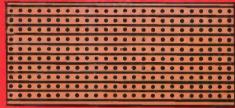
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Mains operated regulater

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110 SWR/Power/FS-10/100W

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1KW max.

(40 MHZ)

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time for callers

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8 3 3

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£45.94 TM352 Hand held, DC 10A, Hile test, Continuity test €57 44

TM353 Bench, 2A AC/DC, 1000V AC/DC, 20M ohm. Typical 0.25% TM351 Bench, 10A AC/DC, 1000V AC/DC, 20M ohm

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POLYESTER CAPACITORS: Axial lead type (Values are in nF) 400V: inF 1n5, 2n2, 3n3, 4n7, 4n8 tip: 10n, 15n, 13n, 22n 12p; 33n, 4n, 8n 15p; 100n, 150n 20p; 220n 30p; 330n 42p; 470n 32p; 680n 40p; 1µF 45p; 2u2 2p; 4u7 45p. 180V: 10nF, 12n, 10on 1tp; 150n, 220n 17p; 330n, 470n 30p; 680n 35p; 1µF 42p; 1µ5 45p; 2u2

POLYESTER RADIAL LEAD CAPACITORS (250V)
10nF, 15n, 22n, 27n 6p; 33n, 47n, 68n, 100n 7p; 150n, 220n 10p; 330n,
470n 17p; 680n 19p; 1µF 23p; 1µ5 40p; 2µ2 48p.

TIL211 Grn TIL212 Yellow

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2N5777

Square LED

2N5777 45
7 Seg Displays
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Til.32 inf. Red 52
Til.78 detector 54
Bargraph Red.
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29

85 45

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7491 7492 7493

74107

74110

74121

74122

TTL 74 (TEXAS)

7400 7401

7402

7403

7406 7407

TANTALUM Bead Capacitors 33 V: 0·1µ, 0·22, 0·33 15p; 0·47, 0·68, 1·0, 1·5 15p; 2·2, 3·3 15p; 0·47, 4·7, 6·8 22p; 10·28p. 16V; 2·2, 3·3 16p; 4·7, 6·8, 10·18p; 15·36p; 2·2 30p; 33, 47·40p; 100. 75p. 16V; 15, 22.26p; 33, 47·35p; 100. 55p.

MYLAR FILM CAPACITORS 100V: 1nF, 2n, 4n, 4n7, 10n 6p; 15nF, 22n, 30n, 40n, 47n 7p; 56n, 100n, 200n 9p. 470n/50V 12p.

MINIATURE TYPE TRIMMERS 4-8pF, 2-10pF 22p; 2-25pF, 5-85pF 30p; 10-88pF 35p.

COMPRESSION TRIMMERS 3-40pF, 10-80oF 20p; 20-250oF 28p; 100-580pF 39p; 400-1250pF 48p.

POLYSTYRENE CAPACITORS 10pF to 1nF 8p; 1-8nF to 12nF 10p.

SILVER MICA: 2pF, 3·3, 4·7, 6·8, 8·2, 10, 12, 15, 18, 22, 27, 33, 39, 47, 50, 56, 68, 75, 82, 85, 100, 120, 150, 180 15p. 200, 220, 250, 270, 300, 330, 360, 390, 470, 60), 800, 820 21p. 1000, 1200, 1200, 1300, 2000 30p. 3300, 4700 80p.

CERAMIC CAPACITORS: 50V 0-5pF to 10nF 4p; 22n to 100n 7p. EURO BREADBOARD £5-20.

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1A TO3 +ve -ve
5V 7805 145p 7905 2
12V 7812 145p 7912 2
15V 7815 145p 7915 2
18V 7818 145p -

7818 145p
TO220 Plastic Casing
7805 50p 7905
7812 50p 7912
7815 50p 7915
7818 50p 7918
7824 50p 7924

1A 5V 12V 15V 18V

1A 5V 12V 15V

18V 24V

POTENTIOMETERS: (ROTARY) Carbon Track, 0-25W Log & 0-5W Linear Value. Soo  $\Omega$ , It & 24K (Lin. only) Single 29p 5K-2 M  $\Omega$  single gang 28p 6K-2 M  $\Omega$  single with DP switch 78p 5K-2 M  $\Omega$  double gang 88p OPTO
ELECTRONICS
LEDe plus clips
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SLIDER POTENTIOMETER 0.25W log and linear values  $60 \mathrm{mm}$  8K  $\Omega$ -500K  $\Omega$  single gang 70p 10K  $\Omega$ -500K  $\Omega$  dual gang 110p Self Stick Graduated Bezele 38p

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Vertical & Horizontal
0-1W 50 Ω—51/Ω Miniature
0-25W 100 Ω—3 -3M Ω Horiz
10-25W 200 Ω—4 -7MΩ Vert
160

RESISTORS: Carbon Film, High Stability, Low Noise, Miniature Tolerance 5%. Range Val. 1-99 100+3W 102-4M7 E12 2p 1p 1W 102-4M7 E12 2p 1p 1W 102-10M E12 5p 4p 2% Metal Film 110-1M Ep 4p 1% Metal Film 510-1M Ep 4p 100+ price applies to Resistors of each value not mixed.

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220p 220p 220p

55p 55p 55p 55p 55p

SLIDE 250V: 1A DPDT 14p 1A DP c/off. 16p 1A DPDT 13p

PUSH BUTTON Latching or Momentary. SPST C/Over 99p DPDT C/Over 145p

SWITCHES
TOGGLE 2A 250V
SPST 44p
DPDT 44p
SUB-MIN
TOGGLE
SP changeover 58p
SPST on/off 84p
DPDT 6 tags 75p
DPDT 6/off 68p
DPDT Blased 149p SWITCHES Ministure Non-Locking
Push to Make 15p Push to Break 20p
ROCKER: SPST on/off 10A 250V 28p

ROCKER: Illuminated DPST
Lights when on: 10A 240V
ROTARY: (ADJUSTABLE STOP) 1 pole2-12 way 2p/2-6W, 3p/2-4W, 4p/2-3W.
ROTARY: Mains 250V AC, 4 Amp
86p

DIL SOCKETS (Low Profile - Texas) 8 pin 8p; 14 pin 10p; 16 pin 10p; 18 pin 18p; 20 pin 22p; 24 pin 25p; 28 pin 28p; 40 pin 30p.

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100/300pF
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61 Ball Drive
61 1 Hall Drive
4103
Dilla Drive
4103
E 1/361
T75p
01-365pF
02 385pF
00 2 385pF
00 3 3 85pF
00 3 85pF
00 3 3 85pF

DENCO COILS RDT2 145p
'OP' VALVETYPE RFC 5
Range 1 to 5 Bl., RFC 7 (19mH) Bloop
Rd., Yl. Whit.122p IFT 13; 14; 15;
6-T B.Y.R. 110p 15; 17 12p
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21 × 33" 73p 52p clad boards clad boards 21 × 35" 73p 52p Fibreals 73p 52p 31 × 35" 83p 6 × 36" 83p 79p 5.R.B.P. 31 × 17" 326p 211p 41 × 17" 426p 7kt. of 100 plns 50p Spot Face Cutter 118p Pln insertion Tool Pin insertion Tool 162p Dalo Pen 90p

Range 2V7 to 39V 400mW 8p each Range 3V3 to 33V, 1-3W 15p each AA129 22 BA100 15 BY127 12 CRO33 250 OA9 40 OA70 12 OA90 8 OA90 9 O Olode 195 BRIDGE RECTIFIERS (plastic case) 1A/50V 20 1A/100V 22 1A/200V

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1A/400V 1 A /800 V 2A/50V 2A/200V 40 IN4148 4 IS44 9 3A/100V 18 3A/400V 16 3A/500V 17 3A/1000V 30 6A/400V50p 2A/400V 2A/600V 48 6A/100V 6A/400V 95 6A/400V 95 10A/200V 215 10A/600V 315 25A/200V 215 25A/600V 395 BY184 56 VM18 DIL 55 We stock a wide selection of Electronic Books and Majazines

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4561 4518 4519 4520 80 38 36 45 74141 74142 74144 74145 74147 74148 74150 74151 74155 74156 74157 74159 74160 74161 74163 74163 74163 74163 74163 LS91 LS92 LS93 LS95 LS96 250 70 99 75 80 45 45 75 75 75 LM3909N LM3911 LM3914 LM3915 LINEAR IC's LS107 LS109 LS112 709 C 8 pin 7107 709 C 8 pin 7107 709 C 8 pin 7107 733 8 pin 713 8 pin 717 C 14 pin 718 C 8 pin 718 C LM3916 LM13600 48 LM1-3600

14 M253AA

14 M253AA

14 M253AA

16 M253AA

18 MC13047

18 MC1303

38 MC13047

199 MC1489

190 MC1489

191 MC1489

192 MC1896

193 MC1896

193 MC1896

194 MC1896

195 MC1896

196 MC1896

197 MC3302

198 MC3302

198 MC3304

199 MC3403

190 MC3403

191 MC544

191 MC503

191 M 15118 15122 LS132 LS136 LS136 LS138 LS139 LS151 62 64 64 65 185 168 290 85 72 72 55 75 88 85 140 LS153 LS155 LS156 LS157 74:66 74:167 74:167 74:170 74:173 74:174 74:175 74:176 74:177 74:177 74:179 74:181 74:182 74:183 74:184 74:185 74:189 74:191 74:192 74:193 74:194 74:195 74:197 LS158 LS160 LS161 LS162 LS163 LS163 LS164 LS165 LS166 LS170 LS173 LS174 LS175 LS181 4039 4040 4041 4042 4043 4044 4046 4047 4048 4049 4050 75 99 99 290 70 70 70 85 75 65 65 LS191 LS192 LS193 LS194 LS195 LS196 LS197 LS200 LS221 LS243 LS243 LS244 LS245 LS251 LS253 LS257 LS258 LS259 LS266 LS273 LS2579 LS268 275 365 71 214 195 213 65 190 215 375 150 98 48 SAB3209 SAB3210 SG3402 SN76013 SN76023N SN76477 SP8629 TAA621 399 22 26 26 20 20 20 20 20 20 26 21 65 70 146 43 188 99 99 725 695 695 770 TAD100
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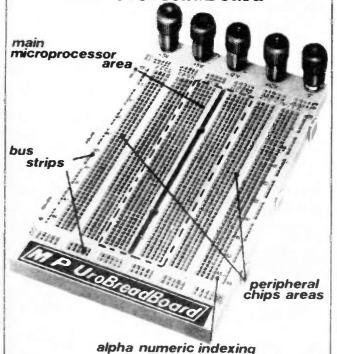
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BC212 BC2121 BC2121 BC2131 BC2131 BC214 BC214 BC235 BC237 BC307B BC337 BC307B BC337 BC307B BC337 BC307B BC411 BC461 BC461 BC477 BC516/18 BC557/8 BC557/8 BC557/8 BD133 BC557/8 BD133 BC38 BC338 BC338 BC338 BC338 BC338 BC338 BC338 BC337 BC41 BC41 BC41 BC41 BC41 BC57/8 BD133 BC57/9/11 BD133 BD134 BD135 BD137 ZTX301/2
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BFX85/86 28
BFX85/8 28
BFX85/8 28
BFY85/6 23
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BFY84 35
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#### LINEAR

★741 15p	LM324 40p	<b>★</b> RC4136
CA3080 70p	LM380 80p	70p
★CA3130	LM381 120p	SN76477
75p	LM382 120p	150p
★CA3140	LM387 120p	★TL081 30p
45p	LM1458 40p	★TL084 90p
*ICM7555	LM3900 50p	XR2206
q08	★LM3914	300p
★ ICM7555	200p	ZN414 100p
80p	★LM3915	ZN425E
LF351 45p	200 p	350p
LF356 90p	★NE55580p	TBA800
LM301 A25p	NE556 55p	80p

#### CONNECTORS

2·5mm 10p 10p 3·5mm 10p 9p Stand'd 12p 20p Stereo 24p 25p	2 pin 9p 9p 3 pin 12p 10p 5 pin 180 13p 11p	
LEDS	VEROBOARO	
★3mm red 7p	Size 0 1in Mahix	
#3mm green 12p 3mm yellow 14p	2·5 × 1" 22	р
3mm yellow 14p ★5mm red 8p	2.5 × 3.75" 75	р
★5mm green 12p	2·5 × 5" 85	р
5mm yellow 14p	3 · 75 × 5" 95	P
10 3 or 5mm 30p	Veropins per 100	
LED clips	Single sided 50	р
FND500 70p	Double sided 60	p

#### 70p

I WALL COLD I A	11.0	
AC128 25p AC176 25p AD161 40p AD162 40p BC107 10p ★BC108 8p ★BC108 C 10p BC109 10p	★BC182L8p ★BC184L7p BC212L 10p ★BC214L8p BC547 10p BD131 35p BD132 35p BD139 35p	BD140 35p ZTX300 14p 2N2646 45p 2N3053 23p 2N3055 50p \$2N3702 6p \$2N3704 6p \$2N3905 6p

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Antex 25W Soldering Iron		10p
10m pack 22swg solder	6	0p
5 PP3 battery connectors	3	Op
Min push to make switch	1	2p
Min push to break switch	1	5p
64mm 64 ohm speaker	7	00
20mm panel fuseholder	. 2	5p
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MN25, 50 asstd, pop rivets.
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MN33, 20 coil formers, ceramic, plastic, reed relay etc.
MN36, 10 asstd, switches, toggie, slide, micro, etc.

micro, etc. MN37, 10 assid. audio connectors. Din

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Can be hidden behind a panel, door, wallpaper, etc.
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magnetisable, the first rede closes a relay, the secondary
contacts of which will light the lamp or whatever
device is secretly controlled and it would also latch itself
on. The second reed will unlatch the relay. Complete

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#### INSTRUMENT BOX WITH KEY

NOTIFICIAL DOA WITH REY
Very strongly made (ally-wood sides with hard-board top and
bottom). This is black grained effect, viny! covered, very pleasing
appearance, Internal dimensions 12." inong, 4"x" wiee, 6" deep.
Ideal for carrying your multi range meter and small tools and for
keeping them in a safe place. £2.30. Post paid if ordered with
other goods, otherwise £1.00.

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COMPUTER RET SWITLES [make your own keyboard]
These are for making up on a p.c.b. and consist of a vertical mount
ing computer type reed switch, which makes circuit when a magnet
passes over it. The magnet is located in
the plastic plunger which in turn is
depressed by a push rod, to
which the legended top is
fixed. These are made up in banks of 6, price £2.30 per bank of 6

OUR CAR STARTER AND CHARGER KIT has no doubt saved many motorists from embartassment in an emergency you can start car off mains or bring your battery up to full charge in a couple of hours. The kit comprises: 250w mains transformer, two 10 amp bridge rectifiers, start/charge switch and full Instructions. You can assemble this in the evening, box it up or leave it on the shelf in the garage, whichever suits you best. Price £11,50 + £2,50 post.

(including to

GPO HIGH GAIN AMP/SIGNAL TRACER. In case measuring only 5½in x 3½in x 1½in is an extremely high gain (70dB) solid state amplifier designed for use as a signal tracer on GPO cables, etc With a radio it functions very well as a signal tracer. By connecting a simple coil to the input socket a useful mains eable tracer can be made. Buns on standard 4½ w battery and has input, output sockets and on-off volume control, mounted flush on the top. Many other uses include general burchose amp cueins and etc. An absolute uses include general purpose amp, queing amp, etc. An absolute bargain at only £1.85. Sultable 800hm earpiece 69p.

MINI MONO AMP on p.c.b., slze 4"x 2 approx. Fitted volume control and a hole for a tone control should you require it. The amplifier has three

transistors and we estimate the output to be 3W rms More technical data will be included with the amplifier Brand new, perfect condition, offered at the very low price of £1.15 each, or 10 for £10.00.



economical. We offer an inverter for 21" 13 watt cent tube. £3,45



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#### TANGENTIAL BLOW HEATER

2.5 Kw quiet, efficient instant heating from 230/240 volt ains Kit consists strated, 2.5 Kw

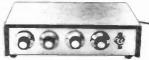


#### MOTORISED DISCO SWITCH

With 10 amp changeover switches, Multi-adjustable switches all rated at 10 amps, this would provide a magnificent display. For mains operated 8 switch model 10 switch model £6.75, 12 switch

#### 3 CHANNEL SOUND TO LIGHT KIT

Complete kit of parts for a three-channel sound to light unit controll ing over 2000 watts of light ing. Use this



at home If you wish but it is plenty rugged enough for disco work. The unit is housed in an attractive two-tone metal case and has controls for each channel, and a master on/off. The audio input and output are by %" sockets and three panel mounting fuse holders provide thy ristor.

A four-rise in plan and socket facilitate ease of connects. ng lamps. Special snip price is £14.95 in kit form or £19.95 assembled and tested

#### THIS MONTH'S SNIP

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Complete in module form with electronics Including |
Printer Synchro Signal Amplifier & Printer Reset |
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#### a for μA

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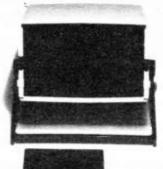
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68	3	1.5	3 - 46	1 - 43	125	6	3	11 - 78	1.90	69	250	7-54	1 - 43
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We congratulate the original inventors and the company managers whose bold step turned out to be no gamble.

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With newcomers to electronics particularly in mind, we include this month a fine selection of Ten Popular Designs, any one of which can be accommodated comfortably on the free piece of Veroboard. As anniversaries are in the air, we chose one design from each complete year of EVERYDAY ELECTRONICS. That makes us just ten, come next month.

FRED BENNETT

Our November Issue will be published on Friday, October 16. See page 683 far detalls.



Readers' Enquiries

We cannot undertake to answer readers' letters requesting modifications, designs or Information on commercial equipment or subjects not published by us. All letters requiring a personal reply should be accompanied by a stamped self-addressed envelope.

We cannot undertake to engage in discussions on the telephone.

Component Supplies

Readers should note that we do not supply electronic components for building the projects featured in EVERYDAY ELECTRONICS, but these requirements can be met by our advertisers.

All reasonable precautions are taken to ensure that the advice and data given to readers are reliable. We cannot however guarantee it, and we cannot accept legal responsibility for it. Prices quoted are those current as we go to press.

# ELECTRONICS

CONSTRUCTIONAL PROJECTS 654 SUSTAIN UNIT For electric guitar by R. A. Penfold 664 EE MINILAB Breadboarding unit with built in electronics by O. N. Bishop TEN POPULAR DESIGNS 671 Snap Indicator......672 Tape Noise Limiter......674 Heads and Tails Game ......675 Photo Flash Slave ......677 Continuity Tester ......676 Fuzz Box ......678 Opto Alarm ......679 Soil Moisture Unit ............680 Ice Alarm ......681 CAPACITANCE METER A desirable test instrument by J. R. W. Barnes 690 GENERAL FEATURES 652 EDITORIAL Come and Join Us; A Sure Foundation; Ten Popular Designs TEACH-IN '82 Part 1: Conduction in Materials and Devices by O. N. Bishop 658 663 COUNTER INTELLIGENCE A retailer comments by Paul Young 670 SHOP TALK Product news and component buying by Dave Barringfon EVERYDAY NEWS What's happening in the world of electronics 684 685 FOR YOUR ENTERTAINMENT Pacemaker, Talking Memories, Chatter Box by Barry Fox RADIO WORLD Armchair Travellers, Television News by Pat Hawker G3VA 686

INTRODUCTION TO LOGIC Part 6: AND, OR, NAND, Symbols and Truth Tables by J. Crowther

Back Issues

VOL. 10 NO. 10

Certain back Issues\* of EVERYDAY ELECTRONICS are available worldwide price 80p inclusive of postage and packing per copy. Enquiries with remittance should be sent to Post Sales Department, IPC Magazines Ltd., Lavington House, 25 Lavington Street, London SE1 0PF. In the event of non-availability remittances will be returned.

\* Not available: October 1978 to May 1979.

JACK PLUG AND FAMILY Cartoon by Doug Baker

SQUARE ONE Beginners Page: Fixed value resistors

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OCTOBER 1981

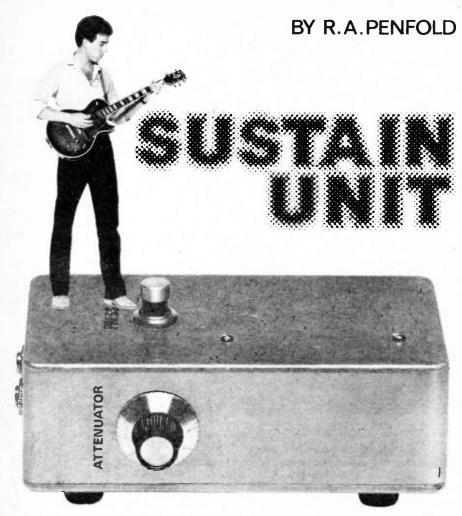
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S ustain units are primarily intended for use with electric guitars, and as the name implies, devices of this type extend the maximum duration of the note from the instrument.

The output signal from a guitar normally has a high initial ampltude which quickly drops to a much lower level, and then steadily dies away to virtually nothing over a few seconds. A sustain unit boosts the signal as it dies away, giving an almost constant output amplitude and effectively increasing the maximum time for which a note can be sustained.

#### METHODS OF OPERATION

There are two reasonably simple ways of obtaining a sustain effect, the most simple one being to use a clipping circuit of the type outlined in the block diagram of Fig. 1(a). Here the input signal is first considerably amplified and then fed to a clipping circuit. The latter normally uses a couple of silicon diodes to clip the signal at about  $\pm 0.6$ V.

Thus the output amplitude is set at this level by the clipping action and is largely independent of the input level. In fact, however large the input level may be, the output will not rise above the clipping level.

It can fall below the clipping level if the input signal is inadequate, but the input amplifier is given a high enough gain to ensure that this does not happen until a few seconds after the commencement of a note.

#### DRAWBACKS

One major drawback to this system is that by clipping the signal it is considerably distorted. Strong harmonics of the fundamental note are produced, giving the same effect as a "Fuzz" unit. A low pass filter is therefore normally included at the output of the circuit in order to attenuate the harmonics that would otherwise give an undesired effect.

This does not completely solve the problem though, since strong intermodulation distortion is caused by the clipping, and this produces a very unpleasant sounding output if more than one note at a time is played. Most of the intermodulation distortion products will not be attenuated by the low filter as they are mainly in the same frequency range as the desired signals.

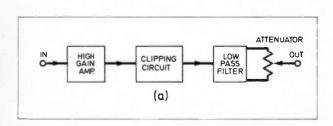
#### SECOND METHOD

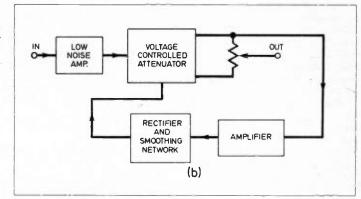
The second method, which is shown in the block diagram of Fig. 1(b), is slightly more complicated but is capable of much improved results. As before, the signal is first fed to an amplifier. It is then fed to a voltage controlled attenuator (v.c.a.), which is a circuit that reduces the amplitude of the input signal by an amount which depends upon the voltage fed to its control terminal.

In this type of circuit the v.c.a. gives little attenuation with only a small control voltage, and increasing attenuation as the control voltage is raised.

The control voltage is derived from the output of the v.c.a. by way of a rectifier and smoothing network, and a certain amount of additional amplification is also needed in order to obtain a sufficiently large control voltage.

Fig. 1 (below). Simple methods for obtaining sustain effect. (a) clipping circuit; (b) voltage controlled attenuator or compressor type of circuit.





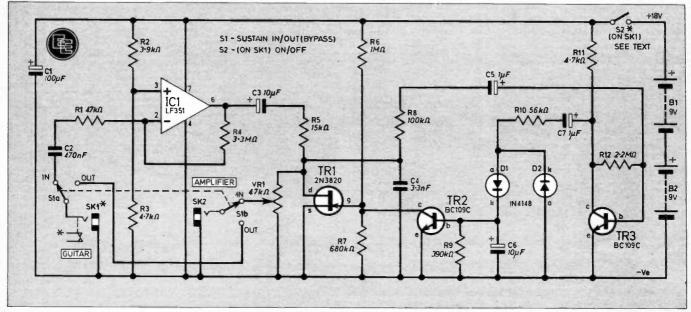


Fig. 2. The complete circuit diagram of the Sustain Unit.

Under quiescent condition or with only a very low input level there will be at most only a very small control voltage, and a greatly amplified signal appears at the output. At higher input levels a significant control voltage is produced, and the signal receives a lower level of amplification in consequence.

At still higher input signal amplitudes the control voltage is increased further, and little or no amplification is provided by the unit. This has the effect of holding the output signal at a certain level, provided the input is sufficiently strong to produce this output level. Raising the input above this minimum level causes the gain of the circuit to drop and results in very little increase in the output level.

This gives the required sustain effect because as the input level decays, the gain of the circuit increases so as to maintain the output level.

#### ATTACK AND DECAY TIMES

The attack and decay times of the smoothing circuit must be carefully chosen so that the unit will respond with suitable rapidity to changes in input amplitude, but not so rapidly that the waveform of the processed signal is altered and significant distortion is generated.

The unit will not properly respond to the transient at the beginning of each note as it is too brief, but this does not matter and is really of benefit as it maintains the guitars natural attack, which is lost when the clipping method is employed.

With either system a variable attenuator is included at the output

so that the signal level from the unit can be adjusted to one which is comparable to the output from a guitar.

#### THE CIRCUIT

Fig. 2 shows the complete circuit diagram for the unit finally evolved, and this is of the second type, which is usually called a "compressor".

The input amplifier is based on operational amplifier IC1. R2 and R3 bias the non-inverting input to slightly more than half the supply voltage, and R4 biases the inverting input and output of the device to the same potential due to a negative feedback action. This bias voltage gives the circuit optimum unclipped peak to peak output voltage swing capability.

The voltage gain of the amplifier is determined by the ratio of R1 to R4, and is approximately equal to R4 divided by R1, or about 70 in other words. This gives sufficient sensitivity to produce a good sustain effect with any normal instrument.

It is essential for the input amplifier to have a low noise level as the circuit will produce maximum output from input levels below ImV. A poor noise performance would result in low input levels being swamped in noise. IC1 provides a suitably high performance in this respect as it is a bifet type which has a low noise jfet input stage. The distortion performance is also very good, and these devices are ideal for critical audio applications.

## VOLTAGE CONTROLLED ATTENUATOR

The v.c.a. is based on p-channel jfet TR1, and it is its drain to source resistance plus R5 that form the two

elements of the attenuator. C3 couples the output from IC1 to the attenuator.

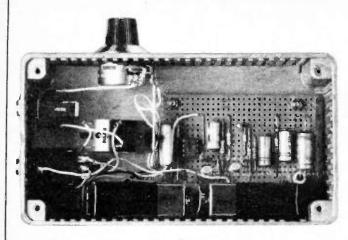
Under quiescent conditions R6 and R7 strongly reverse bias TR1 so that it is fully switched off and exhibits a drain to source resistance of typically about 1,000 megohms. It therefore causes no significant losses through R5, although output attenuator VR1 and other loads on the output do cause losses of a few dB. However, the circuit still exhibits a voltage gain of more than 50.

Some of the output signal is coupled by C5 to a common emitter amplifier which uses TR3 in a conventional configuration. The gain of the amplifier is far higher than is necessary, and so R8 is inserted in the signal path to reduce the gain to a more appropriate level. This resistor also boosts the input impedance of the stage and thus helps to reduce loading on the output.

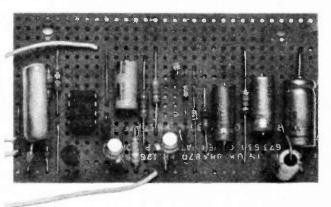
The output from TR3 is fed by C7 and R10 to the smoothing and rectifier circuitry which is comprised of D1, D2. C6 and R9.

If the input signal is sufficiently strong, the positive bias produced by the rectifier and smoothing circuit will be high enough to switch on TR2 to some extent. TR2 then reduces the bias on TR1, causing it to conduct more heavily and increasing the losses through R5. The stronger the input signal is made, the lower the bias voltage fed to TR1 and the greater the losses through R5. This gives the required stabilisation of the output signal amplitude.

C4 rolls-off the high frequency response of the circuit (above about 4 to 5kHz) under quiescent or low signal level conditions. This helps to give an improved signal-to-noise ratio







Completed circuit board removed from case.

under the conditions where any noise will be most noticeable.

C4 has little effect at medium and high input levels where the resistance of TR1 is quite low, and the shunting effect of C4 even at the highest audio frequencies is of far less significance.

#### BYPASS SWITCH

S1 can be used to bypass the circuit so that the sustain effect can be switched out when it is not required. S1 is a heavy duty push button type switch having a sequential action, so that it can be operated by foot.

S2 is the on/off switch, and on the prototype it is actually a make-contact on the input socket (SK1) so that the unit is automatically switched on when the guitar is plugged into it. This is quite common practice with effects units.

An 18 volt supply is needed in order to enable the unit to handle high input signal levels without clipping occuring, and this is provided by two small 9 volt batteries (PP3 size) connected in series. The unit only consumes about 5mA.

#### PERFORMANCE

The graph of Fig. 3 shows input voltage versus output voltage for the prototype. As can be seen from this, the compression commences at less than 1mV, and increasing the input to as much as 50mV or so causes only a marginal increase (about 15 per cent) in the output. At higher input levels TR1 saturates and the gain of the unit stays at approximately unity.

As the output from a guitar is only likely to reach a high enough level to saturate TR1 during the initial transient (if at all), when full compression is not necessary or achieved anyway, in practice the unit gives the

desired effect with the initial attack followed by an almost constant output until the input signal decays to an insufficient level.



#### HOUSING THE UNIT

Due to the high sensitivity of the unit it is advisable to use a metal case so that the circuitry is screened from sources of electrical interference. The case must also be very strong due to the manner in which the unit will be used. A diecast aluminium box measuring about 152 x 82 x 50mm makes an ideal housing for the project.

S1 is mounted on the top panel of the case, and this panel should be left clear of other controls, or sockets, which could otherwise cause confusion

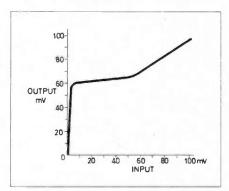


Fig. 3. Input voltage versus output voltage characteristic of the Sustain Unit.

or get in the way when operating S1. Apart from this the physical layout of the unit is not critical.

#### CIRCUIT BOARD

The circuit is built on a piece of 0·lin matrix stripboard, size 31 holes by 17 strips. The component layout and the breaks needed in the copper strips on the underside of the panel are shown in Fig. 4. The two mounting holes are 3·3mm in diameter and accept either 6BA or M3 mounting bolts.

Before mounting the completed panel connect flying leads of sufficient lengths at the four appropriate points. Spacers must, of course, be used over the mounting bolts to hold the connections on the underside of the panel clear of the metal casing.

There is a small amount of point to point fashion wiring, and this is illustrated in Fig. 5. SK1 has d.p.d.t. contacts, but four tags are unused in this application as only a s.p.s.t. switch is required. One of the four spare tags is used as a convenient point to provide the interconnection between the two battery clips.

#### USING THE UNIT

Best results from the unit will probably be obtained with the volume control on the guitar set at maximum. VR1 is merely adjusted to give the same volume level with the unit switched in as when it is bypassed.

The unit provides a considerable increase in gain, and this could lead to problems with stray mechanical or magnetic feedback between the loud-speaker and pick-up if the latter is not a humbucking type. In such cases it is essential to ensure that the guitar is as far away from the loud-speaker as can be arranged, and if necessary the volume control on the guitar can be backed off slightly. H

# SUSTA

#### COMPONE

#### Resistors

47k R1 R2 3.9k R3 4-7k R4 3-3M R5 15k

R6 1M R7 680k R8 100k

R9 390k R10 56k R11 4 · 7k 2·2M R12

All miniature 1 watt ±5% (±10% over 1M)

page 670

#### Potentiometer

VR1 47kΩ log. carbon

#### Capacitors

100µF 25 V elect. C1 C2 C3 C4 C5 C6 470nF type C280 10µF 25 V elect. 3.3nF ceramic plate 1μF 25V elect. 10µF 25 V elect.

1µF 25 V elect.

#### **Semiconductors**

IC1 LF351 f.e.t. input op-amp 2N3820 n-channel f.e.t. TR2 BC109C silicon npn BC109C silicon npn TR3

1N4148 small signal silicon D1 diode

D2 1N4148 small signal silicon diode

#### **Switches**

S1 d.p.d.t. sequential heavy duty push button type **S**2 Part of SK1

#### Sockets

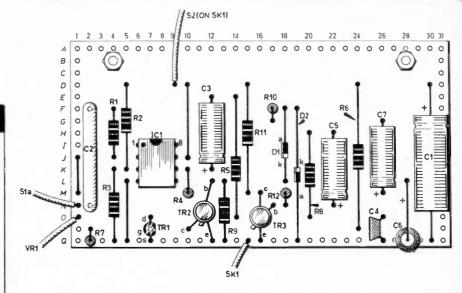
Standard (6.3mm) jack SK1 socket with d.p.d.t. contacts

SK<sub>2</sub> Standard (6.3mm) jack socket

#### Miscellaneous

Diecast aluminium box measuring about 152 × 82 × 50mm. 0.1in matrix stripboard. Control knob. Two PP3 size batteries and connectors to suit. Wire, solder.

£8.50



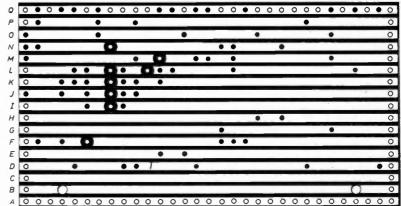
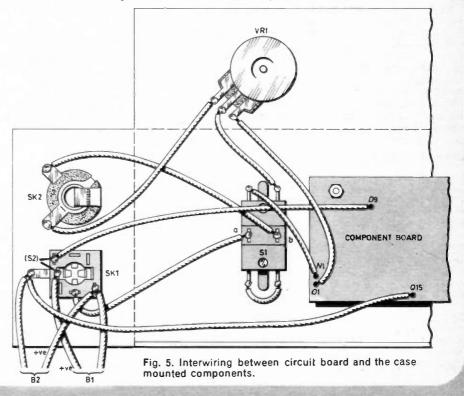


Fig. 4. Circuit board details, topside and underside.





BASIC ELECTRONIC THEORY WITH EXPERIMENTS

#### CONDUCTION IN MATERIALS AND DEVICES

ELECTRONICS is the study of devices in which electric charge is moved. The movement of charge is called conduction, so it makes sense to begin our course in electronics by studying this. We will then be ready to look next month at some of the devices in which charge is conducted.

#### ELECTRIC CHARGE

First of all we should ask "What is electric charge?" The surprising answer to this is that we do not really know what it is, in spite of the fact that it is a major factor in all our lives today.

We know that electrons and protons carry electric charge and that these charges are of two different kinds. We call them negative and positive, but the use of the word "negative" does not mean that electrons are lacking in charge when we say they carry negative charge.

Negative charge is just as real as positive charge. The two kinds of charge are simply opposites in the sense that one unit of positive charge (as carried by a proton) exactly cancels out one unit of negative charge (as carried by an electron). We will leave out of this discussion the fractions of unit charge carried by quarks and similar sub-atomic particles.

The only way we can tell that a particle is charged is when we can detect a force caused by the charge. The force between two similarly charged particles drives them apart. The force between two oppositely charged particles attracts them together.

The amount of charge on an electron is exceedingly small. For practical use we need a unit of charge much larger than this. Our unit is the coulomb (symbol C) which equals the charge on about six million million million electrons.

COMPONENTS required for experiments during the first 6-Parts of Teach-In 82. Complete kits of these (LIST 2) may be obtained from the retailers listed on page 670.

Resistors   Value   1   5 - 6Ω   1   39Ω   3   100Ω   2   180Ω   1   330Ω   2   470Ω   1   560Ω   1   156Ω   2   470Ω   1   150κΩ   1   150κΩ   1   150κΩ   1   150κΩ   1   220κΩ   1   270κΩ   1   270κΩ   1   470κΩ   1	COLUMN TO THE STATE OF THE STAT	£10	Semicon Quantity 1 1 1 2 1 1 1 1 1 1 1	
with short pre	eformed leads	s are not	Miscellar	neous
All leads on	components 1	to be be-	Quantity	Description
tween 0.5 and specified bread	0.8mm diam	eter to fit	1	Verostrip: 0·1 inch matrix 75 strips size 213 × 38mm (Vero 200-21086K)
Capacitors			1	Crystal Microphone Insert
Quantity Value			1	Toggle Switch: standard size
	-ceramic plat		1	s.p.s.t.  Rotary Switch 1-pole, 6-way,
1 47n F-	-metallised p rd C280, ITT I	DMT2R or	'	break before make contacts
simila		147 1211 01	1	Knob: to flt rotary switch and
3		16V types	00	match those in Minilab kit
		preferred,	20	Terminal Pins single-sided to
2 10μ F		axial or radial	_	fit Verostrip (Vero half-pin 200-21017B)
1 47μF	electrolytic	leads. Short	5m	P.V.C. Covered Wire: stranded 7/0-2mm: 1 metre of
2 220μF	electrolytic	lead-out types are not	1m	each of 5 different colours Tinned Copper Wire: 20 s.w.g. Hardware:
All to be suita Very large tobtained.		suitable. ng at 12V. not be	2 2 6	4BA bolt (25mm) 4BA shakeproof washer 4BA nut

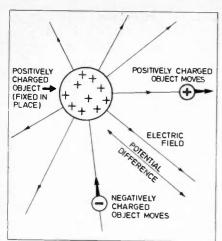


Fig. 1.1. Electric field around a charged object. Potential is high close to the object, but low at a distance.

#### CONDUCTION

If charge is to be carried from one place to another, there are three requirements:

(1) Charge carriers

(2) A medium in or through which the carriers can move

(3) An electric field to provide the force to make them move.

Electrons have already been mentioned as examples of charge carriers. These are the type we find most often in electronics, but there are other kinds, as we shall see later.

An electric field exists when there is a potential difference between one point and another. In Fig. 1.1 the charged object contains many particles which are positively charged. There is an electrical field around the object. Positively charged bodies are repelled by it, negatively charged objects are attracted by it.

If these charged objects are free to move, they will move. There will be movement of electric charge—which is what electronics is all about.

When we talk about the movement of charge we do not usually think of the amount of charge being moved. We think of the rate at which charge is moved. Our unit for rate of movement of charge is the ampere, (symbol A). A current of one ampere carries charge from one place to another at the rate of 1 coulomb per second.

The motion of charge causes a magnetic field, and an ammeter uses this effect to measure current.

The strength of an electric field is measured by the potential difference (p.d.) between two points in the field. Roughly speaking, the p.d. tells us how much energy is needed to move a charged object from one point in the field to the other. The unit for measuring p.d. is the volt (symbol, V).

Now that we have looked at some ideas on conduction, let us see how well different kinds of material can

conduct by carrying out some simple experiments.

It is expected that the Minilab (see page 664) will take a little time to complete, but this should not prevent you from carrying out the experiments immediately. The only facilities we shall be using from the Minilab in the early experiments will be the battery supply, the meter and of course the Verobloc.

You can see from the experiment layouts, in Fig. 1.3 for example, that connections to the meter and battery are made via screw terminal blocks. Essentially this is all that is required, especially for the simple experiments. Even the terminal blocks may be dispensed with initially with the meter and battery leads "plugged" directly into the Verobloc.

The experimental layouts therefore are suitable for use with and without Minilab.

# EXPERIMENT 1.1 How well do different materials conduct?

Look at the circuit diagram (Fig. 1.2) and how to set it out on the breadboard in Fig. 1.3. You are using

the electric field produced by a 3V battery. In other words, the p.d. between the ends of wires A and B is 3V. The meter is included in the circuit to measure the current. A sensitive meter is used because the currents will be small.

When the needle is deflected fully to the right the current is  $100\mu A$ . The symbol " $\mu A$ " is read as "microampere", or "microamp" for short. A microampere is one millionth of an

ampere.

The aim of this experiment is to measure the current that flows through different materials when a p.d. of 3V is applied. Use the two bare ends of wires X and Y as probes to touch against the different materials and devices listed below. Measure the current (if any), and keep a note of its value. Be ready to take the probes away quickly if the needle should swing beyond the right-hand end of the scale. Here are some tests to try:

(a) Hold one bare wire end X and Y in Fig. 1.3 in each hand, gripping its metal firmly. Measure the current flowing through your body (seen on the meter). Remember this point in future, and always hold the wire by its insulation. Remember this also when handling plugs—always hold

#### **EXPERIMENT 1.1**

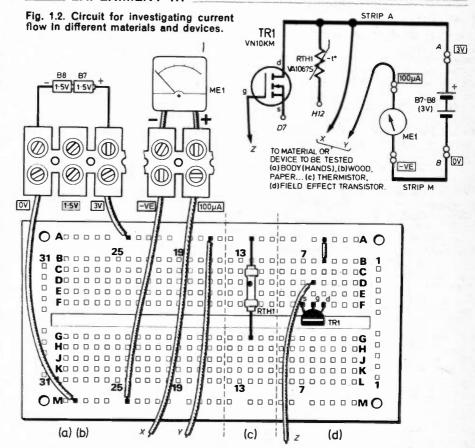
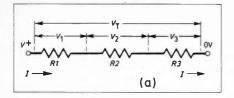


Fig. 1.3. A suggested layout on the breadboard and wiring to meter and batteries for the circuit in Fig. 1.2.



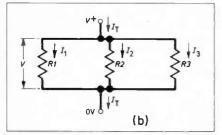


Fig. 1.6. Two ways of connecting resistors. In (a) they are said to be in series and in (b) they are connected in parallel.

#### OHM'S LAW

In the early 19th century a German physicist, George Ohm, did some experiments rather like the ones you have just done. He did not have semiconductors to work with, but he measured the current flowing through wires of various kinds. He summarised his results like this:

"The current passing through a wire at constant temperature is proportional to the p.d. between its ends."

This has since become known as Ohm's Law. Nowadays we use this idea to define our unit for measuring resistance that a wire or other piece of material offers to the flow of electric current. We say that if the p.d. between two ends of a wire (or other material) is 1V and the current is 1A the resistance of the wire is 1 ohm (symbol,  $\Omega$ ).

We can write this as a formula which we shall use many times during this course:

$$V = IR$$

In this formula, V is the p.d., measured in volts; I is the current, measured in amperes; and R is the resistance, measured in ohms. We can write the formula in two other ways:

$$I = \frac{V}{R}$$
 and  $R = \frac{V}{I}$ 

Given any two of the quantities, we can calculate the third.

#### COMBINING RESISTANCES

There are two ways in which resistances can be combined, in series and in parallel. Fig. 1.6 shows what these terms mean.

In Fig 1.6a the same current I passes through each resistance. So we can write:

 $V_1 = IRI$  and  $V_2 = IR2$  and  $V_3 = IR3$ The total p.d. across all three resistances is:

$$V_T = V_1 + V_2 + V_3 = I(RI + R2 + R3)$$

If the combined resistance of all three is R<sub>T</sub>, then:

$$R_{T} = \frac{V_{T}}{I} = \frac{I(RI + R2 + R3)}{I}$$
  
=  $RI + R2 + R3$ 

In short, their combined resistance equals the sum of their separate resistances.

With resistances in parallel, there is the same p.d. across all three resistances, but different currents may flow through each. The total current

$$I_{\mathsf{T}} = \frac{V}{RI} + \frac{V}{R2} + \frac{V}{R3}$$
$$= V\left(\frac{1}{RI} + \frac{1}{R2} + \frac{1}{R3}\right)$$

Their total resistance can be found from the formula:

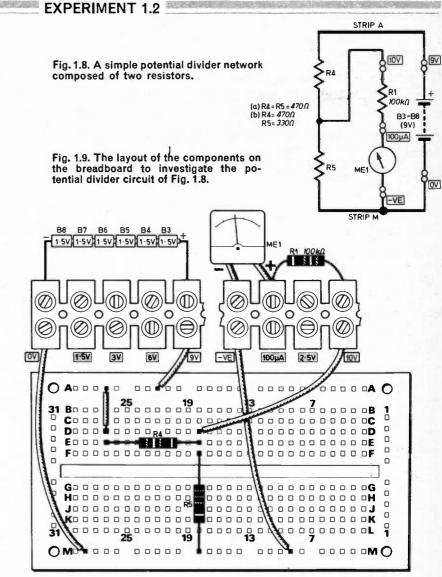
$$\frac{1}{R_{\rm T}} = \frac{I_{\rm T}}{V} = \frac{1}{RI} + \frac{1}{R2} + \frac{1}{R3}$$

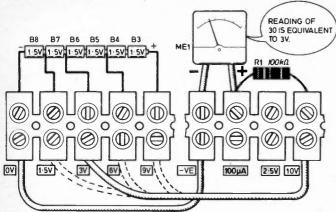
#### MEASURING P.D.

Given a known current and a known resistance we can calculate the p.d. across the resistance. We make use of this fact in electronics to measure p.d. by measuring current.

The meter in Minilab has resistors R1, R2 and R3 connected in series with it, see Fig. 1 on page 665. The resistance of the meter in the prototype is  $4000\Omega$  (4 kilohm, or  $4k\Omega$ ), and the resistance of R1 is  $100k\Omega$ . Their combined resistance in series is  $104k\Omega$ , according to the formula above. If we connect a supply voltage, say  $V_1$  to -ve and lov to pass a current of 100 µA through the meter and R1, the p.d. between the sockets is  $V_1 = IR = 1000 \times 10^{-6} \times 104 \times 10^6 =$ 10.4V.

The factor 10-6 is used because current is in microamps  $(1\mu A = 10^{-6}A)$ , and the factor 103 is used because the resistance is in kilohms  $(1k\Omega = 10^3\Omega)$ .





The meter shows full scale deflection (f.s.d.) when the p.d. is 10·4V, which is close enough to 10V for our

purposes.

The series resistor converts the microammeter to a voltmeter reading 10V f.s.d. Readings of other p.d.s are

Fig. 1.7. Testing the meter connected as a 10 volt full scale voltmeter. The wire shown in 3V position should be placed in each of the positions: shown dotted.

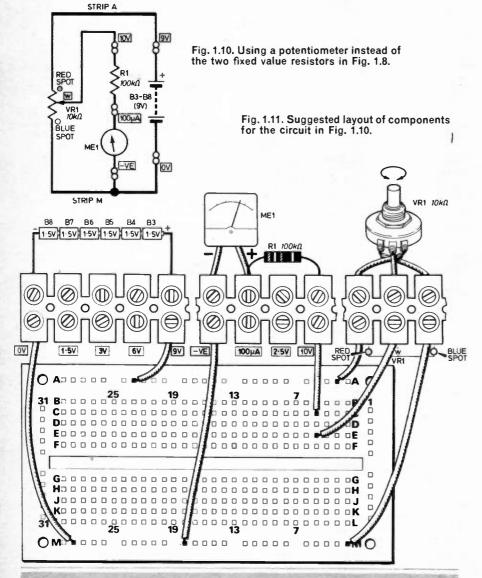
in proportion: when the needle points to "40", the p.d. is 4V.

To test the voltmeter, connect up as in Fig. 1.7 and put the meter 10V lead in to each of the battery ter-

minal block positions in turn except 12v.

When we connect to -vE and 2.5V the meter is in series with R2 and R3. Their total resistance is  $4k\Omega + 20k\Omega + 1k\Omega = 25k\Omega$ . For a f.d.s. current of  $100\mu$ A, the p.d. between sockets is

EXPERIMENT 1.2



 $V = 100 \times 10^{-6} \times 25 \times 10^{3} = 2.5V.$ 

This gives a 2.5V f.s.d. voltmeter, useful for measuring lower p.d.s. with greater accuracy.

#### POTENTIAL DIVIDER

If we require a p.d. which is less than any of the fixed p.d.s from the battery, we can use two resistors in series as shown in Fig. 1.8. If the p.d. across the two resistors is  $V_{\rm IN}$ , the current is  $I=V_{\rm IN}/(R4+R5)$ . The p.d. across the lower resistor is  $V_{\rm OUT}$  and  $I=V_{\rm OUT}/R5$ . Since I is the same for both equations:

$$\frac{V_{\rm IN}}{R4 + R5} = \frac{V_{\rm OUT}}{R4}$$

giving:

$$V_{\text{OUT}} = \frac{V_{\text{IN}} R5}{R4 + R5}$$

The network of resistors has divided the potential, so we call it a potential divider. The calculations assume that no current flows from the divider into the meter or into any other circuit that may be connected at the same point. In practice a small current does flow to the meter, so the actual p.d. across R5 is always a little less than the calculated value.

However, provided the outflowing current is small, we need not worry about this. In general, if the current leaving the junction of the resistors is less than one tenth of the current flowing through the network, we ignore its effects.

#### **EXPERIMENT 1.2**

To make and investigate a potential divider network

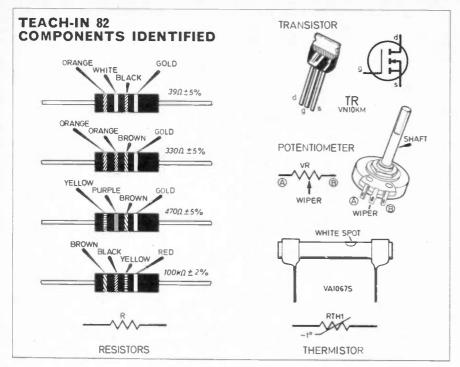
The circuit for this experiment is shown in Fig. 1.8 and the suggested layout on the breadboard in Fig. 1.9. The voltage across R5 can be read from the meter.

Measure the output p.d., using the meter. Does this agree with the result obtained by calculation? Try again with a  $330\Omega$  resistor in place of R4. Then make R4=470 $\Omega$  and R5=  $330\Omega$  and try again.

If we connect a variable resistor as in the circuit of Fig. 1.10, it acts like two resistors in series, with the wiper at their junction. The resistor becomes a potential divider which can be set to give any p.d. between 0V and  $V_{\rm IN}$ . Investigate its action, using the layout shown in Fig. 1.11.

#### PRACTICE SPOT

In Experiment 1.1 you measured current through various materials and devices when the p.d. was 3V. Calculate the resistance of (a) your body, hand to hand; (b) the wet tissue; (c) the lamp; (d) the thermistor warm and cold; (e) the transistor warm and cold.



The components used in this month's experiments are shown above. Here we show a physical drawing of the component, its circuit symbol and its circuit reference.

There are four different resistor values required. The value of each is

coded on the body in the form of a number of coloured rings. See page 697 for full account of resistor coding.

The transistor is a special type, a vmos power field effect transistor. This must be connected exactly as shown in circuit diagram—otherwise

the device may suffer damage.

The potentiometer can be considered as a resistor which can be tapped anywhere between its extremes. As the shaft is rotated clockwise, the wiper moves towards B.

The VA1067S rod thermistor is a heat sensitive resistor made of semi-conductor material. May be connected in circuit either way round.

To be continued

#### **QUESTION TIME**

1.1. Which part of an atom carries negative charge?

1.2. What kind of force exists between two positively charged objects?

1.3. What is the practical unit of electrical charge?

1.4. Name three examples of charge carrier.

1.5. If a current of 2 amperes flows for 3 seconds, how much charge is carried?

1.6. Which kinds of material make the best conductors?

1.7. What does f.s.d. stand for?1.8. The bands on a resistor are brown, red, orange. What is its

resistance?

1.9. When a resistor has a p.d. of 3.4V across it, the current flowing

through it is 5mA. What colours are the bands on the resistor?

1.10. If the potential divider of Fig. 1.8. has  $R1=150k\Omega$  and R2=

15 $\Omega$  and V<sub>IN</sub>=11V, what is V<sub>OUT</sub>? **Answers in Part 2.** 

two days and most likely there will be an intervening weekend.

Many firms have a fixed postal charge which means you lose on the small parcels and gain on the big ones. Last, but most important of all, write your name and address clearly on your order.

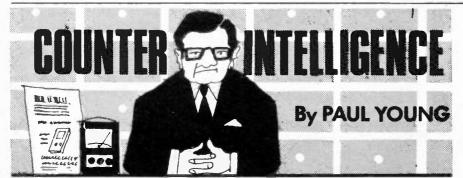
#### Component Ratings

Probably the two most used building blocks in construction are resistors and capacitors, so a few words on these would not be out of place. The two factors you need to know with capacitors are the capacity given in either Picofarads (pF), Nanofarads (nF) or Microfarads ( $\mu$ F) and the voltage.

The designer usually states quite clearly the type he expects to be used, that is, ceramic, silver mica, polyester or electrolytic, but remember that with electrolytics the tolerance is usually +50 per cent and -20 per cent. As far as voltage goes, it is usually safe to go above the suggested one, but not below, remembering of course that as the voltage increases, so does the physical sizel

The two factors given with resistors are wattage and tolerance but here we have a curious anomaly. It might be thought that the wattage is given to indicate the current passing through them, which in term means it must dissipate a certain number of watts, but when the value given is below one watt it is usually given as an indication of physical slze; larger ratings and in some cases even smaller wattage types may be used.

As regards tolerance, these days most resistors are 5 per cent which is adequate for the majority of purposes.



#### **Ordering Components**

Around this time of the year many new recruits are commencing the most fascinating of all hobbies, "Electronics", and it is my agreeable task to guide their faltering footsteps in the mysteries of purchasing components. This being the case I must ask my regular reader to bear with me, that is, unless he or she wishes to refresh his or her memory on these procedures.

You may be extremely lucky and have a little man round the corner who will attend to some of your requirements, but many of you will have to rely on Mail Order. Don't be dismayed by this, there are plenty of us dying to fulfil your needs and while I and my colleagues have always been eager to help you, the recession has put a keener edge on our performance. In other words it is a "Buyers Market".

I suggest your first move is to send for about four or five catalogues and then study them carefully. Make a list, showing

who has the best selection of capacitors, resistors, etc., and so on. For example, dealer "A" might have the best selection of variable capacitors, and dealer "B" the largest selection of transistors, this will be a useful guide when ordering parts for a project. Let me say, at this point, never fail to read the *Shop Talk* page which gives invaluable information especially on the sources of supply of unusual components.

Let us assume you have reached the point where you have made your list of what you wish to buy. Most mail order companies supply order forms, always use them when provided. Paperwork is time consuming and costly and not many of us keep copies of your orders. Instead we ask you to return the original, in case of a query, so make sure you keep the order until you have checked that all the goods are there and as ordered.

Naturally, you are keen to start on your project and any delays are irritating but always remember your letter may take two or three days to reach us and even with a by return service you must allow another



THE Minilab was designed to provide a portable electronic experimental deck for those who are starting to follow the Teach-In 82 series beginning this month, see page 658. With the facilities provided by Minilab and the additional components listed on page 658, you will be able to undertake every experiment described in Parts 1 to 6 of this series.

When the series is over, Minilab will continue to be of great use to you. In fact it is a generally handy workplace for any electronics enthusiast, ranging from the beginner to one who has reached the stage of wanting to design and build his or her own circuits.

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Everyday Electronics

October 1981

#### **FEATURES**

The general purpose circuitry within the basic Minilab is shown in Fig. 1 and provides the facilities listed below. All facilities are accessible through terminal strips TB1, TB2, TB3, mounted immediately behind the Work Panel.

#### Power Supply

With the younger and least experienced experimenters in mind we have decided that a battery is the safest and most convenient form of supply. The circuits to be built have low current requirements, for which a battery is perfectly adequate. There

is also the great advantage that the Minilab is comportable pletely and can be used in any room, without need for a mains power socket.

The battery supply consists of eight dry cells, B1 to B8 providing a maximum voltage of 12 volts(V) d.c. It is tapped at several points to give 1.5V, 3V, 6V, 9V, as well as a balanced ±6V supply for operational amplifiers.

The batteries are housed in the case Minilab, gether with all the other facilities.

#### COMPONENTS

BY O.N.BISHOP

A complete kit of the components listed below may be ordered as LIST 1 from the retailers appearing on page 670.

#### Resistors

 $100k\Omega$ R1  $20k\Omega$  see text R2 R3

All 1 W carbon or metal film ±2%

#### **Potentiometers**

VR1 10kΩ carbon lin. law VR2 100kΩ carbon lin, law

#### Capacitor

500pF miniature variable capacitor, solid dielectric

#### **Semiconductors**

TIL209, TIL220 or D1, 2, 3 similar red light emitting diodes with fixing clip/ bush (3 off)

#### **Switches**

push-to-make, release-S1. S2 to break miniature (2off) **S3** push-to-break, releaseto-make miniature **S4** single-pole doublethrow standard toggle

#### Miscellaneous

ME1 100µA d.c. moving coil panel meter-see text LS<sub>1</sub> 80hm moving coil loudspeaker, 45 to 80mm dia.

B1-B8 HP2 1.5 volt cells (8 off)

TB1-TB3 2-amp 12-way screw terminal blocks (3 off)

Verobloc solderless breadboard or-similar-see text; battery holders to suit four HP2 type cells (2 off); control knobs, skirted with index line to fit VR1, VR2 and C1 (3 off); p.v.c. covered stranded wire, 7/0.2mm, 1 metre each of six different colours (6 metres); OBA solder tags (3 off)—only for in-line battery holders; 2A 12-way terminal block (for modules); materials for case, see CUTTING LIST on page 667.

Guldance only £15 excl. batteries Approx cost and case

#### Meter

A single low-cost microammeter ME1 is provided for measuring currents up to 100 microamps. The series resistors (R1, R2 and R3) allow it to be used to measure voltages up to 2.5V or 10V.

The values of the resistors, in particular R2 and R3, shown in the circuit diagram have been selected for use with a particular meter having an internal resistance of 4 kilohms. Teach-In 82 kit suppliers will be selecting resistor values to suit the meter supplied which may be other than 4 kilohm. This selection is only likely to affect the values of R2 and R3. It is possible that only a single resistor will be required and supplied.

By means of suitable shunt resistors it can also be used for measuring currents larger than 100 microamps. The way to use this meter will be fully explained during the series.

#### Light emitting diodes

Small lamps have many uses as visual indicators as will be explained during the series. The Minilab is equipped with three visual indicators in the form of light emitting diodes

(l.e.d.s) D1, D2 and D3 mounted on the Control Panel. These are semiconductor devices which when excited by the application of a voltage, emit a red light.

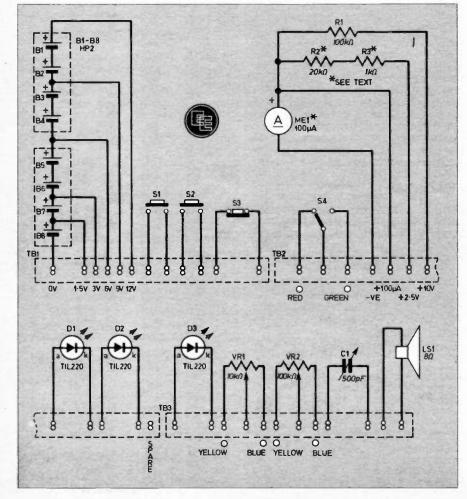
#### **Switches**

There are four switches S1, S2, S3 and S4 having three different functions: S1 and S2 are push-to make, release to break button types; S3 is the inverse of the previous being push-to-break, release-to-make push button type; S4 is known as a single pole double-throw (s.p.d.t.) or single-pole two-way. Here the centre terminal makes with either of the other two depending on its position.

#### **Potentiometers**

There are two potentiometers VR1, VR2 having values of 10 kilohm and 100 kilohm respectively. These devices have three terminals. The extremes have a specially prepared carbon track between them forming a resistance in these instances of 10 kilohm and 100 kilohm. The centre terminal, known as the wiper is in contact with the track and its position along the

Fig. 1. Circuit diagram of the basic Minilab.



track is determined by the angular position of its shaft. Variable resistors may be formed by using the wiper and one other terminal.

#### Variable Capacitor

A variable capacitor CI having a capacitance suitable for radio tuning circuits is included.

The capacitance is a function of the fixed distance between, and overlap area of, two sets of parallel plates. Angular rotation of its shaft moves one set of plates relative to the other varying the overlap and thus its capacitance.

#### Loudspeaker

A miniature low power loudspeaker is fitted to the base of the Minilab for audio experiments. This transducer converts electrical signals to sound waves.

#### Modules

The Minilab described in this issue is the basic version. It will be noted that there are vacant spaces on the Control Panel. These are left ready to receive the controls and readout of three special modules, that are to be built into Minilab at a later date during the series. The circuits and the way the modules work will be described in the series as it progresses.

Each module is simple to construct and is a useful experimental and design tool. The three modules to be included are:

(a) a low-noise signal amplifier

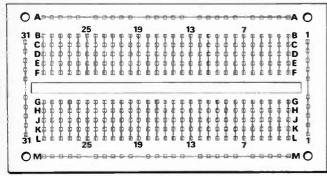
(b) a square-wave oscillator which can be set to any one of four frequencies

(c) a 7-segment l.e.d. display.

#### BREADBOARD

A "solderless breadboard" is used for interwiring the components according to the circuit diagram. It has a matrix of "sockets" arranged on a 0.1 inch grid and these are grouped (wired internally) to provide blocks of common sockets, in all 58 five-input sockets with four separate bus bars on the perimeter in the case of the Verobloc used here. The centre section sockets each allow up to five components to be interconnected at one point and should therefore be suitable for many circuits. This number can be extended by linking two or more groups as required.

The leads of the components and wires are simply pushed into the sockets to make a good electrical connection. This is an economical and quick way of wiring up components for experiments and project testing. When the experiment is completed, they can be quickly "unplugged" ready for re-use in other experiments.



How the contacts are interwired inside the Verobloc.

The breadboard is fitted to the Work Panel with double-sided adhesive tape. There is ample working space around the breadboard with plenty of room for mounting an additional board or boards should your enthusiasm take you in the direction of expansion.

The breadboard used in Teach-In 82 is the Verobloc, which is one of the smaller and least expensive of the 0.1 inch matrix boards. Should you already possess a different board of this type (for example EuroBread-Board or Protoboard) it can be used instead of the Verobloc.

Note that integrated circuits are to be used on several occasions during the series, so it is essential that whatever breadboard you install is capable of holding at least two 16-pin dual-inline integrated circuits.



#### CASE

Details for constructing the Minilab case are shown in Fig. 2. Also contained on this page is the cutting list for all parts. All dimensions are in millimetres.

All parts should be cut to size according to this list and any rough edges arising from saw cuts rubbed smooth with glasspaper. Nails and woodworking adhesive, Resin W. were used for all fixings. The nails were later punched below the surface and filled in with Polyfilla and then sanded

Fix the batten, control panel support and rear panel fixing to each frame side panel followed by the batten on the frame front member. When set, apply adhesive to the ends of the frame front member, frame divider and frame top member and

The Verobloc may be used as a "handle" to used as a "handle the Work remove Panel allowing access to the storage compartments.

secure each in position with nails using the base panel to keep the assembly square. When set, glue and nail the base in position, having previously made the array of holes to suit the loudspeaker. This can be followed by the horizontal rear panel support. The centre support can finally be glued in position to complete the case frame.

In the prototype the Work Panel and Control Panel were made up from 3mm plywood and covered with Formica. These may be painted or varnished if preferred.

The bare plywood panels, should be tried in position on the framework, and if necessary their dimensions modified to fit neatly in their final positions. A gap of about 4mm should exist between Control Panel lower edge and frame divider to allow the wires to pass from the terminal blocks to control panel components. A slightly oversized piece of plastic laminate can be glued to the panels using Evo-Stik or Thixo Fix and when set filed down to size.

Only the drilling of the rear and control panels remains. With the latter, use of a sheet of self adhesive paper or a number of self adhesive labels strategically placed are advised for marking out the positions of the holes on this panel. Besides providing protection to the surface during this operation, the paper will also reduce any likelihood of the drill bit slipping as drilling starts.

#### NOTES ON DRILLING

It is advisable to wait until you have the Minilab components to hand before drilling the holes in the Control Panel. The diameters of the holes shown are for the particular components at hand when building our Minilab. It is possible that you may obtain alternative types of component which may not be exactly the same size as specified. Do not be tempted to rearrange the layout on the Con-



modules to be added later.

Panel mounted components for the modules will not be at hand until later in the series, but it is advised that the holes to accommodate these are made at the same time as the others, before any components are mounted.

We believe it will be safe to go ahead and drill the three uppermost right hand holes the same size as that you drill for S4. Also the hole between VR2 and C1 the same size as VR2 fixing hole.

The rectangular cut-out in the panel is to suit a particular 7-segment display that will be uniquely defined later in the series. It is in order therefore to cut this aperature to the size shown

#### METER CUT-OUT

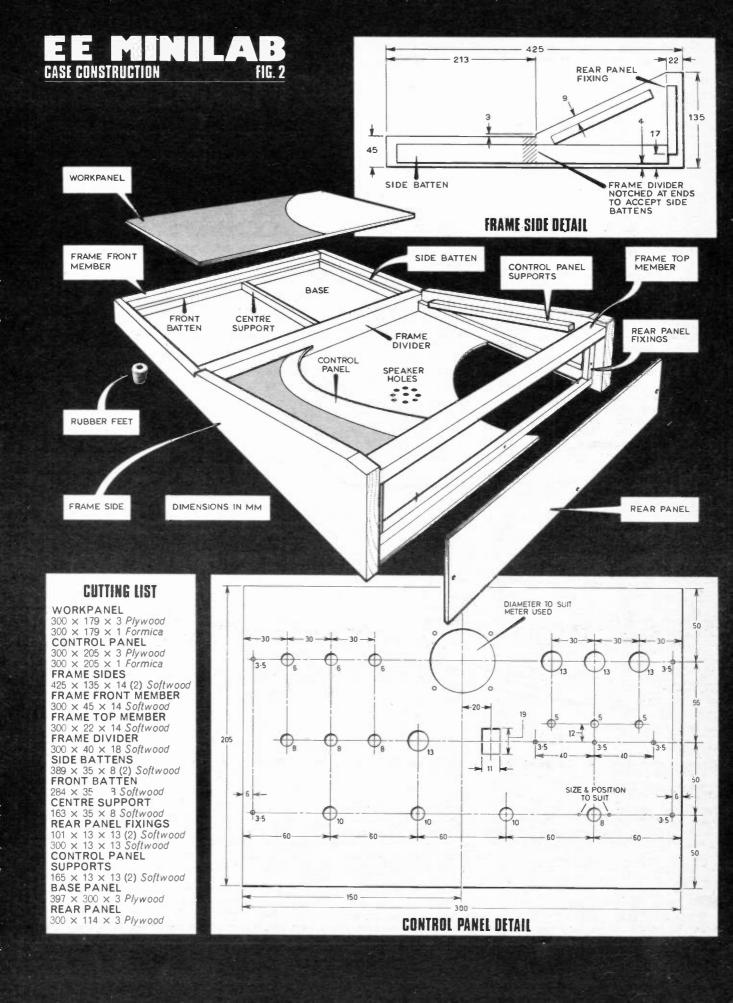
A large circular cutter will be required for making the meter cut-out, its diameter to accommodate the meter rear moulding. Alternatively a ring of small holes may be drilled side by side around the internal circumference of the cut-out and a small saw blade used to cut between adjacent holes and then filed smooth using a half-round file. A fret saw may also be used to make the cut-out.

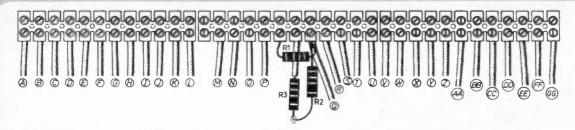
Fixing hole positions can then be determined and made with the meter sitting in the cut-out.

The rectangular cut-out is to accommodate a seven-segment display to be fitted later in the series and is easiest made by drilling two holes about 10mm diameter and then filing square. A file with an uncut edge is advised.

The two fixing holes for the variable capacitor should be marked and drilled using measurements taken from the capacitor supplied in your kit. Alternatively it can be glued using an adhesive such as Araldite.

Finally countersink the four panel fixing holes.





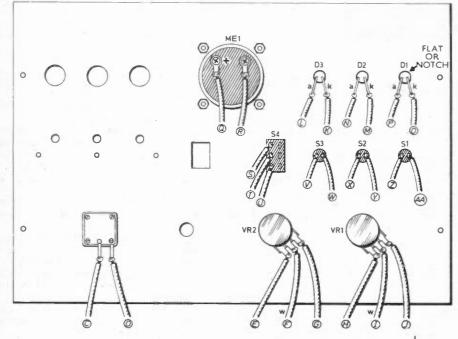


Fig. 3. Overhead view of screw terminal blocks showing wiring positions of leads from Control Panel mounted components. You can see the meter series resistors R1, R2 and R3 connected directly to the terminal block and secured by the terminal block screws. Solder R2 and R3 after they are secured to the block.

Fig. 4. Underside view of the Control Panel with all its components in place and wiring details to the terminal blocks. Connect A to A, B to B and so on.

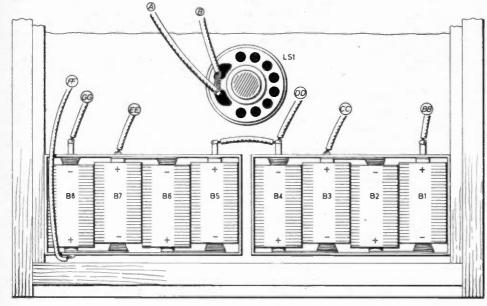
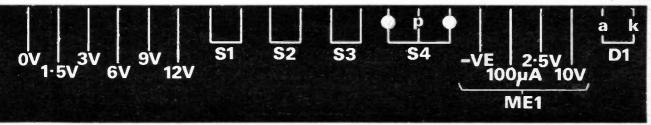
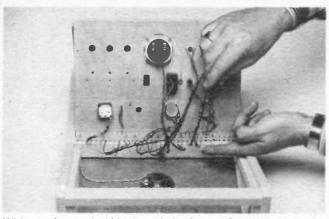


Fig. 5. Wiring details when using the lateral style battery holders showing connections to be made to the terminal block, see Fig. 3.

Fig. 7. Full-size label that may be cut out or copied and then glued on the Work Panel top edge to identify functions of terminal block positions.





Wiring to the terminal blocks with the Control Panel conveniently supported in an upright position.

For neater appearance, the underside of the control panel top edge was chamfered so that it fitted close up against the frame top member.

When this has been done, the fixing holes for this panel should be made in the panel supports using the Panel as a template.

The external framework can now be prepared and painted; gloss or varnish is considered more suitable than emulsion paint.

#### CONTROL PANEL LABELLING

It is recommended that all controls/ indicators on the Control Panel are clearly labelled so that you know exactly which control to use during your experiments. This can be carried out with Letraset or similar rub-down lettering. The result with this method is seen in the photographs of our Minilab.

This is more easily done before mounting the components on the panel, but make allowance for protruding bushes, nuts and knobs which might cover the letwhen the tering items are in position. The lettering should be protected from wear by spraying with a clear lacquer such as Letracote.

Embossed labelling tape (for example, Label-Mate) may be used instead, but this and other "panel type labels" are better added after the items are mounted.

The coloured spots on the Control Panel are intended to help identify the settings of S4, VR1 and VR2. The spots may be painted on or you can punch discs from coloured card and stick them in place.

#### COMPONENT ASSEMBLY

Depending on which battery holders you have, wire up according to Fig. 5 or Fig. 6. Connect long enough leads to these to reach the terminal block positions when holders are in place. The holders should touch the inside face of rear panel support. These may

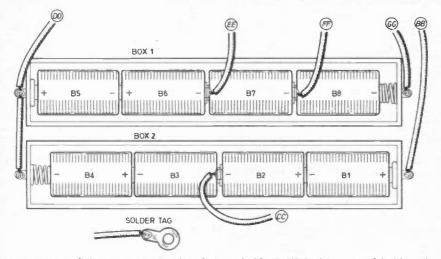
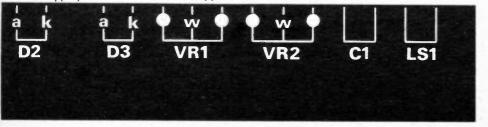


Fig. 6. Wiring of the alternative, in-line battery holders. With this type of holder, the tapping leads need to be fitted with a solder tag as shown. This is then pushed between the appropriate cells to make the tapped connection.



then be screwed or glued to the base panel. Do likewise with the speaker.

With reference to Fig. 4 mount all the components in place on the Control Panel. It may be necessary to recess the Control Panel underside at locations D1, D2 and D3 to properly fix the l.e.d. clips/bushes, and S1 S2, S3 to allow the fixing nut of these switches to be screwed on.

Connect sufficient lengths of wire to these components to comfortably reach the terminal blocks.

All terminal block positions are to be labelled on the Work Panel top side. A two-part label, actual size is provided in Fig. 7 and may be cut-out (or copied) and glued centrally on the Work Panel. With this done, the terminal blocks may be screwed to the frame divider to align with these.

With the Control Panel supported as seen in the photograph, all leads should be trimmed, insulation cut back and the stranded wire twisted and tinned (coated with solder). Referring now to Figs. 3 to 6, secure each of the leads, including battery and speaker, to their terminal block positions.

It is important to connect the meter and l.e.d.s the correct way round. The meter will be marked with a "+" and a "-" to correspond with "100µA" and "-ve" on the terminal block

The two leads from an l.e.d. are polarity conscious being known as anode(a) and cathode(k).

Also pay attention to connections from VR1, VR2 and S4 to avoid confusion during experiments later on.

The wires to C1 should be as short as possible and be as far as possible from any other wires.

Once you have checked and double checked the wiring, the Control Panel may be screwed in place and the batteries inserted. Pay special attention to cell polarity when doing this, see Fig. 5 (or 6). The rear panel may now be screwed on to complete the unit.

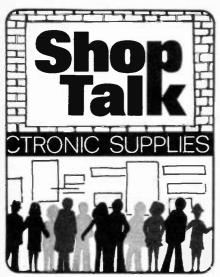
#### LEAD SET

A set of leads need to be made up to provide a means of interconnecting components on the breadboard and to the Control Panel components via the terminal blocks, see table below.

Length (cm)
20
20 15
10
6
4

We recommend 7/0.2mm p.v.c. covered wire.

About 5mm of insulation should be stripped from each end, and the exposed strands twisted tightly together. Both ends should next be tinned to produce a suitable "plug" for inserting in the Verobloc "sockets".



By Dave Barrington

#### Special Offer

We strongly advise all our readers to take advantage of our Special Offer (see page 664) of a £1 off a Verobloc solderless

breadboarding system module.

Apart from its use in the new Teach-In series, this Verobloc is suitable for most instant circuit experiments before being committed to a final "hardwired" version. Also it will repay for its self many times over in respect of damaged components through one cause and another.

#### Component Storage

When considering items for the workshop one item that seems to be regularly overlooked is the component storage bin or rack. However, if we are to make serious attempts at keeping the work area "clutter free" then these are well worth the investment.

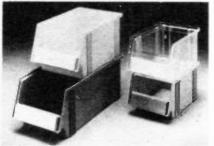
The price will, of course, vary according to whether you choose a single plastics or cardboard tray, or one of the multidrawer cabinets available from most of

our advertisers.

Available in three sizes and three colours, red, blue and yellow, the latest range of polypropylene storage bins from Link-Hampson are easily stackable and feature a corrugated Inner base. The new base, it is claimed, makes it easier to pick up smaller components, such as glass diodes and transistors, from the bottom of the bins.

A moulded handle on the front of each "tray" has a slot for inserting an identification tab. Also, see-through bins are

Storage bins from Link-Hampson.



included in the range and have the advantage of displaying their contents at a glance.

The largest of the polypropylene bins,  $400\text{mm} \times 156\text{mm} \times 186\text{mm}$ ; deep, cost £2·05 and the medium size,  $300\text{mm} \times 156\text{mm} \times 186\text{mm}$  deep, £1·65. The smallest bins, 250mm and 130mm  $\times$  149mm deep, cost 90p each.

The storage bins can be ordered direct or for addresses of nearest stockists write to Link-Hampson Ltd., Dept EE, 5 Bone Lane, Newbury, Berks.

Using their experience in providing document storage systems for banks, Bankers Box are now producing a range of fibreboard bins which make excellent low cost component storage trays.

This firm started by making document storage systems for banks, hence its name, but seeing possibilities in other fields it has applied its technical know how to broaden its range of products. The result is a flat-pack tray which folds into shape using no clips or staples. The folds are designed to provide plenty of strength and are grease resistant.

The bins come in seven sizes from 51mm wide by 102mm high by 305mm long up to 203mm  $\times$  102mm  $\times$  457mm. Available in packs of 10 the 102mm  $\times$  102mm  $\times$  305mm size usually retails at £3.75 per pack.

Further information can be obtained from Bankers Box, Dept EE, Doncaster Road, Kirk Sandall, Doncaster, DN3 1HT.

#### CONSTRUCTIONAL PROJECTS

Sustain Unit

Some of the components called for in the Guitar Sustain could cause readers

buying problems.

First investigations reveal that it would appear only Watford and Electrovalue Electronics stock the 2N3820 field effect transistor (f.e.t.). The f.e.t. input op-amp type LF351 is stocked by Rapid, Watford, Maplin and Electrovalue.

Although the jack socket SK1 is listed as a standard type incorporating a double-pole double-throw switch this item is in very short supply. It does not appear in the Watford catalogue but we understand that they are able to supply this component.

Of course, there is no reason why a single-pole on/off toggle switch and a standard jack socket should not be used. The layout of components within the case will, of course, have to be altered.

Capacitance Meter

The multiturn cermet trimmer potentiometer (VR1) is available from most advertisers but check that it is of the correct physical dimensions for mounting on the circuit board. This component is fairly expensive and could be replaced by a standard skeleton miniature preset type of the same value.

However, for the sake of accuracy we recommend that readers use the components specified in the parts list.

EE Mini Lab and Teach-In '82

For those readers who will have to master the art of woodworking when tackling the *EE MiniLab*, there is no need for dispair as most of the components required for the first few experiments in the *Teach-In 82* series can be wired direct to the Verobloc "instant" circuit module.

A list of complete kit suppliers for the EE MiniLab (List 11) and components for the first six instalments of Teach-In 82 experiments (List 2) is set out in table

below.

One of the reasons for the price variations (see relevent advertisements) is caused by component sourcing and the quality and type of the final selection of components used in making up the kits. This applies particularly to the meter, variable capacitor and battery holders.

Popular Designs

Most of the components called for in Popular Designs should be readily available from advertisers. We see that some of them offer an identical size piece of board in packs of five.

#### **TEACH-IN 82 KIT SUPPLIERS**

Supplier	LIST 1	LIST 2	LISTS 1 & 2
Bi-Pak (p. 645)	£15.65	£8-94	£23·00
Electrovalue (p. 709)	£16.56	£6.32	£21 · 79
Greenweld (p. 699)	£18.50	£8-10	£25.00
Magenta Electronics			
(p. 700)	£16.40	£9.34	£24.98
A. Marshall Ltd (p. 696)	£17.00	£12.50	£28·50
T. Powell (p. 708)	£19·75	£10.50	£26.50
T.K. Electronics (p. 705) Watford Electronics	£19·50	£9·50	£27·50
(p. 641)	£18.08	£9.33	£25·54

#### NOTE

- (1) All prices are inclusive of VAT, postage and packing.
- (2) If Vero coupon is enclosed deduct £1 from List 1 price quoted.
- (3) Price quoted in last column applies only when Lists 1 and 2 are ordered at the same time. Deduct £1 from this price if Vero coupon is enclosed.
- (4) For suppliers full address refer to page number following company name.





A COLLECTION of ten simple projects—comprising one from each volume of EVERYDAY ELECTRONICS 1971-1981. These have been selected to provide a wide range of subjects and fields of interest.

Any of these designs can be built on the piece of Veroboard (Stripboard) given with this issue.

# CONTENTS

1	Snap Indicator	672
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4	Heads and Tails Game	675
5	Continuity Tester	676
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9	Soil Moisture Unit	680
10	Ice Alarm	681

Newcomers to electronics should refer to page 682 for general information concerning construction of these projects.

# **SNAP INDICATOR**

1971/2

In some games, such as snap and question games, it is necessary to know which person or team is first, If recognition of pairs or readiness to answer is declared by voice, it is not always clear who was first.

The device described here was built to avoid this difficulty. Its simple electronic circuit is arranged so that when one person has pressed a button, a later response by the opponent is blocked; and an indicator lamp shows who was first. The circuit automatically returns to its original condition when the push-buttons are released, and hence is ready for the next turn.

#### CIRCUIT OPERATION

Transistors TR1 and TR2 act as switches for LP1 and LP2 that are in the collector circuits of the transistors.

One push-button is operated by each player and S3 is the on/off switch. Normally S1 and S2 are open and the transistor bases are held off and no collector current flows.

If S1 is now pressed, connecting R1 to the base of TR1, this shifts TR1 into conduction so that the indicator lamp LP1 lights. Almost the whole supply voltage is dropped across LP1, so that the supply voltage between the negative line and TR1 collector, R3 junction is very small. If S2 is now closed, TR2 will not be turned on since its base will not be taken positive enough. Hence LP2 will not light.

Should S2 be closed first, LP2 lights and LP1 cannot be lit.

Push-buttons S1 and S2 are bell-pushes, suitably placed for each competitor. The button is pressed and held down, to show recognition of pairs or readiness to answer. LP1 and LP2 are low-consumption bulbs (0.06A 6V) which limit collector current to a low value.

When both pushes are released, the circuit returns to the normal condition.

#### COMPONENTS

#### Resistors

#### All $\frac{1}{4}$ W $\pm$ 10% carbon **Transistors**

TR1, TR2 2N2926 silicon npn

#### Lamps

LP1, LP2 0.06A 6V bulb and holder

#### Switches

S1, S2 S.P.S.T. push button (bell push)
S3 S.P.S.T. toggle (on/off)

#### Miscellaneous

B1 4.5V torch battery

Veroboard: 10 strips by 12 holes 0:1in, matrix P.V.C. covered connecting wire (7 strand coloured)

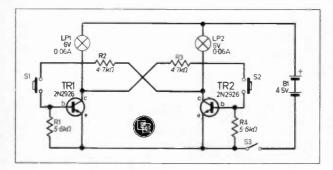
#### ASSEMBLY

The complete unit is housed in a small box, ours was a wooden box made up for the job, but almost any small box will do. Holes are cut for the two indicator lamps (LPI and LP2), the on/off switch, S3, and the wires to S1 and S2.

The wires used to connect the circuit to S1 and S2 are twin core mains or bell wire.

If a metal case is used it is a good idea to cover the metal area under the component board with insulation tape.

For team games, leads to S1 can run to two or more pushes connected in parallel, and leads to S2 similarly to the required number of pushes.



Flg. 1. Circuit diagram of the Snap Sequence Indicator.

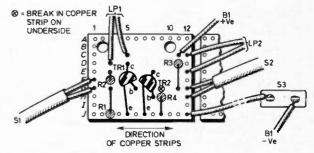


Fig. 2. Veroboard layout and wiring diagram.

# DAMP LOCATOR

11973

THE device to be described is an inexpensive, simple, but effective, damp locator that will reveal damp in a variety of normally non electrically conducting materials such as stone, concrete, plaster, papers and textiles.

Since the device is intended to indicate a low level of dampness that is normally hard to determine by hand or eye, the use of a meter was not merited, and indeed by using an l.e.d. a fair portion of space and cost are saved.

#### CIRCUIT DETAILS

The two transistors, TR1 and TR2, are arranged in a configuration known as a Darlington pair. This arrangement has the property of providing extremely high gain—approximately equal to the multiplied gains of TR1 and TR2—with very high input impedance. Thus a small current flowing between TR1 base/emitter causes a very much larger current to flow between TR2 collector and emitter.

When the Damp Locator is placed upon a damp surface, a minute current is able to pass via that surface between the sensors. This small current provides the base bias for TR1 causing it to be in the conducting state and allows a larger current to pass through its collector and emitter to the base of TR2. This in turn allows TR2 to reach a state of conduction allowing a current to flow through the light emitting diode, D1, damp is indicated by the resultant glow.

In the absence of a conducting media between the sensors, the circuit draws negligible current from the battery, thus an on/off switch was not used on the prototype, the battery being removed during long periods of non use.

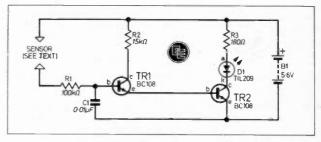


Fig. 1. The circuit diagram of the Damp Locator.

#### ASSEMBLY

The unit should be housed in a plastic box. A metal case is *NOT* suitable. The sensor is made from two parallel strips of 22 s.w.g. wire threaded through four holes drilled in the base of the box. The finished unit should appear to have a "railway track" running across the bottom outer surface of the case.

#### TEST

Carry out the following test. Hold the battery flying leads on the battery and touch the sensor lead (from B4) on the positive terminal of the battery. The diode should glow brightly. If this does not happen re-check and rectify.

#### **COMPONENTS**

#### Resistors

R1  $100k\Omega$ R2  $15k\Omega$  R3  $180\Omega$ 

1 watt ± 10% carbon

Capacitors

C1 0.01μF sub-miniature type

Semiconductors

TR1, TR2 BC108 or similar silicon npn
TIL209 or similar light emitting
diode

Miscellaneous

B1 5.6 volt Mercury type PX-23 Veroboard 10 strips 12 holes 0.1in. matrix; small transparent plastic box; 22 s.w.g. tinned copper wire. Connecting wire.

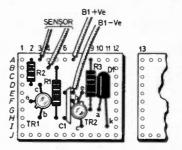


Fig. 2. The layout of the components on the Veroboard.

# TAPE NOISE LIMITER

1974

NEXPENSIVE portable cassette tape recorders are extremely popular, but they usually have only a rather limited output power, driving a small internal loudspeaker, and the noise level is rather high.

The noise level cannot be entirely overcome, but a worthwhile improvement can be made by connecting the unit described in this article between the output of the recorder and the amplifier input.

#### CIRCUIT DESCRIPTION

Resistors R1 and R2 form a potential divider across the supply, and produce a low voltage supply for the field effect transistor, TR1. With only a small supply potential such as this, the d-s connections of the f.e.t. act as a simple resistor. With the gate terminal tied to earth via R4, its resistance is very low, at about 100 ohms, or even less.

Resistor R3, together with the d-s resistance of TR1, R5, and C3, forms an attenuator. The main complication is C3, as its reactance (its resistance to a.c.) varies with frequency. It has a reactance of approximately 50 kilohms at 100Hz, but of only approximately 500 ohms at 10kHz.

This means that the attenuation factor of the circuit changes with frequency. If one calculates the attenuation factor of the circuit at 100Hz, and 1kHz, it will be found to be a little over one, which is barely noticeable.

If it is calculated for 10kHz, it will be found to be almost exactly three, which is of course considerable, and will increase still further at high frequencies. This gives the required treble cut.

Some of the input signal is fed via C4, VR1, C6, and R6, to the base of TR2. This is a high gain common emitter amplifier, and the amplified signal is fed from TR2 collector, via C8 and R9, to a voltage doubling rectifier circuit, D1, D2, and C7. The resultant negative d.c. bias is fed to the gate of TR1.

# 

Fig. 1. Circuit diagram of the Tape Noise Limiter.

On low level signals, this bias voltage will have little or no effect, but on strong signals it will be large enough to cause the resistance of TR1 to be greatly increased, to as much as a few megohms. Components C3 and R5 are virtually switched out of circuit, and the treble cut is thus removed.

The tape hiss is less noticeable in the presence of high frequencies, than in the presence of low or middle frequencies; C4, C6, and C8 are given rather low values, so that the circuit responds more readily to high frequencies.

#### A SSEMBLY

Solder all the flying leads to the board; the leads should be about 70mm long insulated wire. The board should now be mounted in a suitable case by means of two 12mm long 6BA nuts and bolts. Finally, wire up the flying leads to the case-mounted components.

#### ADJUSTMENT AND USE

The "tape" or "radio" input of most amplifiers or record players has a fairly high input impedance (50 to 100 kilohms), and should be used. The unit can be used with amplifiers having fairly low input impedances (5 to 10 kilohms).

Potentiometer VR1 adjusts the level at which the treble cut is removed. Experiment a little using various settings. If this is set too high, tape hiss will be heard on low level signals.

# 

Fig. 2. Layout of components on the Veroboard.

#### **COMPONENTS**

Res	istors
-----	--------

R1	$68k\Omega$	R4	100k $\Omega$	R7	$1M\Omega$
R2	18k $\Omega$	R5	15k $\Omega$	R8	$2 \cdot 7k\Omega$
R3	1 · 2kΩ	R6	2·2kΩ	R9	2·2kΩ

#### All 1 watt carbon + 10%

#### Potentiometer

VR1  $50k\Omega$  sub-miniature preset, horizontal Capacitors

#### C1 100. E alac

C1	100μF elect. 10V	C5	10μF elect. 10V
C2	10μF elect. 10V	C6	0.033µF
C3	0·033μF	C7	2.2µF elect. 10V
C4	0·1μF	C8	0·1µF

#### Semiconductors

TR1 2N3819 n channel f.e.t.
TR2 BC107 silicon npn
D1, D2 OA91 (2 off)

#### Miscellaneous

B1 PP3 9V battery SK1, 2 3.5mm Jack socket (2 off)

S1 s.p.s.t. rotary switch Veroboard: 0-1in. matrix; battery clips for

PP3; aluminium case  $135 \times 70 \times 40$ mm with removeable lid; control knob.

# HEADS & TAILS GAME

1975

This simple little novelty device demonstrates how it is possible to electronically simulate the tossing of a coin.

The circuit is arranged so that it is purely a matter of chance whether the heads or tails lamp is illuminated when the push button is pressed, the circuit thus providing the same effect as tossing a coin.

#### CIRCUIT OPERATION

The two indicator lamps are two light emitting diodes (l.e.d.s), D1 and D2. These are protected against passing excessive currents by the series resistors R1 and R4. Diode D1 is illuminated when TR1 is turned on, and D2 will be illuminated when TR2 is turned on.

If VR1 is ignored for the time being, when the push button switch, S1, is depressed, the positive supply will be connected to the circuit. The transistors will obviously both begin to turn on, TR1 receiving its base bias current via R4, D2, and R3, and TR2 receiving its base current via R1, D1, and R2.

It is, however, not possible for both transistors to turn hard on at the same time, as if TR1 is turned hard on, only a fraction of a volt will appear at its collector, and TR2 cannot receive the necessary bias current to turn hard on. If TR2 turns hard on then the same is true for TR1.

What happens when the supply is connected is that both transistors begin to turn on, but due partially to chance, and partially to a slight unbalance in the resistor values, transistor gains, etc. of each half of the circuit, one will begin to turn on faster than the other. In doing so it tends to starve the other transistor of base current as its own collector swings towards earth potential.

On the other hand this enables it to obtain a heavy base current from the collector of the other transistor, as this has its collector still at virtually the full supply potential. This regenerative action results in one transistor being biased to saturation, and the other being cut off. Obviously only only of the lamps will light up.

#### **COMPONENTS**

#### Resistors

R1 470Ω R3 82kΩ R2 82kΩ R4 470Ω

All 1W carbon ± 5%

Potentiometer

VR1 1kΩ horizontal skeleton preset

Semiconductors

TR1, TR2 BC109 silicon npn

D1, D2 T1L209 or similar l.e.d. with holder

Miscellaneous

S1 push-to-make, release-to-break pushbutton switch

B1 9V PP3 with connector

Veroboard: 0-1 inch matrix size 10 strips by 24 holes; case.

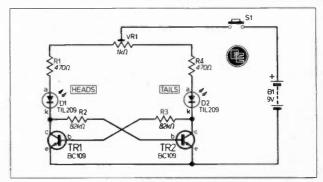


Fig. 1. The circuit diagram of the Heads & Tails Game.

VR1, compensates for the component tolerances by supplying a higher supply voltage to one or other side of the circuit, and is adjusted by trial and error.

#### **ASSEMBLY**

Connect up the l.e.d.s and S1 using approx. 75mm lengths of thin insulated wire. Make quite sure that the l.e.d.s are connected with the correct polarity.

The front panel of a suitable case should next be drilled to take S1 and the diodes. Switch S1 is mounted in the centre of the panel, and the l.e.d.s are mounted to the right of this, one above the other.

#### A DJUSTMENT

Start with the slider of VR1 at a central position and then press S1 a number of times (25 or more). It will probably be found that one lamp lights up much more often than the other. If D1 lights up more frequently, then VR1 should be adjusted slightly in an anticlockwise direction to compensate for this. If D2 lights up more frequently, then VR1 needs to be adjusted slightly in a clockwise direction.

Repeat this procedure until the circuit is properly balanced with each lamp lighting up approximately

the same number of times.

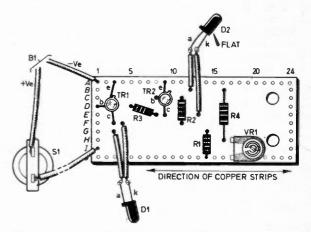


Fig. 2. Stripboard component layout.

# CONTINUITY TESTER

1976

When building and servicing electronic equipment, there are numerous occasions when some form of electrical continuity tester is required. It may be when tracing out the wiring around a complicated array of switches, or perhaps something more everyday such as checking for a break in a mains cable; sorting out the contacts of a wavechange switch; or checking a fuse which one suspects has blown.

The very simple tester described here produces an audible tone to indicate continuity. The unit is very simple to construct and is also quite inexpensive as few components are used.

# CIRCUIT OPERATION

The circuit consists of a relaxation oscillator, feeding a miniature speaker.

A unijunction transistor, TR1 forms the active component in the oscillator. Unijunction transistors have little in common with ordinary bipolar transistors except that they are also three terminal devices. The terminals are named differently though, being called base 1 (b1), base 2 (b2) and emitter (e).

With no voltage present at the emitter, the base 1 and base 2 terminals have a resistance of about three to 10 kilohms across them. Therefore, when the test prods are short-circuited, a current of about a couple of milliamps will flow through the loud-speaker via the unijunction.

This does not, of course, take into account that C1 will have charged to the supply potential within a fraction of a second of the battery being connected and so there is about 9V at the emitter of TR1.

If more than about half the supply potential is present at this terminal the emitter input impedance (which is otherwise extremely high) suddenly falls to a very low level and the base 1 to base 2 resistance of the device falls to about half its previous level.

Thus, at the instant the test prods are touched together, C1 discharges into the emitter of TR1 and a pulse of current is fed to the loudspeaker via the b1, b2 terminals of TR1.

Once C1 has largely discharged, TR1 operates as previously described until C1 is charged via R1 to the trigger voltage once again. Then C1 will again discharge and another pulse of current will be fed

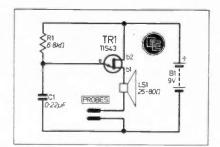


Fig. 1. The complete circuit diagram of the Continuity Tester.

# **COMPONENTS**

# Resistor

R1  $6.8k\Omega \pm 10\% \frac{1}{4}W$  carbon

## Capacitor

C1 0.22µF polyester type C280

# Transistor

TR1 TIS43 unijunction transistor

# Miscellaneous

LS1 25 to 80  $\Omega$  speaker

B1 9V PP3 battery

Veroboard 0.1in matrix; 10 strips × 11 holes; aluminium box; test leads and prods; speaker fret; PP3 battery connector.

to the loudspeaker. This will continue in rapid succession causing a continuous tone to be emitted from the loudspeaker as long as the test prods are connected together.

If a resistance of more than a few hundred ohms is present between the two test prods, TR1 will cease to function and no audio tone will be generated.

## **ASSEMBLY**

Virtually any small case can be used to house the unit, the minimum suitable size being about  $100 \times 70 \times 25$ mm.

# CONCLUSION

When using the unit it should be borne in mind that if the circuit under test has a resistance of perhaps as much as several hundred ohms the unit will still produce an audio tone even though there is not true continuity. This is not a major drawback however as, if the resistance between the prods is more than just a few ohms, the volume of the tone drops and the type of note changes noticeably.

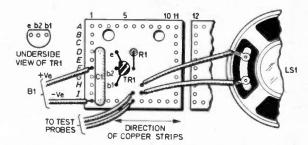


Fig. 2. Layout of components on Veroboard and complete wiring details.

# PHOTO FLASH SLAVE

1977

THERE now seems to be a very wide range of electronic aids available to photographers, and most such aids appear to be fairly complex devices. Although a photo-flash slave unit is probably the most simple of these devices it is nevertheless one of the most useful.

A flash slave unit triggers a secondary flashgun when it receives a pulse of light from the main flashgun.

# CIRCUIT OPERATION

A thyristor is a unilateral device, and it will therefore only function properly if it is fed with a voltage of the correct polarity. In order to avoid any problems here, a bridge rectifier consisting of D2 to D5 has been added in series with the flash lead, and this provides a signal of the correct polarity to the thyristor (CSR1) regardless of which way round the flash lead is connected.

Resistors R2, R3, and diode D1 form a simple Zener shunt regulator, and these limit the voltage which can be fed to the gate of CSR1 via the photo-Darlington amplifier TR1. Capacitor C1 is needed in order to provide the current required to trigger CSR1.

While Cl charges up it is receiving a current of only a few microamps for about one second. When the light from the main flashgun is received by the photocell (TR1), it's normally very high resistance falls to a level of only about a few hundred ohms. This happens extremely quickly, and as a result Cl almost instantaneously discharges through TR1 and into the gate of CSR1. Because the charge on Cl is released so rapidly, the discharge current is many times greater than the charge current. It is thus sufficient to switch on CSR1 which then places a low impedance path across the flash lead and so fires the flashgun.

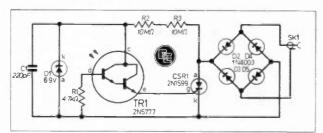


Fig. 1. Complete circuit diagram of the Photoflash Slave Unit.

# **COMPONENTS**

# Resistors

R1  $4.7 k\Omega$  R2  $10 M\Omega$  R3  $10 M\Omega$  All resistors are carbon  $\frac{1}{4}W \pm 10\%$ 

# Capacitors

C1 220nF polyester

# Semiconductors

TR1 2N5777 photo darlington npn CSR1 2N1599 200V 1A invristor or equiv-

alent

D1 BZY88C6V8 6·8V 400m W Zener

D2-D5 IN4003 rectifier (4 off)

# Miscellaneous

SK1 extension lead for flash gun Small transparent plastic case, 50mm × 40mm × 25mm; stripboard 0·1 inch matrix 10 strips × 12 holes; small rubber grommet;

solder.

After the flashgun has gone off the current through CSR1 falls to virtually zero and this component switches off. The circuit is then ready to start a fresh cycle when the flashgun is recharged again.

# **ASSEMBLY**

The unit should be housed in a small transparent plastic case.

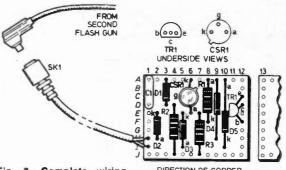


Fig. 2. Complete wiring diagram of the unit.

DIRECTION OF COPPER STRIPS ON UNDERSIDE

# FUZZ BOX

1973

THE POP musician is being constantly bombarded with new and improved musical effects units, but still polling high in the popularity charts is the "old" Fuzz Box.

# CIRCUIT DESCRIPTION

Input to the unit is at SK1 a stereo jack socket wired to complete the d.c. power circuit when the input jack is inserted. The signal then passes to the op-amp via d.c. blocking capacitor Cl. Resistor R1 sets the input impedance at 100 kilohms which should suit most guitars and electronic organs.

The gain of the sustain 741 op-amp is controlled

The gain of the sustain 741 op amp is controlled by feedback and is equal to: (VR1 + R2 + R3)/R2. With specified values gain is from 6 to 92, depending on the setting of VR1. However, all output signals are limited to 600mV peak by the effect of D1, D2 in the feedback chain. VR1 setting controls the rise time of the signals towards the diode clipping level and thus the high order harmonic content in the signal which determines the depth of fuzz.

The resulting signal from the op-amp is reduced in amplitude by the potential divide action of R6 and R7, giving an attenuation factor of approximately four. Thus the maximum output signal via C3 available for inputting to an amplifier is about 150mV. This level will be maintained during the period of clipping (fuzz) and will then decay naturally to zero.

The 741 requires a split supply and this is derived by the potential divide action of R4 and R5 producing  $\pm 4.5$  volts with respect to the op-amp reference line which is decoupled by C2.

A foot-switch S1 is incorporated to allow the unit to be readily by-passed when desired.

#### **ASSEMBLY**

The assembled stripboard, Fig. 2 was mounted horizontally in a die-cast aluminium box.

# **COMPONENTS**

# Resistors

# Capacitors

C1, C3 0·1µF plastic or ceramic C2· 10µF 6V elect.

# Semiconductors

IC1 741 operational amplifier 8 pin d.i.l. D1, D2 1N4148 or similar silicon diode

# Miscellaneous

SK1 standard stereo jack socket SK2 standard jack socket S1 s.p.d.t. successional action foot-switch

B1 9V PP3

VR1 100 kilohm carbon lin.

Stripboard: 0·1 inch matrix 13 strips × 21 holes; PP3 battery clip; aluminium diecast box; knob for VR1; connecting wire.

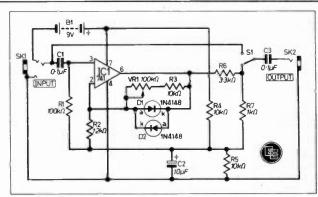


Fig. 1. The complete circuit diagram of the Fuzz Box.

Ø = BREAK IN COPPER STRIP ON UNDERSIDE OF BOARD (12)

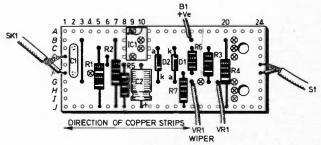


Fig. 2. Layout of components on the stripboard.

Attach the flying leads including the battery connector, Fig. 3, and then proceed with drilling the case.

Screened lead should be used for input and output connections.

A much harsher fuzz can be produced by connecting a 1 kilohm resistor across SK2, signal to earth.

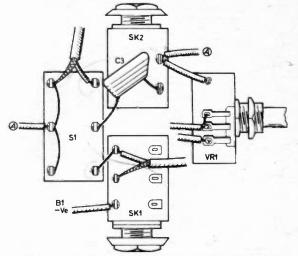


Fig. 3. Interwiring details. To be fitted in a metal box.

# OPTO ALAR

T His simple single-transistor circuit is designed to sound a miniature audible warning device when light falls on to a photocell. The photocell is normally mounted in a dark room and the alarm is triggered when either the room lights are switched on or possibly when light from an intruder's torch falls directly on to the photocell.

# CIRCUIT DESCRIPTION

The photocell PCC1 is an ORP12 light-dependent resistor which is located in the room to be protected, and is connected by means of PL1 and SK1. Together with R1, PCC1 forms a potential divider: the voltage at the junction of R1 and PCC1 varies with the amount of light striking the l.d.r.

In absolute darkness the resistance of an ORP12 is at least 10 megohms, and so the voltage at the junction of R1/PCC1 is very nearly that of the supply rail, 9V. Transistor TR1 is therefore firmly switched off as its base is not biased.

When light falls on PCC1, its resistance drops (albeit relatively slowly) and this causes TR1 to switch on. A triggering pulse is therefore delivered to the gate of CSR1 and this component conducts. The audible warning device (WD1) will therefore sound.

The thyristor will now remain in this low impedance state even if the triggering signal is removed. The only way to reset CSR1 and mute the alarm is to switch off the power supply. Resistor R5 will ensure that a minimum holding current is flowing in the anode-cathode circuit of the triggered thyristor, and so preventing any undesirable resetting.

# **ASSEMBLY**

The prototype was built into an ABS "Bimbox" type 4003. This measures approximately  $85 \times 55 \times 35$ mm and has an aluminium front panel.

There should be no problems with the construction of the circuit; Fig. 2 illustrates the recommended arrangement of components.

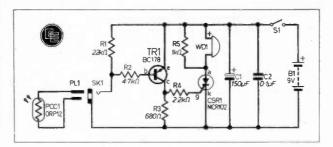


Fig. 1. The circuit diagram of the Opto Alarm.

# **COMPONENTS**

# Resistors

R1 22k $\Omega$ R3  $680\Omega$ R5  $1k\Omega$ R2  $4 \cdot 7k\Omega$ R4 2·2kΩ

All 1W carbon ± 5%

# Capacitors

C1 150µF16V elect. C2 0·1µF polyester C280 or similar

# Semiconductors

TR<sub>1</sub> BC178 silicon pnp

CSR1 MCR102 thyristor rated 30V 0.8A or at least 9V 100mA

PCC1 ORP12 or similar light dependent resistor

# Miscellaneous

SK1 3.5mm jack socket (2 off)

3.5mm jack plug PL1

WD1 miniature 9V audible warning device Stripboard: 0.1 inch matrix, 10 strips × 24 holes; case BIM 4003 or similar; twin-core flex; stranded connecting wire; 6BA fixings including 5mm spacers; 9 volt battery and connector; on/off switch.

All interconnections between the component board and front panel can be completed with stranded flexible hook-up wire.

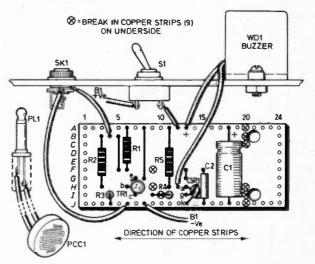


Fig. 2. The layout of the components on the topside of the stripboard

# SOIL MOISTURE UNIT

[] \( \text{O} \text{

This device has partly a novelty value, but also definitely does give some indication of when the soil that it is measuring is "wet" or "dry". It may therefore help to give more consistent and successful results, assisting those who don't have much luck with potted plants.

# CIRCUIT DESCRIPTION

Most of the work is done by ICl, a cheap and

readily available 741C op-amp.

The variable resistor VR1, is wired between the supply lines and its wiper is connected to the inverting input. The setting of VR1 therefore determines the voltage at pin 2, and this can be altered from +9V to 0V.

At the non-inverting input we have the same sort of thing. The two probes, when inserted into soil, in effect form a resistor. The value of this "resistor" is dependent upon the moisture within the soil: the more moisture there is, the lower the value of this resistance.

Assuming that VR1 is at mid-position, when the soil is wet, we can say that the voltage at pin 3 will be lower than at pin 2. Therefore the output of IC1 is low. Current can therefore flow through R2 and D1, and "sink" into the output pin causing the green l.e.d. to light up. This is labelled WET.

Similarly with dry soil, the high resistance of the soil ensures that pin 3 is at a greater voltage than pin 2. The output pin therefore swings high, and it allows current to flow through the red l.e.d. D2 and R3 to 0V lighting up this l.e.d. This is labelled DRY.

# **ASSEMBLY**

A Bimbox type BIM2002/12 houses the unit. This handy-sized box measures 100×50×25mm.

# **COMPONENTS**

#### Resistors

R1  $5.6k\Omega$  R2  $470\Omega$  R3  $680\Omega$ 

All  $\frac{1}{4}$ W carbon  $\pm$  5%.

# Semiconductors

IC1 741C 8-pin d.i.l. operational amplifier

D1 TIL221 0.2 inch green l.e.d.

D2 TIL220 0.2 inch red l.e.d.

# Miscellaneous

VR1 10kΩ miniature horizontal skeleton preset

S1 single-pole push-to-make, release-tobreak

B1 9V type PP3

0.1 inch matrix stripboard; 10 strips by 24 holes; case,  $100 \times 50 \times 25$ mm, Bimbox BIM2002/12 or similar; battery connector; 4BA fittings, threaded brass rod for probes; 8 pin d.i.l. socket; connecting wire; mounting clips for D1 and D2.

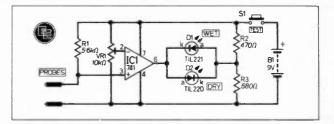


Fig. 1. Circuit diagram of the Soil Moisture Unit.

Any other plastic case can be used.

The two probes are made of 4BA threaded brass rod about 120mm long. Connections to the probes are made by solder tags placed under the mounting nuts within the case.

The two light-emitting diodes can be secured in position with either an appropriately-coloured lens-

clip or a standard plastic fixing clip.

#### SETTING UP

With construction completed, set VR1 to approximately midway, connect up a battery and press S1. The red l.e.d. should glow. Bend the two probes together at their tips so that they short together: the red lamp should extinguish and the green l.e.d. illuminate.

If this happens the unit is ready to use. Set VR1 to give the desired switchover point of the two indicators. Here it may prove useful if you have some small containers of soil available. The individual samples should have various levels of water content, ranging from dry to saturated. It should then be possible to eventually adjust VR1 until a desired sensitivity is obtained.

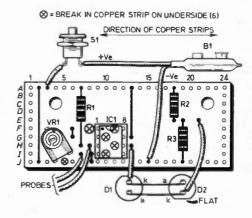


Fig. 2. Stripboard layout and interwiring diagram.

# ICE ALARM

I93I

This Ice Alarm warns the driver when possible conditions exist for the formation of black ice. It does this by monitoring the temperature outside the car. When this drops to about two or three degrees Celsius, slightly above freezing, the unit flashes a warning lamp on the dashboard.

#### CIRCUIT

The heart of the unit is ICl, a cmos multivibrator which has been wired up as a gated astable. This means that the device oscillates only when pin 5 is high, otherwise it is inoperative.

Along with R1 and VR1, the thermistor forms a potential divider, the output of which is connected to the base of TR1.

If TR1 is on, pin 5 is high and so the i.c. oscillates freely. If the transistor is off, however, the reset pin is grounded through R2 and so the i.c. is disabled.

As the temperature of RTH1 decreases towards 0°C, its resistance will increase and the voltage at TR1 base will be reduced. Eventually a point is reached where the base terminal is 1.2V less than the emitter and so TR1 must turn on. Pin 5 of IC1 goes high, permitting it to oscillate normally; pin 11 then presents a square wave signal to TR2 and this causes the indicator lamp to flash.

Note that the lamp is normally fully alight to show that the Ice Alarm is on, but it flashes when RTH1 detects a low temperature.

Connections for the power feed and thermistor are taken by flying leads from the stripboard, through the case to a four-way screw terminal block mounted outside the case.

# INSTALLATION

Ascertain whether the car chassis is positive or negative earth and connect this to the positive or negative terminal on the terminal block. The other supply wire should come from an ignition-controlled circuit (possibly at the fusebox), so that the Ice Alarm is not inadvertently left switched on when the ignition is switched off.

The position of the thermistor module may be rather a trial and error affair. The unit is obviously not waterproof and so it must not be exposed to spray or road filth. Furthermore it needs to be placed away from the car's exhaust system and cooling

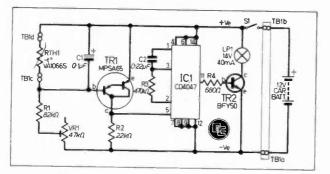


Fig. 1. Full circuit diagram of the Ice Alarm.

# **COMPONENTS**

## Resistors

R1 82k $\Omega$  R3 470k $\Omega$  R4 680 $\Omega$ R2 22k $\Omega$  All  $\frac{1}{4}$ W carbon 10%

#### Capacitors

C1 0·1μF tantalum bead 35V C2 0·22μF polyester C280

# Semiconductors

IC1 CD4047 CMOS mono/astable multivibrator

TR1 MPSA65 pnp silicon Darlington

TR2 BFY50 npn silicon

# Miscellaneous

VR1  $47k\Omega$  miniature horizontal preset potentiometer

RTH1 VA1066S negative coefficient rod thermistor

LP1 14V 40mA integral type MA lamp (amber)

S1 push-on, push-off single pole switch Case,  $110 \times 60 \times 30$ mm, Bimbox type 2003/13 or similar; stripboard,  $0\cdot 1$  inch pitch, 10 strips  $\times$  24 holes; 14 pin d.i.l. socket for IC1; fourway screw terminal block; piece of tagstrip for mounting RTH1; twin core flex, mounting hardware for circuit board; lamp holder for LP1.

system—parts which get hot during normal opera-

Final positioning must vary from car to car. A suggestion is behind (i.e. inside) the front bumper.

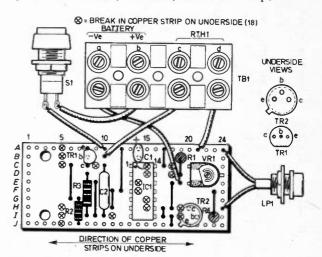
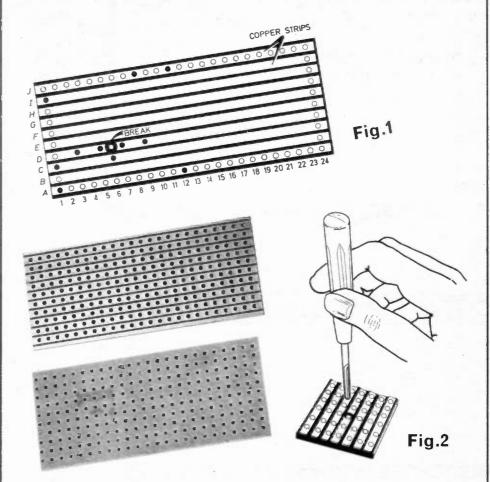


Fig. 2. Circuit board layout and component inter-wiring. Note that the thermistor RTH1 is not shown in this diagram but is connected to the terminal block via a long cable and located at some remote location.



# CONSTRUCTION GUIDE

- 1 Any one of these **Popular Designs** can be built on a piece of stripboard (Veroboard) measuring 1½ in by 2¾ in, or 10 strips by 24 holes, such as the piece of board given free with this issue. Further boards of this size can be obtained from retailers advertising in this magazine.
- 2 Before attempting to build one of these **Popular Designs**, read carefully the article and study the circuit diagram and the component layout diagram. Note the method of coding the strips (letters) and holes (numbers) of the stripboard. This helps positive identification of connection points on the board. See Fig. 1.
- 3 Make any required breaks in the copper strips as indicated in the diagrams. This operation is performed using a twist drill (hand held); or with the aid of a special tool which can be purchased. See Fig. 2.



- 4 Solder in position the resistors, capacitors and semiconductors. Carefully check the identity of the semiconductor lead out wires, also the polarity of electrolytic capacitors, before placing in position. It is advisable to use a heatsink when soldering semiconductor leads. I.C. sockets are recommended for the mounting of these devices.
- 5 Any other components such as potentiometers and switches that are mounted externally to the stripboard should be connected to the appropriate points on the stripboard as indicated. Use insulated (preferably flexible) wire of sufficient length to suit the mounting arrangements and general assembly of the complete unit.
- 6 Note any special recommendations regarding the housing of the project, and the fixing arrangements for the stripboard within the case. Where a plastic (non-conductive) case is specified do NOT use a metal one.

# **Everyday ELECTRONICS**

# SIMPLE INFRA RED REMOTE CONTROL

Pulsed beam gives high efficiency and strong peak output using a single l.e.d. Receiver uses a special i.r. photodiode that does not respond to visible light. The receiver energises a relay which can be used for switching electrically operated apparatus. The i.r. remote control could also be applied as a burglar alarm.

# TRANSISTOR IGNITION

This solid state replacement for the conventional car ignition system will improve m.p.g., extend spark plug life, minimise points wear and reduce demands on starter motor and battery.

# LOUDHAILER

A self-contained electronic megaphone unit built around an LM380N audio i.c. and designed to feed either a small conventional loudspeaker or a horn speaker. The output power of 1W is adequate for many applications and enables the unit to operate for long periods from small batteries.

NOVEMBER 1981 ISSUE ON SALE FRIDAY, OCTOBER 16

# **EXPERIMENTAL CRYSTAL SET**

A fascinating starter project for beginners of all ages. Easy to build and inexpensive in parts. Home-made coils cover medium and short waves.

# Everyday News

# ELECTRONICS AND MICROELECTRONICS

RADIO ELECTRONICS FOR SCHOOLS

With a new school term just starting it's a bumper month for the hard pressed physics teacher who, due to lack of qualified electronics teachers, is the one person invariably "press-ganged" into running the school's electronics classes.

Not to worry! Hard on the heels of our new beginners Teach-In 82 series comes news that, under the School Radio banner, the BBC Radio 4 will be starting on September 22 a sound plus vision series for the teaching of electronics in secondary schools, including practical work. With our new series and the radio broadcasts, this should keep the pupils happy and relieve some of the pressure from the teacher.

The arrival of this more ambitious project from BBC School Radio entitled Electronics and Microelectronics is aimed at the 14 to 16 year olds and is designed to introduce some of the developments which have taken place in the last ten years.

The series consists of ten programmes and are accompanied by five Radiovision film strips. Also complete pupils' kits of component parts incorporating three transistors, two integrated circuits and a special softwood base (which take woodscrews for non-soldering

experiments) are available. It is claimed that each kit is suitable for 3 or 4 pupils and costs £7.95.

Full details of how to obtain the kit and the film-strips (which cost £5·70 each) are included in a 24-page Teacher's Notes booklet. These notes which contain master copies of pupils' worksheets, are available Free from "Electronics and Microelectronics" BBC School Radio, 1 Portland Place, London W1A 1AA, on receipt of a A4 self-addressed envelope stamped at 20p.

In collaboration with this BBC School Radio project



Electronics and Microelectronics is produced by Arthur Vialls. The series can be heard on Radio 4 VHF on Tuesdays, at 2.20-2.40 p.m., starting September 22.



tical circuits.

Wetherby,

LS23 7EH.

The book, price £2.75, is

West Yorkshire

available from BP Educational Service, PO Box 9,

At the end of September '81 Maplin are taking the Atari personal computers and their new Matinee Organ to five cities in the UK. This is a golden opportunity for mail order customers or anyone for that matter to actually see and handle these popular products and ask questions on the spot.

A warm welcome (between 6 pm and 10 pm) awaits anyone wishing to enjoy a pleasant informal evening which is completely free, so make a note of the following dates in your diary:

Newcastle upon Tyne, Friday, September 25, at the Grainger Room, Newcastle Centre Hotel, New Bridge Street, Newcastle upon Tyne.

Edinburgh, Saturday, September 26, at the Rosebery Room, Grosvenor Centre Hotel, Grosvenor Street, Edinburgh.

Manchester, Sunday, September 27, at the Ullswater Room, Portland Hotel, 3 Portland Street, Piccadilly Gardens, Manchester.

Birmingham, Monday, September 28, at the Malvern Suite, Birmingham Centre Hotel, New Street, Birmingham.

Norwich, Tuesday, September 29, at the Riverside Suite, Hotel Nelson, Prince of Wales Road, Norwich.





**Talking Memories** 

A radio receiver and a television set which took the art of speech synthesis and recognition a stage further was demonstrated by Toshiba at a recent trade show. Although the prototypes are not in themselves to be taken too seriously they provide a useful pointer to the future.

The radio receiver is a fairly conventional hi fi unit, but it can be operated under volce control. Although interesting this, in itself, isn't new. Last year Toshiba was showing a television set that operated under voice control. You tell it to switch on, switch off and change channels. The new receiver is important because it can memorise, and mimic, up to ten phrases, each of two seconds in length.

To "teach" the receiver to do what you want, you speak command phrases into a microphone while microprocessor and memory chip store a digital coding of the words and dialect. Each stored phrase is then delegated to a switched

function of the receiver.

For instance if you speak the phrase "BBC Radio 4", the microprocessor and memory store the sound of your voice saying those words. If you delegate that stored command to switch to the frequency for BBC Radio 4, then whenever you subsequently shout "BBC Radio 4" at the receiver it will recognise the command and switch to the appropriate frequency.

The breakthrough is in storage capacity. Ten phrases of two seconds each is a 20 second speech store. And as well as recognising the memorised phrases, the receiver can also speak them out as a

check on what is memorised.

This points to the imminence of much larger solid state speech stores, programmable by the user. Imagine for instance a telephone answering machine on which the master message is recorded on a solid state memory, rather than a clumsy endless tape loop.

# Talking TV

The other Toshiba development is a TV set which talks to the viewer. This is a digitally stored voice which is pre-programmed to speak appropriate phrases at

user-selected times. For instance a timer will switch the TV on in the morning with an alarm call and a spoken "Good morning" at any chosen time.

At night the set will switch itself off at a pre-appointed hour with a "Now I'll fade out—have a goodnight's sleep". If you turn the volume up too loud the solid state voice (in the prototype a very female voice) advises you to "remember the neighbours, lower the volume".

An ultrasonic ranging device (not photoelectric as wrongly reported elsewhere) senses if anyone in the room is too close to the set. The voice then warns "Watch from a distance, for your eyes'

sake".

The ultrasonic sensor also detects when everyone has either left the room, or moved out of range of the set. "Now I'll fade out", says the synthesised voice and the set switches itself off.

# Chatter Box

It now looks as if we shall get legal CB radio on 27MHz f.m. any day now.

If recent experience in both Germany and Austria is anything to go by, it will rapidly become a craze, but a short-lived craze. The sad truth is that CB—even on f.m.—only "works" while the system is illegal and its use is restricted to those who are prepared to risk breaking the law. As soon as the system becomes legal, the electronics industry floods the market with cheap sets and these are heavily advertised. The number of CB users increases dramatically and very soon no one can get a word in edgeways over the air.

Journalists from Germany and Austria have told me that the airwaves there are now so cluttered with CB shouting that useful communication is impossible.

# Pacemaker

When King Khaled visited London in the early summer of 1981, there was much publicity given for his heart ailment and the need for constant communication with his private surgeon in the USA. The word went round Fleet Street that King Khaled wore a miniature transmitter which provided a direct link with a machine in an American hospital continually monitoring his heart beat.

It was rumoured that King Khaled's personal transmitter beamed signals up to a satellite, from where they were beamed down again to the hospital. "Money can buy anything these days", said the political correspondents, "Nonsense" said the science correspondents," satellite transmitters are too big to hide in your clothes."

We'll probably never know whether King Khaled was or was not hooked up to a heart monitoring machine in America, but it was certainly technically possible, despite what the science correspondents told their editors. The clue is to be found in an erudite article published in *The Radio and Electronic Engineer*, official journal of the IERE (Institution of Electronic and Radio Engineers). The article, titled "Manpack Satellite Communications Earth Station" was by coincidence published at almost exactly the same time as King Khaled visited the UK.

# Manpack Station

To cut a long, technical story short the Royal Signals and Radar Establishment in Worcestershire has now built a prototype ground station for satellite communication which can be carried as a military back-pack. Normally satellite linked gound stations are either permanent, with large dish aerials secured to the ground, or are mounted on a lorry. This is because the dish for such a ground station is around 2 metres in diameter and the transmitter is of between 30 and 60 watts power needing either a mains or generator supply to drive it.

The RSRE manpack station is just 45cm  $\times$  45cm  $\times$  20cm in size, weight 17 kilos and draws only 30 watts of power from a battery. The aerial dish is built into the case and measures only 45cm in diameter. The system transmits either low quality analogue speech or a 50 baud telegraph signal.

Although the unit is small enough for one man to carry around on his back, it isn't used as a mobile back-pack. To transmit, the unit is dumped onto the ground and aligned with a sky satellite using its built-in compass and signal strength meter.

It would have been perfectly possible for for King Khaled to wear a radio microphone, of the type now routinely used by performers in TV studios, strapped close to his heart. This radio mic would transmit heart beat signals over a short distance to a back pack satellite earth station in constant contact with the American heart hospital.

As I said we've no idea whether or not King Khaled actually used such a system, but it was certainly dangerous for Fleet Street science correspondents to dismiss the idea out of hand. But then, my experience of Fleet Street science correspondents is that some of them are remarkably naive about up-to-date science.

# RADIO WORLD

# Armchair travellers

Whether or not one believes wholeheartedly in the modern cult of amateur radio "DXpeditions"—the setting up of stations on islands from which there is little or no regular operation-there is no doubt that two-way contacts, even brief ones, with out-of-the-way places still provides an operator with a real sense of achievement, no matter how illogical this may be. Then again, compact h.f. transceivers allow holiday-makers and tourists to set up stations from hotel rooms in, for example, the smaller European countries such as Monaco, Liechtenstein, Andorra, Aland Islands and the like-and then be sure of plenty of contacts.

In Europe there are indeed very few places from which amateur operation does not take place; only Albania remains reluctant to permit either the indigenous population or tourists to use amateur radio "rigs" in that very private country.

The poor early summer this year seems to have made holiday operation more popular than ever. My own log shows that Liechenstein is proving a popular venue for German amateurs: a Dutch amateur

# By Pat Hawker, G3VA

with one of the small 10 watt transceivers has been telling me of the pleasures (and the mosquitoes) of the Northern Italian lakes while sitting on the terrace outside his holiday villa and a Greek amateur, normally located in Athens, has been praising the sunny beaches of Rhodes.

Several Finnish amateurs have crossed over to Aland Islands in the Baltic and, as usual, I have been receiving strong signals from the Royal Navy Amateur Radio Society's station on HMS Belfast In the Pool of London with the special callsign GB4RW to mark the Royal Wedding.

# Ascension Island

The recent contact that gave me the most pleasure was with Dave, ZD8DM on Ascension Island, a thousand miles off the coast of Africa in the Southern Atlantic. A tiny island that had always previously eluded my calls, although back in 1967 I was lucky enough to make a fascinating 8000 mile trip to Ascension on what was possibly the one and only press visit in the island's history.

This was at the time of the opening of the first Cable & Wireless 42ft satellite

population but it supports changing communities of Britons, Americans and contract-workers from "neighbouring" St Helena, some 800 miles away. The island, in fact, is a remarkable mixture of the old and the new. Space and

communications terminal bullt on the island by the Marconi Company as part of the NASA Apollo network for the moon

landings. Small though it is, Ascension supports more aerials to the acre than

Ascension, a tiny volcanic dot in the vast South Atlantic, has no indigenous

almost anywhere else on Earth.

missile tracking stations (at that time it was near the end of the main American missile testing range), a large BBC overseas relay base with numbers of 250kW h.f. transmitters, a long-established Cable & Wireless telegraph cable station linking South America to Europe (and more recently on the route of the South African telephone cable) with a mass of glittering brass telegraph instruments, local medium-wave broadcasting stations and other futuristic electronics co-exist with vast colonies of wideawake terns and enormous green turtles.

While on the island I spoke over the satellite circuit to an American amateur in the NASA base In Maine-and many of the BBC and C&W staff who spend just a few years on this remote island are attracted there by the chance to operate with the rare ZD8 prefix. The American community also often includes a few

radio amateurs.

# Television Makes News

This year's NAB and Montreaux television equipment exhibitions foreshadowed further significant developments in the field of electronic news gathering (the use of electronic rather than film cameras in news reporting). At both exhibitions were shown the first integrated colourcamera/video-recorder equipments, including a prototype unit based on a

single-tube colour camera. The "Hawkeye" unit unit developed and Matsushita jointly by RCA and Matsushita ("Panasonic") uses 12-inch tape cassettes similar in size to the VHS-format domestic VTRs and providing 20 minutes of broadcast recording time on each cassette. The single-tube Sony prototype unit is based on a new high-speed Saticon Trinicon form of pick-up tube and promises to eliminate many of the problems inherent in the accurate registration of the usual colour cameras which have three pick-up tubes; it should also reduce battery consumption which is still a problem with many of the lightweight cameras taking 25 to 35 watts.

With integrated camera/recorders, the electronic news man could be virtually as "action ready" when he arrives on the scene as the traditional film cameraman with no interconnection plugs, sockets and cables to worry about. But it may be some time yet before we see such equipment suitable for use in the European 625line system rather than the American 525-line standard.

Trip Wires

One problem for television broadcasters that seems to be on the increase in Europe is that of aircraft flying into the high television transmitting masts or the guy wires which surround them.

A 1000ft mast in Luxembourg was snapped in two recently with fatalities both in the plane and among the technical staff at the transmitting station when an aircraft flew into it. In the UK a light aircraft from France recently crashed with fatal consequences when it hit the guy wires of the Dover television mast.

A few years ago an American fighter aircraft hit one of the guy wires of the Caldbeck television mast in Cumbria and sliced off part of its wing. The military plane must have been very sturdily built, the pilot managed to fly it a couple of hundred miles back to its base.

Poor Horace

I suppose that in some ways it was rather sad that Horace, the BBC's sub-titling computer found it so diffi-cult to cope "live" with the Royal Wedding commentary—but it added a touch of light relief and Horace deserves an Equity card as the year's best up-and-coming comedian.

It was a worthy try and I am sure that there must have been many hard-of-hearing viewers who benefited from the open sub-titling and who found Horace not only funny but also of tremendous help.

Defence and H.F. Radio

It is becoming increasing clear that there has been a real revival of interest in h.f. radio communications systems among British and NATO Defence planners. According to a recent article by C. R. M. Noonan in the Marconi journal "Communication & Broadcasting", the NATO countries are currently planning to spend over £100 million on such long-established communications techniques as relatively low-speed radio-teleprinters, unprocessed analogue speech (ssb) and hand Morse.

It is recognised that this picture does not conform to the now accepted image of modern radio communications and may even seem "old hat" when compared to wideband, high-speed data and multichannel digitally-encrypted links for speech. But h.f. is seen as having a number of special, almost unique, advantages while there is a growing belief that communications based on space satellites are vulnerable to enemy action. It is also accepted that with hand Morse operating one can continue to pass traffic in conditions which would defeat other systems.

There is also, it would appear, an increasing concern with the possible effects of the electromagnetic pulse (e.m.p.) that would follow a nuclear explosion in the upper atmosphere. It has been suggested that a single explosion, about 300 miles above the Earth's surface, could put out of action solidstate radio and electronic control equipment (including computers) over much of a continent unless more effective ways of "hardening" equipment are adopted.

For this reason we may see a revival not only of h.f. but also of miniature valves in place of the much more e.m.p.-vulnerable solid state devices. There is some evidence that this course is already being followed in the USSR.

# The WORLD of RADIO & LECTRONICS CONCISE PARTS CATALOGUE

# **OUT NOW!**

# AMBIT'S NEW CATALOGUE

In an attempt to collate and organize our burgeoning ranges of 'stock' components (now over 7000 line items), we have at last produced a 'concise' 80 page parts catalogue to supplement the popular 'Tecknowledgey' series of 'wordy' applications catalogues, which now lists a wide range of basic components - as well as our unique RF and Communications components. The World of Radio and Electronics contains everything the informed electronics user needs, at prices which we guarantee will match the lowest on the market for equivalent product.

Prices appear on the page alongside the part numbers, and the catalogue is now updated quarterly - available either direct from here (60p all inc) or at most newsagents and bookstalls where you can find electronics publications. So as well as all the 'run-of-mill' items like resistor, capacitors, hardware, solder etc - you now have the first genuinely complete parts source for the radio, communication, electronics, computer user.

Ambit International 200, North Service Road, Brentwood, Essex CM14 4SG



# INTRODUCTION TO

PART 6 BY J. CROWTHER

# THE "AND" GATE

The AND gate is equivalent to relays (switches) in series, see Fig. 6.1.

An output will be obtained if A AND B are energised, that is if A AND B are at logic 1.

# **Boolean Equation**

AB=S for two inputs ABC=S for three inputs

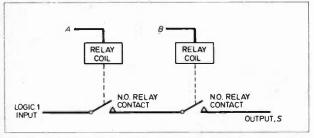


Fig. 6.1. A 2-input AND gate made up using relays.

#### Symbols

A number of different symbols are currently in use for representing logic gates; Fig. 6.2. shows most commonly found AND symbols, Types shown in (a) in each "symbol" diagram are those used in E.E.

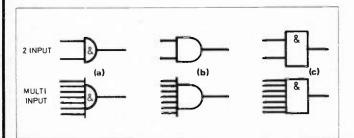


Fig. 6.2. Three different ways of representing AND gates in a circuit.

# Truth Tables

In order to see at a glance what combination of inputs give an output, Truth Tables for a particular gate or circuit are constructed.

# Truth Table for a Two Input "AND" Gate

in	outs	output
A	В	AB=S
0	0	0
0	1	0
1	0	0
1	1	1

The more inputs a gate has, the more complicated the truth table becomes. For example, if a gate has ten inputs the number of possible combinations of the inputs will be 1,024. In order to be certain that none of the possible combinations have been omitted the following procedure is useful:

The number of possible combinations is given by the formula: Number of combinations =  $2^n$  where n is the number of inputs.

Once the number of combinations has been established, allow this number of lines, and number them in binary, starting at 0.

#### example

To construct a Truth Table for a three input "AND" gate. Since there are three inputs the number of combinations are:  $2^3=8$ 

Now draw eight rows and number them in binary, from 0 to 7 as shown.

	Inp A	uts <i>B</i>	С	Outputs ABC=S
0	0	0	0	0
1	0	0	1	0
2	0	1	0	0
3	0	1	1	0
4	1	0	0	0
5	1	1	0	0
5	1	1	0	0
7	1	1	1	1

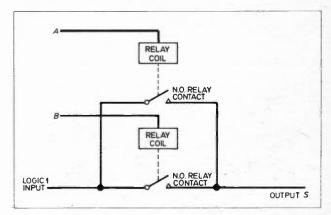


Fig. 6.3. Making up a 2-input OR gate from two relays connected in parallel.

# THE "OR" GATE

The OR gate is equivalent to relays (switches), in parallel. See Fig. 6.3.

# **Boolean Equation**

A+B=S For two inputs A+B+C=S For three inputs

# Symbols

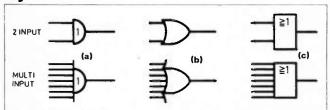


Fig. 6.4. Three different ways of representing OR gates in a circuit.

# Truth Tables

**Two Inputs** 

Ing	uts	Output
A	В	A+B=S
0	0	0
0	1	1
1	0	1
1	1	1

T	hree	Inputs

Inputs			Output
A	В	C	A+B+C=S
0	0	0	0
0	0	1	1
0	1	0	1
0	1	1	1
1	0	0	1
1	0	1	1
1	1	0	1
1	1	1	1

# THE "NOT" OR "INVERTER" GATE

The NOT OF INVERTER gate is the special case of one input only, and it turns logic 0 into logic 1 and vice-versa.

Therefore it turns A into  $\overline{A}$ , and  $\overline{A}$  into A.

In other words we get an output if there is no input, so it is equivalent to a normally closed relay or switch.

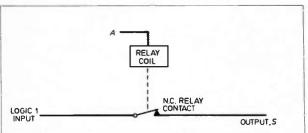


Fig. 6.5. Shows how a relay may be used to realise a NOT (or INVERT) function.

# Truth Table

Inputs A	Output A=S	
0	1	
1	0	

**Boolean Equation** 

 $\vec{A} = S$ 

# **Symbols**

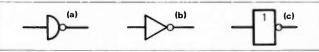


Fig. 6.6. Some symbols seen in use for representing a NOT gate in a circuit.

# THE "NAND" GATE

NAND means "NOT AND", that is there is no output if AND B are both at logic 1.

# **Boolean Equation**

Since AB means A and B, and a bar means not, it follows that:

AB means NOT A AND B.

The Boolean equations for a "NAND" gate are:

 $\overline{AB} = S$  for two inputs  $\overline{ABC} = S$  for three inputs

If we apply Demorgan's Theorem to  $\overrightarrow{AB}$  we get  $\overrightarrow{A} + \overrightarrow{B} = S$ This represents two normally closed switches or relays in parallel. Therefore a NAND gate is equivalent to Fig. 6.7.

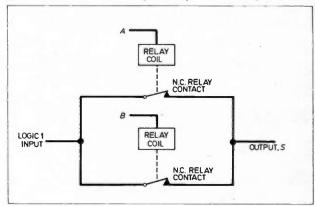


Fig. 6.7. A 2-input NAND function may be realised by using two relays (with normally closed contacts) in parallel.

# **Truth Tables**

Two Inputs

	1 WO Imputs					
Inp A	outs B	Output AB=S				
0	0	1				
0	1	1				
.1	0	1				
1	1	0				

Three Innuts

	Turce inputs						
A	nput <i>B</i>	s C	Output ABC=S				
0	0	0	1				
0	0	1	1				
0	1	0	1 1				
0	- 1	1	1				
1	0	0	1				
1	0	1	1				
1	1	0	1				
1	- 1	1	0				

By comparing the "NAND" truth table with that for the "AND" gate we will see that it is the inverse, in other words it is an "AND" gate followed by a "NOT" gate, and the symbols show this.

# Symbols

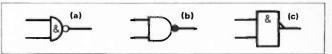
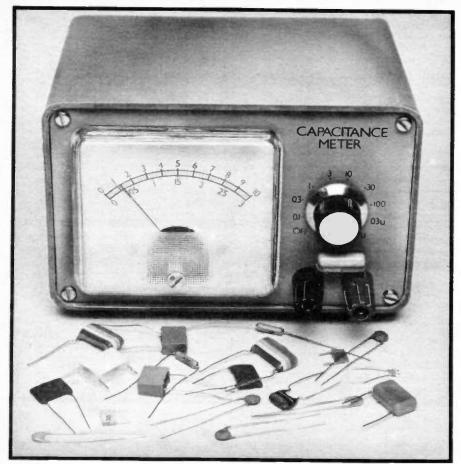


Fig. 6.8. Common symbols for representing NAND gates.

# BE CONTINUE



# CAPACITANCE METER BY J.R.W.BARNES

T is inevitable that at some stage most constructors will be faced by an unmarked or suspect capacitor. How do you measure capacitance? Most multimeters don't have capacitance ranges, the traditional L.C.R. bridge is expensive and difficult to use. Surely a direct reading capacitance meter could be designed. As is the first step in any design, the desired specifications were chosen.

(1) The ability to measure most capacitors encountered in electronics excluding electrolytics.

Useful measurements can be made with the prototype of capacitors between 50pF and  $1\mu$ F.

(2) A linear scale is very desirable.

(3) An accuracy of around 3 to 5 per cent can be obtained depending on the calibration standard (see later).

(4) The decision to make the unit mains powered was taken because, unlike a multimeter, a capacitance meter has few uses outside a workshop.

# **CIRCUIT**

The full circuit of the Capacitance Meter is shown in Fig. 1.

To ease explanation and promote understanding the circuit can be split into four sections: pulse generator, monostable multivibrator, meter circuit and power supply.

The pulse generator is based on a unijunction relaxation oscillator. Cl charges through R10 and VR1 until the voltage across C1 equals the supply voltage multiplied by the intrinsic stand-off ratio of the unijunction, approximately equal to 7V. The capacitor is then discharged through the emitter/base 1 junction of the unijunction transistor TR1 resulting in positive pulse across R12. This turn TR2 on, reducing the voltage at pin 2 of IC1 below 13 Vcc, and so triggering the monostable. VR1 is the calibration control; it controls the frequency with which the monostable is triggered.

# TIMING RESISTOR

The widely used 555 i.c. was selected for the monostable primarily for its ability to handle wider variation of components in its timing circuit. Upon application of a trigger pulse, the unknown capacitor connected to SK1 and SK2 is charged through the timing resistor (one from R1 to R9 as selected by the range switch S1a). When the voltage across the capacitor reaches  $^2$ <sub>3</sub>  $V_{\rm CC}$  the capacitor is discharged. For the duration of the timing cycle the output pin 3 on IC1 goes high, this period being equal to  $1\cdot1$  R<sub>t</sub>C (where R<sub>t</sub> is one from R1 to R9).

When the output is high D1 conducts allowing C3 to charge through R15. The voltage across C3 is indicated on the meter ME1. The deflection

# **COMPONENTS**

1.	C212[(	715		
	R1	$10M\Omega$	R6	$33k\Omega$
	R2	$3.3M\Omega$	R7	10kΩ
	R3	$1M\Omega$	R8	$3.3k\Omega$
	R4	330kΩ	R9	1kΩ
	R5	100kΩ		
	All 1	W metal ox	ide ± 2%	
	(e.g.	Electrosil to	ype TR5)	
	R10	10kΩ		
	R11	$220\Omega$	See	
	R12	$220\Omega$	Ch	00
	R13	2 · 2kΩ	311	UU
	R14	2 · 2kΩ		
	R15	10kΩ		
	All 1	W carbon	р	age 670

#### Potentiometer

VR1 1MΩ in square multiturn cermet trimmer

# Capacitors

Recietore

apac	tors
±21	10nF polyester, axial
C2	100nF ceramic disc
C3	100µF 16V elect., axial
C4	100µF 25V elect., axial
C5	10nF ceramic disc

# Semiconductors

D1-3 1N4148 small signal silicon diode (3 off)
TR1 2N2646 unijunction

transistor
TR2 2N3904 silicon npn

IC1 NE555 timer i.c. IC2 μA7812 regulator 12V 1A

# Miscellaneous

ME1 meter 500μ A (SEW type MR65P)

SK1 insulated screw terminal, black

SK2 insulated screw terminal, red

S1 miniature Maka switch assembly and two 12-way 1-pole wafers

T1 mains transformer: 12-012V 100m A secondary
Stripboard 32 holes by 23 strips
0-1in matrix. Case 100 × 100 ×
150mm. Miniature tag board 9 way.
6B A nuts (8 off) and bolts (4 off).
Veropins. Three core mains lead.
Knob, skirted with index mark.
Grommet for mains lead.

Approx. cost £16 excluding Guidance only

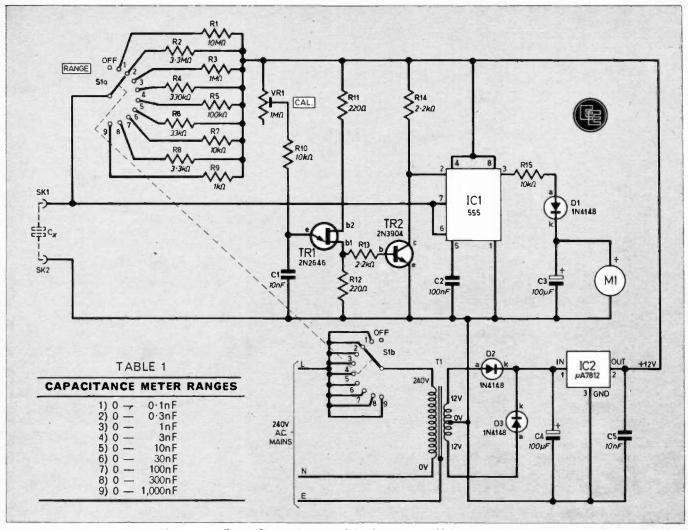


Fig. 1. Circuit diagram of the Capacitance Meter.

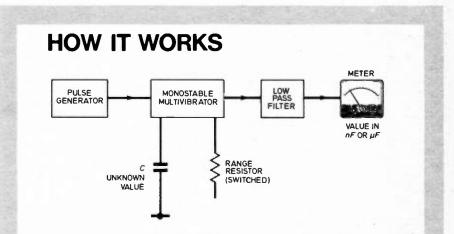
of the meter is proportional to the value of the unknown capacitor.

The power supply consists of T1, D2, D3 and C4 providing raw d.c. and IC2 providing a regulated 12V.

# starts here

# LAYOUT IMPORTANT

Begin construction with the stripboard; the layout given was carefully designed to minimise stray capacitance, and should be strictly adhered to. The stray capacitance shows up as residual reading of around 30pF on the lowest two ranges. This however poses no problems, it is simply subtracted from the measured reading.



The unknown capacitor along with the resistor selected by the range switch form the timing elements in a monostable multivibrator, the length of the output pulse being proportional to the unknown capacitor. The monostable is triggered at regular intervals by the pulse generator built around the unijunction transistor. The output of the monostable is fed to a low pass filter to produce a voltage proportional to the unknown capacitor. This voltage is displayed on the meter.

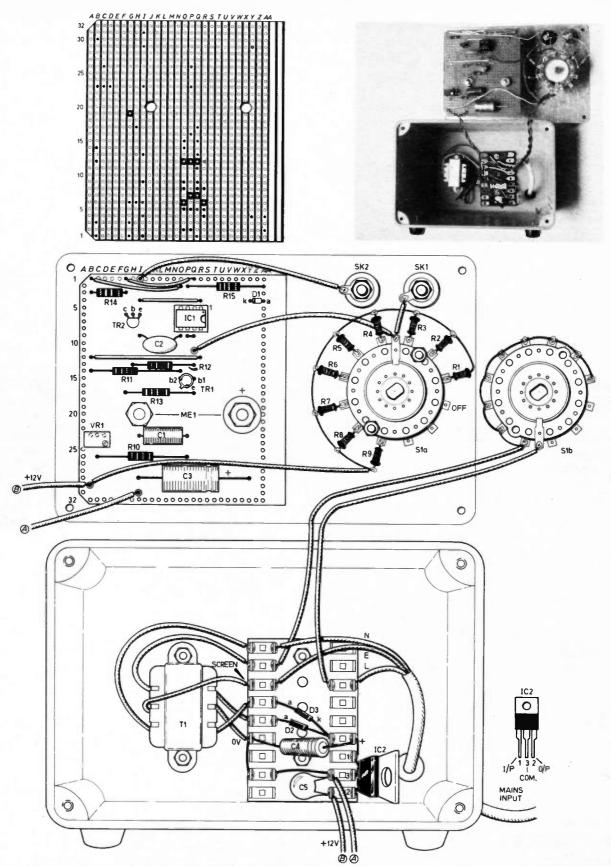


Fig. 2. Constructional details of the Capacitance Meter. The front panel has been lifted clear of the case to expose all components and wiring. At the top is an underside view of the stripboard with details of breaks in copper strips and drilling for meter terminals.



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The normal practice of cutting the board to size, drilling the large holes for the meter terminals and breaking the strips, should be followed. Then insert the Veropins, i.c. socket and wire links. The resistors and capacitors are then added. Finally the diodes and transistors can be soldered in place. Take care to ensure the diode, integrated circuit and the electrolytic capacitor are the right way round. A drawing of the stripboard is shown in Fig. 2.

The prototype used a Sew panel meter type MR65P. If a different meter is used the position of the mounting holes may need adjusting; simply move the mounting hole and the positive ends of both C3 and D1

accordingly.

Mount the range switch, two screw terminals and the meter onto the front panel.

# WIRING THE SWITCH

The wiring details of the range switch are shown in Fig. 2. Care should be taken to avoid over heating the range resistors R1-R9 as this will change their values and subsequently effect the calibration.

Mount the main circuit board onto rear of meter and secure with

nuts on the terminal screws.

The power supply was built on a small piece of tag board as shown in Fig. 2. Construction is straight forward, just take care to observe the polarity of the diodes and the electrolytic capacitor.

Solder the mains lead to the power supply board then feed this lead through the grommet hole at the rear

of the case.

Complete all the interwiring between the range switch, test terminals, main circuit board and power supply board as shown in Fig. 2.

The tag board is mounted together with the transformer on the rear

panel of the case, with 6BA nuts and bolts.

Place the front panel assembly in position and secure to the case with four screws.

# FRONT PANEL

The large hole for the meter is best made by drilling a series of smaller holes and filing the resultant hole. Alternatively an "Abrafile" may be

#### CALIBRATION

In order to calibrate the instrument a "standard" capacitor is required. It is best to use a standard of either  $0.01\mu F$  or  $0.1\mu F$  preferably 1 per cent tolerance. Simply connect the capacitor using short leads to the terminals, select the appropriate range and adjust VR1 for full scale deflection. This single process calibrates all the ranges.

The meter scale should be linearly divided into ten, with five sub-divisions between each. The main divisions should be marked 0 to 10 at the top of the scale. The bottom edge of the scale should be marked at the appropriate points "0", "0.5", "1", "1.5", "2", "2.5" and "3".

This will facilitate readings to be made on all the nine ranges. See Table 1 for range coverage. The values have been specified in nanofarads throughout, though in practise μFs would be often used, certainly for the larger values. Table 2 provides at a glance conversion from nanofarads and picofarads to microfarads.

# USING THE CAPACITANCE METER

When measuring capacitors take them out of the circuit by disconnecting one lead. This is most important because any parallel components

will effect the reading. It is also advisable to discharge any capacitor before measurement as any stored charge could damage the instrument.

TABLE 2

# CAPACITANCE UNIT CONVERSION Each line indicates equivalents

nanofarad (10-9F)	picofarad (10-12F)	microfarac (10-6F)
0·1nF	100pF	0.0001#E
0.3nF	300pF	0.0003µF
1nF	1,000pF	0.001#F
3nF	3,000pF	0.003µF
10nF	10,000pF	0-01µF
30nF	30,000pF	0.03µF
100nF	100,000pF	0-1µF
300n F	300,000pF	0.3µF
1,000nF	1,000,000pF	1µF

# OTHER APPLICATION-**BROKEN CABLE**

An ohmmeter will tell you if a cable is broken, a capacitance meter will tell you where.

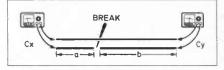


Fig. 3. Locating the break in a cable.

Measure the capacitance between the two wires or the core and the screen at both ends. The ratio of the two capacitances is also the ratio of the length from the break. In the

 $C_X$  a diagram Fig. 6 - = -

П

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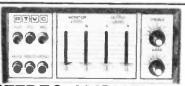
# **BY DOUG BAKER**







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# FOR BEGINNERS

This regular feature provides an easy guide to circuit components and materials and the techniques involved in building electronic projects.

This month we look at resistors.

These are the most commonplace of components.

Resistors fall into the following classes:

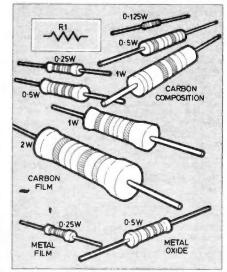
(a) fixed value

(b) variable (potentiometers).

# FIXED VALUE RESISTORS

Some fixed value resistors are wire wound. These are only used for certain special applications.

The most frequently used fixed value resistors are small tubular components with lead out wires emerging from either end. There are four principal types: carbon composition,



carbon film, metal film and metal oxide. All of these are available in a very wide range of resistance values—expressed in ohms  $(\Omega)$ —from 10 ohm to 1 megohm and even higher.

The generally available "preferred" values are 10, 12, 15, 18, 22, 27, 33, 39, 47, 56, 68, 82 ohms, and decadal multiples of these.

39, 390, 3,900, 39,000, 390,000, 3,900,000 ohms. These are expressed in more practical terms as  $39\Omega$ ,  $390\Omega$ ,  $3.9k\Omega$ ,  $39k\Omega$ ,  $390k\Omega$ ,  $3.9M\Omega$ .

It will be seen that k (kilo) means 103 and M (mega) means 106.

Fig. 1. Fixed value resistors. Shown here are carbon composition, carbon film and metal oxide resistors of typical wattage ratings. All are drawn to scale and are actual size. The circuit symbol for a fixed value resistor is also shown.

# RESISTOR COLOUR CODE

		BAND					
Band (ring) Colour	(1st Figure)	(2nd Figure)	3 (Multiplier)	(Tolerance %)			
BLACK	0	0	1-17	_			
BROWN	1	1	10	1			
RED	2	2	100	2			
ORANGE	3	3	1,000	3			
YELLOW	4	4	10,000	4			
GREEN	5	5	100,000	-			
BLUE	6	6	1,000,000	-			
VIOLET	7	•7	10,000,000	-			
GREY	8	8	10 0,000,000	-			
WHITE	9	9	1000,000,000	-			
GOLD			0-1	5			
SILVER			0.01	10			

Examples:							
1 BAND	COLOUR			② E	BAND	COLOUR	
1	RED	2			1	ORANGE	
2	VIOLET	7			2	ORANGE	
3	YELLOW	x 10,000	= 270,000 OHMS		3	RED	= 3·3kfL
			OR 270kΩ	ΔΙ	RSENCE	OF BAND	4 = ±20% TOLERANC
4	GOLD	±5%	±5% TOLERANCE		0021102	. 0. 0.	4-120 % TOLL TOLL

# COLOUR CODE

The resistance value is indicated by coloured bands on the body of the resistor. This colour code is explained in the illustration.

It is an excellent idea for beginners to obtain an assortment of resistors and practise working out the code.

# WATTAGE

The physical size of these fixed value resistors determines the power they can safely carry, without over heating with possible degradation of rated ohmic value. Power (volts × amps) is expressed in watts (W).

The forementioned values are available in the following wattage ratings. Carbon composition: 0.125W, 0.5W,

1W

Carbon film: 0.25W, 0.5W, 1W, 2W Metal film: 0.25W

Metal film: 0.25W Metal Oxide: 0.5W

# EAGER TO START?

If you are handy with a lightweight soldering iron, try one of our Popular Designs. Read the notes on page 682.

How to Solder will be the subject of next month's Square One.

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2527 Red relay panel—contains 2 × 6V reeds, 6 × 25030 or 25230, 6 × 400V rects + Rs. 59.

2529 Pack of ex-computer panels containing 14 series ICs. Lots of different gates and complex logic. All ICs are marked with typ no. or code for which an identification sheet is suppiled. 20 ICs £1.00; 100 ICs £4.00.

A504 Black case 50 × 50 × 78mm with octal base. PCB inside has 24V reed relay, 200 V A SCR, 4 × 5A 200V rects, etc. 69.

2535 RAM panel—36 2102A-4 static RAM's.

2535 RAM panel—36 2102A-4 static RAM's, also 28 other chips inc 7 × LS75, 4 × 74368,3 × 74180 etc. Only £5-00. 2536 As above, but extra 15 74LS chips £8-00.

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A standard keyboard version of the published Elektor 30-note chorus synthesiser with an amazing variety of sounds ranging from violin to cello and flute to clarinet amogst many others. Kit plus keyboard & contacts SET 100

#### FORMANT SYNTHESISER

#### P.E. MINISONIC SYNTHESISER

A very versatile 3-octave portable mains operated synthesiser, with 2 oscillators, voltage controlled filter, 2 envelope shapers, ring modulator, noise generator, mixer, power supply and aub-min loggle switches to select the functions. A case is excluded, but the text gives comprehensive constructional details. SET 38 £169-69

Kit plus keyboard & contacts

PRICES INCLUDE VAT @ 15% & U.K. P. & P.

# NEW KIT MAKE-UP -SEE BELOW

# 128-NOTE SEQUENCER

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Kit plus keyboard & contacts SET 76 £114.00

#### 16-NOTE SEQUENCER

Sequences of up to 16 notes long may be pre-programmed by the panel controls and fed into most voltage controlled synthesisers. The notes and rhythms may be changed whilst playing, making it more versatile than the name would suggest. Kit order code

# **DIGITAL REVERB UNIT**

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Compatible with the Formant and most other synthe sizers.

Kit order code SET 87 £11-69

# WAVEFORM CONVERTER

Converts saw-tooth waveform into sinewave, mark-space sawtooth, regular triangle, or squarewave with variable mark-space. Ideally one should be used with each synthesiser oscillator.

Kit order code SET 67 £28-13

# **BASIC COMPONENT SETS**

Include specially designed drilled & tinned fibreglass printed circuit boards, all necessary realistors, capacitors, semi-conductors, potentiometers, and transformers. They also include basic hardware such as knobs, sockets, switches, a nominal amount of wire and solder, a photocopy of the original published text, and unless otherwise stated, a robust aluminium box. Most parts may be bought separately. For fuller kit and component details see our currentlists.

Kits originate from projects published in PE, EE, and Elektor.

# RHYTHM GENERATORS

Two different kits—The control units are designed around the M252 and M253 rhythm-gen chips which produce pre-programmed switch-selectable rhythms driving 10 effects instrument generators feeding into a mixer.

SET 103-253 £64-19 SET 103-252 £57-26

#### -CHANNEL MIXER

A high specification stereo mixer with variable input impedances. Specs given in our lists. The kit excludes some SW's—see lists for selection. The extension gives two extra channels.

#### -CHANNEL STEREO MIXER

Full level control on left and right or each channel, and with master output control and headphone monitor.

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Kit order code SET 46 £34-56

# TUNING INDICATOR

A simple octave frequency comparitor for use with synthesisers where the full versatility of KIT46 is not needed.

Kit order code SET 69 £14-41

# **PULSE GENERATOR**

Produces controllable pulse widths from 100NS to 2Sec.
Variable frequency range of 0-1Hz to 100KHz.
Kit order code SET 115 £21-45

#### SIGNAL TRACER & GENERATOR

Allows audio signals to be injected into circuits under test, and for tracing their continuity, includes frequency & level SET 109 £15-31 Kit order code

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Provides sine, square and triangular wave outputs variable between 1Hz & 100KHz up to 10V P-P. SET 112 £21-58

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DLZ1000K
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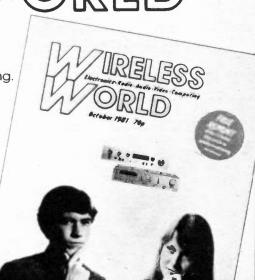
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1980 saw a genuine breakthrough – the Sinclair ZX80, world's first complete personal computer for under £100. Not surprisingly, over 50,000 were sold

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It uses the same micro-processor, but incorporates a new, more powerful 8K BASIC ROM – the 'trained intelligence' of the computer. This chip works in decimals, handles logs and trig, allows you to plot graphs, and builds up animated displays.

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# Built: £69.95

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# 16K-byte RAM pack for massive add-on memory.

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With the RAM pack, you can also run some of the more sophisticated ZX Software – the Business & Household management systems for example.

# SINCIDIC ZX81

6 Kings Parade, Cambridge, Cambs., CB2 1SN. Tel: (0276) 66104 & 21282. Designed exclusively for use with the ZX81 (and ZX80 with 8K BASIC ROM), the printer offers full alphanumerics and highly sophisticated graphics.

A special feature is COPY, which prints out exactly what is on the whole TV screen without the need for further intructions.

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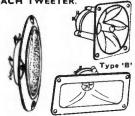
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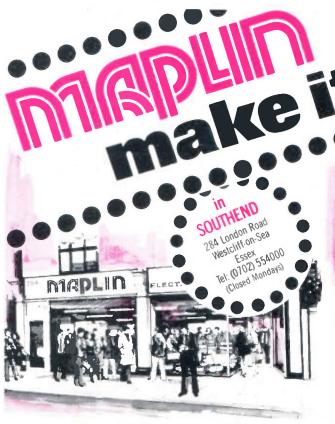
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