"

b7→b5	000	001	010	011	100	101	110	111
b4→b1								
0000	NUL	DLE	SP	0	@	P		p
0001	SOH	DC1	!	1	A	Q	a	q
0010	STX	DC2	"	2	B	R	b	r
0011	ETX	DC3	#	3	C	S	c	s
0100	EOT	DC4	\$	4	D	T	d	t
0101	ENQ	NAK	%	5	E	U	e	u
0110	ACK	SYN	&	6	F	V	f	v
0111	BEL	ETB	'	7	G	W	g	w
1000	BS	CAN	(8	H	X	h	x
1001	HT	EM)	9	I	Y	i	y
1010	LF	SUB	:	J	Z	j	j	z
1011	VT	ESC	+	K	[[{	{
1100	FF	FS	,	<	L	\		
1101	CR	GS	-	=	M]	m	}
1110	SO	RS	period	>	N	^	n	~
1111	SI	US	/	?	O	_	o	DElete

Fig. 2. The ASCII code, showing the binary equivalents of each character

responds with an output voltage positive going. It will in fact saturate in the positive direction at about 4.4 volts, providing an adequate logic level to become our Data output. C5 is chosen for determining the lock range of PLL, while R19, C6, R20, C7, R21, C9 form a filter network preventing the possibility of the VCO frequency, or harmonics, upsetting the output comparator.

Fig. 4. Test oscillator for FSK transmitter

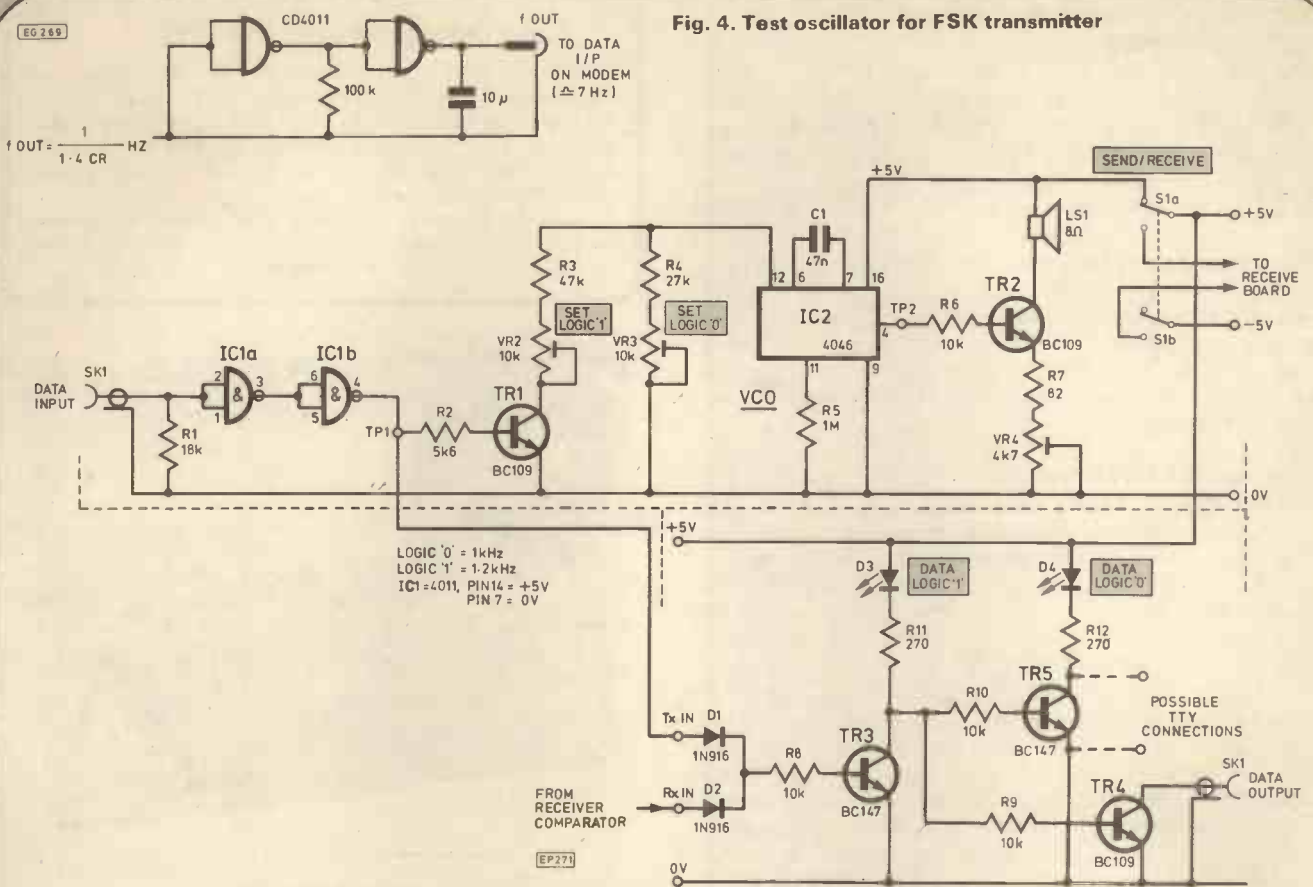


Fig. 5. FSK transmitter of Acoustic Modem

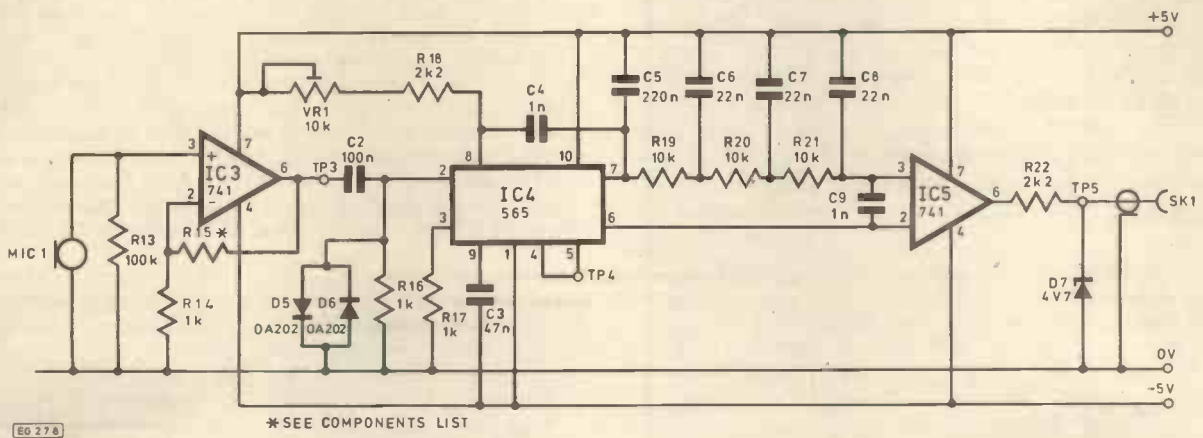


Fig. 6. FSK receiver section

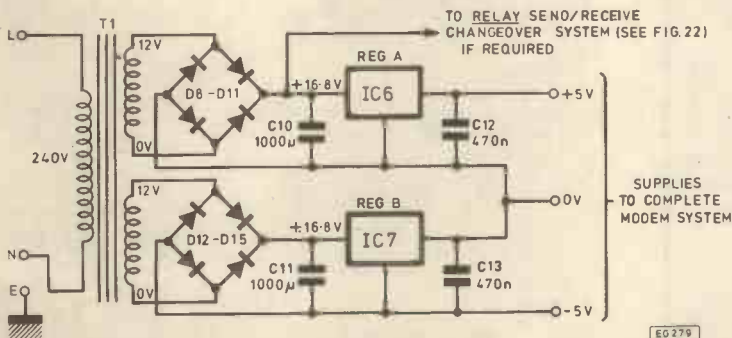


Fig. 7. PSU. ±5V at 100mA

COMPONENTS . . .

Resistors

R1	18k
R2	5k6
R3	47k
R4	27k
R5	1M
R6, R8-R10, R19-R21	10k (7 off)
R7	82
R11, R12	270 (2 off)
R13	100k
R14, R16, R17	1k (3 off)
R15	100k nom. (subject to trimming)
R18, R22	2k2 (2 off)

All resistors 5% $\frac{1}{4}$ W unless otherwise stated

Capacitors

C1, C3	47n WIMA FKC-3 temp. stable or equiv. (2 off)
C2	100n
C4, C9	1000p (2 off)
C5	220n
C6-C8	22n (3 off)
C10, C11	1000 μ /25V (2 off)
C12, C13	470n (2 off)

Ratings and types not critical unless specified

Potentiometers

VR1, VR2, VR3	10k preset (3 off)
VR4	4k7 preset

Transistors and Diodes

TR1, TR2, TR4	BC109 (3 off)
TR3, TR5	BC147 (2 off)
D1, D2	1N916 (2 off)
D3	i.e.d. (red)
D4	i.e.d. (green)
D5, D6	OA202 (2 off)
D7	4V7 Zener (BZY 88)
D8-D11, D12-D15	1A 50V rectifier bridge (2 off)

Integrated Circuits

IC1	CD4011
IC2	4046
IC3, IC5	741
IC4	565
IC6, IC7	5V 1A regulator (plastic)

Miscellaneous

Crystal microphone insert
 Miniature loudspeaker 8 Ω 100mW
 Transformer 0-12V: 0-12V: 3VA
 I/O d.i.n. socket (3-pin)
 Double-pole, double-throw switch (send/receive)
 Veroboard 137 x 64mm (5 x 2.5ins.)
 I.c. board type: PCB421 available from **West Hyde Developments Ltd.**

CONSTRUCTOR'S NOTE

A complete set of components for the Acoustically Coupled Telephone Modem is available from Watford Electronics (see advertisers' index).

MODEM OWNERS' REGISTER

Having got you to build the modem, your computer will be on the look-out for other machines to talk to, and so it is our endeavour to compile a list of all PE Modem owners who wish to "ASCII-municate". We hope response will be sufficient to provide a useful list on request, so send us your details (listed below) and a S.A.E.

Suggested details: *Name, address, telephone number, system, micro' used, peripherals and reason for incorporating the modem (in brief please).*



PO REGULATIONS

We believe the specification of the PE Acoustic Modem conforms to the Post Office regulations governing the use of devices acoustically coupled via the Public Switched Telephone Network. However, it should be noted that under the Post Office Act 1969 acoustically coupled devices should not be used with Post Office maintained plant without the prior consent of the Post Office.

*Applications for the evaluation of private equipment (for which a charge in excess of £50 will be payable) may be made to the Post Office once the supplier is satisfied that the equipment meets the requirements given in Technical Guide No. 32, available from: **Post Office Telecommunications Headquarters, Service Department Sv 1.1.3.3, Tenter House, 45 Moorfields, London EC2Y 9TH.***



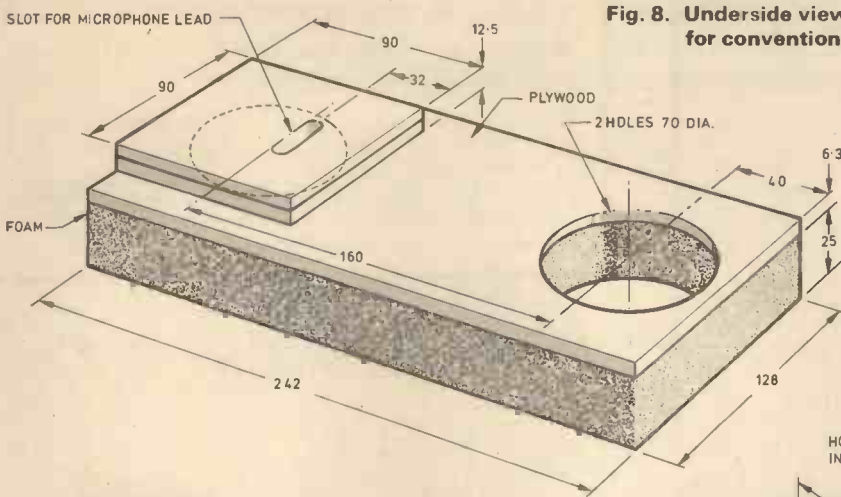


Fig. 8. Underside view of handset mount for conventional telephone

DIMENSIONS IN mm.

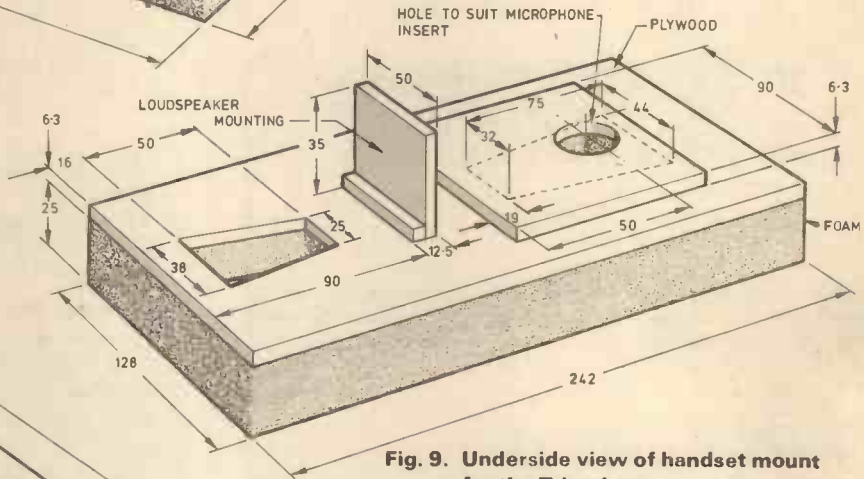
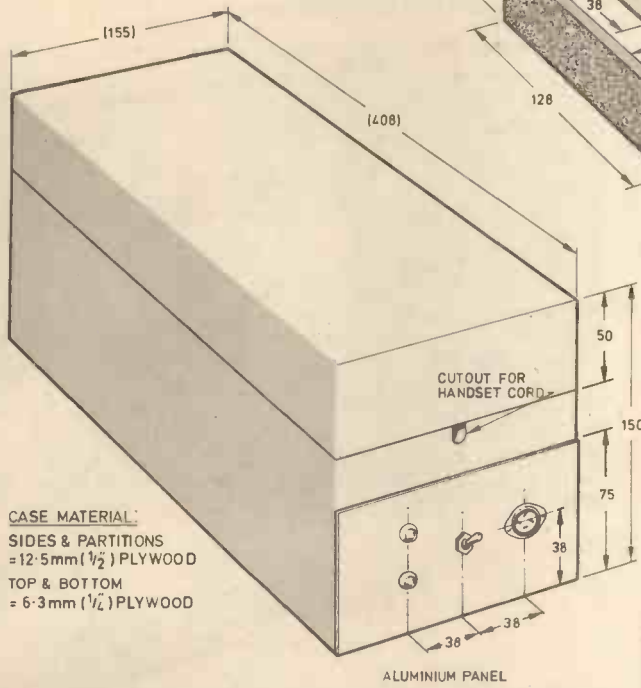


Fig. 9. Underside view of handset mount for the Trimphone



CASE MATERIAL:
 SIDES & PARTITIONS
 = 12.5mm (1/2") PLYWOOD
 TOP & BOTTOM
 = 6.3mm (1/4") PLYWOOD

ALUMINIUM PANEL

Fig. 10. Overall dimensions of the acoustic cabinet

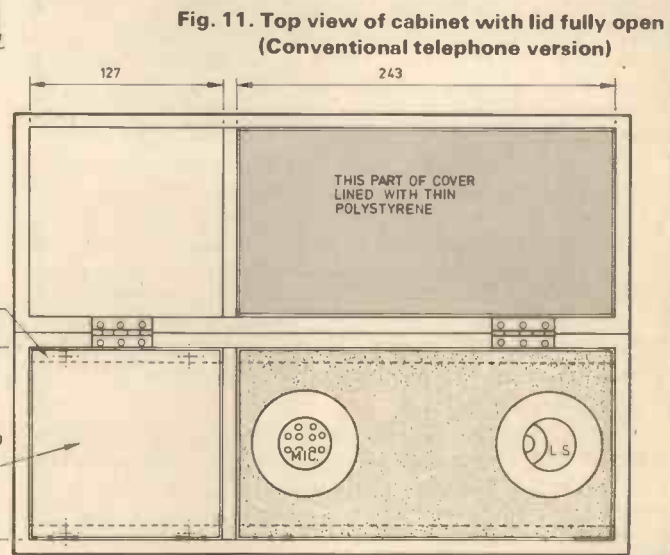


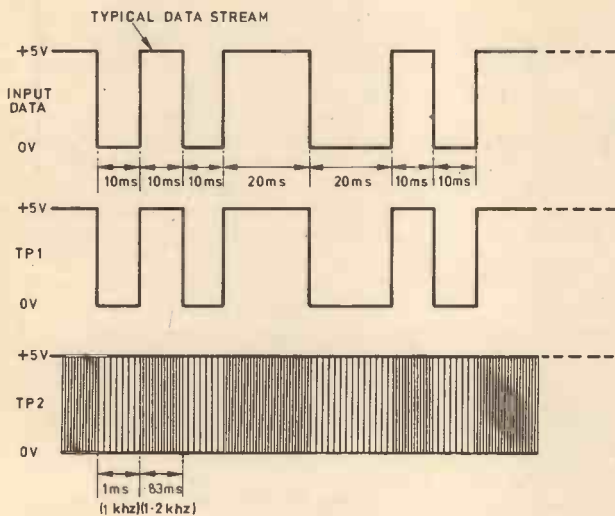
Fig. 11. Top view of cabinet with lid fully open (Conventional telephone version)

10mm THICK BEADING FITTED TO CASE SIDES 6.5mm BELOW SURFACE ON POWER SUPPLY SECTION & 25mm BELOW SURFACE ON HANDSET SECTION

6.3 PLYWOOD COVER

EP250

ACOUSTIC MODEM



EQ 208

Fig. 12. Transmitter Test Point waveforms. When no input is applied and the Modem is switched to Transmit, the 1Ms square wave will appear continuously at Test Point 2

RECEIVER CHECK

- Look at signal on TP3 to ensure it is at least 250mV peak to peak when receiving from distant modem.
- If not, the gain of IC3 may have to be raised by increasing R15 (100k).
- With no signal being received, adjust VR1 until the frequency meter placed on TP4 reads 1100Hz.
- With Data being received observe TP5 and refer to receiver waveforms information.

DATA INDICATOR AND OUTPUT OPTIONS

Signals from both the transmitter and receiver are fed to i.e.d.s D3 and D4. This allows us to monitor the data both when we are sending and also receiving. It should be made clear that the modem will not be doing both at once, being half-duplex.

TR3 switches on when the output of gate (IC1) goes high (Transmit Mode) or when Pin 6 of IC5 goes positive (Receive Mode). This will result in (D3) turning on and indicating a logic "1" level being present. As TR3 is turned on, R12 is virtually connected to ground, thus turning off TR5. The transistors TR3 and TR5 form a "see-saw" stage. When one is off the other is on, so that when no base current flows in TR3 it is a sure sign that a logic "0" is being received or transmitted, thus TR5 will be turned on and the "0" i.e.d. will be illuminated.

TP5 would normally be the receiver output point. It might be that a 180° signal is required to drive the device coupled to the modem. If this is so, connect the output of the modem to the open collector of TR4. i.e. TR4 will be turned on when "0" is being received. This is suitable for coupling to a teletypewriter.

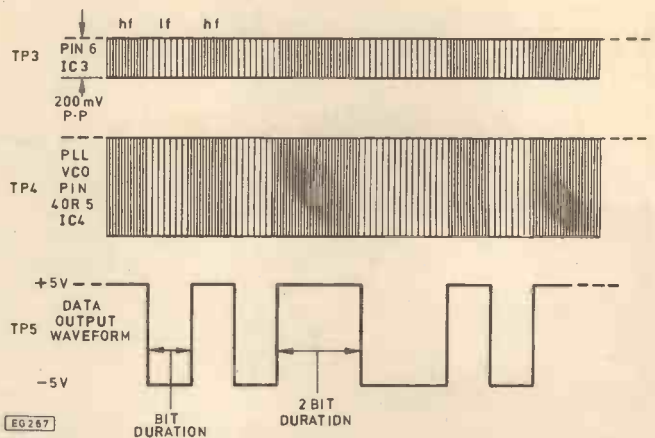
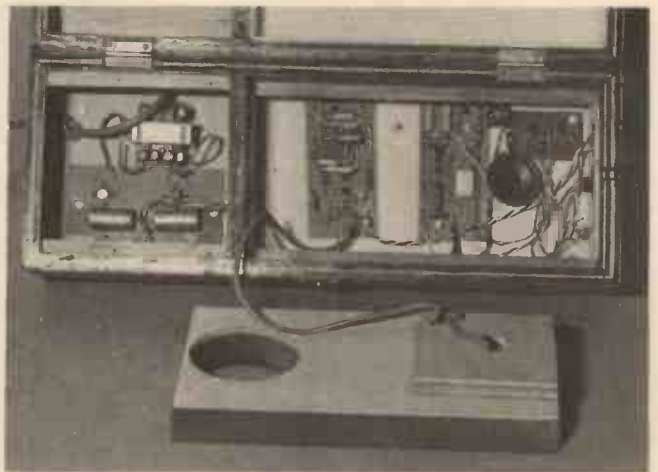
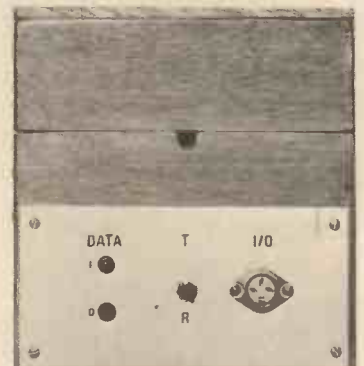


Fig. 13. Receiver waveforms. Example shown is the ASCII character "M". The l.f. pulses, i.e. logic 0, are a burst of 1ms pulses for the duration of the bit. The h.f. burst (logic 1) contains bit blocks 0.83ms wide. For Teletype operation the output bit duration is about 10ms



NEXT
MONTH



In the concluding article next month we shall give stripboard layout diagrams for the three boards, plus a considerable amount of applications information and device data showing how the modem can be improved and/or adapted to suit particular link-up requirements

Strictly

Instrumental

by K. Lenton-Smith

A NEW instrument was presented at the recent Trade Fair—the Sharma HX80. The manufacturers, Keith Hitchcock, have been mentioned previously in this column in connection with their Doppler-effect speaker systems. Early last year I heard the prototype in action and had an opportunity to examine its very compact circuitry. Played through a rotary speaker, it was a Hammond tone-wheel organ recreated.

At first sight, the HX80 is a fairly conventional portable organ, drawbar controlled with 49 notes on the upper and 37 notes on the lower manual. The 18-note pedal clavier has either 8' or 10' pitch available. Generation is by digital pulse trains and, by splitting the waveform into segments at differing voltage levels, versatile tone synthesis is achieved. The sine waves produced by this means are mixed (Fourier synthesis) in the normal way.

The group of nine drawbars on the upper manual has been arranged so that 5-1/3' is replaced by 1-1/7'—which is a very odd frequency in relation to the chromatic compass but is possible because of the tone generator used. It is seven times the fundamental frequency and if the latter is taken as A 440Hz, 3080Hz falls between F and G! Use of the seventh harmonic is not new (one of the overtones in a square wave in fact) but this unusual mutation drawbar provides extra tone colouring.

Other drawbars control 8' and 4' string tone, variable sustain, attack, decay and ambience (a form of reverberation). A single switch alters the compass of the manuals down one octave so that 32' to 2' pitches are substituted. The lower manual has no mutations but separate sustain covers the 16' 8' 4' 2' and 1' pitches. Presets, three vibrato controls and a transposer also feature. Priced at under £700 including VAT, the HX80 will interest those looking for a comprehensive but fully portable instrument (though a tone cabinet is required). Basically this is a 'straight' organ without too many gimmicks and is made in the UK.

PIANOCORDER

The Pianola is a rarity these days and, like the fairground organ, uses punched paper rolls. The Kemble Piano Co. has the UK marketing rights for the first electronic version—the Pianocorder. Using cassette tape, digital pulses are recorded as a pianist plays the original. On playback, pulses from the tape are decoded and applied to solenoids on playing key or expression pedal. For teaching purposes, this could be

a most useful aid as, unlike hearing a magnetic tape recording of the conventional type, the Pianocorder's keys are seen to move—and the playback speed can be varied at the teacher's will.

As this article goes to press, the Pianocorder is being demonstrated at the Chappell Music Centre, New Bond Street, London W1.

NO GO

An interest in electronic music can have its problems as friends may ask for your help in sorting out a small problem caused by transportation or rough usage. Of course, the cause is always obscure, inspection lighting inadequate and time is of essence! If the reader is inadvertently involved, a few guidelines may help:

Modern instruments, and organs in particular, have circuitry that is far more complex than a decade ago. A service manual (or at least block diagram) is essential as CMOS devices are often the order of the day—and it is wise to know what you are dealing with. Instruments in the home are often on nylon carpeting so circuit boards should not be treated with abandon. Although protection diodes are normally employed, the board should be considered as an extension to the i.c. Some special purpose devices are difficult to find so if in doubt earth yourself and obviously use a low leakage iron, also earthed.

If the instrument has been moved without its back panel in place, printed circuit boards can easily get damaged. Equally, forcing it into position over misaligned guides can split the board. If a crack is diagnosed, remove p.c.b. sockets with great care as solid conductors are often used. Run a little cyanoacrylate adhesive into the crack and clamp the board flat for the few minutes it takes to set. After removing the clamp, broken tracks can be patched: clean the copper strip locally with a small, sharp blade (a scalpel is ideal) and solder small bridges of tinned copper across the breaks.

Commercial instruments are designed for easy replacement of sub-assemblies and thus contain many connectors and a wiring harness similar to those used in the car industry. Slide-on tags for multiple connection to a common point are often used and are also reminiscent of BL products. They may be perfectly good for cars, but can cause noise and breakthrough in organs if used for signal earth purposes. Vibration from internal speakers or in transportation can cause them to get noisy or to simply

disconnect themselves: remedies being either to tighten to the maximum or risk the wrath of the professional serviceman by chopping off the tags and making a really sound job by soldering!

TESTING

Digital meters can be very useful at times but, due to the time taken to sample, are not satisfactory for reading moving voltages/currents, e.g. percussion circuits. Although this also applies to analogue meters to some degree, the reaction of a pointer gives a much better idea of the circuit's action. Incidentally, I would suggest that any prospective purchaser takes a good look at a detailed specification of the digital meter in mind as advertisements often omit essential points. Despite claims of high accuracy, current readings are often subject to voltage burdens which, in some models, will cause errors of 25 per cent or so. My own digital meter, which is widely advertised, often refuses to auto-zero itself (and it's no fluke!). At least these meters normally have very high input impedance, but analogue meters should be capable of 20k Ω /V.

A box containing an i.c. amplifier, speaker and battery is a useful tool for tracing audio paths. My version has a switched jack socket so that the battery is connected when the test lead jack is inserted: I don't need to remember to switch off before putting it back on the shelf. The connecting cable has a longer earth wire, terminated by a croc clip, so that the probe (inner of the screened lead) can be taken from end to end of an instrument without having to find a new signal earth point repeatedly.

An oscilloscope is perhaps something of a luxury for fault-finding and more useful at the design stage. Assuming that there is no intention of checking the waveform supplied to a top octave synthesiser, which will be in the MHz range and a simple square wave, the timebase range needs to be modest only. Vibrato will call for about 7Hz as the slowest trace, though unless there is good persistence, flicker makes the display's usefulness questionable. About 15kHz will be just about the maximum required for the X plate, but a simple 'scope will show whether the waveform is what it is supposed to be and if tone filters are having the required effect. Be careful where the 'scope is connected into the instrument, that the input coupling capacitor is above suspicion and if it should be earthed for that particular application. If in doubt with expensive CMOS devices, test elsewhere!

I have not mentioned a CR bridge as most meters can cope with resistor measurement fairly well. Some can also handle capacitors but if not, a single 4001 package and a few discrete components can be assembled into a direct reading capacitor adaptor. Where valve circuitry is concerned—and plenty still exist and work well—it pays to check values of discrete components. Resistors can change their values alarmingly when age and high voltages come into play, so don't always believe the colour of the multiplier band: 10k resistors can easily become 100k in the course of time.



ELECTROSTAT

P. DAKIN

There are many types of thermostat in use around the home and in industry. The majority make use of the bi-metal strip or bar and for general use there is nothing wrong with them. However, where a wide range, close tolerance, or economical use of energy is required an electronic thermostat is far superior.

The advantage of the circuit to be described here is that the temperature of a body can be held, if necessary, to within 0.1°C (ideal for photographic and similar uses) and the economic value is that it does not overheat the body wasting energy in the process. (Fig. 1).

The circuit is designed to give bursts of full power, of sufficient duration to overcome heat losses, and maintain a constant temperature.

The circuit is built around the SGS-ATES L121 i.c. This device incorporates a zero voltage detector, amplifier, comparator, internal power supplies with reference voltages, output stage for triggering triacs or thyristors, and will work from any single phase 50-60Hz supply.

CIRCUIT DESCRIPTION

The block diagram of the L121 is shown in Fig. 2. The supply to the i.c. is via R_s which limits the supply current to approximately 30mA. The mains frequency signal is clipped to $\pm 12\text{V}$ at the clipping and rectification stage. Pins 8 and 10 are the i.c.s dual internal supply smoothing. This smoothed supply is regulated by the voltage regulating stage, and used to supply the other internal functions, and also produce a 1.5 volt reference at pin 4, and a positive supply output at pin 6.

The zero cross detector determines when the supply on pin 9 crosses zero volts relative to pins 12 and 13, and whether it is positive, or negative going. Signals from this detector are fed to the ramp generator, control logic and output logic stages. The ramp generator produces a linear rising voltage, the rate of which is determined by C1 (pin 1) and R1 (pin 16). This is referred to as the time base. The voltage of this ramp is from less than 1V to approximately 6V, and is fed directly to the comparator.

The amplifier is a high gain d.c. operational amplifier, which is used to amplify the sensor output voltages to a suitable level for the comparator.

The comparator compares the relative levels of the amplifier output (pin 2) with the time base signal (pin 1) and will enable the chopper and control logic when the ramp voltage exceeds that of the amplifier output, as shown in Fig. 3. This has the effect of turning the heater on and off at a rate set by the time base.

The output logic stage determines the gate pulse polarity

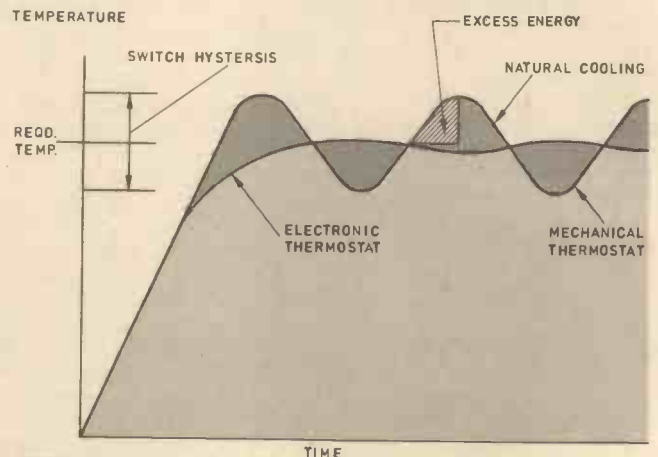


Fig. 1. Comparison between an electrical and a mechanical thermostat

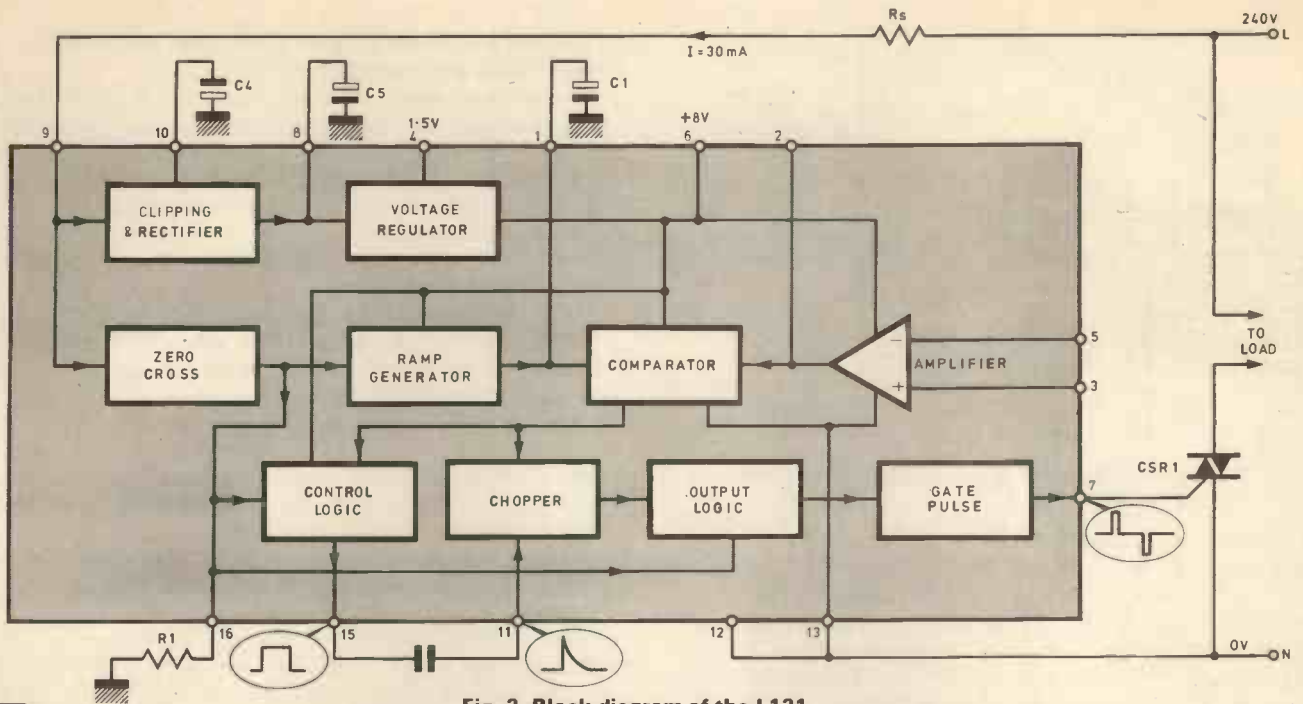


Fig. 2. Block diagram of the L121

EG255

and also synchronises it with the zero voltage of the mains (zero cross detector signal), and this reduces radio frequency interference (R.F.I.) produced by the triac to a very low level.

The control logic is such that it triggers the chopper stage when the voltage from the ramp exceeds the voltage at pin 2, and for a time dependent on C2, which means that the triac switches at a mains voltage of approximately $\pm 10V$.

To eliminate this transistor TR2 can be included, as in Fig. 4. TR2 inverts the control logic output and feeds it directly to

the chopper. This has the effect of triggering the triac while the mains voltage is between zero and $\pm 10V$, which will reduce R.F.I. to an absolute minimum.

Fig. 4 shows the circuit diagram of the Electrostat with a range of $20^{\circ}C$ to $80^{\circ}C$, and a variable time base of 1 second to 1 minute. The minimum time base of the system is set by R1 and C1, with TR1 increasing the value of C1 by a factor dependent on the gain of the transistor as the resistance of VR1 is increased.

Diodes D2 and D3 are extra clipping diodes so that under fault conditions, no high voltages can appear on pin 9 of the L121. The sensor of the system is a thermistor connected to the inverting input of the amplifier as shown in Fig. 5. The 1.5 volt reference from pin 4 is compared with the potential difference across the thermistor, the difference amplified, and fed to the comparator as described earlier. If the amplifier gain is high i.e. 50 or more, any small change in

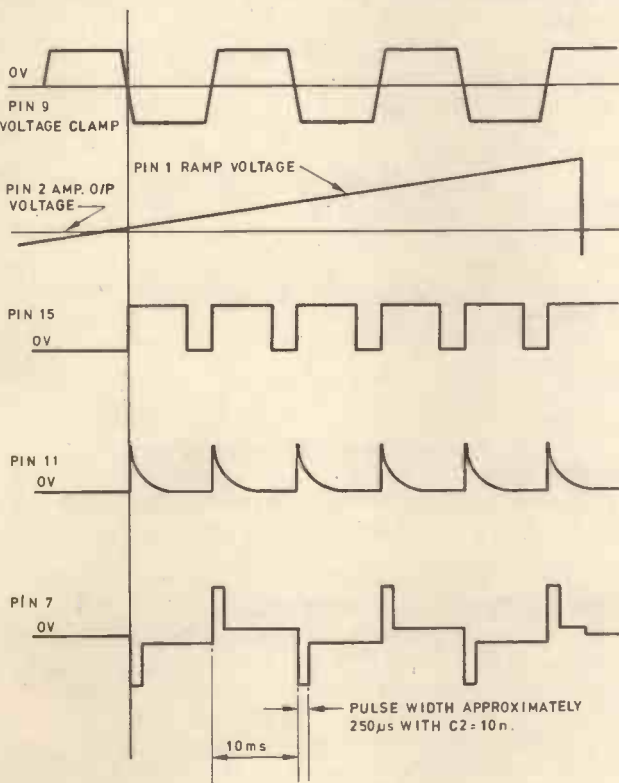


Fig. 3. Waveforms and phase relationships

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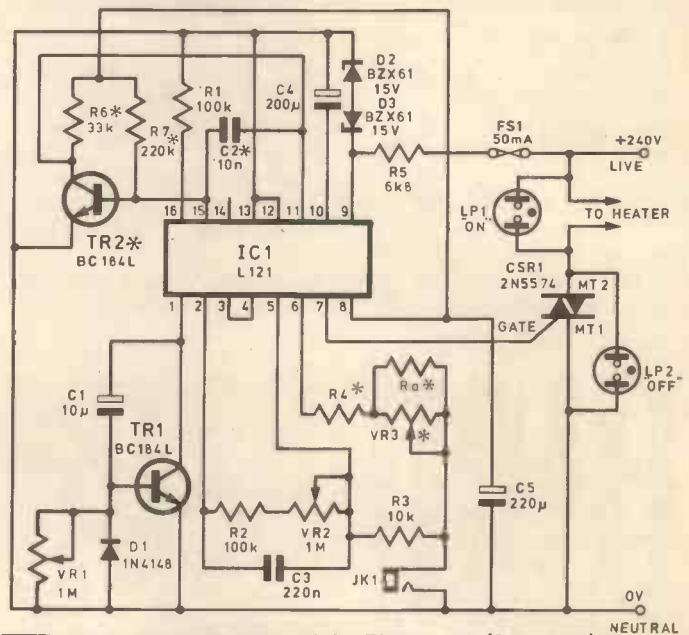


Fig. 4. Circuit diagram of the Electrostat (*see text)

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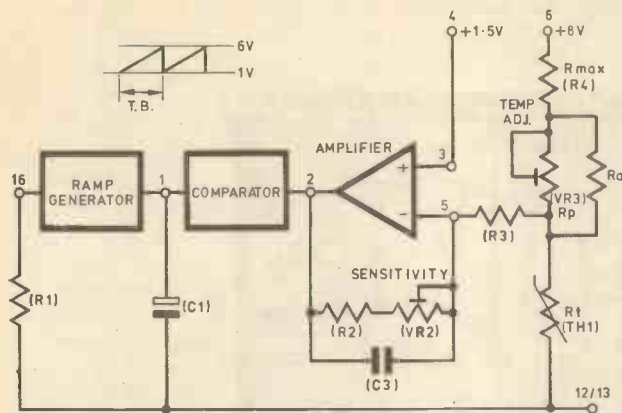


Fig. 5. Block diagram of the amplifier, comparator and ramp generator

thermistor potential will result in a large change in output from the amplifier, which will be acted upon by the other circuit functions, to return the sensor voltage near to the reference level. This situation is ideal for photographic work where a temperature can be controlled to within 0.1°C or better, depending upon the location of the sensor, and the thermal resistance of the object being heated.

The exact value of time base and temperature setting components depends upon every particular application. Table 1 lists some possible uses and suggested values.

Application	Temp. Range Centigrade	Time Base and C1 Value	Amplifier Gain R4 & VR3 Value
Photographic Heater	20-40	1s-5s	50-100
Fermentation Heater	15-30	10μ-47μ	500k-1M
Fish Tank Heater	20-30	1s-30s	20-50
Room Heater	20-30	10μ-330μ	200k-500k
Room Heater	15-25	30s-120s	20-50
Room Heater	15-25	60s-180s	20-50
Immersion Heater	40-80	330μ-1200μ	200k-500k
Heater	40-80	15s-120s	20-70
Heater	40-80	150μ-1200μ	200k-700k

TABLE 1

TIME BASE SELECTION

The time base is set by R1 and C1. This time must be shorter than the time it takes for the heat to travel through the medium to the thermistor (thermal time constant). Transistor TR1 is used to increase the effective value of C1 to several hundreds of times its true value when long time bases are required, and where a large capacitor would be impractical. Typical time bases range from 1 second to several minutes.

The manufacturers of the i.c. recommend that R1 be set at 100k and C1 selected to give the required time using the formula:

$$C1 = \frac{1.2 TB}{R1}$$

where; VR1=0Ω

R1=100k

C1 in Farads

TB in seconds (1 second minimum)

When the time base needs to be 1 minute or more, T1 is used to magnify C1:

$$C1 \text{ (effective)} = C1 \text{ (true value)} \times TR1 \text{ hfe}$$

where TR1 is a BC 184L

hfe = 250 to 900

and VR1 value is greater than 1M or omitted.

THERMISTOR AND TEMPERATURE SETTING RESISTOR SELECTION

A thermistor similar to the types GM 473 or VA 3410 is recommended as their operating range is -60°C to +200°C, and their resistance is fairly large at mid-range (approximately 47k at 25°C). The range of resistances at various temperatures is given in Table 2. Referring to Fig. 5, it can be shown that:

$$R(\max) = 5.33 R_t(\max \text{ temp}) \quad (1)$$

and Rp paralleled with Ra

$$R_p/R_a = (5.33 \times R_t(\min \text{ temp}) - R(\max)) \quad (2)$$

If Rp parallel with Ra is a preferred value ±20 per cent, use the preferred value. If not use the next biggest preferred value and calculate Ra:

$$R_a = \frac{(R_{pv}/R_a) \times R_{pv}}{R_{pv} - (R_{pv}/R_a)} \quad (3)$$

Select the range of temperature required and substitute in the equations the relevant values of thermistor resistances from Table 2.

Temperature (Centigrade)	Approx Resistance (Ohms)
-60	10M
-30	1M
0	159k7
10	100k
20	59k
25	47k
30	37k7
40	24k8
50	16k7
60	11k5
70	8k1
80	5k9
90	4k3
100	3k2
150	1k0

TABLE 2

Example

Photographic Heater

Preferred range 20°C to 40°C

$$R(\max) = 5.33 \times R_t(\max)$$

$$= 5.33 \times 24.8k$$

$$= 132.2k$$

Nearest preferred value 120k

$$R_p/R_a = (5.33 \times R_t(\min) - R(\max))$$

$$= (5.33 \times 59k) - 120k$$

$$= 194.5k$$

Nearest preferred value 220k

$$R_a = \frac{(R_{pv}/R_a) \times R_{pv}}{R_{pv} - (R_{pv}/R_a)}$$

$$R_a = \frac{194.5k \times 250k}{250k - 194.5k}$$

$$= 876k$$

Nearest preferred value 1M5

Amplifier Gain

The amplifier gain is set by the ratio of R2, VR2, and R3:

$$\text{Gain} = \frac{R2 + VR2}{R3} \quad (4)$$

CONSTRUCTION AND TESTING

The p.c.b. design and component layout for the Electrostat are shown in Figs. 6 and 7. Triac CSR1 should be of an adequate rating and mounted on a heatsink using a suitable insulating kit.

Care must be taken when assembling the triac on to the heatsink, and ensure the neutral is connected to the MT1 terminal. For 110V supplies R5 must be reduced to 3.3k 4W. Fuse FS1 is 50mA, and the load should be fused at a suitable value.

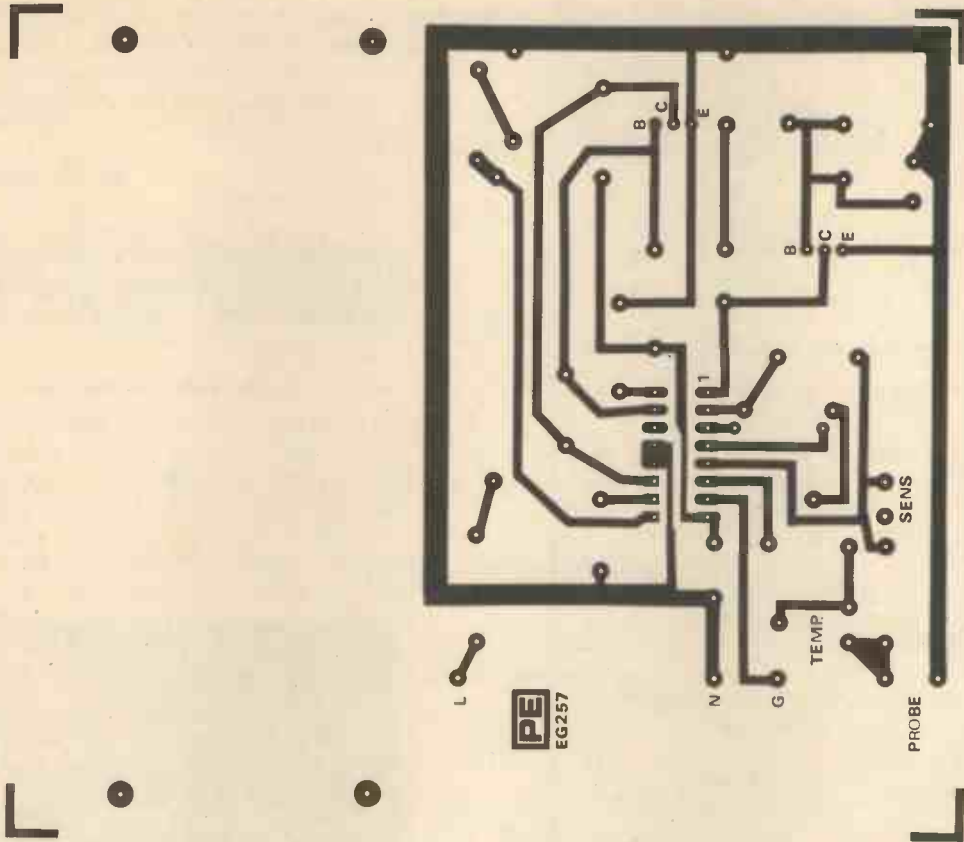


Fig. 6. P.c.b. design

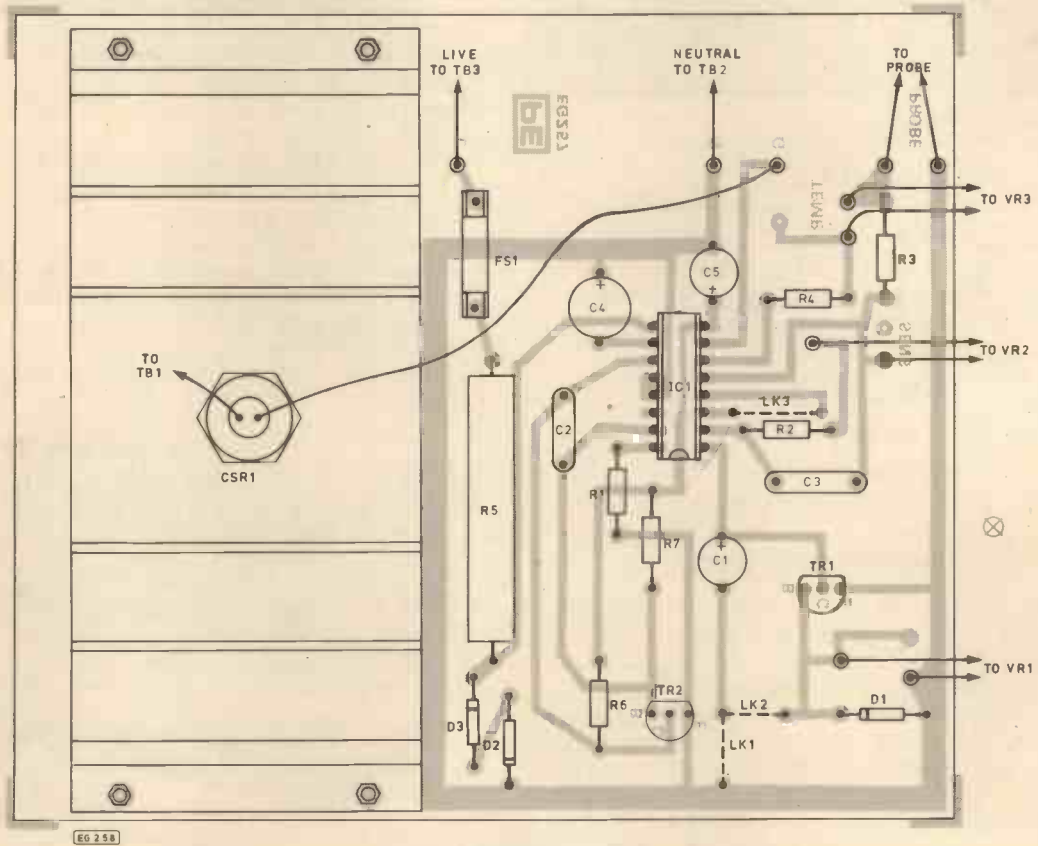


Fig. 7. Component layout

COMPONENTS . . .

Resistors

R1, R2	100k (2 off)
R3	10k
*R4	
R5	6k8 7W (3k3 4W for 110V)
R6	33k
R7	220k
*Ra	

All resistors $\frac{1}{4}$ W 10% except otherwise stated.

Potentiometers

VR1, VR2	1M Lin. (2 off)
*VR3	

Capacitors

*C1	
C2	10n
C3	220n
C4, C5	220 μ 16V (2 off) elect.

Semiconductors

D1	1N4148
D2, D3	BZX61 15V (2 off)
TR1, TR2	BC184L (2 off)
CSR1	2N5574 (or any 400V triac with a suitable current rating for the load)
TH1	VA3410
IC1	L121 (Quarndon Electronics, Slack Lane, Derby.)

Miscellaneous

LP1, LP2	Mini mains neon (2 off)
FS1	50mA fuse
P.c.b.	
Connector block	
Heatsink to suit triac	
Suitable case	
JK1 3.5mm jack socket	
*See text	

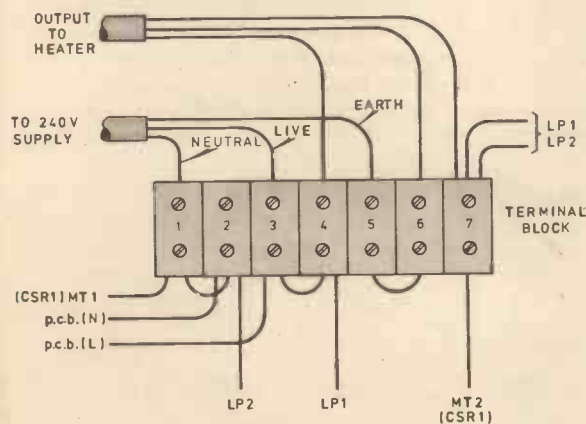


Fig. 8. Terminal block wiring

The wiring for the terminal block is shown in Fig. 8. The thermistor can be built into a probe as suggested in Fig. 9 and is connected to the circuit via JK1.

Having checked the circuit for correct placing of components, carefully examine around the i.c. for solder splashes and joined tracks. When all is satisfactory, connect the load and the supply, and switch on. If the thermistor is "cold", relative to the temperature setting, the "off" neon indicator LP2 will light immediately. If the "on" neon LP1 lights adjust VR3 to minimum, and check the "off" neon lights.

If neither neon lights, turn off and check all connections. Also check pin 8 for +12V, and pin 10 for -12V, with respect to pins 12 and 13. Next check pin 1 for ramp voltage. If satisfactory check pin 2 for voltage swing by adjusting VR3. If an oscilloscope is available, check pins 11 and 15 for control pulses, and pin 7 for output pulses.

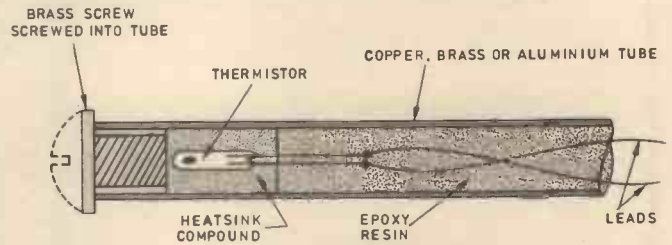
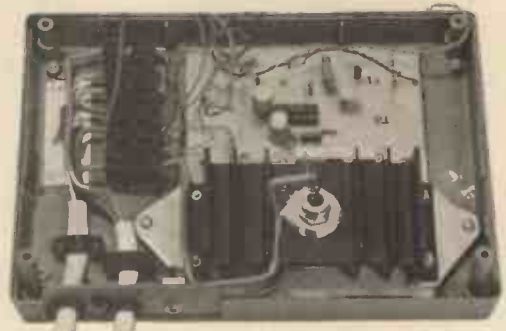


Fig. 9. Probe construction

The setting of the sensitivity resistor VR2 will depend on the thermistor location within the medium and the time base. For maximum sensitivity, a time base much shorter than the thermal time constant of the body is preferable.

The sensitivity is best set by trial and error, or if a storage oscilloscope is available, monitor pin 2 for several minutes and adjust VR2 until a still flat trace is obtained.



If the sensitivity is set too high, the system will be under damped and the temperature could vary by a large amount around the required value. Similar results will be seen if the sensitivity is set far too low.

If in use R.F.I. is present, try adding TR2, R6 and R7, as shown in Fig. 4, and omit capacitor C2.

Having checked that the unit works, it can now be built into a unit to suit your own application. If a metal case or front panel is used ensure it is earthed. Remember, for safety DO NOT connect the earth to the neutral.

Where calibration of VR3 is required, it should be against a digital thermometer, or a laboratory standard mercury thermometer, located in the same area of the body as the units sensor. ★





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CMOS

4001	13p	4025	13p	4069	13p
4002	13p	4026	90p	4070	13p
4007	13p	4027	28p	4071	13p
4009	30p	4028	45p	4072	13p
4011	13p	4029	50p	4081	13p
4012	13p	4040	55p	4093	36p
4013	28p	4041	55p	4510	60p
4015	50p	4042	55p	4511	60p
4016	28p	4043	50p	4518	65p
4017	47p	4046	90p	4520	60p
4018	55p	4049	25p	4528	60p

FULL DETAILS IN CATALOGUE!

TTL

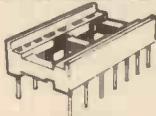
7400	10p	7475	25p	74148	90p
7401	10p	7476	20p	74150	55p
7402	10p	7485	55p	74151	40p
7404	12p	7486	20p	74154	65p
7406	22p	7489	135p	74157	40p
7408	12p	7490	25p	74164	55p
7410	10p	7492	30p	74165	55p
7413	22p	7493	25p	74170	100p
7414	12p	7494	45p	74174	55p
7420	12p	7495	35p	74177	50p
7427	20p	7496	45p	74190	50p
7430	12p	74121	25p	74191	50p
7432	18p	74122	35p	74192	50p
7442	38p	74123	38p	74193	50p
7447	45p	74125	35p	74196	50p
7448	50p	74126	35p	74197	50p
7454	12p	74132	45p	74199	90p

OPTO

LED's	0.125in.	0.2in	each	100+
Red	TIL209	TIL220	9p	7.5p
Green	TIL211	TIL221	13p	12p
Yellow	TIL213	TIL223	13p	12p
Clips	3p	3p		

DISPLAYS				
DL704	0.3 in CC		130p	120p
DL707	0.3 in CA		130p	120p
FND500	0.5 in CC		100p	80p

SKTS



Low profile by Texas

8pin	8p	18pin	14p	24pin	18p
14pin	10p	20pin	16p	28pin	22p
16pin	11p	22pin	17p	40pin	32p

3 lead T018 or T05 socket. 10p each
Soldercon pins: 100:50p 1000:370p

PCBS

VEROBOARD			
Size in.	0.1in.	0.15in.	Vero Cutter 80p.
2.5 x 1	14p		
2.5 x 3.75	45p	45p	
2.5 x 5	54p	54p	Pin insertion tool 108p
3.75 x 5	64p	64p	
3.75 x 17	205p	185p	

Single sided pins per 100 40p 40p
Top quality fibre glass copper board. Single sided. Size 203 x 95mm. 60p each.
'Dalo' pens. 75p each.
Five mixed sheets of Alfac. 145p per pack.

RESISTORS

Carbon film resistors. High stability, low noise 5%.

E12 series. 4.7 ohms to 10M. Any mix:			
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AD161	38p	BD139	35p	ZN3054	50p
BC107	4p	BD140	35p	ZN3442	135p
BC108	8p	BFY51	15p	ZN3702	8p
BC108C	10p	BFY52	15p	ZN3703	8p
BC109	8p	MJ2955	98p	ZN3706	9p
BC109C	10p	MPSA56	20p	ZN3707	9p
BC147	7p	TIP29C	60p	ZN3708	8p
BC148	7p	TIP30C	70p	ZN3819	15p
BC177	14p	TIP31C	65p	ZN3820	44p
BC178	14p	TIP32C	80p	ZN3904	8p
BC179	14p	TIP2955	65p	ZN3905	8p
BC182	10p	TIP3055	55p	ZN3906	8p
BC182L	10p	ZTX107	14p	ZN4058	12p
BC184	10p	ZTX108	14p	ZN5457	32p
BC184L	10p	ZTX300	16p	ZN5459	32p
BC212	10p			ZN5777	50p

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BC477	19p	1N914	3p	1N4006	6p
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100						8p
		220				10p
			470			15p
1000						23p

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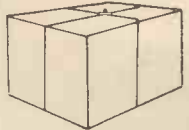
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DC Current	250uA 250mA
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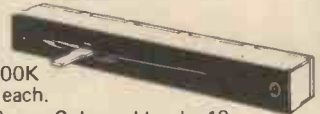
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MICRO-BUS

Compiled by DJD.

Appearing every two months, Micro-Bus will present ideas, applications, and programs for the most popular microprocessors; ones that you are unlikely to find in the manufacturers' data books. The most original ideas will probably come from readers working on their own microcomputer systems, and payment will be made for any contribution featured here. This is also the place to air your views, in general, on this new technology, so let's be hearing from you!

MICRO MAGICIAN

THE following unusual program turns the computer into a magician, enabling it to find a card chosen secretly by a spectator. Two versions of the program are given; one in BASIC, and one in machine code for the 6502.

The presentation of the trick is as follows: only the thirteen cards of one suit are used for the trick. The cards are fanned out face down, and the spectator removes one and remembers its value. The remaining pile of cards is cut once, and the spectator replaces the selected card wherever he likes. The pile is then divided into two, and the two halves are shuffled together. Finally the cards are fanned out face-up on the table, and the order of the cards is typed into the computer. After a brief pause the computer announces which was the chosen card.

Although the trick is based on a simple principle, it leads people to believe that the computer's powers extend to mind-reading, and this belief is strengthened by occasional failures of the computer to guess the card correctly.

By way of illustration, try locating the chosen card in the sequence shown in Fig. 1. The cards were originally in order, ace up to king from left to right, and the sequence shown is the result of following the procedure described above.

HOW THE PROGRAMS WORK

The programs work by comparing the new order of the cards with their previous order; for each card a number is calculated which represents how far the processes of shuffling and cutting have moved that card away from its previous neighbours. The higher this score, the more out of sequence is the card concerned. After calculating this score for each of the 13 cards, the card with the highest score is taken to be the one that was chosen, and in most cases this will be correct. However, in some cases the computer cannot be certain about which card was chosen; for example, if the card is returned to its original position then any card could have been chosen. Also, less obviously, if the card is re-inserted one place to one side of its original position it is impossible to tell whether it, or its neighbour, was the chosen card since the same sequence of

cards would result in each case. The reason for cutting the pack before the card is replaced is to encourage the spectator to replace the card in a different position.

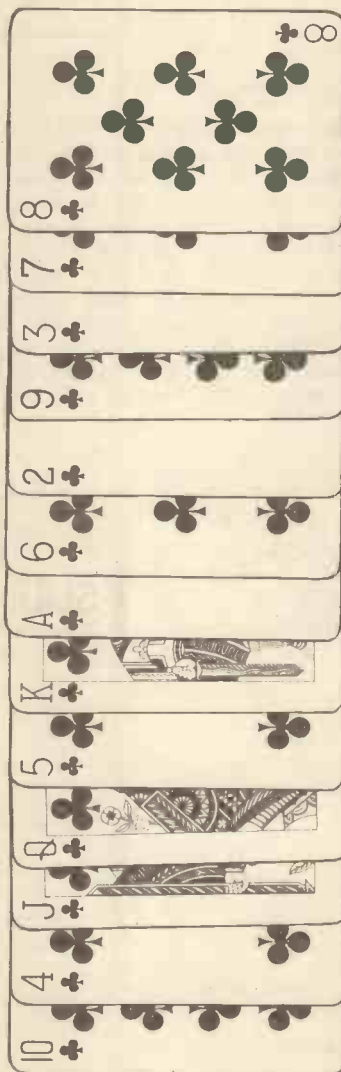


Fig. 1. Find the chosen card in this sequence, which results from a presentation of the trick described in the text

When the programs are first executed they assume that the cards were originally in numerical order, ace up to king. If the cards are not initially in order the program will probably get the first attempt wrong, but in some ways it is more impressive to do the trick without arranging the cards first, and attribute the initial failure to "warming up"! On subsequent operations of the program the initial order is replaced by the new order of the cards, as typed in; the order of the cards should not be disturbed when the trick is repeated.

CARD TRICK PROGRAMS

The two versions of the card-trick program work in comparable ways. The cards are represented internally by the numbers 1 to 13. The programs are divided into three sections. First, the new sequence of cards is read in. Secondly, each of the cards in the previous sequence is searched for in the new sequence, and its position there is subtracted from the positions of each of the cards that were its neighbours in the previous sequence. The sequence is considered to be circular, so if the difference between two positions turns out negative then 13 is added to it. The card's distance from one neighbour, plus its distance from the other neighbour, is saved as that card's score. Finally, the card with the maximum score is found and displayed as the chosen card.

The critical part of the trick is that the cards should only be shuffled once, and the shuffle should be of the sort that divides the packet of cards into two halves and merges the two halves back into one pile (e.g. a riffle shuffle). The cards can be cut at any time, and as many times as wished, but each time the packet of cards should only be cut into two piles.

BASIC VERSION

The BASIC version of the card trick, Fig. 2, closely follows the above description. The subroutine at 500 looks up the card T in the array containing the previous sequence of cards, A, and returns in X the position of T in that array. The program uses only integer arithmetic, and so can work with integer-only BASIC interpreters.

```

0010 REM *** CARD TRICK ***
0020 DIM A(13),B(13),S(13)
0030 FOR J=1 TO 13: A(J)=J: NEXT J
0040 PRINT "ENTER YOUR CARDS"
0050 FOR J=1 TO 13: S(J)=0
0060 INPUT B(J): NEXT J
0070 T=A(13): GOSUB 500
0080 FOR J=1 TO 13: L=X: R=T
0090 T=A(J): GOSUB 500
0100 Q=X-L: IF Q<0 THEN Q=Q+13
0110 S(T)=S(T)+Q: S(R)=S(R)+Q
0120 NEXT J
0130 M=0
0140 FOR J=1 TO 13: A(J)=B(J)
0150 IF S(J)<M GOTO 170
0160 Z=J: M=S(J)
0170 NEXT J
0180 PRINT "YOU PICKED THE "; Z
0190 GOTO 40
0500 FOR K=1 TO 13
0510 IF T=B(K) THEN X=K
0520 NEXT K: RETURN

```

Fig. 2. BASIC version of the card trick enables the computer to find a chosen card

6502 VERSION

The program for the 6502, Fig. 3, was developed on an Acorn system and uses two routines in the Acorn monitor, so if the program is used with another 6502 system these will have to be modified. Subroutine DISPLAY is a display and keyboard-scanning routine which displays the segment patterns from locations 0010-0017, and waits for a keypress (although it can be made to give a single sweep of the display without waiting). It returns with the hex value of the key pressed in the A-register, the X-register is preserved, and the Y-register is zeroed. Subroutine HEXTD generates the segment pattern for the hex number in the lower four bits of the A-register, and stores this in display location Y.

When the program is executed at 0200 the display will go blank. The sequence of cards should then be entered using the hex keys, with ace = 1, ten = A, jack = B, queen = C, and king = D. When the last card is entered the micro will display its guess at the chosen card. The program is then ready to repeat the trick.

<pre> ; CARD TRICK ; **** ***** DISP = \$0010 DISPLY = \$FE0C HEXTD = \$FE7A ; 0000 .= \$0020 0020 OLD .= +14 002E NEW .= +14 003C SUM .= +14 004A LAST .= +1 004B POSN .= +1 004C MAX .= +1 004D .= \$0200 0200 DB BEGIN 0201 AO 00 LDY EO 0203 A2 0D LDX E13 0205 8A SET TXA 0206 95 20 STA OLD,X 0208 94 0F STY DISP-1,X 020A CA DEX 020B D0 F8 BNE SET 020D E8 GET INX 020E 20 0C FE JSR DISPLY 0211 95 2E STA NEW,X 0213 94 3C STY SUM,X 0215 84 12 STY DISP+2 0217 20 7A FE JSR HEXTD 021A E0 0D CPX E13 021C D0 EF BNE GET 021E A5 21 LDA OLD+1 0220 85 4A STA LAST 0222 A0 0D LDY E13 0224 B9 20 00 LOOP LDA OLD,Y 0227 20 6A 02 JSR SCAN 022A 86 4B STX POSN 022C A5 4A LDA LAST 022E 20 6A 02 JSR SCAN 0231 8A TXA 0232 38 SEC </pre>	<pre> ACORN DISPLAY BUFFER DISPLAY ROUTINE HEX DIGIT TO BUFFER ; FREE ZERO PAGE PREVIOUS SEQUENCE NEW SEQUENCE VALUES CALCULATED ; START OF PROGRAM ; ASSUME CARDS ARE IN ORDER. ; CLEAR DISPLAY ; START AT 1 ; GET CARD CLEAR SUM CLEAR RESULT KEY TO DISP+0 ALL DONE? ; FIRST CARD ; DO NEXT CARD ; SAVE POSITION ; SUBROUTINE SCAN SCAN LDX E13 SCAN1 CMP NEW,X SCAN1 BEQ RETN SCAN1 DEX SCAN1 BNE SCAN1 RTS ; END </pre>	<pre> 0233 E5 4B SBC POSN 0235 B0 02 BCS OK 0237 69 0D ADC E13 0239 48 PHA 023A 18 CLC 023B A6 4A LDX LAST 023D 75 3C ADC SUM,X 023F 95 3C STA SUM,X 0241 B6 20 LDX OLD,Y 0243 86 4A STX LAST 0245 68 PLA 0246 75 3C ADC SUM,X 0248 95 3C STA SUM,X 024A 88 DEY 024B D0 D7 BNE LOOP 024D A2 0D LDX E13 024F A9 00 LDA EO 0251 B4 2E FIND LDY NEW,X 0253 94 20 STY OLD,X 0255 D5 3C CMP SUM,X 0257 10 04 BPL NOTGT 0259 86 4C STX MAX 025B B5 3C LDA SUM,X 025D CA NOTGT DEX 025E D0 F1 BNE FIND 0260 A5 4C LDA MAX 0262 A0 7A LDY E2 0264 20 7A FE JSR HEXTD 0267 4C 0D 02 JMP </pre>	<pre> MAKE IN RANGE 0-12 ; REPEAT FOR OTHER NEIGHBOUR. ; DO FOR ALL 13 CARDS. LOOK FOR MAX TRANSFER NEW ORDER TO OLD. ; GREATEST SO FAR UPDATE WITH VALUE ; DO FOR ALL 13 CARDS. BEST CARD DISPLAY IN POSITION 2. REPEAT WITH NEW ORDER ; LOOK FOR A IN ARRAY NEW. ; POSITION IS X </pre>
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Fig. 3. Card trick program for the 6502 micro

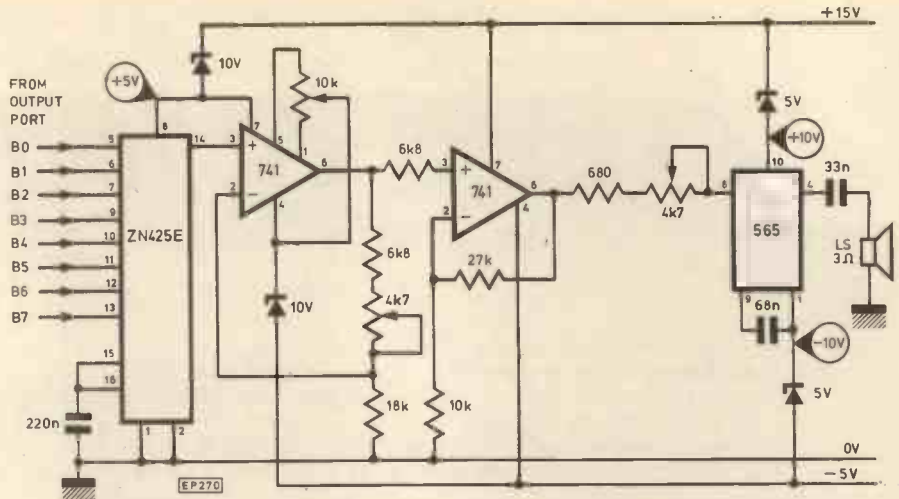


Fig. 4. Programmable frequency generator; the frequency is determined by the number at the output port

For simplicity, 14 bytes are allocated for each of the arrays, the zeroth byte not being used. Subroutine SCAN performs the same function as the subroutine in the BASIC version.

MODIFICATIONS

The programs can also guess two, or more, chosen cards, though with less reliability; for example, to find two chosen cards the programs should be modified to display the cards with the two highest scores. The programs can also be modified to work with any number of cards, and as the number of cards is increased the likelihood of the computer getting the card wrong diminishes. Thirteen cards is a happy medium; it is not too time-consuming to type in the order, while at the same time the computer's powers are indisputable!

FREQUENCY GENERATOR

The following circuit enables a micro to act as a programmable frequency generator. It uses a low-cost digital-to-analogue converter to drive a voltage-controlled oscillator, and to illustrate its operation a program is given which will play monophonic melodies. The circuit and program were submitted by *Mr. P. V. Bayley* of Newcastle, and what follows is based on his description:

"The circuit, Fig. 4, was designed for use with an 8080 microprocessor, but is suitable for any 8-bit micro. It can generate frequencies covering an overall range of at least three octaves, and incorporates a variable pitch control.

CIRCUIT DESCRIPTION

"Data from an 8-bit output port is fed to the inputs of the ZN425E 8-bit digital-to-


```

; MUSIC PROGRAM
0000 010001 MUSIC LXIB 0100H ;SET ADDRESS PTR.
0003 110008 LOOP LXID 0800H ;TEMPO FACTOR
0006 0A LDAX B
0007 D300 OUT O ;OUTPUT TO PORT
0009 C600 ADI O ;TEST ACC.
000B CA1800 JZ STOP
000E AF XRA A ;CLEAR ACC.
000F 1B DELAY DCX D
0010 BA CMP D
0011 C20F00 JNZ DELAY
0014 03 INX B ;POINT TO NEXT
0015 C30300 JMP LOOP
0018 76 STOP HLT

```

Fig. 5. Program for the 8080 micro which, with the circuit of Fig. 4, plays tunes stored in memory

```

C8,C8,C4,C4,C2,C2,BC,BC,08,08,BC,08,BC,08,BC,C2,C2,08,C2,08,C2,08.
C8,C8,C4,C4,C2,C2,C4,C4,BF,BF,08,BF,08,BF,C4,C4,CC,CC.
C8,C8,C4,C4,C2,C2,BC,BC,08,BC,08,BC,08,BC,C2,C2,08,C2,08,C2,08.
C8,C8,C4,C4,C2,C2,C4,C4,BF,BF,08,BF,08,BF,08,BF,08,BF.
C8,C8,C4,C4,C2,C2,BC,BC,08,BC,08,BC,08,BC,BC,BC,08,08.
C8,BC,D4,D4,08,D4,08,D4,08,D4,08,D4,D1,08,D4,EO,EO,DA,DA,D4,D4.
D1,D1,08,D1,08,D1,08,D1,08,D1,DA,DA,D4,D4,D1,D1.
D4,D4,08,D4,08,D4,08,CC,CC,08,CC,08,CC,08,C8,08,C8,08,C8,C2,C2.
08,C2,C2,C3,C3,BE,BE,08,BE,08,BE,C8,C8,C4,C4,C2,C2,BC,BC,08,BC,08.
BC,BC,BC,08,08.

```

Fig. 6. Data for a simple tune to illustrate how the frequency generator can be controlled by a micro

```

; TWO BYTE JUMP ROUTINE
0030 E3 RST6 XTHL ;TOP OF STACK TO HL
0031 6E MOV L,M ;L = DESTINATION
0032 E3 XTHL ;RESTORE HL, STACK
0033 C9 RET ;JUMP TO DESTINATION

```

Fig. 7. (7a above, 7b right) Routines for the 8080 which implement two-byte jump instructions

```

; TWO BYTE CONDITIONAL JUMP
; ON NO CARRY.
0038 E3 RST7 XTHL ;TOP OF STACK TO HL
0039 23 INX H ;SKIP ADDRESS BYTE
003A E3 XTHL ;RESTORE HL
003B D8 RC ;RETURN IF NOT MET
003C E3 XTHL ;GET ADDRESS BACK
003D 2B DCX H ;GET JUMP ADDRESS
003E 6E MOV L,M ;L = DESTINATION
003F E3 XTHL ;RESTORE HL, STACK
0040 C9 RET ;JUMP TO DESTINATION

```

analogue converter. The output feeds the non-inverting input of a 741 op-amp used as a buffer amplifier, and this should be calibrated as follows: first VR1 is used to give an output of OV at pin 6 when OOH is output to the port; then VR2 is used to adjust for an output of 3.6V when FFH is output to the port.

"A second 741 amplifier drives a voltage-controlled oscillator; this is the oscillator section of a 565 phase-locked loop. VR3 is used as an overall pitch control and can be adjusted to provide a maximum frequency of over 6kHz. At a mid-range setting the circuit will cover the range 200Hz to 2400Hz with good linearity. Note that the 565 operates from $\pm 10V$ supplies, derived from $\pm 15V$ with two zener diodes.

MUSIC PROGRAM

"The circuit was used with the program shown in Fig. 5, which produces simple tunes. The values required to produce a musical scale are first determined, and then the tune is written as a sequence of these values, stored in memory starting at 0100H. Different note durations can be obtained by repeating the

same note code, and blank intervals are obtained by any low-value code, such as 08H. A zero byte denotes the end of the tune.

"The program was run on a Limrose Electronics MPT8080 microtutor, and to enter data for the tune it is necessary to change the first instruction of the program to C3H (jump), and then single-step the first three bytes. After entering the data for the tune make sure that the next location contains OOH, and then change the first instruction back to 01H.

"The data in Fig. 6 are for a lively tune which should be fairly well known."

TWO-BYTE JUMP FOR 8080

Jumps in the 8080, whether conditional or unconditional, are all three bytes long: one byte for the op-code, and a two-byte destination address. The following programs, submitted by *M. R. Reynolds* of Surrey, show how to implement two-byte jump instructions which can be used to save memory in a small system. The second byte of the instruction gives the low-order byte of the destination address.

"The program of Fig. 7a uses just four

bytes of memory to implement a two-byte unconditional jump instruction. It makes use of one of the eight restart instructions, which are one-byte subroutine calls to fixed addresses. They have the format '11AAA111' (binary) and cause the return address, the address after the restart instruction, to be pushed onto the stack, and control passed to the subroutine starting at '00AAA000' (binary). The program in Fig. 7a uses the F7 instruction which jumps to a subroutine at 0030H, and the routine simply substitutes the byte following the F7 instruction for the low-order byte of the return address. The return instruction then causes a transfer to that address. Note that this is not a relative jump, although one could be implemented with a longer subroutine.

"A conditional jump can be implemented as shown in Fig. 7b. The extra increment and decrement are needed to skip the address byte if the condition is not met. This example uses the FF instruction, which causes a jump to 0038H. The condition in the conditional return instruction is the inverse of the condition required for the overall jump; for example, RNZ would be required for a 'jump if zero' instruction."

News Briefs

45,000 ICs for £52.50

THE 1980 edition of the IC Master contains more IC data than ever before. Even though some 5,000 redundant parts have been deleted, it nevertheless provides details of more than 45,000 different ICs including 5,000 new devices.

Included in the price of £52.50 is postage and packing and comprehensive up-dates which are issued automatically at 3-monthly intervals. The 1980 IC Master is available from Paterson Steadman and Partners Ltd., 34/36 High Street, Saffron Walden, Essex CB10 1EP.

GIRL OF THE YEAR

THE 1979 Girl Technician Engineer of the Year is Mrs Ann Cox-Horton, age 26, an electrical contracts engineer from Chertsey, Surrey. At a recent ceremony in London she was presented with the prize of £250 and an inscribed rose bowl by Sir Montague Finnieston, FRS, Chairman of the Committee of Inquiry into the Engineering Profession.

Sponsored by The Caroline Haslett Memorial Trust and the IBETE, this award aims to focus attention on electrical and electronic engineering as a worthwhile professional career for women.

Ann Cox-Horton is employed by T Clarke & Co Limited, a London firm of electrical contractors. She is responsible for contracts valued at up to £1½ million, including the work of up to 50 people.

The runner-up, Mrs Barbara Needham, 27, a senior research engineer from Harlow, Essex, received a special award of £150.

Barbara Needham works for Standard Telecommunication Laboratories Limited in Harlow, Essex.

4 CHANNEL MEMORY

C.R. Harding

Experimenter's design for equipping the standard oscilloscope with a charge storage facility so that pulses can be examined at leisure.

WITH the recent advent of microprocessor systems and the availability of cheap digital electronics, more home constructors are using these to replace conventional analogue circuitry. In the design, development and de-bugging of these circuits the need often arises to display a low frequency pulse train, or trains, on an oscilloscope for visual analysis. This is quite difficult on a normal oscilloscope because of the nature of the time base scan, especially if the pulses are narrow. Also charge storage and memory scopes tend to be prohibitively expensive for the amateur.

It was decided therefore to produce a unit which would display pulses of TTL level (+5V) on a normal oscilloscope with no modifications to the oscilloscope. This would include the facility for storing a train of pulses and freezing them on the screen for closer scrutiny. To extend its usefulness further it was designed to display four separate inputs on the oscilloscope as four separate traces.

BLOCK DIAGRAM

Fig. 1 shows a block diagram of the complete unit. The four input signals are fed into a quad tri-state buffer the output of which is multiplexed by a four-way data selector. This is clocked by clock 1 via a divide by four counter and the resulting waveform is available from an output stage in a single wire format.

Outputs from the tri-state buffer are also routed to the memory storage section. This consists of a 256 x 4 RAM which is controlled by a fast clock (clock 2) and an eight bit binary counter. Another clock (clock 3) operating at a slow rate steps another eight bit counter. The outputs of the counters are fed to a sixteen bit comparator which produces controlling pulses for the output trigger and the read/write phase of the RAM.

CIRCUIT DESCRIPTION

The input channels (Fig. 2) are fed into a quad tri-state buffer IC1. This is controlled by a signal from the comparator circuitry which enables information to appear at the output



SPECIFICATION

- 4 TTL compatible channels
- 3 ranges of sweep time (1) 0-1s
(2) 0-10s
(3) 0-100s
- Manual trigger mode or continuous memory cycling
- Trigger output for oscilloscope with l.e.d. indicator
- Mains power supply
- Frequency response using all channels
 - (1) Store mode—0-250Hz
 - (2) Free run mode—0-10MHz

of the buffers with a high level. With no enable present the output of the buffers are high impedance and do not interfere with the read phase of the RAM. These outputs are then multiplexed by IC2, a four way data selector. This is controlled by a two stage counter (IC3) and a clock (IC4) which is formed from a Schmitt trigger and runs at approximately 500kHz.

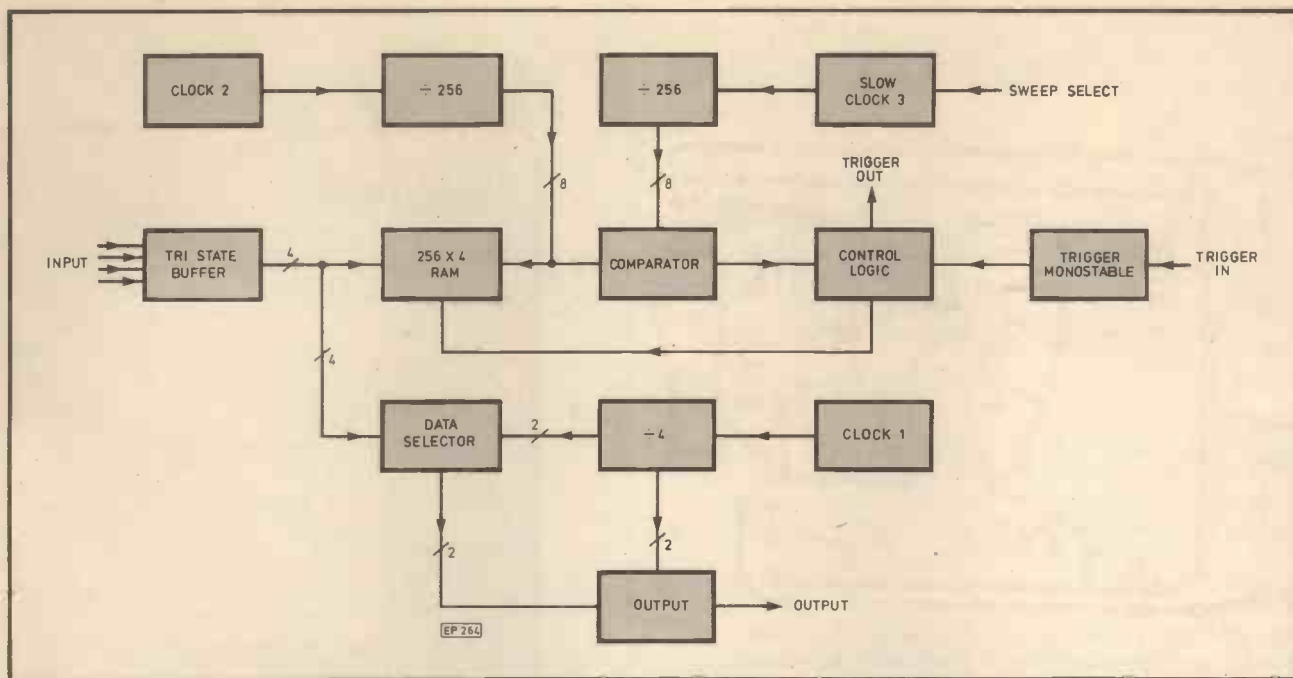


Fig. 1 Block diagram

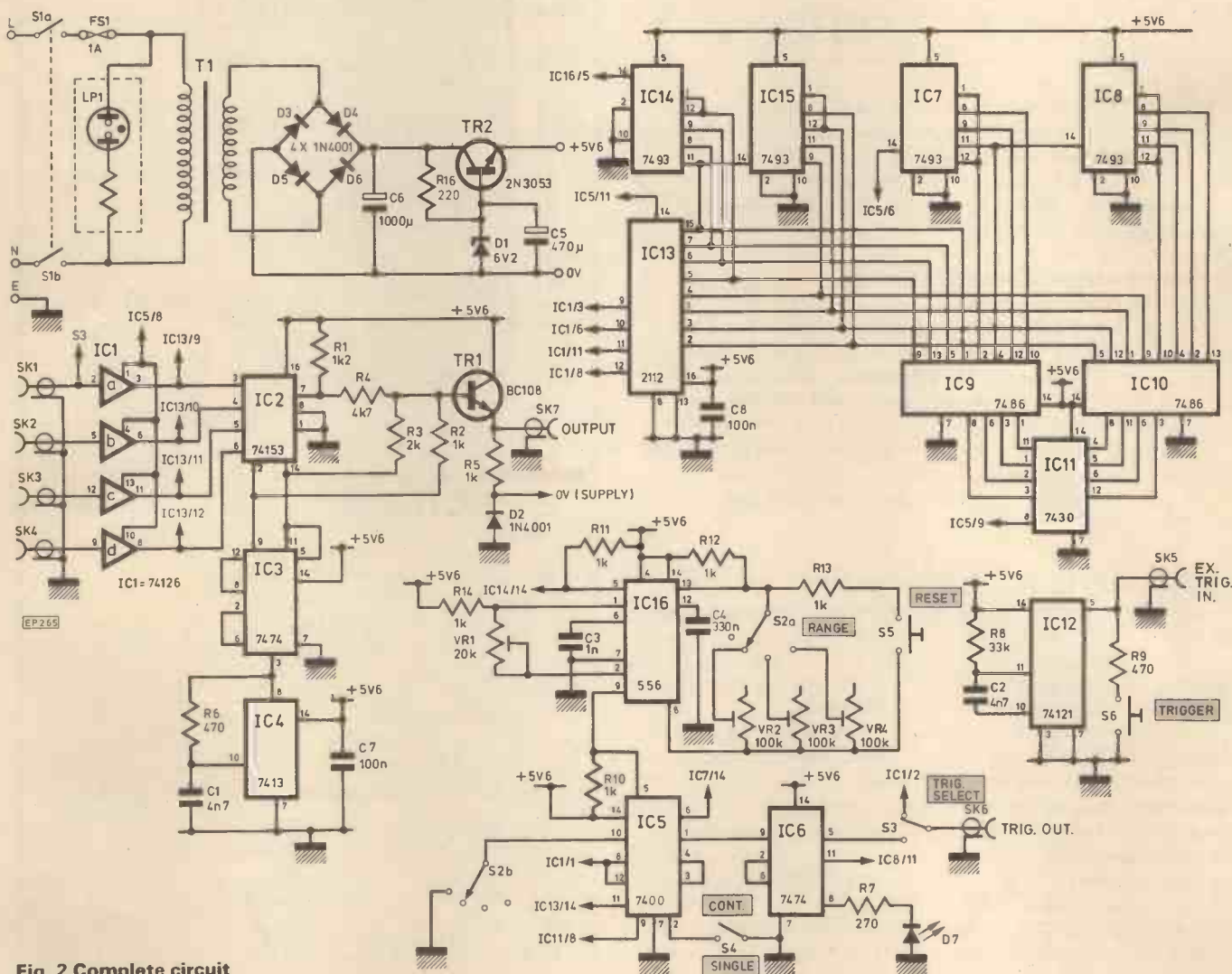


Fig. 2 Complete circuit

TR1 and its associated components form a simple three bit digital to analogue converter. The three inputs to this R2, R3 and R4 are connected to the signal output of IC2 and its two data selector input lines. Information on these last two lines effectively shifts the level of the output from TR1, and gives four distinct traces on the oscilloscope screen. TR1 also provides a buffered, low impedance output and is biased in the emitter follower mode by D2 which has its anode connected to chassis potential (as do the ground connections of the TTL i.c.s) and the cathode to power supply zero volts. This effectively produces a negative bias on the emitter of TR1 and eliminates the need for a negative supply rail.

STORAGE

The storage section of the unit consists of a 256 x 4 RAM IC13, and this is also connected to the outputs of the tri-state buffers. The address lines on the RAM are controlled by an eight bit counter IC14 and IC15. These are clocked by a counter running at 256 kHz.

These address lines are also fed into a sixteen bit comparator (IC9 and IC10).

The other eight inputs of the comparator are provided by the outputs of another eight bit counter (IC7 and IC8) which is clocked by a calibrated oscillator. This is the sweep oscillator which provides the scanning times.

Thus the signals from the two eight bit counters are compared and IC11 produces a pulse only when the outputs from these are coincidental. This pulse is then used to produce a write instruction to the RAM which then stores input data.

The frequency of the slow clock (clock 3) is set to values of 256Hz, 25.6Hz to give variation in storage rate. Once the waveform has been stored, at an appropriate scanning rate, the output scanning rate can be increased to expand out the trace to see more detail. Triggering in the unit is provided by IC5 and IC6. This is available in three modes:

1. Single shot
2. External or internal trigger
3. Continuous run

In the first mode the monostable IC12 gives a pulse from push button switch S6 from a manual trigger when desired. In the external or internal trigger mode triggering is achieved from an external pulse or from input channel 1. An l.e.d. is used to indicate correct triggering.

S4 selects the continuous triggering mode which causes the unit to cycle information into the memory continuously. The output trigger pulse is available through S3 via SK6 and is of TTL level.

POWER SUPPLY

This is a conventional transformer derived mains power supply with a full wave diode bridge rectifier.

Zener diode D1 provides the reference voltage and TR1 supplies the necessary current, which is approximately 300mA, and should have a heatsink attached.

CONSTRUCTION

The prototype was built on two pieces of Veroboard, which were stacked horizontally, one containing the memory and address dividers and the other the oscillators, trigger circuit and power supply. However, the layout of these is not critical and everything could be accommodated on one board.

A case was made for the unit out of aluminium but proprietary cases can be used.

The prototype was made compatible in appearance with

COMPONENTS

Resistors

R1	1k2
R2	1k
R3	2k
R4	4k7
R5	1k
R6	470
R7	270
R8	33k
R9	470
R10	1k
R11	1k
R12	1k
R13	1k
R14	1k
R15	910k
R16	220

0.5W 5% carbon

Capacitors

C1	4n7	mylar
C2	4n7	mylar
C3	1n	mylar
C4	330n	carbonate
C5	470 μ	10V elect
C6	1000 μ	10V elect
C7	100n	mylar
C8	100n	mylar

Integrated Circuits

IC1	74126	IC9	7486
IC2	74153	IC10	7486
IC3	7474	IC11	7430
IC4	7413	IC12	74121
IC5	7400	IC13	2112
IC6	7474	IC14	7493
IC7	7493	IC15	7493
IC8	7493	IC16	556

Transistors

TR1	BC108
TR2	2N3053

Diodes

D1	BZY88 — 6.2 400 mW Zener
D2-D6	1N4001 (5 off)

Potentiometers

VR1	20k
VR2-VR4	100k (3 off)

Switches

S1	Mains on/off double pole
S2	Two pole four way
S3	Single pole change-over
S4	Single pole on-off
S5-S6	Press to make

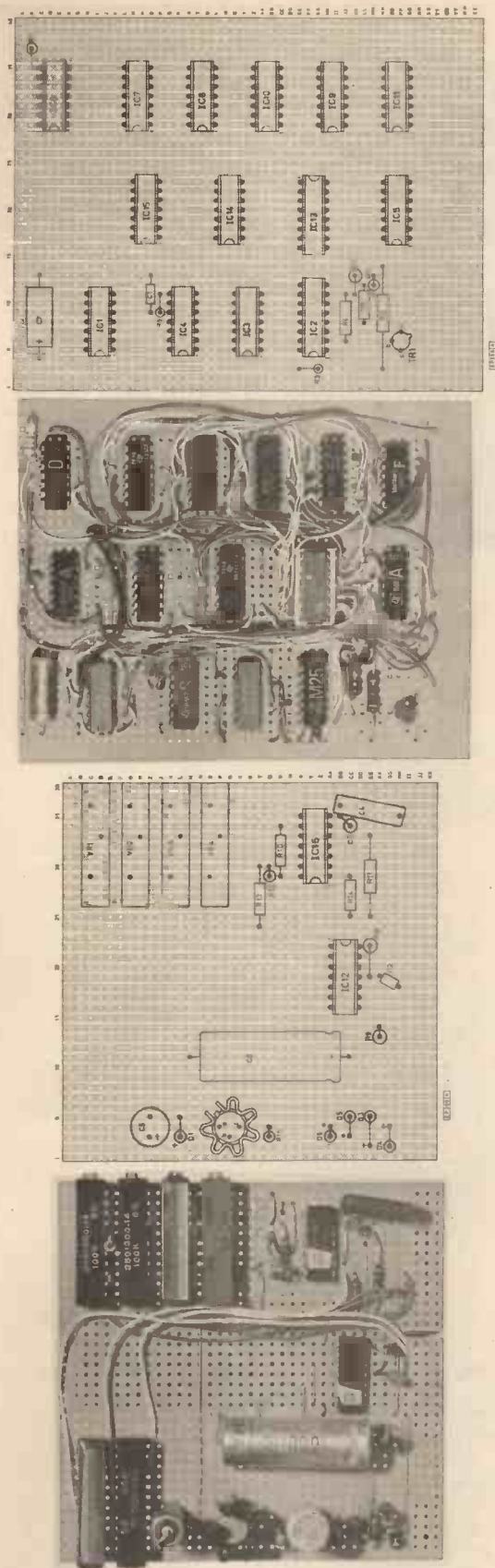


Fig. 3 Component layout for prototype boards. Since the unit is primarily a piece of experimenters test gear which will readily translate for stripboard assembly no cutting or interwiring details are given just a guide for component placement

the author's oscilloscope with the front panel containing input sockets and selection switches. The back panel houses the output and trigger sockets and mains input lead switch and fuse.

SETTING UP AND CALIBRATION

For accurate calibration a digital frequency meter should be used, however, a signal generator can be used for frequency comparison.

The following adjustments should be made:

Adjust	S2 range	Measure	Measurement pins
VR1	—	256 kHz	IC16/5
VR2	1s	256Hz	IC16/9
VR3	10s	25.6Hz	IC16/9
VR4	100s	2.56Hz (390ms)	IC16/9

USING THE UNIT

The unit should be connected to the Y input of one of the oscilloscope's channels (or channel) and the external trigger of the oscilloscope connected to the trigger output of the unit. The Y gain of the oscilloscope should be set to approximately 0.5 volts per centimetre and the timebase to 100 milliseconds per centimetre.

Four traces should appear on the screen with random information on each one. This can be cleared by pressing the sweep trigger button leaving four straight lines.

Connect a low frequency pulse generator (1Hz approximately) to the input (channel 1) and switch the unit to the 10 second sweep range. A trace of 1Hz pulses will slowly be produced on the top trace from left to right and halt when the scan reaches the right-hand margin of the screen.

The input can now be removed and the waveform will remain until the power is removed to the unit.

The unit can also be used in continuous mode where the memory will constantly refresh at the end of each scan. Also the trace can be returned to the time origin by use of the sweep trigger button.

LIMITATIONS

When using the unit care should be taken not to exceed higher frequencies than the maximum stated.

It is useful to note that the trace is actually broken up into 256 sections and at for example 1 second sweep rate the minimum pulse duration that can be recorded is 4ms in 5. If a too high frequency is fed in, depending on its harmonic relationship with the number of bits a lower frequency is produced on the screen, therefore it is useful to know the approximate value of the frequency of the waveform to be measured.

Since the unit writes in all four channels simultaneously one waveform cannot be served while 3 new ones are recorded.

MODIFICATIONS

Although designed for TTL level signals, by using a simple CMOS buffer on the input the usefulness is extended by increasing the input impedance and the level of input. Longer sweep times can be produced by using more ranges with extra capacitance on clock 3. As the sweep time is 256 times longer than the time interval of the 556, sweep periods of up to 100,000 seconds are feasible. ★



A selection of readers' original circuit ideas. It should be emphasised that these designs have not been proven by us. They will at any rate stimulate further thought.

Why not submit your idea? Any idea published will be awarded payment according to its merits.

Articles submitted for publication should conform to the usual practices of this journal, e.g. with regard to abbreviations and circuit symbols. Diagrams should be on separate sheets, not inserted in the text.

Each idea submitted must be accompanied by a declaration to the effect that it is the original work of the undersigned, and that it has not been accepted for publication elsewhere.

TAPE/SLIDE SYNC

SHOWN are circuits for a Tape/Slide Synchroniser. First of all the commentary is recorded onto the tape in the normal way and whenever it is necessary for the slide projector to advance one slide, S1 is pressed which records a burst of ultrasonic sound onto the tape. When the recorder is played back (connected to the projector via the 'tuned amplifier' Fig. 2) the relay closes and advances the slide projector one slide every time an ultrasonic sound is detected.

The oscillator should be tuned to the right frequency simply by connecting the output to the input of the tuned amplifier, and adjusting VR1 until RLA1 closes. The circuit only amplifies frequencies of about 48.2 kHz.

The output of the 741 is amplified by TR1 and 2 which switches on the relay. The circuit works well despite the fact that the recorder's response is very low at 48.2 kHz. A small mono recorder works better than an expensive stereo tape deck because the latter usually have bias oscillators running at about 60-90 kHz which might beat with the synchroniser's output.

R.N. Johnson.
Coulsdon
Surrey

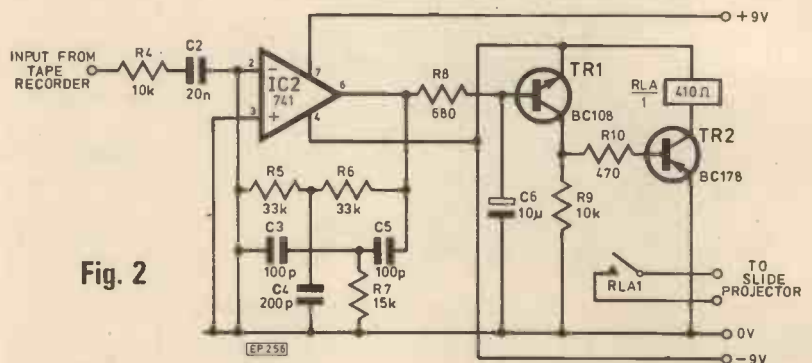
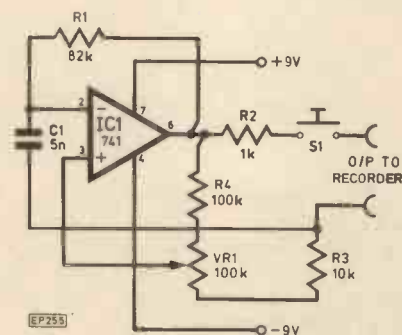


Fig. 2

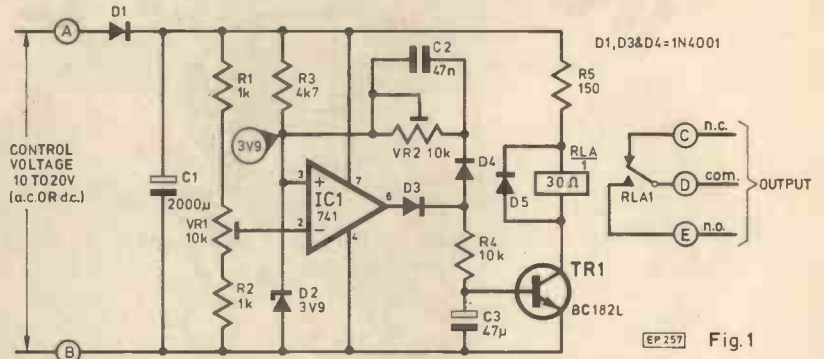
VOLTAGE SENSITIVE RELAY

The circuit diagram as shown right is an electronic relay with two wires on the input side and three wires at the output. The input leads are connected to the voltage to be sensed. Its range is from 10V a.c./d.c. to 20V a.c./d.c. The relay is a 6 volt/30 ohm coil type with changeover contacts rated for 5A/230V a.c.

The heart of the control system is a general purpose operational amplifier type 741. The control voltage is rectified by diode D1 and filtered and stored by capacitor C1. The non-inverting input (pin 3) is fed with a Zener regulated fixed potential of 3.9 volts, whereas the inverting input (pin 2) is variable and can be set by a preset potentiometer VR1 to any desired voltage. It is only by adjustment of this pot that a control point is achieved.

The output of the 741 changes state with input potential difference of as low as 1 millivolt. The positive going output at pin 6 biases a switching transistor through a diode and 10 kilohm resistor. The transistor in turn energises a relay whose normally open and normally closed contacts could be used in a number of applications. The differential at which the relay is desired to be operated is provided by positive feedback to non-inverting input, by a 10k preset pot VR2 in parallel with C2.

Any desired hysteresis can be set by VR2. An electrolytic is used at the base of switching transistor TR1 to avoid chattering of the relay on a.c. operation.



EP 257 Fig. 1

Automatic Temperature controller

This application is shown below. Here a slight modification in the control circuit of the VSR is needed. R1 is removed and a thermistor with long leads substituted. The thermistor is placed near the heating source. Its resistance will change with heat resulting in control initiation. The preset potentiometer VR1 now can be set for the desired temperature. The n.c. contacts of the relay can be used for opening the heater circuit when desired temperature has reached. In this case the input supply to the VSR is obtained by a low voltage transformer and regulated by a Zener.

Twilight switch

Alternative connections to the VSR (shown in outline) provide a light switch. Here R1 is replaced by an l.d.r. The resistance of this increases with a decrease in ambient light and thus initiates the

switching operation. The n.o. (normally open) contacts are connected in series with the load and supply. Thus the lamp lights up on fall of ambient light at preset value.

M. S. Dhingra,
Chittaranjan,
India.

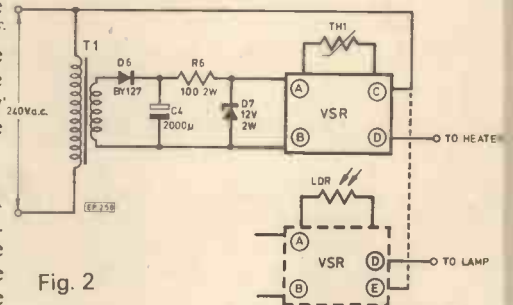


Fig. 2

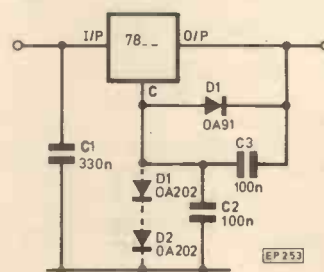
WHEN a precise voltage source is required for a particular application, it is common practice to modify the output voltage of a voltage regulator by "jacking-up" the common, i.e. raising its potential above ground. This useful method has its disadvantages. In the event of a short-circuit the common would become reverse biased and the regulator would lose its protection, thus resulting in its destruction.

In the circuit shown the germanium diode D1 prevents the common becoming more than 0.2 volt reverse biased under short circuit conditions and thereby protecting the regulator from damage.

The addition of C2 is to ensure stability. Diodes D2 and D3 (as an example) raise the output voltage by about 1.3 volts.

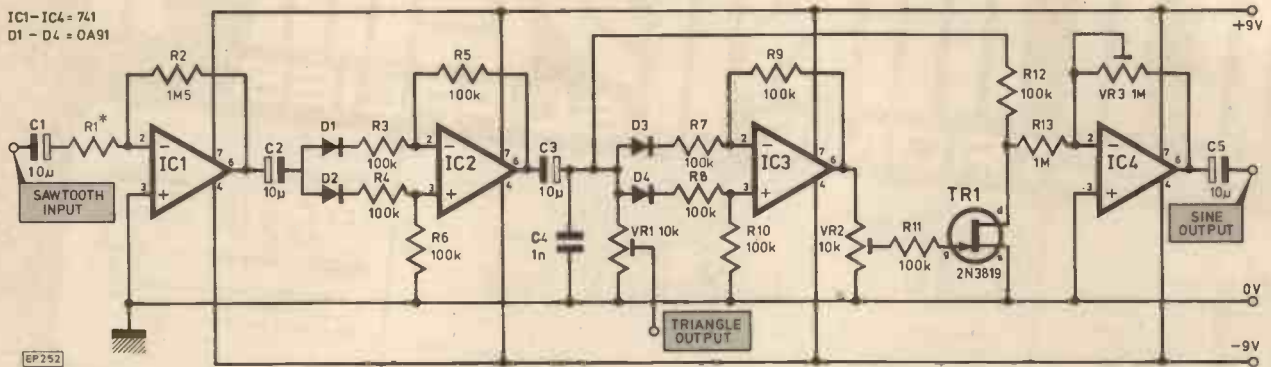
J.A. Barrow
Rugby
Warwickshire

'JACKED UP' REGULATOR



EP 253

WAVEFORM CONVERTOR FOR MINISONIC



THIS circuit will convert an input of sawtooth shape into a triangle shape and then convert the triangle waveform into a synthesised sine waveform. IC1 amplifies the input signal to a level of 15 V peak to peak. The value of R1 should be chosen according to the input level. R1 should in fact be chosen as (100 x peak-to-peak input voltage) kilohms. For Minisonics 1 and 2, R1 should be 39 kilohms, and 100 kilohms respectively. IC2 is a differential amplifier which inverts the positive going half cycle with respect to the negative going half cycle. With a certain amount of smoothing from C4 this produces an output of about 7.2 V peak-to-peak which is triangular in shape. VR1 should be adjusted to give a triangle output of 1 V peak-to-peak for Minisonic 2. Of course the level can be any value up to 7.2 V depending on what suits the constructor's application.

The triangle output is then fed into IC3 which is another differential amplifier and identical to IC2's circuitry. This produces an output of about 3.5 V peak-to-peak which is triangular in shape and twice the frequency of the input triangle waveform. This double frequency triangle wave controls the amount of attenuation in the voltage controlled attenuator built around TR1 and IC4.

As the input triangle wave approaches either its positive or negative peaks the gain of the output stage gradually decreases by the action of the double frequency wave at the gate of the f.e.t. TR1. This has the effect of rounding off the peaks. The shape produced is a very good approximation of a sine wave.

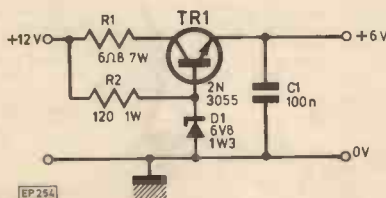
Setting up is quite simple. First the wiper of VR2 must be taken to its 0V end. VR3 is now adjusted to give an output level of around 500 mV r.m.s. VR2 can

now be adjusted until the best approximation of a sine wave is produced. An oscilloscope would be useful here but I have found that the adjustment can be made fairly accurately by ear. With the output connected to an audio amplifier VR2 should be adjusted until the sound contains fewest harmonics. Finally VR3 should be readjusted to give an output of about 350 mV r.m.s. (1 V peak-to-peak).

The circuit can be used as a triangle to sine convertor quite simply by leaving out IC2 and its associated circuitry.

P.G. Ludgate
High Wycombe,
Bucks.

CAR CASSETTE POWER SUPPLY



THIS unit enables a 6V cassette player or recorder to be powered from the 12V car battery, while drawing only 100mA in excess of the cassette player's consumption. Short-circuiting the output of the unit does not damage the circuitry, and only 2A flows, whereas the unit can provide 6V at up to 800mA.

R1 limits the short-circuit current to about 2A, and reduces the power dissipation of the series regulator transistor TR1, which acts as an emitter follower with the base voltage provided by R2/D1. (Note that D1 must be rated at 1W or greater.)

The complete unit fits easily into the smallest size of diecast case, which can also act as a heatsink for TR1.

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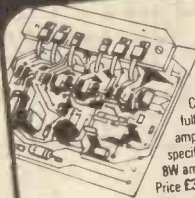
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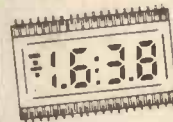


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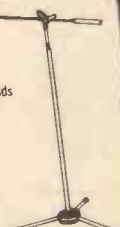


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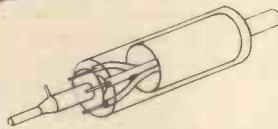
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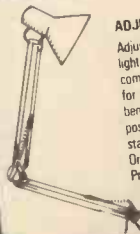
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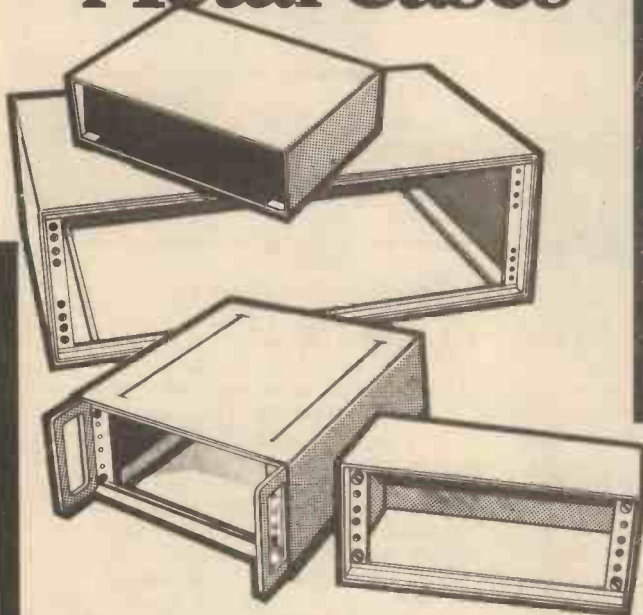
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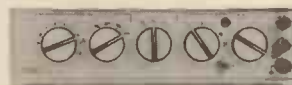
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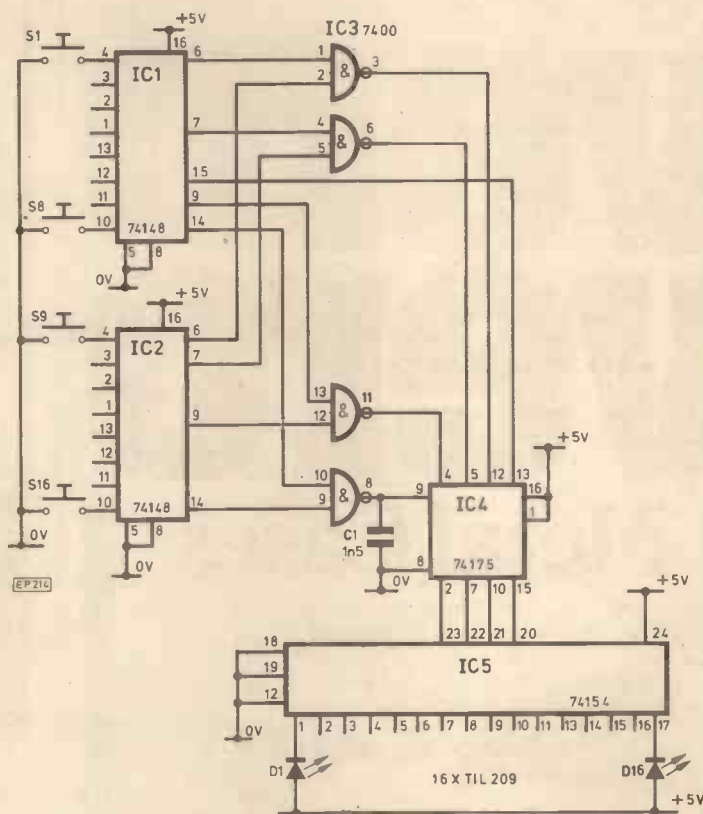
RHYTHM CODE GENERATOR

THE circuit shown was used to generate the 4 bit BCD rhythm select code required by the rhythm generator i.c. M252. It could also be used in any application that requires the generation of a 4 bit BCD word, for example, a hexadecimal keyboard.

If a pushbutton is pressed, the two SN74148 priority encoders code the 16 inputs to a 4 bit BCD word. This word is latched into the SN74175 quad latch by the combined GS outputs (pins 14) of the priority encoders. Capacitor C1 is used to show the edge of the latching signal so that latching takes place when the input data is valid.

The outputs of the quad latch can be decoded back to 16 bits by a SN74154 and the outputs used to drive i.e.d.s to give a visual indication of the button pressed, or as in this example the rhythm selected.

E. J. Weremiuk,
Brantham,
Essex.



AUTOMATIC CAR AERIAL CONTROL

IN February 1977, a circuit was published in "Ingenuity Unlimited" for the automatic control of an electric car aerial. The circuit described here is an alternative approach to the problem. The prototype has been in trouble-free use for over four years. It may be operated by the radio on/off switch so that the aerial is only raised when needed, or, if preferred, by the ignition switch. No additional switch is

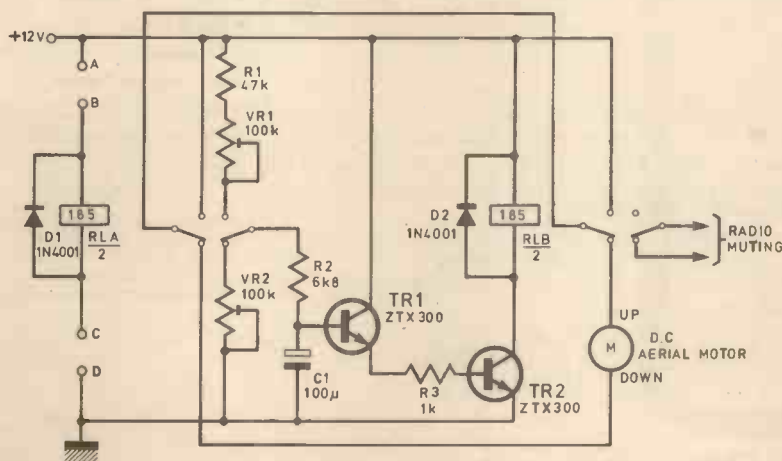
necessary. The device may be left permanently connected to the car battery, since the leakage current when the radio is switched off is very low.

The main control circuit may be used either by connecting terminal B to the radio on/off switch (or the ignition switch) and shorting C and D, or by shorting A and B and connecting C to an additional control circuit. The main circuit contains

two relays, each with two changeover contacts. When the radio is turned on (or the ignition if preferred) RLA switches on, the motor starts to wind up the aerial, C1 charges up and, after a delay (adjusted by the preset VR2), RLB switches on and stops the motor. When RLA switches off, the aerial starts to wind down, C1 discharges and, after a delay (adjusted independently by VR2), RLB 2 switches off and again stops the motor. VR1 and VR2 are adjusted so that the aerial winds completely up and down, the motor making about two revolutions against the slipping clutch for good measure.

If RLB has two changeover contacts, the spare set may be used to mute the radio whilst the aerial is winding up, to prevent crackling from the loudspeaker, e.g. by shorting the volume control. If the radio has a tape recorder socket through which the audio signal passes by means of a shorting plug, that can be used to make the connections to the muting contacts without disturbing the radio internally. A stereo radio would, of course, need a more complex system, or extra contacts in the relay.

D. A. Petty,
Aylesbury,
Bucks.



THERE are many published circuits which enable two lamps to be controlled along two wires giving provision for either lamp or both lamps to be illuminated at any one time.

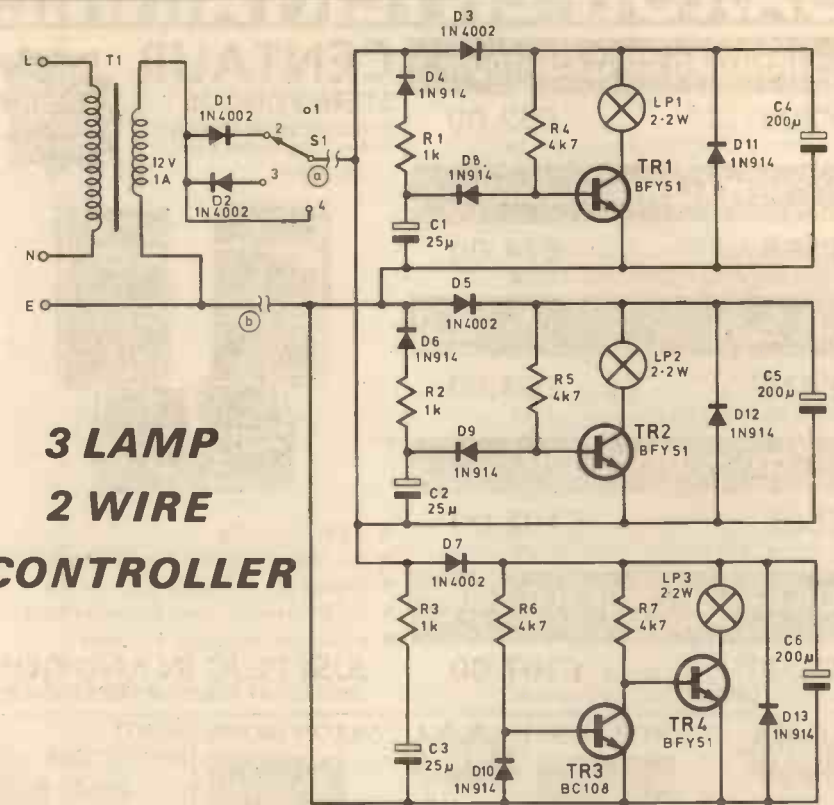
The circuit described below enables three lamps to be controlled along two wires enabling any one lamp to be illuminated at any one time. It was originally designed for a busy office to convey the instructions of 'Engaged', 'Wait' and 'Enter'.

The circuit is powered by 12 volts a.c. which is derived from the mains. The three lamps are controlled by S1 which gives the four possible operations. In position 1 no lamps are on, since no power is being applied to the wires a and b. In position 2, the alternating current from the transformer is half wave rectified by D1, making wire a positive with respect to b. TR1 and its associated components respond to this and so LP1 is illuminated. TR1 is prevented from conducting when S1 is in the a.c. position (position 4) by D4, R1, and D8. These ensure that C1 is charged by the negative half cycles so that TR1 will be made non-conducting.

In position 3 the alternating current is again half wave rectified, but this time wire a is negative with respect to wire b. The circuitry for LP2 is the same as that for LP1 except that it is connected the other way round to the wires a and b.

In position 4 of S1, an alternating voltage is applied to a and b. The circuitry for LP3 works in the following way; with S1 in position 2 (wire a positive) TR3 is conducting so keeping TR4 non-conducting. In position 3 (wire a negative), D7 is reversed biased so that LP3 is again

3 LAMP 2 WIRE CONTROLLER



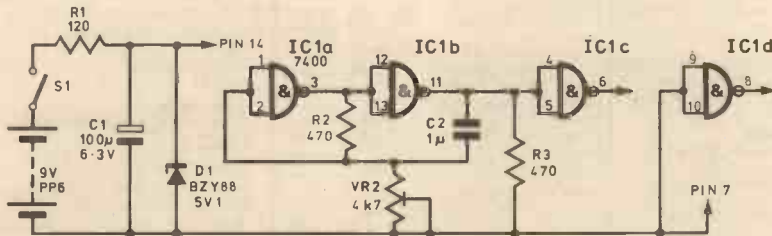
not illuminated. With an alternating voltage, the negative half cycles reduce the voltage across C3 to a sufficiently low level that TR3 is non-conducting. TR4 becomes conducting so allowing LP3 to light.

The 200µF electrolytic capacitors in

each lamp circuit offers a small amount of smoothing to the half rectified a.c., producing approximately 10 volts d.c. across each lamp when alight.

J. P. Kemp,
Kings Norton,
Birmingham.

SCOPE CALIBRATOR



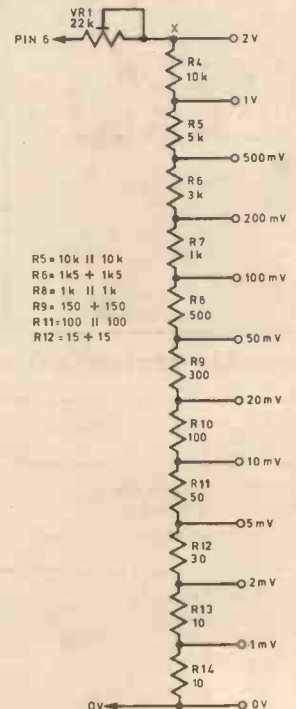
THIS simple circuit can be used to check oscilloscope Y-amplifier Volts/Cm calibration accuracy, and to set the frequency compensation trimmer R4 on a high input impedance oscilloscope probe.

IC1a and IC1b form a simple multivibrator type circuit, and IC1c acts as a pulse shaper. (The unused gate, IC1d, is connected so as to minimise battery current consumption.) The output is an excellent square-wave having fast rise and fall times with flat top and bottom base lines. Frequency is about 1kHz.

The output feeds the simple ladder attenuator (two per cent resistors should be used if high attenuation accuracy is required). VR1 should be set to give 2.0V at point X with respect to 0V.

D1 provides a stabilised 5V for the IC so that the ladder network receives a constant level square wave. Battery current drain is about 25mA.

A. Andrews,
Brighton.



EG 253

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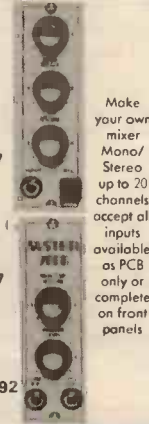
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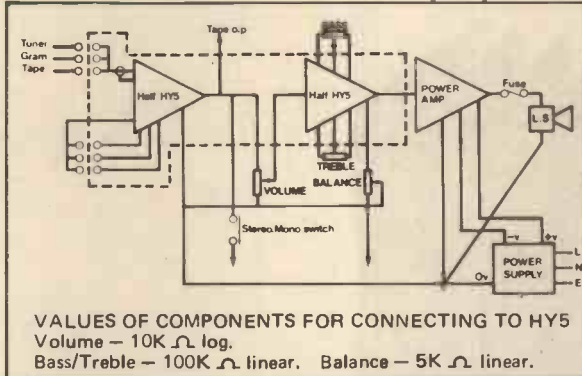
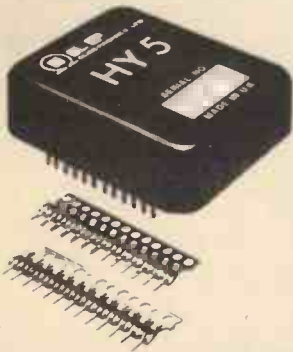
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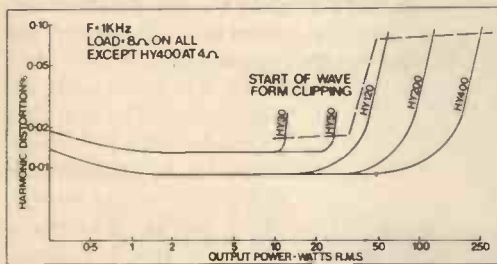


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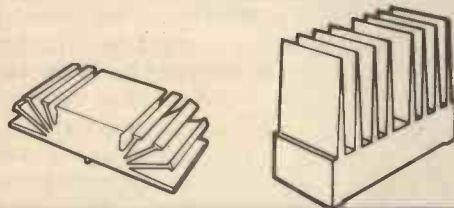
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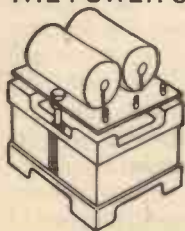


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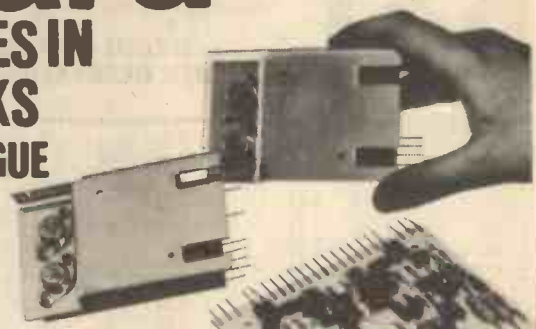
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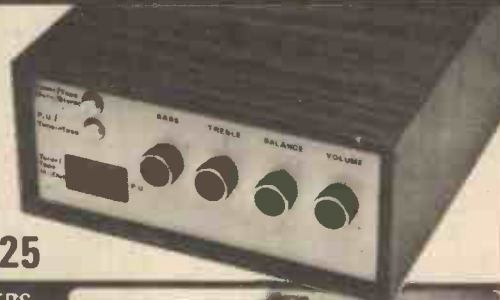
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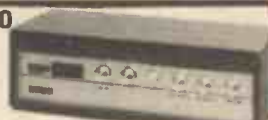
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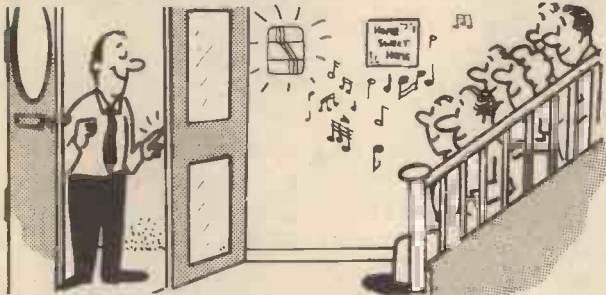
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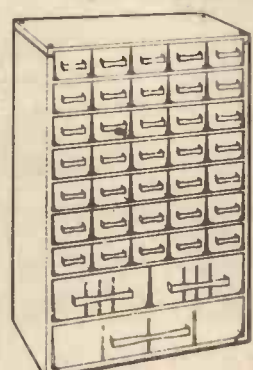
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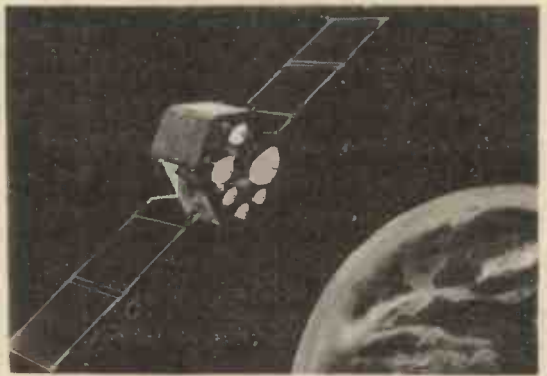
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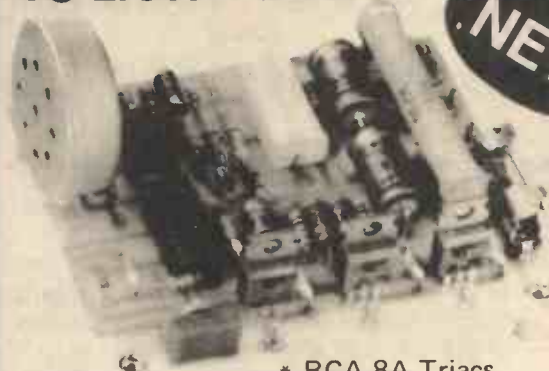


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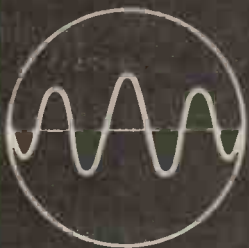


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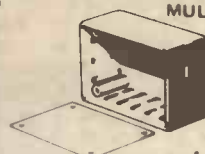
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BIM 4003 (85x56x28.5mm) £1.34
BIM 4004 (111x71x41.5mm) £1.84
BIM 4005 (161x96x52.5mm) £2.48



ALL METAL BIMCONSOLES

All aluminium, 2 piece desk consoles with either 15° or 30° sloping fronts, sit on 4 self-adhesive non-slip rubber feet. Ventilation slots in base and rear panel for excellent cooling. See latest catalogue for new styles and sizes

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BIM7151 (102x140x51[28]mm)	£11.36	BIM7301 (102x140x76[28]mm)	£11.36
BIM7152 (165x140x51[28]mm)	£12.28	BIM7302 (165x140x76[28]mm)	£12.28
BIM7153 (165x216x51[28]mm)	£13.43	BIM7303 (165x183x102[28]mm)	£13.43
BIM7154 (165x211x76[33]mm)	£14.83	BIM7304 (254x140x76[28]mm)	£14.83
BIM7155 (254x211x76[33]mm)	£16.36	BIM7305 (254x183x102[28]mm)	£16.36
BIM7156 (254x287x76[33]mm)	£17.71	BIM7306 (254x259x102[28]mm)	£17.71
BIM7157 (356x211x76[33]mm)	£18.83	BIM7307 (356x183x102[28]mm)	£18.83
BIM7158 (356x287x76[33]mm)	£19.92	BIM7308 (356x259x102[28]mm)	£19.92

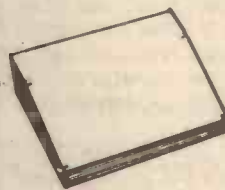


ABS & DIECAST BIMBOXES

6 sizes in ABS or Diecast Aluminium. ABS moulded in Orange, Blue, Black or Grey. Diecast Aluminium in Grey Hammetone or Natural. All boxes incorporate 1.8mm pcb guides, stand-off supports in base and have close fitting flanged lids held by screws into integral brass bushes (ABS) or tapped holes (Diecast).

	ABS		Diecast	Hammetone	Natural
(50x50x25mm)	N/A		BIM5001/11	£1.54	£1.23
(100x50x25mm)	BIM2002/12	£1.09	BIM5002/12	£1.66	£1.32
(112x62x31mm)	BIM2003/13	£1.27	BIM5003/13	£2.24	£1.70
(120x65x40mm)	BIM2004/14	£1.51	BIM5004/14	£2.81	£2.11
(150x80x50mm)	BIM2005/15	£1.72	BIM5005/15	£3.19	£2.72
(190x110x60mm)	BIM2006/16	£2.69	BIM5006/16	£4.94	£3.96

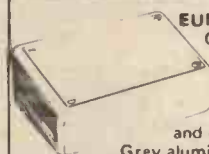
Also available in Grey Polystyrene with no slots and self-tapping screws
BIM 2007/17 (112x61x31mm) £1.06



LOW PROFILE BIMCONSOLES

Orange, Blue, Black or Grey ABS body has ventilation slots as well as 1.8mm pcb guides and stand-off bosses in base. Double angle recessed front panel with 4 fixing screws into integral brass bushes. 4 BIMFEET supplied.

BIM 6005 (143 x 105 x 55.5 [31.5]mm) £2.76
BIM 6006 (143 x 170 x 55.5 [31.5]mm) £3.58
BIM 6007 (214 x 170 x 82.0 [31.5]mm) £4.83



EUROCARD BIMCONSOLES

Orange, Blue, Black or Grey ABS body accepts full or 1/2 size Eurocards, with bosses in the base for direct fixing. 1.8mm wide pcb guides incorporated and 4 BIMFEET supplied. 1mm Grey aluminium lid sits flush with body top and held by 4 screws into integral brass bushes.

BIM 8005 (169x127x70[45]mm) £4.71
BIM 8007 (243x187x103[66]mm) £6.70

BIMTOOLS + BIMACCESSORIES



MAINS BIMDRILLS

Small, powerful 240V hand drill complete with 2 metres of cable and 2 pin DIN plug. Accepts all tools with 1mm, 2mm or .125" dia. shanks. Drills brass, steel, aluminium and pcb's. Under 250g, off load speed 7500 rpm. Orange ABS, high impact, fully insulated body with integral on/off switch £11.21

Mains Accessory Kit 1 includes 1mm, 2mm, .125" twist drills, 5 burrs and 2.4mm collet £2.64

Mains Kit 2 includes Mains BIMDRILL as above, 20 assorted drills, mops, burrs, grinding wheels and mounted points, 1mm, 2mm, 2.4mm and .125" collets. Complete in transparent case measuring 230x130x58mm £23.57

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Allows pcb's to be flat mounted sandwich fashion in BIMBOXES, BIMCONSOLES, and all other enclosures having 1.5mm wide vertical guide slots. One plastic BIMDAPTOR on each corner of pcb(s) enables assembly to be simply slid into place. 54mm long, 10 slots on 5mm spacing and can be simply snapped off to length. £1.15 per pack of 25.



BIMFEET

11mm dia. 3mm high, grey rubber self-adhesive enclosure feet. £0.81 per pack of 24.

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Accessory Kits 1 have appropriate drills and collets as above plus 20 assorted tools. Mini Kit 1 - £16.10, Major Kit 1 - £20.70. Accessory Kits 2 have appropriate drills, collets plus 40 tools and mains-12V dc adaptor. Mini Kit 2 - £36.22, Major Kit 2 - £41.97. Accessory Kits 3 as appropriate Kits 2 plus stand/lathe unit. Mini Kit 3 - £48.30, Major Kit 3 - £54.05.

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BIMPUMP Minor (150mm long) £7.24

BIMIRONS



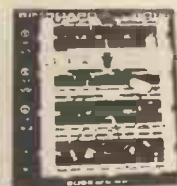
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27 watt iron with long life, rapid change element, screw on tip, stainless steel shaft and clip on hook. Styled handle with neon. £4.37

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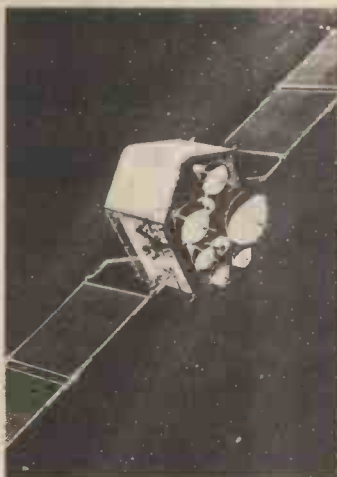
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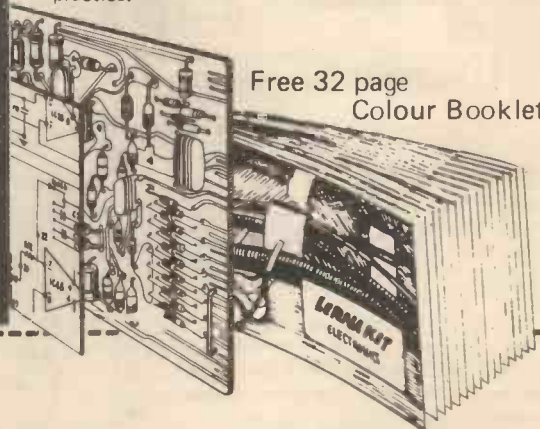
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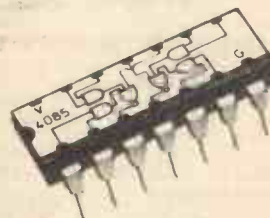
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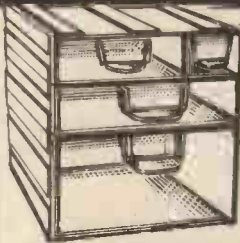
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