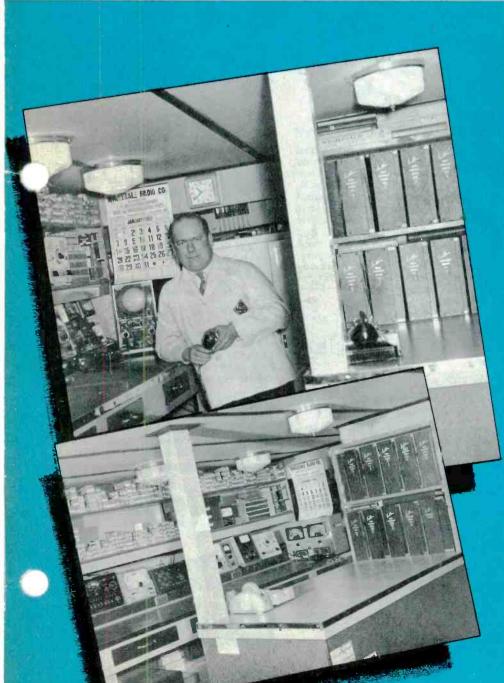
AND TECHNICAL DIGEST



MAY • 1951 including INDEX No. 26 COVERING PHOTOFACT FOLDER SETS I THRU 134

CONTENTS

Shop Talk	
Milton S. Kiver	- 4
Video IF Amplifiers	
W. William Hensler	5
Converting the RCA Victor 730TV-1	
Robert B. Dunham.	11
Dollar and Sense Servicing	
John Markus	17
As I See It	
Walter R. Jones	21
Keyed AGC Application	
R. B. Dunham and W. Wm. Hensler.	23
Experiments in Audio	
James R. Ronk	27
PHOTOFACT CUMULATIVE INDEX	
No. 26 Covering PHOTOFACT	
Folder Sets Nos. 1-154 Inclusive	33
Picture Tube Replacement Chart	59
+More or Less-	62

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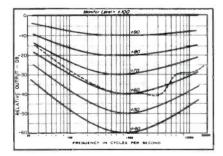
THIRD IN A SERIES OF LRC TECHNIC-AIDS

HOW TO ASSEMBLE LOUDNESS CONTROLS FOR HI-FI TONE AT ANY VOLUME LEVEL WITH IRC Q CONTROLS AND MULTISECTIONS

IRC Continuously Variable Loudness Control for Accurate Compensation in Most AM, FM, TV Sets

Here, at last, is a simple Loudness Control that actually lets the listener hear every tone with almost perfect balance —even at whisper level! It is available as a completely assembled unit, stock No. LCI or you can assemble it yourself economically, in just a few minutes, with a few standard parts obtainable from your IRC Distributor. You can install it in most audio systems as easily as you would an ordinary volume control. And you can use it to "upgrade" your service sales—because it's the very thing your customers have wanted for years.

Compare the Performance of This New Loudness Control with That of Any Other Compensating Device



Tapped volume controls—stepped-type loudness controls—bass and treble boost circuits—you've probably tried them all at one time or another. And you probably found that none of them gave the performance of a true continuously variable loudness control.

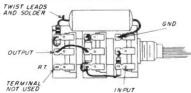
Tapped volume controls afford compensation only when contactor is at the tap. For wider spread of compensation, two or three taps must be used—which makes the controls more difficult and expensive to manufacture. Stepped-type controls permit considerable change of volume between steps, but do not provide full flexibility of adjustment. Also, they are relatively expensive. Bass and treble boost circuits require multiple adjustments with change of volume for ideal compensation.

The new Loudness Control, originated by IRC, does what these other devices have failed to do. It is the only continuously variable loudness control that can be easily and inexpensively assembled from standard parts. With it, you boost highs and lows automatically as volume is decreased—maintain depth and brilliance of tone—without expensive taps or multiple adjustments. The chart at left shows response curves for control at various listening levels.



Here Are All the Parts You Need to Assemble the New IRC Loudness Control

IRC's small ¹⁵⁶" Q Control and original Multisections combine with 2 Advanced BT Resistors and 2 capacitors to form the most efficient loudness control you've ever seen. This is the same Q Control with adaptable fixed shaft feature that technicians have widely used to satisfy most replacement requirements. And Multisections for months have provided an easy answer to ganged control requirements. More than likely you've already used IRC's Q Controls and Multisections to assemble standard duals.

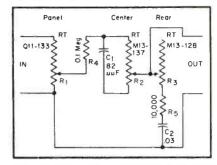


Here's How Easy It is to Assemble the New IRC Loudness Control

Assembling the new IRC Loudness Control is simplicity itself. A glance at the pictorial schematic above will show you how completely easy it is. Here's all you do...

Fasten two specified Multisections to Q Control, as shown, attaching them just as you would switches. Assemble the additional parts—2 BT resistors and 2 capacitors—and make all connections as shown in the diagram. Type 76-1 or 76-2 switches may be added if required. Cut shaft to required length and wire into any high gain audio amplifier.

That's all there is to it. In a matter of minutes, using no special tools, you've assembled a really efficient loudness control that suits most Radio and TV sets.



Only 3 Connections Needed to Install Control. No Special Taps or Complicated Circuits Required

The above diagram of the IRC Loudness Control shows not only the simplicity of design—but the ease of installation as well. Actually, there are only 3 connections—IN, OUT and C or GROUND. It's as easy to wire into most audio systems as an ordinary volume control would be.

With only a few exceptions, the new IRC Loudness Control can be used to improve tonal quality in record players, amplifiers, AM and FM radio and television sets.



Every customer, who puts up with ordinary uncompensated volume control, is a prospect for the new IRC Loudness Control. This inexpensive, easily-assembled loudness control demonstrator unit will help you convince them and self them. You can build it yourself—quickly and easily—with IRC'S CONCENTRIKIT.

This is the simple kit of universal parts,

which you may already have used to assemble concentric duals. As for instructions, we'll furnish them—free of charge.



Your request on a penny postal card will bring you full step-by-step directions for assembling the IRC Loudness Control Demonstrator Unit. We'll also be glad to send you any further information you may wish on the Loudness Control itself.



Pick of the Trade

ANTIQUE BRASS

"From the Army's top level comes the proposal that, in times of emergency, FM and TV stations should be closed down, and only AM broadcast transmitters be allowed to stay on the air. But we seem to recall that Jap planes shot down in the attack on Pearl Harbor were found to have receivers tuned to one of the local AM stations."

> MILTON B. SLEEPER, Editor FM-TV, Radio Communication See March 1951 Issue

* * *

"During our present emergency the use of only first-class quality material will pay out. It pays big dividends to the radio technician in the end.

"At best from now on the service technician will be continuously harassed and his time will become more precious as long as the emergency lasts."

> HUGO GERNSBACK, Editor Radio-Electronics See April 1951 Issue

"Magnetic Recording Units having a retail value of \$15,000,000 were produced during 1950 by 46 licensees of Armour Research Foundation."

* * *

DISTRIBUTORS AND THE SERVICE MAN

"Too often, the importance of the distributor is discounted by the Service Man. Actually, his value to the shop is inestimable.

"It is the distributor who provides that important component or accessory at the right time. He's the man who follows through on the shipments and sees to it that his shelves are stocked with the merchandise required for that installation or servicing call.

"His trained sales personnel often provide the Service Man with vital application guidance. It is the distributor who even assumes the credit responsibility for the service shop.

"The distributor is a true friend of the Service Man and his shop!"

LEWIS WINNER, Editor Service Magazine See March 1951 Issue



AND TECHNICAL DIGEST

VOL. 1 · NO. 3 MAY, 1951

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CONTENTS

Shop Talk Milton S. Kiver	. 4
Video IF Amplifiers W. William Hensler	. 5
Converting the RCA Victor 730TV-1 Robert B. Dunham	. 11
Dollar and Sense Servicing John Markus	. 17
As I See It Walter R. Jones	. 21
Keyed AGC Application R. B. Dunham and W. Wm. Hensler.	. 23
Experiments in Audio James R. Ronk	. 27
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Picture Tube Replacement Chart	. 59
+More or Less -	. 62



HOWARD W. SAMS, Publisher

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ABOUT THE COVER: The photograph is of R. D. Cichy, proprietor of a service shop in Belding, Michigan. Mr. Cichy writes: "Here is an actual 'unsolicited' letter of praise for Sams' Photofact services. I am the guy who always wonders how much was paid for so-called unsolicited testimonials. To keep it short, my business would soon become very tedious if it were not for your most practical way of furnishing service data. Keep up the good work. Enclosed is a photo of what you might call a Sams' shop. Your full set of manuals is one of the most important parts of my shop and as you can see I like to have all I can of it in view of the public."

MILTON S. KIVER

President, Television Communications Institute

"WHAT TEST EQUIPMENT WILL I NEED AND HOW MUCH MUST I SPEND FOR IT?" is the big problem confronting every technician contemplating television servicing.

The answer to the first question has been given many times and is well known by now. Basically, you need a VTVM, an oscilloscope, an AM signal generator, and a sweep signal generator. If you wish to measure high voltage, a special probe can be bought for the VTVM which will accomplish this job. The same can be said of a high frequency probe. So if you want the absolute minimum, there you have it.

The answer to the second question has not received as much attention as the first, although, to a great extent, it is the more important of the two. While the TV service aspirant can usually figure out what he might need in the way of equipment, he does not know how low he dare go in price and still obtain something worth while. The emphasis here is on low priced equipment because if you can afford a Hickok 610A, or a Simpson Genescope, or a Precision 400-C, or an RCA scope, and others of similar quality, then by all means go out and get them. In test equipment, as in most other items, you get only what you pay for.

The major problem, as the writer sees it, concerns the man who has limited capital to spend, yet who wants to get as many different items for his few dollars as he can. With this in mind, let us examine each of the above mentioned basic instruments and see what you should try to get for your money.

VTVM. The VTVM has been around long enough by now to have become quite standard in design. You will readily discover, after some investigation, that nearly all such units employ some form of bridge circuit and that the differences in prices stem primarily from certain refinements that have been added to the instrument. Thus, one VTVM will contain a zero center scale, another will possess an extra Db scale, a third will have a special high frequency probe capable of RF measurements up to 300 mc, etc.

Most of the refinements are good to have and will, to a certain extent, make your servicing tasks easier. But 90 per cent of your TV service work deals with resistance and voltage checks - and so you have to settle in your own mind whether certain refinements are worth the extra money you have to pay for them. If you feel that you are only concerned with the work this meter will have to perform 90 per cent of the time, then get a unit which has only 7 De volt scale, an AC volt scale, and a resistance scale. And you will find that it answers your purpose. But if you feel that you should have one with the additions mentioned above, then adjust your budget to include one of these.

OSCILLOSCOPES. The range of prices that you will encounter in oscilloscopes will be more extensive than those of VTVM's. Yet here again you are dealing in circuit refinement rather than basic circuit difference. An expensive oscilloscope will possess high gain and wide frequency response. A less expensive unit will have a frequency response (in its vertical amplifiers) perhaps up to 100,000 cycles and require possibly .2 of a volt input for one inch deflection on the screen. This means that more signal will have to be pumped into the circuit in order to obtain a sizeable pattern.

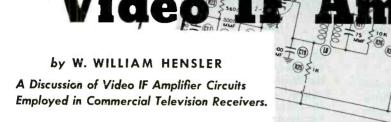
There is considerable controversy about the bandpass of the vertical deflection amplifiers and how wide this should be for television servicing. For ordinary servicing, where you are generally interested more in determining whether the video signal is present rather than how rectangular its sync pulses are, a frequency response to the vicinity of 100,000 cycles will suffice. The same is true of the vertical and horizontal sweep systems of the TV receiver where the fundamental frequency of the deflection voltages are quite low (vertical 60 cycles, horizontal 15,750 cycles). As the bandpass increases, the pulses more nearly approach their rectangular form. But greater bandpass means lowered gain and to offset this reduction, more amplifiers are needed. Which, in turn, raises the cost of the instrument.

The chief difficulty you will find in low priced oscilloscopes is obtaining one which will give you a sizeable deflection for a moderate input voltage. Pay particular attention to this point because nothing will hamper your service work more than dealing with a small pattern and trying to figure out whether it conforms to the recommended shape or not. Forego a little vertical amplifier response - but try to get as high a deflection sensitivity as possible for the amount of money you can afford to spend.

SIGNAL GENERATORS. The problem of the AM signal generator and the sweep generator can be considered at the same time because they serve a common purpose - to help in aligning the various TV circuits. To gain a better appreciation of the job these instruments must do, let us briefly consider the character of the tuned circuits in television receivers.

In the RF and video IF stages you will find wide-band amplifiers designed to pass a band of frequencies from 3 to 6 mc wide) Obviously, to properly

♦ Please turn to page 49 ♦ ♦



ST VIDEO IF AMP.

AII [AI2] (2

In order to understand the requirements of video IF amplifier systems, the nature of the signal which the amplifier must handle should be taken into consideration.

The video carrier is an amplitude modulated signal with approximately 4 mc as the maximum modulation frequency. In conventional double sideband transmission, an 8 mc bandwidth would be required to transmit this signal. Since the allocated TV channel is only 6 mc wide, it is obvious that the double sideband method cannot be employed.

It is possible, however, to transmit this intelligence using single sideband transmission. This is accomplished by filtering out one of the side bands at the transmitter. In actual practice, a high pass filter is usually placed in the transmission line between the transmitter and the transmitting antenna, thus filtering out the lower sideband beyond a point approximately 750 kc from the video carrier frequency. This type of transmission is known as "vestigial" sideband modulation, since only a small part, or "vestige," of one sideband is transmitted.

Figure 2-1 shows the frequency distribution at the output of a transmitter operating on Channel 6. The video carrier is 1.25 mc above the low end of the channel, and the sound carrier is .25 mc below the high end of the channel, making the sound and video carriers 4.5 mc apart.

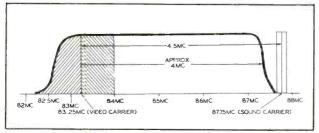


Figure 2-1. Frequency Distribution of a Channel 6 TV Transmitter.

The shaded portion to the right and left of the video carrier represents that part of the signal which has double sideband transmission. In other words, all modulation frequencies below 750 kc will be transmitted with twice the power as compared to those above 750 kc.

5 GAUS

If special precautions are not taken in the design and alignment of receiver video IF amplifiers, the lower frequencies will be over-emphasized, resulting in poor picture reproduction.

Referring again to Figure 2-1, it can be seen that in order to utilize the complete transmitted signal, the video IF amplifier must have a bandpass of 4 mc. and must also incorporate some means of reducing the gain of the signal which carries the lower modulation frequencies.

Although the sound transmission has nothing to do with the transmission of picture information, its signal must be taken into account in the video IF amplifier design for several reasons.

There are two major classifications for video IF amplifiers: (1) The separate channel system, where sound IF and video IF signals are amplified separately, and (2) the intercarrier system, where both of the IF signals are amplified in the intercarrier strip.

In the case of the separate channel system, the sound IF signal is removed through the use of trap circuits in the video IF input, and is then fed to the sound IF channel for amplification. Hence, the nature of the sound transmission must be considered for satisfactory trap design and alignment to make sure that the sound does not get through the video IF strip.

In the intercarrier system, the sound IF signal is allowed to continue through the video IF strip, with its amplitude held at a predetermined percentage of the video IF signal.

VIDEO IF AMPLIFIERS

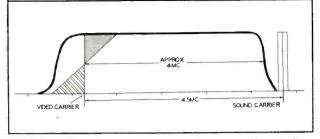


Figure 2-2A. Ideal Separate Channel Video IF Response Curve.

Figure 2-2A shows an ideal separate channel video IF response curve placed over the transmitter output curve. Since this is an ideal response curve, it follows the transmitter output curve at the high end indicating that the bandpass is sufficient to pass all modulation frequencies.

In actual practice, the bandpass may not be the full 4 megacycles wide, but it is shown here as such to indicate the ideal condition.

It should also be kept in mind that the IF response curve will be inverted from that shown in those receivers where the local oscillator operates above the incoming signal.

As mentioned above, some means of preventing over-amplification of the signal representing the lower modulation frequencies must be taken. As can be seen in Figure 2-2A, the video IF amplifier is so aligned that the video carrier is halfway up the slope. The lined area represents the amount of the lower sideband that is passed through the video IF amplifier while the dotted area represents the amount of upper sideband that is not passed. If the video carrier is at the midpoint of the slope, and the slope is straight, the lined and dotted areas will be equal. Under these conditions all frequencies will be amplified an equal amount. If, however, the video carrier is too far up the slope, too much of the lower sideband will be passed, resulting in over-emphasis of the lower frequencies. On the other hand, if the video carrier is too far down the slope, the lower frequencies will be weak, resulting in poor reproduction.

Figure 2-2B shows an ideal response of an intercarrier IF placed over the frequency distribution curve of the transmitter. The response curve is the same as that in Figure 2-2A at the video carrier end but differs at the other end. A "shelf" has been added which allows the sound IF frequency to be amplified in the video IF amplifier, but at a much lower level than the video carrier. In actual practice the "shelf" may not be nearly so pronounced as shown or, as in many cases, the sound carrier is

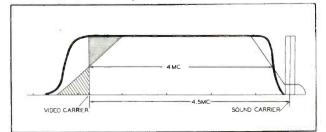


Figure 2-2B. Ideal Intercarrier IF Response Curve.

placed at a predetermined point up the slope and no actual "shelf" will be present. In either case the sound carrier is placed at a point which provides only 5% of total amplitude. At the video detector the beat note of the video and sound carriers produces a 4.5 mc sound IF signal which is trapped out and fed to the sound IF portion of the receiver.

Some means must be provided to control the gain of the video IF amplifier since all signals that are received will not be of the same strength. Otherwise, weak signals would not be amplified enough for the detected signal to be useful. Likewise, if the gain cannot be reduced when receiving strong signals, overloading will occur. Several methods have been incorporated to vary the gain of the IF amplifier. In one case a varying bias is manually applied to the bias line of the IF strip. In other circuits a control in the cathode circuits of the IF amplifiers is manually operated to control the gain.

At the present time, however, the trend is to incorporate some means of automatic gain control where the signal itself is used to develop a bias that is proportional to the amplitude of the signal. Through this means the signal will adjust the gain of the amplifier so that an approximately equal signal is developed at the video detector load whether the signal is weak or strong.

The video IF amplifier must be so designed that there is a minimum of noise developed within the circuit. Special care must be taken in selecting the type of tubes and circuits to be used. Pentode tubes are almost always used because of their high transconductance which makes possible a reasonable gain even though low Q coils are used in the tuned circuits. The amplifier must be stable at all gain settings as any tendency for oscillation will result in poor picture reproduction.

The choice of video IF frequency that is used is important. The higher the frequency, the easier it becomes to obtain full bandpass with a minimum of stages. As the frequency is increased, however, the gain decreases and the possibility of unstable operation increases. Thus it can be seen that a compromise must be made. The most popular frequencies that have been used for video and sound IF is in the region between 21 and 26 mc. At these frequencies a reasonable gain, as well as adequate bandpass, can be achieved. In a receiver using these IF frequencies, however, it is necessary to operate the local oscillator at a frequency which is within some of the television channels. Under these conditions the radiation from the oscillator may interfere with the reception on another receiver.

Considerable work has been done on tuners to keep oscillator radiation at a minimum but the results are not always the same on production models. The present trend is to increase the IF frequency which places the oscillator frequency farther away from the incoming signal. The oscillator frequency is then more easily attenuated in the tuned circuits of the tuner. With the IF in the 41 to 46 mc range, the operating point of the oscillator does not fall within any of the tellevision channels. The recent action of the Federal Communications Commission provides tentative approval of the 41 to 46 mc range. The use of these higher frequencies is made possible by the development of higher gain tubes which will provide adequate gain at these frequencies.

In order to reject unwanted signals, trap circuits are employed in the video IF strip. The traps may be absorption, series or parallel tuned circuits. The type that is employed depends on the nature of the signal that is being rejected and also the circuit in which it is used. The operation of each of these traps will be discussed later.

The video IF amplifier must be selective. That is, it must be able to reject signals that are not within the channel which is being received. Because of past experience in radio, it is a natural tendency to associate selectivity with an amplifier having a very narrow bandpass. By definition selectivity means the ability to discriminate or separate a predetermined band of frequencies from all other signals. Thus applied to television reception, good selectivity would require that a receiver accept only those frequencies within the television band being received and reject all other signals. It would seem that such good selectivity would not be required since adjacent channel assignments are not made in a given locality. It is possible, however, to have a receiver operating at a point between two stations that are operating on adjacent channels where adjacent channel rejection would be required. It is because of this possibility that adjacent channel sound and video traps are incorporated.

Another need for good selectivity is for the rejection of interfering signals that are near the channel which is being received. The results of an interfering signal are much more objectionable in TV reception than in sound reception, since it may degrade the picture and, in some cases, may cause loss of synchronization. Obviously, if the interfering signal is within the frequency limits of the desired channel, it will be accepted as a part of the desired signal and very little can be done to reject it. Good selectivity, however, will make possible the rejection of unwanted signals outside the frequency limits of the desired channel.

From the requirements set forth so far it has been established that a video IF amplifier should provide the following:

1. Adequate Bandpass.

 $\label{eq:compensation} \begin{array}{l} \text{2. Compensation for Vestigial Sideband Transmission.} \end{array}$

- 3. Trap Circuits to Reject Unwanted Signals.
- 4. Variable Gain.
- 5. Stable Operation.
- 6. Sufficient Gain with a Minimum of Noise.
- 7. Good Selectivity.

TUNED CIRCUITS

Some of the requirements above are design problems and do not directly concern the service technician. The problem of obtaining adequate bandpass, however, must be solved quite frequently when an alignment is performed on the video IF strip. The alignment procedure that is followed is governed by the type of tuned circuits that are employed. Three basic types of tuned circuits used in television video IF amplifiers, up to the present time, listed in their order of popularity, are as follows:

- 1. Stagger Tuned.
- 2. Transformer Coupled.
- 3. Bandpass Circuits.

The simplest method of obtaining adequate bandpass in a video IF amplifier is through the use of several single tuned stages. Each of the stages are tuned to different frequencies making possible the wide bandpass of the complete video IF strip. Since the coils are tuned to different frequencies, the tendency for oscillation is decreased, making the production problem easier as far as placement of parts is concerned. The coils themselves need not be high Q units. On the contrary they should be low Q units to provide a wider response. This makes it possible to keep down the cost of the coils resulting in lower production costs.

Figure 2-3 shows the response curve of a single tuned stage, along with a partial schematic of a stage representative of the type used in video IF amplifiers.

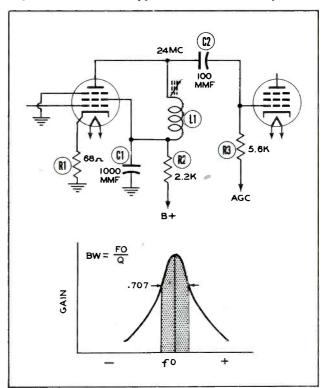


Figure 2-3. Response Curve and Circuit of Single Tuned Stage.

VIDEO IF AMPLIFIERS

A single tuned coil (L1) is placed in the plate circuit of the 6AU6. L1 is tuned by the output and input capacities of the two tubes along with the distributed capacity of the wiring. Use of a minimum capacity makes possible a higher inductance value for L1 over the value which would be permissible if more capacity were placed in the capacitive branch of the tuned circuit. Thus a higher L to C ratio can be maintained, making possible more gain in the circuit.

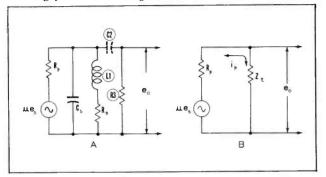


Figure 2-4. Equivalent Circuit of Circuit in Figure 2-3.

The gain of a stage having a plate load which is small in comparison with the plate resistance of the tube, is dependent upon the transconductance of the tube and the impedance of the load. Figure 2-4A shows the equivalent circuit of the video IF amplifier illustrated in Figure 2-3. E_S represents the signal voltage, Rp the plate resistance of the tube, Cs represents the tube capacity along with the distributed capacity, Rs is the series resistance of L1, and R3 is the grid resistor. C2 can be disregarded in the equivalent circuit as its reactance is so low at the operating frequency. Since the signal is developed across the parallel network of C_S, L1 and R3, these three branches can be lumped together and designated as Z_t , the total impedance of the load. This is shown in Figure 2-4B.

Current (I_p) will flow in the closed circuit and the voltage* (mu E_S) will be divided between R_p and Z_t . With Z_t extremely small in comparison to R_p , any change in Z_t (as long as its value is kept low) will not change the amount of current flowing in the circuit. Since the amount of current remains constant, the gain of the stage (the voltage across Z_t) can be changed only by varying the value of Z_t .

In conclusion, it has been said that the gain of a stage having a load which is extremely small in comparison to the plate resistance of the tube, is dependent upon the G_m of the tube and impedance of the load. In order to obtain maximum gain, the L to C ratio should be kept as high as possible and a tube having a high mutual conductance should be employed.

The impedance of a parallel resonant circuit is in proportion to the Q of the circuit. The higher the Q, the higher the impedance and vice versa. Another characteristic of the circuit which is dependent on the Q is the bandwidth. The bandwidth, at the halfpower point, is equal to the resonant frequency divided by the Q of the circuit. Assuming that the resonant frequency of the circuit shown in Figure 2-3 to be 24 mc, and the Q to be 10, the bandwidth of the circuit would be 2.4 mc. By increasing the Q, in order to increase the impedance, the bandwidth of the circuit would be decreased. Some compromise must be made in arriving at the value of the effective Q so that a reasonable gain can be achieved and still maintain adequate bandpass.

The coil illustrated in Figure 2-5 is a tunable choke representative of the type that is used in stagger-tuned video IF systems. A core has been removed from another identical coil and is shown alongside to illustrate its construction. In taking measurements on this coil, the minimum inductance, with the slug all the way out, was found to be 1.75 microhenries, and the maximum inductance, with the slug all the way in, was found to be 4.1 microhenries. The Q of the coil is 75.

The circuit shown in Figure 2-3 employs this type of coil to perform the functions of L1. Note that no additional capacity is placed across L1 other than the tube capacity and the distributed capacity of the wiring. In order to resonate the circuit with the slug set all the way in, to achieve maximum inductance, the maximum allowable capacity in the circuit would be 10 mmf. The input and output capacities of the two tubes are across the coil, and, by referring to the tube manual, the sum of these two capacities is found to be 10.5 mmf. This does not allow for distributed capacity or Miller Effect, so it is obvious that this amount of inductance cannot be used. Assuming that the total capacity in the circuit is 15 mmf., an inductance of approximately 2.9 microhenries would be required. This is well within the range of the tunable choke shown in Figure 2-5.

The reactance of the coil at this setting would be 437 ohms, and assuming that the effective Q of the circuit could remain at 75, the total impedance of the

$$I_p = \frac{mu E_S}{R_p + Z_t}$$
 (Equation 2-1)

The voltage out (E_0) equals the current through the load times the load impedance or:

$$E_0 = I_p X Z_t$$
 (Equation 2-2)

Substituting the equation 2-1 for Ip in equation 2-2 we have:

$$E_0 = \frac{mu E_S X Z_t}{R_n + Z_t}$$
 (Equation 2-3)

The voltage output of any stage equals the voltage input, times the gain or:

$$E_0 = Gain X E_S$$
 (Equation 2-4)

Substituting equation 2-4 for E_0 in equation 2-3 and removing $E_{\rm S}$ from both sides we have:

$$Gain X E_{S} = \frac{mu E_{S} X Z_{t}}{R_{p} + Z_{t}} \text{ or } Gain = \frac{mu X Z_{t}}{R_{p} + Z_{t}}$$
(Equation 2-5)

Since in this case Z_t is extremely small in comparison with $R_p,$ the quantity of R_p plus Z_t may be considered as R_p only, making the equation 2-5 read:

Gain = mu X
$$\frac{Z_t}{R_p}$$
 or Gain = $\frac{mu}{R_p}$ X Z_t (Equation 2-6)

The transconductance (G_{m}) of a tube equals the amplification factor (mu) divided by the plate resistance $(R_{p})\colon$

$$G_m = \frac{mu}{R_m}$$
 (Equation 2-7)

Substituting equation 2-7 for $\frac{mu}{R_p}$ in equation 2-6 we have:

Gain = Gm X Zt

(Equation 2-8)

^{*}The voltage developed in the circuit (Figure 2-4B) is equal to the signal voltage (E_S) times the amplification factor (mu) of the tube or mu E_S. The current that will flow in the circuit is limited by R_p and Z_t, therefore the current in the circuit can be stated as:

Figure 2-5. Tunable Choke Used in Stagger-Tuned Video IF Systems.

load would be 437 times 75 or 32,775 ohms. The mutual conductance of the 6AU6 tube is listed at 4450 micromhos which would provide a gain of around 145. Gains of this quantity cannot be obtained in this type of circuit for several reasons, the most important of which is the need for lowering the Q of the circuit to obtain wider bandpass. With a Q of 75, a tuned circuit of this type would have bandpass of only 320 kc which falls far short of the requirements. If the effective Q were lowered to a value between 6 and 10, however, the bandpass would widen out to between 2.4 and 4 mc, and the stage will give a gain of 10 to 17 which is average. It should be kept in mind that the above calculations do not exactly duplicate the design characteristics of the circuit, but are given as an example to show what problems are involved. By gaining a better understanding of the design problems of this type of circuit, the service technician will be able to make component replacements and be assured of proper performance.

R3, in the partial schematic of Figure 2-3, serves as the grid return for the following stage, and also acts as a shunt across L1 to aid in lowering the Q. A resistance placed across a coil lowers the Q since the resistance is shunting the load resulting in a lower total impedance. In this type of circuit, all the developed signal is coupled to the following stage. If the gain is excessive, regeneration might take place. By varying the value of the shunt in the circuit the gain of the stage can be held below the oscillation point. When making replacement of these

resistors, exact duplicates must be used to maintain proper gain and bandpass in the video IF strip.

Figure 2-6 is a schematic of a video IF strip employing tunable chokes similar to the type illustrated in Figure 2-5.

Four 6AU6 type tubes are employed, having stagger-tuned circuits to obtain adequate bandpass. Note that all the coils are tuned to different frequencies which widens the frequency response and also lessens any tendency for oscillation in the strip. AGC is applied to each stage with a ladder-type decoupling network added to prevent feedback through the AGC line. The plate and screen of each stage is also decoupled by means of an RC network.

Tracing the signal through the circuit, we find that the output of the mixer has a parallel tuned circuit composed of L1 and the output capacity of the mixer tube. The mixer is decoupled by C1 and R2. The signal is coupled to the 1st IF grid by C2, and R3 serves as the grid load of the tube. The value of R3 is 8200 ohms which lowers the effective Q of L1, thereby broadening its bandpass characteristics.

The second and third IF stages are essentially the same as the first IF except for the values of the grid resistor and the decoupling resistor in the plate and screen circuit. Another change is the addition of an unbypassed 68 ohm resistor in the cathode circuit to introduce a very small amount of degeneration. This has the effect of increasing the input resistance of the tube and also lessens the change of input capacity caused by the Miller effect. The small loss in gain can be disregarded, since the advantage of decreasing the capacity change due to Miller effect overshadows this disadvantage.

The tuned circuit of the last IF stage differs from the others in that it has a resistive plate load and the coil is moved to the right of the coupling capacitor. Since it is necessary to use a low value detector load, a low DC resistive path is required at the input side of the diode to prevent a loss of signal at the input side.

It is interesting to note that this IF amplifier has no traps. The selectivity is achieved by proper stagger tuning of the tunable chokes. All of the chokes

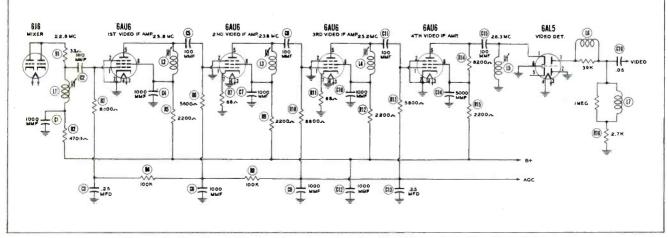
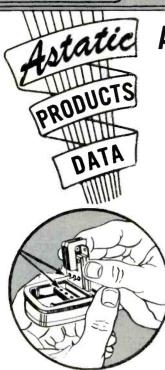


Figure 2-6. Complete Stagger-Tuned Video IF Strip Using Tunable Chokes.

Please turn to page 30



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SIMPLE, SNAP-IN INSTALLATION OF 402-M CARTRIDGE

Installing in the Admiral Arms for which it was designed is a simple matter of inserting the three-prong terminals into the three snap-in receptacles found in these arms. Snap-in action holds the cartridge firmly in place and nothing else need be done.

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				SPECIFICATION.	S			
Model No.	Element Type	List Price	Minimum Needle Pressure	Output Voltage 1000 c.p.s. 0.5 Meg Load	Frequency Range c.p.s.	Needle Type	Approx. Net Wt. in Grams	Code
402-M	Ceramic	\$6.90	12 gr.	0.7* *Audio-tone Test Record	50 to 10,000	G-78 (osmium tip)	8	ASWZN

ASTATIC AC-AG AND AC-J CRYSTAL CARTRIDGES ARE IDEAL REPLACEMENTS FOR ADMIRAL 409A13, 409A13-1 AND 409A300 CARTRIDGES

Model	List Price	Minimum Needle Pressure	Output Voltage 1000 c.p.s. 0.5 Meg Load	SPECIFIC Frequency Range c.p.s.	Needle Type	Application	Code
AC-J	\$8.90	5 gr.	1.0**	50 to 10,000	A-1 (1-mil sapphire tip)	Replaces Admiral 409A300	ASWYJ
AC-AG-J	8.90	6 gr.	1.0**	50 to 10,000	A-AG† (sapphire tip)	Replaces Admiral 409A13 and 409A13-1	ASWYH

*Audio-tone 78-1 Test Record. **RCA 12-5-31V Test Record.

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†"ALL-GROOVE" Needle tip of special design and size to play either 331/3 and 45 RPM (narrow groove) or 78 RPM (standard groove) records.



Astatic Crystal Devices manufactured under Brush Development Co. patents



Converting the RCA Victor 730TV-1 to 14 inch Operation

by Robert B. Dunham

The cabinet and chassis design of the RCA Victor Model 730TV1 (See Figure 1) is such that it may be converted to use a 14" rectangular tube with a minimum of physical changes in the chassis and the cabinet. Figure 2 illustrates this model after conversion to use a 14" tube.

This set incorporated a 10" tube as original equipment and the tube was mounted unusually high above the chassis to position it at a point that would follow the design of the cabinet. The resulting added height of the deflection yoke and focus coil above the chassis, makes it possible to mount the 14" tube in the original bracket and still allow the tube to clear the components on the chassis. The front of the tube is supported by a cutout in a subpanel at the front of the cabinet. It is necessary to cut a new opening in this subpanel to support the new tube. The front panel of the receiver can be removed by taking out the two screws at the top of the panel. This will permit the alteration of the front panel to be performed on the work bench, which is much easier than doing the work within the cabinet itself.

Because of the minor nature of the cabinet work required in this conversion, it can be performed in the service shop without the need of special wood working tools. A complete description of the cabinet work required is given later. It is suggested that no work be done on the cabinet until the electrical con-



Figure 1. RCA Victor Model 730TV1.



Figure 2. The Converted Receiver.

version of the chassis is completed. This practice is wise to follow on any conversion job, so that in the event the chassis cannot be converted for some unforeseen reason, it can be restored to its original circuitry and placed back in the cabinet.

CIRCUIT CHANGES

A partial schematic showing the original wiring in the horizontal circuit is shown in Figure 3. The components located in the high voltage compartment, which need to be removed in this conversion, are shown in Figure 4. These components are the horizontal output transformer, the width coil and the high voltage filter capacitor. After the removable section of the high voltage compartment was taken off, the leads to terminals 1, 4, 5 and 6 were unsoldered and the four screws which mount the horizontal transformer to the HV compartment were removed. The remainder of the high voltage compartment was then taken off which gave access to the high voltage rectifier filament leads. After these filament leads were unsoldered, the horizontal output transformer was removed. The high voltage filter capacitor is mounted in a single clip socket in this model and was removed by slipping the clip off the top connector pin and lifting the capacitor from the socket. The original width coil was then removed from the bracket. This completed the removal of the major components from the HV compartment.

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AUTO VIBRATOR TRANSFORMER SHEET-FORM No. 3, dated DEC. 30, 1950, shows model No., Net, List prices and Specs. of VIBRATOR TRANSFORMERS for FORD-GM-MOTOROLA and MOPAR car radios. Also simple easy-to-read replacement guide covering 30 manufacturers.

MERIT OUTPUT TRANSFORMER CHART-FORM No. 4, single sheet shows proper Merit output transformer for use with all popular output tubes. Both MERIT specific and universal types are shown. Mounting style is included for further convenience.

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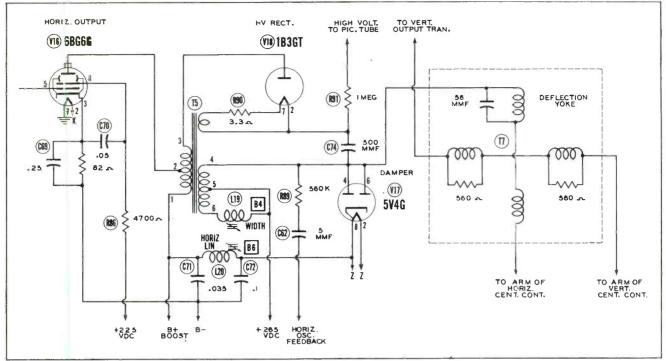


Figure 3. Schematic of Original Horizontal Sweep Circuit.

The new high voltage capacitor was then inserted in the original capacitor socket and the top clip connected. A 15 or 20 KV unit should be used since the original unit has a 10 KV rating and might break down under the voltage developed in the revised circuit. The new horizontal output transformer, a Merit type HVO-6, was then mounted. It was positioned close to the high voltage rectifier socket so that the high voltage rectifier filament leads could be kept as short as possible. This required the drilling of three holes to mount the new unit. The placement

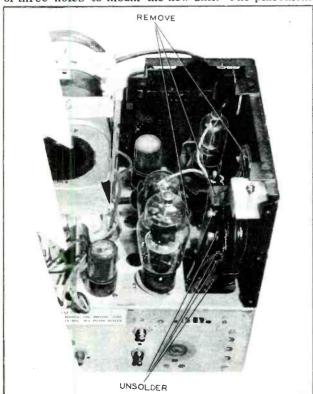


Figure 4. High Voltage Compartment.

of the components is shown in Figure 6. The filament leads were then soldered to the proper tube socket terminals.

Editor's Note: As this goes to press, the Merit Transformer Corporation is starting production on a type HVO-7 horizontal output transformer which is identical electrically to the HVO-6. The type HVO-7, however, has a mounting bracket which is drilled to fit in the same mounting holes used by the original RCA transformer. The leads on this unit are long enough to allow it to be mounted in the same location and position as the original unit. If the HVO-7 transformer is used there will be no need for drilling the new mounting holes previously described in connection with the Merit type HVO-6.

The next step was the mounting of the new width coil, a Merit type MWC-1. This coil is slightly larger than the original unit and requires slight reaming of the hole in the width coil bracket. After the width coil was mounted in the bracket, the side of the high voltage compartment was remounted on the chassis. The leads to the horizontal output transformer were then soldered to the proper terminals as indicated in the schematic of Figure 5.

The next step was the removal of the focus coil and deflection yoke. A Merit type MD-70 deflection yoke which is designed for use with tubes requiring 70 degree deflection, replaces the original yoke. The damping capacitor and resistors were first installed in the yoke per the manufacturer's instructions. The original focus coil was tried with the 14" tube but it was found that the focus point was not quite reached at the limit of the focus control. By using a higher resistance unit, such as the Merit type MF-3, proper focus could be obtained. This type focus coil is the same style as the original unit and will mount in the original bracket with no changes required.



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REPLACING BURNED OUT RESISTORS—With the chassis to be repaired plugged into a LOAD-CHEK MODEL 660, note the wattage reading with the burned out resistor circuit open. Now replace the resistor. Should the increase in watts be greater than that of the resistor rating being installed, it indicates that an extra load has caused the trouble which has not been cleared.

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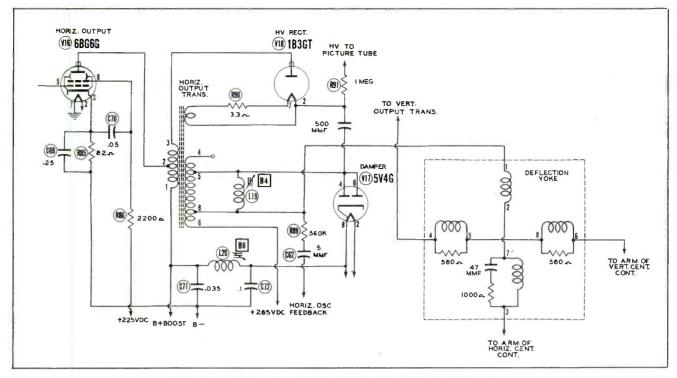


Figure 5. Schematic of Revised Horizontal Sweep Circuit.

In order to insure that the new tube will clear the top adjustment screw on the discriminator transformer, spacers were installed between the deflection yoke and the bracket to which it is mounted. About an eighth of an inch is all the spacing required and conventional steel nuts serve the purpose very well. Select them so that the center hole is slightly large, to slip easily over the studs. This small amount of added height helps considerably in providing adequate clearance.

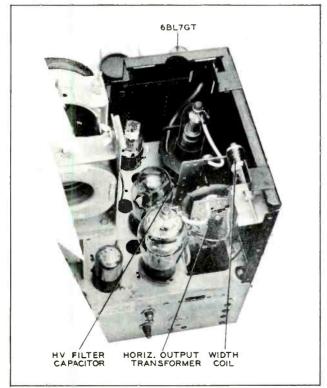


Figure 6. High Voltage Compartment after Conversion.

The screen resistor of the 6BG6G horizontal output tube was changed to a 2200 ohm, 1 watt unit. This change was made to improve horizontal linearity in the new application.

The original ion trap used in this model was an electro-dynamic type having two windings. This type trap cannot be used on the 14'' tube. A 39 ohm, 1 watt resistor was wired in the circuit in place of the ion trap winding. Actually it is not necessary to remove the trap from the circuit; it can be taped to the back of the chassis and left wired in the set.

This completed the wiring and electrical changes in the chassis. The new tube was installed to check the operation. It is necessary to temporarily block up the tube during this check since there is no provision for holding the tube on the chassis. A new single permanent magnet ion trap was installed and the operation on the receiver was checked. The horizontal drive, horizontal linearity and width coil were adjusted to obtain proper horizontal linearity and width.

It was then found that it was impossible to obtain sufficient height by adjusting the height and vertical linearity controls. This model employs a 6SN7GT as a vertical oscillator and vertical output. Several new 6SN7GT tubes were tried in this application, but none would furnish sufficient power to sweep the tube vertically. A 6BL7GT was then tried and it was found to work satisfactorily with no changes required in the socket wiring or in the components. An additional tube could be added to serve as a vertical output tube but this would require the punching of a new tube socket hole and considerable wiring changes. The use of the 6BL7GT solves the vertical deflection problem and its use in this conversion is recommended.

After making all the above changes the receiver should sweep the new tube with good linearity, both

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WHEN TO REPLACE A PHONO-CARTRIDGE

John Markus

Dollar and Sense Servicing

SUMMER SLUMP. Buying of new receivers tapered off early this year, with no likelihood of revival until September at the earliest. Instead of cutting contract prices or granting dealer kickbacks to hold new-installation business up artificially a while longer even though at a loss, smart service organizations are retrenching and settling down to operate profitably with their present volume of business. In one way this business slump is good, because it means time to examine overhead expenses and pare them down, time to improve staff efficiency, time to rebuild shop benches and rearrange things for faster and better rush-season servicing, and time for well-deserved days off and vacations.

POSITIONS OPEN. There's almost a black market in engineers and technicians this spring. Companies are hiring men away from each other, and there's even some farming out of development engineers on a loan basis. At the IRE convention in March, companies hired suites of rooms in hotels, to which would be brought likely prospects for softening with refreshments. A good many dotted lines got signed on before the show was over.

Some succumbed to the lure of a new job just to see new faces and new shop benches. Some saw their chance to try out California living, in response to one famous airplane maker's inducement of "10% more then you're making now, plus moving expenses." Many companies also provide on-the-job training and free night school or home study courses that enable a serviceman to upgrade himself.

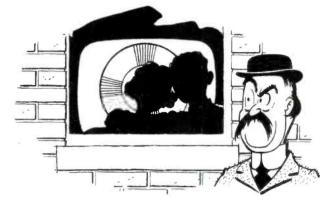
There's even a bit of protective hiring, to get men now for military orders expected later or to improve chances of getting hoped-for contracts. All is good for servicemen, as experience in television servicing makes them highly valuable on radar, guided missile and other military electronic projects.

Hiring of servicemen has been found to pay off, in that it frees scarcer engineers for the toughest research and development jobs. The men who know their circuitry and want to kick their last griping customer out the door can just about take their choice of where to work and who to work for, on military projects that will protect our way of living.

TV NOT GROUNDS FOR EVICTION. A justice of the peace in St. Paul, Minn., ruled recently that a tenant cannot be legally evicted because he installed a TV antenna in the room of a rented house. Despite this, it's still good business to insist that a customer in a rented house show written permission from the landlord for a rooftop TV installation.

COMING DOWN THE LINE. Picture tubes in 14, 17 and 20-inch rectangular sizes using electro-

static focusing, are already on production lines of cathode ray tube makers. This means that they must have orders from manufacturers, so look for the electrostatic tubes in sets sometime during the last half of this year. Reports are that they're the same physical size and shape as magnetically focused tubes and give comparable picture quality. Electrostatic tubes save scarce cobalt, copper and nickel.



TV AND DIVORCE. In England, the judge agreed with one husband's suspicions and granted a divorce, despite his wife's assertion that the living room lights were out only so she and her gentleman friend could see television better.

SAVING TWINLEAD. When the shortage of 300-ohm ribbon loomed, one manufacturer's service division put the heat on their boys to place the antenna as near to the receiver as possible, measure the required length of line accurately and splice together and solder short lengths. This resulted in reducing the average transmission line usage per job from 123 feet to 83 feet. In addition to conserving that much scarce copper, there's a cash-on-theline saving of almost a dollar per job, plus the time and cost of installing standoffs on 40 feet of line.

THREE KINDS OF TIME. In one service organization, there are three different ways for a serviceman to charge off his time during the working day. First is Applied Time, which is directly chargeable to specific service jobs. Second is Accountable Unapplied Time, which covers such things as being out sick, running assigned errands, doing necessary bench and shop cleaning, and suchlike. Third is Unaccountable Unapplied Time, in which the serviceman did nothing useful or can't recall what he did. At the end of each week, the slips are totalized and the results entered on a chart for all to see. The system makes men feel that their productive time receives due publicity. The fellow with the highest figure for Unaccountable Unapplied Time is in for considerable kidding by his co-workers, hence there is rivalry to keep this figure down.



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THE CRYSTAL BALL. Those close to the television receiver market are saying that the present figure of 12 million TV receivers in use is closer to the final peak than we think. Unfreezing the station allocation setup, they say, will make the market only 50% bigger. This means that another 6,000,000 sets will enable the rest of the country to catch up. True, people will continue buying television receivers, but from here on the business will be gradual and seasonal. Therefore, expand your business slowly and carefully. It is far safer to solidify than expand, when in doubt about the future.

HOPE FOR MORE PARTS. To lick shortage of replacement parts for TV sets and radios, NPA early in March permitted use of extra quantities of material above allocation quotas, to be used for making replacement parts at 100 to 150% of rate a company used materials for that purpose during first half of 1950.

REBIRTH. The old capacitance loudspeaker of the 20's is being given another whirl in the research labs, in hopes it can be perfected as a substitute speaker using little or no critical scarce material. Major part in it is a sheet of aluminum foil serving as the diaphragm, moving back and forth in front of a fixed metal plate in accordance with the polarity and strength of the voltage applied between foil and plate.

STRING SAVERS. To overcome a prevalent habit among servicemen of saving used parts that might possibly someday have further use, one service manager has shellacing day once a month. On that day, usually Friday, they stop service work about two hours early and have to shellac the bench top and all shelves. After a few shellacing days their bench drawers get so crammed that they just have to throw the newly acquired junk away.



TENSENESS TEST. Can you go to sleep during a long dull sermon in church? If not, you're too tense - taking yourself or your servicing problems too seriously. But according to one minister, you've got lots of company; his complaint is that he's no longer able to put his congregation to sleep, because they're all too tense these days.

DISCOURAGE TRADE-INS. Smart servicemen are doing everything they can to encourage people to

keep their small-screen set when buying a new set. The small sets are ideal for the basement, where the minority of the family can watch Captain Video or Hopalong in peace. The small sets are equally valuable in bedrooms in time of sickness, for they can be viewed comfortably at arm's length with the controls all in reach. Once a family has had two sets for a while, they become indispensable and provide twice the servicing business of one set.

When sets are traded in, on the other hand, they are seldom resold in the same locality. Instead, they are shipped in bulk to poorer sections of the country, to dealers in poorer sections of cities, or even to Cuban, Mexican and South American markets now that these countries are getting stations. Each such trade-in means the loss of one service card in your customer file.

PICTURE TUBE SIZES. Most popular picture tube in new 1951 sets will be the 17-inch rectangular if production continues at present rate. Last year it was the 16-inch round. Next year, barring a drastic change in the military situation, it could be either the 20-inch or 21-inch rectangular. At Corning, the 20-inch rectangular is already 25% of their total production of glass blanks. There is no likelihood of a shortage of glass for these big tubes, hence the public will get what it demands. You can expect a few 24-inch, 28-inch and 30-inch tubes in 1951, but not many. Engineers consider these big tubes just as safe as the 10-inchers, judging from the way in which the big tubes were displayed without safety glass at the IRE Convention.

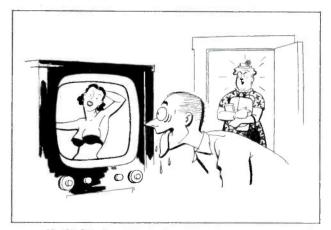
SHOCKING NEWS. Yearly death toll from homemade devices for electrifying fence wires is estimated at 10 to 15 humans and hundreds of animals. Commonest (and probably most lethal - Ed.) scheme used by novices is connecting ungrounded side of 120 volt AC power line to the fence through fuses or electric lights. This arrangement works fairly well during ordinary weather conditions, works poorly in dry weather, and is a killer during good grounding conditions such as after a rain. On the other hand, the approved commercial electric fence controller is today a safe, effective, and economical aid to modern farming. If you come across a homemade installation in your servicing travels, warn the owner tactfully that his system can become as deadly as the electric chair. Keep your customers alive,

ALMOST PERFECTION. In the famous RCA Laboratories at Princeton is one television transmitter-receiver combination that's as perfect as today's dollars and engineering brains can make it. If you hold an 8×10 photo in front of the camera and take an 8×10 photo of the picture on the receiving screen, you can't tell the two pictures apart at arm's length. The equipment uses U. S. standards for picture lines and frames. This proves that we've got the right standards now, and need to concentrate only on improving the equipment.

HABIT. Chimney-strap TV antenna installations can get to be a habit, or maybe some would call it a disease. Seems as if once an installation crew has made a few, it favors that mounting despite its many drawbacks, and goes to great lengths of roof-ridge walking and rooftop gymnastics to put up a chimney job. Check your installation boys and keep score for a while.

WOOD MASTS. In the good old days of radio, wood was the only thing accepted for outdoor antenna masts by customers. Not so today, patriotic servicemen learned to their sorrow when good-quality aluminum or steel masts became scarce. And that's why the wood masts at your jobbers have a shiny coat of aluminum paint; once up, they look like the metal jobs on neighboring homes and last just as long, so there's no complaint. The wood should be ash, oak, or an equally strong wood, though; none of this pine or fir clothes-hanger rod if you want a trouble free job. To a customer complaining about getting a wood mast, one serviceman explained: "We use only the best piece of ash in the cord."

HOMEWORK. Latest figures on per cent of television receivers fixed in homes on first call are 80% for one large manufacturer's service organization that takes in service contracts on their make of set, and 95% for another manufacturer who sends men out only on demand-service calls that for some reason or another cannot be handled by ordinary service organizations. The latter figure is achieved by using top-grade technicians who are almost fullblooded engineers. Each has a delivery truck in which he can cram all the spare parts he wants. Some of these boys come into the office only about once a week, to replenish their stocks and collect the pay check. Each morning they get the day's quota of calls by phone. The chap who gets stumped and has to bring a chassis into the shop gets a lot of friendly advice while troubleshooting with shop instruments, and a lot of ribbing if the trouble turns out to be some simple defect he overlooked in the field. Such high morale among technicians is a real credit to their service manager.



V-NECK BATTLE ON TV. How low can a Vneck gown get on television? Engineers attending the annual IRE Convention in New York City saw the answer on one picture-tube screen - all the way. To demonstrate its new flying-spot picture generator, Telechrone, Inc., used a Hollywood pretty-girl art slide (for artists only; saves costly model fees) in front of the photomultiplier pickup tube, and did a beautiful job of stopping traffic. Incidentally, for less than a thousand bucks you can get the complete equipment along with a low-power oscillator. It's fun to dream about what the next-door neighbor would say if he tuned in that picture - and what his wife would say if she caught him looking at it!

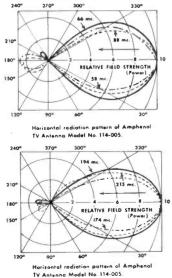


OUTSTANDING MECHANICAL SPECIFICATIONS

Part	Materiol	Yield Strength	Si	z e
		psi	o.d.	Woll
Mast (galv.)	¾" Thinwall Steel Conduit	32,000	0.922"	.049"
Lorge Folded Dipole	35 1/2 H AI.	19,000	.500"	.049"
Small Folded Dipole	35 1/2 H AL	19,000	.375"	.049"
Reflector	35 1/2 H AI.	19,000		.049"
Crossorm	35 H AI.	26,000	.87.5"	.065"
Center Support & T Costing	Al, Alloy 45,000 psi tensile strength			

EXCELLENT RADIATION PATTERNS

These are the radiation patterns of the AMPHENOL Inline antenna at 58 mc., 66 mc., and 88 mc., in the low band, and 174 mc., 194 mc., and 215 mc. in the high band. Notice the uniformity of these lobes at all frequencies. The lack of lobes off the sides and negligible ones off the back maintains high front-to-back and front-to-side ratios necessary for the rejection of various interferences. The



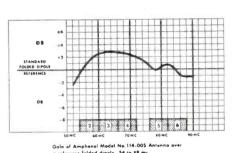
presence of a single forward lobe is usually a very desirable feature, especially when it is wide enough to provide adequate interception area for some differences in transmitter location, changes in the wave front's direction of travel, or physical movement of the antenna in high winds. Furthermore, it is not too critical of orientation. It is necessary only to aim it and forget it.

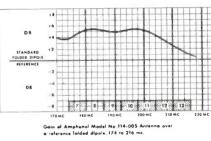
HIGHER GAIN

These gain curves of the AMPHENOL Inline antenna represent the intercepted voltage of the AMPHENOL Inline Antenna as plotted against the intercepted voltage of a reference folded dipole cut to the frequency being compared. There is no channel in either the low band or high band where there is more than a three decible change within the channel that can cause picture modulation or "fuzziness." Gain of the AMPHENOL Inline antenna is quite flat over all channels.

You will find more gain designed into the high band because of greater need for it, due to higher losses at these frequencies. Also, notice the drop-off on channel six. This is at the edge of the FM band and is subject to FM interference, so the Inline's gain is purposely held down at that frequency.

The excellent broadband characteristics, impedance match, single forward lobe radiation patterns on all channels, maximum gain, lightning protection, and superior mechanical features of the AMPHENOL Inline Antenna make it the antenna for greatest TV picture quality!





YOURS FOR THE ASKING

Send for "The Antenna Story" — a sincere discussion of TV antennas based on actual field tests.





WALTER R. JONES

Associate Professor of Electrical Engineering, Cornell University



AC RECEIVER POWER SUPPLY PROBLEMS

Power Supply Impedance Considerations

Have you ever had an AC receiver brought in for servicing that had that easily recognized sickish smell of burned insulation? Usually when this happens the power transformer has become very hot and may even have been permanently damaged. Further examination may indicate a shorted rectifier tube and a ruined first electrolytic capacitor. A survey should be made initially to determine the extent of the damage. A further study should identify immediate cause of the overload, and replacement of the defective part or parts should be made. Undoubtedly most of us would stop here, if the transformer were still capable of satisfactory operation. However, there is a further point which should receive consideration to limit damage in the event of future overloads.

Examination of the circuit diagram in Sams' Photofacts may give a clue. If the resistance from one plate pin of the rectifier tube to the other is not roughly twice that shown in Table I for the rectifier tube employed, then it is highly probable that too low an impedance in the plate circuit would be responsible for the total destruction of the transformer, tube and filter capacitor.

Figure 1 indicates how the required information is given in Photofact Folders. Figure 1(a) shows a schematic of a circuit taken from Set 110, Folder 12, while Figure 1(b) shows the resistance chart. The illustrations indicate several things:

1. The center tap of the transformer is grounded.

2. The rectifier tube is a type 6X5GT.

3. Reference to 1(b) shows resistance from plate pin No. 3 to ground is 95 ohms and that from plate pin No. 5 to ground is also 95 ohms.

4. Table I indicates that rated impedance for each plate of the 6X5GT is 150 ohms. The resistance values shown represent pretty closely the total impedance in the circuit. (Calculated or measured impedance values would vary slightly from DC re-

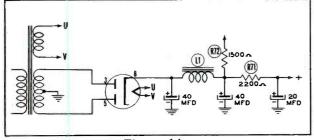


Figure 1A

Item	Tube	Pin 1	Pin 2	Fin 3	Pin 4	Pin 5	Pin ó	Pin 7	Pin 8	Pin S
V I	6AB4	##22KΩ	012	14.20 4.50	DΩ	00	ΩΟ	1000		
¥ 2	6CB6	100	3500	00	‡▲.2Ω #.5Ω	13.5KA	13.5KG	00		
¥ 3	12A T7	ts Inf.	047	2.4KΩ	00	00	†‡inf. †#3.1KG	2.2 Meg.	2.460	‡▲.2Ω ∉.5Ω
¥ 4	IZAT7	13.5KB	3360	0Ω	00	013	f #12KG † a Inf.	1.5 Meg.	330Ω	†▲.2Ω #.5Ω
¥ 5	6CB6	1 Meg.	2200	1.50	002	12.9KD	12.9KR	00		
¥ 6	6CB6	11.10	* 2200	‡ ▲ .2Ω ♦.5Ω	00	* * 2.8KD	# † 2.8KΩ	0Ω		
¥ 7	9001	*100KD	012	1 4 .2Ω 4.5Ω	00	+ 117KG	1 1 17Kg	00		
¥ 8	9001	\$130 Kg	-8	1 A .2Ω 0.5Ω	00	1 1 28KG	1 7 28KD	00		
V 9	BALS	00	100KG	4.50	00	* 200KG	00	100KD		
¥ 10	6A V6	2.2 Meg.	00	2560	25KO	470KΩ	1.6 Meg.	1220KQ		
V 11	12AU7	125KD	6801613	2.980	25KΩ	25KO	11360	470KD	94053	25KΩ
v 12	6X5GT	Ind.	25KO	950	Inf.	950	int.	25KQ	† Δ 16KΩ Φ60KΩ	

Figure 1B

sistance readings, but for the purposes outlined in this article the measured resistance values can be employed satisfactorily.)

5. The 95 ohms per plate is considerably less than the rated value of 150 ohms so that in the event of a short circuit in the power supply, very high currents could flow, resulting in damage to both tube and transformer.

If the transformer is still good, it might be desirable to supply two plate resistors as shown in Figure 2. Care must be taken so that these resistors, which will run pretty hot in service, are located so that they will not cause damage. If the transformer must be replaced, it would be advisable to select one having the proper resistance in the windings, thus eliminating the necessity of adding the plate resistors.

Frequently one wants to change over a receiver which has employed a 5Y3GT tube to a 6X5GT tube. Examination of Table I indicates that the required impedance for 5Y3GT tube is 50 ohms per plate while that required for the 6X5GT is 150 ohms per plate. If one desires to have trouble free results, the plate resistors must be added if the same transformer is to be used unless, of course, the resistance of the windings is reasonably close to 150 ohms per side.

Rectifier Substitution Factors

When given rectifier tube types are not available, it is often necessary to substitute another type in order that the receiver may be kept in operation.

The problems which arise in connection with the substitution of one type for another may be considered as listed below.

1. Filament requirements - voltage and current.

2. Voltage drop - in the tube itself and hence the resulting voltage applied to the filter capacitors.

• • Please turn to page 60 • •



Keyed AGC Application by ROBERT B. DUNHAM and W. WILLIAM HENSLER

A procedure for adding Keyed AGC to the 630-type chassis.

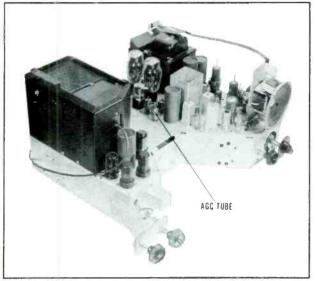


Figure 1. Type 630 Chassis with AGC Tube Mounted.

The application of keyed AGC to the 630-type chassis requires the addition of a 6AU6 tube, an AGC winding on the width coil and minor circuit changes. All the required components are standard replacement items with the exception of the width coil with an AGC winding. Some manufacturers are planning to make available a separate coil which may be slipped over the original width coil. If this type of coil can be obtained, the original width coil need not be replaced. When it is necessary, however, to use a new width coil with an AGC winding, it is imperative that the new width coil have the same inductance range as the original. The new coil should be mounted on the same bracket as the original.

A schematic of the original contrast and video circuits affected in this conversion, is given in Figure 2. Reference may be made to this schematic while rewiring the receiver.

The first step is the mounting of the socket for the AGC tube. Figure 1 shows the chassis with this

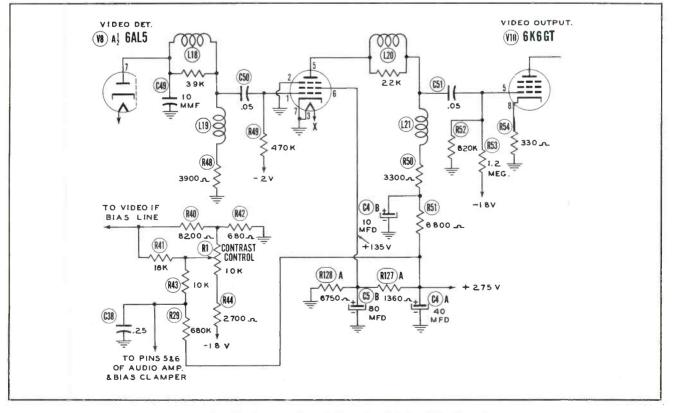


Figure 2. Original Contrast Circuit of Type 630 Chassis.

Frank J. Moch says-"there is no other OSCILLOSCOPE like the NEW Simpson MODEL 476

FRANK J. MOCH, president of the National Alliance of Television an-Electronics Service Associations.

MIRROSCOPE"

new and completely advanced type of oscilloscope - Model 476 MIRROSCOPE - is designed to eliminate certain inherent disadvantages found in the conventional type of oscilloscope by use of the "Mirroscope principle." In this kind of construction the 5-inch cathode ray tube is mounted in a vertical position, thus reducing bench space requirements to an area of only 9" x 8" thereby permitting better concentration of associated equipment for any type of test procedure. The cathode ray image is reflected from an optical type front surfaced mirror mounted in the adjustable cover at the top of the cahinet bringing the viewing surface of instrument near eye level when instrument is used on benches of normal height. The mirror angle is quickly and easily adjusted to any position of the operator. The cover with integral side wings forms an effective shield against external light sources or may be closed down for protection of the tube and mirror when the instrument is not in use. The upright construction permits location of controls and connections for maximum convenience and allows for internal cathode ray tube connections at the front of the panel instead of the rear.

INPUT IMPEDANCE:

SENSITIVITY: Vertical direct......12 volts rms per in. Vertical amplifier.20 millivolts rms per in. Horizontal direct......14 volts rms per in. Horizontal

Vertical direct....10 megohms, 15 mmf. Horizontal direct...10 megohms, 15 mmf. Vertical amplifier.300,000 ohms, 30 mmf. Horizontal

amplifier...... 500,000 ohms, 15 mmf.

Horizontal trace expansion is over 4 times tube diameter. This makes it possible to examine minute portions of a response pattern for finer detail.

Linear Sweep frequency is continuously adjustable in five overlapping ranges from 15 cycles to 60,000 cycles. Internal, external or line frequency synchronization with variable amplitude is available.

Means for intensity or "Z axis" modulation is provided. Approximately 14 volts peak will blank a trace of normal intensity.

The vertical amplifier frequency response is within 3 DB from 20 cycles to over 300,000 cycles and is usable to well over three megacycles. Square wave slant and over-shoot is held to less than 5 per cent of amplitude. This response will be found adequate for all phases of television receiver service including observation and diagnosis of Sync. signals.

TUBE COMPLEMENT:

	5UP4	Cathode Ray Tube.	LINE VO
	4—6J6	Horizontal and Vertical Am- plifiers.	cycles. SIZE: Heig
S. Martin	1-12AU7	Vertical pre-amplifier,	8″ over WEIGHT: Ibs.
ALC: NO	1-6J6	Linear Sweep oscillator and Sync. injector.	Hight Fre
	2-6X4	High voltage rectifiers.	operato
7		and the second	and a start

LINE VOLTAGE: 105-125 volts, 50-60 cycles.	A REAL
SIZE: Height 16¼"; Width 9½"; Depth 8" over all	
WEIGHT: 25 lbs.; Shipping weight 30 lbs. Hight Frequency Crystal Probe\$7.50	Marrie Care
DEALERS NET PRICE including operators manual\$179.50	
A DECEMBER OF THE PARTY OF THE	





SIMPSON ELECTRIC COMPANY 5200 W. Kinzie St., Chicago 44, III. Phone Cc embus 1-1221 In Canada: Bach-Simpson Ltd., London, Ortaria tube mounted in place. The tube is positioned near the video detector and the video amplifier tubes, V8 and V9, so that the length of the connecting leads may be kept to a minimum. After mounting the new tube socket, the following circuit changes were made.

The coupling from the video detector to the first video amplifier was changed to direct coupling by removing C50 and R49 (see Figure 2), and connecting the junction of L18 and L19 directly to pin 1 of V9. R48, the video detector diode load resistor, was changed from 3900 to 4700 ohms. Refer to Figure 4 for placement of parts. The bias on the video amplifier is now derived from the rectified signal or from the rectified noise when no signal is present.

In the contrast circuit, R40, R41, R42, R43, and R44 were removed. These components are located near the contrast control at the front of the chassis. Removing these components leaves the contrast control free from all connections. Pin 6 of V9 was then disconnected from the plus 135 volt line and connected to the arm of the contrast control, and bypassed with a .1 mfd. capacitor (C106). A 10K ohm resistor (R134) was added from the low side of the control to ground and the high side of the control connected to the plus 135 volt line.

In the plate circuit of V9, R51 and the connection to C4B were removed. The low side of R50 was then connected to plus 135 volts.

The next step was the connecting of the keyed AGC tube in the circuit. Pin 3 was grounded and pin 4 was connected to the filament line. R135, a 47K ohm isolation resistor, was connected from pin 1 of the AGC tube to the junction of L21 and R50. Pins 2 and 7 were connected to the plus 135 line. Pin 6 was bypassed to the plus 135 line with a .01 mfd. capacitor (C103), and was also connected to the plus 275 line through a 27K ohm resistor (R136).

Because of the length of the lead required to connect the AGC winding to the keyed AGC circuit, a shielded cable was used. This prevents the coupling of the sharp pulses present to the other circuits. Amphenol 21-138 microphone cable, or similar type, can be used in this application. In order that the phase relationship of the pulses fed to the AGC tube are correct, it may be necessary to reverse the connections on the AGC winding after the set is put in operation. The center conductor of the cable was connected to pin 5 of the AGC tube. The shield portion of the cable was connected to chassis through two 100K ohm resistors (R139 and R140) which were bypassed by C104. The junction of R139 and R140 was bypassed by C105 and connected to the video IF bias line where the original contrast bias voltage was applied. A lead was then connected from the junction of C104 and R140 to a terminal strip which was added near the contrast control. R137 was then wired from the termination of this lead to pins 5 and 6 of the bias clamper tube. The end of R29 which connects to the terminal strip was unsoldered and a 4.7 meg. resistor (R138) was added in series with it. The other end of the 4.7 meg. resistor was then connected to the plus 135 volt line as shown in Figure 3. Actually R29 may be removed from the circuit but due to the close spacing of components around it, it is much easier to merely connect R138 in series with it. The net result is the same.

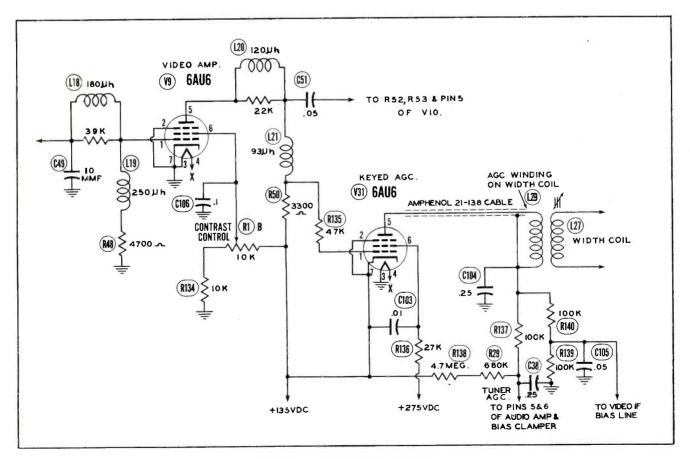


Figure 3. Keyed AGC Circuit.



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- Popular types in quantities correctly proportioned as determined by popularity of sales for replacement.
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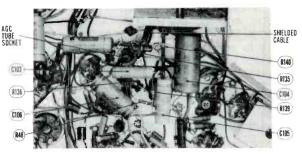


Figure 4. Bottom View of Converted Chassis.

On some models the bias line to the tuner was grounded. If this is the case on the particular model which is being converted, this line should be removed from chassis and connected to pins 5 and 6 of the bias clamper tube.

This completed the circuit changes required for the addition of the keyed AGC circuit. The set was then turned on and checked for operation. If the circuit fails to operate properly, it may be necessary to reverse the connections on the AGC coil as previously mentioned.

Although this circuit is directly adaptable to the 630-type chassis, it may be incorporated in many other models with certain modifications. The prime requisite, however, is that a positive going video signal be present in the plate circuit of the video amplifier.

ENGINEERING VIEWPOINT

A group of engineers were coming home from a party one night. They stood in front of the house of one of their number and called for the father.

"Will you please do ush a favor?" asked one.

"What do you want?" replied the father.

"Will you please come out here and pick out Jimmie so the rest of ush can go home."

- The CORNELL ENGINEER



"Now that all our neighbors have their own TV Sets, let's get that out of here."

Experiments in Audio ____

by James R. Ronk

SQUARE WAVE GENERATION

A simple device for converting a sine wave signal into a square wave for general audio experimentation.

EXPERIMENTS IN AUDIO

The last few years have witnessed a widening interest in the use of visual testing and indicating methods. Television servicing requirements have undoubtedly been responsible for the major impetus in this direction, but even prior to its mass adoption, there had been a steady trend in the audio field toward visual performance checks as compared to the highly fallible listening tests. Aural impressions too frequently combine individual hearing deficiencies and personal taste, to ever contribute much to acceptable standards.

The general worth and utility of the square waveform in audio system evaluation has been well established. Properly used. and with adequate accessories, it can readily identify frequency response, phase shift, and to some extent transient behavior. In other words, it provides a rapid comprehensive picture of the amplifier system.

This article deals with a method of obtaining square waveforms from a sine wave source. Most service shops have such a source available in their standard servicing equipment. With the unit to be described, the audio output of the source need not be great, say in a range up to 5 or 6 volts, and can have single, multiple or completely variable frequency.

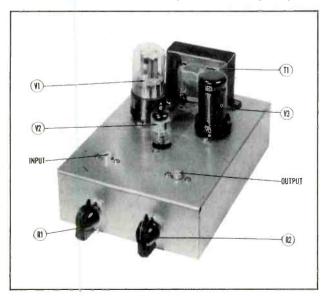


Fig. 1. Clipper Unit, Top Chassis View.

It might be wise to first review the design of the unit and then, with operating principles in mind, study its general application.

The square wave generator, or rather, clipper, is pictured in Figures 1 and 2, and the schematic diagram appears in Figure 3.

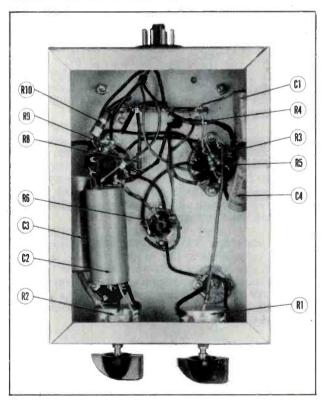


Fig. 2. Clipper Unit, Bottom Chassis View.

As indicated in these illustrations, the design does not include a self-contained power supply since our own application was in conjunction with already existing chassis and rack assemblies containing a master supply. However, power supply voltages are not critical. Standard values of 6.3 volts @ 1.2 amperes for the heaters and 275 volts @ approximately 40 milliamperes for the plate supply were used on the original.

The plate supply voltage should be well filtered since appreciable hum voltage will make comparison

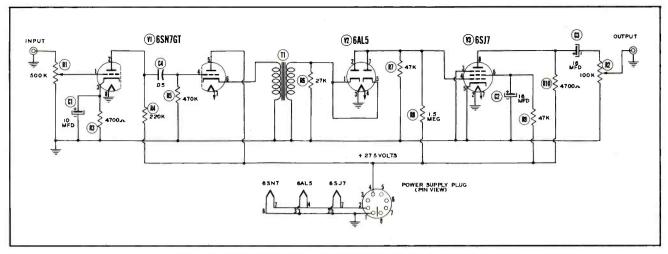


Fig. 3. Schematic Diagrams of Square Wave Clipper.

of waveforms difficult, especially at low audio frequencies.

One-half of the type 6SN7GT (V1) is used as a conventional voltage amplifier stage. Sine wave voltage input to this stage is controlled by R1. With component values as shown, the stage has an approximate gain of 17.

The remaining half of V1 is employed as a cathode follower which, through transformer T1, couples the output of the voltage amplifier to the diode rectifier.

An earlier design with which we experimented had the 6SN7GT connected as a combination voltage amplifier-rectifier stage; however, the low impedance load of the diode circuit affected the voltage amplifier to such extent that it was impossible to develop proper rectified signal for the clipper stage, with reasonable input voltage to V1.

Transformer T1 is an interstage unit having a primary to secondary ratio of 1 to 2.6. The unit employed was selected from several similar types on hand, because the DC resistance of the primary winding of 1120 ohms provides the proper bias to limit plate current to the 10 milliampere maximum recommended.

The shunt resistor R6 across the transformer secondary is included to prevent winding resonance from causing spurious "pips" or traces in the waveform. Its value, 27K ohms in this instance, should probably be arrived at by cut and try with the individual unit chosen.

Resistor R7 - 47,000 ohms - acts as the diode load, and Resistor R8 - 1.5 megohm - serves to clamp the grid of the clipper (V3) to prevent "bounce" which might occur as it returns to the zero point. Note that the grid of V3 is directly coupled to the diode output and that this is the bias source for V3. The 6SJ7 clipper tube, with a nominal 6 or 7-volt cutoff, is driven far beyond the cutoff point on the negative excursion, with the result of square wave formation in the plate circuit.

The clipper output is coupled to the 100K ohm output control by electrolytic capacitor C3. The 16

mfd. capacity of C3 is not actually required – lower values (down to 5 mfd.) will do, but these units were at hand and worked satisfactorily.

Immediately following are some measurements taken during typical operation:

Frequency	100 Cycles	1000 Cycles
Sine Wave Input (Volts)		
(1st Grid, Pin 1 of V1)	4.0	3.4
Signal Voltage - 2nd Grid		
(Pin 4 of V1)	71	61
Signal Voltage-Cathode		
(Pin 7 of V1)	<u>~ 61</u>	52
Signal Voltage-Diode		
(Pins 1, 5 of V2)	100	89
Maximum Square Wave		
Output (Volts)	75	71

This table shows that the square wave output is sufficient for single stage analysis in conventional amplifiers.

As a general consideration, the greatest steepness of wavefront occurs with maximum sine wave signal amplitude; however, it is possible with this device to reach an overload point. Consequently, R1 should be adjusted for maximum symmetry of square wave output.

Before proceeding to application, we would like to say that no originality is claimed for the design of this clipper unit. We investigated several designs and found that, for the application which involves comparatively low sine wave voltage availability, the circuit shown here has the best operation and possibilities.

For those who are interested in experimenting with this clipper, a Parts List is included at the end of this article.

As previously mentioned, the clipper described is intended for use in the audio field. In actual use here, we had highly satisfactory results with input frequencies ranging from 100 to 10,000 cycles. Most audio square wave experimental work is done at either 400 or 1,000 cycles. It is possible, because of the richness of harmonic content of the square wave,

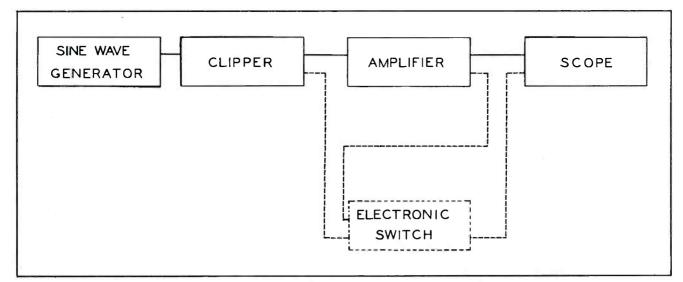


Fig. 4. Equipment Setup for Use of the Square Wave Clipper.

to observe extended frequency response, using only these two frequencies; however, for what it may be worth in future experiments, this clipper unit can provide extremely wide-range fundamentals, as the above frequency range shows.

APPLICATION

Figure 4 shows a representative setup of equipment for use in square wave analysis of amplifiers. As shown, the sine wave generator is fed into the clipper stage, the output of the clipper is applied to the amplifier, and the amplifier output is observed on the scope. The dotted portion of the figure represents optional use of an electronic switch. If at all possible, the electronic switch should be included, since it provides the basis for instantaneous comparative performance of differing or successive stages, or the ability to view the square wave output to the amplifier, and the amplifier output, simultaneously.

A word or two about the use of the scope for interpretation of the wave forms might help a little. First, keep the waveform limited to the center area of the scope. If the scope should have any nonlinearity in either the horizontal or vertical direction, it will tend to distort or amplify any variations from the square waveform, if they occur at the extreme range of the scope. Also, in observing the square waveforms, always have a minimum of two complete cycles for accurate observation.

Figure 5 is an illustration of an ideal square waveform, and the output of the clipper should very closely approximate this figure.

Figures 6 through 9 show waveforms obtained through a representative amplifier.

Figure 6 indicates poor high-frequency response of an amplifier, while Figure 7 shows an improved high-frequency response.

Figure 8 is indicative of poor low-frequency response, while Figure 9 shows an improvement in the low-frequency characteristics.

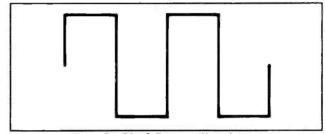


Fig. 5. Ideal Square Waveform.

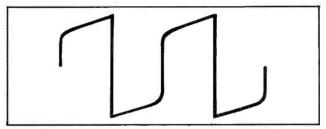


Fig. 6. Amplifier with Poor High Frequency Response.

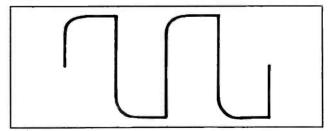


Fig. 7. Amplifier with Improved High Frequency Response.

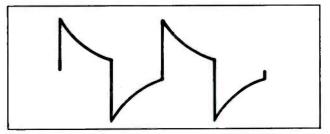


Fig. 8. Amplifier with Poor Low Frequency Response.

Please turn to page 61

VIDEO IF AMPLIFIERS

♦ Continued from page 9

are identical but they have sufficient range to obtain resonance from 22.8 to 26.3 mc. Each of the chokes are tuned to the different frequencies which are marked on the schematic.

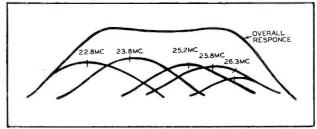


Figure 2-7. Overall Response Curve of Stagger-Tuned IF Amplifier.

Figure 2-7 illustrates the overall bandpass that is obtained through stagger tuning. The five resonant curves represent the response curve of each of the tuned chokes in the circuit. These curves are not drawn to exact scale but serve to show how the sum of the individual response curves are additive. The variation in gain is due to a difference of loading on the chokes.

Another type circuit, which has some advantages over the capacitively coupled circuit of Figure 2-6, employs dual winding transformers that are single tuned. By employing inductive coupling of this type, the gain of the stages can be controlled by varying the spacing between the windings or by changing the turns ratio itself. The response curve of this type of transformer is essentially the same as the tunable choke just described. Either the primary or secondary, or both, can be shunted to obtain the desired effective Q, thus controlling gain and bandpass.

Another advantage of this type of coupling is the low DC resistance in the grid circuit and the low capacity between stages. When the signal is capacitively coupled to a grid having a resistive load, the discharge time is much longer than the charge time, which may cause white dots or streaks to appear in the picture after noise pulses.

In order to better understand what causes this, refer to Figure 2-6 and assume that a noise pulse is received at the grid of the second IF amplifier. On the positive half of the signal the noise pulse may be strong enough to cause grid current to flow, charging

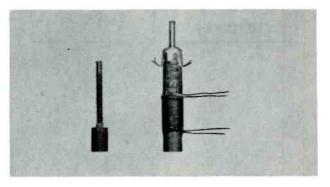


Figure 2-9. Dual Winding Transformer Having Single Tuning.

the coupling capacitor with an excess of electrons on the right side. After the noise pulse, the capacitor will discharge and the electron flow will be down through the grid resistor which will decrease the gain of the stage for an instant or may even momentarily drive the grid to cut-off. Since the discharge current of the coupling capacitor is in the reverse direction from the noise pulse, which corresponds to a "blacker than black" signal, a white dot or streak follows the noise pulse. This is especially noticeable in cases of ignition interference and in weak signal areas where there is a poor signal-to-noise ratio, resulting in "snow." Some manufacturers have placed a small coil across the grid resistor to reduce the DC resistance in the grid circuit. These coils are sometimes referred to as "grass cutters" since they tend to decrease the "snow" or "grass" in the picture.

The circuit of Figure 2-8 shows a typical video IF strip using dual winding transformers with single tuning. Three stages are used and the four IF transformers are stagger tuned to provide adequate bandpass. The grids of the first two stages are returned to the AGC line to control the gain of the IF strip. Since a change of gain of these stages causes the input capacity to vary, due to the Miller effect, an unbypassed 47 ohm resistor is placed in the cathode circuit to introduce a small amount of degeneration. This degeneration decreases the input capacity of the stage which results in a smaller change in input capacity as the gain of the stage is varied. The grid of the third stage is returned to ground and since its gain is not varied, the input capacity will remain con-

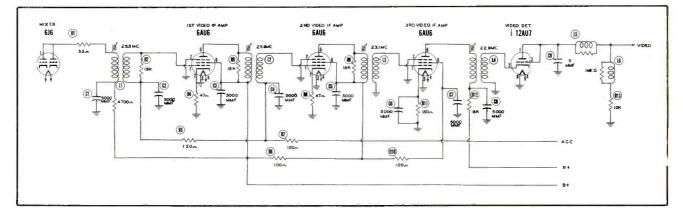


Figure 2-8. Complete Stagger-Tuned Video IF Strip Using Dual Winding Transformers with Single Tuning.

stant. This makes it possible to bypass the cathode resistor and realize more gain from the stage. The value of the cathode resistor is increased to provide adequate bias for the stage. Figure 2-9 illustrates a dual winding transformer representative of the type used in this circuit.

The alignment of either of the circuits shown in Figures 2-6 and 2-8 is comparatively simple. Each of the alignment frequencies is fed to the mixer while the appropriate coil is adjusted for maximum output at the detector load. The overall response curve can then be checked by the use of an oscilloscope and sweep driven generator. If the circuit is normal, only slight touch-up adjustments are required after checking the overall response curve.

The pre-alignment instructions should always be read carefully so that the proper bias can be applied to the circuits. In some cases, even though all components are okay, oscillation will take place in the IF amplifier. This is usually caused by misadjustment of the coils so that adjacent stages are tuned to the same frequency. In most cases readjustment will stop the oscillation, but should it continue, it may be necessary to inject the signal near the end of the IF strip and remove the tubes ahead of that point. After the adjustment of that stage is made, the tube in the stage ahead is replaced and the signal is injected at this point. Adjustment can then be made on this stage. By progressing forward in this manner, the IF strip can be aligned to obtain stable operation. In receivers having series filament operation, the coils can be shunted which will accomplish the same thing as far as oscillation is concerned.

Another method of obtaining wide bandpass is through the use of double tuned, over-coupled transformers. By closely spacing the primary and secondary windings, and tuning both to the same frequency, a double humped response can be obtained. Figure 2-10 shows a partial schematic of a stage employing a double tuned over-coupled transformer along with the response curves that are obtained from such a circuit. By tuning both primary and secondary to the resonant frequency, the band width can be varied by changing the spacing, or coupling, between the two windings. The three response curves shown are obtained with variations in coupling. The response that is obtained with loose coupling, less than critical, is similar to that of the single tuned choke. If, however, the spacing between the windings is decreased, the response curve starts to flatten at the top as shown in Figure 2-10. By further increasing the coupling, the bandpass increases but there is a dip in the curve at the resonant frequency. In some cases two such stages are used along with additional single tuned stages which are so aligned to fill in the dip in the response curve.

One of the advantages in the use of overcoupled transformers is the fact that the gain of the stage can be controlled so that any tendency for oscillation can be prevented. Also by referring to Figure 2-10, it can be seen that the output capacity of one stage and the input capacity of the following stage are separated by the transformer, since the two capacities are across the primary and secondary, respectively. Thus the inductance of the windings of the transformer can

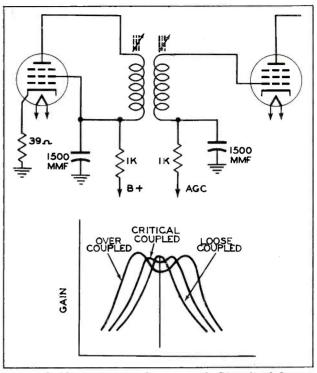


Figure 2-10. Response Curve and Circuit of Overcoupled Stage.

be made greater than the single winding tunable choke, since in this case the input and output capacities were in parallel, requiring a lower value of inductance to obtain resonance.

In some instances, a small capacitor, .5 to 5 mmf., may be added between plate and grid to provide a small amount of high side coupling to hold up the gain at the high end of the pass band.

When aligning over-coupled IF stages, a special procedure must be followed to assure proper alignment. Several methods are employed, but the most popular is through the use of a sweep signal generator and oscilloscope. With this method the response curve can be checked as alignment progresses, and any deviation from the desired results will show up immediately.

One procedure for aligning over-coupled IF stages calls for the adjustment of the transformers progressively starting from the last stage and working toward the mixer stage. The scope is connected to the video detector output and the signal is injected into the IF grids, progressing forward in the circuit. At each point, adjustments are made to obtain the desired response curve at that point. If any components are defective, they will show up in the ability to obtain the proper response at that stage. By progressing forward in the IF strip, the final response curve is obtained.

Another procedure for the alignment of overcoupled IF stages, is the adjustment of each stage separately through the use of a detector probe. The signal is injected at the grid of the stage and the response curve of that stage can be seen by connecting an oscilloscope, through a detector probe, to the

◆ Please turn to page 53 ◆ ◆

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ADMIRAL		Tel. Rec	(See Ch. 6A2)103	Rec. (See Ch. 21FI) 127-1A	27K17 Tel. Rec.
Chassis UL5K1	_	Chassis 21F1, 21G1	Model 6C11 (See Ch. 6C1) 53	Models 19A115, 19A11SN,	(See Ch. 21F1) 127-1A Models 27K25, 27K26,
(See Chassis 5K1).	30	Tel. Rec	Model 6C71 (See Ch. 10A1) 3 Models 6F10, 6F11, 6F12 *	19A12S, 19A12SN, 19A15S, 19A15SN (See	27K27 Tel. Rec.
Chassis UL7C1	25	Rec. (See Ch. 21B1)118	Model 6P32 (See Ch. 6E1,	Ch. 19A1} Tel. Rec 59	(See Ch. 21F1)127-1A
(See Chassis 7C1). Chassis 3A1	2-24	Chassis 21K1, 21L1 Tel.	6EIN) 6	Models 20X11, 20X12 Tel.	Models 27K35, 27K36
Chassis 3C1 (See Ch.		Rec. (See Ch. 21F1) 127–1A Chassis-21P1, 21Q1 Tel.	Models 6Q11, 6Q12, 6Q13, 6Q14 (See Ch. 6Q1) 78	Rec. (See Ch. 20X1)100 Model 20X122 Tel. Rec.	Tel, Rec. (See Ch. 21F1) 127 -1A
20T1)		Rec. (See Model 21F1) . 127-1A	Model 6R11 (See Ch. 6R1) 54	(See Ch. 20X1)100	Models 27K46 Tel. Rec.
Chassis 3C1 Prod. Chg Bul. 15	126_1	Chassis 24D1, 24E1, 24F1,	Model 6RP48, 6RP49,	Model 20X136 Tel. Rec.	(See Ch. 21F1) 127-1A
Chassis 4A1	3-31	24G1, 24H1 Tel. Rec103—2 Chassis 24D1, 24E1, 24F1,	6RP50 (See Ch. 3A1) 2 Models 6RT41, 6RT42, 6RT43	(See Ch. 20X1) 100 Models 20X145, 20X146,	Models 29X15, 29X16, 29X17 Tel. Rec.
Chassis 4A1 Chassis 4B1 Chassis 4D1	24-1	24G1, 24H1 Tel. Rec.	(See Ch. 5B] Phono) 4	20X147 Tel. Rec.	(See Ch. 24D1)103
Chassis 4D1	49—1	Prod. Chge. Bul. 9114-1	Model 6RT41A, 6RT42A,	(See Ch. 20X1)	Models 29X25, 29X26,
Chassis 4H1 (See Chassis 30B1).		Chassis 30A1 Tel. Receiver 57-2	6RT43A (See Ch. 5BIA) 18	Model 22X12 Tel. Rec.	29X27 Tel. Rec. (See Ch. 24D1)
Chassis 4J1, 4E1		Chassis 30B1, 30C1, 30D1 Tel. Rec 71-2	Model 6RT44 (See Ch. 7B1) 18 Models 6S11, 6S12	(See Ch. 20X1)100 Models 22X25, 22X26,	Model 29X25A Tel. Rec.
(See Ch. 2041) Chassis 4L1, 451		Model 4D11, 4D12, 4D13	(See Ch. 651)	22X27 Tel. Rec.	(See Ch. 21B1)118
Chassis 4L1, 451 Chassis 4R1	109 1	(See Ch. 4D1) 49	Model 6T01 1—19	(See Ch. 20X1)100	Models 30A12, 30A13
Chassis 5B1		Models 4H15, 4H16, 4H17	Model 6T02, 6T04 1-20 Model 6T05 (See Ch. 6A1) 1	Models 24A11, 24A12 Tel. Rec. (See Ch. 20A1) 77	(S or SN) Tel. Rec. (See Ch. 30A1) 57
(See Model 6T02).		{A or B} Tel. Rec. {See Ch. 20A1} 77	Model 6105 (See Ch. 6A1) 1 Model 6106, 6107	Madel 24A125 Tel. Rec.	Models 30A14, 30A15,
Chassis 5B1 Phono Chassis 5B1A Chassis 5B2	4-24	Models 4H15, 4H16, 4H17,	(See Ch. 4A1) 3	(See Ch. 20A1) 77	30A16, Television Re-
Chassis 582	100-1	4H18, 4H19 (S or SN)	Model 6T11	Model 24A125AN Tel. Rec.	caivers (See Ch. 30A1). 57 Models 30B15, 30B16,
Chassis 5D2		Tel. Rec. (See Chassis 30B1)	(See Model 6T02) 1 Model 6T12 (See Ch. 4A1) 3	(See Ch. 20X1)100 Models 24A126, 24A127	30B17 (S or SN)
(See Ch. 21B1)		Models 4H18, 4H19 {C or	Models 6V11, 6V12	(See Ch. 20A1)	Tel. Rec. (See Ch. 30B1) 71
Chassis 5F1 Chassis 5H1	57-1	CN) Tel. Rec. (See Ch.	(See Ch. 6V1)	Tel, Rec 77	Models 30C15, 30C16,
Chassis 5K1	301	20A1} 77	Models 6W11, 6W12 (See	Models 24C15, 24C16, 24C17 Tel. Rec.	30C17 (S ar SN) Tel. Rec. (See Ch. 30B1) 71
Chassis 5N1		Models 4H115, 4H116, 4H117 (S or SN)	Chassis 6W1)	(See Ch. 20A1)	Models 30F15, A. 30F16.
Chossis 5R1	59-1	Tel. Rec. (See Ch. 30B1) 71	Chassis 6Y1)	Models 24R11, 24R12 Tel.	A, 30F17, A Tel. Rec.
Chassis 5T1 Chassis 5W1	79-2	Models 4H126A, B, C or	Model 7C60B, 7C60M,	Rec. (See Ch. 2011)117	(See Ch. 20A1) 77 Models 32X15, 32X16 Tel.
Chassis 5X1	76-3	CN Tel. Rec. (See Ch.	7C60W (See Ch. 681) 48 Model 7C61, 7C62, 7C62UL	Models 24X15, 24X15S, 24X16, 24X16S, 24X17S	Rac. (See Ch. 20X1
Chassis 6A1	,	20A1)	(See Ch. 6M1) 25	Tel. Rec. (See Ch.	and 451)
(See Model 6T01) - Chassis 6A2		Tel. Rec. (See Ch. 30B1) 71	Model 7C63, 7C63-UL	20X1 and 4L1)	Models 32X26, 32X27 Tel. Rec. (See Ch. 20X1
Chassis 6B1	48-2	Models 4H137 (A or B)	(See Ch. 7C1) 25	Models 25A15, 25A16, 25A17 Tel. Rec.	and 582}
Chassis 6C1	53—1	Tel. Rec. (See Ch. 20A1) 77	Model 7C64	(See Ch. 20A1) 77	Models 32X35, 32X36 Tel.
Chassis 6E1, 6E1N		Model 4H137 (S or SN) Tel. Rec. (See Ch. 30B1) 71	7C65W (See Ch. 7E1) 36	Models 26R11, 26R12 Tel.	Rec. (See Ch. 20X1
Chassis 6F1 Chassis 6L1	26-2	Models 4H146, 4H147	Model 7C73 (See Ch. 9A1) 32	Rec. (See Ch. 21B1)118 Models 26R25, 26R26 Tel.	and 5B2)
Chassis 6A1 Chassis 6Q1 Chassis 6Q1 Chassis 6R1 Chassis 6S1	25-1	(A or B) Tel. Rec.	Models 7G11, 7G12, 7G14, 7G15, 7G16	Rec. (See Ch. 24D1)103	A Tel. Rec. (See Ch. 2011)
Chassis 6Q1	54-1	(See Ch. 20A1)	(See Ch. 7G1) 54	Models 26R25A, 26R26A	(See Ch. 2011)
Chassis 651		(C or CN) Tel. Rec.	Model 7P32, 7P33, 7P34,	Tel. Rec. (See Ch. 21B1) 118	(See Ch. 21B1)
		(See Ch. 20A1) 77	7P35 (See Ch. 5HI) 26 Model 7RT41, 7RT42,	Models 26R35, 26R36,	Models 36R45, 36R46 Tel.
Chossis 6W1	75-1	Models 4H145, 4H146, 4H147 (S or SN) Tel.	7RT43 (See Ch. 6L1) 26	26R37 Tel. Rec. (See	Rec. (See Ch. 2181)118 Models 36X35, A, 36X36,
Chassis 6Y1 Chassis 7B1	18-2	Rec. (See Chassis 30B1). 71	Models 7T01, 7T01M-UL,	Ch. 24D1)	A, 36X37, A Tel. Rec.
Chassis 7C1 Chassis 7E1	25-2	Models 4H155, 4H156,	7T04, 7T04-UL (See	26R37A Tel. Rec.	[See Ch. 24D1 (Set 103)
Chassis 7G1	54-2	4H157 (A or B) Tel. Rec. (See Ch. 20A1) 77	Ch. 5N1) 31 Model 7T06, 7T12	(See Ch. 21B1)118	and Radio Ch. 5B2 (Set 100)]
Chassis 8B1		Models 4H155, 4H156,	(See Ch. 4B1) 24	Models 26X35, 26X36, 26X37 Tel. Rec. (See	Models 37K15, 37K16 Tel.
Chassis 8C1 (See Ch.	8D1) 67	4H157 (S or SN) Tel.	Model 7110, 7114, 7115	Ch. 24D1)	Rec. (See Ch. 21F1) 127-14
Chassis 8D1 Chassis 9A1	32_1	Rec. (See Chassis 30B1). 71	(See Ch. 5K1) 30	Models 26X45, 26X46 Tel.	Models 37K27, 37K28 Tel. Rec. (See Ch. 21F1)127-1/
Chassis 9B1	49-2	Models 4H165, 4H166, 4H167 (A or B) Tel.	Models 8C11, 8C12, 8C13 [See Chassis 30A1 (Set	Rec. (See Ch. 24D1)103	Models 37K35, 37K36 Tel.
Chassis 9E1		Rec. (See Ch. 20A1) 77	57) and 8D1 (Set 67)]	Models 26X55, 26X56, 26X57 Tel. Rec.	Rec. (See Ch. 21F1) 127-1/
Chassis 10A1 Chassis 19A1 Tal. Re	59_2	Models 4H165, 4H167	Tel. Rec.	(See Ch. 24D1)103	Models 39X16, 39X16A, B, 39X17A, B Tel. Rec. [See
Chassis 19A1 Tel. Re-	c.	(C or CN) Tel. Rec. (See Ch. 20A1)	Models 8C14, 8C15, 8C16, 8C17 (See Ch. 8D1) 67	Models 26X55A, 26X56A,	Ch. 24D1 (Set 103) and
Prod. Chge. Bul. 5	i106—1	Models 4H165, 4H166,	Models 8D15, 8D16	26X57A Tel. Rec. (See Ch. 21B1)118	Radio Ch. 582 (Set 100)]
Chassis 20A1, 20B1,	77_1	4HI67 (S or SN) Tel.	(See Ch. 8D1) 67	Models 26X65, 26X66,	Model 39X17C Tel. Rec. (See Ch. 21B1)118
Tel. Rec. Chassis 2011 Tel. Re-	117-2	Rec. (See Chassis 30B1). 71	Model 8RP46 (See Chassis 3A1) 2	26X67 Tel. Rec.	Models 39X25A, 39X26A
Chassis 2011 Ial. Ke	C.	Models 4R11, 4R12 (See Ch. 4R1)	(See Chassis 3A1) # Model 9B14, 9B15, 9B16	(See Ch. 24D1) 103	Tel, Rec.
Prod. Chge. Bul. 15 Chassis 20V1 Tel. Red	5 . 1. 126-1	Models 5F11, 5F12 57	(See Ch. 9B1) 49	Models 26X65A, 26X66A, 26X67A Tel. Rec.	(See Ch. 2181)118 Models 39X35, 39X36,
(See Ch. 2011)		(See Ch. 5F1)	Models 9E15, 9E16,	(See Ch. 21B1)	39X37 Tel. Rec.
Chassis 20V1 Tel. Re	c.	Models 5R11, 5R12, 5R13,	9E17 (See Ch. 9E1) 68	Models 26X75, 26X76 Tel.	(See Ch. 2181)118
Prod. Chge. Bul. 1	5126-1	5R14 (See Ch. 5R1) 59 Model 5T12 (Ch. 5T1) 68	Models 12X11, 12X12 Tel. Rec. (See Ch. 20X1)100	Rec. {See Ch. 24D1} 103 Models 26X75A, 26X76A	Models 221K16, 221K16A
Chassis 20X1, 20Y1, 2 Tel. Rec.	100-1	Models 5W11, 5W12 (See	Models 14R11, 14R12 Tel.	Tel, Rec.	Tel. Rec. (See Ch. 21F1)
Chassis 2021 Tel. Rec		Ch. 5W1) 79	Rec. (See Ch. 20T1)117	(See Ch. 21B1)118	Models 221K26, 221K28
Prod. Chge. Sul. 7 Chassis 21A1 Tel. Re	110-1	Models 5X11, 5X12, 5X13,	Models 16R11, 16R12 Tel. Rec. (See Ch. 21B1)118	Models 27K12 Tel. Rec. (See Ch. 21F1)127-1A	Tel. Rec. (See Ch. 21F1)127-1
		5X14 (See Ch. 5X1) 76	Rec. Laee Un. 21011	1 (See UR. ZIT!)	

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ADMIRAL-ARVIN

ADMIRAL-Cont.	AIRCASTLE-Cont.	AIRLINE-Cont.	AIRLINE-Cont.
Models 221K35, 221K36	139144 59-4	648R-1503B, 648R-1504B	74WG-2705A, 74WG-
Tel. Rec. (See	147114	(See Models 54BR-	2705B (See Model
Ch. 21F1)	149654, 150084 71-4 159144 (See Model	1503A, B, C; 54BR- 1504A, B, C)	74WG-2505A) 18 74WG-2709A 26—
321K18 Tel, Rec.	139144)	64BR-1513A, B;	74WG-2711A (See Model
(See Ch. 21F1)	AIR CHIEF (See Firestone)	64BR-1514A, B 244	74WG.2505A) 18
Models 321K27 Tel. Rec. (See Ch. 21F1) 127-1A	Aik chief (Jee Firestone)	648R-1808A 16-5 648R-2200A (See Model	84BR-1065A 84BR-1503D, 84BR-1504D 84BR-1515A, 84BR-1516A 84BR-1517A, 84BR-1518A 84BR-1815B, 84BR-1816B. 55—
Models 321K35, 321K36	AIR KING	64BR-1208A)	8488-1515A 8488-1516A *
Tel. Rec. (See	A-400 (Ch. 470) 23—1 A-403 20—2 A-410 34—1	64BR-7000A 51-2	84BR-1517A, 84BR-1518A *
Ch. 21F1)	A-403 20-2	648R-7100A, 648R-7110A, 648R-7120A 57-5	84BR-1815B, 84BR-1816B . 55-
32K49 Tel. Rec.	A-410 (Revised) 40-1	64BR-7300A, 64BR-7310A,	848R-2015A ************************************
(See Ch. 21F1) 127-1A	A-426 43—1 A-450		84BR-2719A *
AERMOTIVE	A-450	64BR-7810A, 64BR-7820A 53-3	84BR-27268 *
181-AD 12—1	A-501, A-502 (Ch. 465-4) 31-3 A-510 24-3	64BR-7320A 54-4 64BR-7810A, 64BR-7820A 53-3 64WG-1050A 10-2 64WG-1050B, 64WG-10-2	84BR-3004 Tel. Rec
AIRADIO	A-510	1050C, 64WG-1050D	84GCB-1062A 52
SU-41D	A-520 49-4	(See Model	84GDC-963B 51-
SU-52A, B, C (Receiver) 13—2 TRA-1A, B, C (Transmitter) 13—1	A-600	64WG-1050A) 10	84GDC-987A
3100 37—1	A-625	64WG-1052A	8488.27268 8488.304 Fel. 6 8487.304 Fel. 8 6 7 7 7 8405.276 S1 52 5405 5405 5405 5405 5405 5405 5405 5405 5405 5405 5405 5405 5505 5405 5505
AIRCASTLE	A-625	64WG-1052A)	
DM-700 851	800	64WG-1207B 18-5 64WG-1511A, 64WG-	84GSE-2731A
EV-760 (See Model	A-1000, A-1001 Tel. Receiver 58—3 A1001A Tel. Rec. 75—2 A101A Tel. Rec. 91—2		84HA-1527A, 84HA-1528A
DM-700) 85 G-516, G-518 48-3	A1001A Tel. Rec 75-2	1511B, 64WG-1512A, 64WG-1512B	(See Model 94HA-1527C) 67 84HA1529A, 84HA1530A . 85-
G-521	A1016 181. Rec	64WG-1801C (See Models	84HA-1810A
G-521	A2000, A2001, A2002 Tel. Rec. (See Model	54WG-1801A, B) 4	84HA-1810C 69-
G-725 50—1	A1001A) 75	64WG-1804A, B	84HA-2727A 84HA-3002A, 84HA-3002B
P.20 71_3	A2010 Tel. Rec. (See	64WG-1804A)	Tel. Rec. 99-
K1	Model A1001A)	64WG-1807A,	Tel. Rec. 99- 84HA3007A, B, C Tel. Rec. *
PAM.4 101_1	Model A1001A)	64WG-18078 5-4 64WG-1809A, 64WG-	84HA-3010A. B C
PC-8, PC-358	16C1, 16C2, 16C3	1809B (See Models	Tel. Rec. 94- 84HA-3010A, B, C Tel.
PM-358	Tel. Rec	64WG-1511A, B; 64WG-	Rec. Prod. Chge.
PX	16M1 Tel. Rec. (See Model 16C1)121	1512A, 8) 5 64WG-2007A.	Bul. 11
REV248	16T1, Tel. Rec.	64WG-2007A, 64WG-2007B	84KR-1520A
SC-448	[See Model 16C1]121	64WG-2009A,	84KR-2511A
WFU.262 91_1	19C1 Tel. Rec. (See Model 16C1)121	64WG-2009B 6-2	BUL 11 B4KR-1209A B4KR-1520A B4KR-2511A B4WG-1060A
WRA1-A 47-1 WRA-4M 60-1 XB702, XB703 Tel. Rec. 93A-1	718R Tel. Rec.	64WG-20108 18-6 64WG-2500A (See	84WG-1060C (See Model 84WG-1060A) 42
X8702 X8703 Tel Parc 92A 1	(See Model 16C1) 121	Model 54WG-2500A) 4	84WG-2015A
AL/30, AP//3 161, Kec 934-1	2017R Tel. Rec	64WG-2700A,	84WG-2506 (See Model 84WG-2721A)
78 52-1	4603	64WG-2700B (See Models 54WG-2500A;	84WG-2721A}
9	4603	54WG-2700A) 4	84WG-25068 58- 84WG-2712A 43-
	4604D (See Model 4604) 4	74BR-916B 17-5	
15 67—2	4609, 4610 Early	748R-1053A	84WG-2712A) 43
15 67—2 16C, 16T Tel. Rec	(See Model 4607)	74BR-1055A * 74BR-1501B, 74BR-1502B * 74BR-1507, 74BR-1508A * 74BR-1513B, 74BR-1514B	84WG-2714A
102B	4609, 4610 11-2 4625	74BR-1507, 74BR-1508A. *	84WG-2718A, 84WG-
106B 133	4700 391	74BR-1513B, 74BR-1514B (See Models 64BR-	2718B, 84WG-2720A 45
150, 153	4704 12-2 4705, 4706 9-1	1513A, B; 64BR-	84WG-2721A, 8 46- 84WG-2724A (See
198 83-1	4705, 4706	1514A, B)	Model 84WG-2718A) 45
201 81-1		745R-1812A (See Model 745R-1812B)	84WG-2728A (See Models 84WG-2718A, B;
211 65 —1 212 68 —3	AIR KNIGHT (SKY KNIGHT)	74BR-1812B	84WG-2720A) 45
213 63-1	CA-500 17-4 CB-500P 17-31	748R-2001A (See Model	84WG-2732A, 8 (See
2271, 227W	N5-RD291	74BR-2001B)	Model \$4WG-2712A, 8) 43 84WG-2734A (See Models
312 Tel. Rec		74BR-2003A	84WG-2718A, B;
358VM 127—3 412 Tel. Rec	AIRLINE	74BR-2003A 74BR-2701A 74BR-2702A (See Model	84WG-2720A) 45
412 Tel. Rec	05BR-3021B Tel. Rec * 05BR-3021C Tel. Rec *	745R-2702A (See Model 745P-2702B) 25	84WG-3006, 84WG- 3008 84WG-3009
	05BR-3024B Tel. Rec *	74BR-2702B	3008, 84WG-3009, (See Model 94WG 3006A) Tel. Rec 72
572	05BR-3024C Tel. Rec * 05BR-3027A Tel. Rec *	74BR-2707A	3006A) Tel. Rec 72
602-182144 114 _2	05BR-3027A Tel. Rec * 05BR-3027B Tel. Rec *	7488-2708A	948R-1525A, 9488-1526A
	058R-3034A Tel. Rec *	74BR-2708A 74BR-2715A 74BR-2717A	94BR-1526A
606-400WB 119-2 607-314, 607-315 122-2 621 (Ch. FJ-91) 14-2 626 18-3		74GSG-8400A,	
607-314, 607-315 122 -2	05BR-3044A Tel. Rec	74GSG-8700A 60—3 74GSG-8810A,	2741A, B
626	05GCB-1540A,	74GSG-8820A 52-2	2741A, B
	05GCB-1541A	74HA-8200A 58-4 74KR-1210A 41-1	94BR-3017A Tel. Rec 89-
651 15 1 9151, W	05GCB-3019A Tel. Rec 116-2	74KR-1210A 41-1 74KR-27068 35-1	94BR-3017B Tel. Rec. Prod. Chge. Bul. 7110-
935	05GHM-1061A	74KR-2713A 43-2	948R-3021, 948R-3024A
5000 5001 16-2	05GSE-3037A Tel. Rec *	74WG-925A	Tel. Rec *
5002 19-1 5003, 5004, 5005, 5006. 20-1 5008, 5009 46-1	05WG-1811B (See Model 94WG-1811A)	Model 64WG-1050A) 10	94GAA3654A
5008, 5009 46-1	05WG-1813A	74WG-1052B (See	94GCB-3023A, B, C
5010, 5011, 5012	05WG-2748C. D. E (See	Models 64WG-1052A, B) 9 74WG-1054A 22 1	Tel, Rec. (See Model
(Ch. 110)	Model 94WG-2748A) 90 05WG-2749D	74WG-1054A	05GCB-3019A)
5020	05WG-2752 100-3	74WG-1054A) 22	94GDC-989A * 94GSE- *
5020	05WG-3016A, B Tel. Rec.	74WG-1056A	2736A 72—
5024	(See Model 94WG-3006A Set 72 and Set 110	74WG-1207B (See Model	94GSE-3011, B (See Model
5027	Folder 2)	64WG-1207B) 18	84GSE-3011A}
5027 49_3 5028 44_1 5029 51_1	05WG-3030A Tel. Rec119-3 05WG-3031A Tel. Rec109-1	74WG-1509A, 74WG-1510A 27—1	94GSE-3018A Tel. Rec 93A-
5035 46 <u>-</u> 2	05WG-3031B Tel. Rec. *	74WG-1511B, 74WG-	94GSE-3025A Tel. Rec *
5035	05WG-3036A, B Tel. Rec. * 05WG-3038A Tel. Rec 129-4	1512B (See Models	
5044	05WG-3038A Tel. Rec 129-4 05WG-3039A, 8 Tel. Rec. *	64WG-1511A, B; 64WG-1512A, B} 5	94HA-1527C, 94HA-1528C 67- 94HA1529A, 94HA1530A (See Model 84HA1529A) 85
5052	05WG-3040 Tel, Rec *	74WG-1802A	(See Model 84HA1529A) 85
5056-A	05WG-3042A Tel. Rec *	74WG-1803A (See Model	94WG.1059A 75_
	05WG-3045A Tel. Rec. (See Model	74WG-1802A)	94WG-1804D
6053 97-1 6514 18-4 6541 17-2	05WG-3038A)	64WG-1804A, B) 4	94WG-2742A, C, D 71-
6514 18-4	15GHM-936A,	74WG-1807A, 74WG- 1807B (See Models	94WG-2742A, C, D 71— 94WG-2745A
6541	15GHM-937A 134-2 15GSE-3043A Tel. Rec 15GSE-3047A B Tel Rec	1807B (See Models 64WG-1807A R) E	94WG-2746A, B; 94WG-2747A (See
6541) 17 6611, 6612, 6613, 6630,		64WG-1807A, B) 5 74WG-2002A 26-4	Model 94WG-2742A) . 71
6611, 6612, 6613, 6630,	15WG-2745C 130-2 15WG-2765A (See Model	74WG-2004A	94WG-2748A, 94WG-
6631, 6632, 6634, 6635 15 -2 7000, 7001 14 -3	15WG-2745C1	74WG-2007B, 74WG- 2007C (See Models	2749A
7000, 7001	54RP.1501A 54RP.1502A 7	64WG-2007A, B1 5	94WG-2748A) 90
7014, 7015 57-3	54BR-1503A, B, C, 54BR-1504A, B, C 3-4	74WG-2009B (See Models 64WG-2009A, B) 6	94WG-3006A Tel. Rec 72- 94WG-3006B Tel. Rec 85-
7015 Early	340K-1303A, B; 346K-	64WG-2009A, B) 6 74WG-2010A (See Model	94WG-3006B Tel. Rec 85-
90081,9008w 99-2 90091,9009w 97-2 90121,9012w 94-1 10002 56-1	1506A, B 2-34	64WG-2010B) 18	94WG-3008A, 94WG- 3009A Tel. Rec. (See
90091, 9009W	54KP-1209A, B	74WG-2010B 186	Model 94WG-3006A) 72
10002	18018	74WG-2500A (See Model	94WG-3009BTel. Rec. (See
	54WG-2500A, 54WG-	54WG-2500A) 4 74WG-2504A	Model 94WG-3006B) 85 94WG-3016A, B, C Tel. Rec.
10021-1, 10022-1	2700A	74WG-2504B 74WG-	(See Model 94WG-3006A
10005	64BR-916B (See Model	2504C (See Model 74WG-2504A) 28	Set 72 and Model 05WG- 3016A Set 110 Folder 2)
10024-1	74BR-916B) 17	74WG-2505A	94WG-3022A Tel, Rec. (See
121104	64BR-917A 10-1 64BR917B (See Model	74WG-2700A, 74WG-	Model 94WG-3006B) 85
10024-1	64BR917A) 10	2700B (See Model	94WG-3026A Tel, Rec. (See
127084 55—2 131504 60—2 132564 69—1 138104 54—3	64BR-1051A	54WG-2700A) 4 74WG-2704A, 74WG-	Model 94WG-3006B) 85 94WG-3028A Tel, Rec. (See
132564 69-1	64BR1051A) 2	2704B, 74WG-2704C	Model 94WG-3006) 72
138104	64BR-1205A, 64BR-1206A. 10-3	(See Model	94WG-3029A Tel. Rec. (See
138124 64 —1	648R-1208A 16-4	74WG-2504A) 28	Model 94WG-3006B) 85
0			

AIRCASTLE-Cont.
139144 59-4 147114 56-3 149654, 150084 71-4
149654, 150084 71-4 159144 (See Model 139144) 59
159144 (See Model 139144) 59
AIR CHIEF (See Firestone)
AIR KING
A-400 (Ch. 470) 23—1 A-403 20—2 A-410 34—1
A-403
A-410 (Revised) 40-1
A-426
A-501, A-502 (Ch. 465-4) 31-3
A-510
A-511, A-512 30-2 A-520 49-4
A-600
A-604 81-2 A-625 50-3
A. 650 454
800 66—1
800 66-1 A-1000 66-1 Tel. Receiver 31001A Tel. A2000, A2001 75-2 A1016 Tel. Rec. 75-2 75-2 A1016 A1. 75 A2010 Tel. Rec. Model A1001A) 75
A1001A Tel. Rec 75-2
A1016 Tel. Rec
Tel. Rec. (See Model
A1001A)
Model A1001A)
A-2012 Tel. Rec. (See
Model A1001A) 75
Tel. Rec
16M1 Tel. Rec. {See Model 16C11121
16T1, Tel. Rec.
Model A1001A)
(See Model 16(1) 121
718R Tel. Rec. (See Model 16C1) 121
(See Model 16C1) 121 2017R Tel. Rec
4601 (See Model 4609) 11
4603
4604D (See Model 4604) 4
4604 4—25 4604D (See Model 4604) 4 4607, 4608 3—1 4609, 4610 Early
(See Model 4007) 3
4609, 4610 11-2 4625
4/00
4704
4705, 4706 9-1 4708 (See Model 4704) 12
AIR KNIGHT (SKY KNIGHT)
CA 600 17 /
CB-500P 17-31
AIRLINE
05BR-30218 Tel. Rec
058R-30218 Tel. Rec * 058R-3021C Tel. Rec * 058R-30248 Tel. Rec *
05BR-3021B Tel. Rec
05BR-3021B Tel. Rec 05BR-3021C Tel. Rec 05BR-3024B Tel. Rec 05BR-3024C Tel. Rec 05BR-3027A Tel. Rec 05BR-3027A Tel. Rec
0588-30218 Tel. Rec 0588-30248 Tel. Rec 0588-30248 Tel. Rec 0588-30247 Tel. Rec 0588-30277 Tel. Rec 0588-30276 Tel. Rec 0588-30278 Tel. Rec
058R-30218 Tel. Rec
058R-30218 Tel. Rec
058R-3021B Tel. Rec • 058R-3024E Tel. Rec • 058R-3024E Tel. Rec • 058R-3024C Tel. Rec • 058R-3027A Tel. Rec • 058R-3027A Tel. Rec • 058R-3024A Tel. Rec • 058R-304A Tel. Rec • 125-2 056CA-1540A 121-2
0588.30218 Tel. Rec
058R-30218 Tel. Rec
058R-30218 Tel. Rec
0588.30218 Tel. Rac. • 0588.30214 Tel. Rac. • 0588.30248 Tel. Rac. • 0588.30246 Tel. Rac. • 0588.30246 Tel. Rac. • 0588.30274 Tel. Rac. • 0588.30274 Tel. Rac. • 0588.30276 Tel. Rac. • 0588.30276 Tel. Rac. • 0588.30276 Tel. Rac. • 0588.30276 Tel. Rac. • 0588.3024A Tel. Rac. • 0588.30264 Tel. Rac. • 056CA-1992A - 125-2 056CB-1541A 131-2 056CB-3019A 056CB-3019A Tel. Rac. 113-3 056CB-3020A Tel. Rac. 133-3 056CB-3020A Tel. Rac. * 050CG-18118 (See Model) * 94WG-1811A
OSBR-30242 Tel. Rec. OSBR-30242 Tel. Rec. OSBR-30247 Tel. Rec. OSBR-3041A Tel. Rec. OSBR-3044A Tel. Rec. OSGCB-1540A Tel. Rec. OSGCB-1541B See Model 94WG-1811A) 99
OSBR-30242 Tel. Rec. OSBR-30242 Tel. Rec. OSBR-30247 Tel. Rec. OSBR-30248 Tel. Rec. OSBR-3041A Tel. Rec. OSBR-3044A Tel. Rec. OSGR-3044 Tel. Rec. OSGCB-1541A Tal2 OSGCB-1541A Tal2 OSGCB-1541A Tal2 OSGCB-1541A Tal2 OSGCB-1541A Tal2 OSGCB-1541B Tel. Rec. Tel. Rec. Tal3 OSGSE-3037A Tel. Rec. OSWG-274811A) 99 OSWG-274811A 127-4 OSWG-2746C, D, E. (See Model 94
OSBR-30242 Tel. Rec. OSBR-30242 Tel. Rec. OSBR-30247 Tel. Rec. OSBR-30248 Tel. Rec. OSBR-3041A Tel. Rec. OSBR-3044A Tel. Rec. OSGR-3044 Tel. Rec. OSGCB-1541A Tal2 OSGCB-1541A Tal2 OSGCB-1541A Tal2 OSGCB-1541A Tal2 OSGCB-1541A Tal2 OSGCB-1541B Tel. Rec. Tel. Rec. Tal3 OSGSE-3037A Tel. Rec. OSWG-274811A) 99 OSWG-274811A 127-4 OSWG-2746C, D, E. (See Model 94
OSBR-30242 Tel. Rec. OSBR-30242 Tel. Rec. OSBR-30247 Tel. Rec. OSBR-30248 Tel. Rec. OSBR-3041A Tel. Rec. OSBR-3044A Tel. Rec. OSGR-3044 Tel. Rec. OSGCB-1541A Tal2 OSGCB-1541A Tal2 OSGCB-1541A Tal2 OSGCB-1541A Tal2 OSGCB-1541A Tal2 OSGCB-1541B Tel. Rec. Tel. Rec. Tal3 OSGSE-3037A Tel. Rec. OSWG-274811A) 99 OSWG-274811A 127-4 OSWG-2746C, D, E. (See Model 94
OSBR-30242 Tel. Rec. OSBR-30242 Tel. Rec. OSBR-30247 Tel. Rec. OSBR-30248 Tel. Rec. OSBR-3041A Tel. Rec. OSBR-3044A Tel. Rec. OSGR-3044 Tel. Rec. OSGCB-1541A Tal2 OSGCB-1541A Tal2 OSGCB-1541A Tal2 OSGCB-1541A Tal2 OSGCB-1541A Tal2 OSGCB-1541B Tel. Rec. Tel. Rec. Tal3 OSGSE-3037A Tel. Rec. OSWG-274811A) 99 OSWG-274811A 127-4 OSWG-2746C, D, E. (See Model 94
036R.302/L Fei. Rec. • 036R.301/L Fei. Rec. • 036R.301/L Fei. Rec. • 036CB-1360/L Fei. Rec. • 036CB-1360/L Fei. Rec. • 036CB-1360/L Fei. Rec. • 036CB-137A Tei. Rec. • 036CB-131A • 036CB-131A • 036CB-131A • 036CB-131A • 036CB-131A • 03WG-2742C, D. E (See Model 94WG-2748A). 90 05WG-2742D • 05WG-2742A). 129-3 05WG-2742A). 120-3 05WG-2744A). 90 05WG-2744A). 90 05WG-3016A, B Tei. Rec. (See Mode) 94WG-3006A
036R.302/L Fei. Rec. • 036R.301/L Fei. Rec. • 036R.301/L Fei. Rec. • 036CB-1360/L Fei. Rec. • 036CB-1360/L Fei. Rec. • 036CB-1360/L Fei. Rec. • 036CB-137A Tei. Rec. • 036CB-131A • 036CB-131A • 036CB-131A • 036CB-131A • 036CB-131A • 03WG-2742C, D. E (See Model 94WG-2748A). 90 05WG-2742D • 05WG-2742A). 129-3 05WG-2742A). 120-3 05WG-2744A). 90 05WG-2744A). 90 05WG-3016A, B Tei. Rec. (See Mode) 94WG-3006A
036R.302/L Fei. Rec. • 036R.301/L Fei. Rec. • 036R.301/L Fei. Rec. • 036CB-1360/L Fei. Rec. • 036CB-1360/L Fei. Rec. • 036CB-1360/L Fei. Rec. • 036CB-137A Tei. Rec. • 036CB-131A • 036CB-131A • 036CB-131A • 036CB-131A • 036CB-131A • 03WG-2742C, D. E (See Model 94WG-2748A). 90 05WG-2742D • 05WG-2742A). 129-3 05WG-2742A). 120-3 05WG-2744A). 90 05WG-2744A). 90 05WG-3016A, B Tei. Rec. (See Mode) 94WG-3006A
038R-30342 Tei. Rec. • 038R-30342 Tei. Rec. • 038R-30342 Tei. Rec. • 038R-30342 Tei. Rec. • 038R-30344 Tei. Rec. • 036CA-9974 Rec. • 036CA-10374 Tei. Rec. • 036CA-103774 I. Rec. • 036CA-103774 Tei. Rec. • 036CB-134074 • • 036CB-134774 I. Rec. • 036CB-134774 I. Rec. • 036CB-134774 I. Rec. • 03WG-21482 J. E. Rec. • 03WG-27420 I. 29-3 05WG-27420 I. 29-3 05WG-20164, 8 Tei. Rec. (See
038R-30342 Tei. Rec. • 038R-30342 Tei. Rec. • 038R-30342 Tei. Rec. • 038R-30342 Tei. Rec. • 038R-30344 Tei. Rec. • 036CA-9974 Rec. • 036CA-10374 Tei. Rec. • 036CA-103774 I. Rec. • 036CA-103774 Tei. Rec. • 036CB-134074 • • 036CB-134774 I. Rec. • 036CB-134774 I. Rec. • 036CB-134774 I. Rec. • 03WG-21482 J. E. Rec. • 03WG-27420 I. 29-3 05WG-27420 I. 29-3 05WG-20164, 8 Tei. Rec. (See
038R-30342 Tei. Rec. • 038R-30342 Tei. Rec. • 038R-30342 Tei. Rec. • 038R-30342 Tei. Rec. • 038R-30344 Tei. Rec. • 036CA-9974 Rec. • 036CA-10374 Tei. Rec. • 036CA-103774 I. Rec. • 036CA-103774 Tei. Rec. • 036CB-134074 • • 036CB-134774 I. Rec. • 036CB-134774 I. Rec. • 036CB-134774 I. Rec. • 03WG-21482 J. E. Rec. • 03WG-27420 I. 29-3 05WG-27420 I. 29-3 05WG-20164, 8 Tei. Rec. (See
038x.302/L2 Teil. Rec
038x.302/L2 Teil. Rec
036R.30212 Tei. Rec. • 036R.30242 Tei. Rec. • 036R.30242 Tei. Rec. • 036R.30242 Tei. Rec. • 036R.30244 Tei. Rec. • 036R.30144 Tei. Rec. • 036C8-1360A 1312 • 036C8-1360A 1312 • 036C8-1367A Tei. Rec. • 05WG-3016A, B Tei. Rec. •
036R.30212 Tei. Rec. • 036R.30242 Tei. Rec. • 036R.30242 Tei. Rec. • 036R.30242 Tei. Rec. • 036R.30244 Tei. Rec. • 036R.30144 Tei. Rec. • 036C8-1360A 1312 • 036C8-1360A 1312 • 036C8-1367A Tei. Rec. • 05WG-3016A, B Tei. Rec. •
036R.30212 Tei. Rec. • 036R.30242 Tei. Rec. • 036R.30242 Tei. Rec. • 036R.30242 Tei. Rec. • 036R.30244 Tei. Rec. • 036R.30144 Tei. Rec. • 036C8-1360A 1312 • 036C8-1360A 1312 • 036C8-1367A Tei. Rec. • 05WG-3016A, B Tei. Rec. •
036R.30212 Tei. Rec. • 036R.30242 Tei. Rec. • 036R.30242 Tei. Rec. • 036R.30242 Tei. Rec. • 036R.30244 Tei. Rec. • 036R.30144 Tei. Rec. • 036C8-1360A 1312 • 036C8-1360A 1312 • 036C8-1367A Tei. Rec. • 05WG-3016A, B Tei. Rec. •
038R.30342 Teil. Rec. • 038R.30344 Teil. Rec. • 035G.303044 Teil. Rec. • 035G.51314 131-2 • 035G.51317 Teil. Rec. • 035G.81307A Teil. Rec. • 035WG-3036A, B Teil. Rec. • 05WG-3036A, B Teil. Rec. 129-3 05WG-3036A, B Teil. Rec. • 05WG-3036A, B Tei
038.7.302/C 16 16 038.302/A 16 16 036CB-1340A 13 12 036CB-1340A 13 13 036CB-1340A 13 13 036CB-1340A 13 13 036CB-1340A 18 16 036CB-131A 18 18 036CB-131A 18 18 036CB-2047A 18 10 036CB-2047A
ODBR.302/LC Tel. Rec. ODBR.304/LC Tel. Rec. ODBR.304/LC Tel. Rec. OSCGB.154/LA OSGCB.154/LA OSGC
038R.30242 Tei. Rec. • 038R.30242 Tei. Rec. • 038R.30242 Tei. Rec. • 038R.30247 Tei. Rec. • 036C8.1540A 1312 • 036C8.1541A 1312 • 036C8.1014A Tei. Rec. 1333 036C8.1017A Tei. Rec. 1333 036C8.1811A) 99 • 05WG-7142C, D. E. (See Model 94WG-7248A). 90 • 05WG-2742D 1293 • • 05WG-2742C, D. E. (See Model 94WG-7248A). 90 • 05WG-3030A Tei. Rec. 1293 • 05WG-3030A Tei. Rec. 129-3 • 05WG-3030A Tei. Rec. 129-3 05WG-3030A
038R.30242 Tei. Rec. • 038R.30242 Tei. Rec. • 038R.30242 Tei. Rec. • 038R.30247 Tei. Rec. • 036C8.1540A 1312 • 036C8.1541A 1312 • 036C8.1014A Tei. Rec. 1333 036C8.1017A Tei. Rec. 1333 036C8.1811A) 99 • 05WG-7142C, D. E. (See Model 94WG-7248A). 90 • 05WG-2742D 1293 • • 05WG-2742C, D. E. (See Model 94WG-7248A). 90 • 05WG-3030A Tei. Rec. 1293 • 05WG-3030A Tei. Rec. 129-3 • 05WG-3030A Tei. Rec. 129-3 05WG-3030A
038x.302/12 Tei. Nec
038x.302/12 Tei. Rec. 038x.302/14 Tei. Rec. 038x.302/14 Tei. Rec. 038x.302/14 Tei. Rec. 036x.302/14 Tei. Rec. 036x.302/15 Tei. Rec. 036x.302/14 Tei. Rec. <t< td=""></t<>
038.8.30242 Tel. Rec. 038.8.30242 Tel. Rec. 038.8.30247 Tel. Rec. 036.8.30247 Tel. Rec. 036.8.30247 Tel. Rec. 036.8.30204 Tel. Rec. 036.8.30204 Tel. Rec. 036.8.30204 Tel. Rec. 036.8.30204 Tel. Rec. 94.90.18118 (Se Model 94.90.18118 (Se Model 94.90.18118 (Se Model 95.90.2752 129-3 05.90.2752 129-3 05.90.20330A Tel. Rec. 129-3
ODBR.3021C Tel. Rec. ODBR.3024C Tel. Rec. ODBR.3024A Tel. Rec. ODBR.3024A Tel. Rec. ODBR.304A Tel. Rec. ODBR.304A Tel. Rec. ODSCB.1340A, ISDCCB.1340A, OSCCB.1340A, ISDCCB.1340A, OSCCB.1340A, ISDCSE.3002A OSCCB.1340A, ISDCSE.3002A OSCCB.134A, ISDCSE.3002A OSWG-IBISA, ISDCSE.3002A OSWG-2748D, ISDCSE.3006A Sew G-301AA, B Tel. Rec. 100-3 OSWG-2749D, ISD-3 OSWG-2749D, ISD-3 OSWG-2748C, D, E (See Model 940G-3748A). OSWG-3030A Tel. Rec. 100-3 OSWG-3030A Tel. Rec. 105WG-3030A, B Tel. Rec. OSWG-3030A, B Tel. Rec. 105WG-3033A, B Tel. Rec. OSWG-3033A, B Tel. Rec. 105WG-3033A, B Te
038.7.302/12 Teil. Rec. • 038.7.302/14 Teil. Rec. • 038.7.302/14 Teil. Rec. • 038.7.302/14 Teil. Rec. • 0306.2.302/74 Teil. Rec. • 0306.2.303/74 Teil. Rec. • 0306.2.303 Teil. Rec. • 0306.2.303 Teil. Rec. • 0306.2.303 Teil. Rec. • 0306.2.303 Teil. Rec. • 0306.2.30
038.7.302/12 Teil. Rec. • 038.7.302/14 Teil. Rec. • 038.7.302/14 Teil. Rec. • 038.7.302/14 Teil. Rec. • 0306.2.302/74 Teil. Rec. • 0306.2.303/74 Teil. Rec. • 0306.2.303 Teil. Rec. • 0306.2.303 Teil. Rec. • 0306.2.303 Teil. Rec. • 0306.2.303 Teil. Rec. • 0306.2.30
038.7.302/1C Teil. Rec. 038.7.302/4C Teil. Rec. 038.7.302/4A Teil. Rec. 038.7.302/4A Teil. Rec. 038.7.302/4A Teil. Rec. 038.7.302/4A Teil. Rec. 036CB.1340A, USCGE.1340A, 03CGE.3037A Teil. Rec. 03CGE.3037A Teil. Rec. 03GSE.3037A Teil. Rec. 03WG-1811A) 90 03WG-2748D, Feil. Rec. 03WG-3013A, B Teil. Rec. 03WG-3013A, B Teil. Rec. 03WG-303A, Teil. Rec. 109-3 03WG-303A, Teil. Rec. 130 03WG-303
038.7.302/1C Teil. Rec. 038.7.302/4C Teil. Rec. 038.7.302/4A Teil. Rec. 038.7.302/4A Teil. Rec. 038.7.302/4A Teil. Rec. 038.7.302/4A Teil. Rec. 036CB.1340A, USCGE.1340A, 03CGE.3037A Teil. Rec. 03CGE.3037A Teil. Rec. 03GSE.3037A Teil. Rec. 03WG-1811A) 90 03WG-2748D, Feil. Rec. 03WG-3013A, B Teil. Rec. 03WG-3013A, B Teil. Rec. 03WG-303A, Teil. Rec. 109-3 03WG-303A, Teil. Rec. 130 03WG-303
038R.30242 Tei. Rec. • 038R.30244 Tei. Rec. • 038R.3024A Tei. Rec. • 038R.3014A Tei. Rec. • 038R.3014A Tei. Rec. • 036CB-1540A 1312 • 036CB-1541A 1312 • 036CB-137A Tei. Rec. 1333 036CB-137A Tei. Rec. 1333 05GE-3020A Tei. Rec. • 05WG-3742C, D. E. (See Model 94WG-2748A). 90 05WG-30316A, B. Tei. Rec. 1293 05WG-3030A 05WG-30316A, B. Tei. Rec. 1293 05WG-3030A 05WG-3030A Tei. Rec. 1293 05WG-3030A 05WG-3030A, B. Tei. Rec. 129-3 05WG-3030A, Tei. Rec. 05WG-3030A, B. Tei. Rec. 129-3

AIRLINE-Cont.	
4488.15038 4488.15048	
AIRLINE—Cont. 64BR-1503B, 64BR-1504B (See Models 54BR- 1503A, B, C, 54BR- 1504A, B, C, 54BR- 1504A, B, C, 64BR-1514A, B 64BR-13154A, B 64BR-1206A (See Model 64BR-7200A (See Model 64BR-7100A, 64BR-7110A, 64BR-7120A 64BR-710A, 64BR-7110A, 64BR-7300A, 64BR-7210A, 64BR-7300A, 64BR-7300A, 64BR-7300A, 64BR-7300A, 64BR-7300A, 64BR-7300A, 64BR-7300A, 64BR-7300A, 64BR-7300A, 64BR-7300A, 64BR-7300A, 64BR-7300A, 64BR-7300A, 64BR-7300A, 64BR-7300A, 64BR-7300A, 64BR-7300A, 64BR-730A, 64BR-7300A, 64BR-730A, 64BR-	
1503A, B, C; 54BR-	
64BR-1513A, B;	3
64BR-1514A, B	24-4
648R-1808A	165
64BR-1208A)	16 51—2
648R-7000A	51-2
64BR-7120A	57—5
64BR-7300A, 64BR-7310A,	54 4
64BR-7810A, 64BR-7820A	53-3
64WG-1050A	10-2
1050C, 64WG-1050D	
(See Model	10
64WG-1052A	9-2
(See Model 64WG-1050A) 64WG-1052A 64WG-1052A 64WG-1052A) 64WG-1052A) 64WG-1052A) 64WG-1511A, 64WG-1512A, 64WG-1512B 1511B, 64WG-1512A, 64WG-1801A, B) 64WG-180A, B 64WG-180A, See Models 64WG-180A	•
64WG-1207B	18-5
64WG-1511A, 64WG-	
64WG-1512B	5-5
64WG-1801C (See Models	
64WG-1804A, B	4-27
64WG-1804C (See Model	
44M/C 19074	
64WG-18078	5-4
1809B (See Models	
64WG-1511A, B; 64WG-	
64WG-2007A.	5
64WG-2007B	5—6
64WG-2009B	6-2
64WG-20108	18-6
64WG-1807A 64WG-1807A 1807B (See Models 64WG-1511A, B; 64WG- 1512A, B; 64WG-2007A, 64WG-2007B 64WG-2007B 64WG-2007B 64WG-2007B 64WG-2008 64WG-2008 64WG-2700A, 64WG-2700A, 64WG-2700A, 64WG-2700A, 74BR-1053A	4
64WG-2700A,	
o4wG-2700B (See Models 54WG-2500A	
54WG-2700A)	.4 .
748R-9168 748R-1053A 748R-1055A 748R-15018, 748R-15028, 748R-15018, 748R-15028, 748R-1513A, 1513A, 6; 648R- 1513A, 6; 6468R- 1514A, 6; 6468R- 1514A, 6; 6468R- 1514A, 6; 6468R- 1548R-18128) 748R-18128; 748R-2001A (See Model 748R-2001B).	17-5
74BR-1055A	•
748R-15018, 748R-15028. 748R-1507 748R-1508A	
74BR-1513B, 74BR-1514B	
(See Models 64BR- 1513A, B: 64BR-	
1514A, B}	24
745R-1812A (See Model 745R-18128)	22
74BR-1812B	22 22—2
748R-2001A (See Model 748R-2001B)	23
74BR-2001B	23-2
748R-2003A	24-5
74BR-2702A (See Model	
74BR-2702B	25
74BR-2702B 74BR-2702B 74BR-2707A	25 253
748R-27028) 748R-27028 748R-2707A 748R-2708A 748R-2715A	25 253
74BR-2702B) 74BR-2702B 74BR-2707A 74BR-2707A 74BR-2708A 74BR-2715A 74BR-2717A	25 253
748R-27028) 748R-27028 748R-2707A 748R-2708A 748R-2715A 748R-2717A 746R-2717A 74GSG-8400A, 74GSG-8400A,	25 253
/488-27028) /488-27028 /488-2707A /488-2705A /488-2715A /488-2715A /488-2717A /488-2717A /46SG-8400A, /46SG-8400A, /46SG-8810A,	25 253
748R-2001A (See Model 748R-2001B) 748R-2001B) 748R-2003A 748R-2701A 748R-2701A 748R-2701A 748R-2702A 748R-2702A 748R-2702A 748R-2707A 748R-2707A 748R-2715A 748R-2715A 748R-2715A 748R-2715A 748R-2715A 748S-28400A, 74GSG-8810A, 74GSG-8810A, 74GSG-8810A	25 253 603 522 58-4
/488.27028) /488.27028 /488.27038 /488.27038 /488.27138 /488.27138 /488.27138 /4636.84004, /4636.84004, /4636.88104 /4636.88104 /4638.8304	58—4 41—1
74HA-8200A 74KR-1210A 74KR-27068	58-4
74HA-8200A 74KR-1210A 74KR-27068	58—4 41—1
74HA-8200A 74KR-1210A 74KR-27068	58—4 41—1 35—1 43—2 24—6
74HA-8200A 74KR-1210A 74KR-2706B 74KR-2713A 74WG-925A 74WG-1050C, D (See Model 64WG-1050A)	58-4 41-1 35-1 43-2 24-6 10
74HA 8200A 74KR -1210A 74KR -2706B 74KR -2710A 74KR -2713A 74WG -025A Model 64WG -1050A) 74WG -1050E (See Model 64WG -1052A, B) 74WG -1054B (See Model 74WG -1054A 74WG -1054A 74WG -1057A	58-4 41-1 35-1 43-2 24-6 10 9 22-1 29-1 32-2
74HA 8200A 74KR - 1210A 74KR - 2706B 74KR - 2706B Model 64WG - 1050A) 74WG - 1050C, D (See Model 64WG - 1050A) 74WG - 1054B (See Model 74WG - 1054B (See Model 74WG - 1054A 74WG - 1054A	58-4 41-1 35-1 43-2 24-6 10 9 22-1 29-1 32-2
74HA 8200A 74KR - 1210A 74KR - 2706B 74KR - 2706B Model 64WG - 1050A) 74WG - 1050C, D (See Model 64WG - 1050A) 74WG - 1054B (See Model 74WG - 1054B (See Model 74WG - 1054A 74WG - 1054A	58-4 41-1 35-1 43-2 24-6 10 9 22-1 29-1 32-2
74HA 8200A 74KR - 1210A 74KR - 2706B 74KR - 2706B Model 64WG - 1050A) 74WG - 1050C, D (See Model 64WG - 1050A) 74WG - 1054B (See Model 74WG - 1054B (See Model 74WG - 1054A 74WG - 1054A	58-4 41-1 35-1 43-2 24-6 10 9 22-1 29-1 32-2
74HA-8200A 74KR-1210A 74KR-2706B 74KR-2710A 74KR-2713A 74WG-1050C, D (See Model 64WG-1050A) 74WG-1052B (See Models 64WG-1052A, B) 74WG-1054B (See Model 74WG-1054A 74WG-1054A 74WG-1054A 74WG-1057A 74WG-1057A 74WG-1057A 74WG-1510A 74WG-1510A 74WG-1511A, B;	58-4 41-1 35-1 43-2 24-6 10 9 22-1 22 29-1 32-2 18 27-1
74HA-8200A 74KR-1210A 74KR-2706B 74KR-2710A 74KR-2713A 74WG-1050C, D (See Model 64WG-1050A) 74WG-1052B (See Models 64WG-1052A, B) 74WG-1054B (See Model 74WG-1054A 74WG-1054A 74WG-1054A 74WG-1057A 74WG-1057A 74WG-1057A 74WG-1510A 74WG-1510A 74WG-1511A, B;	58-4 41-1 35-1 43-2 24-6 10 9 22-1 22 29-1 32-2 18 27-1
74HA-8200A 74KR-1210A 74KR-2706B 74KR-2710A 74KR-2713A 74WG-1050C, D (See Model 64WG-1050A) 74WG-1052B (See Models 64WG-1052A, B) 74WG-1054B (See Model 74WG-1054A 74WG-1054A 74WG-1054A 74WG-1057A 74WG-1057A 74WG-1057A 74WG-1510A 74WG-1510A 74WG-1511A, B;	58-4 41-1 35-1 43-2 24-6 10 9 22-1 22 29-1 32-2 18 27-1
74HA-8200A 74KR-1210A 74KR-2706B 74KR-2710A 74KR-2713A 74WG-1050C, D (See Model 64WG-1050A) 74WG-1052B (See Models 64WG-1052A, B) 74WG-1054B (See Model 74WG-1054A 74WG-1054A 74WG-1054A 74WG-1057A 74WG-1057A 74WG-1057A 74WG-1510A 74WG-1510A 74WG-1511A, B;	58-4 41-1 35-1 43-2 24-6 10 9 22-1 22 29-1 32-2 18 27-1
74HA-8200A 74KR-1210A 74KR-2706B 74KR-2710A 74KR-2713A 74WG-1050C, D (See Model 64WG-1050A) 74WG-1052B (See Models 64WG-1052A, B) 74WG-1054B (See Model 74WG-1054A 74WG-1054A 74WG-1054A 74WG-1057A 74WG-1057A 74WG-1057A 74WG-1510A 74WG-1510A 74WG-1511A, B;	58-4 41-1 35-1 43-2 24-6 10 9 22-1 22 29-1 32-2 18 27-1
74HA-8200A 74KR-1210A 74KR-2706B 74KR-2710A 74KR-2713A 74WG-1050C, D (See Model 64WG-1050A) 74WG-1052B (See Models 64WG-1052A, B) 74WG-1054B (See Model 74WG-1054A 74WG-1054A 74WG-1054A 74WG-1057A 74WG-1057A 74WG-1057A 74WG-1510A 74WG-1510A 74WG-1511A, B;	58-4 41-1 35-1 43-2 24-6 10 9 22-1 22 29-1 32-2 18 27-1
74HA-8200A 74KR-1210A 74KR-27068 74KR-2713A 74WG-1030C, D (See Model 64WG-1030A) 74WG-1032B (See Models 64WG-1030A) 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1037A 74WG-1307A 74WG-1307A 74WG-131LA, B 74WG-131LA, B 74WG-1307A 74WG-1307A, 74WG- 1807A 74WG-1807A, 74WG- 1807A 74WG-1807A, 74WG- 1807A 74WG-1807A, 74WG- 1807A	58 - 4 - 1 - 1 - 35 - 1 - 1 - 35 - 1 - 4
74HA-8200A 74KR-1210A 74KR-27068 74KR-2713A 74WG-1030C, D (See Model 64WG-1030A) 74WG-1032B (See Models 64WG-1030A) 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1037A 74WG-1307A 74WG-1307A 74WG-131LA, B 74WG-131LA, B 74WG-1307A 74WG-1307A, 74WG- 1807A 74WG-1807A, 74WG- 1807A 74WG-1807A, 74WG- 1807A 74WG-1807A, 74WG- 1807A	58 - 4 - 1 - 1 - 35 - 1 - 1 - 35 - 1 - 4
74HA-8200A 74KR-1210A 74KR-27068 74KR-2713A 74WG-1030C, D (See Model 64WG-1030A) 74WG-1032B (See Models 64WG-1030A) 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1037A 74WG-1307A 74WG-1307A 74WG-131LA, B 74WG-131LA, B 74WG-1307A 74WG-1307A, 74WG- 1807A 74WG-1807A, 74WG- 1807A 74WG-1807A, 74WG- 1807A 74WG-1807A, 74WG- 1807A	58 - 4 - 1 - 1 - 35 - 1 - 1 - 35 - 1 - 4
74HA-8200A 74KR-1210A 74KR-27068 74KR-2710A 74KR-2713A 74WG-1030C, D (See Model 64WG-1030A), T 74WG-1032B (See Models 64WG-1052A, B) 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-135A 74WG-1310A 74WG-1310A 74WG-1310A 74WG-1310A 74WG-1312A, B) 74WG-1312A, B) 74WG-1312A, B) 74WG-1312A, B) 74WG-1312A, B) 74WG-1302A 74WG-1802A 74WG-1802A, B) 74WG-1802A, B) 74WG-1802A, B) 74WG-1802A, B) 74WG-1802A, B) 74WG-1802A, B) 74WG-1802A, B) 74WG-1802A, B) 74WG-1802A, B) 74WG-1802A, B) 74WG-1807A, B) 74WG-1807A, B) 74WG-2004A 74WG-2004B, 74WG- 2007C (See Models 64WG-2007A, B)	58 - 4 - 1 - 1 - 35 - 1 - 1 - 35 - 1 - 4
74HA 8200A 74K8 - 210A 74K8 - 2706B 74K8 - 2710A 74K8 - 2706B 74K8 - 2713A 74WG - 1030C, D (See Model 64WG - 1030A), T 74WG - 1035A 74WG - 1037A 74WG - 1007A, 74WG 1007B (See Models 64WG - 1007A, 74WG 2007C (See Models 64WG - 2007A, B) 74WG - 2007A 74WG - 2007A, B) 74WG - 2007A 74WG - 2007A, B) 74WG - 2007A 74WG - 2007	58 - 4 - 1 - 1 - 35 - 1 - 1 - 35 - 1 - 4
74HA.8200A 74KR-1210A 74KR-2706B 74KR-2710A 74KR-2710A 74KR-2710A 74WG-1050C, D (See Model 64WG-1050A), See Models 64WG-1052A, B) 74WG-1054A 74WG-1054A 74WG-1054A 74WG-1054A) 74WG-1054A 74WG-1054A) 74WG-1054A 74WG-1054A) 74WG-1054A 74WG-1054A) 74WG-1054A 74WG-1054A) 74WG-1510A 74WG-1510A 74WG-1510A 74WG-1510A 74WG-1510A 74WG-1510A 74WG-1510A 74WG-1510A 74WG-1510A 74WG-1507A, B) 74WG-1607A, B) 74WG-1007A, 74WG- 1807A (See Models 64WG-2007A, CSee Models 64WG-2007A, CSEE 74WG-2007A, CSEE 74WG-2007A, CSEE 74WG-2007A, CSEE 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, SEE 74WG-2007A, B) 74WG-2007A, SEE 74WG-2007A, B) 74WG-2007A, SEE 74WG-2007A, SEE 74WG-2007A, CSEE 74WG-2007A, CSEE 74WG-2007A, CSEE 74WG-2007A, CSEE 74WG-2007A, CSEE 74WG-2007A, CSEE 74WG-2007A, SEE 74WG-2007A, SEE 74WG-2007A 74WG-200	58 - 4 - 1 - 1 - 35 - 1 - 1 - 35 - 1 - 1 - 35 - 1 - 22 - 24 - 6 - 1 - 0 - 9 - 22 - 1 - 1 - 22 - 22 - 1 - 1 - 22 - 2 -
74HA.8200A 74KR-1210A 74KR-2710A 74KR-2710A 74KR-2710A 74KR-2713A 74WG-1030C, D (See Model 64WG-1030A) 74WG-1032B (See Models 64WG-1052A, B) 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1311A, B 64WG-1311A, B 64WG-1311A, B 64WG-1311A, B 64WG-1311A, B 64WG-1312A, See Models 64WG-1312A, B 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-2004A, 72WG- 74WG-2004A, 72WG- 74WG-2007A, B 74WG-2007A, B 74WG-2007A, B 74WG-2007A, S 74WG-2007A, S 74W	58 - 4 - 1 - 1 - 35 - 1 - 1 - 35 - 1 - 1 - 35 - 1 - 2 - 24 - 6 - 1 - 0 - 2 - 2 - 2 - 2 - 2 - 2 - 2 - 2 - 2
74HA.8200A 74KR-1210A 74KR-2710A 74KR-2710A 74KR-2710A 74KR-2713A 74WG-1030C, D (See Model 64WG-1030A) 74WG-1032B (See Models 64WG-1052A, B) 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1311A, B 64WG-1311A, B 64WG-1311A, B 64WG-1311A, B 64WG-1311A, B 64WG-1312A, See Models 64WG-1312A, B 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-1802A 74WG-2004A, 72WG- 74WG-2004A, 72WG- 74WG-2007A, B 74WG-2007A, B 74WG-2007A, B 74WG-2007A, S 74WG-2007A, S 74W	58 - 4 - 1 - 1 - 35 - 1 - 1 - 35 - 1 - 1 - 35 - 1 - 22 - 24 - 6 - 1 - 0 - 9 - 22 - 1 - 1 - 22 - 22 - 1 - 1 - 22 - 2 -
74HA.8200A 74KR-210A 74KR-2706B 74KR-2706B 74KR-2706B 74KR-2705A 74WG-1050C, D (See Model 64WG-1050A) 74WG-1052B (See Models 64WG-1052A, B) 74WG-1054A 74WG-1054A 74WG-1054A 74WG-1054A 74WG-1054A 74WG-1054A 74WG-1054A 74WG-1054A 74WG-1054A 74WG-1510A 74WG-1510A 74WG-1510A 74WG-1511A, B) 64WG-1511A, B) 74WG-1511A, B) 74WG-1512A, B) 74WG-1802A 74WG-1802A 74WG-1802A, 74WG- 1807A (See Models 64WG-1804A, B) 74WG-1802A, 74WG- 1807A (See Models 64WG-1804A, B) 74WG-1802A, 74WG- 1807A (See Models 64WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, C) 74WG-2007A, C) 74WG-2007	58 - 4 - 1 - 1 - 35 - 1 - 1 - 35 - 1 - 1 - 35 - 1 - 2 - 24 - 6 - 1 - 2 - 2 - 2 - 2 - 2 - 2 - 2 - 2 - 2
74HA.8200A 74K8-1210A 74K8-27068 74K8-27068 74K8-2713A 74WG-1030C, D (See Model 64WG-1030A) 74WG-1032B (See Models 64WG-1052A, B) 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1357A 74WG-1357A 74WG-1311A, B) 64WG-1311A, B) 64WG-1311A, B) 74WG-1312A, B) 74WG-180A 74WG-200A 74WG-2	58 - 4 - 1 - 1 - 35 - 1 - 1 - 35 - 1 - 1 - 35 - 1 - 2 - 24 - 6 - 1 - 2 - 2 - 2 - 2 - 2 - 2 - 2 - 2 - 2
74HA.8200A 74K8-1210A 74K8-27068 74K8-27068 74K8-2713A 74WG-1030C, D (See Model 64WG-1030A) 74WG-1032B (See Models 64WG-1052A, B) 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1035A 74WG-1357A 74WG-1357A 74WG-1311A, B) 64WG-1311A, B) 64WG-1311A, B) 74WG-1312A, B) 74WG-180A 74WG-200A 74WG-2	58-4 41-1 35-1 43-2 24-6 10 9 22-1 22 18 27-1 18 25-4 25 4 5 26-4 27-2 5 6 18
74HA.8200A 74KR-1210A 74KR-2706B 74KR-2710A 74KR-2706B 74KR-2713A 74WG-1050C, D (See Model 64WG-1050A) 74WG-1052B (See Models 64WG-1052A, B) 74WG-1054A 74WG-1054A 74WG-1054A 74WG-1054A 74WG-1054A 74WG-1054A 74WG-1057A 74WG-1057A 74WG-1510A 74WG-1510A 74WG-1510A 74WG-1511A, B) 64WG-1511A, B) 74WG-1511A, B) 74WG-1512A, B) 74WG-1512A, B) 74WG-1802A 74WG-1802A 74WG-1802A, 74WG- 1807A (See Models 64WG-1807A, B) 74WG-1807A, B) 74WG-1807A, B) 74WG-1807A, B) 74WG-1807A, B) 74WG-1807A, B) 74WG-1807A, B) 74WG-1807A, 74WG- 18078 (See Models 64WG-2007B, 74WG- 18078 (See Models 64WG-2007B, 74WG- 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, C) 74WG-2007A, C) 7	58-4 41-1 35-1 43-2 24-6 10 9 22-1 22 18 27-1 18 25-4 25 4 5 26-4 27-2 5 6 18
74HA.8200A 74K8-210A 74K8-27068 74K8-27068 74K8-27068 74K8-27068 74K8-27068 74WG-1050C, D (See Model 64WG-1050A) 74WG-1052B (See Models 64WG-1052A, B) 74WG-1054A) 74WG-1054A 74WG-1054A) 74WG-1054A) 74WG-1054A) 74WG-1054A) 74WG-1054A) 74WG-1054A) 74WG-1054A) 74WG-1054A) 74WG-1510A 74WG-1510A 74WG-1510A 74WG-1510A 74WG-1511A, B) 64WG-1511A, B) 74WG-1511A, B) 74WG-1802A 74WG-1802A 74WG-1802A, 74WG- 1807A (See Models 64WG-1807A, B) 74WG-1807A, B) 74WG-1807A, B) 74WG-1807A, B) 74WG-1807A, B) 74WG-2007B, 74WG- 1807B (See Models 64WG-2007B, 74WG- 1807B (See Models 64WG-2007B, 74WG- 1807B (See Models 64WG-2007B, 74WG- 1807A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, See Models 64WG-2007A, B) 74WG-2007A, B) 74WG-2007A, See Models 64WG-2008 (See Models 64WG-2008 (See Models 64WG-2008, See Models 64WG-2008, See Models 64WG-2008, See Models 74WG-2008, See Models 74WG-2506A, See Model 74WG-2506A, See Models 74WG-2506A, See Models 74WG-2506A, See Models 74WG-2506A, See Models 74WG-2506A, See Models	$58-4 \\ 41-1 \\ 35-1 \\ 43-2 \\ 24-6 \\ 10 \\ 9 \\ 22-1 \\ 22 \\ 29-1 \\ 32-2 \\ 18 \\ 27-1 \\ 18 \\ 27-1 \\ 5 \\ 4 \\ 27-1 \\ 5 \\ 6 \\ 18 \\ 18-6 \\ 4 \\ 28-1 \\ 18 \\ 7 \\ 18 \\ 7 \\ 18 \\ 7 \\ 18 \\ 18 \\ 7 \\ 18 \\ 18 \\ 18 \\ 7 \\ 18 \\ 18 \\ 18 \\ 18 \\ 18 \\ 18 \\ 18 \\ 18$
74HA.8200A 74K8-210A 74K8-27068 74K8-27068 74K8-27068 74K8-27068 74K8-27068 74WG-1050C, D (See Model 64WG-1050A) 74WG-1052B (See Models 64WG-1052A, B) 74WG-1054A) 74WG-1054A 74WG-1054A) 74WG-1054A) 74WG-1054A) 74WG-1054A) 74WG-1054A) 74WG-1054A) 74WG-1054A) 74WG-1054A) 74WG-1510A 74WG-1510A 74WG-1510A 74WG-1510A 74WG-1511A, B) 64WG-1511A, B) 74WG-1511A, B) 74WG-1802A 74WG-1802A 74WG-1802A, 74WG- 1807A (See Models 64WG-1807A, B) 74WG-1807A, B) 74WG-1807A, B) 74WG-1807A, B) 74WG-1807A, B) 74WG-2007B, 74WG- 1807B (See Models 64WG-2007B, 74WG- 1807B (See Models 64WG-2007B, 74WG- 1807B (See Models 64WG-2007B, 74WG- 1807A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, See Models 64WG-2007A, B) 74WG-2007A, B) 74WG-2007A, See Models 64WG-2008 (See Models 64WG-2008 (See Models 64WG-2008, See Models 64WG-2008, See Models 64WG-2008, See Models 74WG-2008, See Models 74WG-2506A, See Model 74WG-2506A, See Models 74WG-2506A, See Models 74WG-2506A, See Models 74WG-2506A, See Models 74WG-2506A, See Models	$58-4 \\ 41-1 \\ 35-1 \\ 43-2 \\ 24-6 \\ 10 \\ 9 \\ 22-1 \\ 22 \\ 29-1 \\ 32-2 \\ 18 \\ 27-1 \\ 18 \\ 27-1 \\ 5 \\ 4 \\ 27-1 \\ 5 \\ 6 \\ 18 \\ 18-6 \\ 4 \\ 28-1 \\ 18 \\ 7 \\ 18 \\ 7 \\ 18 \\ 7 \\ 18 \\ 18 \\ 7 \\ 18 \\ 18 \\ 18 \\ 7 \\ 18 \\ 18 \\ 18 \\ 18 \\ 18 \\ 18 \\ 18 \\ 18$
74HA.8200A 74K8-210A 74K8-27068 74K8-27068 74K8-27068 74K8-27068 74K8-27068 74WG-1050C, D (See Model 64WG-1050A) 74WG-1052B (See Models 64WG-1052A, B) 74WG-1054A) 74WG-1054A 74WG-1054A) 74WG-1054A) 74WG-1054A) 74WG-1054A) 74WG-1054A) 74WG-1054A) 74WG-1054A) 74WG-1054A) 74WG-1510A 74WG-1510A 74WG-1510A 74WG-1510A 74WG-1511A, B) 64WG-1511A, B) 74WG-1511A, B) 74WG-1802A 74WG-1802A 74WG-1802A, 74WG- 1807A (See Models 64WG-1807A, B) 74WG-1807A, B) 74WG-1807A, B) 74WG-1807A, B) 74WG-1807A, B) 74WG-2007B, 74WG- 1807B (See Models 64WG-2007B, 74WG- 1807B (See Models 64WG-2007B, 74WG- 1807B (See Models 64WG-2007B, 74WG- 1807A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, See Models 64WG-2007A, B) 74WG-2007A, B) 74WG-2007A, See Models 64WG-2008 (See Models 64WG-2008 (See Models 64WG-2008, See Models 64WG-2008, See Models 64WG-2008, See Models 74WG-2008, See Models 74WG-2506A, See Model 74WG-2506A, See Models 74WG-2506A, See Models 74WG-2506A, See Models 74WG-2506A, See Models 74WG-2506A, See Models	$58-4 \\ 41-1 \\ 35-1 \\ 43-2 \\ 24-6 \\ 10 \\ 9 \\ 22-1 \\ 22 \\ 29-1 \\ 32-2 \\ 18 \\ 27-1 \\ 18 \\ 27-1 \\ 5 \\ 4 \\ 27-1 \\ 5 \\ 6 \\ 18 \\ 18-6 \\ 4 \\ 28-1 \\ 18 \\ 7 \\ 18 \\ 7 \\ 18 \\ 7 \\ 18 \\ 18 \\ 7 \\ 18 \\ 18 \\ 18 \\ 7 \\ 18 \\ 18 \\ 7 \\ 18 \\ 18 \\ 7 \\ 18 \\ 18 \\ 7 \\ 18 \\ 18 \\ 7 \\ 18 \\ 18 \\ 18 \\ 7 \\ 18 \\ 18 \\ 18 \\ 18 \\ 18 \\ 7 \\ 18 \\ 18 \\ 18 \\ 18 \\ 18 \\ 18 \\ 18 \\ 18$
74HA.8200A 74KR-1210A 74KR-2706B 74KR-2710A 74KR-2706B 74KR-2713A 74WG-1050C, D (See Model 64WG-1050A) 74WG-1052B (See Models 64WG-1052A, B) 74WG-1054A 74WG-1054A 74WG-1054A 74WG-1054A 74WG-1054A 74WG-1054A 74WG-1057A 74WG-1057A 74WG-1510A 74WG-1510A 74WG-1510A 74WG-1511A, B) 64WG-1511A, B) 74WG-1511A, B) 74WG-1512A, B) 74WG-1512A, B) 74WG-1802A 74WG-1802A 74WG-1802A, 74WG- 1807A (See Models 64WG-1807A, B) 74WG-1807A, B) 74WG-1807A, B) 74WG-1807A, B) 74WG-1807A, B) 74WG-1807A, B) 74WG-1807A, B) 74WG-1807A, 74WG- 18078 (See Models 64WG-2007B, 74WG- 18078 (See Models 64WG-2007B, 74WG- 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, B) 74WG-2007A, C) 74WG-2007A, C) 7	$58-4 \\ 41-1 \\ 35-1 \\ 43-2 \\ 24-6 \\ 10 \\ 9 \\ 22-1 \\ 22 \\ 29-1 \\ 32-2 \\ 18 \\ 27-1 \\ 18 \\ 27-1 \\ 5 \\ 4 \\ 27-1 \\ 5 \\ 6 \\ 18 \\ 18-6 \\ 4 \\ 28-1 \\ 18 \\ 7 \\ 18 \\ 7 \\ 18 \\ 7 \\ 18 \\ 18 \\ 7 \\ 18 \\ 18 \\ 18 \\ 7 \\ 18 \\ 18 \\ 7 \\ 18 \\ 18 \\ 7 \\ 18 \\ 18 \\ 7 \\ 18 \\ 18 \\ 7 \\ 18 \\ 18 \\ 18 \\ 7 \\ 18 \\ 18 \\ 18 \\ 18 \\ 18 \\ 7 \\ 18 \\ 18 \\ 18 \\ 18 \\ 18 \\ 18 \\ 18 \\ 18$

AIRLINE-Cont.
2705B (See Model
74WG-2505A) 18
AIRLINE-Cont. 74WG-27058, 74WG- 27058 [See Model 74WG-2505A) 18 74WG-2709A 26—5 74WG-2505A) 18 74WG-2505A) 18 848R-1065A 18 848R-1065A 18 848R-1065A 488R-1516A 848R-1516A 488R-1516A 848R-1815B, 848R-1816A 55—3 848R-1815B, 848R-1816B 55—3
74WG-2505A) 18
84BR-1065A
84BR-1503D, 84BR-1504D *
8458-1515A, 8458-1516A *
84BR-18158, 84BR-1816B . 55-3
848R-27158
84BR-2719A
84BR-3004 Tel. Rec *
84GAA3967A
84GDC-963B
84GDC-963B 51-3 84GDC-987A 53-4
Calourto/A Colourto/A Colourto/A 84GH4-9268 55-4 84GE 700-1 84GSE-2730A 70-1 84GSE-2731A 70-1 84GSE-2731A 70-1 84GSE-2731A 70-1 84GSE-2731A 84A-152A 70-1 82-1 84HA-1527A 84HA-152C1 67 84HA-152C2 67 84HA-1810A, 84HA-1830A 85-2 84HA-1810A 69-2
84GSE-2730A, 84GSE-2731A
84GSE-2731A
84HA-1527A, 84HA-1528A
(See Model 94HA-1527C) 67
84HA1529A, 84HA1530A . 85-2
84HA-1810C 69-2
84HA-2727A
84HA-3002A, 84HA-3002B
84HA-1810A, 84HA-1810C
84HA-3010A B C
Tel. Rec
84HA-3010A B C Tel
Bul, 11
84KR-1209A
84KR-1520A
Rec. Prod. Chge. Bul. 11 1181 84KR-1209A 6 4 84KR-1520A 564 84KR-2511A 84KR-2511A 684 84WG-1060A 421
84WG-1060A 42-1 84WG-1060C (See Model
84WG-1060A) 42
84WG-2015A 381
84WG-2506 (See Model
84WG-2506B 58-5
84WG-2712A
84WG-2712B (See Model
84WG-2714A
84WG-2714F, G, H, I 56-5
84WG-2718A, 84WG-
84WG.2721A B 46-3
84WG-2724A (See
Model 84WG-2718A) 45
84WG-2728A (See Models 84WG-2718A B.
84WG-2720A)
84WG-2732A, 8 (See
Model \$4WG-2712A, B} 43
84WG-2718A B-
BAWG-2016A GO-1 BAWG-1060C (See Model 82-1 BAWG-1060CA 42 BAWG-2015A 38-1 BAWG-2015A 38-1 BAWG-2005 (See Model 84 BAWG-2712A 38-1 BAWG-2712A 43 BAWG-2712A 36-2 BAWG-2712A 36-2 BAWG-2712A 36-2 BAWG-2712A, See Model 43 BAWG-2712A, See Models 44-3 BAWG-2712A, See Models 84WG-2712A, See Models BAWG-272AA, See Models 84WG-2712A, See Models BAWG-272AA, See Models 84WG-272AA, See Models BAWG-272AA, See Models 84WG-272AA, See Models BAWG-272AA, See Models 84WG-2718A, B BAWG-272AA, See Models 84WG-2718A, B
84WG-3006, 84WG-
3008, 84WG-3009,
3006A) Tel. Rec 72
94BR-1525A,
948R-1526A
9488-2740A. 9488-
2741A, B 89—1
94BR3004, C,
94BR3005, C Tel. Rec. 91A-3 94BR-3017A Tel. Rec. 89-2
94BR-3017B Tel. Rec.
Prod. Chge. Bul. 7110-1
94BR-3021, 94BR-3024A
9488-2740A, 9488- 2741A, 8
94GCB-1064A 96-2
94GCB-3023A, B, C
161. Rec. 95—1 94GAA3654A 95—1 94GAA3654A 96—2 94GCB-3023A, B, C 76—2 74GCB-3023A, B, C 76—1 76GCB-3023A, B, C 76 76GCB-3023A, B, C 76 94GCCB-3023A, B, C 76 76GCB-3023A, B, C 76 772A 24/25
94GDC-989A
94GDC-989A 94GSE-2735A, 94GSE-
2736A
94GSE-3011, B (See Model
94GSE-3011, B (See Model 84GSE-3011A)
94GSE-3011, B (See Model 84GSE-3011A)
94GSE-3011, B (See Model 84GSE-3011A)
94GSE3011,8(See Model 84GSE3011,8(See Model 84GSE3011A)
94GSE-3011, 8 (See Model 84GSE-3011, 8 (See Model 84GSE-3011A)
94058-3011,8158e Model 84658-3011A) - 82 94058-3015A Tei, Rec 107—2 94058-3015A Tei, Rec 93A-2 94058-3035A Tei, Rec 94058-3033A Tei, Rec 94051-3026A Tei, Rec 9404A-1527C, 941A-1528C 67—3 941A-1527A, 941A-1528A 67 (See Model 841A1529A) 85
94C35E-3011, B [See Model 84C3E-3011A)
94WG-1804D 86—2
94WG-1804D

ALGENE
AR5U 22—3 AR6U 22—4
ALTEC LANSING
ALC-101
ALC-205, ALC-206 Tel. Rec
A3238
Rec. 1053 A3238 662 A-323C (See Model ALC-101) 84
AMC
125P
AMERICAN COMMUNICATIONS (See Liberty)
AMPLIFIER CORP.
OF AMERICA ACA-100DC, ACA-100GE. 632
810DV*
AMPLIPHONE 10
10 21 —1 20 21 —12
ANDREA
ANDREA BT-VK12 Teil, Rec
CO-VK15, COVK16 (Ch. VK1516) Tel. Rec
CO-VK16 Tel. Rec.
COVK-125 Tel. Rec.
(See Model BI-VK12) 76 COVL-16 (Ch. VL16)
Tel. Rec
CO-VK16) Prod. Chge.
Bul. 8
(See Model BT-VK12)., 76 CVL-16 (Ch. VL16)
1e1. kec. (Supp. to C-VK19 Te1. kec. (Supp. to C0.VK16) Prod. Chge. 112—1 CVK-126 Te1. Rec. (See Model BT-VK12) 76 CVL-16 (Ch. VL16) Te1. Rec. Te1. Rec. (See Model COVL-16)
P-163 (Ch. 163)
T-UI6
(See Model BT-VK12) 76
(See Model 8T-VK12) 76
TVL-12 Tel. Rec
Tel. Rec. (See Model COVL-16) 125
VJ-12, VJ-12-2 Tel. Rec *
Ch. VK1516 (See Model
T-U15 24-7 T-U16 21-3 T-VK12 Teil, Rec. 21-3 T-VK12 Teil, Rec. 76 TVK-127E, M Teil, Rec. 123-3 TVL-16 (Ch. VL-16) 123-3 TVL-16 (Ch. VL-16) 125 VJ-15 Teil, Rec. 425 VJ-12, VJ-12-2 Teil, Rec. 425 Ch. VK1516 (See Model COVK15) COVK15) 103 Ch. VL16 (See Model 125
COVL-16)
32
53 24 8
701 Tel. Rec 71-6 APEX
485
192A 17
192A 17-6 APPROVED ELECTRONIC INSTRUMENT CORP.
192A
192A 17—6 APPROVED ELECTRONIC INSTRUMENT CORP. FM Tuner 41—2 ARC 601 25—5
192A 17-6 APPROVED ELECTRONIC INSTRUMENT CORP. FM Tuner 11-2 ARC
192A 17—6 APPROVED ELECTRONIC INSTRUMENT CORP. FM Tuner 41—2 ARC 601 25—5 ARCADIA 37D14-600 9—3 ARIA 4 4
192A 17—6 APPROVED ELECTRONIC INSTRUMENT CORP. FM Tuner 41—2 ARC 601 601 25—5 ARCADIA 9—3 ARIA 554-1-61A
192A 17—6 APPROVED ELECTRONIC INSTRUMENT CORP. FM Tuner 41—2 ARC 601 601 25—5 ARCADIA 9—3 ARIA 554-1-61A 554-1-61A 7—2 ARTHUR ANSLEY 14
192A 17—6 APPROVED ELECTRONIC INSTRUMENT CORP. FM Tuner 41—2 ARC 601 601 25—5 ARCADIA 9—3 ARIA 554-1-61A 554-1-61A 7—2 ARTHUR ANSLEY 14
192A 17—6 APPROVED ELECTRONIC INSTRUMENT CORP. FM Tuner 41—2 ARC 601 601 25—5 ARCADIA 9—3 ARIA 554-1-61A 554-1-61A 7—2 ARTHUR ANSLEY 14
122A 17—6 APPROVED ELECTRONIC INSTRUMENT CORP. FM Tuner 41—2 ARC 601 601 25—5 ARCADIA 9—3 ARIA 554-1-61A 554-1-61A 7—2 ARTHUR ANSLEY IP-2, IP-3 IP-4A 82—2 IP-5 108 IP-7 134-3 P-5 60-4
192A 17—6 APPROVED ELECTRONIC INSTRUMENT CORP. FM Tuner 41—2 ARC 601 601 25—5 ARCADIA 9—3 37D14-600 9—3 ARIA 7—2 ARHUR ANSLEY 1P-2. IP-3 62—4 IP-7 134—3 P-1 60—4
192A 17—6 APPROVED ELECTRONIC INSTRUMENT CORP. FM Tuner 41—2 ARC 601 601 25—5 ARCADIA 37D14-600 37D14-600 9—3 ARIA 7—2 ARTHUR ANSLEY IP-2, IP-3 IP-4 82—2 IP-5 [See Model P-5] 108 4 SP-1 60—4 SP-1 60—4 SP-1 60—4 SP-1 60—4 SP-1 50—1 324 76—6
192A 17—6 APPROVED ELECTRONIC INSRUMENT CORP. FM Tuner 41—2 ARC 601 25—5 ARCADIA 9—3 ARIA 554-1-61A 554-1-61A 7—2 ARTHUR ANSLEY IP-2, LP-3 IP-4 82—2 LP-5 (58e Model P-5) 108 487-1 ARTONE AR-23TV-1 Tel. Rec. 80—1 524 140-P (Ch, RE-209) 25—6
192A 17—6 APPROVED ELECTRONIC INSRUMENT CORP. FM Tuner 41—2 ARC 601 25—5 ARCADIA 9—3 ARIA 554-1-61A 554-1-61A 7—2 ARTHUR ANSLEY IP-2, LP-3 IP-4 82—2 LP-5 (58e Model P-5) 108 487-1 ARTONE AR-23TV-1 Tel. Rec. 80—1 524 140-P (Ch, RE-209) 25—6
192A 17—6 APPROVED ELECTRONIC INSRUMENT CORP. FM Tuner 41—2 ARC 601 25—5 ARCADIA 9—3 ARIA 554-1-61A 554-1-61A 7—2 ARTHUR ANSLEY IP-2, LP-3 IP-4 82—2 LP-5 (58e Model P-5) 108 487-1 ARTONE AR-23TV-1 Tel. Rec. 80—1 524 140-P (Ch, RE-209) 25—6
192A 17—6 APPROVED ELECTRONIC INSRUMENT CORP. FM Tuner 41—2 ARC 601 25—5 ARCADIA 9—3 ARIA 554-1-61A 554-1-61A 7—2 ARTHUR ANSLEY IP-2, LP-3 IP-4 82—2 LP-5 (58e Model P-5) 108 487-1 ARTONE AR-23TV-1 Tel. Rec. 80—1 524 140-P (Ch, RE-209) 25—6
192A 17—6 APPROVED ELECTRONIC INSRUMENT CORP. FM Tuner 41—2 ARC 601 25—5 ARCADIA 9—3 ARIA 554-1-61A 554-1-61A 7—2 ARTHUR ANSLEY IP-2, LP-3 IP-4 82—2 LP-5 (58e Model P-5) 108 487-1 ARTONE AR-23TV-1 Tel. Rec. 80—1 524 140-P (Ch, RE-209) 25—6
192A 17—6 APPROVED ELECTRONIC INSRUMENT CORP. FM Tuner 41—2 ARC 601 25—5 ARCADIA 9—3 ARIA 554-1-61A 554-1-61A 7—2 ARTHUR ANSLEY IP-2, LP-3 IP-4 82—2 LP-5 (58e Model P-5) 108 487-1 ARTONE AR-23TV-1 Tel. Rec. 80—1 524 140-P (Ch, RE-209) 25—6
192A 17—6 APPROVED ELECTRONIC INSRUMENT CORP. FM Tuner 41—2 ARC 601 25—5 ARCADIA 9—3 ARIA 554-1-61A 554-1-61A 7—2 ARTHUR ANSLEY IP-2, LP-3 IP-4 82—2 LP-5 (58e Model P-5) 108 487-1 ARTONE AR-23TV-1 Tel. Rec. 80—1 524 140-P (Ch, RE-209) 25—6
192A 17—6 APPROVED ELECTRONIC INSRUMENT CORP. FM Tuner 41—2 ARC 601 25—5 ARCADIA 9—3 ARIA 554-1-61A 554-1-61A 7—2 ARTHUR ANSLEY IP-2, LP-3 IP-4 82—2 LP-5 (58e Model P-5) 108 487-1 ARTONE AR-23TV-1 Tel. Rec. 80—1 524 140-P (Ch, RE-209) 25—6
192A 17—6 APPROVED ELECTRONIC INSRUMENT CORP. FM Tuner 41—2 ARC 601 25—5 ARCADIA 9—3 ARIA 554-1-61A 554-1-61A 7—2 ARTHUR ANSLEY IP-2, LP-3 IP-4 82—2 LP-5 (58e Model P-5) 108 487-1 ARTONE AR-23TV-1 Tel. Rec. 80—1 524 140-P (Ch, RE-209) 25—6
192A 17—6 APPROVED ELECTRONIC INSRUMENT CORP. FM Tuner 41—2 ARC 601 25—5 ARCADIA 9—3 ARIA 554-1-61A 554-1-61A 7—2 ARTHUR ANSLEY IP-2, LP-3 IP-4 82—2 LP-5 (58e Model P-5) 108 487-1 ARTONE AR-23TV-1 Tel. Rec. 80—1 524 140-P (Ch, RE-209) 25—6
192A 17—6 APPROVED ELECTRONIC INSRUMENT CORP. FM Tuner 41—2 ARC 601 25—5 ARCADIA 9—3 ARIA 554-1-61A 554-1-61A 7—2 ARTHUR ANSLEY IP-2, LP-3 IP-4 82—2 LP-5 (58e Model P-5) 108 487-1 ARTONE AR-23TV-1 Tel. Rec. 80—1 524 140-P (Ch, RE-209) 25—6
192A 17—6 APPROVED ELECTRONIC INSRUMENT CORP. FM Tuner 41—2 ARC 601 25—5 ARCADIA 9—3 ARIA 554-1-61A 554-1-61A 7—2 ARTHUR ANSLEY IP-2, LP-3 IP-4 82—2 LP-5 (58e Model P-5) 108 487-1 ARTONE AR-23TV-1 Tel. Rec. 80—1 524 140-P (Ch, RE-209) 25—6
192A 17—6 APPROVED ELECTRONIC INSRUMENT CORP. FM Tuner 41—2 ARC 601 25—5 ARCADIA 9—3 ARIA 554-1-61A 554-1-61A 7—2 ARTHUR ANSLEY IP-2, LP-3 IP-4 82—2 LP-5 (58e Model P-5) 108 487-1 ARTONE AR-23TV-1 Tel. Rec. 80—1 524 140-P (Ch, RE-209) 25—6
192A 17—6 APPROVED ELECTRONIC INSRUMENT CORP. FM Tuner 41—2 ARC 601 25—5 ARCADIA 9—3 ARIA 554-1-61A 554-1-61A 7—2 ARTHUR ANSLEY IP-2, LP-3 IP-4 82—2 LP-5 (58e Model P-5) 108 487-1 ARTONE AR-23TV-1 Tel. Rec. 80—1 524 140-P (Ch, RE-209) 25—6
192A 17—6 APPROVED ELECTRONIC INSRUMENT CORP. FM Tuner 41—2 ARC 601 25—5 ARCADIA 9—3 ARIA 554-1-61A 554-1-61A 7—2 ARTHUR ANSLEY IP-2, LP-3 IP-4 82—2 LP-5 (58e Model P-5) 108 487-1 ARTONE AR-23TV-1 Tel. Rec. 80—1 524 140-P (Ch, RE-209) 25—6
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122A 17—6 APPROVED ELECTRONIC INSTRUMENT CORP. FM Tuner 41—2 ARC 601 25—5 ARCADIA 9—3 37D14-600 9—3 ARIA 7—2 S54-1-61A 7—2 ARTHUR ANSLEY [P-3 [P-7 134—3 [AP-7 140-9 (Ch. RE-209)
122A 17—6 APPROVED ELECTRONIC INSTRUMENT CORP. FM Tuner 41—2 ARC 601 25—5 ARCADIA 9—3 37D14-600 9—3 ARIA 7—2 S54-1-61A 7—2 ARTHUR ANSLEY [P-3 IP-3 62—4 IP-4A 82—2 IP-5 [See Model P-5] IGE MC ARTONE 60—1 AR-23TV-1 Tel. Rec. 60—1 AR-23TV-1 Tel. Rec. 90—1 140-P (Ch. RE-209). 25—6 150-TC, 151-TC (Ch. RE-228) (Lote). 25—7 150TC, 151-TC (Ch. RE-228) (Lote). 25—7 150TC, 151-TC (Ch. RE-228)
$\begin{array}{c c c c c c c c c c c c c c c c c c c $
122A 17—6 APPROVED ELECTRONIC INSTRUMENT CORP. FM Tuner 41—2 ARC 601 25—5 ARCADIA 9—3 37D14-600 9—3 ARIA 7—2 S54-1-61A 7—2 ARTHUR ANSLEY [P-3 IP-3 62—4 IP-4A 82—2 IP-5 [See Model P-5] IGE MC ARTONE 60—1 AR-23TV-1 Tel. Rec. 60—1 AR-23TV-1 Tel. Rec. 90—1 140-P (Ch. RE-209). 25—6 150-TC, 151-TC (Ch. RE-228) (Lote). 25—7 150TC, 151-TC (Ch. RE-228) (Lote). 25—7 150TC, 151-TC (Ch. RE-228)

Nec. SZ2) Tel. A081 Tel. Rec. 104—2 A081 Tel. See Model 40807 104 162CM (Ch. TE-286) 130—3 Ch. RE-70 [See Model 442] 34 Ch. RE-200 (See Model 442) 34 44) 1 RE-200M [See Model 44M) 444 Ch. RE 444 Ch. RE 544 44M) 23 RE-201 (See Model 5.58 Ch. RE-229 (See Model 665) 18 Ch. RE-221 (See Model 552AN) 13 Ch. RE-231 (See Model 1607) 13 Ch. RE-232 (See Model 1527) 13 Ch. RE-233 (See Model 1827FM) 32 Ch. RE-237 (See Model 1827FM) 32 Ch. RE-242 (See Model 240P) 32 Ch. RE-244 (See Model 241P) 42 Ch. RE-244 (See Model 24271) 52 Ch. RE-248 (See Model 2337) 53 Ch. RE-253 (See Model 2337) 52 Ch. RE-254 (See Model 24271) 52 Ch. RE-253 (See Model 24307FM) 53 Ch. RE-254 (See Model 2407FM) 70 Ch. RE-254 (See Model 2407FM) 70 Ch. RE-254 (See Model 2417) 72 Ch. RE-254 (See Model 2417) 72 Ch. RE-254 (See Model 2417) 72 Ch. RE-254 (See Model 2417) 74 Ch. RE-254 (See Model 2417) 74 Ch. RE-254 (See Model 2417) 74 Ch. RE-264 (See Model 2417) 74 Ch. RE-264 (See Model 2417) 74 Ch. RE-264 (See Model 2417) 74

Ch

Ch

Ch

Ch Ch

Сh

Ch

Ch

Ch. Rt. 264T) 95.

ïs C

Ch

Ch

3500

RE-267 (See Mode!

356T) Ch. RE-274 (See Model 341T)

. 64

. 69

.100

78

ARVIN-Cont. h. RE-284/See Model 460T). 107 h. RE-287-1 (See Model 462-CB). 116 h. RE-288-1 (See Model 482CFB). 117 h. TE-272-1, 2 (See Model 3100TB). 80 h. TE-276 (See Model 3160CM) 93 h. TE282 (See Model 4080T) 104 h. TE-286 (See Model 4162CM). 130 Ch Ch Ch Ch Ch ASTRASONIC ATLAS AB-45 14-5 AUDAR
 AUDAR
 26-5

 P-IA
 5-10

 P-4A
 19-3

 P-5
 5-10

 P-5
 13-10

 PR-6
 19-4

 RE-8A
 19-4

 Teivar BM-25, BMP-25
 62-5

 Teivar RE-12
 35-2

 Teivar RE-12
 35-2
 'Bingo Amp.''.... 26-
 Automatic
 Automatic

 Tom Boy
 27-4

 Tom Thumb Buddy
 53-7

 Tom Thumb Comera-Radio 49-5
 26-7

 Tom Thumb Comera-Radio 49-5
 26-7

 Tom Thumb Personal ATTP 23-4
 26-7

 Catom Thumb Personal ATTP 23-4
 24-10

 C-63X (See Model C-60X)
 24

 C300
 102-1

 D200
 102-1

 D200
 104-3

 F-100
 103-6

 F-790
 23-6

 TV-707, TV-709, TV-707
 57-4

 Teil, Rec.
 60-5

 (See Model TV-1207)...
 60

 TV-1205 Teil, Rec.
 103

 TV-1205 Teil, Rec.
 103

 See Model TV-1249
 103

 TV-1245 Teil, Rec.
 105-17

 TV-124 Teil, Rec.
 103--5

 TV-124 Teil, Rec.
 103--5

 TV-5116k Tel, Rec.

 [See Model TV-5020].

 TV-5160 Tel, Rec.

 (See Model TV-5020].

 TVX313 Tel, Rec.

 (See Model TV-707).

 60

 601, 602 (Series A).

 134

 501, 602 (Series A).

 601, 602 (Series B).

 601, 602 (Series B).

 613X (See Model 612/301)

 614X, 616X

 8-2

 620

 12-3

 640, series B.

 10-4

 660, 662, 22--6

 677

 720

 21--4

 AVIOLA

 720

 AVIOLA

 509

 511

 601

 612

 612 (See Model 601)

 15

 16
 BELL SOUND SYSTEMS
 BELL SOUND SYSTEMS

 B-23
 75-4

 RC-47 (RE-CORD-O-FONE)
 30-4

 A401, 4405
 *Belfone

 2075
 10-5

 2122
 77-3

 2159

 3725
 22-8

 3725
 22-9

 3728
 22-9

 3728
 22-9

 3728
 24-11

 3730
 31-5

 FELLTONE
 BELLTONE 5-33 RV-11 (See Model RV-10) 46 500

BELMONT (Also See Raytheon)
A-6D110 17—7 3AW7 10—7
3AW7 10-7
4B115
4B112, 4B113 (Series A) 10-6
5D110
5P19 (Series A)
5P113 "Boulevard" 28-2
6D111
6D120
21A21 Tel. Rec 93A-4
SPI13 "Boulevard" 28-2 6D11 2-33 60120 24-12 8A59 6-1 24-12 24 21A21 Tel. Rec. 93A.4 22A21, 22AX21, 22AX22, 25AX22 Television Receiver. 55-5 55-5
BENDIX
C172 Tel. Rec
(See Model 2051) 111
C200 Tel, Rec.
(See Model C172)134
T170 Tel. Rec. {See Model 2051}111
(See Model 2051) 111
0526A, 0526B, 0526C, 0526D, 0526E, 0526F. 1-22
PAR 80
celo celo ceno ceno El 4
55X4
65P4
75B5, 75M5, 75M8,
75P6, 75W5 59—5 79M7 66—3
9583, 95M3, 95M9 60-7
110, 110W, 111, 111W,
112, 114, 115 41-3
MA. MB. MC MD1
7355, 7367, 7395, 7396, 7396, 7396, 7396, 7396, 66-3 59-5 79M7 66-3 9583, 95M3, 95M9, 60-7 60-7 710, 110W, 111, 111W, 112, 113, 114, 115, 114, 114
300, 300W, 301, 302 40-2
416A
A13 40_3
626-A (0626A) 12-4
636A, 636C
040A 2
656A
687A 61-3 697A 26-8 736B 10-8 847-8 27-5 847-5 'Focto Meter' 217D 29-4 1217D 1217D 1518 1519 1521
736B 10—8
847-B
1217 12178 1217D 29-4
1217D (Late) 46-5
1518, 1519, 1524, 1525. 37-3 1521 42-4
1521
2001, 2002 Tel. Rec 84-4
1521
[See Model 2001] 84
2025 Tel. Rec
2051 Tel. Rec.
Prod. Chge, Bul. 16 126-1
(See Model 2051)111
2060 Tel. Rec.
(See Model 2051)
2071 Tel. Rec.
2071 Tel. Rec. (See Model 2051)111
Prod. Chge. Bul. 16 120 2071 Tel. Rec. (See Model 2051) 111 2071 Tel. Rec. (See Model 2051)
Prod. Chge. Bul. 16 126 2071 Tel. Rec. (See Model 2051) 111 2071 Tel. Rec. (See Model 2051) 126
Prod. Chge. Bul. 16 120 2071 Tel. Rec. (See Model 2051) 111 2071 Tel. Rec. (See Model 2051) Prod. Chge. Bul. 16 126 3001, 3002 Tel. Rec.
Prod. Chge. Bul. 16 126 2071 Teil. Rec. (See Model 2051) 111 2071 Teil. Rec. (See Model 2051) Prod. Chge. Bul. 16 126 3001, 3002 Teil. Rec. (See Model 2001) 84 3030 3031 Teil. Rec
Prod. Lnge. Bul. 16 126 2071 Tel. Rec. [See Model 2051] 111 2071 Tel. Rec. [See Model 2051] Prod. Chege. Bul. 16 126 3001, 3002 Tel. Rec. [See Model 2001] 84 3030, 3031 Tel. Rec. [See Model 2001] 84
2001, 2002 Tel. Rec. 84-4 2020, 2021 Tel. Rec. 99-5 2023 Tel. Rec. 99-5 2031 Tel. Rec. 99-5 2031 Tel. Rec. 111-3 2040 Tel. Rec. 111-3 2051 Tel. Rec. 111-3 2060 Tel. Rec. 126-1 2060 Tel. Rec. 126-1 2060 Tel. Rec. [See Model 2051] Prod. Chage. Bul. 16 126 2071 Tel. Rec. [See Model 2051] [See Model 2051] 111 2071 Tel. Rec. [See Model 2051] [See Model 2051] 111 2071 Tel. Rec. 16 (See Model 2051] 111 2071 Tel. Rec. 16 (See Model 2051] 111 2071 Tel. Rec. 16 (See Model 2051] 113 207 Tel. Rec. 126 305 Tel. Rec. 126 305 Tel. Rec. 136 305 Tel. Rec. 136 305 Tel. Rec. 136 305 Tel. Rec. 39
3051 Tel. Rec.
(See Model 2051) 111 3051 Tel. Rec. (See Model 2051)111 3051 Tel. Rec. (See Model 2051) Prod. Chge. Bul. 16126 6001 Tel. Rec.
(Jee Model 2051)111 3051 Tel. Rec. (See Model 2051)111 3051 Tel. Rec. (See Model 2051)111 6001 Tel. Rec. (See Model 2051)111
See Model 2051)111 3051 Tel. Rec. [See Model 2051)111 3051 Tel. Rec. [See Model 2051] Prod. Chege Bul. 16126 6001 Tel. Rec. [See Model 2051]111 6001 Tel. Rec.
See Model 2051)111 3051 Tel. Rec. [See Model 2051)111 3051 Tel. Rec. [See Model 2051] Prod. Chege Bul. 16126 6001 Tel. Rec. [See Model 2051]111 6001 Tel. Rec.
See Model 2051)111 3051 Tel. Rec. [See Model 2051)111 3051 Tel. Rec. [See Model 2051] Prod. Chege Bul. 16126 6001 Tel. Rec. [See Model 2051]111 6001 Tel. Rec.
See Model 2051)111 3051 Tel. Rec. [See Model 2051)111 3051 Tel. Rec. [See Model 2051] Prod. Chege Bul. 16126 6001 Tel. Rec. [See Model 2051]111 6001 Tel. Rec.
See Model 2051)111 3051 Tel. Rec. [See Model 2051)111 3051 Tel. Rec. [See Model 2051] Prod. Chege Bul. 16126 6001 Tel. Rec. [See Model 2051]111 6001 Tel. Rec.
3051 Tel. Rec.
3051 Tel. Rec. 3052 Tel. Rec. 3053 Tel. Rec. 3053 Tel. Rec. 3054 Rec. 3053 Tel. Rec. 3053 Tel. Rec. See Model 2051)
3051 Tel. Rec. 3052 Tel. Rec. 3053 Tel. Rec. 3053 Tel. Rec. 3054 Rec. 3053 Tel. Rec. 3053 Tel. Rec. See Model 2051)
3051 Tel. Rec. 3052 Tel. Rec. 3053 Tel. Rec. 3053 Tel. Rec. 3054 Rec. 3053 Tel. Rec. 3053 Tel. Rec. See Model 2051)
3051 Tel. Rec. 3052 Tel. Rec. 3053 Tel. Rec. 3053 Tel. Rec. 3054 Rec. 3053 Tel. Rec. 3053 Tel. Rec. See Model 2051)
3051 Tel. Rec. 3052 Tel. Rec. 3053 Tel. Rec. 3053 Tel. Rec. 3054 Rec. 3053 Tel. Rec. 3053 Tel. Rec. See Model 2051)
3051 Tel. Rec. 3052 Tel. Rec. 3053 Tel. Rec. 3053 Tel. Rec. 3054 Rec. 3053 Tel. Rec. 3053 Tel. Rec. See Model 2051)
3051 Tel. Rec. 3052 Tel. Rec. 3053 Tel. Rec. 3053 Tel. Rec. 3054 Rec. 3053 Tel. Rec. 3053 Tel. Rec. See Model 2051)
358 Mole 2031 11 3051 Tel. Rec. 111 601 rel. Ches Bull. 16 126 6001 rel. Rec. 111 6001 rel. Rec. 111 6001 rel. Rec. 111 6001 rel. Rec. 126 6002 Tel. Rec. 99 6003 Tel. Rec. 111 6003 Tel. Rec. 111 6003 Tel. Rec. 126 1000 Tel. Rec. 126 7001 Tel. Rec. 126 75
3051 Tel. Rec. See Model 2051). 111 3051 Tel. Rec. See Model 2051). 111 6001 Tel. Rec. See Model 2051). 111 6001 Tel. Rec. See Model 2051). Prod. Chae. Bul. 16 126 6002 Tel. Rec. (See Model 2051). 111 6003 Tel. Rec. (See Model 2051). 111 6003 Tel. Rec. (See Model 2051). 111 6003 Tel. Rec. (See Model 2051). 111 6100 Tel. Rec. (See Model 2051). 111 6100 Tel. Rec. (See Model 2051). 111 6100 Tel. Rec. (See Model 2051). 111 7001 Tel. Rec. (See Model 2051). 111 7001 Tel. Rec.
3051 Tel. Rec. See Model 2051). 111 3051 Tel. Rec. See Model 2051). 111 6001 Tel. Rec. See Model 2051). 111 6001 Tel. Rec. See Model 2051). Prod. Chage. Bul. 16 6002 Tel. Rec. (See Model 2051). 111 6003 Tel. Rec. (See Model 2051). 111 6003 Tel. Rec. (See Model 2051). 111 6003 Tel. Rec. (See Model 2051). 111 6100 Tel. Rec. (See Model 2051). Prod. Chage. Bul. 16 (See Model 2051). Prod. Chage. Bul. 16 (See Model 2051). Prod. Chage. Bul. 16 See Model 2051). Prod. Chage. Bul. 16 See Model 2051). Prod. Chage. Bul. 16 See Model 2051).
3051 Tel. Rec. 3051 Tel. Rec. 15051 Tel. Rec. 15050 Tel. Rec. 1505 T
3051 Tel. Rec. Ste Model 2051) 111 3051 Tel. Rec. Ste Model 2051) <
3051 Tel. Rec. See Model 2051). 111 3051 Tel. Rec. See Model 2051). 111 6001 Tel. Rec. See Model 2051). 111 6001 Tel. Rec. See Model 2051). Prod. Chae. Bul. 16 126 6002 Tel. Rec. (See Model 2051). 111 6003 Tel. Rec. (See Model 2051). 111 6003 Tel. Rec. (See Model 2051). 111 6003 Tel. Rec. (See Model 2051). Prod. Chae. Bul. 16 See Model 2051) Prod. Chae. Bul. 16 See Model 2051) <td< td=""></td<>
3051 Tel. Rec. 3053 Tel. Rec. 3054 Tel. Rec. 3055 Tel. Rec. 3056 Tel. Rec. 3057 Tel. Rec. 3058 Tel. Rec. 3058 Tel. Rec. 3059 Tel. Rec. 3051 Tel. Rec. 3058 Tel. Rec. 3058 Tel. Rec. 3058 Tel. Rec. 3059 Tel. Rec. 3051 Tel. Rec. 3050 Tel. Rec.
3051 Tel. Rec. 3053 Tel. Rec. 3054 Tel. Rec. 3055 Tel. Rec. 3056 Tel. Rec. 3057 Tel. Rec. 3058 Tel. Rec. 3058 Tel. Rec. 3059 Tel. Rec. 3051 Tel. Rec. 3058 Tel. Rec. 3058 Tel. Rec. 3058 Tel. Rec. 3059 Tel. Rec. 3051 Tel. Rec. 3050 Tel. Rec.
3051 Tel. Rec. 3052 Tel. Rec. 3053 Tel. Rec. 3053 Tel. Rec. 3053 Tel. Rec. 3053 Tel. Rec. 3054 Tel. Rec. 3055 Tel. Rec. 3056 Tel. Rec. 3057 Tel. Rec. 3058 Tel. Rec. 3058 Tel. Rec. 3059 Tel. Rec. 3051 Tel. Rec.
3051 Tel. Rec. 3052 Tel. Rec. 3053 Tel. Rec. 3053 Tel. Rec. 3053 Tel. Rec. 3054 Tel. Rec. 3055 Tel. Rec. 3056 Tel. Rec. 3057 Tel. Rec. 3058 Tel. Rec. 3058 Tel. Rec. 3059 Tel. Rec. 3051 Tel. Rec.
Observed See Model 2051)
Observed See Model 2051)
3051 Tel. Rec. 3050 Tel. Rec.
3051 Tel. Rec. 3051 Tel. Rec. 3051 Tel. Rec. 3051 Tel. Rec. See Model 2051). 111 3051 Tel. Rec. See Model 2051). Prod. Chage. Bul. 16 See Model 2051). See Model 2051). Ster Model 2051). See Model 2051). See Model 2051). See Model 2051). Prod. Chage. Bul. 16 15 rel. Kec. (See Model 2051). See Model 2051). Prod. Chage. Bul. 16 16 See Model 2051). Prod. Chage. Bul. 16 176 See Model 2051). Prod. Chage. Bul. 16 18 Revore Model 2051) Prod. Chage. Bul. 16 126 See Model 2051) Prod. See David Bogen) BREWSTER 900 See Model 10C). 10C 10C 12A 12A 89-3 BROWN
3051 Tel. Rec. 3050 Tel. Rec.

BRUNSWICK BJ-8836 'Tuxcany,'' C-3300 'Darby' D-1000, D-1100 56 D-8767 'Buckingham'' See Model T-40001, 29 'Buckingham'' 1-4000, T-40001/s 'Buckingham'' 1-4000, T-40001/s 'Buckingham'' 1-60005, T-600055, T-60001, 29 1-2000, T-8001, Carbon, T-60001, 29 'Somodel T-4000, 29 1-2000, T-8001, Carbon, T-60001, 29 'Somodel T-4000, 29 1-2000, T-8001, Carbon, T-8000, 29 'Somodel T-4000, 29 1-2000, T-8000, Carbon, T-8000, 29 'Somodel T-4000, 29 1-2000, T-8000, Carbon, Carbon, 29 'Somodel T-4000, 29 1-2000, T-8000, Carbon, 20 'Somodel T-4000, 29 1-2000, T-8000, Carbon, 20 'Somodel T-4000, 29 1-201, Carbon, 20 'Somodel T-4000, 29 1-201, Carbon, 20 'Somodel T-4000, 29 1-201, Carbon, 20 'Somodel T-4000, 20 1-201, Carbon, 20 'Somodel T-4000, 20 1-201, Carbon, 20 'Somodel T-4000, 20 1-201, Carbon, 20 'Somodel T-4000, 20</td BRUNSWICK 28—4 56—7 29-5 61_4 922B, M Iel, Kec. 5000 5125 Tel, Rec. 6165 Tel, Rec. 8125 Tel, Rec. 8125 Tel, Rec. 42-5 BUICK 980540, 980733 18—9 980744, 980745 19—5 980747 62—6 980747 59—6 980782 62—6 980784 59—6 980868 104—4 BUTLER BROS. 58 CADILLAC * 7253203 * 7253203 * 7258155 * 7258255 109—2 7260205 (See Model 7258755) 109 CALLMASTER (See Lyman) CAPEMART BUICK CALLMASTER (See Lyman) CAPEMART B-504.P16 Tel. Rec. (See Model 401P Set 87 and 35P7 Set 93A) P0, P10 TC-20 (Ch. C-297).....132--4 19N4, 21P4, 24N4, 24P4, 26N4, 29P4, 30P4, 31N4, 31P4, 24N4, 24P4, 26N4, 29P4, 30P4, 31N4, 31P4, 24N4, 24P4, 26N4, 29P4, 30P4, 31P0, 11P4, 24N4, 24P4, 26N4, 29P4, 30P4, 11B74 (See Model 327P). 64 35P7 (Ch. P7).....93A.4 11B74 (See Model 327M)...112 320-B, 320-M (Ch. CX-33L) Tel. Rec. [See Model 323M)...112 320-B, 320-M (Ch. CX-33L) Tel. Rec. [See Model 323M)...112 320-B, 320-M (Ch. CX-33L) Tel. Rec. [See Model 323M]...122--1 321-B, 321-M, 322-B, 322-M (Ch. CX-33) Tel. Rec. (See Model 323M) 112 321-B, 321-M, 322-B, 322-M (Ch. CX-33) Tel. Rec. (See Model 323M) 112 321-B, 321-M, 322-B, 322-M (Ch. CX-33) Tel. Rec. (See Model 323M) 112 324-M, 325-F, 325-M (Ch. CX-331) Tel. Rec. (See Model 323M) 112 324-M, 325-F, 325-M (Ch. CX-331) Tel. Rec. (See Model 323M) 112 324-M, 325-F, 325-M (Ch. CX-331) Tel. Rec. (See Model 323M) 112 324-M, 325-F, 325-M (Ch. CX-331) Tel. Rec. (See Model 323M) 112 324-M, 325-F, 325-M (Ch. CX-331) Tel. Rec. (See Model 323M) 112 324-M, 325-F, 325-M (Ch. CX-331) Tel. Rec. (See Model 323M) 112 332-B, 332-M, 334-M (Ch. CX-331) Tel. Rec. [See Model 115P2]... 67 601P, 602P12 Tel. Rec. 97-2 501P, 502P, 504P Tel. Rec. (See Model 401P Set 87 and 35P7 Set 93A) 601-M (Ch. CX-3) Fred. 93A-5 3005-M (Ch. CX-3), Fred. C-2721 Tel. Rec. 93A-5 3005-M (Ch. CX-3), Fred. C-2721 Tel. Rec. (See Model 323M)... 12 304-100-4M (Ch. CX-3), Fred. C-2741 Tel. Rec. (See Model 3004-M) 93A 3007 (Ch. CX-3), Fred. C-2741 Tel. Rec. See Model 3004-M) 93A 3007 (Ch. CX-3), Fred. C-2741 Tel. Rec. See Model 3004-M) 93A 4002-M (Ch. CX-3), Fred. C-2741 Tel. Rec. See Model 3004-M) 93A 4002-M (Ch. CX-3), Fred. C-2741 Tel. Rec. See Model 3004-M) 93A CAPITOL 30—4 28—5 29—6 D-17 T-13 U-24 CARDWELL, ALLEN D.

CE-26

..... 14-6

ARVIN-CONCORD

CENTURY (Also See Industrial Television) 226, 326 (Ch. IT-26R, IT-35R, IT-37R, IT-46R) Tel. Rec. 721, 821, 921, 1021 (Ch. IT-21R) Tel. Rec. 994.7 97 A-8 12 21-5 21-6 CHALLENGER
 CHALLENGER

 CC8

 CC30

 CC30

 CC64

 CC63

 CC64

 CC64

 CO6

 200

 600

 CSee Model 208]

 600 (See Model 608)
 63-4 67 ____7 68---6 70---3 66-4 69---5 62---7 69 62 CHANCELLOR (See Radionic) 35P 30-25 CHEVROLET 985792 985793 985986 986067 986146 986146 986240 986241 986241 986241 19-0 90---2 28---6 75---5 58--7 CHRYSLER (See Mopar) CISCO 1A5 9A5 37-4 9A5 CLARION C100 C101 C102 C103 9-ó C103 C104 C105 (See Model C104). C105A C108 (Ch. 101). 150 155 155 157 Tel. Rec. 11011 11305 11411-N 11801 11801 11801 11801 11801 6---6 1---4 -7 6—7 5—8 95A-1 17-8 18-11 23-6
 11802V-M (See Model
 23

 11801)
 23

 12100M
 54

 12300
 34

 12708
 41

 12801
 61

 13101
 46

 13201
 1303

 4601
 60

 14905
 66

 16703 Tel. Rec.
 102

 CLARK
 102
 -5 31-6 61-5 46-7 62---8 60---9 66-5 10/03 fei. kec..... CLARX PA-10 PA-10A PA-20 PA-20 PA-20A PA-30 12-6 18-12 13-12 18-13 19-7 CLEARSONIC See U. S. Television)
 See U. 5. Television)

 COLLINS AUDIO PRODUCTS

 FMA-6
 99-6

 45-D
 72-6

 COLLINS RADIO
 72-6

 COLINS RADIO
 72-6

 COLINS RADIO
 72-6

 COMANDER INDUSTRIES
 Commander 3 Tube

 Commander 3 Tube
 Record Player

 Record Player
 17-10

 CONCORD
 19-9

 IN343 (IN435, IN436
 [Similar to Chassis] 121-2

 IN434 (Similar to Chassis] 38-5
 1N351 (Similar to Chassis] 38-5

 IN351 (Similar to Chassis) 38-5
 IN354, IN555

 Isimilar to Chassis] 38-5
 IN351 (Similar to Chassis) 38-5

 INS51 [Similar to Chassis]
 38—6

 INS54, INS55
 ISimilar to Chassis].
 55—10

 INS56, INS57
 ISimilar to Chassis]
 90—7

 INS56 (Similar to Chassis]
 90—7

 INS56 (Similar to Chassis]
 90—7

 INS60 (Similar to Chassis]
 90—7

 INS61, INS62
 19—8

 GCS1W (See Model GCS1B)
 90—6

 GR3ARC
 21—7

 7R3APW (See Model GCS2
 20—5

 1.402, 1.403
 45—5

 1.501 (See GES1B)
 20
 21 22---1 20--5 45---6 48--5 20 1-411 1-501 (See 6E51B)..... 1-301 [See 6C318] 1-504 1-506, 1-510 [See 6C318] 1-516, 1-517 1-601, 1-602, 1-603 [See 7G26C] 1-606 1-608 [See 6F26W] 1-609 [See 6F26W] 1-609 [See 6F26W] 1-601 1-201 2-105 [See 315W] 55-6 504 49-7 20 _7 19 -8 55-7 1-1201 2-105 (See 315WL).... 53 54---6 2-106 2-200, 2-201, 2-218, 2-219, 2-232, 2-235, 2-236, 2-237, 2-238, 2-239, 2-240 315WL, 315WM 325WL, 325WM (See 2-106) 2-106 62—9 53—8 54

CONRAC-EMERSON

CONRAC-EMERSON
CONRAC
36, 39, Tel. Rec110-4
(See Skyweight)
CONVERSA-FONE
MS-5 (Master Station)
SS-5 (Sub-Station) 16-7
6AWC2, 6AWC3,
6AWC2, 6AWC3, 6A47WCR, 6A47WT, 6A47WTR
CORONADO
FA43-8965 (See Model
FA43-8966 Tel. Rec *
TV43-8908 Tel. Rec *
05RA1-43-7755A, 05RA1- 43-7755B
05RA1-43-7901A
05RA4-43-8935A Tel. Rec. *
05RA4-43-9876A 103-7 05RA33-43-8120A110-6
05RA33-43-8120A 110-6 05RA37-43-8360A 1023 05TV1-43-8945A Tel Rec *
CORONADO FA43.8965 (See Model 43.8965 [Sel, Rec
05TV1-43-9014A Tel. Rec. 128-4
05TV2-43-8950A Tel. Rec. * 05TV2-43-9010A Tel. Rec. *
05TV6-43-8935A Tel. Rec. * 15RA1-43-7902A
15TV1-43-9008A Tel. Rec. *
43-9013A Tel. Rec *
43-2027
43-6485
43-6485
43-7601, 43-7602 (See Model 43-7601B) 10
43-8685) 11 43-7601, 43.7602 (See Model 43.76018) 10 43-76018 10-11 14 43-7651 9-7 14.7651 43-7651 9-7 14.7651 43-7651 9-7 14.7651
43-7652 (See Model
43-7651)
43-8101 (See Model 94RA31-43-8115A) 81
43-8130C, 43-8131C (See Model 94RA33-
43.813001 83
43-8177 (See Model
43-8178) 21 43-8178 21—8
43-8180 10-12 43-8190 19-11
43-8201 (See 43-8178) 21 43-8213 7-5
43.8240 43.8241 12_8
43-8305
43-8330 43-8351, 43-8352 19—12 12—9
43-8353, 43-8354 28 -7 43-8420
43-8420 (See Model
43-8305)
43-8312A) 8 43-8576B
43-8685
43-9196
43-9201
94RA1-43-6945A 69-6 94RA1-43-7605A 65-5 94RA1-43-7656A, 65-5
94PA1-43-7657A 73-2
94RA1-43-8510A.
94RA1-43-8511A 71—7 94RA1-43-8510B, 94RA1-43-8511B 75—6
94PA4_43-8129A
94RA4-43-8130A, 94RA4-43-8130B
94RA4-43-8130A, 94RA4-43-8130B, 94RA4-43-8131A, 94RA4-43-8131A, 94RA4-43-8131B,
94RA31-43-8115A, B,
94RA31-43-8116A 81—5 94RA31-43-9841A 79—3
94RA33-43-8130C, 94RA33-43-8131C 82—3
94KA33-43-8131C 82—3 94TV1-43-8940A Tel. Rec. * 94TV1-43-9002A Tel. Rec. * 94TV2-43-8970A, 94TV2-43-8971A, 94TV2-43-8972A,
94TV2-43-8970A,
941V2-43-89/1A,
94TV2-43-8972A,
94TV2-43-8972A, 94TV2-43-8973A, 94TV2-43-8985A,
941 V2-43-8973A, 94T V2-43-8985A, 94T V2-43-8986A, 94T V2-43-8986A,
941 V2-43-8973A, 94T V2-43-8985A, 94T V2-43-8986A, 94T V2-43-8986A,
941V2-43-8975A, 941V2-43-8985A, 941V2-43-8986A, 941V2-43-8987A, 941V2-43-8993A, 941V2-43-8993A, 941V2-43-8995A
9417243.873A, 947243.8785A, 947243.8785A, 947243.8787A, 947243.8797A, 947243.8797A, 947243.8797A, 7477243.8795A, 78-4 9474.8795A, 78-4 9474.8795A, 78-4 9474.8795A, 78-4 9474.8795A, 78-4 9474.8795A, 78-4 9474.8795A, 78-4 9474.8795A, 78-4 9474.8795A, 78-4 9474.9795A, 78-4 9474.9795A, 78-4 9474.9795A, 78-4 9474.9795A, 78-4 9474.9795A, 78-4 9474.9795A, 78-4 9474.9795A, 78-4 9474.9795A, 78-4 9474.9795A, 78-4 9474.9795A, 78-4 9474.9795A, 78-4 9474.9795A, 94754, 9475
9417243-873A, 947243-8785A, 947243-8785A, 947243-8787A, 947243-8797A, 947243-8797A, 947243-8797A, 747243-8795A, 78-4 9474-800523-10-8-78-4
V41V2-43-87/3A, 94TV2-43-8985A, 94TV2-43-8986A, 94TV2-43-8993A, 94TV2-43-8993A, 94TV2-43-8993A, 94TV2-43-8993A, 94TV2-43-8953A Tel, Rec. 106-3 165 (See Model 94R841-43-8115A) 81
Va1V2.43-87/3A, 94TV2.43-8985A, 94TV2.43-8986A, 94TV2.43-8987A, 94TV2.43-8993A, 94TV2.43-8995A Tel, Rec. 78-4 94TV6.43-8953A Tel, Rec. 106-3 165 (See Model 94RA31-43-8115A) 81 197, 197U (See Model 94RA31-43-8115A) 81
V41V2-43-87/3A, 94TV2-43-8985A, 94TV2-43-8986A, 94TV2-43-8993A, 94TV2-43-8993A, 94TV2-43-8993A, 94TV2-43-8993A, 94TV2-43-8953A Tel, Rec. 106-3 165 (See Model 94R841-43-8115A) 81
Va1V2.43-87/3A, 94TV2.43-8985A, 94TV2.43-8986A, 94TV2.43-8987A, 94TV2.43-8993A, 94TV2.43-8995A, 94TV2.43-8995A Tel, Rec
vitv.43.803A, vitv.43.803A, vitv.43.803A, vitv.43.803A, vitv.43.803A, vitv.43.803A, vitv.43.803A, vitv.43.805A, 81 vitv.43.8015A, 81 vitv.43.8015A, 81 vitv.43.8015A, 81 vitv.43.8015A, 81 vitv.43.8015A, 81 vitv.43.8015A, 81 vitv.43.8015A, <td< td=""></td<>
va1v2.43-87/3A, va1v2.43-87/3A, va1v2.43-8985A, va1v2.43-8987A, va1v2.43-8987A, va1v2.43-8997A, va1v2.43-8973A, va1v2.43-8973A, va1v2.43-8974A,
varv2.43-br/3A.
varv2.43-br/3A.

CONRAC 36, 39, Tel. Rec	CROSLEY-Cont. 9-103, 9-104W	CR
CONTINENTAL ELECTRONICS	9-105, 9-105W 5970 9-113, 9-114W 53-9	D/ Ba
(See Skyweight) CONVERSA-FONE		ME
MS-5 (Master Station) SS-5 (Sub-Station) 16-7	9-118W (See Model 9-102) 50 9-119, 9-120W	10
CO-OP	9-204, 9-205M	DB
6AWC2, 6AWC3, 6A47WCR, 6A47WT, 6A47WTR	9-209, 9-212M 53-10 9-213B (See Model 9-209) 53	E6 F7
CORONADO	9-214M, 9-214ML	EI
FA43-8965 (See Model 43-8965) Tel. Rec 86 FA43-8966 Tel. Pec.	Rec	EX G-
FA43-8965 (See Model 43-8965) Tel. Rec 86 FA43-8966 Tel. Rec * TV43-8908 Tel. Rec * TV43-8908 Tel. Rec * CSPA1 43 - 2756A *	(See Model 9-403M) 79 9-407, 9-407M-1,	GC
05RA1-43-7755A, 05RA1- 43-7755A, 05RA1- 43-7755B, 012 05RA2-43-8515A, 1152 05RA2-43-8935A, 10-5 05RA2-43-8935A, 10-5 05RA3-43-8976A, 1037 05RA3-43-8120A, 105 05RA3-43-8945A, 10-, Rec. * 05TV1-43-9045A, 10-, Rec. * 05TV1-43-905A, 10-, Rec. * 05TV2-43-8950A, 10-, Rec. * 05TV2-43-8950A, 10-, Rec. * 05TV2-43-8950A, 10-, Rec. * 05TV2-43-8950A, 10-, Rec. * 05TV2-43-9014A, 1	7-40 am 1ct. Rec. 79 (See Model 9-403M) 79 9-407.9.407M-1, 9 66-6 9-407M-2 Tel. Rec 94-3 9-413B.9.2-414B 74-14B 74-14B Tel. Rec. (See Model 6-416B 79	H1 H3
05RA1-43-7901A	9-4138, 9-4138-2, 9-4148 Tel. Rec. (See Model 9-403M)	H5 H6
05RA4-43-9876A 103-7 05RA33-43-8120A 110-6	1e1. ret. (See Model 9-403M)	HC HC HC
05RA37-43-8360A 1023 05TVI-43-8945A Tei. Rec.	9-419M3-LD Tel, Rec. (See Model 9-409M3) 94	нх
05TV1-43-9005A, 05TV1- 43-9006A Tel. Rec	9-420M Tel. Rec. (See Model 9-403M) 79	ιc
051V1-43-9014A Tel. Rec. 128-4 05TV2-43-8950A Tel. Rec. *	9-422M, 9-422MA Tel. Rec. 81—6 9-423M Tel. Rec	PH
05TV6-43-8935A Tel. Rec. * 15RA1-43-7902A	(See Model 9.403M) 79	PX PX R5
13RA1-43-7902A 134—6 15TV1-43-9008A Tel, Rec. * 15TV2-43-9012A, 15TV2- 43-9013A Tel, Rec *	9-425 Tel. Rec. 95A-2 10-135, 10-136E, 10-137, 10-138, 10-139, 10-140 93—3 10-307M, 10-308, 10-309 80—4 10-401 V-1 Part	R6 UP
43-2027 11-3	10-30/M, 10-308, 10-309 804 10-401 Tel. Rec	2A 11
43-5005 28-36 43-6301 7-4 43-6451 10-10 43-6485 46-9	Tel. Rec	11 11: 21
43-0/30 (See Model	10-307/M, 10-308, 10-307 804 10-401 Tel. Rec. 952 10-404MU, 10-404M1U Tel. Rec. 1143 10-412MU Tel. Rec. (See Model 10-404MU).114 10-414MU Tel. Rec. (See Model 10-414MU).116	21
43-8685)	10-416MU Tel. Rec. (See Model 10-414MU) 116 10-418MU Tel. Rec. (See Model 10-404MU).114 10 4104U Tel. Rec.	10
43-7601B 10-11 43-7651	(See Model 10-404MU).114 10-419MU Tel. Rec104-6	DP
43-7652 (See Model 43-7651)	10-419MU Tel. Rec 104-6 10-420MU Tel. Rec. (See Model 10-404MU).114	DP
43-8101 (See Model	10-421MU Tel. Rec 106-4 10-427MU Tel. Rec 125-1A	DI
	10-429MU Tel. Rec. (See Model 10-414MU) 116	R- R- R-
43-8130C, 43-8131C (See Model 94RA33- 43-8130C)	11-100U, 11-101U, 11-102U, 11-103U,	R-
43-8178)	[See Modei 10-404MU].114 10-421MU Tel. Rec 106-4 10-427MU Tel. Rec 125-14 10-428MU Tel. Rec 129-5 10-429MU Tel. Rec [See Model 10-414MU] 116 11-100U, 11-101U, 11-104U, 11-105U, 11-104U, 11-105U, 11-126U, 11-127U, 11-126U, 11-1	R- R- R-
43-8180 10—12 43-8190	11-1280, 11-1290 (Ch. 312)	P. 1
43-8213		R- R- R-
43-8305 8-3	11-3050, 11-3040 11-3050, 11-3040 11-41740, 11-3050 11-44220, 11-3050 11-44220, 11-3050 11-44220, 11-4422, 126 14-4420, 11-4422, 126 14-4420, 11-4422, 126 14-4420, 11-4422, 11	R- R- R-
43-8312A 8-4 43-8330 19-12 43-8351, 43-8352 12-9 43-8353, 43-8354 28-7 43-8420 24-13 43-8470 (See Model	11-443MU Tel. Rec. (See Model 11-442) 126	R-
43-8353, 43-8354 28 —7 43-8420	11-445MU, 11-446MU,	RI TV TV
43-8470 (See Model 43-8305)	(See Model 11-442MU) 126 11-453MU Tel. Rec. (See Model 11-442MU)126	
43-8312A)	11-459MIU, MU, 11-460MU Tel. Rec.	De
43-8965 Tel. Rec 86-3		55
43-9201	(See Model 11-442MU) 120-4 11-461WU Tel. Rec. * (See Model 11-442MU) 126 11-470BU Tel. Rec. (See Model 11-442MU) 126	55 56 57
94RA1-43-6945A 69—6 94RA1-43-7605A 65—5 94RA1-43-7656A,	Modei 11-442MU) 126 11-471BU Tel. Rec	57
94RA1-43-7657A 73—2 94RA1-43-7751A 87—3	11-472BU Tel. Rec. {See Model 11-442MU} 126 11-473BU Tel. Rec.	57 57
94RA1-43-8510A, 94RA1-43-8511A 71-7 94RA1-43-8510B,	(See Model 11-442)120 11-4758U, 11-4768U,	57 57 58
94RA1-43-8511B 75—6 94RA4-43-8129A,	11-4778U Tel. Rec. (See Model 11-442MU) 126	61
94RA4-43-8130A, 94RA4-43-8130B,	11-483BU Tel. Rec. (See Model 11-442MU) 126 46FA. 46FB 15-5	62
94RA4-43-8131A, 94RA4-43-8131B 62—10 94RA31-43-8115A, B,	56FA, 56FB, 56FC 31-7 56PA, 56PB	72 D
94RA31-43-8116A 81—5 94RA31-43-9841A 79—3 94RA33-43-8130C,	56TD	AS AS
94RA33-43-8131C 82—3	56TG	A-
94TV1-43-8940A Tel. Rec. * 94TV1-43-9002A Tel. Rec. * 94TV2-43-8970A,	Models 56TA-L, 56TC-L) 4	A- Ad
94TV2-43-8971A, 94TV2-43-8972A,	56TP 8—5 56TO, 56TZ 33—2 56TR, 56TS 17—11 56TU 10—13	8- B-
94TV2-43-8973A, 94TV2-43-8985A, 94TV2-43-8986A,	56TU	8- B-
94TV2-43-8987A, 94TV2-43-8993A,	58TK	B- B-
94TV2-43-8994A, 94TV2-43-8995A	58TW	8- B- B-
Tel. Rec. 78-4 94TV6-43-8953A Tel. Rec. 106-3 165 (See Model	66CS, 66CSM	B- B-
165 (See Model 94RA31-43-8115A) 81 197, 197U (See Model		BT C- C-
94RA31-43-8115A) 81 CORONET	8/CQ [Kevised Models	с- С1
C2 6—8	88CR (See Model 87CQ) 36 88TA, 88TC	CI
CRESCENT H-16A1 76—8	88TA, 88TC (Revised) 43-8 106CP 106CS 76	D- D-
CROMWELL (Mercantile Stores)	146CS	D- D: D:
1010 88—2 1020 89—5	148CP, 148CQ, 148CR 42 —6 307TA Tel. Rec 348CP-TR1, 348CP-TR2, 348CP-TR3 Tel. Rec *	D- D1 D1
CROSLEY 9-101	CROSLEY CAR	D
9-102 50-4	5MX080 ****	

RYSTAL PRODUCTS (See Coronet) ALBAR	
arcombo Jr., Barcambo Sr. 8 ''Tonomatic'' 20-1000 Series	10—14 8—34 10—15 9—9
AVID BOGEN	02-4 854 832
X35 K-326 -50 O-50	76 —9 30 —6 26 —9 22 —12 25 —11 80 —6 79 —5 78 —6
O-125 X50 15 30, H150, H2150, 223 DH, HOL D50 D125 X50 DH, LOL (See Model HOH) 16 16 10 10 10 10 10 10 10 10 10 10	71—8 80—5 84—5 87—4 82—4 75—7 80
116 116 110 110 110 110 110 110	86 4 73 3 68 5 72 7 33 3 67 8 86 28 8 86 28 8 77 5 76 10 77 2 77 76 74
EARBORN	2213
ECCA P-11 P-29	24—15 19—13 25—12
ELCO -705 -1227 P.1228 P.1229	42—7 15—6
1230-A, R-1231-A, R-1232-A	14-33
ELCO 705 727, R-1228, R-1229, 1207, R-1228, R-1229, 1230, R-1231-A, R-1232-A 1233, 1234, R-1235 1234, R-1235 1241 1242 1243, R-1245, R-1246, 1244, R-1245, R-1246, 1246, R-1249, R-1250, 1251, R-1252 1268, R-1409 1410 -/71, TV-71A Tel. Rec +/00, TV-102 Tel. Rec +/00, TV-102 Tel. Rec +/00, TV-102 Tel. Rec +/00, TV-102 Tel. Rec +/00, Tel. Rec	42-8 7-7 29-7 38-4 62-11 31-8 32-4 52-6 66-7 21-10 47-7 15-7 99A-3 88-3 85-5 59-8
ETROLA 54-1-61A (See Aria	
Model 534-1-61A) 58-1-49A 58-13-221D 71, 571A, 571B, 571L, 571AL, 571BL 71X, 571AX, 571BX 72-220-226A	7 78 910 1016 911 86
72-220-226A 79 79-2-58B (See Model 579) 82 10-A 11-A 26 Series 156 270	8-7 7-9 7 19-14 55-8 50-6 11-5 48-6 16-8
FILLAL D	
Few ALD 500, A 5001, A 500W, A 501, A 502, A 503 .504, A 505 .507 .514 .602, A 605 .604, A 505 .604, A 505 .602, A 605 .604, Scee Model A 6021. .401 .402 .403 .504 .504 .506 .506 .506 .506 .506 .506 .506 .506 .512	38-5
510 512 513 614 614 616 617 617 	35-4 63-6 42-9 56-9 79-6 64-4 69-7 79
	82—5 106—5 131—4 100—5 38 102—5 100—6 82
(See Model DT-120)	100

DEWALD-Cont. DT-162, DT-163 Tel, Rec., 118—5 Tel, Rec., 1163 Tel, Rec., 118—5 Tel, Rec., 118—5 Tel, Rec., 118 DT-190 Tel, Rec., 118 DT-190D Tel, Rec., 118 DT-190D Tel, Rec., 118 DT-190D Tel, Rec., 100 DT-1020, DT-1020A Tel, Rec. (See Model DT-120) 100 DT-X-160 Tel, Rec., 128—5 ET-140, ET-141 Tel, Rec., 128—5 ET-140, ET-141 Tel, Rec., 118 ET-140R, ET-141 Rec., 118 ET-140R, ET-141 Rec., 148 ET-170, ET-171 Tel, Rec., 149 Still, 149 DODGE (See Moper) DODGE (See Mopar)
 DODGE (See Moper)

 DUMONT

 RA-101 Tel, Rec.

 RA-102B1, RA-10282,

 RA-1032 Tel, Rec.

 RA-1033 Tel, Rec.

 Prod. Chge, Bul, 6...108—1

 RA-1032 Tel, Rec.

 Prod. Chge, Bul, 6...108—1

 RA-1032 Tel, Rec.

 Prod. Chge, Bul, 6...108—1

 RA-1032 Tel, Rec.

 See, Prod. Chge, Bul, 9. 114—1

 RA-1035 Tel, Rec.

 Prod. Chge, Bul, 9. 114—1

 RA-105 Tel, Rec.

 Prod. Chge, Bul, 9. 114—1

 RA-105 Tel, Rec.

 Prod. Chge, Bul, 6...108—1

 RA-106 Tel, Rec.

 Prod. Chge, Bul, 6...108—1

 RA-106 Tel, Rec.

 Prod. Chge, Bul, 6...108—1

 RA-106 Tel, Rec.

 RA-106 Tel, Rec.

 See Model RA-105B.

 RA-106 Tel, Rec.

 Prod. Chge, Bul, 6...108—1

 RA-106 Tel, Rec.

 Prod. Chge, Bul, 6...108—1
 DUMONT DUOSONIC K1, K2 K3, K4 19—15 19—16 DYNAVOX M-510 Swingmaster 3-P-801 28-_9 15.--8 27---7 36-3 ECA 101 (Ch. AA)...... 102 104 105 106 106 FCA 1-25 14-7 13-14 16-11 7-10 3-6 13-15 16-12 45-9 15-9 32-5 108 121 131 132 201 204 ECHOPHONE (Also See Hallicrafters) EC-1A EC113 3-13 14-8 22-14 4-18 64-5 EC-306 EC-403, EC-404 EC-600 EC-600 EX-102, EX-103 EX-306 (See Model EC-306) 14 EDWARDS Fidelotuner 33-4 FLCAR 5---19 602 . . . ELECTONE T5TS3 12—34 ELECTRO 14—9 B20 ELECTROMATIC APH301-A, APH301-C ... 7—11 606A, 607A 5—32 ELECTRO-TONE 555 13-706, 712 (See Model 555) 13 ELECTRONIC CORP. OF AMERICA (See ECA) ELECTRONIC SPECIALTY CO. (See Ranger) E/L (ELECTRONIC LABS.) EMERSON 501, 502 (Ch. 120000, 120029)

.. 2-1

 EMERSON-Cont.

 503 (Ch. 120000, 120029)

 (See Models 501, 502).

 2

 505 (Ch. 120001) (See

 Model 523).

 506

 507

 508 (Ch. 120008).

 7-12

 509 (See Model 507).

 8

 510, 510A

 (Ch. 120000, 120029).

 511 (Ch. 120000, 120029).

 512 (Ch. 120006).

 9-12

 512 (Ch. 120006).

 9-12

 512 (Ch. 120006).

 9-12

 512 (Ch. 1200056).

 26-11

 514 (Ch. 1200050).

 (See Model 507).

 8

 515 (516 (Ch. 1200056)

 (See Model 507).

 8

 517 (Ch. 1200050)

 (See Model 507).

 8

 519 (Ch. 1200050)

 30-7

 520 (Ch. 1200030).

 319 (Ch. 1200030).

 321 (Ch. 1200031, 120021)

 5323 (Ch. EMERSON-Cont. .13 20-10 16-13 7
 540A
 [Ch. 120042]
 20-10

 541
 16-13

 542
 (See Model S21)
 7

 543
 (See (Ch. 120042)
 19-30

 545
 (Ch. 120047)
 11-15

 546
 (Ch. 120047)
 21-15

 547
 (Ch. 120051)
 30-8

 549
 (Ch. 120051)
 30-8

 540
 (Ch. 120051)
 30-8

 540
 (Ch. 120051)
 30-8

 540
 (Ch. 120051)
 26

 550
 (Ch. 120050)
 26

 551
 (See Model S23)
 20

 553
 (See Model S23)
 20

 554
 (Ch. 120038)
 70-4

 555
 (Ch. 120038)
 70-4

 558
 (Ch. 120038)
 70-4

 558
 (Ch. 120038)
 70-4

 56
 (Ch. 120038)
 70-4

 578
 (Ch. 120038)
 70

 56
 (Ch. 120037)
 73-4

 56
 (Ch. 120038)
 70

 56
 (Ch. 12 586 (Ch. 1200238, 1200338) 587 (Ch. 120033A, B)... 588 (See Model 547A). 590 (Ch. 120101A, B)... 591 (Chassis 1200638) (See Models 563). 594, 595 (Ch. 120071A) (See Model 581]. 597 (Ch. 1200738). 597 (Ch. 1200738). 590 (Ch. 320378). 72—9 71—10 25 87—5 67—9 73 68 61 90-69-_8 600 (Chassis 120103-B) 87-6 Tel. Rec.

EMERSON-Cont. 600 (Ch. 120103-B) Tel. Rec. Prod. Chge. Bul. 9. 114-1 601 (Chassis 120075B) 56—10
 But Construction
 But Construction<

 01 B (Ch. 1200702) Tel.

 61 P (Ch. 1200702) Tel.

 Rec. (3ee Model 571

 Ch. 1200808)

 71 Ch. 1200808)

 757 Ch. 1200808)

 76

 621 (Ch. 1200910-QD)

 77 Ch. 1200808)

 76

 621 (Ch. 1200980)

 76

 621 (Ch. 1200808)

 76

 622 (Ch. 1200878)

 Tel. Rec. (See Model 501)

 71 Ch. 12008780)

 824 (Ch. 12008780)

 76

 623 (Ch. 1200878)

 76

 624 (Ch. 12008780)

 76

 625 (Ch. 120184)

 76

 626 (Ch. 1201848)

 76

 627 (Ch. 1201848)

 76

 628 (Ch. 12019078)

 71 Ch. 1200888)

 76

 628 (Ch. 12019078)

 77 Ch. 1200888)

 76

 628 (Ch. 12019078)

 77 Ch. 1200888)

 76

 627 (Ch. 1201978)

 77 Ch. 1200888)

 76

 630 (Ch. 1201978)

 76

EMERSON-Cont FADA-Cont.
 120127-01161. rec.

 [See Model 660%]
 126

 1003
 [See Model 607%]

 11003
 [See Model 607%]

 12003
 [See Model 607%]

 137
 Ch. 120033

 120033
 [See Model 607%]

 137
 Ch. 120033

 137
 Ch. 120033

 137
 Ch. 120033

 138
 Ch. 120033

 138
 Ch. 120033

 138
 Ch. 120033

 139
 Ch. 120033

 140
 Ch. 120033</t ESPEY (Also see Philharmonic)
 Ester
 (Also see Frindrimonic)

 R13, R13, ...
 13—17

 76
 47—8

 188
 90—7

 31
 103—9

 512
 68—8

 513, 514
 63—8

 31
 03-1

 512
 512

 513
 514

 524
 (See Model 18B)

 90
 581

 641
 14

 621
 10

 643
 651

 751
 (See Model 18B)

 90
 581

 651
 651

 751
 (See Model 18B)

 90
 6511, 6515

 6514, 6516, 6517, 6520,

 6514, 6516, 6517, 6520,

 6510, 6517, 6520,

 6511

 6514

 6514

 6514

 6514

 6514

 6514

 6514

 6514

 6514

 6514

 6514

 6540

 6541

 8
 -10 -17 -11 -14 ESQUIRE 60-10, 65-4 14-11 FADA -

PIDA-LOHT. P100 R-1025 Tel. Rec. (See Model R-1025)1 S4C20 Tel. Rec. S4C40 Tel. Rec. S4T15 Tel. Rec. S4T30 Tel. Rec. S4C35 Tel. Rec	27-10
10	14-4
(See Model R-1025)	14
S4C40 Tel. Rec.	:
S4T30 Tel. Rec	*
S6C70 Tel. Rec.	
[See Model S6C55]] S6T65 Tel. Rec.	34
(See Model \$6C55)1 \$7C70 Tel. Rec. (See Model \$6C55)1 \$7T65 Tel. Rec. (See Model \$6C55)1 \$9C10 Tel. Rec. (See Model \$6C55)1 \$1015 Tel. Rec. \$1015 Tel. Rec. Model \$1015)1	34
(See Model S6C35)1	34
(See Model S6C55)1	34
(See Model S6C55)1	34
S1015 Tel. Rec. (See	09-4
Model S1015)1 S1030 Tel. Rec. (See	09
Model S1015)	09
(See Model S6C55)] S1065 Tel. Rec. (See Model S6C55)] TV30 Tel. Rec	34
(See Model S6C55)1	34
TV30 Tel. Rec. 602 605, 606 Series. 609, 610 Series. 633 637	14—12 1—13
609, 610 Series	1-15
637	17-14
637	123 327 2810
711, 740	2810 646
795 799 Tel. Rec. (See Model TV30)	36—6
Model TV30)	74 97—5
845	97-6
880 Tel. Rec	92-2 95A-5
880 Tel. Rec. 899 Tel. Rec. (See Model TV30)	74
Model TV30) 925 (See Model G-925). 930, 940 Tel. Rec. (See Model TV30) 945 (See Model G-925)	89
	74 89
1000 Series 1001	1-17
EADNEWODTH	
EC-260 EK-081, EK-082, EK-083.	7
EK-262, EK-263BL, E-263WL, E-264BL	
EC-260 EK-081, EK-082, EK-083 EK-262, EK-263BL, E-263WL, E-264BL, EK-264WL, EK-264BL, EK-264WL, EK-265 (See Model EC-260) EK-681 (See Model EK-081)	7
EK-681 (See Model EK-081) ET-060 ET-061 ET-063	26 6—11
ET-064, ET-065, ET-066.	42
GK-100, GK-102, GK-103, GK-104	238
GK-114, GK-115	60-11
GK-140, GK-141, GR-142, GK-143, GK-144	24-18
GT-050, GT-051, GT-052.	
GI-060, GI-061, GI-064,	35-5
GT-060, GT-061, GT-064, GT-065 GV220, GV240, GV260	35—5 35—6
GT-060, GT-061, GT-064, GT-065 GV220, GV240, GV260 Tel. Rec. K-267, K-669 (See Model	35—5 35—6
GT-060, GT-061, GT-064, GT-065 GV220, GV240, GV260 Tel. Rec. K-267, K-669 (See Model EC-260). Ch. 150 (See Model	35—5 35—6
EK-681 (See Model EK-081) ET-064, ET-063, ET-063, ET-064, ET-064, ET-065, ET-066, GK-102, GK-102, GK-104, GK-114, GK-112, GK-114, GK-112, GK-144, GK-145, GK-144, GK-150, GK-164, GT-064, GT-0	35—5 35—6
	35—5 35—6 * 7
ET-060) Ch. 152, 153 (See Model EC-260) Ch. 156, 157 (See Model EK-081) Ch. 159 (See Model	35—5 35—6 * 7 6
ET-060) Ch. 152, 153 (See Model EC-260) Ch. 156, 157 (See Model EK-081) Ch. 159 (See Model	35—5 35—6 * 7 6 7
ET-060) Ch. 152, 153 (See Model EC-260) Ch. 156, 157 (See Model EK-081) Ch. 159 (See Model	355 356 * 7 6 7 26
ET-060] Ch. 152, 153 (See Model EC-260) Ch. 156, 157 (See Model EK-081) Ch. 158, 159 (See Model ET-064) Ch. 162 (See Model EC-260) Ch. 170 (See Model GK-100) Ch. 170 (See Model GK-100)	355 356 * 7 6 7 26 4 7
ET-060] Ch. 152, 153 (See Model EC-260) Ch. 156, 157 (See Model EK-081) Ch. 158, 159 (See Model ET-064) Ch. 162 (See Model EC-260) Ch. 170 (See Model CK-100) Ch. 133 (See Model	355 356 * 7 6 7 26 4 7
ET-060] Ch. 152, 153 (See Model EC-260) Ch. 156, 157 (See Model EK-081) Ch. 158, 159 (See Model ET-064) Ch. 162 (See Model EC-260) Ch. 170 (See Model CK-100) Ch. 133 (See Model	355 356 * 7 6 7 26 4 7 23
ET-060] Ch. 152, 153 (See Model EC-260) Ch. 156, 157 (See Model EK-081) Ch. 162 (See Model ET-064) Ch. 162 (See Model EC-260) Ch. 170 (See Model CK-100) Ch. 193 (See Model EK-081) Ch. 194, 201, 216 (See Model GK-100) FEDERAL MFG. CO.	35-5 35-6 * 7 6 7 26 4 7 23 26 23
ET-060] Ch. 152, 153 (See Model EC-260) Ch. 156, 157 (See Model EK-081) Ch. 162 (See Model ET-064) Ch. 162 (See Model EC-260) Ch. 170 (See Model CK-100) Ch. 193 (See Model EK-081) Ch. 194, 201, 216 (See Model GK-100) FEDERAL MFG. CO.	35-5 35-6 * 7 6 7 26 4 7 23 26
ET-060] Ch. 152, 153 (See Model EC-260) Ch. 156, 157 (See Model EK-081) Ch. 158, 159 (See Model ET-064) Ch. 120 (See Model EC-260) Ch. 120 (See Model EK-100) Ch. 120 (See Model EK-081) Ch. 124, 201, 216 (See Model GK-100) FEDERAL MFG.CO. 104 (Select-A-Call) 135 (Select-A-Call)	35-5 35-6 * 7 6 7 26 4 7 23 26 23 18-17 11-7 CORP.
ET-060] Ch. 152, 153 (See Model EC-260) Ch. 156, 157 (See Model EK-081) Ch. 158, 159 (See Model ET-064) Ch. 120 (See Model EC-260) Ch. 120 (See Model GK-100) Ch. 120 (See Model EK-081) FEDERAL MFG. CO. 104 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 136 (Select-A-Call) 137 (See Model 10307). 10907	35-5 35-6 * 7 6 7 26 4 7 23 26 23 18-17 11-7
ET-060] Ch. 152, 153 (See Model EC-260) Ch. 156, 157 (See Model EK-081) Ch. 158, 159 (See Model ET-064) Ch. 120 (See Model EC-260) Ch. 120 (See Model GK-100) Ch. 120 (See Model EK-081) FEDERAL MFG. CO. 104 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 136 (Select-A-Call) 137 (See Model 10307). 10907	35-5 35-6 7 6 7 26 4 7 23 26 23 18-17 11-7 8 8-13 8
ET-060] Ch. 152, 153 (See Model EC-260) Ch. 156, 157 (See Model EK-081) Ch. 158, 159 (See Model ET-064) Ch. 120 (See Model EC-260) Ch. 120 (See Model GK-100) Ch. 120 (See Model EK-081) FEDERAL MFG. CO. 104 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 136 (Select-A-Call) 137 (See Model 10307). 10907	35-5 35-6 * 7 6 7 26 4 7 23 26 23 18-17 11-7 CORP. 8 8-13
ET-060] Ch. 152, 153 (See Model EC-260) Ch. 156, 157 (See Model EK-081) Ch. 158, 159 (See Model ET-064) Ch. 126 (See Model EC-260) Ch. 120 (See Model Ch. 120 (See Model Ch. 120 (See Model EK-081) FEDERAL MFG. CO. 104 (Select-A-Coll) 135 (Select-A-Coll) 135 (Select-A-Coll) 135 (Select-A-Coll) 135 (Select-A-Coll) 136 (Select-A-Coll) 137 (See Model 10307) 10307 10307 10307 10307 10307 10307 10307 10307 15407 (See Model 10307) 15407 (See Model 10307) FERRAR	35-5 35-6 * 7 6 7 26 4 7 23 26 23 18-17 11-7 8 8-13 8 23-9 8
ET-060] Ch. 152, 153 (See Model EC-260) Ch. 156, 157 (See Model EK-081) Ch. 158, 159 (See Model ET-064) Ch. 126 (See Model EC-260) Ch. 120 (See Model Ch. 120 (See Model Ch. 120 (See Model EK-081) FEDERAL MFG. CO. 104 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 143 (Select-A-Call) 143 (Select-A-Call) 143 (Select-A-Call) 143 (Select-A-Call) 143 (Select-A-Call) 150 (Select-A-Call) 10307 10307 10307 10307 (See Model 10307) . 15407 (See Model 10307) . FERRAR C-81-8 -618	35-5 35-6 * 7 6 7 26 4 7 23 26 23 18-17 11-7 8 8-13 8 23-9 8
ET-060] Ch. 152, 153 (See Model EC-260) Ch. 156, 157 (See Model EK-081) Ch. 158, 159 (See Model ET-064) Ch. 126 (See Model EC-260) Ch. 120 (See Model Ch. 120 (See Model Ch. 120 (See Model EK-081) FEDERAL MFG. CO. 104 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 143 (Select-A-Call) 135 (Select-A-Call) 136 (Select-A-Call) 137 (See Model 10307) 10307 10307 10307 (See Model 10307) 10407 10407 [See Model 10307] 15407 (See Model 10307) FERRAR C-81-8 -618 WR-11	35-5 35-6 * 7 6 7 26 4 7 23 26 23 18-17 11-7 CORP. 8 8-13 8 23-9
ET-060] Ch. 152, 153 (See Model EC-260) Ch. 156, 157 (See Model EK-081) Ch. 158, 159 (See Model ET-064) Ch. 120 (See Model EC-260) Ch. 120 (See Model GK-100) FEDERAL MFG. CO. 104 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 133 (Select-A-Call) 133 (Select-A-Call) 133 (Select-A-Call) 133 (Select-A-Call) 133 (Select-A-Call) 133 (Select-A-Call) 133 (Select-A-Call) 134 (Salect-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) FERRAL MCAL 5407 (See Model 10307) FERRAR C-81-B T-61B WR-11	35-5 35-6 * 7 6 7 26 4 7 23 26 23 18-17 11-7 CORP. $\frac{8}{8}-13$ $\frac{8}{39-4}$ 15-10
ET-060] Ch. 152, 153 (See Model EC-260) Ch. 156, 157 (See Model EK-081) Ch. 158, 159 (See Model ET-064) Ch. 120 (See Model EC-260) Ch. 120 (See Model GK-100) FEDERAL MFG. CO. 104 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 133 (Select-A-Call) 133 (Select-A-Call) 133 (Select-A-Call) 133 (Select-A-Call) 133 (Select-A-Call) 133 (Select-A-Call) 133 (Select-A-Call) 134 (Salect-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) FERRAL MCAL 5407 (See Model 10307) FERRAR C-81-B T-61B WR-11	35-5 35-6 * 7 6 7 26 4 7 23 26 23 18-17 11-7 CORP. 8 8-13 8 8-13 8 -9 8 -9 8 -9 -9 -9 -9 -9 -9 -9 -9 -9 -9
ET-060] Ch. 152, 153 (See Model EC-260) Ch. 156, 157 (See Model EK-081) Ch. 158, 159 (See Model ET-064) Ch. 120 (See Model EC-260) Ch. 120 (See Model GK-100) FEDERAL MFG. CO. 104 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 133 (Select-A-Call) 133 (Select-A-Call) 133 (Select-A-Call) 133 (Select-A-Call) 133 (Select-A-Call) 133 (Select-A-Call) 133 (Select-A-Call) 134 (Salect-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) FERRAL MCAL 5407 (See Model 10307) FERRAR C-81-B T-61B WR-11	35-5 35-6 * 7 6 7 26 4 7 23 26 23 18-17 11-7 CORP. 8 8-13 8 8-13 8 -9 8 -9 8 -9 -9 -9 -9 -9 -9 -9 -9 -9 -9
ET-060] Ch. 152, 153 (See Model EC-260) Ch. 156, 157 (See Model EK-081) Ch. 158, 159 (See Model ET-064) Ch. 120 (See Model EC-260) Ch. 120 (See Model GK-100) FEDERAL MFG. CO. 104 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 133 (Select-A-Call) 133 (Select-A-Call) 133 (Select-A-Call) 133 (Select-A-Call) 133 (Select-A-Call) 133 (Select-A-Call) 133 (Select-A-Call) 134 (Salect-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) FERRAL MCAL 5407 (See Model 10307) FERRAR C-81-B T-61B WR-11	35-5 35-6 * 7 6 7 26 4 7 23 26 23 18-17 11-7 CORP. 8 8-13 8 8-13 8 -9 8 -9 8 -9 -9 -9 -9 -9 -9 -9 -9 -9 -9
ET-060] Ch. 152, 153 (See Model EC-260) Ch. 156, 157 (See Model EK-081) Ch. 158, 159 (See Model ET-064) Ch. 128 (See Model EC-260) Ch. 120 (See Model Ch. 120 (See Model Ch. 120 (See Model Ch. 124, 201, 216 (See Model CK-100). FEDERAL MEG. CO. 104 (Select-A-Call) TEDERAL MEG. CO. 104 (Select-A-Call) FEDERAL TEL. & RADIO 1021 (See Model 1030T). 1030T 1032 (See Model 1030T). 1030T 1032 (See Model 1030T). 1040T, 1040TB 1540T (See Model 1030T). FERAR C-81-8 T-618 T-618 MR-11 FIRESTONE (AIR CHIEF) 4-A-2 (Code No. 297-6.LMMU-143) -4-A-10 (Code No. 297-7.RN228] -4-A-11 (Code	35-5 35-6 * 7 6 7 26 4 7 23 26 23 18-17 11-7 CORP. 8 8-13 8 8-13 8 -9 8 -9 8 -9 -9 -9 -9 -9 -9 -9 -9 -9 -9
ET-060] Ch. 152, 153 (See Model EC-260) Ch. 156, 157 (See Model EK-081) Ch. 158, 159 (See Model ET-064) Ch. 128 (See Model EC-260) Ch. 120 (See Model Ch. 120 (See Model Ch. 120 (See Model Ch. 124, 201, 216 (See Model CK-100). FEDERAL MEG. CO. 104 (Select-A-Call) TEDERAL MEG. CO. 104 (Select-A-Call) FEDERAL TEL. & RADIO 1021 (See Model 1030T). 1030T 1032 (See Model 1030T). 1030T 1032 (See Model 1030T). 1040T, 1040TB 1540T (See Model 1030T). FERAR C-81-8 T-618 T-618 MR-11 FIRESTONE (AIR CHIEF) 4-A-2 (Code No. 297-6.LMMU-143) -4-A-10 (Code No. 297-7.RN228] -4-A-11 (Code	35-5 35-6 * 7 6 7 26 4 7 23 26 23 18-17 17-16 8-13 8-13 8-17 17-16 39-4 15-10 14-4 31-13 28-11 41-7
ET-060] Ch. 152, 153 (See Model EC-260) Ch. 156, 157 (See Model EK-081) Ch. 158, 159 (See Model ET-064) Ch. 128 (See Model EC-260) Ch. 120 (See Model Ch. 120 (See Model Ch. 120 (See Model Ch. 124, 201, 216 (See Model CK-100). FEDERAL MEG. CO. 104 (Select-A-Call) TEDERAL MEG. CO. 104 (Select-A-Call) FEDERAL TEL. & RADIO 1021 (See Model 1030T). 1030T 1032 (See Model 1030T). 1030T 1032 (See Model 1030T). 1040T, 1040TB 1540T (See Model 1030T). FERAR C-81-8 T-618 T-618 MR-11 FIRESTONE (AIR CHIEF) 4-A-2 (Code No. 297-6.LMMU-143) -4-A-10 (Code No. 297-7.RN228] -4-A-11 (Code	35-5 35-6 * 7 6 7 26 4 7 23 26 23 18-17 11-7 CORP. 8 8-13 8 8-13 8 -9 8 -9 8 -9 -9 -9 -9 -9 -9 -9 -9 -9 -9
ET-060] Ch. 152, 153 (See Model EC-260) Ch. 156, 157 (See Model EK-081) Ch. 158, 159 (See Model ET-064) Ch. 128 (See Model EC-260) Ch. 120 (See Model Ch. 120 (See Model Ch. 120 (See Model Ch. 124, 201, 216 (See Model CK-100). FEDERAL MEG. CO. 104 (Select-A-Call) TEDERAL MEG. CO. 104 (Select-A-Call) FEDERAL TEL. & RADIO 1021 (See Model 1030T). 1030T 1032 (See Model 1030T). 1030T 1032 (See Model 1030T). 1040T, 1040TB 1540T (See Model 1030T). FERAR C-81-8 T-618 T-618 MR-11 FIRESTONE (AIR CHIEF) 4-A-2 (Code No. 297-6.LMMU-143) -4-A-10 (Code No. 297-7.RN228] -4-A-11 (Code	35-5 35-6 * 7 6 7 26 4 7 23 26 23 18-17 17-16 8-13 8-13 8-17 17-16 39-4 15-10 14-4 31-13 28-11 41-7
ET-060] ET-060] ECh 152, 153 (See Model EC-260) ECh 156, 157 (See Model EK-081) ECh 158, 159 (See Model EC-064) ECh 120 (See Model EC-260) Ch 120 (See Model EC-260) Ch 120 (See Model EK-1001) FEOFRAL VILL & RADIO 1021 (See Model 10307] 10307 10317 10307 10307 10317 10307	35-5 35-6 * 7 6 7 26 4 7 23 26 23 18-17 17-7 CORP. $\frac{8}{8-13}$ $\frac{8}{39-4}$ 15-10 14-4 31-13 28-11 41-7 49-8 36-7 35-7 15-11
ET-060] ET-060] ECh 152, 153 (See Model EC-260) ECh 156, 157 (See Model EK-081) ECh 158, 159 (See Model EC-064) ECh 120 (See Model EC-260) Ch 120 (See Model EC-260) Ch 120 (See Model EK-1001) FEOFRAL VILL & RADIO 1021 (See Model 10307] 10307 10317 10307 10307 10317 10307	35-5 35-6 35-6 7 26 4 7 26 4 7 23 26 23 18-17 1-7 CORP. $\frac{8}{8}-13$ $\frac{8}{3}-9$ $\frac{8}{3}-4$ 15-10 14-4 31-13 28-11 41-7 49-8 36-7 35-7 15-11 11-19 2-29
ET-060] Ch. 152, 153 (See Model EC-260) Ch. 156, 157 (See Model EK-081) Ch. 158, 159 (See Model ET-064) Ch. 126 (See Model EC-260) Ch. 120 (See Model Ch. 120 (See Model Ch. 120 (See Model Ch. 124, 201, 216 (See Model CK-100) FEDERAL MFG. CO. 104 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 135 (Select-A-Call) 136 (Select-A-Call) 136 (Select-A-Call) 137 (Select-A-Call) 138 (Select-A-Call) 139 (Select-A-Call) 139 (Select-A-Call) 139 (Select-A-Call) 139 (Select-A-Call) 139 (Select-A-Call) 130 (Select-A-Call) 130 (Select-A-Call) 130 (Select-A-Call) 130 (Select-A-Call) 130 (Select-A-Call) 131 (Select-A-Call) 133 (Select-A-Call) 134 (Select-A-Call) 134 (Select-A-Call) 134 (Select-A-Call) 134 (Select-A-Call) 134 (Select-A-Call) 134 (Select-A-Call) 135 (Select-A-Call) 134 (Select-A-Call) 134 (Select-A-Call) 134 (Select-A-Call) 134 (Select-A-Call) 134 (Select-A-Call) 134 (Select-A-Call) 134 (Select-A-Call) 134 (Select-A-Call) 134 (Select-A-Call) 154 (Select-A-Ca	35-5 35-6 * 7 6 7 26 4 7 23 26 23 18-17 17-16 8 -17 17-16 39-4 15-10 14-4 31-13 28-11 41-7 49-8 36-7 35-7 15-11 11-19

1

EMERSON-GENERAL ELECTRIC

FIRESTONE-Cont. FIRESTONE-Cont. 4.A-25 (Code 29)-6-572) 13-6 4.A-26 (Code 307-6-9030.A) 33-5 4.A-27 4.A-30 4.A-31 (Code 4-A-30 4-A-31 (Code No. 177-5-4A31) 11-20 4-A-37 (Code 177-5-4A37) 13-7 4-A-40 4-A-41 (Code 291-7-576). 52-8 53—11 4-A-69 (Code No. 155-8-85) 4.A.89 (See Model 4.A.85)...118 4.B.1 (Code 7.6-PM15)...7.--1 4.B.2 (Code 7.6-PM14)...18.--18 4.B.21 (Code 7.6-PM14)...29.--8 No. 177-7-PM18).....29---8 4.B-21 86-5

 13.G-5 (Code 291.9-651)

 Tel. Rec.

 31.G-33 Tel. Rec.

 13.G-45 Tel. Rec.

 13.G-46, 13.G-45 Tel. Rec.

 *

 13.G-46 Tel. Rec.

 *

 13.G-46 Tel. Rec.

 *

 13.G-46 Tel. Rec.

 *

 13.G-46 Tel. Rec.

 *

 13.G-48 Tel. Rec.

 *

 13.G-49, 13.G-50 Tel. Rec.

 83-3 FLUSH WALL FORD FREED EISEMAN 46 11—8 54, 55, 56, 58 (Ch. 1620C) Tel. Rec. 113–1A GALVIN (See Motorola)
 GAMBLE-SKOGMO (See Coronado)

 GARCD (Also See MAJESTIC)

 4A.1, 4A.2
 29--9

 4B.1
 51--6

 5A.2
 5--28

 5A.3
 44--5

 5D, 5D-2
 12--12

 5D, 5D-2
 12--12

 5D, 3D-2
 12--12

 5D, 3D-2
 12--12

 5D, 3, 5D-3A
 22--16

 5D-4; 5D-5
 33--7

 5RC-1
 36--8

 6A2
 28--13

 6AU-1
 TBE Sendor

 6A2
 28--13

 10724, 10722, 10723, 10724, 10727, 10724, 10727, 10724, 10727, 10724, 10727, 10724, 10727, 10724, 10727, 10724, 10727, 10724, 1272, 12724, 12725, 12726A, 12727, 12724, 13727, 13726, 15727

 127220, 12721, 12721, 12722, 12723, 12724, 12723, 12724, 137220, 13727, 13726, 15727

 127220, 12721, 12722, 12722, 12723, 12724, 137224, 137224, 137220, 137224, 137224, 137225, 137226, 137224, 137225, 137226, 137224, 137225, 137226, 137227, 13724, 137225, 137226, 137226, 137227, 137226, 137227, 137226, 137227, 13723, 13726, 137226, 137227, 13723, 13726, 137226, 137227, 13723, 13726, 137226, 137227, 13723, 13726, 137226, 137227, 13723, 13726, 137226, 137227, 13723, 13726, 137226, 137227, 13723, 13726, 137226, 137227, 13723, 13726, 137226, 137227, 13723, 13726, 137226, 137227, 13723, 13726, 137226, 137227, 13723, 13726, 137226, 137227, 13723, 13726, 137226, 137227, 13726, 137226, 137227 GAMBLE-SKOGMO (See Coronado)

...... 29-10
 025
 29-10

 036
 48-8

 900, 1000 Series
 50-7

 1alevision Roceiver
 50-7

 10425, 1043G Tel. Rec.
 99A.5

 1000 Series Tel. Rec.
 93A-7

 1000 Series Tel. Rec.
 50

 1200 Series Tel. Rec.
 50

 12445, 1245G Tel. Rec.
 54

 15445, 1545 Tel. Rec.
 5456, 15475, 15486, 15456, 15475, 15486, 15457, 154817, 154817, 154817, 154817, 154817, 154817, 154817, 154817, 154817, 154817, 154817, 15481, 1649

 15445, 15471, 15481, 1649
 74.8

 1671 (98 Series) Tel. Rec., *
 *

 1671 (98 Series) Tel. Rec., *
 *

 1900 97 Series Tel. Rec., *
 *
 306 900, 1000 Series 48-8 GENERAL ELECTRIC

 1213, 12738, 1274, 12748

 Tel, Rec. (See

 Model 12(107)
 125

 1217 Tel, Rec.
 99A, 5

 14
 35-8

 14(12, 1473 Tel, Rec.
 123

 16(102, 14(103 Tel, Rec.
 123

 16(102, 14(103 Tel, Rec.
 123

 16(101, 16(111 Tel, Rec.
 123

 16(102, 14(102), 123
 16(102), 123

 16(115, 16(116, 16(117 Tel, Rec.
 123

 16(115, 16(116, 16(102), 123
 16(115, 16(116, 16(102), 123)

 16(115, 16(116, 16(102), 123)
 123

 17(101, 17(102 Tel, Rec.
 124

 19(101 Tel, Rec.
 123

 19(101 Tel, Rec.
 124

 19(01 Tel, Rec.
 123

 19(101 Tel, Rec.
 124

 100, 101
 6-13

 102, 102W
 41-8

 100, 101
 6-13

 102, 102W
 41-8

 103, 105 (See Models
 106

 102, 102W
 41-8

 113
 140, 115, 115W

 166
 8-14

 107, 107W (See Models
 107-7

 113, 144, 115, 115W
 152

 114, 114w, 115, 11 99A-6 32---8 7----16 36---9 -4 -13 6 8---14 41 39—5 97—7 81—8 30—10 75—9 60—13 56—11 56—12 89—7 20—11 57—7 180 186-4 200, 201, 202, 203, 205, 205M 210, 211, 212 218, 218 "H" 219, 220, 221 224 8-15 51-8 121-5 4-1 91-5
 219
 220, 221
 4-1

 226
 230 [See Kaiser-Frozer
 91-5

 230001
 35
 32

 250
 4-13
 2-9

 260
 15-13
 32

 260
 23-10
 303

 303
 18-19
 32-10

 321
 32-20
 32-10

 324
 32-20
 32-10

 324
 32-20
 32-10

 324
 32-20
 32-10

 324
 32-20
 32-10

 324
 32-20
 32-10

 324
 32-10
 32-10

 324
 32-20
 32-10

 324
 32-10
 32-10

 324
 32-10
 32-10

 324
 32
 30-11

 326, 327
 330 (See Model 324)
 64

 329, 330 (See Model 324)
 64

 329, 336, 357, 358
 37-6
 4-13 32-9 15-13 23-10 18-19 32-10 3-26 64-7

37

GENERAL ELECTRIC-MAGNAVOX

GENERAL ELECTRIC-MA	ONAVOA
GENERAL ELECTRIC-Cont.	GRANTLINE-Cont.
GENERAL ELECTRIC-Cont. 376, 377, 378 45-11 400, 401 118-8 404, 405 1216 408 1166 410 (See Model 404) 121 411 (See Model 400) 118 417 1615	605, 606
404, 405	651
408	641 12—12 651 11—9 5610 35—11 6547 11—10
411 (See Model 400)118	HALLICRAFTERS
300. 301 ISee Model 041 70	(Also See Echophone)
502 35-9 505, 506, 507, 508, 509	CA-2, CA-2A
(See Model 64)	S-38
(See Model 64)	S-38B
530 (See Model 64) 98	S-40A
600	S-40B
650	S-47
752, 753	S-51
800A, B, C, D Tel. Rec.	S-53
310, 311 114-5 320, 522, 522 98 600 109-6 601, 603, 604 109-6 601, 603, 604 109-6 600, 603, 604 113-3 752, 753 123-5 755 130-6 800, 8, C, D Tel. Rec. 130-6 801 Tel. Rec. 978 803 Tel. Rec. 974.4 805 Series Tel. Rec. 78-7 801 Tel. Receiver. 53-12 811 Tel. Receiver. 63-9 811 Tel. Receiver. 63-9 815 Tel. Rec. 97A-5 815 Tel. Rec. 97A-5 817 Tel. Rec. 78	$\begin{array}{cccccccccccccccccccccccccccccccccccc$
(Photofact Servicer) 78	5-59
802 Tel. Rec	\$78
805, 806, 807, 809	ST-74
Series Tel. Rec	SX-43
811 Tel. Receiver 63-9	SX-62
814 Iel. Rec 09-9 815 Tel. Rec 97A-5	SX-71 1116 T-54 Tel. Receiver. 481(T-54 (Late) Tel. Rec. 916 T-60 Tel. Receiver. 631(T-61, T-64, T-67 Tel. Rec. 657 T-68 (Tel. Rec.) 43
817 Tel. Rec.	T-54 (Late) Tel. Rec 91-6
817 Tel. Rec. (See Model 805)	T-61, T-64, T-67 Tel. Rec. 65-7
820 Tel. Rec	T-68 (Tel. Rec.) (See Model T-60)
(See Model 805) 78	T-69 Tel. Rec
830 Early, Tel. Rec	5R11, 5R12, 5R13, 5R14, 129-7
	See Model T-60) 63 T-69 Tel. Rec. 30—7 SR10 130—7 SR11, SR12, SR13, SR14. 129—7 400, 406, 409, 410, 411, 412 52—9
840 Tel. Rec. (See Model 830 Early) 81	412 52—9 505, 506 Tel. Rec.
901 Tel. Rec	(See Model T-54) 48
910 Tel. Rec. (See Model 901)	412
GENERAL IMPLEMENT	509, 510 Tel. Rec. (See Model T61) 65-7
9A5 37—7	511 Tel. Rec
GENERAL MOTORS	512C, 513 Tel. Rec 80-7 514 Tel. Rec. (See
CORP. (GMC) 2233029 93-6	(See Model 161)
GENERAL TELEVISION 1A5, 2A5, 3A5, 5A5 (Ch. 1-1)	(see model size) ov
1A5, 2A5, 3A5, 5A5	518, 519, 520 Tel. Rec 92-3 520E Tel. Rec.
4B5	(See Model 512C) 80
5B5G, 5B5Y 27—12 9A5 39—6	Model 518) 92
1A5, 2A5, 3A5, 5A5 (Ch, 1-1) 1AB5 27-11 5B5G, 5B5Y 27-2 9A5 9B4P 39-6 9B4P 36-10 15A5 (Ch, 1-1) (See	(See Model 512C)
14A4F	524 Tel. Rec. (See
Models 1A5, 2A5, 3A5, 5A5) 1	Model 512C)
17A5	Tel. Rec. (See Model
17A5 5—22 19A5 (Ch. 1-1) (See Models 1A5, 2A5, 3A5, 5A5) 1	Tel. Rec. (See Model 518)
3A5, 5A5) 1 21A4	680, 681 Tel. Rec
21A4 12-14 22A5C 13-19	
23A6 14—14 2484	715 Tel. Rec. (See Model 680)
21A4 12-14 22A5C 13-19 23A6 14-14 2486 37-8 2585 26-15 2485 29-11	730, 731 (Run 1) Tel. Rec.
2685	715 Tel, Rec. (See Model 680)113 730, 731 (Run 1) Tel. Rec. (See Model 680)113 740, 741 (Run 1) Tel. Rec. (See Model 680)113 745 Tel. Rec. 750, 751, Tel. Rec. (See Model 745)
GILFILLAN	740, 741 (Run 1) Tel. Rec. (See Model 680) 113-3
56A, 56B 1—27	745 Tel. Rec
56A, 56B 1-27 56BC1, 56BCR, 56C, 56D, 56E (See Model 56A) 1	750, 751, Tel. Rec. (See Model 745)
	760, 761 Tel. Rec. (See
66A, 66AM 8-16 66B "The Overland" 8-17 66D, 66DM (See Model	805, 806 Tel. Rec 125-1A
66D, 66DM (See Model 66A)	810A, 811 Tel. Rec 124-6 810C Tel. Rec.
66P, 66PM	(See Model 805) 125-14 815 Tel. Rec.
"The El Dorado" 9-15 68B-D 46-10 68F 46-11 68-48 61-10 86C, 86P, 86U (86 Series) 26-16 108,48 9-10	(San Model 810A) 174
68F	818, 820, 822 Tel. Rec. (See Model 810A)124
86C, 86P, 86U (86 Series) 26-16	
100.48	860, 861 Tel. Rec. (See Model 810A)124 870, 871 Tel. Rec.
GLOBE 58P1 18—20	870, 871 Tel. Rec. (See Model 810A)124
SBP1 18—20 6AP1 (See Model 6P1) 20 6D1 20—13 20—13 791 20—12 20	880 Tel. Rec.
6P1 20-12 6U1 (See Model 6D1) 20	(See Model 810A) 124 HAMILTON ELECTRONICS
7CP-1	H-15-5 16—1
51	H-15-S 16—1 H-50-2S 16—1
62C 19-19 85 49-9 454 41-9 456 40-7 457 39-7 500 21-18 517 21-17 551 16-16 552 27-13 553 28-15 559 50-8	HAMILTON RADIO CORP. (See Olympic)
454	HAMMARLUND
457	HQ-129-X
517 21 —17	SP-400-X
551 16—16 552	
553 28 —15	ATR-3-6, ATR-3-12 36-1
559 50—8	HEATH HBR-5 24-2
GODFRET	HBR-5 24-2 HOFFMAN
6AD	A-200 (Ch 103) 4-2
GON-SET	A-202 (Ch. 119) 11-1 A-300
3-30 Meter Converter 61—11 10-11 Meter Converter 37—9	1 000 (0) 1101
B. F. GOODRICH (See Mantola)	(See Model A-202) 11 A-401 (Ch. 102) 11—1
GOODELL	A-309 (Ch. 119) (See Model A-202) 11 A-401 (Ch. 102) 11-1 A-500 (Ch. 107) 4-3 A-501 (Ch. 108T) 3-3 A-700 (Ch. 110S) 12-1 B-400
ATB-3	A-700 (Ch. 1105) 12-1
NSA-20	B-1000
W. T. GRANT (See Grantline)	C-501
GRANTLINE 300 (Series B) 9	C-501 48—1 C-502 51—9 C-503 50—9 C-504 (Ch. 123) 47—1
500, 501 (Series A) 9-17	
300 (Series B) 916 500, 501 (Series A) 917 501-7 3510 504-7 2119 508-7 34-8	C509, C510
500, 501 (Series A) 9—17 501, 501 35—10 504, 502 34—8 504, 502 34—8 508, 502 34—8 510-A 24—19	C-506 (.C-507

RANTLINE-Cont.	
	HOFF
5 404 9_17	C-513 C-514
11-9	C-518
33, 500 12-15 31 11-9 3510 35-11 547 11-10	C710 C1006
ALLICRAFTERS	CT-80
	CT- 522,
Also See Echophone) A-2, CA-2A 30-12 A-4 38 38 398 201 21-7 40 2-19 40A 33-10 406 4166, S-41W 10-19 47 46-12 52 48-9	600, Tel 610 (612 (
38	610 (
40	156
40A	(Se 613 (f (Se 630, Tel 630, Tel 632
41G, S-41W	630,
47 46 12 51 40 8 52 48 9 53 39 8 55, S-56 55 9 58 57 8	630
52	Tel
53	
	1el 632, Tel 820, Tel 826, Tel 830
59 58—10 72 82—6	820,
78	Tel
125-8 K-42 44-6 K-43 45-13 K-62 61-12 K-71 111-6 S4 Tel. Receiver 48-10 S4 (Lote) Tel. Rec. 91-6 60 Tel. Recciver 63-10 61, T-64, T-67 Tel. Rec. 65-7 68 (Tel. Rec.) 63 69 Tel. Rec. 130-7 R10 130-7 R0, 5812, 5813, 5814, 129-7 10, 404, 407, 411, 412	Tel
45 13 (-62	830, Rec
x-71	832 ((Se
54 Tel. Receiver	(Se 836,
60 Tel, Receiver	836, Rec 840 (
68 (Tel. Rec.)	(Se 846 ((Se 847, Tel
(See Model T-60) 63	846 (
R10	847,
R11, 5R12, 5R13, 5R14129-7	860, Tei
00, 406, 409, 410, 411, 412 52—9	Tei 847
(See Model T-54) 48	866, Tel
05, 506 (Late) (See Model T-54 (ate) 91	870,
09, 510 Tel. Rec.	870, Tel
(See Model 161) 65—/ 11 Tel. Rec	Tel 876, Tel 890
12C, 513 Tel. Rec 80-7	Tel
Model T-54 Late) 91	912.
15 Tel. Rec. (See Model 512C)	Rec 914,
18, 519, 520 Tel. Rec 92-3	Per C
(See Model 512C) 80	917, Rec
21 Tel. Rec. (See	
21E Tel. Rec. (See	(Se 946, Tel
24 Tel. Rec. (See	
Model 512C)	950, 951 (CI
Tel. Rec. (See Model	(CI 960,
05, 606 Tel. Rec 107—5	1 101
80, 681 Tel. Rec	Mo Chass
(See Model 680)113	(Se
15 Tel. Rec. (See Model 680)	Chass (Se
30, 731 (Run 1) Tel. Rec.	Chass (Se
32, 733 Tel. Rec	Chass
40, 741 (Run 1) Tel. Rec. (See Model 680)	(Se Chass
45 Tel. Rec 105-4	(Se Chass
50, 751, 1el. Rec. (See	(Se
Model 745) 105	
Model 745)	Chase
Model 745)	Chase
Model 745)	Chase
Model 745)	Chass (Se Chass (Se Ch. 1
Model 745)	Chass (Se Chass (Se Ch. 1
18 820 822 Tel Rec	Chass (Se Chass (Se Ch. 1
Model 745)	Chass (Se Chass (Se Ch. 1
(See Model 810A) 124 32, 833 Tel. Rec 121-1A 60, 861 Tel. Rec. (See Model 810A) 124	Chass (Se Chass (Se Ch. 1
(See Model 810A) 124 32, 833 Tel. Rec 121-1A 60, 861 Tel. Rec. (See Model 810A) 124	Chass (Se Chass (Se Ch. 1
(See Model 810A) 124 32, 833 Tel. Rec 121–1A 60, 861 Tel. Rec. (See Model 810A) 124 70, 871 Tel. Rec. (See Model 810A) 124 80 Tel. Rec.	Chass (Se Chass (Se Ch. 1
(See Model 810A) 124 32, 833 Tel. Rec 121–1A 60, 861 Tel. Rec. (See Model 810A) 124 70, 871 Tel. Rec. (See Model 810A) 124 80 Tel. Rec. (See Model 810A) 124	Chass (Se Chass) (Se Ch. 1 91 Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch.
(See Model 810A) 124 32, 833 Tel. Rec 121-1A 60, 861 Tel. Rec 124 (See Model 810A) 124 (See Model 810A) 124 80 Tel. Rec 124 (See Model 810A) 124 (See Model 810A) 124 (AdMILTON ELECTRONICS	Chass (Se Chass) (Se Ch. 1 91 Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch.
[See Model 810A]	Chass (Se Chass (Se Ch. 1 Ch. 1 Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch.
(See Model 810A)	Chass (Se Chass) (Se Ch. 1 Oh. Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch. C
(See Model 810A)	Chass (Se Chass) (Se Ch. 1 Oh. Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch. C
(See Model 810A)	Chass (Se Chass) (Se Ch. 1 Oh. Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch. C
(See Model 810A)	Chass (See Choise (See Ch. 1 Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch.
(See Model 810A)	Chass (Se Chass (Se Ch. 1) Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch.
(See Model 810A)	Choss (683) (61) Choss
(See Model 810A)	Choss (56) Choss (5, 1) Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch.
(See Model 810A)	Choss Choss
(See Model 810A)	Choss (58 Choss Chos
(See Model 810A)	Choise (Se Choise) (Se Choise) (Se Choise) (Se Choise) (Se Choise) (Choise)
(See Model 810A)	Choise (Se Choise) (Se Choise) (Se Choise) (Se Choise) Choise Cho
(See Model 810A)	Choise (Se Choise) (Se Choise) (Se Choise) (Se Choise) (Se Choise) (Choise)
(See Model 810A)	Choise (Se Choise) (Se Choise) (Se Choise) (Se Choise) (C
(See Model 810A)	Choss (Se Choss) (Se Choss) (Se Choss) Choss Cho
(See Model 810A)	Chossis (Se Chossis) (Se Chossis) (Se Chossis) Chossis
(See Model 810A)	Choss (Se Choss) (Se Choss) (Se Choss) Choss Cho
(See Model 810A)	Chossis (Se Chossis) (Se Chossis) (Se Chossis) Chossis
(See Model 810A)	Choss (Se Choss) (Se Choss) (Se Choss) Choss Cho
(See Model 810A)	Chosts (Se Chosts) (Se Chosts) Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch. Ch.

HOFFMAN-Cont.
C-513 (See Model C-503) 50
C.514 ISan Model C-5041 47
C1-13 (100 model c1-00 model c
C-518 61—13 C710 (Ch. 133) * C10046, C1007 54—9 C1-800, CT-801, CT-900, CT-901 C12, 524 (Ch. 138) * 600, 601 (Ch. 136) * 610 (Ch. 140) * 610 (Ch. 140) *
CT-800, CT-801, CT-900, CT-901 (Tel. Rec.) 63-11
522, 524 (Ch. 138) *
600, 601 (Ch. 154, 155) Tel. Rec. 95A-8
Tell. Rec. 95A-8 610 (Ch. 140) Tell. Rec. 97A-6 612 (Ch. 142) Tell. Rec. (See Model 610)
612 (Ch. 142) Tel. Rec. (See Model 610)
613 (Ch. 149) Tel. Rec.
(See Model 610) 97A
630, 631 (Ch. 159) Tel. Rec. *
Tel. Rec. 530, 631 (Ch. 170) Tel. Rec. 1170-1A 632, 633 (Ch. 160) 117-1A 781, Rec. 117-1A 632, 633 (Ch. 171) 117-1A 781, Rec. 117-1A 820, 821, 822 (Ch. 146) 117-1A
632, 633 (Ch. 160)
632 633 (Cb 171)
Tel. Rec
820, 821, 822 (Ch. 146)
Tel. Rec
Tel. Rec
Rec
Tel. Rec. 95A-8 830, 831 (Ch. 151) Tel. Rec. 97A-6 832 (Ch. 151) Tel. Rec. 1000 (See Model 830) 97A 836, 837 (Ch. 153) Tel. 03A 03A
836, 837 (Ch. 153) Tel.
Rec
(See Model 836) 93A
836, 837 (Ch. 153) 101. Rec. 93A-8 840 (Ch. 153) 101. Rec. (See Model 836). 93A 846 (Ch. 151) 101. Rec. (See Model 836). 93A 847, Ch. 151) 101. Rec. (See Model 836). 97A 847, 848, 849 (Ch. 156) 97A 7
847, 848, 849 (Ch. 156)
Tel. Rec
Tel. Rec. (See Model
(366 model 350)
Tel. Rec
8/0, 8/1, 8/2 (Ch. 1/0) Tel. Rec. 117–1A
876, 877, 878 (Ch. 171) Tel. Rec
890, 891, 892 (Ch. 175)
8900, 891, 892 (Ch. 175) Tel. Rec
Rec. (See Model 826) 95A
914, 915 (Ch. 150) Tel. Rec. (See Model 610) 97A
917, 918 (Ch. 152) Tel.
Rec. (See Model 830) 97A 920 (Ch. 152) Tel. Rec.
(See Model 830) 97A
740, 747, 748 (Ch. 164) Tel. Rec. (See Model
847)
950A, 951A, 952A
(Ch. 174) Tel. Rec 127-6
lei, Kec. 117-1A 912, 913 (Ch. 147) Teil. Rec. (See Model 826) 95A 914, 915 (Ch. 150) Teil. Rec. (See Model 830) 97A 920 (Ch. 152) Teil. Rec. (See Model 830) 97A 920 (Ch. 152) Teil. Rec. (See Model 830) 97A 946, 947, 948 (Ch. 164) Teil. Rec. (See Model 847) 950, 951, 952 (Ch. 172), 950A, 951A, 952A (Ch. 174) Teil. Rec 127—6 960, 961, 962, (Ch. 176) Teil. Rec. (See Model 950)
Chassis 102 (See Model A401) 11 Chassis 103
Chassis 103 (See Model A200)
(See Model A200) 4
[See Model A200] 4 Chassis 107 (See Model A500] 4
[See Model A200] 4 Chassis 107 (See Model A500) 4 Chassis 108ST (See Model A501) 3
See Model A2001 4 Chassis 107 [See Model A500] 4 Chassis 108ST [See Model A501] 3 Chassis 110S 3
[See Model A200]
[See Model A200]
[See Model A200] 4 Chassis 107 (See Model A500) 4 Chassis 1085T (See Model A501)
[bee Model A200]4 Chassis 107 (See Model A500)4 Chassis 1085T (See Model A501)3 Chassis 1005 [See Model A700]12 Chassis 1105 [See Model A700]20 Chassis 114 (See Model A202)11 (See Model A202)11
[bee Model A200]4 Chassis 107 (See Model A500)4 Chassis 1085T (See Model A501)3 Chassis 1005 [See Model A700]12 Chassis 1105 [See Model A700]20 Chassis 114 (See Model A202)11 (See Model A202)11
[bee Model A200]4 Chassis 107 [See Model A500]4 Chassis 107 [See Model A501]3 Chassis 105 [See Model A500]12 Chassis 1105 [See Model A500]12 Chassis 114 [See Model A500]12 Chassis 114 [See Model A202]11 [Chassis 114 A202]11 [Chass Model A202]11 [Chass Model A202]11 [Chass Model A202]11 [Chass Model A202]11 [See Model C504]47 [Chass Model A202]11
[See Model A200]
[See Model A200]
Lee Model A200} 4 Chassis 102 3 Chassis 102 12 Chassis 1105 12 Chassis 114 81000) 20 Chassis 112 CS64 47 Ch. 138 (See Model 500) 97A Ch. 142 (See Model 610) 97A Ch. 142 (See Model 610) 97A Ch. 142 (See Model 820) 5A Ch. 144 (See Model 820) 5A Ch. 145 (See Model 820) 5A Ch. 146 (See Model 820) 5A Ch. 147 (See Model 820) 5A Ch. 147 (See Model 820) 5A
Lee Model A200;
Lese Model A200)
Lese Model A200;
Lee Model A200;
Lese Model A200)
Lese Model A200;

HOFFMAN-Cont. C-513 (See Model C-503) 50 C-514 (See Model C-504) 47	INDUSTRIAL ELECTRONIC CORP. (See Simplon)
C-514 (See Model C-504) 47 C-518	INDUSTRIAL TELEVISION
C710 (Ch. 133)	(Also See Century) IT-40R, IT-42R (Ch. IT-26R,
C.518 (1006 1100 1100 1100 1100 1100 1100 110	IT-35R, IT-39R, IT-46R) Tel. Rec
522, 524 (Ch. 138)*	IT-48R Tel. Rec*
Tel. Rec	JACKSON 10C, 10T Tel. Rec
610 (Ch. 140) Tel. Rec 97A-6 612 (Ch. 142) Tel. Rec.	10C, 10T Tel. Rec
(See Model 610)	(See Model 10C)132 14C, 14T Tel. Rec.
(See Model 610) 97A	(See Model 10C)
630, 631 (Ch. 159) Tel. Rec	(See Model 10C)132
630, 631 (Ch. 170) Tel. Rec	29C Tel. Rec. * 150 1308 153 (See Model 150) 130 312 Tel. Rec. 130
632, 633 (Ch. 160)	312 Tel. Rec.
632, 633 (Ch. 171) Tel. Rec	(See Model 10C)132 316 Tel. Rec.
820, 821, 822 (Ch. 146)	(See Model 10C)132
Tel. Rec. * 826, 827, 828 (Ch. 143)	350
Tel. Rec	(See Model 10C)132 416 Tel. Rec.
Rec	(See Model 10C)132 5000, 5050 Tel. Rec88—5 5200, 5250 (See Model 5000) Tel. Rec
(See Model 830) 97A 836, 837 (Ch. 153) Tel.	5200, 5250 (See Model
836, 837 (Ch. 153) Tel. Rec	
Rec. 93A-8 840 (Ch. 153) Tel. Rec. (See Model 836)	5000) Tel. Rec 88
846 (Ch. 151) Tel. Rec.	JEFFER5ON-TRAVIS MR-2B
(See Model 830) 97A 847, 848, 849 (Ch. 156)	MR-2B
Tel. Rec. 97A-7 860, 861, 862 (Ch. 157)	JEWEL
Tel Per (See Model	300 23—11 304 35—12
847)	500A, B, C; 501A, B, C; 502A, B, C: 503A, B, C;
870, 871, 872 (Ch. 170)	304 35—12 500A, B, C; 501A, B, C; 502A, B, C; 503A, B, C; 504A, B, C; 503A, B, C; 504A, B, C; 505A, B, C 15—14 505**Pin-Up************************************
Tel. Rec	801 (Trixie)
Tel. Rec	
Tel. Rec	915 (See Model 910) 99
912, 913 (Ch. 147) Tel. Rec. (See Model 826) 95A 914, 915 (Ch. 150) Tel.	935 (See Model 920) 55 949 105-5
Kec. (See Model OIU) Y/A	955 98—5
917 918 (Ch 152) Tel	960
920 (Ch. 152) Tel. Rec. (See Model 830) 97A	985 (See Model 910)
946, 947, 948 (Ch. 164)	
Tel. Kec. (See Model	KAISER-FRAZER 100170
950, 951, 952 (Ch. 172), 950A, 951A, 952A	200001
(Ch. 174) Tel. Rec 127-6 960, 961, 962, (Ch. 176)	KAPPLER
Tel. Rec. (See Model 950)	102T
Chassis 102	KARADIO
(See Model A401) 11 Chassis 103	80-C
(See Model A200) 4 Chassis 107	1276
(See Model A500) 4	231, 232, 233, 234, 235,
Chassis 1085T (See Model A501) 3 Chassis 1105	231, 232, 233, 234, 235, 236, 237, 238, 239, 240, 241 Tel. Rec *
[See Model A700] 12	242 Tel. Rec
Chassis 114 (See Model B1000) 20	733 Tel. Rec
Chassis 119 (See Model A202) 11	240, 241 lei. Kec
Chassis 123 (See Model C504) 47	KAY MUSICAL
(See Model C504) 47 Ch. 138 (See Models	INSTRUMENT CO.
Ch. 138 (See Models 912, 913) Ch. 140 (See Model A10) 97A Ch. 142 (See Model A21) 97A Ch. 143 (See Model 826) 95A Ch. 144 (See Model 820) Ch. 147 (See Model 820) Ch. 147 (See Model 820) Ch. 147 (See Model 820) 95A Ch. 150 (See Model 820) 97A Ch. 151 (See Model 830) 97A Ch. 153 (See Model 830) 95A Ch. 154 (See Model 830) 95A Ch. 155 (See Model 840) 95A Ch. 156 (See Model 840) 95A Ch. 157 (See Model 840) 97A Ch. 157 (See Model 840) 97A	77 42—13
Ch. 142 (See Model 612) 97A Ch. 143 (See Model 826) 95A	5 Tube Radio 6-14
Ch. 146 (See Model 820) *	KNIGHT
Ch. 149 (See Model 613) 97A	4D-450
Ch. 150 (See Model 914) 97A Ch. 151 (See Model 830) 97A	4D-430 409 4G420 886 5A150 5A154 1217 5A190 1415 18-160 5B-175, 5B-176 2015 5B-175, 5B-176 2015 5C-290 52-51 5D-455 349 5D-455 349 5E-250, 5E-251 349
Ch. 152 (See Model 917) 97A Ch. 153 (See Model 836) 93A	5B-160
Ch. 154 (See Model 600) 95A Ch. 155 (See Model 600) 95A	5B-185
Ch. 156 (See Model 847). 97A Ch. 157 (See Model 847). 97A	5C-290
Ch. 164 (See Model 946). 97A	5D-455
(See Model 630)117-1A	(Similar to Chassis) 36-25 SE-457 (Similar to Chassis) 53-23 SF-525, SF-526 53-13 SF-565 55-12 SG-563 (Similar to Chassis) 97-1
(See Model 630)117-1A Ch. 172 (See Model 950).127 Ch. 173 (See Model 630).117-1A Ch. 174 (See Model 950).127 Ch. 175 (See Model 630).117-1A Ch. 176 (See Model 950).127	5F-525, 5F-526 53-13
Ch. 174 (See Model 950). 127 Ch. 175 (See Model 630) 117-14	5G-563 (Similar to Chassis) 97-1
	5H607, 5H-608 (Similar to Chassis)
	(Similar to Chassis) 97—15 5H-678, 5H-679
HOWARD 472AC, 472AF, 31—14 474	
475TV Tel. Rec.	6A-122 9—18
Photofact Servicer 84 481B, 481C, 481M 67-11	6A-127
481B, 481C, 481M	68-122 (See Model 6A-122) 9
901A-M, 901A-W (See	Similar to Chossis). 109-/ SH-700 1237 6A-122 918 6A-127 918 6A-122 919 6b-127 919
901A Series 1-8	6C-225, 6D-225, 6D-226 30-14
901AP	6D-235
906, 906C	6G-400 (See Model 449). 83 6H580 126 7
901A.S.r(is) 1 901A.Series 1 901A.Series 1 901A.Series 1 901A.Period 10 902 17 909M 25 920 5	78-220
HUDSON	8B-210
DB47 (Fact. No. 6MH089) 25-16 DB48 (Fact. No. 6MH889) 39-9	bb-121 g bb-127 See Model g bb-127 See Model g bc-225 bb-226 30-14 bb-320 See Model 49 bb-323 See Model 49 bb-326 See Model 40 bb-326 See Model 40 bb-326 See Model 40 bb-326 See Model 42 bb-326 See Model 42 bb-326 See Model 42 bb-326 See Model 42 bb-326 See Model 42
HUDSON ELECTRONICS 332-H	9V-101 Tel. Rec
332.H 192 /	

KNIGHT-Cont. 14F490, 14F495, 14F496, 63-12 15H609 (See Model 5118) 125 19F492, 19F497, 19F498, 58-11 93-017
14F490, 14F495, 14F496, 63-12 15H609 (See Model 5118) 125
19F492, 19F497, 19F498. 58-11
93-024 32—13 93-103 31—16 93-146 36—15
93-140
93-191
93.330 99_0
93-350 / 6 -13
449
LAFAYETTE FA15W FA15Y 15_15
FA15W, FA15Y 15—15 J62, J62C
MC10B, MC10Y 14-16
MC11
MCI3
MC16
(Similar to Chassis) 98-5
IN437 (Similar to Chassis) 121-2 IN549 (Similar to Chassis) 38-5
IN551 (Similar to Chassis) 38-6
IN554, IN555 (Similar to Chassis) 55 —10
IN556 IN557
(Similar to Chassis)109—7 IN559 (Similar to Chassis) 90—7
IN560 (Similar to Chassis) 1097
(Similar to Chassis) 97-8
IN819 (Similar to Chassis) 69-7
LAMCO
1000 16 —20
LEARADIO Chassis R-971 51—11
RM-402C (Learavian) 42-15
561, 562, 563 I—26 565, 565BL, 566, 567, 568 9—20
1281-PC (Ch. 78) 49-11
6610PC, 6611PC, 6612PC . 9-21 6614, 6615, 6616, 6619 . 3-18
6617PC 16-22
LEE TONE
AP-100 16—23
LEW YT
605 615A 711 42—16
LEXINGTON 6545 13—20
LIBERTY
A6K, A6P, 6K 20-18
AON, AOP, OK 20-18
A6K, A6P, 6K 20—18 507A 20—19
507A
507A
S07A 20—19 LINCOLN (Auto Radio) 7ML080 (5EH-18805-A), 7ML081 (5EH-18805-B), 66—11
S07A 20—19 LINCOLN (Auto Radio) 7ML080 (5EH-18805-A), 7ML081 (5EH-18805-B), 66—11
S07A 20—19 LINCOLN (Auto Radio) 7ML080 (5EH-18805-A), 7ML081 (5EH-18805-B), 66—11
S07A 20—19 LINCOLN (Auto Radio) 7ML080 (5EH-18805-A), 7ML081 (5EH-18805-B), 66—11
507A 20—19 LINCOLN (Auto Radio) 7ML080 (5EH-18805-A), 7ML080 (5EH-18805-A), 8ML882 (8L-18805-A), 8ML8822 (8H-18805-A), 8ML9852 (8L-18805-A), 8ML9852 (8L-18805-B), 8ML9852 (8H-18805-B), 8ML9852 (8H-18805-B), 8ML9852 (8H-18805-B), 8ML9852 (8H-18805), 83—4
507A 20—19 LINCOLN (Auto Radio) 7ML080 (5EH-18805-A), 7ML081 (5EH-18805-A), 8ML882 (8H-18805-A), 8ML8822 (8H-18805-A), 66—11 8ML8825 (8H-18805-A), 8M-7 8ML9855 (8L-18805-A), 8ML9852 (8H-18805-A), 8ML9852 (8H-18805-A), 8ML9852 (8H-18805-A), 8ML9852 (8H-18805-A), 8M-47 INK00LN 83-4
507A 20—19 LINCOLN (Auto Radio) 7ML080 (5EH-18805-A), 7ML081 (5EH-18805-A), 8ML882 (8L-18805-A), 8ML8822 (8H-18805-A), 8ML8822 (8H-18805-A), 8ML8825 (8H-18805-A), 8ML983 (8L-18805-A), 8ML985 (8L-18805-A), 8ML985 (8L-18805-A), 8ML9852 (8H-18805-A), 8M-9852 (8H-18805-A), 8ML9852 (8H-18805-A), 8M-9852 (8H-18805-A), 8ML9852 (8H-18805) 8 3-4 LINCOLN 2—10 LINCOLN 2—10
507A 20—19 LINCOLN (Auto Radio) 7ML080 (5EH-18805-A), 7ML081 (5EH-18805-A), 66—11 SML882 (8L-18805-A), 8ML8822 (8H-18805-A), 8ML8825 (8H-18805-A), 8ML9855 (8L-18805-A), 8ML9852 (8H-18805-A), 8ML9852 (8H-18805-A), 8ML9852 (8H-18805-A), 8M-9852 (8H-18805-A), 8ML9852 (8H-18805-A), 83—4 LINCOLN 513L-B 2—10 LINCOLN (ALLIED RADIO CORP.)
507A 20—19 LINCOLN (Auto Radio) 7ML080 (5EH-18805-A), 7ML080 (5EH-18805-A), 8ML882 (5H-18805-A), 8ML8822 (8H-18805-A), 8ML8822 (8H-18805-A), 8ML982 (8H-18805-A), 8ML982 (8H-18805-A), 8ML982 (8H-18805-A), 8ML985 (8H-18805-A), 8ML985 (8H-18805-A), 8ML9852 (8H-18805-A), 8ML9852 (8H-18805-A), 8M-9452 (8H-18805-A), 9ML9852 (8H-18805-A),
507A 20—19 LINCOLN (Auto Radio) 7ML080 (5EH-18805-A), 7ML080 (5EH-18805-A), 8ML882 (8L-18805-A), 8ML8822 (8H-18805-A), 8ML8822 (8H-18805-A), 8ML982 (8L-18805-A), 8ML982 (8L-18805-A), 8ML982 (8L-18805-A), 8ML9825 (8L-18805-A), 8ML982 (8L-18805-A), 8ML9825 (8L-18805-A), 8ML9825 (8L-18805-A), 8M-9452 (8L-18805-A), 9ML9825 (8L-18805-A),
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML081 (SEH-18805-A), 8ML882 (SI-18805-A), 8ML882 (SI-18805-A), 8ML882 (SI-18805-A), 8ML882 (SI-18805-A), 8ML985 (SI-18805-A), 8ML985 (SI-18805-A), 8ML985 (SI-18805-A), 8ML985 (SI-18805-A), 8M-985 (SI-18805-A), 8ML985 (SI-18805-A), 8M-985 (SI-18805-A), 8ML9852 (SH-18805-A), SI-4 LINCOLN 2—10 (ALLIED RADIO CORP.) 5A-110 5A-110 5—34 LINPAN (See Supreme) LIPAN (See Supreme)
507A 20—19 LINCOLN (Auto Radio) 7ML080 (5EH-18805-A), 7ML081 (5EH-18805-A), 8ML882 (8L-18805-A), 8ML882 (8L-18805-A), 8ML882 (8L-18805-A), 8ML985 (8L-18805-A), 8ML985 (8L-18805-A), 8ML985 (8L-18805-A), 8ML9852 (8H-18805-A), 8ML9852 (8H-18805-A), 8M-7 8ML9852 (8H-18805), 8M-4 LINCOLN 5M-34 LINCOLN (ALLIED RADIO CORP.) 5-34 LIPAN (See Supreme) LULLABY (See Mitchell)
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML080 (SEH-18805-A), 8ML882 (SH-18805-A), 8ML882 (SH-18805-A), 8ML882 (SH-18805-A), 8ML882 (SH-18805-A), 8ML985 (SH-18805-A), 8ML982 (SH-18805-A), 8ML985 (SH-18805-A), 8ML985 (SH-18805-A), 8ML985 (SH-18805-A), 8ML985 (SH-18805-A), 8ML985 (SH-18805-A), 8ML985 (SH-18805-A), 8M-4 LINCOLN SH3-4 LINCOLN 2—10 LINCOLN 5A-110 SA-110 5—34 LINDEX CORP. (See Swank) LIPAN (See Supreme) LULABY (See Mitchell) LYMAN
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML080 (SEH-18805-A), 8ML882 (BL-18805-A), 8ML882 (BL-18805-A), 8ML882 (BL-18805-A), 8ML985 (BL-18805-A), 8ML985 (BL-18805-A), 8ML985 (BL-18805-A), 8ML985 (BL-18805-A), 8ML985 (BL-18805-A), 8M-985 (BL-18805-A), 8ML985 (BL-18805-A), 8M-985 (BL-18805-A), 8ML9852 (BH-18805) 83-4 LINCOLN 2—10 LINCOLN 5-34 LINDEX CORP. (See Swank) LIPAN (See Supreme) LULLABY (See Mitchell) LYMAN CM10, CM20 44-8
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML080 (SEH-18805-A), 66—11 8ML882 (SL-18805-A), 8ML882 (SL-18805-A), 8ML882 (SL-18805-A), 8ML982 (SL-18805-A), 8ML982 (SL-18805-A), 8M.9852 (SL-18805-A), 8ML982 (SL-18805-A), 8M.9852 (SL-18805-A), 8ML982 (SL-18805-A), 8M.9852 (SL-18805-A), 8ML982 (SL-18805-A), 83—4 LINCOLN 5131-B 2—10 LINCOLN 5A-110 5—34 LINDEX CORP. (See Swank) LIPAN (See Supreme) LULLABY (See Mitchell) LTMAN CM10, CM20 44—8 LYRIC (Also See Rauland) 3407, 54617, 54617, 54617, 54717
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML080 (SEH-18805-A), 8ML882 (SEI-18805-A), 8ML882 (SEI-18805-A), 8ML882 (SEI-18805-A), 8ML985 (SEI-18805-A), 8ML985 (SEI-18805-A), 8ML985 (SEI-18805-A), 8ML985 (SEI-18805-A), 8ML985 (SEI-18805-A), 8M-985 (SEI-18805-A), 8ML9852 (SEI-18805-A), 8M-9852 (SEI-18805-A), 8ML9852 (SEI-18805-A), 8M-4 LINCOLN SI3L-B 2—10 LINCOLN SA-110 5—34 LINDEX CORP. (See Swank) LIPAN (See Supreme) LULLAEY (See Mitchell) LYMAN CM10, CM20 44—8 LYRIC (Also See Rauland) 5461, 5461Y, 5461Y 5461, 5461Y, 5461Y 7—17
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML080 (SEH-18805-A), 8ML882 (SL-18805-A), 8ML882 (SL-18805-A), 8ML882 (SL-18805-A), 8ML882 (SL-18805-A), 8ML985 (SL-18805-A), 8ML985 (SL-18805-A), 8ML985 (SL-18805-A), 8ML985 (SL-18805-A), 8M-985 (SL-18805-A), 8ML985 (SL-18805-A), 8M-985 (SL-18805-A), 8ML985 (SL-18805-A), 8M-4 LINCOLN 2—10 LINCOLN (ALLED RADIO CORP.) 5A-110 5—34 LINDEX CORP. (See Swank) LIPAN (See Supreme) LULLABY (See Mitchell) LYMAN CM10, CM20 44—8 LYRIC (Also See Rauland) 5467, 5461W 504 (Battla Reprised) 2—18
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML080 (SEH-18805-A), 8ML882 (SL-18805-A), 8ML882 (SL-18805-A), 8ML882 (SL-18805-A), 8ML882 (SL-18805-A), 8ML985 (SL-18805-A), 8ML985 (SL-18805-A), 8ML985 (SL-18805-A), 8ML985 (SL-18805-A), 8M-985 (SL-18805-A), 8ML985 (SL-18805-A), 8M-985 (SL-18805-A), 8ML985 (SL-18805-A), 8M-4 LINCOLN 2—10 LINCOLN (ALLED RADIO CORP.) 5A-110 5—34 LINDEX CORP. (See Swank) LIPAN (See Supreme) LULLABY (See Mitchell) LYMAN CM10, CM20 44—8 LYRIC (Also See Rauland) 5467, 5461W 504 (Battla Reprised) 2—18
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML080 (SEH-18805-A), 8ML882 (SL-18805-A), 8ML882 (SL-18805-A), 8ML882 (SL-18805-A), 8ML985 (SL-18805-A), 8ML985 (SL-18805-A), 8ML985 (SL-18805-A), 8M-985 (SL-18805-A), 8ML985 (SL-18805-A), 8M-985 (SL-18805-A), 8ML985 (SL-18805-A), 8M-985 (SL-18805-A), 8ML985 (SL-18805-A), 8M-4 LINCOLN SL (ALLIED RADIO CORP.) 5A-110 5A-110 534 LINCOLN (ALLIED RADIO CORP.) 5A-34 LINEX CORP. (See Swank) LIPAN (See Supreme) LULLABY (See Mitchell) LYMAN CM10, CM20 44—8 LYRIC (Also See Rauland) 5467, 5461W, 7-17 MAGIC TONE 5004 (BoHte Receiver) 22—18 508 (Keg Radio) 38-9 510
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML080 (SEH-18805-A), 8ML882 (SEI-18805-A), 8ML882 (SEI-18805-A), 8ML882 (SEI-18805-A), 8ML882 (SEI-18805-A), 8ML985 (SEI-18805-A), 8ML985 (SEI-18805-A), 8ML985 (SEI-18805-A), 8ML985 (SEI-18805-A), 8M-985 (SEI-18805-A), 8ML985 (SEI-18805-A), 8M-9 100 CON 2—10 LINCOLN 2—10 LINCOLN 5—34 LINCOLN 5—34 LINEX CORP. (See Swank) LIPAN (See Supreme) LULLABY (See Mitchell) LYMAN CM10, CM20 44—8 LYRIC (Also See Rauland) 54617, 54617% 506 (Keg Radio) 38—9 508 (Keg Radio) 38—9 500 (See Model 508) 38
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML080 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEE (SEE (SEE (SEE (SEE (SEE (SEE (SE
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML080 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8M.982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8M.982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8M-4 LINCOLN S131-B 2—10 LINCOLN 5.4110 5—34 LINDEX CORP. (See Swank) LIPAN (See Supreme) LULLABY (See Mitchell) LTMAN CM10, CM20 44—8 LYRIC (Also See Rauland) 5447, 5447W, 7—17 MAGIC TONE 5—40 5004 (Bottle Receiver) 22—18 508 (Keg Radio) 38 —9 500 (See Model 508) 38 MAGNAVOX Chossis AMP-101A,
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML080 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEE (SEE (SEE (SEE (SEE (SEE (SEE (SE
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML080 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEE (SEE (SEE (SEE (SEE (SEE (SEE (SE
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML080 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEE (SEE (SEE (SEE (SEE (SEE (SEE (SE
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML080 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEE (SEE (SEE (SEE (SEE (SEE (SEE (SE
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML080 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEE (SEE (SEE (SEE (SEE (SEE (SEE (SE
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML080 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML9852 (SEH-18805-A), 8ML9852 (SEH-18805-A), 8M-952 (SEH-18805-A), 8ML9852 (SEH-18805-A), 8M-952 (SEH-18805-A), 18ML982 (SEE (SEH-18805-A), 8M-9510, 11NCOLN 5-34 LINCOLN (ALLIED RADIO CORP.) 5-34 JINDEX CORP. (See Swank) LIPAN (See Supreme) LULLABY (See Mitchell) LIMAN CM10, CM20 44—8 LYRIC (Also See Rauland) 5440, 5447V, 5447W, 7-17 MAGIC TONE 5-40 504 (BotHe Receiver) 22—18 508 (Keg Radio) 38-9 510 5-40 504 (BotHe Receiver) 22-18 508 (Keg Radio) 38-9 510 5-40 504 (BotHe Receiver) 22-18 508 (
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML080 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML985 (SEH-18805-A), 8ML9852 (SEH-18805-A), 8ML9852 (SEH-18805-A), 8M-952 (SEH-18805-A), 8ML9852 (SEH-18805-A), 8M-952 (SEH-18805-A), 8ML9852 (SEH-18805-A), 8M-4 LINCOLN S131-B 2—10 LINCOLN S131-B 2—10 LINCOLN S131-B 2—10 LINCOLN S131-B 2—10 LINCOLN SA-110 5—34 LINDEX CORP. (See Swank) LIPAN (See Supreme) LULLBY (See Mitchell) LTRIC (Also See Rauland) S407, S4347, S447W, 7-17 MAGIC TONE S003 (Set Model 508) 38 MAGNAVOX Chostis AMP-100, 52—10 S004 (Bottle Receiver) 22—18 S08 (Keg Radio) 38 MAGNAVOX Chostis AMP-104, 4
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML080 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML985 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML9852 (SEH-18805-A), 8ML9852 (SEH-18805-A), 8M.9852 (SEH-18805-A), 18ML9852 (SEE (SH-18805-A), SB.3-4 LINCOLN SI31-8 2—10 LINCOLN SAL10 SEM (ALLIED RADIO CORP.) SAL10 SAL10 SAL10 CORP. (SEE Murchell) LULABY (SEE Murchell) LINDEX CORP. (See Murchell) SAL10 SAL10 Cholo (SEE Model SO8) SB SAL10 SO0 (Sol (SCI TONE S-40 SOL10
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML080 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML985 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML9852 (SEH-18805-A), 8ML9852 (SEH-18805-A), 8M.9852 (SEH-18805-A), 18ML9852 (SEE (SH-18805-A), SB.3-4 LINCOLN SI31-8 2—10 LINCOLN SAL10 SEM (ALLIED RADIO CORP.) SAL10 SAL10 SAL10 CORP. (SEE Murchell) LULABY (SEE Murchell) LINDEX CORP. (See Murchell) SAL10 SAL10 Cholo (SEE Model SO8) SB SAL10 SO0 (Sol (SCI TONE S-40 SOL10
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML081 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8M-952 (SEH-18805-A), 8ML9852 (SEH-18805-A), 8M-952 (SEH-18805-A), 8ML9852 (SEH-18805-A), 8M-4 LINCOLN S131-8 2—10 LINCOLN SA-110 5—34 LINDEX CORP. (See Swank) LIPAN (See Supreme) LULLABY (See Mitchell) LYRIC (Also See Rauland) S407, 54347V, 54347W 7—17 MAGIC TONE 5—40 S034 (Bottle Receiver) 32—10 900 (See Model 508) 38 MAGNAVOX Chossis AMP-108 43—12 Chossis AMP-109 *
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML081 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8M-952 (SEH-18805-A), 8ML9852 (SEH-18805-A), 8M-952 (SEH-18805-A), 8ML9852 (SEH-18805-A), 8M-4 LINCOLN S131-8 2—10 LINCOLN SA-110 5—34 LINDEX CORP. (See Swank) LIPAN (See Supreme) LULLABY (See Mitchell) LYRIC (Also See Rauland) S407, 54347V, 54347W 7—17 MAGIC TONE 5—40 S034 (Bottle Receiver) 32—10 900 (See Model 508) 38 MAGNAVOX Chossis AMP-108 43—12 Chossis AMP-109 *
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML081 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8M-952 (SEH-18805-A), 8ML9852 (SEH-18805-A), 8M-952 (SEH-18805-A), 8ML9852 (SEH-18805-A), 8M-4 LINCOLN S131-8 2—10 LINCOLN SA-110 5—34 LINDEX CORP. (See Swank) LIPAN (See Supreme) LULLABY (See Mitchell) LYRIC (Also See Rauland) S407, 54347V, 54347W 7—17 MAGIC TONE 5—40 S034 (Bottle Receiver) 32—10 900 (See Model 508) 38 MAGNAVOX Chossis AMP-108 43—12 Chossis AMP-109 *
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML081 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML882 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8ML982 (SEH-18805-A), 8M-952 (SEH-18805-A), 8ML9852 (SEH-18805-A), 8M-952 (SEH-18805-A), 8ML9852 (SEH-18805-A), 8M-4 LINCOLN S131-8 2—10 LINCOLN SA-110 5—34 LINDEX CORP. (See Swank) LIPAN (See Supreme) LULLABY (See Mitchell) LYRIC (Also See Rauland) S407, 54347V, 54347W 7—17 MAGIC TONE 5—40 S034 (Bottle Receiver) 32—10 900 (See Model 508) 38 MAGNAVOX Chossis AMP-108 43—12 Chossis AMP-109 *
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML081 (SEH-18805-A), 8ML8822 (BH-18805-A), 8ML8822 (BH-18805-A), 8ML8825 (BH-18805-A), 8ML985 (BL-18805-A), 8ML9852 (BH-18805-B), 8ML9852 (BH-18805-B), 9ML9852 (BH-18805-B), 8ML9852 (BH-18805-B), 8ML985 (BH-18805-B), 8ML985 (BH-18805-B), 8ML985 (BH-18805-B), 8ML985 (BH-18805-B), 8ML985 (BH-18805-B), 8ML995 (BH-19805-B), 8ML995 (BH-1995 (BH-1805-B), 710 (BH-18805-B), 710 (BH-18805-B), 711 (Chassis AMP-1102, AMP-1102, B, C. Chassis CR-1926 (SH-19805-B), 711 (Chassis CR-1926 (SH-19805-B), 711 (Chassis CR-1926 (SH-19805-B), 711 (Chassis CR-1927 (SH-19805-B), 711 (Chassis CR-1926 (SH-19805-B), 711 (Chassis CR-1926 (SH-19805-B), 711 (Chassis CR-1927 (SH-19805-B), 711 (Chassis CR-1926 (SH-19805-B), 711 (Chassis CR-1927 (SH-19805-B), 711 (Chassis CR-1926
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML080 (SEH-18805-A), 8ML882 (SL-18805-A), 8ML882 (SL-18805-A), 8ML882 (SL-18805-A), 8ML985 (SL-18805-A), 8ML985 (SL-18805-B), 8ML985 (SL-18805-B), 8MAGNAVOX Chassis AMP-100, 8MAGNAVOX Chassis AMP-100, 8MAGNAVOX Chassis AMP-100, 8ML995 (SL-1982, 8ML995 (SL-198
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML080 (SEH-18805-A), 8ML882 (SL-18805-A), 8ML882 (SL-18805-A), 8ML882 (SL-18805-A), 8ML985 (SL-18805-A), 8ML985 (SL-18805-B), 8ML985 (SL-18805-B), 8MAGNAVOX Chassis AMP-100, 8MAGNAVOX Chassis AMP-100, 8MAGNAVOX Chassis AMP-100, 8ML995 (SL-1982, 8ML995 (SL-198
507A 20—19 LINCOLN (Auto Radio) 7ML080 (SEH-18805-A), 7ML081 (SEH-18805-A), 8ML8822 (BH-18805-A), 8ML8822 (BH-18805-A), 8ML8825 (BH-18805-A), 8ML985 (BL-18805-A), 8ML9852 (BH-18805-B), 8ML9852 (BH-18805-B), 9ML9852 (BH-18805-B), 8ML9852 (BH-18805-B), 8ML985 (BH-18805-B), 8ML985 (BH-18805-B), 8ML985 (BH-18805-B), 8ML985 (BH-18805-B), 8ML985 (BH-18805-B), 8ML995 (BH-19805-B), 8ML995 (BH-1995 (BH-1805-B), 710 (BH-18805-B), 710 (BH-18805-B), 711 (Chassis AMP-1102, AMP-1102, B, C. Chassis CR-1926 (SH-19805-B), 711 (Chassis CR-1926 (SH-19805-B), 711 (Chassis CR-1926 (SH-19805-B), 711 (Chassis CR-1927 (SH-19805-B), 711 (Chassis CR-1926 (SH-19805-B), 711 (Chassis CR-1926 (SH-19805-B), 711 (Chassis CR-1927 (SH-19805-B), 711 (Chassis CR-1926 (SH-19805-B), 711 (Chassis CR-1927 (SH-19805-B), 711 (Chassis CR-1926

 MAGNAVOX-Cont.

 Chossis CR-213

 Chossis CR-214

 Chossis CR-215

 Chossis CR-216

 Chossis CR-217

 Chossis CR-218

 Chossis CR-219

 See Chorler 10

 See Chorler 11

 Rec

 Stee Chorler 129

 Chossis CR-221 Tell Rec.

 See Chorler 129

 See Mole CI-2321

 See Mole CI-2321

 See Chorler 129

 MAGNAVOX-Cont.
 Chossis MCT228 Tell. Rec.
 95A-9

 MAGUIRE
 50081, 5008W, 500D1, 5008W
 6-15

 56181, 5618W, 561D1, 5618W
 6-16
 571

 5610
 6-12
 571

 661, 661A
 12-18
 700A

 700A
 7-18
 700E
 MAJESTIC
 MAJESTIC.
 133—8

 G-414 Tel. Rec.
 133—8

 G-614 Tel. Rec.
 133

 G-624 Tel. Rec.
 133

 G-614 Tel. Rec.
 133

 G-624 Tel. Rec.
 133

 G-644 Tel. Rec.
 133

 G-644 Tel. Rec.
 133

 Sea Model G-414).
 133

 SA430 (Ch. 4501).
 1--30

 SA430 (Ch. 4504).
 1--30

 SA431 (Ch. 4501).
 23--12

 SAK711.
 27-17

 SAK731.
 5A4780.

 (Ch. 5005A).
 28--19

 SLA5. SLA6
 130--9

 G-M773 (Ch. 681D).
 57--10

 G-M777 (Ch. 680D).
 50--10

 G-M777 (Ch. 680D).
 50--10

 G-M777 (Ch. 7006).
 27--12

 7/K778 (Ch. 47006).
 27--12

 7/K778 (Ch. 47006).
 27--13

 7/K777 (Ch. 7006).
 26--14

 7/K730 (Ch. 7006).
 22--19

 7/K730 (Ch. 7007).
 24--17

 7/K730 (Ch. 7006A).
 29--13

 7/K752 (Ch. 7809A).
 29--13

MAJESTIC-Cont
 MAJESTIC-Cont.

 141, 1415 (Ch. 100), 142B (Ch. 100), 142, 142B (Ch. 100), 142, 142B (Ch. 100), 142, 142B (Ch. 100), 142, Rec. (See Model 17DA)...127

 150, 160B, 162, 163 (Ch. 101) Tel. Rec. (See Model 17DA)....127

 170, (Ch. 101) Tel. Rec. (See Model 17DA)....127

 902, 903 (Ch. 103) Tel. Rec. (See Model 17DA)....127

 910, 911 (Ch. 103) Tel. Rec. (See Model 17DA)....127

 1042, G, GU, T, 1043, G, GU, T Tel. Rec. (See Model 12C4)....108

 1142, 1143 Tel. Rec. (See Model 12C4)....108

 1244, G, GU, T, XI Tel. Rec. (See Model 12C4)....108

 1348 Tel. Rec. (See Model 12C4)....108

 1348 Tel. Rec. (See Model 12C4)....108

 1440, 14008 (Ch. 100), 1401 (Ch. 105) Tel. Rec. (See Model 17DA)

 127

 1546, G, GU, T, TX Tel. Rec. (See Model 12C4)....108

 1400, 1400B (Ch. 100), 1401 (Ch. 105) Tel. Rec. (See Model 17DA)

 127

 1505, 16058 (Ch. 102) Tel. Rec. (See Model 17DA)

 128

 1400, 1600B (Ch. 102) Tel. Rec. (See Model 17DA)

 127

 1605, 16058 (Ch. 102) Tel. Rec. (See Model 17DA)

 128

 129

 1348

 129

 130

 1607, 1608 (Ch. 102) Tel. Rec. (See Model 17DA)

 127

 (36e model /r/m62/)... 30 (5.e Model 7/18866)... 60 (5.e Model 8FM744)... 30 (5.e Model 8FM744)... 30 (5.e Model 8FM776)... 29 (5.e Model 8FM775)... 29 (5.e Model 8FM775)... 54 (5.e Model 8FM889)... 54 (5.e Model 10FM981)... 65 (5.e Model 10FM981)... 65 Ch. (5 (See Model 10FMyor),... h. 12826E [See Model 12FM475]... 28 h. 12C22E [See Model 12FM895]... 59 [See Model 12FM895]... 59 СЬ

 [388 Model 12FMA23].
 28

 [386 Model 12FMA23].
 59

 [386 Model 12FM895].
 59

 [386 Model 71450]...
 *

 [458 Model 74450]...
 1

 [586 Model 5A410]...
 1

 [586 Model 5A430]...
 1

 [586 Model 5A430]...
 1

 [586 Model 5A430]...
 23

 [586 Model 5A430]...
 24

 [587 Model 75433]...
 22

 [586 Model 75433]...
 26

 [586 Model 76437]...
 14

 [586 Model 74473]...
 14

 [586 Model 74477]...
 14

 [586 Model 748778].
 27

 [586 Model 74875]...
 8

 [586 Model 81885]...
 47

 [586 Model 814885]...
 47

 [586 Model 84 8475]...
 28

 </tbr>
</tabr> Ch MANTOLA (B. F. Goodrich Co.)
 MANTOLA (B. F. Goodrich Co.)

 R630-RP
 3--22

 R643-PM (See Model
 4

 R643V)
 4

 R643VH
 4-29

 R652, R652N
 9--22

 R654.PM, R643W
 4-29

 R654.PM, R654.PV
 3--5

 R655W
 9--22

 R654.PM, R654.PV
 3--33

 R654.PM, R654.PV
 3--33

 R664.PM, R664.PV,
 3-33

 R64.PM, R664.PV,
 3-33

 R74.3W (See Model
 8-643-W)

 R-754.3
 18-23

 R-751.43
 39-12

 R-743-W
 (See Model

 R-643-W)
 4

 R-7543
 18—23

 R-75143
 39—12

 R-75152
 38—10

MANTOLA-Cont.	
	MECK-Cont. XL750 Tel. Rec
R-75343 (See Model 75143) 39 R-76143 (See Model 2486) 25 R-76162 40-10	XL750 Tel. Rec
R-76162	Model XF-777) 101
R-76262 (Fact. No. 7160-17)	XOB Tel. Rec. (See Model MM510T)
11 701	XP-775, XQ-776, XQA-776 Tel Rec (See Model
24B6	XF-777)
92-520, 92-521, 92-522 68-11 92502 (See Model	XQA, XQR Tel. Rec. (See
R043 W	Model MM510T)
92503, 92504 (See Models R654FM, PV) 3	Model MM510T)110 XR-778, XS-786, XT-785 Tel. Rec. (See Model
92505, 92506 (See	Tel. Rec. (See Model
Models R664PM, PV,	AF-///
PW) 23 92516, 92517 92752	XSA Tel. Rec. (See Model MM510T)
92752 ****	XSB Tel. Rec.
MARK SIMPSON (See Masco)	(See Model MM614C)117 XSB Tel. Rec.
MASCO	XSB Tel. Rec. Prod. Chge. Bul. 12120-1
IM-5 41—13	
IM-5	(See Model MM510T)110 XTA, XTR Tel. Rec. (See
JM-5 (Master Station), JR (Sub-Station) 42—18	model mm3101)
JA-3 (Master Station), 42—18 MA-5NO 45—15 MA-8N. 119—8 MA-10HF 112—4 MA-10EK 113—4 MA-12HF 51—13 MA-17N 14—32 MA-17N 50—114	XX900 Tel. Rec. (See Model MM510T)
MA-8N	4B7*
MA-10EX	4C7
MA-12HF 51—13	5D7/WL18 21-22
MA-17	6A6-W4
MA-17P (See Model MA-17) 14	514C, T (Ch. 9018) Tel. Rec. * 616C, T (Ch. 9018) Tel. Rec. *
MA-17PN (See Model MA-17N)	616C, T (Ch. 9018)
	619C (Ch. 9018)
MA-25	Tel. Rec *
MA-25HF	MEDCO (See Telesonic)
MA-25N 43-14	MEISSNER
MA-25P (See Model MA-25) 16	TV-1 (Ch. 24TV) Tel. Rec. 56-15
	5A (See Maguire Model 571)
MA-25PN (See Model MA-35N) 43 MA-35N 41-20 MA-35N 44-11 MA-35RC (See Model MA-35) 21 MA-50 30-16	
MA-35N	oH (See Maguire Model 661, 661, 661A) 12 8C
MA-35RC (See Model MA-35)	9AJ
MA-33)	9-1065
MA-50N (See Model	9-1091A, 9-1091B 35—15 9-1091C
MA-5N0)	9-1093 55-13
MA-60	24TV Tel. Rec
MA-75	16A
MA-121	25TV Tel. Rec
MA-808 26—18	571)
MAP-13	571)
MAP-105	661)
MAP-105N	MERCURY
MAP-120N 46-15	6MM790 (See Ford
MB-50N	Model 6MF780) 62
MA-50N (See Model MA-75N (See Model MAP-12 (See Model MAP-105 (See Model MAP-100 (See Model MAP-120 (See Model MB-60 (See Model MC-10 (See Model MC-10 (See Model	8MM890 (Ch. 8E90) (8M-18805-8) 49-13
MC-10	8MM990 (8M-18805-B) 69-10
MC-25, MC-25P	(3M-18805-B)
MB-30N 38—12 MB-60 127—8 MB-75 61—13 MC-10 47—12 MC-25, MC-25P 17—21 MC-25, MC-25P 17—21 MC-25PN, MC-25PC, MC-25N, MC-25PC, MC-25NN, MC-25PC, 111—8 MHP-110 114—6 MHP-110X 115—5 Midgetolk 116—7 MC-5 137—6 RK-5 33—11 T-16 123—8 TP-16 120—8 TP-16 20—80 MASON 20—21	MIDLAND
MC-126, MC-126P 1118 MHP-110	M6B
MHP-110X	MIDWEST
MPA-3, MPT-4	P-6, PB-6 14-19
MCR-5	P-6, PB-6
RK-5 33—11	R-12, RG-12, RT-12 (Ch. RGT-12)
T-16	R-16, RG-16, RT-16
TD-16	
76, 711 20 —20	(Ch. STM-8) 15—19
86, 811 20 —21	S-12, SG-12, ST-12
	(Ch. KG1-16)
45-1A	[Ch. SGI-10]
45-1A	716, 716A (See Model S-16) 21
MAYFAIR	MINERVA
510, 510W, 520, 520W,	1 700 //
530, 530W	L-728, W-728
McGRADE	L-702 (See W-702B) 12 L-728, W-728 1115 W-117, Tropic Master 617 W-117-3 1114 W-702B 1220
M-100 16-27	W-702B 12-20 W710, W710A (W119) 5-25
MECH (Tanit Blanny Blumouth)	W-728 (See Model L-728). 11
	410, 411 41-14
CE-500 (5C5-P12) 34-10	702H, 702H-1 30—18 729 (Portapat) 23—14
Cb-500 (FA-SC3-EV-1Y) 33-12 CE-500 (SC5-P12) 34-10 CM-500 (SD7-W18) 34-11 CW-500 38-11	MIRRORTONE (See Meck)
CW-500 4011	MITCHELL
DA601, DB6021 81—10 EC720	1250, 1251
EC720	1267 (See Model 1268R)127 1268R
Ch. 10003)	MOLDED INSULATION CO.
EV-/00	(Also see Viz)
MM3101, MM3121,	MR-6 (Wiretone) 41—15
Tel. Rec	MONITOR
MM614C, T Tel. Rec	M-403 (Fact. No. 470-2) . 22—20 M-500 (Fact. No. 475) 22—23
MM614C, T Tel. Rec. Prod. Chge. Bul. 12 120 1 MM616C, T Tel. Rec. (See Model MM614C1 117	M-510 (Fact. No. 472) 23-15
(See Model MM614C)117	M-3070 29—15
MM616C, T Tel. Rec.	TA56M, TW56M
MM616C, T Tel. Rec. Prod. Chge. Bul. 12120-1 MM619C Tel. Rec. (See Model MM614C)117	M. 3070
(See Model MM614C)117	(See Airline)
	MOPAR
PM-5C5-DW10	602 (Colonial Model
Prod. Chge. Bul. 12120—1 PM-5C5-DW10 2—4 PM-5C5-PW10 12—19 RC-5C5-P 1—9 RC-6A7-P6 31—19	671A) 1920 603 65-9
RC-6A7-P6	604
RC-6A7-P6 31—19 SA-10, SA-20 101—4 XA-701 Tel. Rec. 101—16 XE-705 [See Model 41	606
XE-705 (See Model	802 (Philco C-4608)
XA-701)	(Revised)

MECK-Cont. XOB. Tel. Rec. (See Model MMS101). 110 XP-775, XQ-776, XQA-776 101 XG-775, XQ-776, XQA-776 101 XGA, YGR Tel. Rec. (See Model MMS101). 110 XRA, XRP Tel. Rec. (See Model MMS107). 110 XRA, XRP Tel. Rec. (See Model MMS107). 110 XRA. Tel. Rec. (See Model MMS107). 110 XSA Tel. Rec. (See Model MMS107). 101 XSA Tel. Rec. (See Model MMS107). 110 XSB Tel. Rec. See Model XF-777) XSB Tel. Rec. See Model MMS107). XB Tel. Rec. See Model MMS107). XA, XRP Tel. Rec. See Model MMS107). XA, XRP Tel. Rec. See Model MMS107). XA, XRP Tel. Rec. See Model MMS107). XPO Tel. Rec. See Model MMS107). 110 XP00 Tel. Rec. See Model MMS107). 110 XP00 Tel. Rec. See Model MMS107). 110 XP00 Tel. See Model MMS107). 110 XP00 Tel. See Model MMS107). 110 487 4C7 5A7-P11, 5A7-P811 31-5D7/W118 21-6A6.W4 16-514C, 17 (Ch, 9018) Tel, Rec. 619C (Ch, 9018) Tel, Rec. 519C (Ch, 9018) 35—14 31—18 21—22 16—26 MEDCO (See Telesonic) MEISSNER TV-1 (Ch. 24TV) Tel. Rec., 56-15 5A (See Maguire Model 571) 001, 661A) 12 8C 37—12 9AJ 123—9 9.1065 123—9 9-1065 3-15 9-1091A, 9-1091B 35-15 9-1091C 16-8 9-1093 55-13 16A 105-6 661) 12 2961 Series 27—19 MERCURY MERCURY 6MM790 (See Ford Model 6MF780) 62 8MM890 (Ch. 8E90) 49—13 8MM990 (8M-18805-8) 69—10 8MM990 (8M-18805-8), 69—10 8MM991 (8M-18805-8), 83—4 MIDLAND MIDWEST P-6, PB-6 14—19 R:-12, RG-12, RT-12 44—12 ICh. RGI-12, RT-12 44—12 R:-12, RG-12, RT-12 44—13 R:-16, RG-16, RT-16 45—16 S8, ST-6, RG-16, RT-16 15—19 S1-12, SG-12, ST-12 15—19 ICh. SGT-12, I.S. 15, ST-16 21—23 S1-6, SG-16, S, ST-16 21—24 ICh. SGT-16, S-16) 21—24 ISee Model S-16) 21 MINFEVA 21 MINERVA MINERVA 12 L-702 (See W-7028) 11 L-728 (W-728) 11 W-117, Tropic Master 6-17 W-117, Topic Master 11 W-7028 12-20 W-703 12-14 W-704 (W19) 5-25 W-728 (See Model L-728) 11 410, 411 40-41 702H, 702H-1 30-18 729 (Portopal) 23-14 MIRRORTONE (See Meck) MIRCORIONE Stevensor MITCHELL 55—14 1250, 1251 55—14 1267, [See Model 1268R] 127 1268R 127 MODIDED INSULATION CO. (Also see Viz) MR-6 (Wiretone) 41—15 MONITOR M-403 (Fact. No. 470-2) 22—20 M-500 (Fact. No. 475) 22—23 M-510 (Fact. No. 472) 23—15 M-3070 29—15 RA-50 24—23 TA56M, TW56M 6—18 MONTGOMERY WARD (See Airline) MOPAR 602 (Colonial Model 671A) 603 **19**---20 **65**--9 **106**--9 **133**--9 **18**---24 606 802 (Philco C-4608) 802 (Philco C-4608)

MAGNAVOX-MOTOROLA

MOPAR-Cont. -6

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 For an and an antipart of a set of

MOTOROLA-MUSITRON

MOTOROLA-Cont.

MOTOROLA-Cont. MOTOROLA-Cont. 1755A, 1755BA (Ch. TS-89 & Rodio Ch. HS-261) Tel. Rec. (See Model 14K1BH) 121 1765B (Ch. TS-118) Tel. Rec. (See Model 14K1BH) 121 177B (Ch. TS-118) Tel. Rec. (See Model 14K1BH) 121 177B (Ch. TS-118) Tel. Rec. (See Model 14K1BH) 121 177B (Ch. TS-118) Tel. Rec. (See Model 14K1BH) 121 177B (Ch. TS-118) Tel. Rec. (See Model 14K1BH) 121 177B (Ch. TS-118) Tel. Rec. (See Model 14K1BH) 121 177B (Ch. TS-118) Tel. Rec. (See Model 14K1BH) 121 177B (Ch. TS-118) Tel. Rec. (See Model 14K1BH) 121 177B (Ch. TS-1174) Tel. Rec. (See Model 14K1BH) 121 177B (Ch. TS-1174) Tel. Rec. (See Model 14K1BH) 121 17K1A, 17K1BA (Ch. TS-172) Tel. Rec. See Model 14K1BH) 121 17K3A (Ch. TS-174) Tel. Rec. (See Model 14K1BH) 121 17K3A (Ch. TS-174) Tel. Rec. (See Model 14K1BH) 121 17K3A (Ch. TS-174) Tel. Rec. (See Model 14K1BH) 121 17K3A (Ch. TS-174) Tel. Rec. (See Model 14K1BH) 121 17K3A (Ch. TS-174) Tel. Rec. (See Model 14K1BH) 121 17K4A (Ch. TS-174) Tel. Rec. (See Model 14K1BH) 121 17K4A (Ch. TS-174) Tel. Rec. (See Model 14K1BH) 121 17K4A (Ch. TS-174) Tel. Rec. (See Model 14K1BH) 121 17K4A (Ch. TS-174) Tel. Rec. (See Model 14K1BH) 121 17K4A (Ch. TS-174) Tel. Rec. (See Model 14K1BH) 121 17K5C (Ch. TS-174) Tel. Rec. (See Model 14K1BH) 121 17K5C (Ch. TS-174) Tel. Rec. (See Model 14K1BH) 121 17K5C (Ch. TS-174) Tel. Rec. (See Model 14K1BH) 121 17K3A (Ch. TS-180) Tel. Rec. (See Model 14K1BH) 121 17K3A (Ch. TS-180) Tel. Rec. (See Model 14K1BH) 121 17K3A (Ch. TS-118) Tel. Rec. (See Model 14K1BH) 121 17K3A (Ch. TS-118) Tel. Rec. (See Model 14K1BH) 121 17K3A (Ch. TS-118) Tel. Rec. (See Model 14K1BH) 121 17K3A (Ch. TS-118) Tel. Rec. (See Model 14K1BH) 121 17K3A (Ch. TS-118) Tel. Rec. (See Model 14K1BH) 121 17K3A (Ch. TS-118) Tel. Rec. (See Model 14K1BH) 121 17K3A (Ch. TS-118) Tel. Rec. (See Model 14K1BH) 121 17K4A (Ch. TS-118) Tel. Rec. (See Model 14K1BH) 121 17K4A (Ch. TS-118) Tel. Rec. (See Model 1

MOTOROLA-Cont.	1	1
BR11A, 58R12A, 58R13A, 58R14A, 58R15A, 58R16A, 58R15A, 58R16A (Ch. H5-184) 58X11, 58X12 (Ch. H5-125) 59H11(Ch. H5-188) 59H11(Ch. H5-210) 59H10(Ch. H5-210)	69-11	0
58X11, 58X12 (Ch. HS-125)	53—15 68—12	¢
59F111 (Ch. HS-188). 59F1110, 59H1210, 59H1210, 10(Ch. HS-210). 59L1120, 59U120, 59U121, 55U12, 5	97—9	¢
59L11Q, 59L12Q, 59L14Q (Ch. HS-187)	78 —10	000
59R11, 59R121, 59R13M, 59R14E, 59R15G,	79 —10	¢
59X11, 59X121 (Ch. HS-180)	8111	C
59X21U, 59X221U {Ch. HS-192}	98_6	¢
65F11 (Ch. HS-31) 65F12 (See Model 65F11). 65F21 (Ch. HS-26)	619 6 412	0
65L11, 65L12 (Ch. HS-7) . 65T21, 65T21B	8-22	
(Ch. HS-32)	1_)	
65X14B (Ch. HS-2) 67F11, 67F12, 67F12B,	48	
(Ch. HS-63) 67F14 (Ch. HS-122)	31-20 55-15 44-14	1
67L11 (Ch. HS-59) 67X11, 67X12, 67X13	31-21	
(Ch. HS-58) 67XM21 (Ch. HS-64)	30—20 32—14	
68F11, 68F12, 68F14, 68F14B, 68F14M 68L11 (Ch. HS-119)	58	
68T11 (Ch. HS-144) 68X11, 68X12 (Ch.	54-14	
HS-127), 68X11A, 68X12A (Ch. HS-127A).	56—16 76—15	•
68F148, 68F144 68L11 (Ch. HS-119) 68T11 (Ch. HS-114) 68X11, 68X12 (Ch. HS-127), 68X12 (Ch. 68X124 (Ch. HS-127A). 69X11, 69X121 (Ch. HS-181) (Ch. HS-181)	82-9	
75F21 (Ch. HS-91) 75F31 (Ch. HS-36),	19-21	
69x11, 69x121 (Ch. HS-181) 75F31 (Ch. HS-36), 75F31 (Ch. HS-36), 75F31 (Ch. HS-36), 76F31 (Ch. HS-98) 77FM21 (Ch. HS-98) 77FM22, 77FM22M, 77FM22WM, 77FM23 (Ch. HS-97)	29 —18	,
77FM22, 77FM22M, 77FM22WM, 77FM23		•
77FM22WM, 77FM23 (Ch. HS-97). 77XM21, 77XM22, 77XM228 (Ch. HS-102). 78F11, 78F11M (Ch. HS-150), 78F12M (Ch. HS-155). 78FM21, 78FM21M (Ch.	33—13 34—12	
78F11, 78F11M (Ch. HS-150), 78F12M (Ch.	3412	
HS-155) 78FM21, 78FM21M (Ch.	56 —17	1
(Ch. HS-128) 79FM21, 79FM21B,	59 —13	1
79FM21R (Ch. HS-178). 79XM21, 79XM22	887	P
85F21 (Ch. HS-22) 85K21 (Ch. HS-52)	6-20 53	
H3-150], 78F124M (Ch. H3-155], 78FM22H (Ch. H3-132), 78FM22M (Ch. H3-132), 78FM22M (Ch. H3-128), 79FM21R (Ch. H5-128), 79FM21R (Ch. H5-128), 79FM21R (Ch. H5-128), 79FM21R (Ch. H5-128), 88FM21 (Ch. H5-128), 88FM21 (Ch. H5-129), 88FM21 (Ch. H5-129), 95F318 (Ch. H5-129), 95F318 (Ch. H5-129), 95F318 (Ch. H5-129), 95F318 (Ch. H5-129), 107F318, (Ch. H5-87), 007F318, (Ch. H5-87), 007F318, (Ch. H5-87)	54 —15	
(See Model 19F1) 95F31, 95F31B (Ch. HS-39) 95F33 (Ch. HS-38)	19—22	.
99FM21R (Ch. HS-170) 107F31, 107F31B,	19 —22 80 —10	'
(Ch. HS-87) 309 400	33—14 63—14 99—10	
405 (Ch. AS-13)	3—8	
405M 408 409 (See Model 408)	2125 38-12 38	
500	987 13310	
505 (Ch. AS-14) 508 509 (See Model 508)	4-3/	
600 605 (Ch. AS-15)	9710 51	
608 609 (See Model 608) 700 705 (Ch. AS-16)	39—14 39 100—8	
709 [See Model 708]	40 103—10	
800 Ch. AS-13 (See Model 405) Ch. AS-14 (See Model 505) Ch. AS-15 (See Model 505) Ch. AS-15 (See Model 605) Ch. AS-16 (See Model 705) Ch. AS-22 (See Model	3 4 5	
Ch. AS-16 (See Model 705) Ch. AS-22 (See Model	7 10	
BK-6) Ch. HS-2 (See Model 65X11A)		
Ch. HS-2 (See Model 65X11A) Ch. HS-6 (See Model 5A1) Ch. HS-7 (See Model 65L11) Ch. HS-8 (See Model	2	
Ch. HS-15 (See Model 5A5)	3	
Ch. HS-18 (See Model WR6) Ch. HS-22 (See Model 85F21) Ch. HS-26 (See Model	5	
Ch. HS-26 (See Model 65F21)	4	
65F21) Ch. HS-30 (See Model 55F11) Ch. HS-31 (See Model	4	
65F11}		
Ch. HS-32 (See Model 65T21) Ch. HS-36 (See Model 75F31) Ch. HS-36A (See Model		
7562141	29	
95F33)	10	
	19	
55X11A)	Ź	1

Ch. H5.52 (See Model 5 Ch. H5.58 (See Model 30 Ch. H5.59 (See Model 31 Ch. H5.59 (See Model 31 Ch. H5.60 (See Model 31 Ch. H5.62 (See Model 57(1) S7X1 (See Model 33 Ch. H5.62 (See Model 33 Ch. H5.62 (See Model 33 Ch. H5.62 (See Model 33 Ch. H5.64 (See Model 33 Ch. H5.64 (See Model 33 Ch. H5.64 (See Model 33 Ch. H5.67 (See Model 33 Ch. H5.69 (See Model 33 Ch. H5.79 (See Model 33 Ch. H5.79 (See Model 33 Ch. H5.79 (See Model 34 Ch. H5.19 (See Model 34 Ch. H5.19 (See Model 34 Ch. H5.19 (See Model 34 Ch. H5.10 (See Model 45 Ch. H5.10 (See Model 47 Ch. H5.11 (See Model 47 Ch. H5.12 (See Model 48 Ch. H5.12 (See Model 55 Ch. H5.12 (See Model 56 Ch. H5.13 (See Model 56	MOTOROLA-Cont.	
Ch. HS-39 (See Model 57X11)	Ch. HS-52 (See Model 85K21)	5
Ch. HS-39 (See Model 57X11)	Ch. HS-58 (See Model 67X11)	30
57X11) 28 Ch. HS-62 (See Model SA7) 29 5A7A) 29 Ch. HS-62 (See Model A 29 5A7A) 31 Ch. HS-62 (See Model A 31 Ch. HS-62 (See Model A 33 Ch. HS-62 (See Model A 33 Ch. HS-72 (See Model A 33 Ch. HS-77 (See Model A 33 Ch. HS-76 (See Model A 34 Ch. HS-77 (See Model A 34 Ch. HS-103 (See Model A 34 Ch. HS-104 (See Model A 35 Ch. HS-113 (See Model A 35 Ch. HS-124 (See Model A 35 Ch. HS-127 (See Model A 35 Ch. HS-128 (See Model A 35 Ch. HS-129 (See Model A 35 Ch. HS-121 (See Model A 35 Ch. HS-122 (See Model	Ch. H5-59 [See Model	
Ch. HS-63 (See Model 67714). Ch. HS-64 (See Model 67714). Ch. HS-67 (See Model 67764). Ch. HS-72 (See Model 777672). Ch. HS-97 (See Model 76731). Ch. HS-113 (See Model 45 Ch. HS-113 (See Model 45 Ch. HS-114 (See Model 45 Ch. HS-119 (See Model 45 Ch. HS-122 (See Model 58 Ch. HS-122 (See Model 58 Ch. HS-122 (See Model 58 Ch. HS-124 (See Model 58 Ch. HS-127 (See Model 58 Ch. HS-127 (See Model 68 Ch. HS-137 (See Model 68 Ch. HS-138 (See Model 59 Ch. HS-128 (See	Ch. HS-60 (See Model 57X11)	28
Ch. HS-63 (See Model 67714). Ch. HS-64 (See Model 67714). Ch. HS-67 (See Model 67764). Ch. HS-72 (See Model 777672). Ch. HS-97 (See Model 76731). Ch. HS-113 (See Model 45 Ch. HS-113 (See Model 45 Ch. HS-114 (See Model 45 Ch. HS-119 (See Model 45 Ch. HS-122 (See Model 58 Ch. HS-122 (See Model 58 Ch. HS-122 (See Model 58 Ch. HS-124 (See Model 58 Ch. HS-127 (See Model 58 Ch. HS-127 (See Model 68 Ch. HS-137 (See Model 68 Ch. HS-138 (See Model 59 Ch. HS-128 (See	Ch. HS-62 (See Model 5A/) Ch. HS-62A (See Model	29
Ch. H3: 437 (See Model 29 Ch. H3: 597 (See Model 33 Ch. H3: 597 (See Model 33 Ch. H3: 597 (See Model 33 Ch. H3: 597 (See Model 75721) 75721 (See Model 33 Ch. H3: 597 (See Model 29 Ch. H3: 597 (See Model 29 Ch. H3: 597 (See Model 29 Ch. H3: 103 (See Model 29 Ch. H3: 103 (See Model 29 Ch. H3: 103 (See Model 41 Ch. H3: 101 (See Model 41 Ch. H3: 113 (See Model 45 Ch. H3: 114 (See Model 58 Gh. H3: 122 (See Model 58 Gh. H3: 122 (See Model 58 Gh. H3: 123 (See Model 59 Gh. H3: 123 (See Model 59 Gh. H3: 133 (See Model 58 Gh. H3: 133 (See Model 54 Gh. H3: 133 (See Model 54 Gh. H3: 134 (See Model 56	Ch. HS-63 (See Model	31
Ch. H3: 437 (See Model 29 Ch. H3: 597 (See Model 33 Ch. H3: 597 (See Model 33 Ch. H3: 597 (See Model 33 Ch. H3: 597 (See Model 75721) 75721 (See Model 33 Ch. H3: 597 (See Model 29 Ch. H3: 597 (See Model 29 Ch. H3: 597 (See Model 29 Ch. H3: 103 (See Model 29 Ch. H3: 103 (See Model 29 Ch. H3: 103 (See Model 41 Ch. H3: 101 (See Model 41 Ch. H3: 113 (See Model 45 Ch. H3: 114 (See Model 58 Gh. H3: 122 (See Model 58 Gh. H3: 122 (See Model 58 Gh. H3: 123 (See Model 59 Gh. H3: 123 (See Model 59 Gh. H3: 133 (See Model 58 Gh. H3: 133 (See Model 54 Gh. H3: 133 (See Model 54 Gh. H3: 134 (See Model 56	Ch. HS-64 (See Model 67XM21)	33
Ch. H3: 437 (See Model 29 Ch. H3: 597 (See Model 33 Ch. H3: 597 (See Model 33 Ch. H3: 597 (See Model 33 Ch. H3: 597 (See Model 75721) 75721 (See Model 33 Ch. H3: 597 (See Model 29 Ch. H3: 597 (See Model 29 Ch. H3: 597 (See Model 29 Ch. H3: 103 (See Model 29 Ch. H3: 103 (See Model 29 Ch. H3: 103 (See Model 41 Ch. H3: 101 (See Model 41 Ch. H3: 113 (See Model 45 Ch. H3: 114 (See Model 58 Gh. H3: 122 (See Model 58 Gh. H3: 122 (See Model 58 Gh. H3: 123 (See Model 59 Gh. H3: 123 (See Model 59 Gh. H3: 133 (See Model 58 Gh. H3: 133 (See Model 54 Gh. H3: 133 (See Model 54 Gh. H3: 134 (See Model 56	Ch. HS-69 (See Model 67F61BN)	44
107731 33 77FM21 33 77FM21 33 77FM21 19 Ch. HS-91 (See Model 33 77FM21 19 Ch. HS-94 (See Model 33 Ch. HS-94 (See Model 33 Ch. HS-96 (See Model 34 Ch. HS-98 (See Model 34 Ch. HS-108 (See Model 34 Ch. HS-113 (See Model 47 Ch. HS-114 (See Model 45 Ch. HS-114 (See Model 45 Gh. HS-114 (See Model 55 Ch. HS-115 (See Model 56 Gh. HS-124 (See Model 56 Gh. HS-127 (See Model 56 Gh. HS-128 (See Model 58 Gh. HS-128 (See Model 58 Gh. HS-128 (See Model 58 Gh. HS-138 (See Model 58 <	Ch. HS-72 (See Model 47B11)	29
77FA21) 33 77FA21) 19 75F21) 19 56X11) 28 75F21) 33 75F21) 33 77FA22) 33 77FA22) 33 77FA22) 33 77FA22) 33 77FA22) 33 77FA21) 34 77FA21) 34 77FA22) 34 77FA21) 34 77FA21) 58 78 58 78 58 78 58 78 58 78 58 78 59 78 59 78 59 78 59 78 59 78 59		33
Ch. H5-74 (See Model 28 Ch. H5-77 (See Model 33 Ch. H5-77 (See Model 33 Ch. H5-77 (See Model 34 Ch. H5-78 (See Model 34 Ch. H5-102 (See Model 34 Ch. H5-102 (See Model 34 Ch. H5-102 (See Model 34 Ch. H5-113 (See Model 45 Ch. H5-113 (See Model 45 Ch. H5-114 (See Model 56 Ch. H5-119 (See Model 56 Ch. H5-112 (See Model 56 Ch. H5-122 (See Model 58 Ch. H5-123 (See Model 58 Ch. H5-124 (See Model 58 Ch. H5-127 (See Model 58 Ch. H5-128 (See Model 58 Ch. H5-128 (See Model 58 Ch. H5-137 (See Model 59 Ch. H5-137 (See Model 59 Ch. H5-137 (See Model 59 Ch. H5-138 (See Model 56 Ch. H5-139 (See Model<	77FM21)	33
Solvi 1) 28 Ch. HS-97 (See Model 33 Ch. HS-98 (See Model 33 Ch. HS-98 (See Model 29 Ch. HS-108 (See Model 34 Ch. HS-108 (See Model 34 Ch. HS-108 (See Model 34 Ch. HS-113 (See Model 34 Ch. HS-114 (See Model 47 Ch. HS-114 (See Model 49 Ch. HS-119 (See Model 49 Ch. HS-119 (See Model 55 Ch. HS-124 (See Model 56 GB111 (See Model 56 Ch. HS-127 (See Model 56 GAN113 (See Model 56 Ch. HS-127 (See Model 57 Ch. HS-128 (See Model 58 Ch. HS-128 (See Model 56 Ch. HS-128 (See Model 58 Ch. HS-133 (See Model 58 Ch. HS-133 (See Model 58 Ch. HS-133 (See Model 58 GENTIA 54 Ch. HS-133 (See Model 58 GENTIA 56 Ch. HS-151 (See Model 58	75F21)	19
76731) 29 77XM21) 34 Ch. HS-102 (See Model VK.101) 34 Ch. HS-1108 (See Model 48111) 31 Ch. HS-113 (See Model 58811) 47 Ch. HS-114 (See Model 58811) 47 Ch. HS-115 (See Model 6811) 45 Ch. HS-116 (See Model 6811) 45 Ch. HS-119 (See Model 6811) 55 Ch. HS-122 (See Model 68711) 58 Ch. HS-128 (See Model 68811) 56 Ch. HS-128 (See Model 68811) 56 Ch. HS-128 (See Model 688742) 57 Ch. HS-132 (See Model 78FM22M) 54 Ch. HS-133 (See Model 78FM21) 54 Ch. HS-155 (See Model 78F12M) 54 Ch. HS-158 (See Model 78F12M) 54 Ch. HS-158 (See Model 78F12M) 54 Ch. HS-158 (See Model 78F11) 54 Ch. HS-158 (See Model 78F12M) 54 Ch. HS-158 (See Model 78F12M) 54 Ch. HS-158 (See Model 78F12M) 56 Ch. HS-160 (See Model 78F12M) 56 Ch. HS-178 (See Model 79FM21) 85 Ch. HS-180 (See	56X11)	28
76731) 29 77XM21) 34 Ch. HS-102 (See Model VK.101) 34 Ch. HS-1108 (See Model 48111) 31 Ch. HS-113 (See Model 58811) 47 Ch. HS-114 (See Model 58811) 47 Ch. HS-115 (See Model 6811) 45 Ch. HS-116 (See Model 6811) 45 Ch. HS-119 (See Model 6811) 55 Ch. HS-122 (See Model 68711) 58 Ch. HS-128 (See Model 68811) 56 Ch. HS-128 (See Model 68811) 56 Ch. HS-128 (See Model 688742) 57 Ch. HS-132 (See Model 78FM22M) 54 Ch. HS-133 (See Model 78FM21) 54 Ch. HS-155 (See Model 78F12M) 54 Ch. HS-158 (See Model 78F12M) 54 Ch. HS-158 (See Model 78F12M) 54 Ch. HS-158 (See Model 78F11) 54 Ch. HS-158 (See Model 78F12M) 54 Ch. HS-158 (See Model 78F12M) 54 Ch. HS-158 (See Model 78F12M) 56 Ch. HS-160 (See Model 78F12M) 56 Ch. HS-178 (See Model 79FM21) 85 Ch. HS-180 (See	77FM22) Ch. HS-98 (See Model	33
Ch. H3-100 [See Model \$1 VK-101 [See Model 48111] 48111] 45 Ch. H3-113 [See Model 45 SB11] 45 Ch. H3-114 [See Model 58 SB11] 45 Ch. H3-112 [See Model 55 Ch. H3-122 [See Model 57 GAT14] 58 Ch. H3-122 [See Model 56 GAT14] 56 Ch. H3-122 [See Model 56 GAT14] 56 GAT13] 56 GAT14] 56 GAT15] 56 GAT13] 56 GAT13] 56 GAT13] 56 GAT13] 56 GAT13] 56 GAT13] 56 GAT14] <td>76F31)</td> <td>29</td>	76F31)	29
Ch. H3-117 (See Model 68L11) 68L11) 67F14) 67F14) 67F14) 68F11) 68F11) 68F11) 68F11) 68F11) 68F11) 68F11) 68F11) 68F11) 68T11 68T11 68T11 68T11 68T11 68T11 68T11 68T12 68T13 68T62 68T62 68T62 68T62 68T62 68T62 68T62 68T62 68T613 788T614 68T613<	77XM21) Ch. HS-108 (See Model	34
Ch. H3-117 (See Model 68L11) 68L11) 67F14) 67F14) 67F14) 68F11) 68F11) 68F11) 68F11) 68F11) 68F11) 68F11) 68F11) 68F11) 68T11 68T11 68T11 68T11 68T11 68T11 68T11 68T12 68T13 68T62 68T62 68T62 68T62 68T62 68T62 68T62 68T62 68T613 788T614 68T613<	VK-101) Ch. HS-113 (See Model	51
Ch. H3-117 (See Model 68L11) 68L11) 67F14) 67F14) 67F14) 68F11) 68F11) 68F11) 68F11) 68F11) 68F11) 68F11) 68F11) 68F11) 68T11 68T11 68T11 68T11 68T11 68T11 68T11 68T12 68T13 68T62 68T62 68T62 68T62 68T62 68T62 68T62 68T62 68T613 788T614 68T613<	48[11] Ch. HS-114 (See Model	4/
Ch. H3-117 (See Model 68L11) 68L11) 67F14) 67F14) 67F14) 68F11) 68F11) 68F11) 68F11) 68F11) 68F11) 68F11) 68F11) 68F11) 68T11 68T11 68T11 68T11 68T11 68T11 68T11 68T12 68T13 68T62 68T62 68T62 68T62 68T62 68T62 68T62 68T62 68T613 788T614 68T613<	Ch, HS-116 (See Model	40
Ch. HS. J. 4 (See Model 6B-1) SB Ch. HS. 12 (See Model 5B(1) SB Ch. HS. 127 (See Model 6B(1) SB Ch. HS. 128 (See Model 78FM21) SP Ch. HS. 132 (See Model 88FM21) SP Ch. HS. 132 (See Model 88FM21) SP Ch. HS. 137 (See Model 88FM21) SP Ch. HS. 137 (See Model 78F11) SE Ch. HS. 137 (See Model 78F11) SE Ch. HS. 150 (See Model 78F11) SE Ch. HS. 153 (See Model 78F11) SE Ch. HS. 158 (See Model 58C11) SE Ch. HS. 160 (See Model 58C11) SE Ch. HS. 168 (See Model 59R11) SE Ch. HS. 178 (See Model 58C11A) SE Ch. HS. 178 (See Model 58C11A) SE Ch. HS. 184 (See Model 59F11) SE Ch. HS. 184 (See Model 59F11) SE Ch. HS. 188 (See Model 59F11) SE Ch. HS. 223 (See Model 59111Q) SE Ch. HS. 224 (See Model 59F11) <	Ch. HS-119 (See Model 68111)	45
Ch. HS-132 (See Model 59 78FW21) 54 88FW21) 54 88FW21) 54 RASIM 21) 54 88FW21) 54 Ch. HS-137 (See Model 54 VK101) 51 Ch. HS-137 (See Model 56 Ch. HS-150 (See Model 56 Ch. HS-155 (See Model 56 Ch. HS-158 (See Model 58 S8A11) 52 Ch. HS-160 (See Model 58 S8A11) 64 Ch. HS-160 (See Model 58 S8A11) 64 Ch. HS-168 (See Model 58 S8C11 64 Ch. HS-178 (See Model 59 58 Ge Model 97 79 Ch. HS-178 (See Model 59 58 Ge Model 59 77 Ch. HS-184 (See Model 59 58 Ge Model 59 77 Ch. HS-184 (See Model 59 59 Ch. HS-184 (See Model 59 59 Ch. HS-184 (See Model 59 59 Ch. HS-123 (See Model 59 68 Ch. HS-223 (See Model	Ch. HS-122 (See Model 67F14)	55
Ch. HS-132 (See Model 59 78FW21) 54 88FW21) 54 88FW21) 54 RASIM 21) 54 88FW21) 54 Ch. HS-137 (See Model 54 VK101) 51 Ch. HS-137 (See Model 56 Ch. HS-150 (See Model 56 Ch. HS-155 (See Model 56 Ch. HS-158 (See Model 58 S8A11) 52 Ch. HS-160 (See Model 58 S8A11) 64 Ch. HS-160 (See Model 58 S8A11) 64 Ch. HS-168 (See Model 58 S8C11 64 Ch. HS-178 (See Model 59 58 Ge Model 97 79 Ch. HS-178 (See Model 59 58 Ge Model 59 77 Ch. HS-184 (See Model 59 58 Ge Model 59 77 Ch. HS-184 (See Model 59 59 Ch. HS-184 (See Model 59 59 Ch. HS-184 (See Model 59 59 Ch. HS-123 (See Model 59 68 Ch. HS-223 (See Model	Ch. HS-124 (See Model 68F11)	58
Ch. HS-132 (See Model 59 78FW21) 54 88FW21) 54 88FW21) 54 RASIM 21) 54 88FW21) 54 Ch. HS-137 (See Model 54 VK101) 51 Ch. HS-137 (See Model 56 Ch. HS-150 (See Model 56 Ch. HS-155 (See Model 56 Ch. HS-158 (See Model 58 S8A11) 52 Ch. HS-160 (See Model 58 S8A11) 64 Ch. HS-160 (See Model 58 S8A11) 64 Ch. HS-168 (See Model 58 S8C11 64 Ch. HS-178 (See Model 59 58 Ge Model 97 79 Ch. HS-178 (See Model 59 58 Ge Model 59 77 Ch. HS-184 (See Model 59 58 Ge Model 59 77 Ch. HS-184 (See Model 59 59 Ch. HS-184 (See Model 59 59 Ch. HS-184 (See Model 59 59 Ch. HS-123 (See Model 59 68 Ch. HS-223 (See Model	Ch. HS-125 (See Model 58X11)	53
Ch. HS-132 (See Model 59 78FW21) 54 88FW21) 54 88FW21) 54 RASIM 21) 54 88FW21) 54 Ch. HS-137 (See Model 54 VK101) 51 Ch. HS-137 (See Model 56 Ch. HS-150 (See Model 56 Ch. HS-155 (See Model 56 Ch. HS-158 (See Model 58 S8A11) 52 Ch. HS-160 (See Model 58 S8A11) 64 Ch. HS-160 (See Model 58 S8A11) 64 Ch. HS-168 (See Model 58 S8C11 64 Ch. HS-178 (See Model 59 58 Ge Model 97 79 Ch. HS-178 (See Model 59 58 Ge Model 59 77 Ch. HS-184 (See Model 59 58 Ge Model 59 77 Ch. HS-184 (See Model 59 59 Ch. HS-184 (See Model 59 59 Ch. HS-184 (See Model 59 59 Ch. HS-123 (See Model 59 68 Ch. HS-223 (See Model	Ch. HS-127 (See Model 68X11)	56
Ch. HS-132 (See Model 59 78FW21) 54 88FW21) 54 88FW21) 54 RASIM 21) 54 88FW21) 54 Ch. HS-137 (See Model 54 VK101) 51 Ch. HS-137 (See Model 56 Ch. HS-150 (See Model 56 Ch. HS-155 (See Model 56 Ch. HS-158 (See Model 58 S8A11) 52 Ch. HS-160 (See Model 58 S8A11) 64 Ch. HS-160 (See Model 58 S8A11) 64 Ch. HS-168 (See Model 58 S8C11 64 Ch. HS-178 (See Model 59 58 Ge Model 97 79 Ch. HS-178 (See Model 59 58 Ge Model 59 77 Ch. HS-184 (See Model 59 58 Ge Model 59 77 Ch. HS-184 (See Model 59 59 Ch. HS-184 (See Model 59 59 Ch. HS-184 (See Model 59 59 Ch. HS-123 (See Model 59 68 Ch. HS-223 (See Model	68X11A)	56
0011/j 00 76712 56 76712 56 76712 56 76712 56 76712 56 76712 56 76712 56 76712 56 76 58 76 58 76 58 77 58 77 58 77 58 77 58 77 58 77 58 77 58 77 58 77 58 78 78 79 58 70 58 70 58 70 58 70 58 70 58 70 58 70 58 70 58 70 58 70 58 70 58 <tr< td=""><td>78FM22M) Ch. HS-132 (See Model</td><td>59</td></tr<>	78FM22M) Ch. HS-132 (See Model	59
0011/j 00 76712 56 76712 56 76712 56 76712 56 76712 56 76712 56 76712 56 76712 56 76 58 76 58 76 58 77 58 77 58 77 58 77 58 77 58 77 58 77 58 77 58 77 58 78 78 79 58 70 58 70 58 70 58 70 58 70 58 70 58 70 58 70 58 70 58 70 58 70 58 <tr< td=""><td>78FM21) Ch. HS-133 (See Model</td><td>59</td></tr<>	78FM21) Ch. HS-133 (See Model	59
0011/j 00 76712 56 76712 56 76712 56 76712 56 76712 56 76712 56 76712 56 76712 56 76 58 76 58 76 58 77 58 77 58 77 58 77 58 77 58 77 58 77 58 77 58 77 58 78 78 79 58 70 58 70 58 70 58 70 58 70 58 70 58 70 58 70 58 70 58 70 58 70 58 <tr< td=""><td>88FM21) Ch. HS-137 (See Model</td><td>54</td></tr<>	88FM21) Ch. HS-137 (See Model	54
0011/j 00 76712 56 76712 56 76712 56 76712 56 76712 56 76712 56 76712 56 76712 56 76 58 76 58 76 58 77 58 77 58 77 58 77 58 77 58 77 58 77 58 77 58 77 58 78 78 79 58 70 58 70 58 70 58 70 58 70 58 70 58 70 58 70 58 70 58 70 58 70 58 <tr< td=""><td>VK101) Ch. HS-144 (See Model</td><td>51</td></tr<>	VK101) Ch. HS-144 (See Model	51
SBA11)	68111) Ch. HS-150 (See Model	54
SBA11)	Ch. HS-155 (See Model 78F12M1	56
SigG(1) 64 Ch. HS-167 698.11	Ch. HS-158 (See Model 58A11)	52
Ch. HS-180 (See Model 59X11) 81 (See Model 59X11) 82 (h HS-181 (See Model 69X11) 82 (h HS-184 (See Model 491110) 77 Ch. HS-184 (See Model S8811A) 69 (See Model 40110) 78 (See Model 40110) 78 (See Model 40110) 78 (See Model 40110) 78 (See Model 59110) 78 (See Model 59110) 78 (See Model 59110) 78 (See Model 59110) 97 (See Model 59110) 97 (See Model 59110) 97 (See Model 59110) 100 (Ch. HS-224 (See Model 101 Ch. HS-224 (See Model 102 Ch. HS-224 (See Model 102 Ch. HS-242 (See Model 102 Ch. HS-249 (See Model 101 Ch. HS-239 (See Model 101) (See Model 1771) 121 Ch. HS-280 (See Model 101) Ch. HS-280 (See Model 101) (See Model 1771) 121	Ch. HS-160 (See Model 58G11)	64
Ch. HS-180 (See Model 59X11) 81 (See Model 59X11) 82 (h HS-181 (See Model 69X11) 82 (h HS-184 (See Model 491110) 77 Ch. HS-184 (See Model S8811A) 69 (See Model 40110) 78 (See Model 40110) 78 (See Model 40110) 78 (See Model 40110) 78 (See Model 59110) 78 (See Model 59110) 78 (See Model 59110) 78 (See Model 59110) 97 (See Model 59110) 97 (See Model 59110) 97 (See Model 59110) 100 (Ch. HS-224 (See Model 101 Ch. HS-224 (See Model 102 Ch. HS-224 (See Model 102 Ch. HS-242 (See Model 102 Ch. HS-249 (See Model 101 Ch. HS-239 (See Model 101) (See Model 1771) 121 Ch. HS-280 (See Model 101) Ch. HS-280 (See Model 101) (See Model 1771) 121	Ch. HS-167 (See Model 59R11)	79
Ch. HS-180 (See Model 59X11) 81 (See Model 59X11) 82 (h HS-181 (See Model 69X11) 82 (h HS-184 (See Model 491110) 77 Ch. HS-184 (See Model S8811A) 69 (See Model 40110) 78 (See Model 40110) 78 (See Model 40110) 78 (See Model 40110) 78 (See Model 59110) 78 (See Model 59110) 78 (See Model 59110) 78 (See Model 59110) 97 (See Model 59110) 97 (See Model 59110) 97 (See Model 59110) 100 (Ch. HS-224 (See Model 101 Ch. HS-224 (See Model 102 Ch. HS-224 (See Model 102 Ch. HS-242 (See Model 102 Ch. HS-249 (See Model 101 Ch. HS-239 (See Model 101) (See Model 1771) 121 Ch. HS-280 (See Model 101) Ch. HS-280 (See Model 101) (See Model 1771) 121	Ch. HS-168 (See Model 79XM21)	85
Ch. HS-180 (See Model 59X11) 81 (See Model 59X11) 82 (See Model 59X110) 77 Ch. HS-184 (See Model S8811A 69 (See Model 59X10) 98 (See Model 59X10) 98 (See Model 59X10) 97 (See Model 5Y10) 100 (Ch. HS-224 (See Model SM1) 102 (See Model 5X110) 111 Ch. HS-224 (See Model 102 Ch. HS-224 (See Model 102 Ch. HS-224 (See Model 103 (See Model 5X110) 114 (See Model 5X110) 114 (Ch. HS-249 (See Model SM1)	(See Model 99FM21.R) Ch. HS-175	80
Ch. HS-180 (See Model 59X11) 81 (See Model 59X11) 82 (See Model 59X110) 77 Ch. HS-184 (See Model S8811A 69 (See Model 59X10) 98 (See Model 59X10) 98 (See Model 59X10) 97 (See Model 5Y10) 100 (Ch. HS-224 (See Model SM1) 102 (See Model 5X110) 111 Ch. HS-224 (See Model 102 Ch. HS-224 (See Model 102 Ch. HS-224 (See Model 103 (See Model 5X110) 114 (See Model 5X110) 114 (Ch. HS-249 (See Model SM1)	(See Model 69L11) Ch. HS-178 (See	76
Ch. HS-181 (See Model 69X11)	Model 79FM21) Ch. HS-180	88 /
Ch. Gae Model 49(110)77 Ch. HS. 184 (Sce Model SBR11A) Ch. HS. 187 Ch. HS. 187 Ch. HS. 188 (Sce Model See Model 59(110)78 Ch. HS. 192 Ch. HS. 192 Ch. HS. 192 See Model 59(21U)98 Ch. HS. 210 See Model 59(110)97 Ch. HS. 223 (Sce Model Sh11) Ch. HS. 223 (Sce Model Gee Model 511)100 Ch. HS. 224 (Sce Model GSee Model 19F1).111 Ch. HS. 228 (See Model 19F1).111 Ch. HS. 228 (See Model 19F1).111 Ch. HS. 228 (See Model 5110)102 Ch. HS. 228 (See Model 5110)115 Ch. HS. 228 (See Model 5110)111 Ch. HS. 228 (See Model 5110)112 Ch. HS. 224 (See Model 5110)112 Ch. HS. 242 (See Model 5110)112 Ch. HS. 249 (See Model (See Model 511)101 (See Model 511)101	(See Model 59X11) Ch. HS-181	
Ch. VIIS. 147 (See Model 59111Q) 78 (See Model 59111Q) 78 (See Model 59221U) 68 Ch. HS. 12 68 Ch. HS. 210 68 Ch. HS. 221 68 Ch. HS. 223 101 Ch. HS. 223 101 Ch. HS. 224 101 Ch. HS. 225 102 Ch. HS. 224 101 Ch. HS. 225 102 Ch. HS. 226 102 Ch. HS. 224 102 Ch. HS. 224 <t< td=""><td>(See Model 491110)</td><td></td></t<>	(See Model 491110)	
Ch. HS-187 (See Model 59111Q)78 (See Model 59111Q)78 (She Model 59211)98 (See Model 59211)97 (See Model 59211)97 (See Model 5911)101 (See Model 5911)101 (See Model 5911)102 (See Model 5911)102 (See Model 511)102 (See Model 511)102 (See Model 511)102 (See Model 511)102 (See Model 511)102 (See Model 511)111 (She HS-224 (See Model 1611)111 (She Model 511)112 (See Model 51111)113 (See Model 51111)114 (She Model 51111)114 (She Model 51111)114 (She Model 51111)112 (She Model 51111)112 (She Model 5111)112 (She Model 5111)112 (She Model 511)112 (She Model 511)112	Ch. HS-184 (See Model 58R11A)	69
Ch. n3-172 [See Model 59X21U)	Ch. HS-187 (See Model 59L11Q)	78
Ch. n3-172 [See Model 59X21U)	Ch. HS-188 (See Model 59F11)	68
(See Model S9H1U) 97 Ch. HS-223 (See Model SM1] 101 Ch. HS-224 (See Model (See Model S1). 100 Ch. HS-226 (See Model (See Model S1). 102 Ch. HS-228 (See Model 116 HS-230 (See Model 19F1). 111 Ch. HS-2424 (See Model 102 Ch. HS-2434 (See Model 101 Ch. HS-2434 (See Model 111 Ch. HS-2434 (See Model 111 Ch. HS-246 (See Model 111 Ch. HS-247 (See Model 111 Ch. HS-248 (See Model 111 Ch. HS-249 (See Model 114 Ch. HS-249 (See Model 111 Ch. HS-249 (See Model 111 Ch. HS-249 (See Model 111 Ch. HS-239 (See Model 111 Ch. HS-239 (See Model 111 Ch. HS-239 (See Model 111 Ch. HS-238 (See Model 111) 121	(See Model 59X21U)	98
(bee Model Spl) 100 (Ch. HS.228 (See Model 102) 101 (See Model Sci) 116 HS.231 (See Model 19F1) 111 (Ch. HS.242 (See Model 19F1) 102 (A. HS.234 (See Model 19F1) 102 (A. HS.243 (See Model 19F1) 102 (A. HS.243 (See Model 19F1) 102 (A. HS.243 (See Model 5X110) 115 (Ch. HS.246 (See Model 5X110) 114 (Ch. HS.247 (See Model 5X110) 112 (Ch. HS.249 (See Model 5M1) 114 (Ch. HS.247 (See Model 5M1) 101 (Ch. HS.249 (See Model 5M1) 101 (Ch. HS.249 (See Model 5M1) 101 (Ch. HS.249 (See Model 5M1) 101 (See Model 5T1) 101 (See Model 5T1) 101 (See Model 5T1) 102 (See Model 5T1) 101 (See Model 5C1) 101	(See Model 59H11U)	97
(bee Model Spl) 100 (Ch. HS.228 (See Model 102) 101 (See Model Sci) 116 HS.231 (See Model 19F1) 111 (Ch. HS.242 (See Model 19F1) 102 (A. HS.234 (See Model 19F1) 102 (A. HS.243 (See Model 19F1) 102 (A. HS.243 (See Model 19F1) 102 (A. HS.243 (See Model 5X110) 115 (Ch. HS.246 (See Model 5X110) 114 (Ch. HS.247 (See Model 5X110) 112 (Ch. HS.249 (See Model 5M1) 101 (Ch. HS.250 (See Model 5M1) 101 (See Model 5M1) 101 (See Model 5M1) 101 (See Model 5M1) 100 (See Model 5M1) 101 (See Model 5C1) 101 (See Model 5C1) 101	5M1) Ch. HS-224	101
HS-230 Glober Model 19F1) 111 Ch. HS-234 [See Model 102 1611	(See Model 5J1) Ch. HS-226 (See Model	100
HS-230 Glober Model 19F1) 111 Ch. HS-234 [See Model 102 1611	6L1) Ch. HS-228	102
Ch. Sae Akadet SR1[U]115 CH. Sae Madet SR1[U]114 Ch. Sae Madet SR1[U]114 Ch. Sae Madet SR1[U]114 Ch. HS-246 SR1[U]117 Ch. HS-246 SR1[U]112 Ch. HS-246 SR2[]114 Ch. HS-247 (See Model SFM2]]114 Ch. HS-249 (See Model SFM2]]101 Ch. HS-249 (See Model SFM2]]101 Ch. HS-249 (See Model SFM2]]100 Ch. HS-239 (See Model SFM2]]100 Ch. HS-230 (See Model SFM2]]121 Ch. HS-230 (See Model SFM2]]121 Ch. HS-230 (See Model SFM2]]121 Ch. HS-230 (See Model SFM2)121	HS-230 (See Model 19F1).	iii
Ch. HS-243 (See Model 5X11U)114 Ch. HS-244 (See Model 5H11U)117 Ch. HS-245 (See Model 5H11U)112 Ch. HS-246 (See Model 6X11U)114 Ch. HS-247 (See Model 8FM21)114 Ch. HS-249 (See Model 5M1)101 Ch. HS-250 (See Model 51)100 Ch. HS-253 (See Model 17F1)121 Ch. HS-253 (See Model 17F1)121 Ch. HS-258 (See Model 5C1)116 Ch. HS-258 (See Model 5C1)116	16F1) Ch. HS-242	102
(See Model SX110)114 (Ch. HS-244 (See Model SH110)117 (Ch. HS-245 (See Model SX110)112 (Ch. HS-245 (See Model GFM21)114 (Ch. HS-246 (See Model GFM21)114 (Ch. HS-247 (See Model GFM21)121 (Ch. HS-249 (See Model GFM21)	CL US 242	115
Ch. HS-243 (See Model 6X11U)112 Ch. HS-240 (See Model 9FM21)114 Ch. HS-247 (See Model Model 8FM21)101 Ch. HS-250 (See Model 5J1)100 Ch. HS-253 (See Model 17F1)121 Ch. HS-253 (See Model 5C1)116 (See Model 5C1)116	(See Model 5X11U) Ch. HS-244	114
Ch. HS-246 [See Model 9FM21]114 Ch. HS-247 (See Model SM1]101 Ch. HS-249 (See Model SM1]101 Ch. HS-250 (See Model 5J1)100 Ch. HS-253 (See Model 17F1)121 Ch. HS-258 [See Model SC1]116	(See Model 5HI1U) Ch. HS-245	117
Smil/ Comparison Ch. H5-250 (See Model 5J1)	Ch HS.246	
Smil/ Comparison Ch. H5-250 (See Model 5J1)	Ch. HS-247 (See Model 8FM21)	121
(See Model 5J1)	Ch. HS-249 (See Model 5M1)	101
(See Model 5C1)116	Ch. HS-250 (See Model 5J1)	100
(See Model 5C1)116	Ch. HS-253 (See Model 17F1)	121
{See Model 5X21U}120 Ch. HS-261 (See Model 17F5)121	(See Model 5C1)	116
Model 17F5) 121	(See Model 5X21U) Ch. HS-261 (See	120
	Model 17F5)	121

MOTOROLA-Cont. Ch. HS-262 (See Model 5C1).....116 Ch. HS-264 (See Model 6F11).....117 h. HS-265 Ch n. 15-52 (See Model 16K2)..... 93A h. TS-53 121 134—8 46—16 106—10 Ch. 1A Ch. 8A Ch. 10A
 Ch. 10A.
 106—10

 MUNTZ
 300 (Ch. TV-16A1)

 Tel. Rec.
 106—10

 MUNTZ
 301 (Ch. TV-16A1)

 Tel. Rec.
 106—10

 M31 (Ch. TV-16A2)
 Tel. Rec.

 Tel. Rec.
 108

 M31 (Ch. TV/16A2)
 Tel. Rec.

 Tel. Rec.
 116—10

 M31 (Ch. TV/7A2)
 Tel. Rec.

 Rec (See Model M31).
 116

 M32 (Ch. TV/7A2)
 Tel. Rec.

 M32 (Ch. TV/7A2)
 116

 M32 (Ch. TV/7A3)
 116

 M33 (Ch. TV/7A3)
 116

 M41 (M42 (Ch. TV/7A3)
 116

 M45 (Ch. MUNTZ MURPHY 2-1 2-2 2 113 122 (See Model 112).... MUSITRON PT-10 PX SRC-3 (See Model 101). 101 "Piccolo" 103 "Piccolo" 105 202 15-20 13 13 13 15 21 21 21 21 21 27

NASH	OLYMI 7-728
6MN082	7-728 7-925, 7-939
6MN082) 9	7-939 8-451
NATIONAL CO.	8-533V
HFS 62—14 HR0-7R, HR0-7T 50—12 HR0-50 112—7 NC-TV7, NC-TV7M, NC-TV7W Tel, Rec. 67—14	8-618 8-925, 752, 7
HRO-7R, HRO-7T 50—12 HRO-50	752, 7
HR0-50	Tel. 755, 75
NC-TV7W Tel, Rec 67-14 NC-TV-10C, T, W Tel, Rec. 94-5	(See
NC-TV-10C, T, W Tel. Rec. Prod Chae Bul 1	764, 76 (See
NC-TV-12C, W Tel. Rec.	766 Tel
(See Model NC-TV-10C) 94 NC-TV-12C W Tel. Rec.	(See 767 Te
Prod. Chge Bul. 1 103-19	(See
NC-TV-1001 Tel. Rec. (See Model NC-TV-10C) 94	OPERA
NC-TV-1001 Tel. Rec.	1A30 . 1A35 .
Prod. Chge. Bul. 1 103—19 NC-TV-1025 Tel. Rec.	1A45 .
(See Model NC-TV-10C) 94 NC-TV-1025 Tel. Rec.	1A65 1A70-A
Prod. Chge., Bul. 1 103—19	1A140
Prod. Chge. Bul. 1	4A25-E 4A30-A
(See Model NC-TV-10C) 94	4A30-A 4A35, 4A50-A
NC-TV-1201 Tel Rec	Mode
NC-TV-1225, NC-TV-1226	4M25C 11A55
	530, 53 Sol
(See Model NC-TV-10C) 94 NC-TV-1225, NC-TV-1226 Tel. Rec. Prod. Chge.	
Tel. Rec. Prod. Chge. Bul. 1	ORTHO (See
NC-2-40DR NC-2-40DT 41-16	PACK
	PA-382
NC 57 A8_14	PA-393
NC-108R, NC-108T 47—15 NC-173R, NC-173T 40—13	PACK/ C1362
NC-108R, NC-108T 47-15 NC-173R, NC-173T 40-13 NC-183R, NC-183T 49-15 TV-1201 Tel. Rec 119-10 TV-1226 Tel. Rec	C1461
TV-1226 Tel. Rec.	5DA
(See Model IV-1201). ITY	5FP
(See Model TV-1201) 119	100
(See Model TV-1201)119 TV-1625 Tel. Rec. (See Model TV-1201)119	261 471 551 551-D
NATIONAL UNION	551 551-D
G-613 "Commuter" 19—23 G-619	561 563 (S
G-619 11-35 571, 571A, 571B 17-22	566 (S
NEWCOMB	568 571 (S
H-10 14-20	572 .
H-14 KX-30	581 (S 651
NIELSON	661
1018 Tel. Rec	662 673A, 682
1018 Tel. Rec	682 771
NOBLITT SPARKS (See Arvin)	771 861 872
OLDSMOBILE	872 880, 8
982375	Mode
	881-A, 882
982421	884, 8 1052,
982454	1052, 1054B
982544, 982573 96-7	1063
OLYMPIC	1181, 1272
DX-214, DX-215, DX-216 Tel. Rec	1273
DX-619, DX-620, DX-621, DX-622 Tel. Rec. (See	1291TV 1472
Model DX 214)	
Model DX 214)	2001TV 2091,
Model DX-214)106	2101.
	2105, (See
Model DX-214)	(See 2202,
TV 104 TV 107 TV 109	(See 22911\
Tel. Rec. (See Model	2294
Tel. Rec. (See Model TV-104)	2294 2294 2297-T
4See Model TV-104) 67	Stan (See
TV-922 Television Receiver 58—14 TV-9221 Tel. Rec. See Model TV-104)	2298 T
TV-944, TV-945, TV-946	(See 2301-T
Tel. Rec. (See Model TV-104)	2301-T 2601-T 2602 1
TV-947 Tel. Rec 85-10	(See 2692-T
Model TV-104) 67	(See
TV-949, TV-950 Tel. Rec. (See Model TV-947) 85	2801-T
XL-210, XL-211 Tel. Rec. 109-8	Tel. Mod
XL-612, XL-613 Tel Rec (See	2803T
Tel. Rec. (See Model TV-104)	2991TV 3191, 3381 1 4580 1
	4580 1
6-503 4-10 6-501V-U (See Model 6-501W-U) 3	46911
6-501V-U (See Model 6-501W-U)	PHILO
6-504, 6-504L	C-4608 802)
6-604 Series 22-21 6-604V-11th, 6-604V-220,	C-4608
6-604V-110, 6-604V-220, 6-604W-110, 6-604W- 150, 6-604W-220 (See	C-4908
150 4 #04W/ 220 /See	Mod
Model 6 604 Sector) 33	L CP.2
Mode! 6-604 Series} 22	CR-2 CR-4,
6-606 4-36 6-606-A 11-17	CR-2 CR-4, CR-8
Model 6.604 Series) 22 6-606 4-36 6	CR-2 CR-4, CR-8 CR-9 CR-12
Model 6-606 4-36 6-606 4-36 6-606 6-606-A 1117 6-606-U 1118 6-617 4-7 6-617 4-7	CR-2 CR-4, CR-8 CR-9 CR-12 CR-50 CR-50
Model 6-606 4-36 6-606 4-36 6-606 6-606-A 1117 6-606-U 1118 6-617 4-7 6-617 4-7	CR-2 CR-4, CR-8 CR-9 CR-12 CR-50 CR-50 CR-50 P-463
Model 6-606 4-36 6-606 4-36 6-606 6-606-A 1117 6-606-U 1118 6-617 4-7 6-617 4-7	CR-2 CR-4, CR-8 CR-9 CR-12 CR-50 CR-50 CR-50 P-4635 Mod P-4735
Model 6-604 Series) 12 6-606 436 6-606 117 6-607 1118 6-617 47 6-617 47 7.421V, 7.421W, 7.421X, 5713 7.435V 7.435V, 7.435W 3413 7.526 3021 7.532W, 7.532V 3215 7.532, 738 3414	CR-2 CR-4, CR-8 CR-9 CR-12 CR-50 CR-50 P-4635 Mod P-4735 Mod
Model 6-604 Series) 12 6-606 436 6-606 1117 6-607 1118 6-617 47 6-617 47 7-617U (Sire Model 6-617) 4 7-421V, 7-421W, 7-421X, 5713 7.435W 7-326 3413 7-532W 3215	CR-2 CR-4, CR-8 CR-9 CR-12 CR-50 CR-50 CR-50 P-4635 Mod P-4735

OLYMPIC-Cont.	PHILCO
7-728 (See Model 7-724) 2 7-925, 7-934, 7-936, 7-939 3	deboke
0 /51 A	P_15 debok
8-533V, 8-533W	7-14 UN6-100 5-16 UN6-400 5-19 UN6-450
8-618 3 8-925, 8-934, 8-936 4 752, 752U, 753, 753U,	11114 500
755. 755U Tel. Rec.	40-131
(See Model 752)	46-132
(See Model 752) 120 766 Tel. Rec.	
(See Model 752)11 767 Tel. Rec. (See Model 752)12	46-203
OPERADIO	46-250,
1A30	3-15 40-420,
1A45	8-16 46-427 2-14 46-480
1A70-A	7-16 46-1201 6-17 46-1201
4A25-E	1-8 46-1201 2-9 46-1203 46-1209
IA05 > IA70.A 4 IA140 4 4A25:E 10 4A30.A 10 4A35.A 10 4A50.A 4A51.A (See Model Model 4A30.A 4A50.A 20	0-9 46-1213 46-1226
Model 4A30-A) 10 4M25C 9 11A55 11	0 48-1203 46-1213 46-1213 2 46-1226 9 11 47-1227 47-1227 3 6 47.1230 7 -14 48-141, 50
	3-6 47-1230 48-141,
"Soundcaster" 3 ORTHOSONIC	7-14 48-150 48-200, 48-206
(See Electronic Labs.)	48-214 (
PA-382042 2	0-26 48-225, 48-250
PACKARD-BELL	48-360
C1362 1 C1461	2-21 48-460, 2-22 48-461 6-29 48-464
5DA	4-15 48-472,
5FP	1-29 48-472 3-16 48-475 1 29 48-482
261	0-22 48-485
	2 48-1000 7 35 Tel.
561	48-1000
568	9-24 48-1000 Tel
572	2-22 48-1001
651	4-42 (See / Code
	3-22 48-1050
682 5 771	A 1/ Model
861	4-16 4-16 7-23 48-1201 7-23 48-1253 48-1253
880, 880A (See	6 48-1260
	7-17 48-1262 4-6 48-1263
884, 892	826 48-1264
1003	3-23 48-1266 8-25 48-1270 48-1274
1181, 1181A	5-12 48-1282 (See
1091 Tel. Rec	5-12 (See) 6-19 48-1284 48-1286
14/2	8-17 48-1290 48-2500 8-8 Rec. (
2001TV 2002TV Tel. Rec., 9	* 122)
	310 49-101 49-500,
(See Model 2101)12 2202, 2204 Tel. Rec.	3 49-501, 49-503
(See Model 2101) 12 2291TV, 2292TV, 2293TV,	3 49-504, 49-505
2005, 2105 Fe1, Rec. (105, 2105 Fe1, Rec.	2-10 49-506 49-601
Standard Tel. Rec.	49-602
2298-TV Tel. Rec. (See Model 2291-TV) 8	49-605, 49-900-
2301-TV Tel. Rec	6-9 49-901 2-6 49-902
(See Model 2291-TV) & 2 2301-TV Tel. Rec	3 49-905
2692-TV Tel. Rec. (See Model 2601-TV) .12	
2801-TV, 2801A-TV Tel. Rec. (See	49-1002 Tel. F
Model 2301-TV)	9 49-1040 Rec.
2801-TV, 2801A-TV Tel. Rec. (See Model 2301-TV)	49-10
3191, 3192 Tel. Rec 3381 Tel. Rec 4580 Tel. Rec 4691TV Tel. Rec	Tel. 1 49-1075
PHILCO	122) 49-1076 Tet. F
C-4608 (See Mopar Model 802)	18 (See 49-1076
802] C-4608 (Revised) (See Mo- par Model 802 Revised) 4 C-4908 (See Mopar	12 49-10/6 12 49-10 Tel. F
Model 605)	(See
CR-2 CR-4, CR-6	33-17 48-48
CP 0	49-90
CR-503	28-10 123)
P-4635 (See Pockard	Tel. F
P.4735 (See Packard	20 (See 49-1175 57 123)
PD-4908 (See Mopor	Mode So Code
Model 803)	- Code

4676	—Cont. S-4625 (See Stu- ar Model S-4624). S-4627 (See Stu-	21
debok	er Model S-46761.	19
N6-100		1926 3023
N6-450 N6-500	Revised)	18—26 17—24
N6-550		31-24 5-13
6-131 (Revised)	32-16
6-132 6-142		4-20 36-16
200 1	46 201 46 202	1-24
46-203	(See Model Series) 46-250-1, 46-251	1
46-200	0 Series) 46-250-1, 46-251 46-420-1 46-421-1	2 12
5-350	46-420-1	10-24 6-22 5-12
6-421,	46-421-1	2-25
5-480		19-25 4-35
5-1201	(Revised)	20 21
5-1203		6-23 13-24 12-33
6-1213		
7-204,	(Revised) 47-205 48-145 48-200-1	33 —18 25 —22 22 —23 25 —23
7.1230	48-145	22-23
8-141, 8-150	48-145	
8-200, 8-206	48-200-1	
B-214 (See Model 48-200)	33
8-250,	48-250-1	32—17 37—17
8-360 8-360	48-200-1 See Model 48-200) 48-230 48-250-1	37-17
		29 15
8-464	48-472-1 (Revised)	26-20 43-15 48-18
8-472	(Revised)	48-18
8-4/5		30-24
8-485 8-700	Tel. Rec (Code 121)	47—19 68—13
8-1000	, 48-1000-5	
(Code 8-1000	48-1000-5 122) Tel. Rec (Code 125)	53-17
Tel. 1 8-1001	(Code 125) Rec. , 48-1001-5 (Code 122) Te1. Rec. Model 48-1000 122) . 48-1050-5 (Code	*
121 8	122) Tel. Rec.	
Code	122)	53
122)	Tel. Rec. (See	
8-1201		31-25 36-17
8-1255	(See Model 01)	34-18
8-1260 48-12	(See Model 01)	31 35—18
8-1262		35—18 32—18
8-1264		
8-1270	. 48-1276 , 48-1283 Model 48-1262)	36—18 39—15 42—20 41—17
8-1274	, 48-12/6	
(See 1 8-1284	Model 48-1262)	35
(See) 8-1284 8-1286 9-1200	Model 48-1262)	35 45-20 5115
8-1284 8-1286 8-1286		35
8-1284 8-1286 8-1290 8-2500 Rec. (122)	, 48-2500-5 Tel. Codes 121 and	35 45-20 51-15 47-18
8-1284 8-1286 8-1290 8-2500 Rec. (122) 9-101	, 48-2500-5 Tel. Codes 121 and	35 45-20 51-15 47-18
8-1284 8-1286 8-1290 8-2500 Rec. (122) 9-101 9-500,	48-2500-5 Tel. Codes 121 and 49-500-1 49-501-1	35 45-20 5115 4718 8910 878 4819 5618
8-1284 8-1286 8-2500 Rec. (122) 9-101 9-500, 9-501, 9-503 9-504,	, 48-2500-5 Tel. Codes 121 and 49-500-1 49-501-1	35 45_20 51_15 47_18 89_10 87_8 48_19 56_18 52_15 54_17
8-1284 8-1286 8-1286 8-2500 Rec. (122) 9-101 9-500, 9-501, 9-503 9-504, 9-505	, 48-2500-5 Tel. Codes 121 and 49-500-1 49-501-1 49-504-1	35 45-20 51-15 47-18 89-10 87-8 48-19 56-18 52-15 54-17 53-18 48
8-1284 8-1286 8-1290 8-2500 9-101 9-500, 9-501, 9-503 9-504, 9-505 9-5061	, 48-2500-5 Tel. Codes 121 and 49-500-1 49-501-1 49-504-1 See Model 49-500)	35 45-20 51-15 47-18 89-10 87-8 48-19 56-18 52-15 54-17 53-18 48
8-1284 8-1286 8-1290 8-2500 9-101 9-500, 9-501, 9-503 9-504, 9-505 9-506 9-601 9-602	, 48-2500-5 Tel. Codes 121 and 49-500-1 49-501-1 49-504-1 See Model 49-500)	35 45-20 5115 4718 8910 878 48-19 5618 5215 54-17 5318 48 4221 4118 5916
8-1284 8-1286 8-1290 8-2500 9-101 9-500, 9-501, 9-503 9-504, 9-505 9-5061 9-601 9-602	, 48-2500-5 Tel. Codes 121 and 49-500-1 49-501-1 49-504-1 See Model 49-500) 49-607 E 49-900-1	35 45-20 51-15 47-18 89-10 87-8 48-19 56-18 52-15 54-17 53-18 48 42-21 41-18 59-15 58-15 58-15 58-15
8-1284 8-1286 8-1290 8-2500 9-101 9-501, 9-503 9-505 9-505 9-505 9-506 9-601 9-602 9-603 9-605, 9-9001	, 48-2500-5 Tel. Codes 121 and 49-500-1 49-501-1 49-504-1 See Model 49-500) 49-607 E 49-900-1	35 45-20 51-15 47-18 89-10 87-8 48-19 56-18 52-15 54-17 53-18 48 42-21 41-18 59-15 58-15 58-15 56-19
8-1284 8-1286 8-1290 8-2500 9-101 9-501, 9-503 9-504, 9-505 9-506 9-506 9-601 9-602 9-603 9-605, 9-900- 9-902	, 48-2500-5 Tel. Codes 121 and 49-500-1 49-501-1 49-504-1 (See Model 49-500) 49-607 E, 49-900-1	35 45-20 5115 4718 8910 878 8910 878 4819 5618 5215 5417 5318 48 4221 4118 5915 5815 5417 5
8-1284 8-1286 8-1290 8-2500 9-500, 9-500, 9-503 9-504, 9-505 9-506 9-505 9-601 9-603 9-603 9-603, 9-900- 9-900 9-901 9-902 9-905	, 48-2500-5 Tel. (Codes 121 and 49-500-1 49-501-1 49-504-1 (See Model 49-500) 49-607 E, 49-900-1	35 45-20 51-15 47-18 89-10 87-8 89-10 87-8 85-18 52-15 54-17 53-18 48 42-21 41-18 59-15 58-15 58-15 58-15 58-16 55-116 58-16 58-16
8-1284 8-1286 8-2500 Rec. (122) 9-501, 9-503, 9-505, 9-504, 9-505 9-506 9-601 9-602 9-603 9-605, 9-900 9-901 9-902 9-905 9-905 9-906	, 48-2500-5 Tel. Codes 121 and 49-501-1 49-501-1 49-504-1 See Model 49-500) 49-607 E, 49-900-1	35 45-20 5115 4718 8910 878 8910 878 4819 5618 5215 5417 5318 48 4221 4118 5915 5815 5417 5
8-1284 8-1286 8-1290 Rec. (122) 9-501, 9-503, 9-503, 9-505 9-504, 9-505 9-504 9-601 9-602 9-603 9-605, 9-900 9-901 9-902 9-905 9-905 9-906 9-909	43-2500-5 Tel. Codes 121 and 49-500-1 49-501-1 49-504-1 See Model 49-500 49-607 E, 49-900-1	$\begin{array}{c} 35\\ 45-20\\ 51-15\\ 89-10\\ 87-8\\ 48-19\\ 56-18\\ 87-8\\ 48-19\\ 56-18\\ 52-15\\ 55-18\\ 48\\ 42-21\\ 41-18\\ 59-15\\ 58-15\\ 58-16\\ 55-19\\ 51-16\\ 55-16\\ 55-17\\ \end{array}$
8-1284 8-1286 8-1290 8-2500 9-101 9-501, 9-503 9-504, 9-505 9-505 9-505 9-601 9-601 9-603 9-603 9-603 9-603 9-900 9-901 9-902 9-905 9-905 9-906 9-905 9-906 9-905 9-906 9-905 9-906 9-905 9-1002 Tel. 60	, 43-2500-5 Tel. Codes 121 and 49-500-1 49-501-1 49-504-1 See Model 49-500 49-607 E, 49-900-1 E, 49-900-1	$\begin{array}{c} 35\\ 45-20\\ 51-15\\ 89-10\\ 87-8\\ 48-19\\ 56-18\\ 87-8\\ 48-19\\ 56-18\\ 52-15\\ 55-18\\ 48\\ 42-21\\ 41-18\\ 59-15\\ 58-15\\ 58-16\\ 55-19\\ 51-16\\ 55-16\\ 55-17\\ \end{array}$
8-1284 8-1286 8-1290 8-2500 9-101 9-501, 9-503 9-504, 9-505 9-505 9-505 9-601 9-601 9-603 9-603 9-603 9-603 9-900 9-901 9-902 9-905 9-905 9-906 9-905 9-906 9-905 9-906 9-905 9-906 9-905 9-1002 Tel. 60	, 43-2500-5 Tel. Codes 121 and 49-500-1 49-501-1 49-504-1 See Model 49-500 49-607 E, 49-900-1 E, 49-900-1	35 45 - 20 51 - 15 47 - 18 89 - 10 87 - 8 48 - 19 56 - 18 56 - 18 56 - 18 57 - 15 58 - 15 57 - 16 55 - 17 91A - 10
8-1284 8-1260 8-2500 Rec. (122) 9-501, 9-503, 9-505, 9-506, 9-505, 9-506, 9-505, 9-506, 9-506, 9-601 9-602, 9-603, 9-900 9-900 9-902 9-904 9-909 9-904 9-909 9-902 9-904 9-905 9-900	, 48-2500-5 Tel. Codes 121 and 49-500-1 49-501-1 49-504-1 (See Model 49-500) 49-607 E. 49-900-1 (Code 121) (Code 121) (Code 121) (Code 121) (Code 121) (Code 121) (Code 121) (Code 121) (Code 121)	$\begin{array}{c} 35\\ 45-20\\ 51-15\\ 47-18\\ 89-10\\ 87-8\\ 48-19\\ 56-18\\ 56-18\\ 55-15\\ 54-17\\ 53-18\\ 42-21\\ 41-18\\ 58-15\\ 58-15\\ 58-15\\ 58-16\\ 58-16\\ 58-16\\ 58-16\\ 58-16\\ 58-16\\ 55-17\\ 91A-10\\ 91A \end{array}$
8-1284 8-1286 8-1290 Rec. (122) 9-101 9-500, 9-505, 9-506, 9-505, 9-506, 9-505, 9-506, 9-505, 9-506, 9-503, 9-603, 9-603, 9-603, 9-900 9-902, 9-904, 9-902, 9-904, 9-905, 9-900, 9-905,	, 48-2500-5 Tel. Codes 121 and 49-500-1 49-501-1 49-504-1 See Model 49-500) 49-607 E. 49-900-1 (Code 121) tec. (Code 121) Tel. (See Model 123) tec. (Code 123) tec. (Code 123) tec.	$\begin{array}{c} 35\\ 45-20\\ 51-15\\ 47-18\\ 89-10\\ 87-8\\ 48-19\\ 56-18\\ 56-18\\ 55-15\\ 54-17\\ 53-18\\ 42-21\\ 41-18\\ 58-15\\ 58-15\\ 58-15\\ 58-16\\ 58-16\\ 58-16\\ 58-16\\ 58-16\\ 58-16\\ 55-17\\ 91A-10\\ 91A \end{array}$
8-1284 8-1286 8-1290 8-2500, 9-501, 9-500, 9-501, 9-504, 9-505 9-506 9-505 9-506 9-505 9-506 9-505 9-506 9-505 9-506 9-602 9-905 9-9	, 48-2500-5 Tel. (Codes 121 and 49-500-1 49-501-1 49-504-1 (See Model 49-500) 49-607 E, 49-900-1 (Code 121) tec. (Code 121) Tel. (Code 123) tec. (Code 123) tec. (Code 121) and Tel. Rec. (Code 122)	$\begin{array}{c} 35\\ 45-20\\ 51-15\\ 47-18\\ 89-10\\ 87-8\\ 48-19\\ 56-18\\ 56-18\\ 55-15\\ 54-17\\ 53-18\\ 42-21\\ 41-18\\ 58-15\\ 58-15\\ 58-15\\ 58-16\\ 58-16\\ 58-16\\ 58-16\\ 58-16\\ 58-16\\ 55-17\\ 91A-10\\ 91A \end{array}$
8-1284 8-1284 8-1280 8-2500 9-501, 9-501, 9-503, 9-504, 9-505 9-506 9-505 9-506 9-505 9-506 9-505 9-506 9-505 9-506 9-900 9-900 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-90	, 48-2500-5 Tel. Codes 121 and 49-500-1 49-501-1 49-504-1 (See Model 49-500) 49-607 E. 49-900-1 (Code 121) tec. (Code 121) Tel. (See Model 123) Tel. (Code 124) (Code 124) (Code 125) (Code 125)	35 45-20 5115 4718 8910 878 48-19 56-18 55-15 54-17 53-18 42-21 4118 58-15 58-15 58-15 58-15 58-15 58-16 55-17 91A-10 91A 92-5 93A-11
8-1284 8-1284 8-1280 8-2500 9-501, 9-501, 9-503, 9-504, 9-505 9-506 9-505 9-506 9-505 9-506 9-505 9-506 9-505 9-506 9-900 9-900 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-90	, 48-2500-5 Tel. Codes 121 and 49-500-1 49-501-1 49-504-1 (See Model 49-500) 49-607 E. 49-900-1 (Code 121) tec. (Code 121) Tel. (See Model 123) Tel. (Code 124) (Code 124) (Code 125) (Code 125)	35 45-20 5115 4718 8910 878 48-19 56-18 55-15 54-17 53-18 42-21 4118 58-15 58-15 58-15 58-15 58-15 58-16 55-17 91A-10 91A 92-5 93A-11
8-1284 8-1284 8-1280 8-2500 9-501, 9-501, 9-503, 9-504, 9-505 9-506 9-505 9-506 9-505 9-506 9-505 9-506 9-505 9-506 9-900 9-900 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-902 9-904 9-905 9-90	, 48-2500-5 Tel. Codes 121 and 49-500-1 49-501-1 49-504-1 (See Model 49-500) 49-607 E. 49-900-1 (Code 121) tec. (Code 121) Tel. (See Model 123) Tel. (Code 124) (Code 124) (Code 125) (Code 125)	35 45-20 5115 4718 8910 878 48-19 56-18 55-15 54-17 53-18 42-21 4118 58-15 58-15 58-15 58-15 58-15 58-16 55-17 91A-10 91A 92-5 93A-11
8-1284 8-1286 8-1260 8-2500 9-501, 9-501, 9-502, 9-504 9-501, 9-502, 9-504 9-702, 9-7	43-2500-5 Tel. Codes 121 and 49-500-1 49-501-1 49-504-1 See Model 49-500) 49-607 E. 49-900-1 (Code 121) Ec. (Code 121) Tel. (See Model 1021 (Code 122) Tel. (See Model 1021 (Code 122) Tel. (Code 123) Tel. (Code 124) Tel. (Code 124) Tel. (Code 125) Tel. (Code 125) Tel. (Code 127) Tel. (Code 127) Tel. (Code 128) Tel. (Code 128) Tel. (Code 129) Tel. (Code 129) (Code 129)	35 45-20 5115 4718 8910 878 48-19 5618 5215 54-17 5318 44 42-21 4118 55-15 58-15 58-15 58-15 58-15 58-16 55-17 91A-10 91A 92-5 93A-11 93A
8,1224 8,1266 8,1260 8,2500 9,101 9,501, 9,501, 9,501 9,501 9,502 9,504 9,502 9,504 9,503	, 43-2500-5 Tel. (Codes 121 and 49-500-1 49-501-1 49-501-1 49-504-1 (See Model 49-500) (Code 121) (Ec. (Code 121) (Code 121) (Code 121) (Code 122) (Code 122) (C	35 45-20 5115 4718 8910 878 48-19 5618 5215 54-17 5318 44 42-21 4118 55-15 58-15 58-15 58-15 58-15 58-16 55-17 91A-10 91A 92-5 93A-11 93A
8-1284 8-1286 8-1260 8-2500 9-501, 9-501, 9-502, 9-504 9-502 9-504 9-502 9-504 9-502 9-504 9-502 9-504 9-502 9-504 9-502 9-504 9-502 9-504 9-502 9-504 9-502 9-504 9-502 9-504 9-502 9-504 9-502 9-504 9-502 9-504 9-502 9-504 9-502	, 43-2500-5 Tel. (Codes 121 and 49-500-1 49-501-1 49-501-1 49-504-1 (See Model 49-500) (Code 121) (Ec. (Code 121) (Code 121) (Code 121) (Code 122) (Code 122) (C	35 45-20 5115 4718 8910 878 48-19 5618 5215 54-17 5318 44 4221 4418 55-15 58-15 58-15 58-15 58-15 58-15 58-15 58-15 58-15 58-15 58-15 58-15 58-16 57-16 55-17 91A-10 91A 92-5 93A-11 93A
8-1284 8-1286 8-1260 8-2500 9-501, 9-501, 9-502, 9-504 9-502 9-504 9-502 9-504 9-502 9-504 9-502 9-504 9-502 9-504 9-502 9-504 9-502 9-504 9-502 9-504 9-502 9-504 9-502 9-504 9-502 9-504 9-502 9-504 9-502 9-504 9-502 9-504 9-502	, 43-2500-5 Tel. (Codes 121 and 49-500-1 49-501-1 49-501-1 49-504-1 (See Model 49-500) (Code 121) (Ec. (Code 121) (Code 121) (Code 121) (Code 122) (Code 122) (C	35 45-20 5115 4718 8910 878 48-19 5618 5215 54-17 5318 44 4221 4418 55-15 58-15 58-15 58-15 58-15 58-15 58-15 58-15 58-15 58-15 58-15 58-15 58-16 57-16 55-17 91A-10 91A 92-5 93A-11 93A
8-1284 8-1280 8-2500 9-500, 9-501, 9-501, 9-504, 9-501, 9-504, 9-505 9-504 9-505 9-504 9-505 9-504 9-505 9-504 9-505 9-504 9-505 9-504 9-505 9-602 9-602 9-602 9-602 9-602 9-909 9-904 9-905 9-905 9-904 9-905 9-9	, 48-2500-5 Tel. Codes 121 and 49-500-1 49-501-1 49-501-1 (See Model 49-500) 49-607 E, 49-900-1 (Code 121) (Ec. (Code 121) (Code 121) (Code 121) (Code 121) (Code 123) (Code 122) (Code 123) (Code 123) (Code 122) (Code 123) (Code 123) (Cod	35 45-20 5115 4718 8910 878 48-19 56-18 52-15 54-17 53-18 42-21 4118 55-15 58-15 58-15 58-15 58-15 58-16 55-17 91A-10 91A 92-5 93A-11 93A 92 47 55
8.1284 8.1260 8.1260 8.1260 9.101 9.500, 9.600, 9.600, 9.600, 9.600, 9.600, 9.600, 9.600, 9.600, 9.600, 9.600, 9.600, 9.9	49-500-5 Tel. Codes 121 and 49-500-1 49-501-1 49-501-1 49-502-1 See Model 49-500) 49-607 E, 49-607 E,	35 45-20 5115 4718 8910 878 48-19 56-18 52-15 54-17 53-18 42-21 4118 55-15 58-15 58-15 58-15 58-15 58-16 55-17 91A-10 91A 92-5 93A-11 93A 92 47 55
8.1284 8.1260 8.1260 8.1260 9.101 9.500, 9.600, 9.600, 9.600, 9.600, 9.600, 9.600, 9.600, 9.600, 9.600, 9.600, 9.600, 9.9	, 48-2500-5 Tel. Codes 121 and 49-500-1 49-501-1 49-501-1 (See Model 49-500) 49-607 E, 49-900-1 (Code 121) (Ec. (Code 121) (Code 121) (Code 121) (Code 121) (Code 123) (Code 122) (Code 123) (Code 123) (Code 122) (Code 123) (Code 123) (Cod	$\begin{array}{c} 35 \\ 45 \\ -20 \\ 51 \\ -15 \\ 47 \\ -18 \\ 89 \\ -10 \\ 87 \\ -8 \\ 48 \\ -19 \\ 54 \\ -17 \\ -18 \\ 89 \\ -10 \\ 54 \\ -17 \\ -18 \\ 42 \\ -21 \\ 41 \\ -18 \\ 42 \\ -21 \\ 41 \\ -18 \\ 42 \\ -21 \\ 41 \\ -18 \\ 42 \\ -21 \\ -15 \\ -15 \\ -15 \\ -15 \\ -16 \\ 55 \\ -16 \\ 55 \\ -16 \\ 55 \\ -16 \\ 55 \\ -16 \\ 55 \\ -16 \\ 55 \\ -16 \\ 55 \\ -16 \\ 55 \\ -16 \\ 55 \\ -16 \\ 55 \\ -16 \\ 57 \\ -16 \\ 57 \\ -16 \\ 57 \\ -16 \\ 57 \\ -16 \\ 57 \\ -16 \\ 57 \\ -16 \\ -25$

PHILCO-Cont. 49-1175 (Codes 122, 124) Tel. Rec. (See Model 49-1040)... 92 49-1240 (Codes 121, 123) Tel. Rec. (See Model 49-1075)... 93A 49-1240 (Code 124) Tel. Rec. (See Model 49-1075)... 92 49-1275 (Code 121) Tel. Rec. (See Model 49-1075)... 93A 49-1278 (Code 122) Tel. Rec. (See Model 49-1075)... 93A 49-1278 (Code 122) Tel. Rec. See Model 49-1075)... 93A 49-1279 (Code 122) Tel. Rec. (See Model 49-1075)... 93A 49-1279 (Code 122) Tel. Rec. (See Model 49-1075)... 93A 49-1278 (Code 122) Tel. Rec. (See Model 49-1075)... 93A 49-1278 (Code 122) Tel. Rec. (See Model 49-1075)... 93A 49-1278 (Code 122) Tel. Rec. (See Model 49-1075)... 93A 49-1278 (Code 122) Tel. Rec. (See Model 49-1075)... 93A 49-1278 (Code 122) Tel. Rec. (See Model 49-1075)... 93A 49-1278 (Code 122) Tel. Rec. (See Model 49-1075)... 93A 49-1278 (Code 122) Tel. Rec. (See Model 49-1075)... 93A 49-1278 (Code 122) Tel. Rec. (See Model 49-1075)... 93A 49-1278 (Code 122) Tel. Rec. (See Model 49-1075)... 93A 49-1278 (Code 122) Tel. Rec. (See Model 49-1075)... 93A 49-1278 (Code 122) Tel. Rec. (See Model 49-1075)... 93A 49-1278 (Code 122) Tel. Rec. (See Model 49-1075)... 93A 49-1278 (Code 122) Tel. Rec. (See Model 49-1075)... 93A 49-1278 (Code 122) Tel. Rec. (See Model 49-1075)... 93A 49-1278 (Code 122) Tel. Rec. (See Model 49-1075)... 93A 49-1278 (Code 122) Tel. Rec. (See Model 49-1075)... 93A (See Model 49-1075) PHILCO-Cont. 3^{+} , 1200 (Code 121) Tel. Rec. [See Model 49-1040)... 4^{-} , 1404 (See Model 4^{-} , 1405 4^{-} , 1405 4^{-} , 1405 4^{-} , 1405 4^{-} , 1405 4^{-} , 1405 4^{-} , 1405 4^{-} , 1405 4^{-} , 1405 5^{-} , 123A or B, 123T A or 8^{+} , 1200 (Codes 121 A or 8^{+} , 1201 (See Model 49-1450) 77 9^{-} , 1600 4^{-} , 1600 4^{-} , 1600 4^{-} , 1600 4^{-} , 1600 4^{-} , 1600 4^{-} , 1600 4^{-} , 1600 4^{-} , 1600 4^{-} , 1600 4^{-} , 1600 5^{-} , 1600 5^{-} , 50-522, 50-522, 50-522, 50-524 5^{-} , 50-522, 50-522, 50-52 30 Teil, Rec. 94—7 50 T1478, 50 T1477, 94—7 50 T1478, 50 T1477, 128—11 50 T1478, 50 T1479, 128—11 50 T1481, 50 T1482 Teil, Rec. (See Model 80 T1481, 80 T1482 Teil, Rec. (See Model 50 T1481, 80 T1482 Teil, Rec. (See Model 50 T1484 Teil, Rec. 91A-10 50 T1600 Teil, Rec. 91A-10 50 T1600 Code 122) 110—10 50 T1600 Code 121) 91A 50 T1630 Teil, Rec. 99A-8 50 T1630 Code 121) 91A 50 T1630 Code 121) 10 51 P11207, 51 P11208 * 51 P11207, 51 P11208 * 51 P11207, 51 P11208 * 51 P11208, FIR, S1 T1443PM, 123—11 11 T1443P, 13 T1443PM, 123—11 51 P11208, FIR, S1 T1443PM, 123—11 11 T1604 (Code 121, 122) 16 Rec. Prod. Chye 804 20 <tr

NASH-PLYMOUTH

PHILCO-Cont.
PHILCO-Cont. 51-T1606 (Code 131) Tel. Rec. (See Model 50-T1600) 51-T1606 (Code 132) Tel. Par.
51-T1606 (Code 132) Tel. Rec. *
Tel. Rec. 51-T1607 (Code 121) Tel. Rec. (See Model 31-T1601 Code 121) 115 -1A 51-1607 (Code 122)
51-1607 (Code 122) Tet. Rec
51-T1634 (Codes 121, 122) Tel. Rec. (See Model
51-11607 (Code 122) Tel. Rec
Tel. Rec. (See Model 51-71601)115-1A
51-T1634 (Code 124) Tel. Rec
Tel. Rec. *
Tel. Rec. * 51-Ti 800 (Code 122) Tel. Rec. * 51-Ti 830 (Code 121)
lel. Rec.
51-11832 (Code 121) Tel Rec. *
51-T1833 (Code 121) Tel. Rec. * 51-T1834 (Code 121)
51-T1835 (Code 121)
51-T1836 (Code 123)
51-T1870 (Code 121) Tel. Rec. *
1e1. kec. 51.71870 (Code 121) Te1. Rec. * 51.1871 (Code 121) Te1. Rec. * 51.71871 (Code 122)
51-11871 (Code 122) Tel. Rec
Tel. Rec.
51-T1872 (Code 122) Tel. Rec
Tel. Rec
51-T1876 (Code 121)
51-12102 (Code 122) Tel. Rec. 132-10 51-12130 (Code 121)
Tel. Rec. [See Model
51-72102)
Tel Rec
(See Model 51-T2102)132
(See Model 51-72102), 132
Tel, Rec. (See Model 51-T2102) 132 51-T2170 (Code 121)
51-T2170 (Code 121) Tel. Rec. (See Model 51 T2102) 132
51-T2175, 51-T2176 (Code 124) Tel. Rec.
Tel. Rec. (See Model 51.T2102)132 51.T2175, 51.T2175 (Code 124) Tel. Rec. (See Model 51.T2102)132 51.530
51-532 (See Model 51-530)
51 527 51 5271 126-10
51-631 106-12 51-934 102-10 51-1330 130-11 51-1731 124-7
51-1731, 51-1732 124-7 PHILHARMONIC
100C
100T 33—20 149-C, 249-C 55—19 349-C 58—17
1001 33-20 149-C, 249-C 55-19 349-C 58-17 810, 8701, 8702, 8703, 871, 8712 18-27 (Ch. RR14) 18-27 Ch. RR14 (See Model 6810) 18
Ch. RR14 (See Model 6810) 18
PHILLIPS 66 (See Woolaroc) 3-62A (See Woolaroc Model 3-71A)
3-81A
CP-731D Tel. Rec
PHONOLA K-92, K-104 51—17 K-105 79—11
K-92, K-104 51—17 K-105 79—11 K-202, K-263 55—20 TK-134 83—8 TK-234 108—9
PILOT
T-411-U
T510, T511 T-521 T-530 Series T-530 Series T-530 Series T-530 Series T-530 Series T2-24 T-601 "Pilotuner" T28-26
T-601 ''Pilotuner'' 28-26
TV-37 Tel. Rec
PLYMOUTH (See Mopar) PLYMOUTH
(Interstate Stores) 1010
1020 89—5

41

POLICALARM-RCA VICTOR

POLICALARM PR-8	
PR-8	R
	61
PONTIAC	61
984170	
984247 *	61
984248, 984249 * * 984273 *	61
984273 984296, 984570 95-4 PORTO BARADIO (Also See	61
Porto Products)	
PA-510 (9008-A), PB-520 (9008-B) 3316 PA-510, PB-520 (Revised). 4821	61
PA-510, PB-520 (Revised). 4821 PORTO PRODUCTS	6T
SR-600 (Ch 90404	01
''Smokerette'') (See Porto Baradio Model	
PA-510) 33 PREMIER	61
15LW	6T
PURITAN	77
501 (Ch. 5D15WG), 502 (Ch. 5D25WG) 45 501X (Ch. 5D15WG), 502X (Ch. 5D25WG) 426	71
501X (Ch. 5D15WG), 502X (Ch. 5D25WG) 4-26	71
503	71
504 (CL 4425)4(C) E 20	
506 (6D155W),	71
507 (6D25SW) 3-10 506X, 507X (See Model 506)	
308 (Code /A335W) 4-31	8B
509	88
RADIO APPARATUS CORP. (See Policalarm)	88
RCA VICTOR	88
AAPU-1 * A55 {Ch. RC-1087} 109—10 A106 {Ch. RC-622} 97—12 B1-A, B1-8, B1-C {Ch.	88
A106 (Ch. RC-622) 97-12 B1-A, B1-8, B1-C (Ch.	8F
A 105 (Ch. KC-522)	8P
KRS2U-1, KRS21-1) Tel. Rec. * B2-C, B2-F, B2-G, B2-H (Ch. KCS24-1, KRK1-1, KRS20-1, KRS21-1) Tel. Rec. *	
(Ch. KCS24-1, KRK1-1, KRS20-1, KRS21-1)	
Tel. Rec. * B3-A, B3-B	8R
B3-A, B3-B * B4-A, B4-B, B4-C * B5-A, B5-B *	8R
B-411 (Ch. RC1098) 132 —12 BX6 (Ch. RC-1082) 103 —13	8T
BX55 (Ch. RC-1088), BX57 (Ch. RC-1088A) 102-11	8T
MI-12224, MI-12224A 81-12 MI-12236, -A, -B, -C,	8T
B3.A, B3.B	
MI-12239, -A	8T
MI-12289, MI-12290 80—12 MI-12291, MI-12292,	
MI-12293, MI-12294 868 MI-12295	8T
(See Model Mi-12287). 89 Mi-12296, Mi-12298 (See Model Mi-12289). 80 Mi-12299	
(See Model MI-12289). 80 MI-12299	8T
(See Model MI-12287). 89 MI-13159	
MI-13167	8T
PPAU-1 RV151 (Ch. RK121C, RS-123D)	81
S1000 (Ch. KCS31-1, RC617B) Tel. Rec 91A-11 T100 (Ch. KCS-38)	.
leí, Kec	81
T120, T121 (Ch. KCS 34C) Tel. Rec. (See Model	i
T164 (Ch. KCS40)	01
T164 (Ch. KCS40)	8V
T164 (Ch. KCS40) Tel. Rec	
T164 (Ch. KCS40) Tel. Rec	87
T164 (Ch. KCS40) Tel. Rec	8V 8V 8V 8X
TI 64 (Ch. KC540) Tel. Rec	8V 8V 8V 8X 8X 8X
Ti 64 (Ch. KC540) Tel. Rec	8V 8V 8X 8X 8X 8X 8X
Ti 64 (Ch. KC540) Tel. Rec	8V 8V 8X 8X 8X 8X 8X 8X 8X
Ti 64 (Ch. KC540) Tel. Rec	8V 8V 8X 8X 8X 8X 8X
Ti 64 (Ch. KC540) Tel. Rec	8V 8V 8X 8X 8X 8X 8X 8X
TI 64 (Ch. KC540) Tel. Rec	8V 8V 8X 8X 8X 8X 8X 8X 8X 9B
TI 64 (Ch. KC540) Tel. Rec	8V 8V 8X 8X 8X 8X 8X 8X 8X 8X 8X 8X 9B
T164 (Ch. KC540) Tel. Rec. 109—11 TA.129 (Ch. KC542A), TA.129 (Ch. KC541) Tel. Rec. 110—11 T4.61 (Ch. KC543 & Rodio Ch. KC543 & Rodio Tel. Rec. 108—10 TC124, TC125, TC127 (Ch. KC534, 8) Tel. Rec. (See Model T100),	8V 8V 8X 8X 8X 8X 8X 8X 8X 8X 8X 8X 9B
T164 (Ch. KC540) Tel. Rec. 109—11 TA.128 (Ch. KC542A), Tal.128 (Ch. KC541-1) Tel. Rec. 110—11 Tal69 (Ch. KC543 & Rodio Ch. KK135D) Tel. Rec. 108—10 TC124, TC125, TC127 (Ch. KC534, 8D Tel. Rec. (See Model T100). 93 TC165, TC166, TC167, TC168 (Ch. KC543A) Tel. Rec. Model T164). 10978, C) Y211 (Ch. KC543) Tel. Rec. 10878, C) Y211 (Ch. KC543) Tel. Rec. Tel. Rec. 1251 (Ch. KC543) Tel. Rec. Tel	8V 8V 8X 8X 8X 8X 8X 8X 8X 9B 9B
Ti 64 (Ch. KC540) Tel. Rec	8V 8V 8X 8X 8X 8X 8X 8X 8X 8X 8X 983 983 983 983 983 983 983 983 983 983
T164 (Ch. KC540) Tel. Rec. 109—11 TA.128 (Ch. KC542A), Tal.28 (Ch. KC543 & Rdio Ch. KC543 & Rdio Clas. Tc125, Tc127 (Ch. KC534, B) Tel. Rec. (See Model T100). 93 Tc165, C166, Tc167, Tc168 (Ch. KC540A) Tel. Rec. (See Model T164). 109 X551, X552 (Ch. 10898, C) 129—9 X711 (Ch. KC5453) Tel. Rec., Frod. Chege. Bul. 11. Ts1 (Ch. KC5453) Tel. Rec., Prod. Chege. Bul. 11. Ta16. (Ch. KC5454) Tel. Rec. Model Z151).111 2180 (Ch. KC5454) Tel. Tel. Rec. Model Z151).111 2180 (Ch. KC5454) Tel. Chege. Bul. 11. Ta16. (Ch. CC5454) Tel. Chege. Model Z151).111 Z180 (Ch. KC5454) Tel. Chege. Model Z151).111 Tel. Rec. (Sole Model Z151).111 <td>8V 8V 8X 8X 8X 8X 8X 8X 8X 9B 9B</td>	8V 8V 8X 8X 8X 8X 8X 8X 8X 9B 9B
TI64 (Ch. KC540) Tel. Rec. 109—11 TA.128 (Ch. KC542A), Tal.28 (Ch. KC543.1) Tel. Rec. 110—11 Tal69 (Ch. KC543.8 Rodio Ch. KX135D Tel. Rec. 108—10 TC124, TC125, TC127 (Ch. KC534, 8D Tel. Rec. (See Model T100). 93 TC165, TC166, TC167, TC168 (Ch. KC540A) Tel. Rec. (See Model T164). 109 X551, X552 (Ch. 10898, C) 129—9 X711 (Ch. KC5453) Tel. Rec., See Model T164). T51, K523 (Ch. 1289, C, Ch. 129—9 X711 (Ch. KC5453) Tel. Rec., See Model T164). T61, Rec., Prod. Chege. Bul. 11. T61, Rec., Prod. Chege. Bul. 11. T62, Ch. KC5453) Tel. Rec., Prod. Chege. Bul. 11. T61, Rec., CS453, Tel. T61, Rec., CO00) Tel. Rec. T61, Rec., CO00) Tel. Rec. T61, Rec., Co000) Tel. Rec., * T	8V 8V 8V 8X 8X 8X 8X 8X 8X 8X 9B 9B 9P 9P 9P
Ti 64 (Ch. KC540) Tel. Rec	8V 8V 8V 8X 8X 8X 8X 8X 8X 9B3 9P6 9P6 9P7 9P7 9P7 977 977 977 977 977
TI 64 (Ch. KC540) Tel. Rec	8V 8V 8V 8X 8X 8X 8X 8X 8X 8X 8X 8X 8X 9B 9P 9P 9P 9P 9P 9P 9P 9P 9P 9P 9P 9P 9P
TI 64 (Ch. KC540) Tel. Rec	8V 8V 8V 8X 8X 8X 8X 8X 8X 8X 8X 9B 90 90 90 90 90 90 90 90 90 90 90 90 90

RCA VICTOR-Cont. 6764, 6765 (Ch. KCS47,	RCA VICTOR-Cont.
A) Tel. Rec.	9T126 (Ch. KCS49B, C) Tel. Rec. (See Model 7T103)
(See Model 6154)113 614, 6165 (Ch. KCS47, A, Prod. Chge, Bul, 12120	9T128 (Ch. KCS49B, C) Tel. Rec.
Prod. Chge. Bul. 12, 1201 6T71 Ch. KCS47, A) Tel. Por. /See Model 4T541 113	(See Model 71103) 91147 (Ch. KCS60A)
6T71 (Ch. KCS47, A, AT, T) Tel. Rec. Prod.	Tel. Rec. (See Model 7T103) 9T240 (Ch. KCS28, A)
Chge. Bul. 12	9T240 (Ch. KCS28, A) Tel. Rec. (See Model 8T241)
Madal T164) 109	0T246 (Ch KC529C) Tal
(Ch. KCS47, A) Tel, Rec. (See Model 6154)113	Rec. (See Model 8124 91246 (Ch. KCS38) Tel. Rec. (See Model 1100) 91256 (Ch. KCS38C) Tel.
6174, 6175, 6176 (Ch. KCS47, A. AT, T) Tel	91256 (Ch. KCS38C) Tel. Rec. (See Model T100)
6174, 6175, 6176 (Ch. KCS47, A, AT, T) Tel. Rec. Prod. Chge. Buil, 12	Rec. [See Model T100] 91270 (Ch. KCS29, KCS29C) Tel. Rec. [See Model 81270]
6T84 (Ch. KCS48, Radia Ch. RC-1090 ar	Tel. Rec. (See Model
6T86, 6T87 (Ch. KCS48, Badio Ch. PC1080 or	8T241)
RC1092) Tel. Rec * 7T103, 7T104 (Ch.	9TC247 (Ch. KCS34, B) T Rec. (See Model T100)
Buil 12 001 50 00 00 6180 12 001 000 00 RC-1090 00 RC-1090 00 RC-1090 12 12 12 12 12 12 12 12 12 12 12 12 12	91C245 [Ch. KC3348] 1e Rec. [See Model T100] 9TC247 [Ch. KC534, B] T Rec. [See Model T100] 9TC249 (Ch. KC534, B] T Rec. [See Model T100] 9TC272, 9TC275 (Ch. KC529, KC529C) Tel. [See Model 8T270] 9TW309 [Ch. KC541.]
Tel. Rec. (See Model 7T103)	9TC272, 9TC275 (Ch. KCS29, KCS29C) Tel.
71103)	PK125C) Tel Bes
7T103)	9TW333 (Ch. KCS30-1, Rodio Ch. RC616N) Te Rec. (See Model 8T241
Tel. Rec. (See Model 7T103)134 8B41 (Ch. RC-1069),	Rec. (See Model 8T241 9TW390 (Ch. KCS31-1,
8B41 (Ch. RC-1069), 8B42 (Ch. RC-1069A), 8B43 (Ch. RC-1069B) 76 —16	9TW390 (Ch. KCS31-1, RC617A) Tel. Rec. (See Model S1000)
8841 (Ch. RC-1069A), 8842 (Ch. RC-1069A), 8843 (Ch. RC-1069B), 76 —16 8846 (Ch. RC-1069C) [See Model 8841),	(See Model 31000). 9W101, 9W102, 9W103 (Ch. RC-618B), 9W105 (Ch. RC-618C).
88X6 (Ch. RC-1040C) 44-18	9W106 (Ch. RC-622)
88X54, 88X55 (See Model 88X5)	(See Model A106) 9X561 (Ch. RC-10798), 9X562 (Ch. RC-1079C
86X65 (See Model 86X6), 44 8F43 (Ch. RC-1037B),, 97—13	9X561 (Ch. RC-10798), 9X552 (Ch. RC-1079C 9X571 (Ch. RC-1079), 9X572 (Ch. RC-1079), 9X641 (Ch. RC-1080), 9X642 (Ch. RC-1085), 9X652 (Ch. RC-1085), 9X652 (Ch. RC-1085A), 9Y7 (Ch. RC-1078B)
BrA3 (Ch. RC-1037B) 97—13 BPCS41, BPCS41B, BPCS41B, BPCS41B, BPCS41C (Ch. KCS24B-1, KRS1A-1, KRS20B-1, KRS21A-1, KS-123C) Tal. Per. 90_0	9X642 (Ch. RC-1080A 9X651 (Ch. RC-1085),
KRS20A-1, KRS20B-1, KRS21A-1, RS-123C)	9X652 (Ch. RC-1085A 9Y7 (Ch. RC-1057B)
8R71 (Ch. RC-1060),	9X532 (Ch. RC-1063A 9Y7 (Ch. RC-1057B) 9Y51 (Ch. RC-1077) 9Y510 (Ch. RC1077A), 9Y511 (Ch. RC1077B). 45.FY.3
8R72 (Ch. RC-1060A) 53—20 8R74, 8R75, 8R76 (Ch. RC-1060 A)	54B1, 54B1-N, 54B2,
	5483 (Ch PC590)
8T241, 8T243, 8T244 (Ch. KCS28) Tel. Rec 74—8 8T270 (Ch. KCS29, KCS29A) Tel. Rec 85—13	5485 (Ch. RC1047) 55AU (Ch. RC1017) 55U (See Model 55AU) 55F (Ch. RC-1004E)
KCS29A) Tel. Rec 85—13 8TC270, 8TC271 (Ch.	55FA (See Model 55F) 56X, 56X2, 56X3 (Ch. RC-1011)
KC327A [16], Rec 63—13 RC270, 81C271 [ch. KC529, KC529A] Tel. Rec. [See Model 81270] 85 81K29 [ch. KC532, KC5322, KK135, RK135A] Tel. Rec 88—9 81K290 [ch. KC533A]]	
KCS32A, KCS32B, KCS32C, RK135.	56X10 (Ch. RC-10238) 58AV, 58V (Ch. RC-604)
01K020 (CII. KC000A-1)	56X10 (Ch. RC-10238). 58AV, 58V (Ch. RC-604) 59AV1, 59V1 (Ch. RC-604) 59AV1, 59V1 (Ch. RC-60 63E (Ch. RC-127) 64F1, 64F2 (Ch. RC1037A) 64F3 (Ch. RC1037A)
(Radio Ch. RK-135A-1) Tel. Rec. (See Model 8T270)	65889 (Ch PC-1045)
8T270)	65F (See Model 55F) 65AU (Ch. No. RC-1017) 65U, 65U-1 (See Model 65AU)
8TR29 (Ch. KCS32, KCS32A, KCS32B, KCS32C, RK135, RK135A) Tel. Rec.	65AU) 65X1, 65X2 (Ch. RC-103
(See Model 81829) 88 87530 Tel. Rec. (See	65X1, 65X2 (Ch. RC-103 65X1, 65X2 (Ch. RC-106 65X8, 65X9 (See Model
8TV41 (Ch. KCS25D-1	65X1) 66BX (Ch. RC-1040,
KCS25E-2, RK117A, RS-123A) Tel. Rec * 8TV321, 8TV321B,	RC-1040A) 66E (Ch. RS-126) 66X1, 66X2, 66X3, 66X4.
8TV 321, 8TV 321 B, 8TV 323, 8TV 323B (Ch. KC 530-1) (Radio Ch. RC 61 6B, C, J, K) Tel. Rec. (See Model 8T241) 74 8V7 (Ch. RC-615) (See Model 77V1) 38	64Y7 66Y8 44Y0
RC616B, C, J, K) Tel. Rec. (See Model 8T241) 74	[See Modei 66X1] 66X11 (Ch. RC-1046A], 66X12 (Ch. RC-1046A], 66X13, 66X14, 66X15 (Ch. RC-1046B]
Model 77V1)	(Ch. RC-1046B) 67V1, 67AV1
	67V1, 67AV1 {Ch. RC-606} 68R1, 68R2, 68R3, 68R4 {Ch. RC-608}
8V111,8V112(Ch.RC-616) 58-18 8V151 (See Model RV151) 61 8X53 (Ch. RC-1064)	(Ch. RC-608) 75X11, 75X12 (Ch. RC-1050)
8X53 (Ch. RC-1064) 39–17 8X71, 8X72 (RC-1070) 63–15 8X521 (RC-1066),	75X14, 75X15 (Ch. RC-1050) (See Model
8X522 (RC-1066A) 52—17 8X541, 8X542	75X11) 75X16, 75X17, 75X18,
(Ch. RC-1065, RC-1065A) 59 16 8X544, 8X545, 8X546, 8X547 (See Model 8X541) 59	75X11) 75X16, 75X17, 75X18, 75X19 (Ch. RC-10506) (See Model 75X11) 77U (Ch. RC-1057A) 77VI (Ch. RC-1057A)
8X681, 8X682	77V1 (Ch. RC-615) 77V2 (Ch. RC-615) 77V2 (Ch. RC-606-C) 610V1 (Ch. RC610C)
(Ch. RC-1051) 6510 9BX5 (Ch. RC-1059B, C) (See Model 8BX5) 46	610V1 (Ch. RC610C) 610V2 (Ch. RC610)
98X56 (Ch. RC-1068) 79-13	610V1 (Ch. RC610C) 610V2 (Ch. RC610} 612V1, 612V2, 612V3 (Ch. RK-121, RS-123). 612V4 (See Medel (12)V)
9PC41A, B, C (Ch. KCS24C-1, D, KRK-4,	621TS (Ch. KCS21-1)
9EY31, 9EY32	Tel. Rec. (Servicer) 630TCS Tel. Rec. (See Model 630TS)
(See Model 8PCS41) 90 9757 (Ch. KCS49, A,AT,T) Tel. Rec122—8 9777 (Ch. KCS49, A,	630TS Tel. Rec. 641TV (Ch. KCS25A1-1,
AI,I) Iel. Rec.	KCS25C-2, RK117A, RS-123A) Tel. Rec 648PTK (Ch. KCS24-1
(See Model 9157) 122	430TCS Tel. Rec. (See Model 430TS) 630TS Tel. Rec 641TV (Ch. KCS2541-1, KCS25C-2, RK117A, RS-123A) Tel. Rec 64BPTK (Ch. KCS24-1, KR1-1, KR520-1, KR521A-1, RK.121A, RS-123A) Tel. Rec. (See Model 8PCS41) 64BPY (Ch. KCS2A4-1)
(See Model 9757) 122 (See Model 9757) 122	RS-123A) Tel. Rec. (See Model 8PCS41)
9189 (Ch. KCS60, T) Tel. Rec. (See Model 9157). 122	(See Model 8PCS41) 648PV (Ch. KCS24A-1, KRK-1A, KRS20-1, KRS21A-1, RK-121A, RS-123B) Tel. Rec.
9T105 (Ch. KCS49B, C) Tel. Rec. (See Model 7T103) 134	KRS21A-1, RK-121A, RS-123B) Tel. Rec. (See Model 8PCS41)
	(000 model or 0041)

RCA VICTOR-Cont. 9T126 (Ch. KCS49B, C)	
e Ker	
(See Model 71103)134 91128 (Ch. KCS49B, C) Tel Rec	
97128 (Ch. KCS49B, C) Tel. Rec. (See Model 77103) 134 97147 (Ch. KCS60A)	1
Tel. Rec. (See Model 77103)134	1
91240 (Ch. KCS28, A)	
8T241)	
9T246 (Ch. KC528C) Tel. Rec. (See Model 8T241) 74	
9T246 (Ch. KCS38) Tel. Rec. (See Model T100) 93	1
9T256 (Ch. KCS38C) Tel. Rec. (See Model T100) 93	
9T270 (Ch. KCS29, KCS29C) Tel. Rec.	
Tel. Rec. (See Model 81241)	
Tel. Rec. (See Model 87241)	
9TC245 (Ch. KCS34B) Tel. Rec. (See Model T100) 93	
9TC247 (Ch. KCS34, B) Tel. Rec. (See Model T100). 93	
(See Model 81270) 85 91C240 (CK. KC5288) Tel., Rec. (See Model 81241) Rec. (See Model T100). 93 91C242 (Ch. KC5348) Tel. Rec. (See Model T100). 93 91C249 (Ch. KC534, 8) Tel. Rec. (See Model T100). 93 91C249 (Ch. KC534, 8) Tel. Rec. (See Model 81270) 85 91W309 (Ch. KC531-1, Rodio Ch. RC631-1, Rodio Ch. RC631-1, RC617A) Tel. Rec. (See Model 81270) 91A 91W303 (Ch. KC531-1, RC617A) Tel. Rec.	ĺ
9TC272, 9TC275 (Ch.	
(See Model 8T270) 85	
RK135C) Tel. Rec 95A-11	
Rodio Ch. RC616N) Tel.	
9TW390 (Ch. KC\$31-1,	
(See Model S1000) 91A	1
[See Model S1000] 91A 9W101, 9W102, 9W103 [Ch. RC-618B], 9W105 [Ch. RC-618C] 73—10 9W106 [Ch. RC-622] [See Model A106] 97	1
(See Model A106) 97 9X561 (Ch. RC-10798),	
9X562 (Ch. RC-1079C). 101—9 9X571 (Ch. RC-1079),	
9X572 (Ch. RC-1079A) 107-7 9X641 (Ch. RC-1080),	'
9X642 (Ch. RC-1080A). 87—9 9X651 (Ch. RC-1085),	'
9X652 (Ch. RC-1085A). 104-9 9Y7 (Ch. RC-1057B) 75-13	'
9Y51 (Ch. RC-1077)	1
9Y511 (Ch. RC1077B)13113 45-EY-3	1
[Ch. RC-6168], WVI03 73-10 (Ch. RC-618C),	ί.
5485 (Ch. RC1047) 1725 55AU (Ch. RC1017) 216	
55U (See Model 55AU) 2 55F (Ch. RC-1004E) 4-6	
55FA (See Model 55F) 4 56X, 56X2, 56X3	
(Ch. RC-1011) 1—16 56X5 (See Model 56X10) 1	
56X10 (Ch. RC-10238) 112 58AV, 58V (Ch. RC-604). 132	
59AV1, 59V1 (Ch. RC-605) 6-25 63E (Ch. RS-127) 28-28	
64F3 (Ch. RC1037A) 4-16 65BR9 (Ch. RC-1045) 23-16	
65BR9 (Ch. RC-1045) 23-16 65F (See Model 55F) 4	
65AU (Ch. No. RC-1017A) 14-23 65U, 65U-1 (See Model	
SSF (Ch. RC-1004E) 4-6 SSFA (See Model SSF) 4 SSFA (See Model SSF) 4 SK, Sosz, Sosz 1-16 SKS (See Model SSF) 1-16 SKS (See Model SSF) 1-12 SKAV, SSV (Ch. RC-604) 1-32 SAV, SSV (Ch. RC-604) 1-32 SAF (See Model SSF) 28-28 d4F1 (AF2 (Ch. RC-1037), 6-16 65F (See Model SSF) 4 65KJ (See Model SSF)	
65X1, 65X2 (Ch. RC-1064) 31-26 65X8, 65X9 (See Model	
000A (CR. KC-1040,	
RC-1040A)	
66E (Ch. RS-126) 1726 66X1, 66X2, 66X3, 66X4. 723 66X7, 66X8, 66X9 [See Modei 66X1]	
66X11 (Ch. RC-1046A),	
66X13, 66X14, 66X15 (Ch. RC-1046B) 27-20	•
67V1, 67AV1 (Ch. RC-606) 9	•
68R1, 68R2, 68R3, 68R4	
75X11, 75X12	
(Ch. RC-1050) 33—21 75X14, 75X15 (Ch.	
75X14, 75X15 (Ch. RC-1050) (See Model 75X11)	4
75Y10 (CL DC 1050-1	0
	4
77V1 (Ch. RC-615) 38—18 77V2 (Ch. RC-606-C) 39—18	4
610V2 (Ch. RC610) 31-27	0
(Ch. RK-121, RS-123) 17-27 612V4 (See Model 612V1) 17	¢
621TS (Ch. KCS21-1)	c
Tel. Rec. (Servicer) 78 630TCS Tel. Rec. (See Model 630TS) 54	c
(See Model 63015) 54 630TS Tel. Rec	c
	c
RS-123A) Tel. Rec 91A-11 648PTK (Ch. KCS24-1, KRK1-1, KRS20-1,	c
RKSZIA-I, RK-IZIA, PS-123A1 Tel Pec	c
(See Model 8PCS41) 90	c
(See Model 8PCS41) 90 6489V (Ch. KC524A-1, KRK-1A, KRS20-1, KRS21A-1, RK-121A, RS-123B) Tel. Rec.	c
RS-123B) Tel. Rec. (See Model 8PCS41) 90	С

	RCA VICTOR-Cont. 710V2 (Ch. RC-613A) 40-15	R
	RCM VICINE-Cont. 40—15 710V2 (Ch. RC-613A) 40—15 711V2) 22 711V2, 711V3 (Ch. REX-117 & REX-123) 22 24 721rCS (Ch. RC526-1, 2) 22—24 72 1705 (Ch. RC526-1, 2) 161. Rec. (See Model 70	0
	RK-117 & RS-123) 22-24 721TCS (Ch. KCS26-1, 2)	c
	Tei. Rec. (See Model 730TV1)	c
	7301V1) 70 7301V1 (Ch. KCS27, RC610A) Tel. Rec 70—7 7301V2 (Ch. KCS27,	c
	RC610B) Tel. Rec. (See Model 730TV1) 70	C
	741PCS (Ch. KCS24B-1, KRK1A-1, KRS20A-1,	0
	Tel. Rec. (See Model 8PCS41) 90	0
	Ch. KCS-20A-1 (See Model 630TS) 54 Ch. KCS-20B-1	0
	See Model 630TCS 54	c
	Ch. KCS-20J-1 (See Model 8TS30) 54 Ch. KCS21-1	c
	{See Model 621TS} * Ch. KCS24-1	C
	(See Model 8PCS41) 90 Ch. KCS24A-1	C
	(See Model 8PCS41) 90 Ch. KCS24B-1 (See Model 8PCS41) 90 Ch. KCS24C 8PCS41) 90	0
	(See Model 8PCS41) 90	0
•11	(See Model 8PCS41) 90	c
	Ch. KCS25A1-1 (See Model 641TV) * Ch. KCS25C-2	C
	(See Model 641TV) * Ch. KCS25D-1	C
	Ch. KCS25E-2	0
-10	(See Model 8TV41) * Ch. KCS26-1, KCS26-2 (See Model 721TCS) *	
		c
-9 -7	Ch. KC527 (See Model 730TV1) 70 Ch. KC528, A, B, C (See Model 81241) 74 Ch. KC529, KC529A (See Model 81270) 85 Ch. KC529C (See Model	c
-9	(See Model 8T270) 85 Ch. KCS29C (See Model	c
-9 -13	8T270)	C
-11	(See Model 87241) 74 Ch. KCS31-1 (See Model S1000)	c
-13 -11	Ch. KCS32, KCS32A, KCS32B, KCS32C (See	c
-22 25	(See Model 81241) 74 Ch. KCS31-1 (See Model S1000) 91A Ch. KCS32, KCS32A, SCS32A, KCS328, KCS32C (See Model 81K29) 88 Ch. KCS33A-1 (See Model 81Z70) 85 Ch. KCS34 B C	c
-16	(See Model 81270)	c
-6	Ch. KCS-38, C (See Model T100) 93 Ch. KCS40, A. B. (See	c
-16	Model T164)	c
-12 -32 -25	Model TA-128)110 Ch. KCS42A (See Model TA-128) 110	c
-28	Model TA169) 108	c
-16 -16	Ch. KCS45, A (See Model 2T51)111 Ch. KCS47, A, AT, T (See Model 6T54)113 Ch. KCS47B, C	C
-23		c c
-30 -26	(See Model 71103) 134 Ch. KCS48 (See Model 2181)*	c
-20	Ch KCS484	с
-24	(See Model 9157) 122	c
-24 -26 -23	Ch. KCS49B, C (See Model 9T105) 134 Ch. KCS60, T	с
	(See Model 9189) 122 Ch. KCS60A	c
-20	(See Model 97147)134 Ch. KRK-1A (See Model 8PCS41) 90	c c
-27	{See Model 8PCS41) 90 Ch. KRK1-1 (See Model 8PCS41) 90	c
-17	Ch. KRK1A-1	c
21	(See Model 8PCS41) 90 Ch. KRK4 (See Model 8PCS41) 90 Ch. KRS20-1	c
	{See Model 8PCS41} 90 Ch. KRS20A-1	с
	(See Model 8PCS41) 90 Ch. KRS20B-1 (See Model 8PCS41') 90	C
-17 -18	Ch. KRS21A-1 (See Model 8PCS41) 90	CI
18	(See Model 54B1) 7	С
27	Ch. RC-604 (See Model 58AV)	с
	(See Model 59AV1) 6 Ch. RC-606	CI
	(See Model 67V1) 9 Ch. RC-606C (See Model 77V2) 39	CF
18	Ch. RC-608 (See Model 68R1) 23	CH
11	(See Model 610V1) 31	CH CH
ļ	Ch. RC610A, RC610B (See Model 730TV1) 70 Ch. RC610C	CF
	(See Model 610V1) 31 Ch. RC-613A	Ch
	(See Model 710V2) 40 Ch. RC-615	СЪ
	(See Model 77V1) 38 Ch. RC-616 (See Model 8V111) 58	Ch

RCA VICTOR-Cont.
Ch. RC-616A, RC-616H
(See Model 8V01) 56
(See Model 8T241) 74
(See Model 8T241) 74
Ch. RC-616N (See Model 8T241) 74
(See Model 8T241) 74 Ch. RC617A, 8 (See Model S1000) 914
Ch. RC-618, RC-618A
(See Model 8V90) 56 Chassis RC-618 B, C
Chassis RC-618 B, C (See Model 9W101) 73 Ch. RC-622
[See Model AIU6] 9/
(See Model 55F) 4
Ch RC-1017
Ch RC-1017A
(See Model 65AU) 14 Ch. RC-10238
(See Model 56X10) 1
(See Model 65X1) 4
Ch. RC-1037, RC-1037A (See Model 64F1)
Ch. RC-1037B
(See Model 66X1) 7 Ch. RC-1040, RC-1040A
(See Model 66BX) 14
Ch. RC-1040C (See Model 8BX6) 44 Ch. RC-1045
(See Model 6588?)
Ch. RC-1046, A, 8
{See Model 66X11} 27 Ch. RC-1047
(See Model 5485) 17
(See Model 5453) 17 Ch. RC-1050, RC-1050B (See Model 75X11) 33 Ch. RC-1057A (See Model 77U) 38 Charte BC 1057B (See
Ch. RC-1057A (See Model 77U) 38 Chassis RC-1057B (See
Chassis RC-1057B (See Model 9Y7)
(See Model 8BX5) 46 Ch. RC-10598, RC-1059C
(See Model 9BX5) 46
Ch. RC-1059 (See Model 88X5) 46 Ch. RC-1059B, RC-1059C (See Model 98X5) 46 Ch. RC-1060 (See Model 8R71) 53 Ch. RC-1060A
[See Model 8R72] 53
(See Model 8X51)
(See Model 65X1) 31 Ch. RC-1065, RC-1065A
(See Medel 87541) 59
(See Model 8X521) 52
Ch. RC-1066A (See Model 8X522) 52
Ch. RC-1068
(See Model 9BX56) 79 Ch. RC-1069A, B
(See Model 9841) 76 (See Model 8841) 76 Ch. RC-1070 (See Model 8X71) 63
(See Model 8X71) 63 Ch. RC-1070A
(See Model X711)133
(See Model 9Y51) 98
Ch. RC-1079, A (See
Model 9X571)
(See Model 9X561)101
(See Model 9X301)
Ch. PC-1092 (See Model
BX6)
BX6) Ch. RC-1085, RC-1085A (See Model 9X651)104
BX6) Ch. RC-1085, RC-1085A (See Model 9X651)104
BX6) Ch. RC-1085, RC-1085A (See Model 9X651)104
BX6). 103 Ch. RC-1085, RC-1085A (See Model 9X651)104 Ch. RC-1087, RC-1088A (See Model A55)109 Ch. RC-1088, RC-1088A (See Model BX55)102 Ch. RC10898, C. 10898, C
BX0) 103 Ch. RC-1085, RC-1085A [See Model 9X651] 104 (See Model A55) 109 Ch. RC-1088, RC-1088A [See Model X55] 102 Ch. RC-10895, C [See Model X55] 129 Ch. RC-1090 [See Model 781]
BA0)
BA0)
BX0 103 Ch. RC-1085, RC-1085A [See Model 9X651]104 Ch. RC-1087 [See Model A55]102 Ch. RC-1088, RC-1088A [See Model A55]102 Ch. RC-1089, CC [See Model A55]102 Ch. RC-1089, RC-1088A [See Model A55]102 Ch. RC-1098, CC [See Model 783]
BX0 103 Ch. RC-1085, RC-1085A [See Model 9X651]104 Ch. RC-1087 [See Model A55]102 Ch. RC-1088, RC-1088A [See Model A55]102 Ch. RC-1089, CC [See Model A55]102 Ch. RC-1089, RC-1088A [See Model A55]102 Ch. RC-1098, CC [See Model 783]
BX0] 103 Ch. RC-1085, RC-1085A [See Model 9X651]104 (See Model 8X55]107 [See Model 8X55]102 Ch. RC-1088, RC-1088A [See Model 8X55]102 Ch. RC-1090 [See Model X551]129 [See Model X551]129 Ch. RC-1090 [See Model 7181]. * Ch. RC-1092 [See Model 8411]
BX0 103 Ch. RC-1085, RC-1085A [See Model 9X651]. 104 Ch. RC-1087, RC-1087 [See Model AS5]. 107 Ch. RC-1088, RC-1088A [See Model AS5]. 102 Ch. RC-1088, RC-1088A [See Model AS5]. 102 Ch. RC-1098, RC-1088A [See Model AS5]. 102 Ch. RC-1090 [See Model 2781]. * Ch. RC-1092 [See Model 9757 122 Ch. RC-1098 [See Model A11]. 132 [Ch. RK-117 [See Model 717A [See Model 811]. 132 Ch. RK-117A [See Model 8141]. *
BX0) 103 Ch. RC-1085, RC-1085A [See Model \$K51] See Model \$K55] 107 [See Model \$K55] 107 [See Model \$K55] 102 [Ch. RC-1088, RC-1088A [See Model \$K55] [See Model \$K55] 102 [Ch. RC-1098, RC [See Model \$K55] [See Model \$K55] 102 [Ch. RC-1090 (See Model \$7100 (See Model \$757) 122 [See Model \$K-11]
BX0] 103 Ch. RC-1085, RC-1085A [See Model 9X651] 104 Cse Model 9X651] 107 [See Model A55] 102 Ch. RC-1088, RC-1088A [See Model 8X55] 102 [See Model 8X55] 102 Ch. RC-1098, RC-1088A [See Model 8X55] 102 [See Model 781]
BX0] 103 Ch. RC-1085, RC-1085A [See Model 9X651] 104 Cse Model 9X651] 107 [See Model A55] 102 Ch. RC-1088, RC-1088A [See Model 8X55] 102 [See Model 8X55] 102 Ch. RC-1098, RC-1088A [See Model 8X55] 102 [See Model 781]
BX0] 103 Ch. RC-1085, RC-1085A [See Model 9551]. 104 Ch. RC-1087, RC-1087 [See Model A55]. 107 Ch. RC-1088, RC-1088A [See Model A55]. 102 Ch. RC-1088, RC-1088A [See Model A55]. 102 Ch. RC-1098, RC-1088A [See Model X55]. 102 Ch. RC-1090 [See Model 2781]. * Ch. RC-1092 [See Model 9757 122 Ch. RC-1098 [See Model 411]. 132 [See Model 7157 Ch. RK-1092 [See Model 9757 122 Ch. RK-1092 [See Model 9757 122 Ch. RK-117A [See Model 811]. 132 [See Model 717A [See Model 1221]. 17 [See Model 821]. 90 [Ch. RK-133, RK-133, A1 [See Model 8151]. 61 [See Model 8120]. 88 (See Model 8121) 17 [See Model 8151]. 60 (See Model 8121) 17 [See Model 8151]. 61 (See Model 8151) 61 (See Model 8120]. 88 (See Model 81354) <
BA0) 103 Ch. RC-1085, RC-1087A [See Model 855]). 104 Ch. RC-1087 [See Model 855]). 107 See Model A55]. 108 RC-1088A [See Model 855]. 102 [See Model 855]). 102 Ch. RC-1098, RC-1088A [See Model 855]). 129 Ch. RC-1098, RC [See Model 711V2]). 122 Ch. RC-1092 (See Model 9157 122 [See Model 8411]. 132 Ch. RC-1092 (See Model 9157 122 [See Model 8411]. 132 Ch. RC-1092 (See Model 9157 122 [See Model 8411]. 132 Ch. RC-1098 (See Model 811) 132 [See Model 811]. 132 Ch. RC-1098 (See Model 811]. 132 [See Model 81]. 132 Ch. RK-117A [See Model 81]. 132 [See Model 81]. 132 Ch. RK-121A [See Model 87541]
BX0] 103 Ch. RC-1085, RC-1085A [See Model \$K51] 104 (See Model \$K55]
BX0] 103 Ch. RC-1085, RC-1085A [See Model \$K51] 104 (See Model \$K55]
BX0] 103 Ch. RC-1085, RC-1085A [See Model \$K51] 104 (See Model \$K55]
BX0] 103 Ch. RC-1085, RC-1085A [See Model \$K51] 104 (See Model \$K55]
BX0] 103 Ch. RC-1087, RC-1087, 104 See Model X551) 107 See Model X551) 108 Ch. RC-1088, RC-1088A Isee Model RS551) See Model RS551) 102 Ch. RC-1098, RC 1088A Isee Model RS551) 102 Ch. RC-1098, RC 1088A Isee Model RS51) 129 Ch. RC-1098, RC 108 Model 2181] 132 Ch. RC-1090 (See Model 202 (See Model 9157 122 Ch. RC-1092 (See Model 202 (See Model 9157 122 Ch. RK-1072 ISee See Model 81711V2) 22 Ch. RK-117A ISee Model 81711 See Model 817201 17 Ch. RK-1212 ISee Model 8187201 ISee Model 817201 85 Ch. RK-135A ISee Model 817201 ISee Model 817201 85 Ch. RK-135A ISee Model 817201 ISee Model 817201 85 Ch. RK-135A 108
BX0] 103 Ch. RC-1087, RC-1087, 104 See Model X551) 107 See Model X551) 108 Ch. RC-1088, RC-1088A Isee Model RS551) See Model RS551) 102 Ch. RC-1098, RC 1088A Isee Model RS551) 102 Ch. RC-1098, RC 1088A Isee Model RS51) 129 Ch. RC-1098, RC 108 Model 2181] 132 Ch. RC-1090 (See Model 202 (See Model 9157 122 Ch. RC-1092 (See Model 202 (See Model 9157 122 Ch. RK-1072 ISee See Model 81711V2) 22 Ch. RK-117A ISee Model 81711 See Model 817201 17 Ch. RK-1212 ISee Model 8187201 ISee Model 817201 85 Ch. RK-135A ISee Model 817201 ISee Model 817201 85 Ch. RK-135A ISee Model 817201 ISee Model 817201 85 Ch. RK-135A 108
BX0] 103 Ch. RC-1085, RC-1085A Isee Model 9X51]. 104 Ch. RC-1087, RC-1087A Isee Model AS5]. 107 See Model AS5]. 108, RC-1088A Isee Model AS5]. 102 Ch. RC-1088, RC-1088A Isee Model 8X55]. 102 Isee Model 8X55]. 102 Ch. RC-1098, RC-1088A Isee Model 8X51]. 122 Isee Model 9157 122 Ch. RC-1098 Case Model 9157 122 Isee Model 9157 122 Ch. RC-1098 Case Model 8111]. 132 Isee Model 61271]. 132 Ch. RK-10708 Isee Model 81741]. * Ch. RK-117 Isee Model 81741]. * Ch. RK-117 Isee Model 81271]. 17 Isee Model 81271]. 17 Ch. RK-121 Isee Model 81270]. 85 Ch. RK-135A Isee Model 81270]. 85 Ch. RK-135A Isee Model 81270]. 108 Ch. RK-135A, Isee Model 81270]. 108 Ch. RK-135D (See Model 81270]. 108 Isee Model 81270]. 108 Ch. RK-132A, B, C Isee Model 81270]. 108
BX0] 103 Ch. RC-1085, RC-1085A (See Model \$K51) 104 (See Model \$K55) 107 (See Model \$K55) 107 (See Model \$K55) 108, RC-1088A (See Model \$K55) 102 (See Model \$K55) 109 (See Model \$K55) 102 (See Model \$K55) 102 (See Model \$K55) 102 (Ch. RC-1090 (See Model \$K11) 122 (Ch. RC-1090 (See Model \$K11) 122 (Ch. RC-1090 (See Model \$K11) 132 (See Model \$K11) 132 (Ch. RC-1094) (See Model \$K11) 132 (See Model \$K11)

DALE
 KME
 50—14

 DB-22A
 50—14

 HF10-20
 49—17

 VHF 2-11
 79—14

 VHF-152A
 51—18

 584
 13—25

 84
 14—13

 84
 14—13

 RADIOLA
 61-1, 61-2, 61-3

 61-1, 61-2, 61-3
 (Ch. RC-1021)

 15-1 (Ch. RC-1023)
 12—25

 61-10 (Ch. RC-1038)
 12—25

 61-10 (Ch. RC-1038)
 12—25

 61-10 (Ch. RC-1034)
 27—21

 62-2 (See RCA Model
 550-11

 50-11 (Ch. RC-1034)
 14

 752 (Ch. RC-1038)
 14

 752 (Ch. RC-1058)
 14

 762 (Ch. RC-1058)
 14

 68 (Ch. RC-1033)
 14

 Ch. RC-1033, RC-1023B
 12

 16 (RC-1034)
 14

 17 (See Model 61-5)
 12

 18 (See Model 51-5)
 12

 19 (See Model 752U)
 36

 10 (See Model 752U)
 36
 RADIO CRAFTSMAN
 RADIO
 CRAFTSMAN

 RC-1
 {Tuner},
 39-19

 RC-2
 {Audio Amp.}
 39-19

 "Kitchenaire"
 6-14

 RC-8
 66-13

 RC10
 110-12

 RC100-161, Rec.
 96-9

 RC100A Te1, Rec.
 117-11

 RC101 Te1, Rec.
 *

 RC100 Te1, Rec.
 129-1A
 RADIO DEVELOPMENT & RESEARCH CO. (See Magic-Tone) RADIOETTE PR-2 50—15 RADIONIC (See Chancellor) Y62W, Y728 26-22 RANGER 118 28-27 RADIO MFG. ENGINEERS (See RME) RADIO WIRE TELEVISION (See Lafavette) RAULAND RAY ENERGY **RAYTHEON** (Also See Belmont)

RAYTHEON-Cont.

REGAL (TOK-FONE)-C	ont.
CR761	50—16 68—14
L-76	5-18
L-76 W700 (See Model W800) W800, W801	14
W700 (See Model W800) W800, W801 W900, W901 16131 Tel. Rec. 16136 Tel. Rec. 19C36 Tel. Rec. 205 	. 13-28
16131 Tel. Rec 16T36 Tel. Rec.	. 80—14
19C36 Tel. Rec	*
205 208 (See Model W800).	26-23
747 777 1007 Tel. Rec	27-22
777 1007 Tel. Rec	53-21 83-9
1030, 1031 Tel. Rec.	
(See Model 16T31)	80
1049 1107 1207, 1208 Tel. Rec. (See Model 1007)	41-19
1207, 1208 Tel. Rec.	83
(See Model 1007) 1230 Tel. Rec. (See Model 16T31) 1500	. 03
(See Model 16T31)	. 80
1500 1607 Tel. Rec.	. 38—19
(See Model 1007)	83
7152	28-29
7162	69-12
(See Model 16731) 1500	66—14 40—16
REMBRANDT	
80 Tel. Rec	. •
80 Tel. Rec 130 Tel. Rec 721, 1606, 1606-15, 1950 Tel. Rec	
1950 Tel. Rec	. 65-11
MP5-5-3	8-28 23-18
5300B, 5300B1, 53001 . 5310	. 23—18 . 40—17
5400, 5410	44 —19 27 —23
5500 "Scottie Pup"	27—23
Pup'' (See Model 5500)	27
MB5-5-3 3300B, 5300B1, 53001 3400, 5410 5500, 'Scottie Pup' 5505, 5510, 5515 'Scottie Pup' (See Model 5500) 5520, 5530 'Scottie Jun- ior' (See Model 5500) 6000	
6000	27
RENARD	
L-1A, PT-1A, 185T-1	. 9—28
SCOTT (E. H.)	
Musicale Music Control, Dynamic Noise Suppressor	44-20
Music Control, Dynamic Noise Suppressor	4621
6T11, 6T11A Television	
Receiver 6111, 6111A Tel. Rec. Prod. Chge. Bul. 4. 13A Tel. Rec. 16A 300 Tel. Rec. 400 Tel. Rec. (Se. Hudge) CT111	52-19
Prod. Chge. Bul. 4	105-2
13A Tel. Rec	40-18
300 Tel. Rec	40-10
400 Tel. Rec.	5.0
400 Tel. Rec. Prod. Chae.	. 32
Bul. 4	105-2
ВиІ. 4 510 800-В	105—2 103—14 14—27
Bul. 4 510 800-B 800BT Tel. Rec. [See	105—2 103—14 14—27
Bul. 4 510 800-B 800BT Tel. Rec. [See Model 6T11 (Set 52) and Model 800B Set 14]	105—2 103—14 . 14—27
Bul. 4 510 800-B Model 6T11 (Set 52) and Model 8008 Set 14 800BT Tel. Rec. Prod.	105—2 103—14 14—27
Bul. 4 510 800-B Model 6T11 (Set 52) and Model 8008 Set 14 800BT Tel. Rec. Prod. Chge. Bul. 4 SCOTT (M. M.)	105-2 103-14 14-27
210-A 211-A	
210-A	
SEARS-ROEBUCK (See Silvertone)	
SEARS-ROEBUCK (See Silvertone) SENTINEL	79-15 81-14
Scoll (N. N.) 210-A 211-A 211-A SEARS-ROEBUCK (See Silvertone) SENTINEL 10-284GA (See Model 284GA) 10-284N, 10-284NA, 10-284N, 10-284NA, 10-283P (See Model 283P 10-293CT (See Model 293CT) 10-293T, 10-293T,	79-15 81-14
Scoll (N. N.) 210-A 211-A 211-A SEARS-ROEBUCK (See Silvertone) SENTINEL 10-284GA (See Model 284GA) 10-284N, 10-284NA, 10-284N, 10-284NA, 10-283P (See Model 283P 10-293CT (See Model 293CT) 10-293T, 10-293T,	79-15 81-14
Scoll (n. n.) 210-A 211-A 211-A (See Silvertone) SENTINEL U-284GA (See Model 284GA) U-284N, 1U-284NA, 1U-284N, 1U-284N (See Model 2841) U-285P (See Model 285P U-293CT) 1U-293CT (See Model 293CT) 1U-293V (See Model 294 Series) 1U-2941, 1U-294N, 1U-2941 (See Model	. 79-15 81-14
Scoll (n. n.) 210-A 211-A 211-A (See Silvertone) SENTINEL U-284GA (See Model 284GA) U-284N, 1U-284NA, 1U-284N, 1U-284N (See Model 2841) U-285P (See Model 285P U-293CT) 1U-293CT (See Model 293CT) 1U-293V (See Model 294 Series) 1U-2941, 1U-294N, 1U-2941 (See Model	79-15 81-14 22 1 6 . 29 1
Scoll (n. n.) 210-A 211-A 211-A (See Silvertone) SENTINEL U-284GA (See Model 284GA) U-284N, 1U-284NA, 1U-284N, 1U-284N (See Model 2841) U-285P (See Model 285P U-293CT) 1U-293CT (See Model 293CT) 1U-293V (See Model 294 Series) 1U-2941, 1U-294N, 1U-2941 (See Model	79-15 81-14 22 1 6 29 1 103-15
Scoll (n. n.) 210-A 211-A 211-A (See Silvertone) SENTINEL U-284GA (See Model 284GA) U-284N, 1U-284NA, 1U-284N, 1U-284N (See Model 2841) U-285P (See Model 285P U-293CT) 1U-293CT (See Model 293CT) 1U-293V (See Model 294 Series) 1U-2941, 1U-294N, 1U-2941 (See Model	79-15 81-14 22 1 6 . 29 1
Scoll (N. N.) 210-A 211-A 211-A 211-A 21-A Scars-ROEBUCK (See Silvertone) SENTINEL 10-284GA (See Model 284GA) 10-284N, 10-284NA, 10-283P (See Model 283P 10-293CI 10-293CI 10-293CI 10-293CI 10-293CI 10-293CI 10-293CI 10-293V (See Model 294 Series) 10-294L, 10-294N, 10-3124C (See Model 294 Series) 10-314C, 10-312W (See Model 3131). 10-314E, 10-314L, 10-314K (See Model 20-310-312).	$\begin{array}{c} 79 - 15 \\ 81 - 14 \\ 22 \\ 1 \\ 6 \\ 29 \\ 1 \\ 103 - 15 \\ 39 \end{array}$
Scoll (N. N.) 210-A 211-A 211-A 211-A 21-A (See Silvertone) SENTINEL 10-284GA (See Model 284GA) 10-284N, 10-284N, 10-284N, 10-284N (See Model 2841) 10-283P (See Model 283P 10-293V (See Model 293CT) 10-293V (See Model 294 Series) 10-2941, 10-293T, 10-293V (See Model 294 Series) 10-3124C (See Model 294 Series) 10-3124C (See Model 294 Series) 10-3124W (See Model 3131), 10-313W (See Model 3131), 10-314E, 10-3141, 10-314W (See Model	79-15 81-14 22 1 6 29 1 103-15
Scoll (N. N.) 210-A 211-A 211-A 211-A 21-A (See Silvertone) SENTINEL 10-284GA (See Model 284GA) 10-284N, 10-284N, 10-284N, 10-284N (See Model 2841) 10-283P (See Model 283P 10-293V (See Model 293CT) 10-293V (See Model 294 Series) 10-2941, 10-293T, 10-293V (See Model 294 Series) 10-3124C (See Model 294 Series) 10-3124C (See Model 294 Series) 10-3124W (See Model 3131), 10-313W (See Model 3131), 10-314E, 10-3141, 10-314W (See Model	$\begin{array}{c} 79 - 15 \\ 81 - 14 \\ 22 \\ 1 \\ 6 \\ 29 \\ 1 \\ 103 - 15 \\ 39 \\ 38 \\ 48 \end{array}$
Scoll (N. N.) 210-A 211-A 211-A 211-A 21-A (See Silvertone) SENTINEL 10-284GA (See Model 284GA) 10-284N, 10-284N, 10-284N, 10-284N (See Model 2841) 10-283P (See Model 283P 10-293V (See Model 293CT) 10-293V (See Model 294 Series) 10-2941, 10-293T, 10-293V (See Model 294 Series) 10-3124C (See Model 294 Series) 10-3124C (See Model 294 Series) 10-3124W (See Model 3131), 10-313W (See Model 3131), 10-314E, 10-3141, 10-314W (See Model	$\begin{array}{c} 79 - 15 \\ 81 - 14 \\ \cdot 22 \\ \cdot 29 \\ \cdot 1 \\ 103 - 15 \\ \cdot 39 \\ \cdot 38 \end{array}$
SEARS-ROEBUCK (See Silvertone) SENTINEL 10-284GA (See Model 284GA) 10-2841, 10-284NA, 10-2841, 10-284N, 10-284N, 10-284W (See Model 284) 10-2857 (See Model 293CT) 10-2937 (See Model 294, 10-2937, 10-2934, 10-2937, 10-2941, 10-294N, 10-2941, 10-294N, 10-2944, Series) 10-3744, 10-314C, 10-3124C, 10-3124W (See Model 3131) 10-314E, 10-314C, 10-316PM, 10-316PM (S-38-6, N-28-6, N-29-20) (S-38-6, N-28-6, N-29-20) 10-338-1, 10-338-8, 10-338-1, 10-338-8, 10-338-1, 10-338-8, 10-338-1, 10-338-8, 10-338-1, 10-338-1, 10-338-8, 10-338-1, 10-338-8, 10-336-1, 10-338-8, 10-336-1, 10-338-8, 10-336-1, 10-338-8, 10-336-1, 10-338-8, 10-336-1, 10-338-8, 10-336-1, 10-338-8, 10-338-1, 10-338-8, 10-338-1, 10-338-8, 10-338-1, 10-388-1, 10-388-1, 10-388-1, 10-388-1, 10-388-1, 10-388-1, 10-	$\begin{array}{c} 79 - 15 \\ 81 - 14 \\ 22 \\ 1 \\ 6 \\ 29 \\ 1 \\ 103 - 15 \\ 39 \\ 38 \\ 48 \\ 105 - 9 \end{array}$
Scoll (n. n.) 210-A 211-A 211-A 211-A 22-A 2	79 - 15 81 - 14 22 1 6 29 1 103 - 15 39 38 48 105 - 9 122 - 9 111 - 12
Scoll (N. N.) SEARS-ROEBUCK (See Silvertone) SENTINEL 10-284GA (See Model 284GA) 10-284N, 10-284NA, 10-284N, 10-284N (See Model 2841) 10-283P (See Model 293CT) 10-293W (See Model 293CT) 10-293W (See Model 2943, 10-293T, 10-293W (See Model 2943, 10-293W, 10-2941, 10-294N, 10-2941, 10-294N, 10-2941, 10-294N, 10-2944, Series) 10-312FM, 10-313FW 10-316FM, 10-316FT (See Model 3131). 10-316FM, 10-316FT (See Model 316FM). 10-338-1, 10-338-R, 10-339-K 10-346	$\begin{array}{c} 79 - 15 \\ 81 - 14 \\ . 22 \\ . 1 \\ . 6 \\ . 29 \\ . 1 \\ . 103 - 15 \\ . 39 \\ . 38 \\ . 48 \\ . 105 - 9 \\ . 105 - 9 \\ . 122 - 9 \\ . 11 - 12 \\ . 129 - 10 \end{array}$
Scoll (N. N.) SEARS-ROEBUCK (See Silvertone) SENTINEL 10-284GA (See Model 284GA) 10-284N, 10-284NA, 10-284N, 10-284N (See Model 2841) 10-283P (See Model 293CT) 10-293W (See Model 293CT) 10-293W (See Model 2943, 10-293T, 10-293W (See Model 2943, 10-293W, 10-2941, 10-294N, 10-2941, 10-294N, 10-2941, 10-294N, 10-2944, Series) 10-312FM, 10-313FW 10-316FM, 10-316FT (See Model 3131). 10-316FM, 10-316FT (See Model 316FM). 10-338-1, 10-338-R, 10-339-K 10-346	$\begin{array}{c} 79 - 15 \\ 81 - 14 \\ . 22 \\ . 1 \\ . 6 \\ . 29 \\ . 1 \\ . 103 - 15 \\ . 39 \\ . 38 \\ . 48 \\ . 105 - 9 \\ . 105 - 9 \\ . 122 - 9 \\ . 117 - 12 \\ . 105 - 9 \\ . 129 - 10 \\ . 17 - 12 \\ . 15 - 9 \end{array}$
Scoll (N. N.) SEARS-ROEBUCK (See Silvertone) SENTINEL 10-284GA (See Model 284GA) 10-284N, 10-284NA, 10-284N, 10-284N (See Model 2841) 10-283P (See Model 293CT) 10-293W (See Model 293CT) 10-293W (See Model 2943, 10-293T, 10-293W (See Model 2943, 10-293W, 10-2941, 10-294N, 10-2941, 10-294N, 10-2941, 10-294N, 10-2944, Series) 10-312FM, 10-313FW 10-316FM, 10-316FT (See Model 3131). 10-316FM, 10-316FT (See Model 316FM). 10-338-1, 10-338-R, 10-339-K 10-346	$\begin{array}{c} 79 - 15 \\ 81 - 14 \\ \end{array}$
Scoll (N. N.) SEARS-ROEBUCK (See Silvertone) SENTINEL 10-284GA (See Model 284GA) 10-284N, 10-284NA, 10-284N, 10-284N (See Model 2841) 10-283P (See Model 293CT) 10-293W (See Model 293CT) 10-293W (See Model 2943, 10-293T, 10-293W (See Model 2943, 10-293W, 10-2941, 10-294N, 10-2941, 10-294N, 10-2941, 10-294N, 10-2944, Series) 10-312FM, 10-313FW 10-316FM, 10-316FT (See Model 3131). 10-316FM, 10-316FT (See Model 316FM). 10-338-1, 10-338-R, 10-339-K 10-346	$\begin{array}{c} 79 - 15 \\ 81 - 14 \\ . 22 \\ . 1 \\ . 6 \\ . 29 \\ . 1 \\ . 103 - 15 \\ . 39 \\ . 38 \\ . 48 \\ . 105 - 9 \\ . 105 - 9 \\ . 122 - 9 \\ . 117 - 12 \\ . 105 - 9 \\ . 129 - 10 \\ . 17 - 12 \\ . 15 - 9 \end{array}$
Scoll (N. N.) SEARS-ROEBUCK (See Silvertone) SENTINEL 10-284GA (See Model 284GA) 10-284N, 10-284NA, 10-284N, 10-284N (See Model 2841) 10-283P (See Model 293CT) 10-293W (See Model 293CT) 10-293W (See Model 2943, 10-293T, 10-293W (See Model 2943, 10-293W, 10-2941, 10-294N, 10-2941, 10-294N, 10-2941, 10-294N, 10-2944, Series) 10-312FM, 10-313FW 10-316FM, 10-316FT (See Model 3131). 10-316FM, 10-316FT (See Model 316FM). 10-338-1, 10-338-R, 10-339-K 10-346	79-15 81-14 22 1 29 1 103-15 39 38 48 105-9 122-9 111-12 129-10 117-12 115-9 124-9 124-9
Scoll (M. H.) 210-A 211-A 211-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 222-A 22-A 22-A 22-	79-15 81-14 22 1 6 29 1 103-15 39 38 105-9 122-9 115-9 124-9 124 132-1
Scoll (M. H.) 210-A 211-A 211-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 222-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A	$\begin{array}{c} 79 - 15 \\ 81 - 14 \\ \end{array}$
Scoll (M. H.) 210-A 211-A 211-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 222-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A	$\begin{array}{c} 79 - 15 \\ 81 - 14 \\ \end{array}$
Scoll (M. H.) 210-A 211-A 211-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 222-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A	$\begin{array}{c} 79 - 15 \\ 81 - 14 \\ \end{array}$
Scoll (M. H.) 210-A 211-A 211-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 221-A 222-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A 22-A	$\begin{array}{c} 79 - 15 \\ 81 - 14 \\ \end{array}$
Scoll (M. H.) 210-A 211-A SENTINEL U-284A U-284A 10-284F, 10-284A, 10-284H, 10-293T, 10-293T, 10-293T, 10-293T, 10-293T, 10-293T, 10-293T, 10-293T, 10-293T, 10-294H, 10-293T, 10-294H, 10-293T, 10-312W 10-294J, 10-293T, 10-293T, 10-312W 10-314E, 10-312W 10-314E, 10-313W See Model 331] 10-314E, 10-312W 10-314E, 10-312W 10-314E, 10-313W See Model 336PM, 10-316PT 10-384C, PI, PM, PW, 10-387C, PI, PM, PW, 10-387C, PI, PM, PW, 10-387C, PI, PM, PW, 10-338T, 10330*K 10339-K 10339-K 10339-K 10339-K 10339-K 10339-K 10339-K 10339-K 10339-K 10420 <	$\begin{array}{c} 79 - 15 \\ 81 - 14 \\ \end{array}$
Scoll (M. H.) 210-A 210-A 211-A 211-A 211-A 211-A 211-A 221-A 211-A 211-A 211-A 211-A 211-A 211-A 211-A 211-A 211-A 211-A 210-23 221-A 224 (See Model 224 (See Model 224 (See Model 2294), 10-2937, 10-293	$\begin{array}{c} 79 - 15 \\ 81 - 14 \\ \end{array}$
Scoll (N. N.) 210-A 211-A SENTINEL U-284A U-284A U-284N, 1U-284N, 1U-284N U-293U U-293U, ISee Model 284] U-293U Series) U-293U U-293U See Model 2311 U-293U U-293U See Model 311 U-312FC, 10312PW U312FC, 10312PW U312FC, 10312PW U312FC, 10312PW U312FC, 10312PW U312FC, 10312PW U33FC, PI, PRA U33B-K, 1033FC, PI, PRA, PW U33B-K U340E U340E U420B U420B U420B U420B </td <td>$\begin{array}{c} 79 - 15 \\ 81 - 14 \\ \end{array}$</td>	$\begin{array}{c} 79 - 15 \\ 81 - 14 \\ \end{array}$

SENTINEL-Cont. SETCHELL-CARLSON SETCHELL-CARLSON 150 Tel. Rec. * 416 2—14 427 21—29 437 39—22 447 40—20 458-RD 106—13 469 99—15 570 97—15 2500, 2500LP Tel. Rec. * SHERIDAN ELECTRONICS (See Vogue) SIGNAL 37-19 141 241 33-25 341-A 341-T 311 Ch. 132.878). 101—10 33 (Ch. 548.363). 111—13 33 (Ch. 548.363). 111—13 34 (Al. (Ch. 132.887). 112—8 54 (S6 (Ch. 132.8887). 112—8 67 (101.859-1, 101—10 101 (Ch. 132.887). 113—10 64 (S5 (Ch. 132.887). 113—10 101 (Ch. 549.100), 101A 113 101 (Ch. 549.100), 101A 114. Rec. 102—12 105 (Ch. 132.887) 102—12 106 (Ch. 132.887) 102—12 106 (Ch. 132.887) 104—10 108 (Ch. 549.100) Tel. 102 rel. Rec. 115—11 122 (Ch. 478.289) 115—11 720 (Ch. 478.287) 115—11 122 (Ch. 478.287) 124. rel. Rec. 115—11 122 (Ch. 478.287) 115—11 122 (Ch. 478.287) 116—10 133 (Ch. 100.043) 104—10 SILVERTONE Rec. 104-10 133 (Ch. 100.043) rec. 104—10 133 (Ch. 100.043) Tel. Rec. 138 (Ch. 549.100.3) Tel. Rec. 138 (Ch. 549.100.3) Tel. Rec. 143 Tel. Rec. 99A-10 143 Tel. Rec. 99A-10 143 Tel. Rec. 121 143 A (Ch. 100.111) Tel. Rec. Tel. Rec. 121—12 159 (Ch. 478.309) Tel. Rec. (See Model 120). Rec. (See Model 120). 115 160-12 (Ch. 549.100.4) Tel. Rec. Tel. Rec. 97A-12 161-16 (Ch. 100.112) Tel. Rec.

RME-SILVERTONE

SILVERTONE—Cont. 179-16, 180-16 (Ch. 132.890) Tel. Rec.	
(Ch. 132.890) Tel. Rec. 194-16, 195-16 (Ch.	130—12
(See Model 179-16) 210 (Ch. 132.880)	130 109—12
215 (Ch. 528.174) 220 (Ch. 528.173)	117-13 110-13 107-8
239 (Ch. 548.360-1). 245 (Ch. 548.358-1).	115—12 107—9
246 (Ch. 137.906) 249 (Ch. 548.360-1) (See Model 239)	111—14 115
1300 (Ch. 319.200), 1300-1 (Ch. 319.200-1)	90 —10
1301 (Ch. 319.190) 1304 (Ch. 185.706) 6002 (Ch. 132.818)	91 —11 5 —35
6011 (Ch. 132.816), 6012 (Ch. 132.816A)	15-27
6016 (Ch. 132.820) 6050 (Ch. 132.825-4) 6051 (Ch. 110.451).	27—24 15—28
6052 (Ch. 110.452) 6071 (Ch. 132.826-1)	13-29 15-29 13-30
6092 (Ch. 101.672-1B), 6093 (Ch. 101.672-1A)	10-28
SILVERIONE-Cont. 179-16, 180-16 (Ch. 132.890) Tel. Rec. (See Model 179-16) 210 (Ch. 132.880) 215 (Ch. 528.174) 225 (Ch. 528.174) 226 (Ch. 528.174) 227 (Ch. 528.174) 239 (Ch. 548.360-1) 245 (Ch. 548.360-1) 245 (Ch. 548.360-1) 246 (Ch. 137.906) 247 (Ch. 548.360-1) 1300-1 (Ch. 139.200-1) 1300-1 (Ch. 139.200-1) 1300-1 (Ch. 139.200-1) 1300-1 (Ch. 139.200-1) 1300-1 (Ch. 132.818) 6012 (Ch. 132.818) 6012 (Ch. 132.818) 6012 (Ch. 132.818) 6052 (Ch. 132.818) 6053 (Ch. 132.825-4) 6053 (Ch. 101.6422) 6052 (Ch. 101.6424) 6052 (Ch. 101.6424) 6052 (Ch. 101.6424)	6—29 7
6105 (Ch. 101.622-2B) 6106A (Ch. 101.662-4E).	7-26 29-23
6111 (Ch. 101.662-3C) (See Model 6105) 6111A (Ch. 101.662-5F)	7
(See Model 6106A) 6200A (Ch. 101.800-3) 6200A (Ch. 101.800-1) 6203 (Ch. 101.800A) (See Model 6200A)	29 6512
6200A (Ch. 101.800-1) 6203 (Ch. 101.800A) (See Model 6200A)	929 9
6220, 6220A (Ch. Nos. 101.801, 101.801-1A) .	930
6230 (Ch. 101.802), 6230 (Ch. 101.802-1) 6285A (Ch. 101.666-1B).	11-21 20-28
6290 (Ch. 101.677-B) 6293 (Ch. 528.6293-2)	20-29 99-16
6230 [Ch. 101.802], 6230 (Ch. 101.802-1). 6285A (Ch. 101.607-18]. 6290 [Ch. 101.677-8]. 6293 [Ch. 528.6293-2]. 6485 [Ch. 139.150-1], Ch. 139.150-1], Power Shifter,	98-12
7010	15-30
7012	:
7016 7017 7020 (See Model 7021)	16
7021 (Ch. 101.807, 101.807A1	16-31
7025 (Ch. 132.807-2) 7054 (Ch. 101.808) 7070 (Ch. 101.817) 7080 (Ch. 101.809)	29-24 15-31 30-26
7080, 7080A (Ch.	16—32
7085 (Ch. 101.814) 7086 (Ch. 110.466)	30-27
7090 (Ch. 101.810) 7095 (Ch. 101.826) (See Model 7115)	15—32 16
7100 (Ch. 101.811)	17—29 30
7102 (Ch. 101.814-1A), (See Model 7085). 7103 (Ch. 110.466-1) (See Model 7086) 7105, 7106	
7105, 7106 7111 (Ch. 434,140) 7115 (Ch. 101,825), 7116 (Ch. 101,825-1A), 7117 (Ch. 101,825-1B)	30 —28
7116 (Ch. 101.825-1A), 7117 (Ch. 101.825-1B)	16-33
7145 (Ch. 436.200) 7145 (Ch. 431.188),	23-21
7148A (Ch. 431,188-1) 7152 (Ch. 109.626) 7153 (Ch. 109.627)	23-22 25-26 26-30
7165 (Ch. 101.823-A, 1A), 7166 (Ch. 101.823,	10 00
7210 (Ch. 101.820) 7220 (Ch. 161.801-2C)	32-20
7116 (Ch. 101.825-18). 7117 (Ch. 101.825-26). 7145 (Ch. 435.200). 7145 (Ch. 430.200). 7145 (Ch. 431.188). 7145 (Ch. 431.188). 7132 (Ch. 431.188). 7133 (Ch. 431.188.1). 7135 (Ch. 109.627). 7136 (Ch. 101.823.4, 1A). 716 (Ch. 101.823.4, 1A). 710 (Ch. 101.823.4, 1A). 7210 (Ch. 101.820.1). 7220 (Ch. 101.802.2A) (See 6230)	31 —28
(See 6230) 7300 (Ch. 435.240)	11 45—22
7350 (Ch. 435.410) 7351 7352	3822 *
7353 (See Model 7350) 8000 (Ch. 132.838)	38 31-29
8004 (See Model 8003) 8005 (Ch. 132.839)	53 33-26
8010 (Ch. 132.840) 8011 (See Model 8010) 8020 (Ch. 132.841)	40-21 40 43-17
8021 (Ch. 132.868) 8022	70 —10
8024, 8025 (Ch. 478.206-1) 8050 (Ch. 101.813)	80
8051 (Ch. 101.839) 8052 (Ch. 101.808-1C)	49-19 68-15
478.206-1) 478.206-1) 8050 (Ch. 101.833) 8051 (Ch. 101.839) 8052 (Ch. 101.808-1C) 8053 (Ch. 101.808-1C) (See Model 8052) 8070 (Ch. 101.817-1A) (See Model 7070) 8071	68
(See Model 7070) 8071	30-26
8073 (Ch. 135.243) 8080 (Ch. 101.852)	34—19 84—9 52—20
101.809-1A) (See	
Model 7080) 8084, 8084A (Ch. 101,809-18) (See Model 7080)	58
Model 7080)	58

SILVERTONE-SPARTON

ILVERIONE-SPARION	
SILVERTONE-Cont.	SILVERTONE-Cont.
8086 (Ch. 101.814-5C) 61-18	Ch. 100.043 (See Model 133)*
8086A, 8086B (Ch. 101.814-6C) (See	Ch. 100.111 (See Model
Model 8086)	143A)
8092 *	161-16) 99/ Ch. 101.660-1A
(See Mode) / 119) 02	(See Model 6100) 6
8100 (Ch. 101.829) 51—19 8101. 8101A. 8101B.	Ch. 101.662-28, 101.662-2D, 101.662-3C (See Model 6105)
8100 (Ch. 101,829)5119 8101, 8101 A, 8101 B, 8101 C (Ch. 101,809-3C) (See Model 7080)58 8102 (Ch. 101,814-2B) (See Model 8086)61 8102A (Ch. 101,814-3B) (See Model 8086)61	(See Model 6105) 7
(See Model 7080) 58 8102 (Ch. 101.814-28)	Ch. 101.662-4E, 101.662- 5F (See Model 6106A) 29 Ch. 101.666-18
(See Model 8086) 61	Ch. 101.666-1B (See Model 6285A) 20
(See Model 8086) 61 81028 (Ch. 101.814-28)	Ch. 101.672-1A, 101.672-
	1B (See Model 6092) 10 Ch. 101.677B
8103 (Ch. 110.473) 56-21	(See Model 6290) 20
8104 (See Model 8086) 61 8105, 8105A	Ch. 101.773 {See Model 8127} 41
(Ch. 101.833) 35—20	Ch. 101.800-1, 101.800-
	Ch. 101.800-3
Model 8105)	(See Model 6200A) 65 Ch. 101.801, 101.801-1A
Model 8105)	(See Model 6220) 9
	Ch. 101.802, 101.802-1 (See Model 6230) 11 Ch. 101.807, 101.807A (See Model 7021) 16
8112, 8113 (See Model 8115)	Ch. 101.807, 101.807A
8115 (Ch. 101.825-3D), 8115A, B, C (Ch. 101.825-4), 8117 (Ch.	(See Model 7021) 16 Ch. 101.808
101.825-4), 8117 (Ch.	(See Model 7054) 15 Ch. 101.808-1C, 101.808-
101.825-3F), 8117 [Ch. 101.825-3F], 8118 (Ch. 101.825-3F), 8118A, B, C (Ch. 101.825-4) (See Model 7119)	1D (See Model 8052) 68
C (Ch. 101.825-4) (See Model 7119)	Ch. 101.809
	(See Model 7080) 16 Ch. 101.809-1A, B,
101.831A, Ch.	Ch. 101.809-1A, B, 101.809-2, 101.809-3C (See Model 7080) 58
Model 8127) 41	Ch. 101.810
Model 8127) 41 8127, A, B, C (Ch. 101.831A), 8128, A, B, C (Ch. 101.831), Wire	(See Model 7090) 15 Ch. 101.811
C (Ch. 101.831), Wire Recorder Amp. (Ch.	(See Model 7100) 17
101.773) 41 —20	Ch. 101.813 (See Model 8050) 13
101.773) 41-20 8130 Television Receiver. 49-21 8132 (Ch. 101.854)	(See Model 8050) 13 Ch. 101.814, 101.814-1A (See Model 7085) 30
lef. Kec 00	Ch. 101,814-28,
8133 (Ch. 101.829-1, Ch. 101.846) Tel. Rec.	101.814-3B, 101.814-5C, 101.814-6C (See
(See Model 8132)	Model 8086) 61 Ch. 101.817
8145 (Ch. 109.631) 45-23	(See Model 7070) 30
8148 (Ch. 109.632) 44—22 8149 (Ch. 109.633) 48—23	Ch. 101.819A (See Model 7226) 31
8149 (Ch. 109.633) 48-23 8150 (Ch. 109.634) 32-22 8152 (Ch. 109.635)	Ch. 101.820
(See Model 8153) 42	(See Model 7210) 32 Ch. 101.821
8153 (Ch. 109.635), 8153A (Ch. 109.635-1) 42-22	(See Model 8090) 49 Ch. 101.822, 101.822A
8153 (ch. 109.635.) 81534 (ch. 109.635-1) 42 22 8155 (ch. 463.155) 57 17 8160 (ch. 109.636), 81604 (ch. 109.636A). 50 17	(See Model 8270) 57
81604 (Ch. 109.636A). 50-17	Ch. 101.823, 101.823A, 101.823-1, 101.823-1A
8160 (ch. 109.636), 8160A (ch. 109.636A). 50—17 8168 (ch. 109.638)	Ch. 101.823, 101.823.4, 101.823-1, 101.823-1A (See Model 7166) 10 Ch. 101.825, 101.825-1A, 101.825, 101.825-1A,
(See Model 8168) 46	101.825-18 (See
[See model 6200A	Ch. 101.825-18 (See Model 7115)
[Ch 101 800.31] 65	3D, 101.825-3E,
8201 (See Model 6200A). 65 8210 (Ch. 101.820-1A) 71-13	101.825-3F, 101.825-4 (See Model 7119) 62 Ch. 101.829
8220, 8221 (Ch. 101.801-3D), 8222	Ch. 101.829 (See Model 8100) 51
(See 6220)	Ch. 101.829-1
8231 (See Model 8230). 59 8260 (Ch. 101.823-28)	(See Model 8132) 66 Ch. 101.831, 101.831A,
	Ch. 101.831, 101.831A, 101.831-1 (See Model 8127)
8270 (Ch. 101.822), 8270 A (Ch. 101.822A), 9000 (Ch. 132.857),, 65–13 9005, 9006 (Ch. 132.858), 72–11 9022 (Ch. 132.871),, 76–17 9054 (Ch. 132.871),, 76–17	Ch. 101.833
9000 (Ch. 132.857) 65-13	(See Model 8105) 35 Ch. 101.834
9005, 9006 (Ch. 132.858) 72-11 9022 (Ch. 132.871) 76-17	(See Model 8072) 34
9034 [Cn. 101.049] 03-10	Ch. 101.835 (See Model 8230) 59
9073, 9073A (Ch. 135.244), 9073B	Ch. 101.839 (See Model 8051) 49
135.244), 9073B (Ch. 135.244-1) 8310 9073C (Ch. 135.243-1)	Ch. 101.846
(See Model 90/3) 83	(See Model 8132) 66 Ch. 101.849
9082 (Ch. 135.245) (See Model 41)	(See Model 9054) 63 Ch. 101.850
9101 (Ch. 101,809-3C) (See Model 7080) 58	(See Model 9260) 51
9102 (See Model 7080). 58	Ch. 101.851, 101.851-1 (See Model 8107A) 64
	Ch. 101.852 (See Model 8080) 52
9107A (Ch. 101.851-1) (See Model 8107A) 64 9115 (Ch. 478.224), 9116	Ch. 101.854
(Ch. 478.221) Tel. Rec. 97-16	(See Model 8132) 66 Ch. 101.859
(Ch. 478.221) Tel. Rec. 97-16 9119, 9120 (Ch. 101.865) Tel. Rec. *	(See Model 9133) 95
9120A (Ch. 101.865-1)	Ch. 101.859-1, -2 (See Model 64)113
Tel. Rec	Ch. 101.864
Tel. Rec	(See Model 9122) 66 Ch. 101.865
9122 (Ch. 101.864) (See Model 8132)	(See Model 9119) * Ch. 101.865-1
9122A (Ch. 101.868) Tel. Rec. * 9123 (Ch. 110.499), 9124	(See Model 9120A) *
9123 (Ch. 110.499), 9124	Ch. 101.866 (See Model 9133) 95
(Ch. 110.499-1), Tel. Rec	Ch. 101.867
9125 (Ch. 478.252)	(See Model 9121) * Ch. 101.868
Tel. Rec. * 9125A (Ch. 478.253) Tel. Rec. (See Model 125)104	(See Model 9122A) *
Rec. (See Model 125)104 9126 (Ch. 101.499-2)	Ch. 109.626 (See Model 7152) 25
Tel. Rec. (See	Ch. 109.627
Model 9123)	Ch. 109.631
Tei, Rec *	(See Model 8145) 45 Ch. 109.632
9131 (Ch. 478.210) Tel. Rec	(See Model 8148) 44
9133, 9134 (Ch. 101.866,	Ch. 109.633 (See Model 8149) 48
Radio Ch. 101.859) Tel. Rec	Ch. 109.634
9153 (Ch. 435.417) 67-16 9161 (Ch. 548.358) 88-10	(See Model 8150) 32 Ch. 109.635, 109.635-1
9151 (Ch. 333.417)	Ch. 109.635, 109.635-1 (See Model 8153) 42 Ch. 109.636, 109.6364
9270 (Ch. 547.245) 82—11 9280 (Ch. 528.168) 94—9	Ch. 109.636, 109.636A (See Model 8160) 50

RTONE-Cont.	SILVERTONE-Cont.
0.043 Model 133) *	Ch. 109.638 (See Model 8168) 46
0.111 (See Model	Ch. 110.451, 110.452 (See Model 6051) 13
16)	Ch. 110.454 (See Model 6072) 13
1.660-1A Model 6100) 6	Ch. 110.466, 110.466-1 (See Model 7086) 27
1.662-28, 662-2D, 101.662-3C Model 6105) 7	Ch. 110.473 (See Model 8103) 56
1.002-4E, 101.002-	Ch. 110.499 (See Model 9123) 79
See Model 6106A) 29 1.666-1B Model 6285A) 20	Ch. 110.499-1 (See Model 9124) 79
Model 6285A) 20 1.672-1A, 101.672- See Model 6092) 10	Ch. 110.499-2 {See Model 9126} 7.9 Ch. 132.807-2
1.677B Model 6290) 20	(See Model 7025) 29 Ch. 132.816, 132.816A (See Model 6011)
1.773 Model 8127) 41	(See Model 6011) 15 Ch. 132.818
1.800-1, 101.800-	(See Model 6002) 5 Ch. 132.818-1
See Model 6200A). 9 1.800-3 Model 6200A) 65	(See Model 8003) 53 Ch. 132.820
1.801, 101.801-1A	(See Model 6016) 27 Ch. 132.825-4
1.802, 101.802-1 Model 6230	(See Model 6050) 15 Ch. 132.826-1
Model 7021) 16	(See Model 6071) 15 Ch. 132.838
Model 7054) 15 1.808-1C, 101.808-	(See Model 8000) 31 Ch. 132.839
See Model 8052) 68	(See Model 8005) 33 Ch. 132.840
1.809 Model 7080) 16	(See Model 8010) 40 Ch. 132.841 (See Model 8020) 43
1.809-1A, B, 809-2, 101.809-3C Model 7080)58	Ch. 132.858 (See Model 9005) 72
Model 7080) 58 1.810 Model 7090} 15	Ch. 132.868 (See Model 8021) 70
Model 7100) 17	Ch. 132.871 (See Model 9022) 76
1.813	Ch 132.875
1.814, 101.814-1A Model 70851 30	(See Model 9105) 89 Ch. 132.878 (See Model 1)
11.814-28, 814-38, 101.814-5C, 814-6C (See	Ch. 132.880 (See Model 210)
el 8086) 01	Ch. 132.882 (See Model 105)* Ch. 132.887
1.817 Model 7070) 30	Ch. 132,887 (See Model 51)
1.819A Model 7226) 31 1.820	(See Model 54)115 Ch. 132.889
Model 7210) 32	(See Model 106)
Model 8090) 49	(See Model 179-16)130 Cb 135.243
Model 8270] 57 1.823, 101.823A,	(See Model 8073) 84 Ch. 135.243-1
.823-1, 101.823-1A Model 7166) 10	Ch. 135.244, 135.244-1 (See Model 9073) 83 Ch. 135.244, 135.244-1 (See Model 9073) 83 Ch. 135.245
 Model 8270)	Ch. 135.245 (See Model 41)101
lei 7115) 16 11.825-2C, 101.825-	Ch 137 906
.825-3F, 101.825-4 Model 7119) 62	(See Model 246)111 Ch. 139.150, 139.150-1 (See Model 6685)15
01.829	Ch. 185.706 (See Model 1304)* Ch. 319.190
1.829-1	Ch. 319.190 (See Model 1301) 91
(1.831, 101.831A, .831-1 (See lel 8127) 41	Ch. 319, 200, 319, 200-1 91 Ch. 319, 200, 319, 200-1 (See Model 1300)
1.833	Ch. 431.188, 431.188-1 (See Model 7148) 23 Ch. 431.199
Model 8105) 35)1.834 e Model 8072) 34	
Model 8230) 59	(See Model 8130) 49 Ch. 434.140
)1.839 Model 8051) 49	(See Model 7111) 30 Ch. 435.240
1.846 Model 8132) 66	(See Model 7300) 45 Ch. 435.410
1.849 Model 9054) 63	(See Model 7350) 38 Ch. 435.417 (See Model 9153) 67
11.850 Model 9260) 51 01.851, 101.851-1	(See Model 9153) 67 Ch. 436.200 (See Model 7145) 23
Model 8107A) 64	Ch. 463.155 (See Model 8155) 57
Model 8080) 52 01.854	Ch. 478.206-1 (See Model 8024) 80
Model 8132) 66	Ch. 478.210 (See Model 9131) 84
Model 9133) 95 01.859-1, -2	Ch. 478.221 (See Model 9115) 97
e Model 64)113 01.864	Ch. 478.224 (See Model 9115) 97
e Model 9122) 66)1.865	Ch. 478.252 (See Model 9125)* Ch. 478.253
1.865-1	(See Model 125)104 Ch. 478.257
e Model 9120A} * 01.866 9 Model 9133} 95	(See Model 125)104 Ch. 478.289
Model 9133) *	(See Model 112)118 Ch. 478.309
)1.868 • Model 9122A)*	(See Model 120) 115 Ch. 478.311
9.626 Model 7152) 25	(See Model 120)115 Ch. 528.168
)9.627 e Model 7153) 26	(See Model 9280) 94 Ch. 528.171-1 (See Model 225) 107
09.631 9 Model 8145) 45	Ch. 528.171-1 (See Model 225)
09.632 a Model 8148} 44	Ch. 528.174 (See Model 215)
09.633 e Model 8149) 48	Ch. 528.6293-2 (See Model 6293) 99
09.634 • Model 8150) 32	Ch. 528.6295 (See Model 6295)
09.635, 109.635-1 e Model 8153} 42	Ch. 547.245 (See Model 9270) 82 Ch. 548.358
09.636, 109.636A e Model 8160) 50	(See Model 9161) 88

SILVERTONE-Cont.
Ch. 548.358-1 (See
Model 245)
(See Model 33)111 Ch. 549.100, 549.100.1 (See Model 101)102
(See Model 138)
Ch. 549.100-4 (See Model 160-12)
SIMPLON CA-5 22—27
CA-5
SKYRIDER (See Hallicrafters)
SKYROVER N5-RD-250 (9022-N).
N5-RD-250 (9022-N), N5-RD-251 (9022-H) 6-31 N5-RD295 (Ch. 5A7) 21-30
SKY WEIGHT
81B 20—30 82 13—13 SONOGRAPH
BL100
BL100) 122
SONORA RBU-176
RB-207 (See Model RB-176) 5 RCU-208 5-30 RDU-209 3-29
RDU-209 3—29 RET-210 24—24 RGMF-212, RGMF-230 27—26
RKRU-215 (Ch. RKRU) 9-31
(See Model RMR-219) 19
ROU.222 8-23
RWFU-238 23-24 RX-223 19-29 WAU-243 27-27 WBRU-239 32-23 WCU-246 36-22
WBR0-239 32-23 WCU-246 36-22 WDU-233 25-27 WDU-249 37-20
WBL 252 36-23
WLRU-254A
WIR1.21941 37
WLRU-245A (See Model WLRU-219A) 37 WXTU-700, WXTUA-700A
Tei. Rec
41_21
171
172 (See Model 171)109 302, 303 Tel. Rec97A-13 306108-11
306 108—11 401 47—21 402A (See Model RMR-219) 19 402F (See Model
402F (See Model WLRU-219A)
SOUND, INC. "Intersound" 7-27
MBOP3, MBOP30, MB6R4 35—21 MB7E3 28—31 MB7E8 26—24 SR2 28—32
SPARK5-WITHINGTON
(See Sparton) SPARTON
4AW17 (Ch. 417) 50—18 4AW17-A (Ch. 417A) 49—22 5AH06, 5A106 (See
5AH06, 5A106 (See
5AI16 (Ch. 5-16) 30-29 5AM26-PS (Ch. 5-26-PS) . 5-17 5AW06 (Ch. 5-06) 4-17
Model 5AW06)
6AM06 (Ch. 6-06) 34-21 6AM26 (See Model
6AW26PA
7AM46 (Ch. 7-46) 1-31
78W46PA, 8AM46 (See Model 7AM46) 1
200407A, 0AMAG 1 (See Model 7AM4G) 1 10AB76-PA, 10AM76-PA, 10BM76-PA, 10BM76-PA 1 10BW76-PA 1 10D, 101 1 102, 103, 104 1 105, 566 1 106, 500 3
108W76-PA (Ch. 10-76PA) 15-34 100 101 (Ch. 5A7) 38-23
102, 103, 104 (See Model 100) 38 121 (Ch. 819)
121 (Ch. 8L9)
(ch. 8L9) 57—19 122 (See Model 121) 57 130, 132, 135, 139 (ch. 5A10)
(Ch. 5A10) 9410 141 (See Model 121)57 141A (Ch. 8L10)926 141Xx, 142XX (Ch. 8W10) 12612 142 (See Model 121) 57
141xx, 142xx (cn. 126—12 8w10 126—12 142 (See Model 121)
150, 151, 152, 155 (Ch. 4E10) 91—12
1000 1001 1003
1005 1006 1007 1008
1015 (See Model
1020, 1021, 1023
(See Model 1000) 60

SPARTON-Cont.

SPARTON-Cont. Ch. 8.46 (See Model 8AM46).... 1 Ch. 8.57 (See Model 1005) 29 Ch. 918 (See Model 1035) 62 Ch. 918A SPIEGEL (See Aircastle) STARK 410 40-22 1010 88-2 1020 89-5 STARRETT
 STARRETT
 101—12

 Gayham Tel, Rec.
 101—12

 Henry Hudson, Henry Parks
 2—7

 John Hancok Tel, Rec.
 96—10

 Nathan Hale Tel, Rec.
 97—12

 Robert E, Lee Tel, Rec.
 (See Model Henry

 L See Model Henry
 92

 Circle Man
 2
 STEELMAN
 200
 23—25

 303
 19—31

 350, 351
 21—31
 STEWART-WARNER 9100F, 9100G, 9100H Tel. Rec. 75-15 9103.8,-C, 9104.4,-B,-C Tel. Rec. 105-10 91054.8 Tel. Rec. 118-10 91054.8 Tel. Rec. 118-10 91084.8, 91094.8 Tel. Rec. (See Model 9106A) 118 91134 Tel. Rec. (See Model 9106A)...118

STEWART-WARNER-Cont.
 STEWARI-WARNEK-Cont.

 9120-A. -B. -C. -D. -E. -F

 Teil. Rec.
 119-1A

 9151-A.
 106-14

 9152-A.B., -C
 102-14

 9152-A.B., -C
 102-14

 9152-A.B., -C
 102-14

 9153-A.
 108-12

 9200-A.C. -D. -FA.
 108-12

 9200-A.C. -D. -FA.
 132-13
 STRATOVOX 579-1-58A 6-32 STROMBERG-CARLSON

 STROMBERG-CARLSON

 AM-43

 AM-44

 AM-43

 AP-50

 130—14

 AP-50

 AR-37

 130—14

 AP-50

 AM-43

 AP-50

 130—14

 AP-50

 AW-37

 128—11

 AU-32

 133—12

 AU-33

 AU-34

 132—14

 AV-36

 AU-37

 AU-38

 126—13

 TC-10 Tel. Rec.

 70-17

 TC-125 Tel. Rec.

 97—17

 TC-125 Tel. Rec.

 97-17

 TC-125 Tel. Rec.

 120 Tel. Rec.

 120 Tel. Rec.

 1212 Coll Ch. TV-122)

 68-16

 12-23

 117 Series Tel. Rec.

 122

 1135-PFM, 1135-PFM, 1135-PFM, 1135-PFM, 1135-PFW, [Series 10-11] 1202 (Series 10)... 1204 (Ch. 112021). 1210M2-W, 1210M2-W, 1210PFM, 1210PGW [Series 10-11]... 1220 Series 1235 Series 1235 Series 1400 (See Model 1200)... 1409PFM, 1409PG-M, 1409PG-W, 1500 STUDEBAXER 57-20 55-21 34-22 37-23 50-19 49-23 57 58-23 62-20 132-15 133-13 STUDEBAKER S-4624, S-4625 21-32 S-4626, S-4627 19-32 SUPREME (Lipon)
 Survey
 68—17

 711
 68—17

 712S
 63—17

 733
 60—19

 738LP
 64—13

 750
 55—22
 SYLVANIA

SYLVANIA-Cont.	TELE-KING-Cont.
1-245-1, 1-246-1 (Ch.	210 Tel. Rec 310 Tel. Rec
1-245-1, 1-246-1 (Ch. 1-186) Tel. Rec. (See Model 1-125-1)113	310 Tel. Rec 410 Tel. Rec
1-247 (Ch. 1-168) [e]. Rec.	410 Tel. Rec 416 Tel. Rec 510 Tel. Rec.
(See Model 1-090) 99	510 Tel. Rec. {See Model 410}.
1-247-1 (Ch. I-231) Tel, Rec	512 Tel. Rec.
124/-1 (Ch. 1-331) Tel, Rec	(See Model 410). 612 Tel. Rec.
(Ch. 1-215)	(See Model 410). 710 Tel. Rec.
Tel. Rec. (See	710 Tel. Rec.
Model 7110X}	{See Model 410}. 712 Tel. Rec.
model / 1/0/ 124 5108, 510H, 510W (Ch. 1-215) (See Model 1-250)103 5408, 540H, 540M119—11 1210X (Ch. 1-381) 119—11 1210X (Ch. 1-381) 128—16 2221M (Ch. 1-387) *	(See Model 410).
5408, 540H, 540M119-11	716 Tel. Rec. (See Model 162).
1210X (Ch. 1-381)	(See Model 162). 816-3CR Tel. Rec.
Tel. Rec	916C Tel. Rec. {See Model 162}. 919, 919CAF Tel. Re
	919, 919CAF Tel. Re
4120M (Ch. 1-260) Tel. Rec	TELEQUIP
4130B, 4130E, 4130M,	16TR, 19T, 19TR
Tel Kec (See Model	TELEQUIP Ch. 12TR, 14T, 14TR, 16TR, 19T, 19TR Tel. Rec 5135, 5136, 5140A
4120M)	5135, 5136, 5140A
Tel. Rec	TELESONIC (Medco 1635
4120M)	1636
Bul. 17	1642
Bul. 17	TELE-TONE
Rec. (See Model 5130B) 120 5140B. M (Ch. 1-290)	TV149 Television Per
Tel. Rec. Prod. Chge.	TV-170 Tel. Rec TV-208 Tel. Rec TV-208 Tel. Rec TV-209 Tel. Rec TV-209 Tel. Rec
Bul. 17	TV208TR Tel. Rec
	IV-209 Tel. Rec. (See Model TV-24
6110X (Ch. 1-261) Tel. Rec. (See Model	(See Model TV-24 TV-210 Tel. Rec TV-220 Tel. Rec (See Model TV208 TV-245, 246 Tel. R
4120441 124	See Model TV208
	TV-245, 246 Tel. Re TV-249 Television Re
(See Model 4120M) 124	TV-249 Television Re TV-250 Tel Rec
6130B, 6130M, 6130W	TV-254 Tel. Rec.
(See Model 4120M) 124	(See Model TV-250 TV-255, TV-256
[Cn. 1-201] 161, Rec. [See Model 4120M]124 61308, 6130M, 6130W (Ch. 1-261) Tel. Rec. (See Model 4120M]124 6140M, W (Ch. 1-271) Tel. Rec. [See Model 51308) 120 7110X (Ch. 1-366, 1-366-66) Tel, Rec. [See Model 4120M]24 7110X8 (Ch. 1-441)	TV-249 Television R. TV-250 Tel. Rec (See Model TV-250 TV-255, TV-256 (Ch. TS) Tel. Rec TV259 Tel. Rec. (See Model TV248 TV-282 Tel. Rec (See Model TV-283
7110X (Ch. 1-366,	See Model TV249
1-366-66) Tel. Rec. (See	TV-282 Tel. Rec
7110XB (Ch. 1-441)	TV-283 Tel. Rec. (See Model TV-284
Tel. Rec	TV-284 Tel. Rec
Ter, Nec. (See	TV-285 Tel. Rec TV-286 287 288 Te
Model 5150M)131 7111M (Ch. 1-441)	Rec. (See Model T
Tel. Rec	TV-300, TV-301 (Ch. TAB) Tel. Rec
7111M (Ch. 1-441) Tel. Rec	TV-283 Tel. Rec. (See Model TV-281 TV-284 Tel. Rec TV-285 Tel. Rec TV-285 Tel. Rec TV-286, 287, 288 Te Rec. (See Model Tel. Rec. TAB) Tel. Rec. TAB) Tel. Rec. TAB, TV-300, TV-301 (Ch. TW) Tel. Rec TAB) Tel. Rec. (See Model TV-300 (See Model TV-300)
Tel. Rec. (See Model 4120M)	(Ch. TW) Tel. Rec TV-304, TV-305 (Ch.
71208, 7120M, 7120W	TAB) Tel. Rec.
Tel. Rec. (See Model	[See Model 1V-300 TV-304, TV-305 (Ch
4120M)	TV-304, TV-305 (Ch Tel. Rec. (See Model TV-300)
Tel. Rec. (See	TV-306, TV-300)
Tel. Rec. (See Model 5150M)	TV-306, TV-307 (Ch. TY, TZ) Tel. Rec. TV-308 (Ch. TAC) Tel. Rec.
(Ch. 1-366, 1-366-66)	TV-308 (Ch. TAC)
Tel. Rec. (See Model 4120M)	Tel. Rec.
7130MFA (Ch. 1-442)	
Tel. Rec. (See	Tel. Rec. TV-315 (Ch. TAA, T/ Tel. Rec.
Model 5150M)	Tel. Rec TV-317 Tel. Rec
5130B)	Tel. Rec. TV-317 Tel. Rec. TV318 (Ch. TAM) Tel. Rec.
7140MA, 7140WA	TV322, TV323 (Ch. Tel. Rec. (See Mo
(Ch. 1-437) Tel. Rec. (See Model 5150M)131	Tel. Rec. (See Mo
7140MA, 7140WA (Ch. 1-437) Tel. Rec. (See Model 5150M)131 7150M (Ch. 1-357) Tel. Rec. (See Model 5150M) 131	TV318) TV324, TV325, TV32
Model 5150M)	TAP-2) Tel Rec.
7160B (Ch. 1-357)	TV328, TV329 (Ch.
7160B (Ch. 1-357) Tel. Rec. (See Model 5150M)	TV324, TV325, TV32 (Ch. TAP, TAP-1 TAP-2) Tel. Rec. TV328, TV329 (Ch. TAP-1, TAP-2) Tel Rec. (See Model
Ch. 1-139 (See	
Ch. 1-168 (See	TV335, TV336 (Ch. TAP-1, TAP-2) Te (See Model TV32)
Model 1-090)	(See Model TV32
(See Model 1-125-1)113	TAP-2) Tel. Rec.
Ch. 1-215 (See Model 1-250)103	(See Model TV32: TV340 (Ch. TAP, TA TAP-2) Tel. Rec. Model TV324) TV345 (Ch. TAP, TA TAP-2) Tel. Rec. Model TV324) TV-352 Tel. Rec. (See Model TV-32 TV.358 TV.359
Ch. 1-260 (See Model	TAP-2) Tel. Rec.
4120M)	Model TV324) TV-352 Tel Rec
4120M)	(See Model TV-32
Ch. 1-271 (See Model 5130B)120	(See Model TV-32
Ch. 1-274 (See	(Ch. Series A) 109 (Ch. Series J) 109 (Ch. Series J)
	109 (Ch. Series A).
(See Model 5130B)120 Ch. 1-357 (See Model 5150M)131	109 (Ch. Series J) 110 (See Model 117 111, 113 (See Model 117-A (Ch. Series 119, 120 (See Model 117-A) 122 (See Model 123 (See Model)
Model 5150M)	117-A (Ch. Series '
Ch. 1-366, 1-366-66 (See Model 4120M)	119, 120 (See Mode
Ch. 1-381 (See	122, 123 (See Mod
Ch. 1-381 (See Model 1210X)	122, 123 (See Model 117 125 (See Model 100 126 (See Model 117 127, 130, 131 (See Model 100) 132 (See Model 117
Model 5150M)	126 (See Model 117
Ch. 1-442 (See Model 5150M)	127, 130, 131 (See Model 100)
TELECHRON	132 (See Model 117
8H67 ''Musalarm'' 44-23	133
TELECOIN	133 134 135 138 (Ch. Series N) 139, 140, 141 (Ch.
M5TS4 25—28	138 (Ch. Series N) 139, 140, 141 (Ch.
TELE-KING	H) (See Model 13.
C716X Tel. Rec. *	142, 143, 144 (See Model 145)
C716X Tel, Rec. * 1516 Tel, Rec. * 16CD3CR Tel, Rec. * 116 Tel, Rec. * 116, 116C, 117, 117C, 117CA, 117CAF, 117RO T17CAF, 117CAF, 117RO	145 /Ch Series "P
116, 116C, 117, 117C.	149 (Ch. Series 5) . 149 (Ch. Series H)
117CA, 117CAF, 117RO	(See Model 135)
162 Tel. Rec	148 (Ch. Series S) . 149 (Ch. Series S) . (See Model 135) 150 (Ch. Series T) 151 (Ch. Series S)
201, 202 Tel. Rec. 131-16	(See Model 148)

TELE-KING-Cont. 210 Tel. Rec. 310 Tel. Rec. 410 Tel. Rec. 410 Tel. Rec. 510 Tel. Rec. (See Model 410) 512 Tel. Rec. (See Model 410) 512 Tel. Rec. (See Model 410) 512 Tel. Rec. (See Model 410) 510 Tel. Rec. (See Model 40) 510 Tel. Rec. (See Model 40) 510 Tel. Rec. (See Model 40) 510 Tel. Rec. (See Model 162) 129 510-3CR Tel. Rec. * 916 Crel. Rec. * 916 Tel. Rec. * 510 Tel. Rec. * 916 Crel. Rec. * 916 Crel. Rec. TELESONIC (Medco) 1635 20—22 1636 21—33 1642 20—23 1643 21—34 1630 21---33 1642 20--23 1643 20--23 16443 20--23 16443 20--23 16443 20--23 16443 20--23 17 17 1644 20--23 17 1645 17 1646 17 1647 17 1648 Tel. Rec. 125-12 TV-315 (Ch. TAA, TAB)

SPARTON-TEMPLE

TELE-TONE-Cont.

 Part Provession.

 152 (Ch. Series R)

 (See Model 145)

 156 (Ch. Series H)

 (See Model 145)

 157 (Ch. Series H)

 (See Model 145)

 157 (Ch. Series AE)

 157 (Ch. Series AE)

 158 (Ch. Series AA)

 159 (Ch. Series AA)

 156 (Ch. Series AG)

 156 (Ch. Series AG)

 156 (Ch. Series AG)

 157 (Ch. Series CI)

 (See Model 150)

 157 (Ch. Series CI)

 (See Model 150)

 157 (Ch. Series CI)

 (See Model 150)

 157 (Ch. Series CI)

 158 (Ch. Series AI)

 152 (Ch. Series AI)

 TELEVOX
 22-29

 PP
 20-32

 27K-W
 20-33

 27-P-T
 22-28
 TEL-VAR (See Audor) TEMPLE

45

TELE-TONE—WESTINGHOUSE

TELE-TONE-WESTINGHO	JUSE
TELE-TONE-Cont. G-1430	TRUETONE-Cont. D1998A Tel. Rec
G-4108 [See Mode]	D2017, D2018
G-418)	D2020 D2025A (Fact. Mod. 26A95-906)
G-721, G-722, G-723) 24 H-411 47—23	26A95-906} D2027A D2050A Tel. Rec.
H-521 (See Model G-521) 28 H-622 (See Model G-622). 44 H-727 (See Model G-725) 34	D2050A Tel. Rec. D2603 (Factory No.
H-727 (See Model G-725) 34	D2604
TV-1776, TV-1777, TV-1778, TV-1779 Tel. Rec	D2605 (Factory Mod 2AW2) D2606
TEMPOTONE	D2612 (Code SW-90
500 E Series 2-8	D2613 D2615 (Factory
TEMPLETONE (See Temple) THORDARSON	D2616 (Factory
T-30W08A 8-31	Model 6D117) D2616-B
T-31W10A	D2619 (Factory No. D2620
T-31W50A	D2621 D2622
T-32W00, T-32W10 76—18	D2423
TONE PAK AC8HF	D2624 (Factory 27D1 D2626 (Fact. No. 45 D2630 (Factory
TRAD	27014-002 issue
T-20, A Tel. Rec133—14 TRANSVISION	D2634 D2640 (Factory No.
Chassis Model A Tel. Rec. 107-11	D2642 D2644 (Factory No.
Chassis A-3 Tel. Rec130—15 WRS-3 Tel. Rec112—10	D2645 D2661 (Factory 4B19
TRANSVUE	D2663 (Ch. 4C1) D2665 (Factory 4B1)
601 (Ch. 16AX23, 25, 26) Tel. Rec.	Series A)
610 (Ch. 16AX23, 25, 26) Tel. Rec *	D2692 D2709 (Factory No.
TRAV-LER	D2710 (Factory No. 24D22-630BR)
10T Tel. Rec	D2718 (Factory No. 227D14-638IU)
127 Tel. Rec. (See Model 107) 86	D2743 D2745 (See Model D2748 (Ch. 7156)
121 1et. Kec. (See Model 107) 14850, A, 14C50, A Tel. Rec. (See Model 12L50)	D2748 (Ch. 7156) D2806, D2807 (Fact Model 181)
Model 12150)	Model 181) D2810 (Factory No. 24D24-730BB)
Model 12L50)	D2815
Tel. Rec. (See Model 12L50)108	D2819 (Factory No. 26A82-738)
16T Tel. Rec.	D2851 D2906 (Factory No.
5000 (See Model 50001) 11	D2910 D2919 (Fact. No. 6)
5002 Series (Ch. 109) 12-28	D2963 D2982 Tel. Rec
5007, 5008, 5009 (Ch. 104) 136	D2982 Tel. Rec D2985 Tel. Rec
5010, 5011, 5012 (Ch. 105) 2—5	D2987 (See Model 1
5015	Tel. Rec. D2988, D2989 Tel.
5019 23-30 5020 (Ch. 800) 11-28 5021 43-20 5022 101-14 5027 31-30 5028 34-24	D2988, D2989 Tel. D2990 Tel. Rec. D3615 (Factory 25BD D3619 (Factory 5P11 D3619 (Factory 5P11
5022	D3619 (Factory 5P11 D3630, D3630N D3720
5028	D3720 D3721 (Factory 110
5029 33—29 5030, 5031 32—25 5036 54—19	D3722 (Fact. No. 4 D3809 (Factory No.
5049	D3810 D3811 (Fact. No.
5054	114BXH) D3840
5049 45-24 5051 32-26 5054 36-26 5056-A 90-12 5060, 5061 116-11 5066 42-24 6040 49-25 6050 56-23	D3910 (Fact. Model 140611)
6040 49-25 6050 56-23	D4620 (Factory No. D4630 (Factory 260
6050 56—23 7000, 7001 59—21 7003 (Ch. 501) 12—29 7014 (See Model 7000) 59	D4818 (Fact No
7000, 7001 59_21 7003 (Ch. 501) 12_29 7014 (See Model 7000) 59 7016, 7017 84	D4832 (Fact. No. 25C22-82)
7016, 7017	D4842 (Fact. No. 26C21-81)
Chassis 104	2D1088A Tel. Rec. 2D1088B Tel. Rec.
(See Model 5007) 1 Chassis 105 (See Model 5010) 2	2D1089A Tel. Rec 2D1089B Tel. Rec
Chassis 109	2D1093A, 2D1094A
(See Model 5002) 12 Chassis 501	2D1095 Tel. Rec. 2D2052 Tel. Rec.
(See Model 7003) 12 Chassis 800	(See Model 2D109 2D2053 Tel. Rec.
(See Model 5021) 11 TRELA	ULTRADYNE
HW301 14-28	L-46
TRUETONE	UNITED MOTORS (See Delco or B Chevrolet, Oli and Pontiac)
D1034B, C (See Model D1046A)102	and Pontiac)
D1046A	U. S. TELEVISION
(See Model D1046A)102 D1090 Tel. Rec	C-12923P Tel. Rec. C16030 Tel. Rec. C19031 Tel. Rec.
D1612 28-34	(See Model C160) CFM-12823P Tel. Re
D1644 12-30 D1645 (Factory 26476-650) 6-33	CFM-12823P Tel. Re CFM-15925 Tel. Rec. CFM-16926 Tel. Rec.
D1747, D1748	CFM-16926 Tel. Rec. K-25936 Tel. Rec
D1835 (Factory Model 25A86-856)	K-30936 Tel. Rec KFM-25-PIC Tel. Re
25A86-856)	CFM-16926 Tel. Rec K-35936 Tel. Rec KFM-25-PIC Tel. Re KFM-25836 Tel. Rec. KFM-30836 Tel. Rec. KFK-15933 Tel. Rec. KRV-15831P Tel. Re KRV-15831P Tel. Re
	KRF-15933 Tel. Rec. KRV-12831P Tel. Rev
D18464 B C 40_23	KRV-15831P Tel. Rec T-3X4836 Tel. Rec
D1030 (Series A) 31-23	and a start and a start in a start
D1950, D1951 (See	1-525L Let Rec
Model D1850)	T621M, T621P Tel. T-10823 Tel. Rec T 10925 Tel. Rec
74F22) Tel Rec 69-13	T-10925 Tel. Rec T-12823 Tel. Rec T-12923P Tel. Rec
D 994 lel. Rec	T-15823 Tel, Rec
(See Model D2983) 68	T-15925 Tel. Rec T16030 Tel. Rec.
D1997A Tel. Rec *	(See Model C1603

19	TRUETONE-Cont. D1998A Tel Rec.	U. S. TELEVISION-Cont. T19031 Tel. Rec.
	D1998A Tel. Rec	(See Model C16030) 994
	D2025A (Fact. Mod. 26A95-906) 83-14	5A16, 5B16, 5C16 (See Model 5C66 Early). 17 5A66, 5B66, 5C66,
-23	D2027A	5D66MPA 24-
	D2603 (Factory No. 461) . 13-33 D2604 13-34	8-16M (Dumbarton) 26-
	26495 (rdct, mod. 83—14 26495 (906) 83—14 D2050A Tel. 86 D2003 (Factory No. 461) 13—34 D2005 (Factory Model 13—34 D2005 (Factory Model 9—34 D2005 (Factory Model 9—34 D2006 (540 Stream) (540 S	UNITONE 88 5-
-16		V-M
8	D2613	1001-A 10- VAN-CAMP
	Model 6D110) 2-18	576-1-6A 7-
31 30	Model 6D117) 10—32 D2616-B	VIDEO CORP. OF AMERICA (See Videola)
		VIDEODYNE
_33 _34	D2620 1—28 D2621	10FM, 10TV, 12FM, 12TV Tel. Rec
-18	D2621 4-32 D2622 14-30 D2623 11-29 D2624 (Factory 27D14-600) 2-6 D2626 (Foct. No. 457-2) 52-22	VIDEOLA VS-160, VS-161 Tel. Rec 92-
-28		VS-160, VS-161 Tet. Rec. 92- VS-165, VS-166, VS-167, VS-168 Tel. Rec. (See Model VS-160) 92
_14	27D14-602 issue A) 1-10 D2634	(See Model VS-160) 92 VIEWTONE
		RC-201A, RRC-201 11-
-11 -15	D2644 (Factory No. 101C) 11-30 D2645 4-39 D2661 (Factory 4819) 2-23	VIZ RS-1
-10	D2663 (Ch. 4C1) 11-31	VOGUE
	D2665 (Factory 4Bt 14 Series A)	532 A-P
	D2709 (Factory No. 470) . 27-30	WARWICK (See Clarion) WATTERSON
	D2710 (Factory No. 24D22-630BR)	ARC-4591A 16- PA-4585, APA-4587 3-
_11 _13	D2718 (Factory No. 227D14-638IU) 23—32 D2743 25—29	RC-4581 16-
	D2743 25-29 D2745 (See Model D1645) 6 D2748 (Ch. 7156) 26-27	4582 6- 4782 24 4790 16- 4800 43-
	D2800, D280/ (Factory	4790
	Model 181)	
	24D24-730BB) 36—27 D2815 48—25	WEBSTER-CHICAGO 66-1A 34 100-608 121 100-621 113 100-621 113- 100-621 113- 101-1 55- 288 117- 362 105- 760 112- 762 (See Model 362) 105-
	D2819 (Factory No. 26A82-738) 35-24	100-621
	26A82-738)	288
-27 -28	D2910 65-16 D2919 (Fact. No. 6DF21) 59-22 D2963 73-13 D2982 Tel. Rec. 73-13	760 112-
-36	D2982 Tel. Rec	WEBSTER (Telehome)
	D2983 Tel, Rec	W606M
-5 -25 -30	Tel. Rec. 69 D2988, D2989 Tel. Rec. *	WESTERN AUTO (See Trueto
-28 -20 -14 -30	D2988, D2989 Tel. Rec	WESTINGHOUSE H-104, H-105
-14 -30	D3619 (Factory 5P110) 10—33 D3630, D3630N 19—33	WESTINGHOUSE H-104, H-105
24 29 25 19 24	D3720	
-25 -19	D3722 (Fact. No. 472) 51-24 D3809 (Factory No. 178) . 43-22	(See Model H-117) 11
-24 -26 -26	D3811 (Fact. No.	H-117, H-119 11-
-26 -12 -11	114BXH)	H-122 H-122A, H-122B (See Model H-122)6
-24	D3910 (Fact. Model 140611)	H-125, H-126 3- H-130 (See Model H-122) 6
-24 -25 -23 -21 -29	D4620 (Factory No. 5C12) 20-28 D4630 (Factory 26C19-61) 7-28	H-125, H-126 3- H-130 (See Model H-122) 6 H-133 14- H-137 (See Model H-138) 6 H-138 6-
-29	D4818 (Fact, No. 134DX) 43-20 D4832 (Fact, No. 25622 23) 47 25	H-14/
-11 -13 -11	26C21-81)	H-148A (See Model H-148) 15
-11	2D1088A Tel. Rec105-11	H-153, H-153A (Ch. V-2103)
	2D1088B Tel. Rec	
	Tel Rec 119_12	H-153 (See Model H-153) 35 H-157 (Ch. V-2122) 33- H-161 (Ch. V-2118) 34- H-162 (See Model H-117). 11 H-164 (Ch. V-2119.1) 26
	2D1095 Tel. Rec	H-162 (See Model H-117). 11 H-164 (Ch. V-2119-1) 36-
	(See Model 2D1095)134 2D2053 Tel. Rec	H-165 32- H-166, H-167
	ULTRADYNE	(See Model H-164)
-28	L-46 4-21 UNITED MOTORS SERVICE	H-168, H-168A, H-168B (Ch. V-2118) (See Model H-161)
	(See Delco or Buick, Chevrolet, Oldsmobile	Model H-161)
-15	and Pontiac)	(Ch. V-2103) (See Model H-153) 35
	U. S. TELEVISION C-12923P Tel. Rec *	Model H-153) 35 H-178 (Ch. V-2123) 35 H-181 Tel, Rec. * H-182 (Ch. V-2128) * (Ch. V-2128-1) 53 H-183, H-183A 48 H-184 (See Model H-153) 35 H-185 (Ch. V-2123), Y-2131, Y-2131-1) 54
~	C16030 Tel. Rec	H-182 (Ch. V-2128), (Ch. V-2128-1) 53-
-34		H-183, H-183A
-27	CFM-12023F Tel. Rec * CFM-15925 Tel. Rec * CFM-16926 Tel. Rec *	V-2131-1)
25 25	K-30936 Tel. Rec. * KFM-25-PIC Tel. Rec. *	H-186M, H-187 (Ch. V-2132)
-25	KFM-25836 Tel. Rec * KFM-30836 Tel. Rec *	(Ch. V-2132) 60- H-188 (Ch. V-2133) 51- H-190, H-191, H-191A (Ch. V-2134) 59- H-195 (See Model H-184) 54
-23	CFM-16926 Tel, Rec K-25936 Tel, Rec K-30936 Tel, Rec KFM-25.PIC Tel, Rec KFM-30836 Tel, Rec KFM-30836 Tel, Rec KR-15933 Tel, Rec KRV-12831P Tel, Rec KV-12831P Tel, Rec	H-195 (See Model H-184). 54 H-196 Tel. Rec
-31	KRV-15831P Tel. Rec * T-3X4836 Tel. Rec *	H-196A (CHV-2130-1) Tel. Rec. (See Model
-23	T502M, T502P Tel. Rec * T-507, T-507M Tel. Rec *	
	RW-13831P Tel. Rec* T-334836 Tel. Rec* T502M, T502P Tel. Rec* T-507, T-507M Tel. Rec* T-525L Tel. Rec* T621M, T621P Tel. Rec* T-10R23 Tel. Rec*	H196A (DX) (Ch. V-2130-11DX or V-2130-12DX) Tel. Rec. 84-
	T-10823 Tel. Rec 89-15 T-10925 Tel. Rec	H-198 (Ch. V-2137-2) 73- H-199 (Ch. V-2137-1) 69-
-13	T-10925 Tel. Rec	V-2130-120A) Tel. Keč. 84 H-198 (Ch. V-2137-2) 73- H-199 (Ch. V-2137-1) 69- H-202 (Ch. V-2137-1) 69- H-203 (Ch. V-2178-2) 50- H-204 (See Model H-202). 50
-11	T-15925 Tel. Rec	H-204 (See Model H-202). 50 H-207A (Ch. V-2130-1, V-2137) Tel. Rec. (See
	T16030 Tel. Rec. (See Model C16030) 99A	V-2137) Tel, Rec. (See Model H-196)

U. S. TELEVISION-Cont.	
0. 3. FEEVISION-CONT. 11903 1Fel. Rec. (See Model C16030) 99A 5A16, 5816, 5C16 (See Model 3C66 Early). 17 5A66, 5866, 5C66, 5D66MPA	1
5A16, 5B16, 5C16 (See Model 5C66 Early). 17 5A66 5B66, 5C66	
5C66 Early 17-9	
UNITONE	
88	
1001-A	
576-1-6A	
(See Videola) VIDEODYNE	l
10FM, 10TV, 12FM, 12TV Tel. Rec	
VIDEOLA	
VS-160, VS-161 Tel. Rec. 92-9 VS-165, VS-166, VS-167, VS-168 Tel. Rec. (See Model VS-160) 92	
VIEWTONE	ļ
RC-201A, RRC-201 1132	
RS-1 14—31 VOGUE	
532 A-P	
WARWICK (See Clarion) WATTERSON ARC-4591A	
PA-4585 APA-4587 3-2	Ľ
4582	
4782 24—31 4790 16—34 4800 43—23	
WEBSTER-CHICAGO	
34-26 100-608 121-14 100-621 113-11 130 119-13 161-1 55-23	
288	
362 105—12 760 112—12 762 (See Model 362)105	
WEBSTER (Telehome)	
604M 57-23 WESTERN AUTO (See Truetone)	'
WESTINGHOUSE	'
H-104A, H-105A, H-107A, H-108A	
H-108A	
(See Model H-117) 11 H-117, H-119 11-34 H-122 6-35	
H-122A, H-122B (See Model H-122) 6	
H-125, H-126	Ι,
H-138 6-36	
H-147	1
H-1480 (See Model H-148) 15 H-153, H-153A (Ch. V-2103)	1
H-155 (See Model H-153). 35	
H-157 (Ch. V-2122) 33—31 H-161 (Ch. V-2118) 34—27 H-162 (See Model H-117). 11 H-164 (Ch. V-2119-1) 36—28	,
H-164 (Ch. V-2119-1) 36—28 H-165	,
H-166, H-167 (See Model H-164) 36 H-168, H-168A, H-168B	ŀ
H-164 (Ch. V-2119-1) 36-28 H-165 32-29 H-165, H-167 (See Model H-164) 36 H-168, H-168A, H-168B (Ch. V-2118) (See Model H-161) 37 H-169 (Ch. V-2124-1) 37-24 H-171, H-171A, H-171C (Ch. V-2103) (See Model H-153) 35 H-178 (Ch. V-2123) 35 H-178 (Ch. V-2124) 35 H-178 (Ch. V-212	
(Ch. V-2118) (See Model H-161) 34 H-169 (Ch. V-2124-1) 37—24 H-171, H-171A, H-171C (Ch. V-2103) (See Model H-153) 35 H-181 Tel. Rec. 35—26	
H-191 Tel Per *	ŀ
H-182 (Ch. V-2128), (Ch. V-2128-1) 53-25	F
H-183, H-183A 48—26 H-184 (See Model H-153) 35 H-185 (Ch. V-2131, V-2131-1)	F
V-2131-1) 54—20 H-186M, H-187	F
V-2131-1)	F
(Ch. V-2134) 59—23 H-195 (See Model H-184), 54 H-196 Tel, Rec	H
H-196A (CHV-2130-1) Tel. Rec. (See Model H-196)	
H-196) 65 H196A (DX) (Ch. V-2130-11DX or	н
H-196) -, Good	н
V-2130-12DX) Tel. Rec. 84—13 H-198 (Ch. V-2137-2) 73—15 H-199 (Ch. V-2137-1) 69—16 H-202 (Ch. V-218-2) 50—22 H-203 (Ch. V-2137) 62—21	н
H-204 (See Model H-202). 50 H-207A (Ch. V-2130-1, V-2137) Tel. Rec. (See	н

	WESTINGHOUSE-Cont. H207A (DX) (Ch. V-2130-110X or V-2130-120X and Radio Ch. V-2137) Tol., Rec. (San decid w1284)	
-30 -9 -29	Ch. V-2137) Tel. Rec. (See Model H196A (DX))	
-26	Ch. V-2130-22DX and Kodio Ch. V-2137) Tel. Rec. [See Model H196A	
-34	(DX])	
-29		
	Model H-203)	
	V-2146-05, V-2146-45, V-2149-1) Tel. Rec 97A-14	
-15	H-217, H-217A (Ch. V-2146-1, V-2137, V-2149) Tel. Rec	
-9	H-217, H-217A (Ch. 2146-11DX, V-2137,	
	H-214, H-214A (Ch. V-2103-3)	
-32	V-2137, V-2149) Tel. Rec	
-31	H-223 (Ch. V-2150-01, V-2150-02) Tel. Rec 78-14	
-33 -32	H-225 (DX) (Ch. V-2130-31DX or V-2130-32DX) Tel. Rec.	
	(See model III) on	
-36	[DX])	
-35 -32 -34	H-2176)	
-31 -34 -23	V-2137-35 or V-2137-35, V-2149-2) Tel. Rec	
	H-242 (Ch. 2150-31) Tel. Rec. 97A-14 H-251 (Ch. V-2150-81) -82. -841 Tel. Rec. 97A-14 -30015 H30175 H-30015 H30175 88—14 H-302P5 (Ch. V-2151) 91—15 H-303P4 H30474 89—16	
-26 -14 -11	-82, -84) Tel. Rec 99A-14 H300T5, H301T5 (Ch. V-2148) 88—14	
-13 -23 -14	(Ch. V-2148) H-302P5 (Ch. V-2151-1) 91-15 H303P4, H304P4 (Ch. V2153)	
-12 -12		
	V-2136)	
-24 -23	H-310T5, H-310T5U, H-311T5, H-311T5U (Ch. V-2161, V-2161U). 99 —18	
ne)	H-312P4, H-312P4U, H-313P4, H-313P4U,	
-11 -36	H-314P4, H-314P4U, H-315P4, H-315P4U (Ch. V-2153-1)	
-19	V 2/136) 100—13 H-309F5 H-309F5U 101—16 H-31075 H-31015U, H-31175, H-31175U 101—16 H-312P4, H-312P4U, H-312P4, H-312P4U, H-312P4, H-313P4U, H-315P4, H-313P4U, H-315P4, H-313P4U, H-315P4, H-315P4U (Ch. V-2153-1) 98—13 H-316C7 (Ch. V-2136-1) (See Model H316C7)112 H-318T5, U	
-34	History Model H316C7)112 H-31875 H-31875 H-31075 H-310775	
-35	H-32015, U (Ch. V-2157, U) (See Model H-318T5) 117 H-321T5, U, H-322T5, U	
-19 -34	(Ch. V-2157-1, U) (See Model H-318T5)117 H.323T5 H.(Ch. V.2157.2	
	U) (See Model H-318T5) 117 H-324T7, H-325T7, U	
-36 -33 -37	(Ch. V-2136-2)11313 H-326C7 (See Model H-316C7) 112	
-25	H-327T6U (Ch. V-2157-3U)	
	H-336T5U, H-337T5U (Ch. V-2157U)134—12 H-600T16 (Ch. V-2150-61	
-31 -27	A, B) Tel. Rec	
-28 -29	(See Model H-600T16). 98 H-603C12 (Ch. V-2152-01	
	V-2157-3U) 12614 H-336T5U 13412 (Ch. V-2157U) 13412 H-600T16 (Ch. V-2150-61, A, B) Tel. Rec. V-2150-41) Tel. Rec. (See Model H-600T16). 98 H-603C12 (Ch. V-2152-01 & V-2149-3) Tel. Rec. 10014 H-604T10, H-604T10A (Ch. V-2150-71A, -94, -94A) Tel. Rec. (Supp. to H-609T10, Set 95) 99A-14	
	Tel. Rec. (Supp. to H-609T10, Set 95)	
-24	H-603112 (Ch. V-2150-101) Tel. Rec 97-19 H-606K12 (Ch. V-2150-	
-26	V-2150:91A, 94, 94A Tel. Rec. (Supp. to H-609T10, Set 95) 99A-14 H-005T12 (Ch. V-2150-101) Tel. Rec. 97-19 H-606K12 (Ch. V-2150- 111, A) Tel. Rec. 120-12 H-607K12 (Ch. V-2150- 111, A) Tel. Rec. [See Model 606K12]120	
-25	111, A) Tel. Rec. (See Model 606K12)120 H-608C12 (Ch. V-2152-01, V-2149-3) Tel. Rec. (See Model H-603C12)100 H-609T10 (Ch. V-2150-94C) Tel. Rec957 H-610T12 (Ch. V-2150- 136) Tel. Rec10513 H-611C12 (Ch. V-2152-16) Tel. Rec11214	
-26	V-2149-3) Tet. Rec. (See Model H-603C12) 100 H-609T10 (Ch.	
20	V-2150-94C) Tel. Rec 957 H-610T12 (Ch. V-2150- 136) Tel. Rec	
21 25	H-611C12 (Ch. V-2152-16) Tel. Rec	
23	H-613K16 (Ch. V-2150- 146) Tel. Rec. 107-12 H-614T12 (Ch. V-2150- 136) Tel. Rec	
17	toof for koc.	
	Tel. Rec. (See	,
13 15	1/0) Iel. Kec	,
16 22 21	H-617T12 (Ch, V-2150- 176) Tel. Rec	•
	H-618T16 (Ch. V-2150-	•
	186) Tel. Rec. (See Model H-617T12) 103	¢

WESTINGHOUSE-Cont. H-618716 (Ch. V-2150. 186, A, C, CA) Tel. Rec. (See Model H-617712) Prod. Chge. Bul. 10...116 H-619712, U(Ch. V-2150. 176, U) Tel. Rec. (See Model H-617712). 103 H-619712, U(Ch. V-2150. 176, U, V-2150.177U) Tel. Rec. (See Model H-617712). 103 H-620X16 (Ch. V-2150. 186, A, C, CA) Tel. Rec. (See Model H-617712). 103 H-620X16 (Ch. V-2150. 186, A, C, CA) Tel. Rec. (See Model H-617712). 103 H-620X16 (Ch. V-2150. 186, A, C, CA) Tel. Rec. (See Model H-617712). 103 H-620X16 (Ch. V-2150. 186, A, C, CA) Tel. Rec. (See Model H-617712). 103 H-622X16 (Ch. V-2150. 186, A, C, CA) Tel. Rec. (See Model H-617712). 103 H-622X16 (Ch. V-2150. 186, A, C, CA) Tel. Rec. (See Model H-617712). 103 H-622X17 (Ch. V-2150. 186, A, C, CA) Tel. Rec. (See Model H-617712). 116 H-6238116, H-629K-16 (Ch. V-2171) Tel. Rec. (See Model H-626716). 116 H-630117 (Ch. V-2173) Tel. Rec. (See Model H-626716). 116 H-630117 (Ch. V-2173) Tel. Rec. (See Model H-626716). 116 H-630117 (Ch. V-2173) Tel. Rec. (See Model H-626716). 116 H-630117 (Ch. V-2173) Tel. Rec. (See Model H-626716). 116 H-630717 (Ch. V-2173) Tel. Rec. (See Model H-626716). 116 H-630717 (Ch. V-2173) Tel. Rec. (See Model H-626716). 116 H-630717 (Ch. V-2178) Tel. Rec. (See Model H-626716). 116 H-630717 (Ch. V-2178) Tel. Rec. (See Model H-626716). 116 H-630717 (Ch. V-2178) Tel. Rec. (See Model H-626716). 116 H-630717 (Ch. V-2178) Tel. Rec. (See Model H-626716). 116 H-630717 (Ch. V-2178) Tel. Rec. (See Model H-626716). 116 H-630717 (Ch. V-2178) Tel. Rec. (See Model H-626716). 116 H-630717 (Ch. V-2178) Tel. Rec. (See Model H-626716). 116 H-630717 (Ch. V-2178) Tel. Rec. (See Model H-626716). 116 H-630717 (Ch. V-2178) Tel. Rec. (See Model H-626716). 116 H-630717 (Ch. V-2178) Tel. Rec. (See Model H-626716). 116 H-630717 (Ch. V-2178) Tel. Rec. (See Model H-626716). 116 H-630717 (Ch. V-2178) Tel. Rec. (See Model H-626716). 116 H-630717 (Ch. V-2178) Tel. Rec. (See Model H-626716). 116 H-638K20 (Ch. V-2178) Tel. Rec. 129—13 H-639T17 (Ch. V-2192, -1) Tel. Rec. 133—15 $\begin{array}{c} \text{Resc} & \text{Resc} & \text{Resc} & \text{Resc} \\ \text{Tell}, Rec. & \text{Resc} & \text{Resc} \\ \text{(Ch. V2172, Tell, Rec.} \\ \text{(See Model H-639177), 133} \\ \text{H-641K17} (Ch. V-2175-1, -5), \text{H-641K17A} (Ch. V-2175-1, -5), \text{H-641K17A} (Ch. V-2172, -1), -5), \text{H-641K17A} (Ch. V-2172, -1), -3), Tell, Rec. & \text{Resc} & \text{Resc} \\ \text{(See Model H-639177), -133} \\ \text{H-642X20} (Ch. V-2178-1, -3), Tell, Rec. & \text{(See Model H-639177)} & -13, Tell, Rec. & \text{(See Model H-125)} & -13, Tell, Rec. & \text{(See Model H-133)} & -14, Ch. & V-2102, -13, Ch. & V-2102, -13, Ch. & V-2103, -13, Ch. & V-2130, -13, V. & V-2130, -220, V. & See Model H-166, (DX) & -34, Ch. & V-2130, -210, V. & V-2130, -$ Ch. V-2137 (See Model H-203).... 62 V-2137

WESTINGHOUSE-ZENITH

v	VESTINGHOUSE-ZENITH
ZENITH-Cont.	ZENITH-Cont.
6D015, 6D015Y, 6D030	Ch. 6C22
(Ch. 6C05, 6C05Z) 3-24	(See Model 6R087) 7
6D815, 6D815W,	Ch. oC40
6D815Y (Ch. 6E05) 55-24	(See Model 6G001) 3
6G001, 6G001Y	Ch. 6C41
(Ch. 6C40) 3—14	(See Model 6G004Y) 20
6G001YZ1 (See Model	Ch. 6C50
6G001)	(See Model 6G038) 32
6G004Y (Ch. 6C41) 20-35	Ch. 6E02 (See Model 6R886) 34
6G801 (Ch. 6E40) 53-26	Ch. 6E05
6R084 (Ch. 6C21) 20-36	(See Model 6D815) 55 Ch. 6E40
6R08/ (Ch. 6C22)	(See Model 6G801) 53 Ch. 6G01
7H820, 7H820W	(See Model G660) 96
(Ch. 7E01)	Ch. 6G05
7H822 (Ch. 7E02),	(See Model G615) 86
7H822W7 7H8227	Ch. 5G20
(Ch. 7E02Z) 55-25 7H918 (Chassis 7F03) 75-18	(See Model G2957) 98
7H920, 7H920W (Ch.	Ch. 6H01 (See Model H661E)125
74921 (Charris 7504) 73 14	Ch. 7E01 {See Model 7H820} 43
7H922 (Ch. 7F02)	Ch. 7E02, 7E02Z {See Model 7H822} 55
7R070 (Ch. 6C06) 37—25 7R887 (Ch. 7E22) 54—22 8G005Y (Ch. 8C40) 7—33	Ch. 7E22 (See Model 7R887) 54
8G005YT (Z1) (Ch. 8C40T) (Z1), 8G005YT) (Z2) (Ch. 8C40T) (Z2) 53 —27	Ch. 7F01 (See Model
(Ch. 8C40T) (Z2) 53—27 8H023 (Ch. 8C01) 4—40	Ch. 7F02
8H032, 8H033 {Ch. 8C20}	(See Model 7H922) 87 Chassis 7F03 (See Model 7H918) 75
8H034 (See Mode) 8H023) 4	Chassis 7F04
8H050, 8H051, 8H052,	(See Model 7H921) 73
8H061 (See Model	Ch. 7G01
8H032) 1 8H832, 8H861 (Ch. 8E20) 52 —24 9H079, 9H079E, 9H079R, 9H081, 9H082R, 9H085R, 9H088R (Ch. 8C21)7—34	(See Model G725)101 Ch. 7G02
9H079, 9H079E, 9H079R,	(See Model G724)103
9H081, 9H082R, 9H085R,	Ch. 7G04
9H088R (Ch. 8C21) 7-34 9H881, 9H882R, 9H885, 9H888R (Ch. 9E21) 43-25	(See Model G723)104
9H888R (Ch. 9E21) 43-25	Ch. 7H02 (See Model
9H984, 9H984LP	H724)
9H984, 9H984LP	Ch. 7H04 (See Model
(Ch. 9F22)	H723)
12H090, 12H091, 12H092,	Ch. 7H04Z (See Model
12H093, 12H094	H723Z)
(CH. 11C21) 2-20	Ch. 8C01 (See Model 8H023) 4
14H789 (Ch. 13D22) 41-24	Ch. 8C20
27T965R (Ch. 27F20) Tel.	(See Model 8H032) 1
Rec. (See Model G2951) 95	Ch. 8C21
287925 (Chassis 28F22)	{See Model 9H079} 7
Tel. Rec	Ch. 8C40 (See Model 8G005Y) 7
(Chassis 28F25)	Ch. 8C40T(Z1), 8C40T(Z2)
Tel. Rec. (See Model	(See Model 8G005YT(Z1) 53
287925) 64 287960, 287961, 287962, 287963 (Ch. 28F20,	Ch. 8E20 (See Model 8H832) 52
281963 (Ch. 28720, 28F20Z, 28F21) Tel. Rec. (See	Ch. 8G20 (See Model G881) 98
Model 28T925) 64	Ch. 8G20/22 (See Model G3157RZ)
287964R (Chassis	Ch. 8H20 (See
28F23) Tel. Rec	Model H880RZ)
29T926R, E (Ch. 28F25) Tel. Rec. (See Model 28T925) 64	Ch. 8H20 Revised (See
28T925)	Model H880)
Rec. (See Models 421999RLP and 9H995). 74	(See Model 9H995) 74
37T998 RLPU (Chassis	Ch. 9F22
28F20, 9E21Z) Tel. Rec.	(See Model 9H984) 64
[See Model 287925 (Set	Ch. 10H20
64) and Model 9H995	(See Model H2437E)120
(Set 74)]	Ch. 11C21
42T999RLP (Chassis 28F23,	(See Model 12H090) 2
Radio Ch. 13D22) Tel.	Ch. 13D22
Rec. See Model	(See Model 14H789) 41
28T964R) 74	Ch. 22H20 (See
Ch. 4C52	Model H2226R)
(See Model 4K016) 6	Ch. 22H21 (See Model
Ch. 4C53	H2229R)
(See Model 4K035) 6	Ch. 23G22 (See Model G2322) Tel. Rec
Ch. 4E41	Ch. 23G23 (See Model
(See Model 4G800) 35	G2957)
Ch. 4E41Z	Ch. 23G24 (See Model
(See Model 4G800Z) 52	G2322Z)
Ch. 4F40 (See Model	Ch. 23G24Z1
4G903)	(See Model G2322Z1). *
Ch. 5C01, 5C01Z	Ch. 23H22Z
(See Model 5D011) 3	(See Model H2328EZ)118
Ch. 5C02, 5C02Z	Ch. 24G20 (See Model
(See Model 5R080) 4	G2420E)
Ch. 5C04	Ch. 24G20-OX (See Model
(See Model 5R080) 4	G2420E) 93
Ch. 5C40	Ch. 24G21 (See Model
(See Model 5G003) 17	G2454R)
Ch. 5C40Z, 5C40ZZ	Ch. 24G21-OX (See Model
(See Model 5G003Z) 30	G2454-ROX)
Ch. 5C51	Ch. 24G22/23
(See Model 5G036) 30	(See Model G2441R) 98
Ch. 5E02	Ch. 24G24
(See Model 5D810) 54	{See Model G2441} 98
Ch. 5G01 (See Model	Ch. 24G24/25
G511) 85	(See Model 3059R) 98
Ch. 5G02 (See Model	Ch. 24G26
G510) 84	(See Model G2437RZ) 91A
Ch. 5G03 (See	Ch. 24G26Z1
Model G516)	(See Model G2441Z1). *
Ch. 5G40 (See Model	Ch. 24H20, 24H21 {See Model H2437E}120
G500)	Ch. 27F20 (See Model 27T965R) 95
Ch. 6C01	Ch. 28F20, 28F20Z, 28F21,
(See Model 6D014) 9	28F22 (See Model
Ch. 6C05, 6C05Z	281925) 64 Ch. 28F23
Ch. 6C06	(See Model 28T964R) 74 Ch. 28F25
(See Model 7R070) 37	(See Model 28T925) 64
Ch. 6C21	Ch. 29G20
(See Model 6R084) 20	(See Model G2951) 95

WESTINGHOUSE-Cost. Ch. V-2137-1 (See Model H-199).... 69 Chassis V-2137-2 (See Model H-198).... 73 Ch. V-2137-3, V-2137-35 (See Model H-231) 974 h. V-2144, V-2144-1 (See Model H-210).... 61 h. V-2146-05 (See Model ... 97A Ch **7ENITH** n. V-2149 (See Model H-2178)..... **91** n. V-2149-1 (See Model H-216)..... **97A** Ch Ch Ch Ch Ch Ch Ch Сh Ch (See Model H-610T12)...105 Ch. V-2150-146 (See Model H-613X16).....107 Ch. V-2150-176,U (See Model H-617T12)..103 Ch. V-2150-177U (See Model H-612T12, Set 103 and Prod. Chge. Bul. 10, Set 116) Ch. V-2150-185, A, C.CA (See Model H-617T12)..103 Ch. V-2150-1927 (See Model H-617T12).103 Ch. V-2150-197 (See Model M-625T12)..114 Ch. V-2151-1 (See Model H-302P5)...91 Ch. V-2152-01 (See Model H603C12)....100 Ch. V-2152-14 Ch 1. V-2152-16 (See Model H-611C12). 112 Ch. h. V-2156 (See Model H-309P5)...101 Ch. V-2157, U, +1, -1U, -2, -2U (See Model H-318T5)......117 Ch 117 Ch V-2171 n. V-2171 (See Model H-626TT6)..**116** (See Model H-626T16), .116 Ch. V-2175, J. 3, 44, -5 (See Model H-633C17), 122 Ch. V-2175, J. 3, -4, -5 (See Model H-639T17), 133 Ch. V-2175, V-2176, V-2177 V-2177 (See Model H-626T16), ..., 116 Ch. V-2178, -1, -3 (See Model H-638K20), ..., 129 Ch. V-2109, -1, (See WILCOX-GAY (Also See Recordio) WILLY5-OVERLAND 8030 50-23 WILMAK W-446 "DENchum" 21-11

WOOLAROC

3-1A (Ch. 6-9022-J), 3-2A (Ch. 6-9022-K) ... 6-37

 WOOLAROC-Cont.

 3-3A (Code 7-9003.D)
 6-38

 3-5A
 22-32

 3-6A/5
 24-37

 3-70
 311A (Ch. 56A76)
 8-33

 3-12A/3
 3-13A, 3-16A, 3-15A, 3-16A, 34-28
 33

 3-17A, 3-16A, 3-16A, 34-28
 32-70A, 3-16A, 34-28

 3-70A
 2-4-33
 7--31

 3-61A (See Model 3-71A) 36-34
 36-29

 7FNUTU
 36-29
 7FNUTU
 Tel. Rec. G2327Z (Ch. 23G24) Tel. Rec. (See Model G2322Z)... 91A (See Model G23222). G-2340, R (Ch. 23G22) Tel. Rec. (See Model G2322) G2340RZ, Z (Ch. 23G24) Tel. Rec. (See Model G2322Z)... 91A G2340Z1. R71 (CL Tel. Rec. (See Model G2322Z). tsee Model G2322Z)... 91A G2353EZ1 (Ch. 23G24Z1) Tel. Rec. Tel. Rec. G2356EZ (Ch. 23G24) Tel. Rec. (See Model G2322Z)... 91A G2420R (Ch. 24G20) Tel. Rec. (See Model G2420E)... 93 See Model G2420E)... 93 G2420.ROX (Ch. 24G20-OX) Tel. Rec. (See Model G2420E)... 93 G2437RZ, G2439RZ, Z, G2439RZ (Ch. 24G24)... 91A-12 G2441 (Ch. 24G24) Tel. Rec. (See Model G2322) 98 G2441 R (Ch. 24G22/23) Tel. Rec. Tel. Rec. (See Model G2322).... 98 G2441RZ, Z (Ch. 24G26) Tel. Rec. (See Model G2437RZ)... 91A ()See Model G2437K2)... G244121, RZ1 (Ch. 24G26Z1) Tel. Rec.... G2442E, R (Ch. 24G22/23) Tel. Rec. (See Model G2322) ... 98 G2442RZ (Ch. 24G26) G2322) G2448RZ (Ch. 24G26) Tel. Rec. (See Model G2437RZ). 91A Inec., Rec. (See Model G2437RZ).
 91A
 G2448RZ1 (Ch. 24G26Z1)
 Tel., Rec.
 G2454R (Ch. 24G21)
 Tel., Rec.
 (See Model G2420E)...
 93
 G-2454-ROX (Ch. 24G21-OX)
 Tel. Rec.
 (See Model G2420E)...
 93
 G2951, G2951R, G2952R, G2952R, G2952-ROX (Ch. 29G20)
 Tel. Rec.
 (See Model G2420E)...
 95-G2957R (Ch. 23G23 & Radio G G2958
 Rec. (See Model G2322)
 98
 G2958R (Ch. 23G23 & B 95_8 Rec. (See Model G2322) 98 G2958R (Ch. 23G23 & Radio Ch. 6G20) Tel. Rec. (See Model G2322) 98 G3059R (Ch. 24G24/25 & Radio Ch. 6G20) Tel. Rec. (See Model G2322) 98 G3062 (Ch. 24G24/25 & Radio Ch. 6G20) Tel. Rec. (See Model G2322) 98

 ZENITH-Cont.

 G30627 (Ch. 24G26)

 Tel. Rec.

 SG30/22) Tel. Rec.

 SG30/22) Tel. Rec.

 SG31/SZ1, RZ1 (Ch. 23G24,

 SG30/22) Tel. Rec.

 SG31/SRZ (Ch. 23G24,

 SG31/SRZ (Ch. 23G24,

 SG30/22) Tel. Rec.

 SG30/22) Tel. Rec.

 SG80/22) Tel. Rec.

 SG30/22) Tel. Rec.

 SG20/22) Tel. Rec.

 SG31/SRZ (Ch. 23G24,

 SG20/22) Tel. Rec.

 SG20/22) Tel. Rec.

 See Model G3157RZ).

 SG31/SRZ (Ch. 24G24,

 SG20/22) Tel. Rec.

 See Model G3157RZ).

 See Model G3157RZ).

 SG31/SRZ (Ch. 24G26,

 SG20/22) Tel. Rec.

 See Model G3259RZ).

 SG3258RZ (Ch. 24G26,

 SG20/22) Tel. Rec.

 See Model G3259RZ).

 Notice
 Holik
 Izs-I3

 AH01)
 Izs-I3
 Izs-I3

 H665, R, RZ, Z
 (A. 6H01)
 Izs-I3

 V23 (Ch. 7H04)
 Izz-I2

 H723 (Ch. 7H04)
 Izz-I2

 H723 (Ch. 7H04)
 Izz-I2

 H723 (Ch. 7H04)
 Izz-I2

 H723 (Ch. 7H04)
 Izz-I2

 H806, H800 (Ch. 8H20
 Revised)

 Revised)
 Izz-I2

 HB80R, HB80R (Ch. BH20

 Revised)
 127—15

 H2229R, H2230E,R (Ch. 27–1A

 H280R2 (Ch. BH20)
 114—12

 H-1083E (Ch. 10420)
 15ee Madel H2437E)
 120

 H380R2 (Ch. BH30)
 114—12

 H-1083E (Ch. 10420)
 15ee Madel H2437E)
 120

 H380R2 (Ch. 24207E, H2227E, H2227E, H2227E, H2227E, H2227E, H2227E, H2227E, H2227E, H2227E, H2228E, H2228E, H2235E (Ch. 2420) Tel. Rec.
 127–1A

 H2250R, H2235E (Ch. 24120) Tel. Rec.
 114—13
 14226R, H2235E (Ch. 24122)

 Tel. Rec. (See Model
 12228R, H2235E (Ch. 24122)
 127–1A

 H2352R, H2235E (Ch. 24122)
 127–1A
 14235R2, H2235E (Ch. 24122)

 Tel. Rec. (See Model H2228R)
 127–1A

 H2352R, H2235E (Ch. 24122)
 128
 127–1A

 H2352R, H2235E (Ch. 24122)
 128
 127–1A

 H2352R, H2235E (Ch. 24122)
 128
 128

 H2352R, H2235E (Ch. 24122)
 128
 128

 H2352R, H2235E (Ch. 24122)
 128
 120

 H2437E (F. 24421)
 120
 124

 H2437E (F. 24421)
 120
 124

 H2437E (F. 24421)
 120
 124

 H2437E (F. 24421)
 120 H3475R (Ch. 24H20 and Radio Ch. 10H20) Tel. Rec. (See Model H2437E)
 Rec.
 [Sea Model]

 H243751
 120

 H347787 (Ch. 24H21 and Redio Ch. 10H20) Tel.
 120

 H34787 (Ch. 24H21 and Redio Ch. 10H20) Tel.
 120

 H34786 (Ch. 24H21 and Redio Ch. 10H20) Tel.
 120

 H34787 (Ch. 44H21 and Rec. (See Model
 120

 H243761 (Ch. 44H2)
 35-27

 4G800 (Ch. 44H2)
 35-27

 4G8003 (Ch. 44H2)
 35-27

 4G8003 (Ch. 44H2)
 35-27

 4G803 (Ch. 44H2)
 52-23

 4G803 (Ch. 4C53)
 6-39

 4K016 (Ch. 4C53)
 6-39

 4K035 (Ch. 4C53)
 6-40

 5D011, 5D027
 3

 4K035 (Ch. 4C53)
 6-40

 5D011, 5D027
 3-17

 (Ch. 5C01, 5C01)
 3-17

 5B03 (Ch. 5E02)
 54-21

 5G003 (Ch. 5C40)
 17-35

 5G0032 (Ch. 5C402),
 30-31

 5G0034 (Ch. 5C51)
 30-32

 5R080-5R086
 4-4

 6D014, 6D014w, 6D029,
 6D024, 6C01)

 6D024 (Ch. 6C01)
 9-35

7621	a dia I	0400
ee Mi	217	71100
ee Mi sis 98 ee Mi 9F22 ee Mi 10H20 ee Mi 11C21 ee Mi 12D22	odel	9499
9F22		
ee Ma	del	9H98
10H20)	
ee Mo	odel	H243
11021		1 2110
ee M 13D22	odel	12H0
ee M 22H2(1411/
odel h	1222	6R)
22H2	(Se	e Mor
odel H 22H21 2229R	}	
23G2	2 (Se	e Mo
2322)	Tel.	Rec.
23G2	3 (Se	e Mo
22H2 2229R 23G2 23G2 23G2 23G2 23G2 23G2 23G2 23G		
2362	4 (56	e Mo
23222	471	
	odel	623
23H2	77	015
ee M	odel	H232
23H2: ee Ma 24G2(2420E 24G2(24G2) 24G2 24G2 2454F 24G2 2454- 24G2 2454- 24G2) (Se	e Mod
2420E)	
24G2	o-ox	(See
2420E)	
24G2	1 (Se	e Mo
2454R)	
2462	1-03	(See
2434-	KUX	
2402. ee Ma	∠/∡o ⊳del	G244
ee M 24G2	4	0144
ee Me	del	G244
24G2-	4/25	
ee Mo	del 3	3059R
24G2	6	
ee Mo	del	G243
24G2	6Z1	~ ~ ~
66 M	oder	G24
241121	0, 24 Idal	121
27F20	Juer	11243
ee Mo	del 3	27196
28F20	, 281	20Z,
F22 (See	Mode
T925		
28F23	3	
ee M	odel	2879
28125		2070
ee m	Daei	2019
24G2 24G2 24G2 24G2 24G2 24G2 24G2 24G2	del	G205
ve me		0275
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RC-161A (Supplement to	GARRAD	PHILCO	UNIVERSAL CAMERA	346 (CM-3) 100-12
RC-200) (CM-1) 21-37 RC-170, RC-170A. (CM-1) 31-2	RC-60	D10, D10A (CM-1) 14-21 M-4	100(CM-1) 36 —30	356, 357 (CM-3) 106—16
RC-180, RC-181 (CM-2) 76-1 RC-182 Supplement (CM-2) 76-2	GENERAL ELECTRIC	M-7(CM-1) 28-35	UTAH	WESTINGHOUSE
RC-200	P6 (CM-2) 79—8	M-8	550 (CM-1) 8 650 (CM-1) 22-34	V4914 (CM-2) 47—26 V4944 (CM-2) 86—13
(CM-3) 72-1 RC-221, RC-222 (CM-3) 79-1	GENERAL INDUSTRIES	M-12C	7000 (CM-1) 27-31	V6235
RC220, RC221, RC222	RC130L (CM-1) 22-33	RCA	7001(CM-2) 83—15	ZENITH
Changes	GENERAL INSTRUMENT	RP168	V-M	S11468 (CM-1) 23—35 S11680
Model RC220 Changes)	204 (CM-1) 23-34 205 (CM-1) 10	RP-177	200-B (CM-1) 15 —36 400	\$14001 (CM-2) 75-17
RC400	LEAR PC-206A	RP-178	400 (Late) (CM-2) 90-13 402, 400C (CM-2) 82-12	\$13675, \$14002, \$14006, \$14008 (CM-2) 85-15
AERO	MAGUIRE	SEEBURG K	402D, 400D (CM-2) 87-14 404 (See Model 405)	S14004, S14007 (CM-2) 79
46A	ARC-1	L	(CM-3) 73 405 (CM-3) 73—14	\$14022 (CM-3) 112-15
47A (CM-2) 77—2	MARKEL	S, SQ (CM-2) 78-12	406, 407 (CM-3) 102-16	S14023
AVIOLA 100 (CM-1) 33-32	70, 71 (CM-2) 84—8 74, 75 (CM-3) 91—7	SILVERTONE	800	Model S14022) (CM-3) 112 S14026 (See Model
BELMONT	74, 75 Supplement	101.761-2, 101.762-2 [CM-2] 77-10	802	S14023) (CM-3) 105
C-9 (CM-2) 34-31	MILWAUKEE ERWOOD	101.761-3, 101.762-3(CM-2) 83—11	950 (CM-3) 107-13 950 Supplement	514027 (See Model 514022) (CM-3) 112
COLUMBIA 104 124-2	10700 (CM-1) 16-37 11200	101.762, 101.763		MISCELLANEOUS
CRESCENT	11600	SPARTON	WEBSTER 50 (CM-1) 24-35	Series 700F (CM-2) 89-9
C-200 (CM-1) 20-37	MOTOROLA	C48	56 (CM-1) 17-36	Series 700F 33/45 (CM-3) 75-11 Series 700FLP (CM-2) 101-6
6 Series	B24RC, B25RC, B27RC, B28RC (CM-1) 12-35	THORENS	70	Series 700FS (CM-2) 104-8
350 Series (CM-2) 80-3	RC30 (CM-2) 80-9	CD-40 (CM-1) 39-29	148 (CM-2) 86-12	Series 700R (CM-2) 91-8

RECORDERS

AMPRO	CRESCENT-Cont.	GENERAL INDUSTRIES	RCA	ST. GEORGE
730133—4	H-2A1 Series (CM-3) 119-4	R70, R90 (CM-1) 35-28	MI-12875 (CM-2) 85-12	1100 Series
BRUSH SOUND MIRROR BK-401 Tape Recorder [CM-1] 42-25 BK-403	H-19 Series "Steno"	INTERNATIONAL ELECTRONICS PT3(CM-2) 88-4 LEAR DYNAPORT WC-311-D	REELEST CIA123—13 SILVERTONE	Wire Recorder (CM-1) 40-24 WEBSTER-CHICAGO 79-80 Wire Recorder (CM-1) 37-26 178
BRUSH MAIL-A-VOICE BK-501, BK-502, BK-503	CRESTWOOD CP-201 (CM-3) 118-4	MAGNECORD AUDIAD AD-1R(CM-2) 84-7	70 (Ch. 567.230, 567.231)	WEBSTER ELECTRIC Ekotape
CRESCENT H-1A	EICOR 1000	MASCO 375 (CM-3) 117-7	101.774-2, 101.774-4 (CM-3) 114-10	WIRE RECORDING CORP. WP

INDEX TO ADVERTISERS

American Phenolic Corp.																20
The Astatic Corporation																
Camburn, Inc	• •															50
Electro-Voice, Inc						•				٠						16
Erie Resistor Corp																
General Cement Mfg. Co.																52
Insuline Corporation of A	m	er	ic	ea		5										52
International Resistance	Co	•			Ŀ	ns	si	de	e	F	rc	n	t	C	'o'	ver
Littelfuse, Inc]	In	S	id	e	E	Ba	cl	ĸ	C	'o'	ver
Merit Transformer Corp.				•	•	•										12
Park Metalware Co., Inc.	,						•	•		•	•				e.	52
Quam-Nichols Co			•			÷				•	l,					18

Radio Electronics								2	52
Radio Receptor Company, Inc									56
Shure Bros., Inc.									50
Simpson Electric Co								•	24
Sylvania Electric Products Inc						•			32
Technical Appliance Corp			•		•	•			58
Television Communications Institute									50
Thomas Electronics, Inc	•	•				•			22
Triplett Electrical Instrument Co		•			÷		÷		14
V-M Corporation									52
The Ward Products Corp			×.	•	•			•	62

" SHOP TALK " (Continued from page 4)

align such circuits we require a sweep generator. The generator should be capable of generating the frequencies we desire and its output should be constant over the section of the range it is sweeping. (Note that this does not mean its output need be constant over its entire frequency range.)

Basically, the foregoing is all you require of any sweep generator. Any other function that such a unit can perform is desirable (if it helps you shorten your work) but not absolutely essential. Hence, for the fellow with a limited amount of money to spend, only the two basic requirements stated above need be given service consideration.

Built-in marker generators are handy, but seldom essential since a separate AM generator is usually available. The ability to blank out the return trace will make it more convenient to do your job but here, too, is a refinement that the fellow with limited capital can forego.

The choice of an AM signal generator is a little more critical because it is this generator upon which you will lean heavily during nearly all of your alignment work. In stagger-tuned IF systems, this generator will be used to help you accurately peak the various coils prior to an overall check. And, when the sweep check is being made, the same AM generator will be used as a frequency marker, showing you exactly what the response of the circuit is. It will tell you if the video carrier is at the 50 per cent level; it will tell you at what frequency the response drops; it will show you how wide the passband is and it will reveal whether the traps are where they should be.

In short, it will tell you everything that the sweep generator cannot tell you - and if the AM generator is accurately calibrated, it will tell you all this in precise terms.

In view of the importance of this AM generator, any extra money lying around should be sunk into this instrument. You can get by with AM generators possessing an accuracy of only 5%, but if you want to be sure, get one that has an accuracy of 1 per cent. Better still, get an AM generator which uses a crystal cambrator and has a large, easy-to-read dia face. And get one that permits you to read up to 150 mc or better, on fundamentals. Harmonics can and have been used - but you will frequently find that spurious signals will develop which will confuse and mislead you. If possible, stick to fundamentals up to whatever frequency you desire to reach. The foregoing discussion is intended to serve as a guide to those entering the TV service field. It is a result of the author's experience plus the experience of his associates. No attempt was made to indicate specific models because within the same price class there are generally several units of similar quality to choose from. Some men may prefer one instrument, others may prefer a different make, etc., and both groups will be equally vehement about the virtues of their choice.

Choose whichever instrument you prefer, but make sure it serves the purpose you intended.

* * *

REVIEW: This month we are concerned with the review of an article for improving sound reproduction of typical television receivers. The article is as follows:

IMPROVED AUDIO QUALITY FROM STANDARD TV RECEIVERS by G. C. Proud AUDIO ENGINEERING - October 1949

Copyright 1949 by Radio Magazines, Inc. 342 Madison Avenue New York 17, New York

Subscription Price \$3.00 per year in U. S. A., Possessions, and Canada

In every television receiver, the major design emphasis is on the video circuits and the quality of the picture that is produced. The sound section of the receiver is relegated to a secondary position and, as a consequence, it has suffered considerably in quality. The most widely used arrangement, as exemplified by the audio section of the RCA 630TS circuit (shown in Figure 1), consists of a high-mu triode with contact potential bias for the first stage, and a pentode output stage without feedback. As Mr. Proud points out in his article, add to this an output transformer, which has a $5/8 \times 5/8$ inch core, and a small speaker with inadequate baffling, and the sound you hear is bound to be less than ideal.

As a start toward the improvement of the sound quality, it is suggested that the output transformer be changed to one containing more iron in its core. Just what type to get will be governed by the amount of space that you have available on the chassis for mounting the unit.

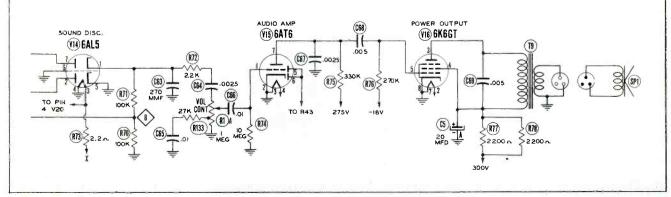


Figure 1. Audio Section RCA Victor 630TS.





The next step is an alteration of the circuit itself involving a change in tubes, replacement of some of the components, and the inclusion of negative feedback. The revised circuit diagram of the 630TS is shown in Figure 2. Principally, the following changes have been made.

1. The 6AT6 audio stage has been replaced by substituting a type 6AU6 for audio amplification and a type 1N34 germanium diode for the diode rectifier.

2. The 6K6 audio output stage was changed to employ a type 6V6GT

3. The interstage coupling capacitor, C67, has been increased in value (to aid low-frequency response). For the same reason, C64 has been changed, too.

4. C67 and C69 have been eliminated.

5. Feedback from the plate of the 6V6 to the cathode of the 6AU6.

6. The arm of the volume control is now connected directly to the grid of the 6AU6, permitting the elimination of C66.

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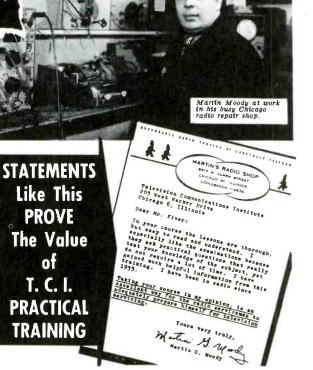
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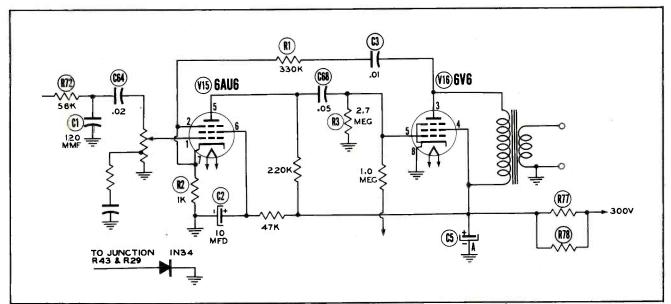


Figure 2. Revised Audio Section RCA Victor 630TS (Type 1).

Since the diode portion of the 6AT6 is used in the contrast control circuit of this particular receiver, provision must be made to replace this when the 6AU6 is substituted for the 6AT6.

A germanium crystal, 1N34, will do this very nicely.

An alternate diagram is suggested by Mr. Proud, one in which the output transformer is included in the feedback loop (see Figure 3).

To determine which side of the secondary winding of the output transformer the resistor R5 (feedback resistor) should be connected to, the following test is suggested.

Make a temporary ground connection to one side of the transformer secondary. Then, with a signal passing through the set, momentarily connect the feedback resistor R5 to the other output lead. If the output is reduced, correct polarity is indicated. If not, reverse the connections to the transformer secondary. When these changes have been made, you will find that additional filtering of the B+ voltage to both stages is desirable in view of the improved low-frequency response. In this particular receiver, the 300 volts B+ is obtained from the power supply just before the filter choke (i. e., on the rectifier side of the choke). If this connection is transferred to the other side of the filter choke, the hum level will decrease. In the 630 receiver, there is an adequate number of filter condensers and no additional units need be added. In other receivers, it may help to include additional filter capacitors of 40 to 80 mfd.

It might be pointed out that there is a considerably more elaborate system outlined in a recent article, "Viewer's Amplifier" by Melvin C. Sprinkle in the January 1951 issue of FM-TV Radio Communications. However, the amplifier described is actually a custom built project from scratch, so to speak, so that it lies a little bit outside typical TV receiver improvement. However, the equipment described is capable of excellent reproduction and should a custom application arise it would be well to consult Mr. Sprinkle's write-up.

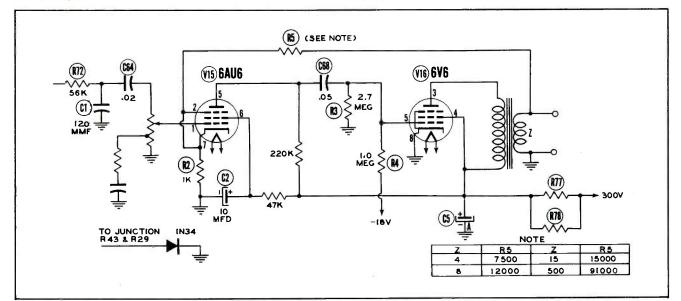


Figure 3. Revised Audio Section RCA Victor 630TS (Type 2).



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Continued from page 31 + +

plate of the following tube. By adjusting each of the stages to the proper response, the over-all response will be correct.

One of the disadvantages of this type of transformer is the inability to vary the mutual coupling between the coils. In other words the bandpass is fixed since it is dependent on the mutual coupling between the coils. In many cases one of the windings is wound on a tubular form which is slipped over the other winding. Should this tube slip or change position in any way, the bandpass of the circuit will be changed.

Another way of obtaining the same results is through the use of a "common impedance" type coupling. In this circuit a capacitor or coil, or both, is placed in the circuit so that it is a part of both the output circuit of one stage and the input circuit of the following stage. By adjusting the common impedance component so that its impedance is equal to that of the mutual inductance in the overcoupled transformer, similar bandpass characteristics are obtained.

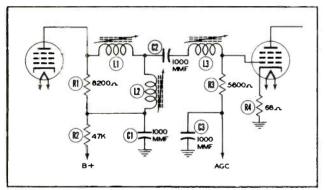


Figure 2-11. "Common Impedance" Coupled Video IF Stage.

Figure 2-11 is a partial schematic showing a "common impedance" coupling. The tuned circuit in the input consists of L1 in series with L2, shunted with the output capacity of the tube along with the distributed capacity in the wiring. The tuned circuit in the output consists of L3 in series with L2, shunted by the input capacity of the tube and the distributed capacity in the wiring. C2 serves as a DC blocking capacitor and its reactance is so low at the IF frequency that it need not be considered as a part of the tuned circuit. Since L2 is a part of both the input and output circuits, the signal is coupled from one stage to the next, and the pass band is dependent on the inductance value of L2. In the schematic of Figure 2-11, L2 is shown as a variable inductance. In many cases, however, the common impedance coil is not variable, its value having been predetermined to provide proper bandwidth. R1 and R3 shunt the input and the output circuits respectively, lowering the Q of the circuit to make possible wider bandpass characteristics.

It is possible to use a capacitor as the "common impedance" component instead of an inductance, but less gain is realized. As a result, the inductance is most commonly used.

TRAP CIRCUITS

Up to this point our discussion of video IF amplifiers has dealt with the circuits which were designed to give amplification of the signal. Equally important are the trap circuits which are used for rejecting certain unwanted signals.

There are four basic type trap circuits employed in the video IF amplifiers. They are:

1. Absorption Traps.

2. Parallel Resonant Traps.

3. Series Resonant Traps.

4. Bridged T Traps.

The three frequencies which are undesirable in the video IF strip are the adjacent channel sound IF, the adjacent channel video IF, and the accompanying sound. Obviously all frequencies that are not a portion of the video signal would be undesirable. The three signals mentioned above, however, are ones which will occur at a certain point when a channel is properly tuned, making it possible to tune traps to reject them. One exception to the above would be in the case of an intercarrier receiver in which case the sound IF is allowed to continue through the video IF strip but at a much lower level than the video IF signal.

As a general rule the traps which are tuned to the most objectionable signals will be located near the front end of the IF strip. Thus the signal can be rejected while it is still at a lower level. If this signal were allowed to go through several amplifying stages before it was trapped, it would be much more difficult to reject.

In many cases two traps that are tuned to the same frequency will be used. This is especially true in the case of the accompanying sound IF signal in receivers having separate sound and video IF channels. The first sound IF trap, which also serves as a sound IF take-off point, is usually located in the plate circuit of the mixer. The second trap is located in the last IF stage and is normally some form of absorption trap.

Figure 2-12 shows a few of the most frequently used trap circuits. The circuit shown in (A) is an absorption type trap. The secondary, or trap winding, is positioned close to the primary so that the two windings are mutually coupled. The trap has a very high Q winding, usually around a Q of 300, and at its resonant frequency will "suck out" a very narrow band of frequencies. The energy that is required to set up parasitic oscillations in the trap winding is taken from the primary of T1, thus it is not passed on to the rest of the circuits. This trap winding is an excellent source for the sound IF signal and the circuit shown in (A) has a lead connected to a tap on the trap winding which is connected to the input of the sound IF.

Circuit (B) shows a similar absorption trap, except that in this case the primary is in the cathode

VIDEO IF AMPLIFIERS

circuit of the tube instead of the plate circuit. At the resonant frequency of the trap (usually the sound IF), degeneration is introduced into the cathode circuit, thus reducing the gain of the stage at this frequency. The Q of this trap winding is usually around 300, which makes possible the rejection of a very narrow band of frequencies. In some cases the trap circuits of (A) and (B) will both be used in the same video IF strip. The small amount of signal which is not rejected by circuit (A) will be rejected by circuit (B). Note that in circuit (B) C2 bypasses R1 to prevent degeneration at frequencies other than the resonant frequency of the trap.

The trap circuit illustrated in (C) is also a form of absorption trap. At first glance it appears to be a parallel resonant circuit which would present maximum impedance at the resonant frequency. If this were the case there would be an increase in the signal level instead of the desired decrease. The answer lies in the value of C1, which in this case is 1.2 mmf. With this low value, the reactance of C1, even at the comparatively high IF frequency, is quite high. Thus the resonant circuit of L1 and C2 can be considered to be loosely coupled to the secondary of T1. This is the same condition which exists when two coils are loosely coupled, in that energy will be coupled from one to the other. Another approach to the subject would be to realize the fact that the energy that is dissipated to ground, due to the parasitic oscillations of the resonant circuit C2 and L1, must be taken from the grid circuit, therefore rejecting the unwanted signal. This type of trap seems to be gaining in popularity, probably due to the fact that it can be added almost anywhere in the circuit without disturbing existing components.

The trap circuit of (D) is a form of parallel resonant trap. A portion of the circuit, C2 and L2, is in series with the plate current path for the preceding stage. When a signal is passed by the preceding stage that corresponds to the resonant frequency of the trap circuit, degeneration will be introduced into the plate circuit. In this manner the undesired signal will be trapped out. It is interesting to note that in many respects this circuit resembles that of (C). In the case of (C) the signal was coupled to the trap by a small capacitor while in the case of (D) the signal is coupled across only a small portion of L2.

The trap circuits shown in (E) are parallel resonant traps and are placed in series with the signal. At the resonant frequency, the trap will present maximum impedance to the signal which will produce a loss in the coupling circuit. In most cases the traps shown are adjusted to the accompanying sound IF and adjacent sound IF signals, but they may be used for rejection of any signals.

The series tuned trap is less frequently used than those previously described. A trap of this type consists of a capacitor and a tunable choke in series and is usually placed between the plate and ground. One application of the series resonant trap is in the plate circuit of the mixer. The trap is tuned to the center of the adjacent channel IF frequency and aids in rejecting any interference from this source.

The bridged T trap in circuit (F) produces good results but is not used very often, probably due to the

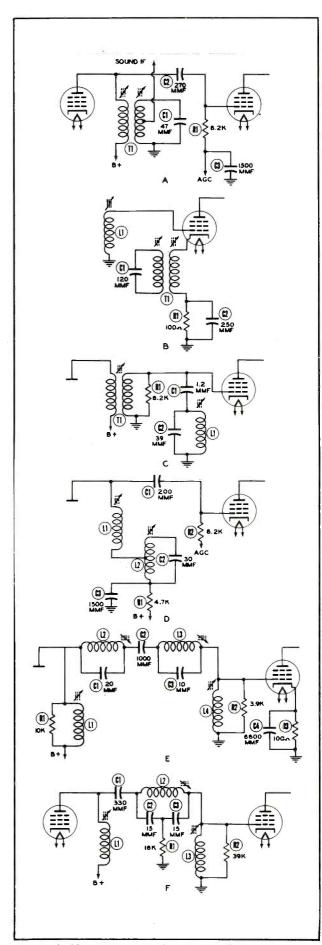


Figure 2-12. Representative Types of Trap Circuits.

expense of the extra components. A circuit of this type will produce a much narrower null point than a parallel resonant trap alone. The trap consists of L2, C2, C3 and the resistor R1 connected between the junction of the two capacitors and ground. L2 is adjustable so that the circuit can be resonated at the desired frequency. The bridge is balanced when the reactance of the capacitive branch equals the reactance of the inductive branch, and the resistor is equal to approximately 1/4 of the parallel resistance of the circuit. In the event of replacement of the capacitors or resistor, the new part should exactly duplicate the original to maintain proper balance of the circuit.

Another version of this same trap consists of a center tapped coil with a single capacitor across it. The operation of the circuit is the same as that of circuit (F).

The adjustment of the traps in the video IF circuit is very critical. Misaligned traps may result

♦ Continued from page 15 ♦ ♦

vertically and horizontally. If it does not, check the wiring carefully for any errors that might have been made.

The design of the new horizontal transformer used in this conversion is such that a much higher degree of efficiency is obtained over that of the original unit. Even though greater power is required in the deflection yoke to sweep the larger tube, the current drain through the 6BG6G tube is less in the revised circuit than it *was in the original circuit. In the original circuit the total current through the 6BG6G was 100 ma. In the revised circuit the current is only 87 ma. The total B plus current drain in the set fell from 250 ma to 225 ma. The added resistance in the focus coil circuit reduced the low B plus voltage approximately 5 volts. This reduction will not affect the operation of the receiver.

CABINET CHANGES

The first step in altering the cabinet to accommodate the new tube was the enlarging of the cutout in the subpanel. The four tube mounting brackets were removed and the subpanel was marked to indicate the points where the cutout was to be made. All dimensions for making this cutout are given in Figure 8. The four corners were cut diagonally to provide additional strength. Cutting the enlarged opening out to square corners may weaken the panel. The subpanel can be sawed very easily with a keyhole saw or hacksaw blade. The dimensions given should be followed very closely so that the mounting brackets can be positioned to hold the new tube. After making the cutout, four holes were drilled in the subpanel to secure the mounting brackets. The positioning of these brackets is not critical but it is suggested that they be mounted in approximately the same position as those in Figure 7, where two brackets are mounted at the bottom and one on each side near the top. This provides adequate support for the tube as well as a means of properly positioning it. These brackets were originally mounted with the ears pointing toin poor picture reproduction, excessive interference in the picture, or loss of the sound signal. The most positive method of adjusting the traps is through the use of sweep alignment. By increasing the level of the sweep signal and the marker, a more positive indication can be obtained on the scope. An even greater degree of accuracy can be obtained by decreasing the sweep width in the generator so that only a narrow band is swept on either side of the trap "notch." This has the effect of increasing the width of the notch making possible a more critical adjustment.

As previously stated, most of the requirements of the video IF amplifier are design problems and are solved at the manufacturing level. The service technician, on the other hand, is very frequently called upon to align the TV receiver. This operation is probably the most exacting of all adjustments and the result may mean the difference between a good picture or an unacceptable one.

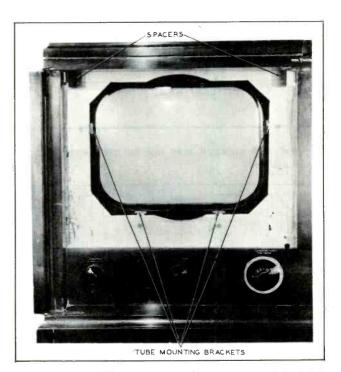


Figure 7. Sub-Panel Cutout and Tube Mounting Arrangement.

ward the front of the cabinet. The new mask, however, made it necessary to move the tube farther back in the cabinet which caused the edge of the tube to miss the brackets. The brackets were turned around with the ears facing to miss the back but even then it was found that the particular mask which was used forced the tube back to a point where the tube rested on the edge of the ears. To overcome this, longer bolts were used in the mounting brackets and spacers were placed between the mounting brackets and the subpanel. These spacers were made of small blocks of 1/4 inch plywood which moved the brackets back enough to support the tube. The position of the tube is dependent on the type of mask that MILLIONS OF "SAFE CENTER" SELETRON RECTIFIERS IN USE IN RADIO AND TV!



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MAX. PEAK MODEL PLATE MAX. INPUT MAX D.C. STACK VOLTAGE R.M.S. INVERSE OUTPUT THICKNESS NO. SIZE 100 MA 1M1 1" sq. 25 75 % 1" 5q. 12" 5q. 12" 5q. 12" 5q. 1" 5] 1" 5] 1" 5] 20 MA* 20 MA* 380 760 8Y1 130 品"" 书书"" · 14" · 14" 16Y1 260 380 65 MA 8J1 130 5M4 130 380 75 MA 5M1 130 100 MA 380 150 MA 5P1 130 380 7⁄a″ 1 1³ 7/8″ 156 130 6P2 456 200 MA 5R1 380 5Q1 11/8" 130 380 250 MA 11/8" 156 156 456 456 601 250 MA 250 MA 602 1% 6Q4 (†) 5QS1 11/2" sq. 130 380 300 MA 11/2" x 2' 11/2" x 2' 11/8* 350 MA 130 380 6052 156 456 350 MA 11/4" 2" sq. 2" sq. 551 11/8" 130 380 500 MA 6S2 1% 156 456 500 MA * This rectifier is rated at 25 MA (†) Stud mounted-overall: 2" when used with a 47 ohm series resistor. RR SELETRON DIVISION RR RADIO RECEPTOR COMPANY, INC. Sales Department: 251 West 19th St., New York 11, N. Y. Factory: 84 North 9th St., Brooklyn II, N. Y.

is used, and spacers may not be necessary in some applications.

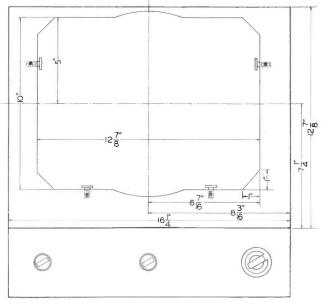


Figure 8. Dimensions of Sub-Panel Cutout.

The next step was that of making the cutout in the front panel. The dimensions of the new opening are given in Figure 9. The dotted line shown in this figure represents the decorative groove that is cut in the front panel. The top and bottom grooves may be used as guide lines while the sides should be cut 3/8

inch beyond the side grooves. This size opening is

proportional to the face of the tube. Note that the corners have a 1/2 inch radius curve which gives a neater appearance than would be obtained if the corners were perfectly square. The edge of the cutout should be sanded and stained. The rounded corners can be sanded by wrapping the sandpaper around a GT type vacuum tube.

The panel was then ready for the mounting of the mask. Figure 10 is a back view of the panel with the mask in place. In order to prevent the face of the plastic mask from being scratched, a piece of glass was placed in front of the mask. The use of this glass is optional. However, not only will it protect the mask, but it will also prevent dirt from falling between the mask and the panel. Since the glass has a much smoother surface than the mask, a better seal to the back of the front panel can be obtained. Three strips were then cut which had the same thickness as the combined thickness of the mask and the glass. These strips were glued to the back of the front panel to form a sort of frame for the mask. After the glue had dried, three more wider strips were mounted by screws so that the inside edge lapped over the mask, holding it in place, as shown in Figure 10. Extreme care must be taken that the screws are not too long so as to extend through the panel and come out the front which would obviously ruin the appearance of the front panel. They should be long enough, however, to extend just into the panel itself so that the support of the mask does not depend upon the glued strips alone.

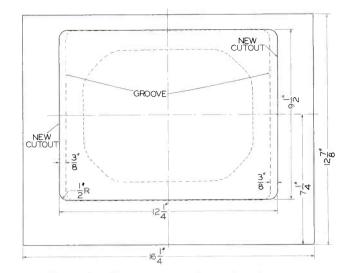


Figure 9. Dimensions of Front Panel Cutout.

With the mounting arrangement described above, the mask and glass can be very easily removed for cleaning by loosening the screws that hold the mounting strips, and sliding the mask and glass up out of the frame.

This completed the work on the front panel. It was then installed on the cabinet to determine what thickness spacers would be required to bring the front panel out flush with the radio panel. These spacers may be mounted as shown in Figure 7 or they may be slipped over the two screws that hold the front panel in position.

The chassis was then installed in the cabinet and the 14" tube was inserted from the front. The bolts which hold the chassis should not be tightened, as yet, to permit slight shifting of the chassis if required. The tube mounting brackets were then adjusted to properly position the tube and the front panel was installed. If the tube is too far forward, it is necessary to loosen the bolts holding the deflection yoke mounting hood, allowing it to be shifted back. This should allow the TV front panel to fit flush with the radio panel. If required, the complete chassis may be shifted a small amount to allow the tube to move back far enough into the cabinet. After the front panel is in position, the chassis bolts should be tightened and the deflection yoke mounting hood should be moved forward as far as possible. This is important, as failure to do so may result in the cutting off of the corners of the picture. When the tube is properly positioned, the tube should be clear forward against the mask, and the deflection yoke should be against the bulb of the tube. Check to see that the tube is not pressing against the high voltage compartment or any of the components at the front of the chassis.

The ion trap was then mounted on the tube and the high voltage and tube socket connections were made. The audio and AC plugs were inserted into the proper sockets in the radio chassis. The set was then turned on and the ion trap setting as well as the required deflection yoke and focus coil adjustments were made. It is suggested that the set be operated for a few hours before delivery is made to check for any possible failure due to lead dress or faulty components. Check especially for any arcing or corona discharge in the high voltage compartment.

All of the wiring changes outlined in this article will also apply to the RCA Victor model 730TV2, which is an identical set except for the cabinet. The chassis incorporated in models 721TS and 721TCS are identical to the TV chassis of the above models except that an audio output is incorporated. Therefore the wiring changes required for conversion will also apply to these models. The cabinet changes outlined apply to the model 730TV1 only and cannot be used on any other model. The instructions for converting the chassis may serve as a guide, however, in making changes on other models.

ADDITIONAL RCA VICTOR TELEVISION RECEIVERS SUITABLE FOR CONVERSION

The following list of RCA Victor TV receiver models have similar deflection circuits and can be converted by following a similar procedure to that outlined under "Wiring changes" in this article.

8T241	8TV323	9TC 247	T121
8T243	9T240	9TC 249	TA128
8T244	9TC 240	9TW309	TA129
8TK29	9T246	9TW333	TC124
8TR29	9T256	T100	TC125
8TV321	9TC245	T120	TC127

No change should be required in the vertical deflection circuit of any of these models since all of them employ a type 6K6GT tube which ordinarily supplies enough power for vertical deflection.

If the horizontal circuit of any of these receivers is rewired as shown in the schematic of Figure 5, it should adequately sweep from 14'' to 20'' picture tubes. It should be kept in mind that the cabinet design of most of these receivers will not accommodate a larger tube. If the chassis is to be mounted in another cabinet, however, such a conversion can be made.

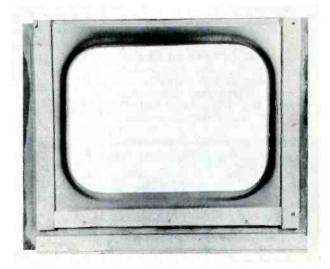
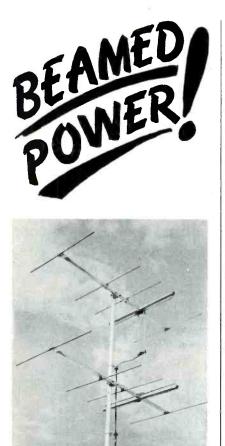


Figure 10. Rear View of Mask Assembly.



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Any methods or techniques that are offered here should be considered as suggestions rather than recommendations, since they may not necessarily be the easiest way to effect the conversion. However, our experiments on these models have enabled us to present data known to be effective.

PARTS LIST

1 - Horiz. Output	Merit HVO-6 or
Trans.	HVO-7
1 - Width Coil	Merit MVC-1
1 - Deflection	(Merit MO-70
Yoke	or MD-70-F
	(Stancor DY-7
1 - Focus Coil	Merit MF-3*
1 - 15KV or 20KC	(Aerovox HV20C
H.V.Filter	(CRLTV2 - 502
Capacitor	(Erie 410-501
1 - 2200 Ohm,	IRC BTA-2200
1 Watt	
Resistor	
1 - 39 Ohm,	IRC BW-1-39
1 Watt	nte bit i bb
Resistor	
1 - 6BL7GT	Sylvania Type
Tube	6BL7GT
1 - Single Magnet	
Ion Trap	
1 - Mask for 14"	
I = MIASK IUL 14	

Rect. Tube

* If this chassis is converted to employ a short-neck type picture tube, a thin focus coil unit such as the Merit type MF-2 should be employed.

TELEVISION: As a solution to appearance problem of indoor TV antennas, Volk-M-Rick Co., Minneapolis, Minn., is offering a flower pot with built-in antenna. (Tide)



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PICTURE TUBE REPLACEMENT DATA

The picture tube replacement chart on the facing page lists the standard replacement type tube which can be used to replace practically all type picture tubes that have been used in television receivers. In some cases it will be found that the original type tube is no longer in production and cannot be obtained. The type tube which is available and will make an effective replacement can be determined from the replacement chart.

Particular attention should be paid to all notes following a replacement listing. Each note indicates a variation of the replacement tube from the original and it should be determined if this variation would prevent the use of the replacement in each particular application. This is especially true on those replacement listings which are followed by Note 5. In these cases the replacement tube is slightly larger than the original and measurements should be made to determine whether the cabinet and mounting structure will accommodate the larger unit.

The presence of Notes 1 and 6 indicates that a different style ion trap magnet is required on the replacement tube than was used on the original unit while Note 3 indicates that no ion trap magnet is required on the replacement tube.

Where Note 2 follows a replacement listing care must be taken after the new tube is installed to see that the outer coating of the tube is grounded. Since the original tube did not have a coating there is a possibility that the manufacturer did not provide a grounding clip. If such is the case a grounding clip should be added to prevent arc-over from the coating to ground. Note 4 indicates that the high voltage connector must be changed and Note 7 indicates that a high voltage capacitor must be added at the tube end of the high voltage filter resistor. This is required because the replacement tube listed does not have an outside coating to perform the required filtering.

Where no listings are given in the replacement columns, no substitution of tube types should be made.

PICTURE TUBE REPLACEMENT DATA

ORIGINAL TUBE	SYLVANIA Replacement	THOMAS Replacement	ORIGINAL TUBE	SYLVANIA Replacement	THOMAS Replacement
7JP4	7JP4		16HP4,A	16JP4,A	16DP4A
8BP4	8BP4		16JP4,A	16JP4A	Note 6 16JP4A
ODP4	UDF4	33 -	16KP4,A	16KP4	16KP4
		10BP4A	16LP4,A	16LP4A	16CP4A
10BP4,A	10BP4,A 10FP4	10BP4A	1011 4,11		Note 6
10CP4		Notes 1,5	16MP4,A	16JP4,A	16DP4A
10004	Notes 4,5	Notes 1,5	100011-1,71	1001 1,11	Note 6
10DP4	10BP4,A	10BP4A	16QP4	16KP4	16KP4
10EP4			10414	Notes 2,6	Notes 2,6
10554 4	Notes 2,4	Notes 2,4 10BP4A	16RP4	16RP4	16KP4
10FP4,A	10FP4A	Note 1	16SP4,A	16WP4A	16DP4A
107704	10004	Note 1	10514,4	10101 111	Notes 5,6
10HP4	10HP4		16TP4	16TP4	16KP4
10MP4,A	10MP4		101 P4	10124	Note 5
			100004	16KP4	16KP4
12AP4			16UP4		Note 2
12CP4			1 0 1	Note 2	16DP4A
12JP4	12KP4	12LP4A	16VP4	16JP4,A	
	Notes 2,3,4	Notes 2,4,5,6		Notes 1,2,5	Notes 2,5
12KP4,A	12KP4A	12 LP4A	16WP4,A	16WP4A	16DP4A
		Notes 5,6		Note 2	Notes 2,5,6
12LP4,A	12 LP4A	12LP4A	16XP4	16KP4	16KP4
		Note 6		Notes 2,6	Notes 2,6
12QP4,A	12LP4,A	12 LP4A	16YP4	16JP4,A	16DP4A
	Notes 1,2,4,5	Notes 2,4,5		Notes 1,5	Note 5
12RP4	12LP4,A	12LP4A	16ZP4	16ZP4	16CP4A
	Notes 1,2,4,5	Notes 2,4,5			Note 6
12TP4	12LP4,A	12LP4A			
	Note 2	Note 2,6	17AP4	17BP4A	17BP4
12UP4,A,B		Glass Equivalent		Note 5	Note 5
	12LP4,A	12LP4A	17BP4,A,B	17BP4A	17BP4
	Notes 2,4	Notes 2,4,6		Note 2	
12VP4,A	12VP4A	10000 2,1,0	17CP4	Glass Equivalent	Glass Equivalent
12WP4	12 11 14			17BP4A	17BP4
14 ₩ ₽ ٩				Notes 2,4	Notes 2,4
14BP4	14BP4	14CP4	17FP4	17FP4	
IIDFI		Note 6			
14CP4	14CP4	14CP4	19AP4,A,B,C,D	19AP4A,B,C,D	Glass Equivalent
14DP4	14CP4	14CP4		, , , ,	19BP4A
IIDFI	Notes 2,6	Notes 2,6			Notes 2,4
14EP4	14CP4	14CP4	19BP4,A		19BP4A
IILFI	Note 5	Note 5	19DP4,A		19BP4A
14GP4	14GP4	Note b			Note 6
14074	1401 1		19EP4		
15AP4	16LP4,A	16DP4A	19FP4		19BP4A
IJAP4	Notes 1, 2,4,5				Notes 2,6
15 CD4	16LP4,A	16DP4A	19GP4		19BP4A
15CP4		Notes 2,5,6	i bai i		Notes 2,5
15004	Notes 2,5	16DP4A	19JP4		
15DP4	16LP4,A				
	Notes 1,2,4,5	Notes 2,4,5	20BP4		
10404 4 0	164044	Glass Equivalent	20CP4,A	20CP4,A	20CP4
16AP4,A,B	16AP4A	_	20DP4,A	20DP4,A	2001 1
		16CP4A	20FP4	2001 1,11	
10004	101 04 4	Notes 2,4,6	20GP4	20GP4	
16CP4,A	16LP4,A	16CP4A	20014	20014	
14001	Notes 2,5	Notes 2,6			
16DP4,A	16JP4,A	16DP4	NOTES:		
	Note 2	Notes 2,6	NOTES.		
16EP4,A,B		Glass Equivalent	1 77 . 1 11.	agent is a har	
	16JP4,A	16DP4A	1. Use double m	-	
	Notes 2,4,5	Notes 2,4,5,6		must be grounded.	
16FP4	16JP4,A	16DP4A	3. Ion trap not u		.1
	Notes 1,2,4,5	Notes 2,4,5,6		connector must be	changed.
16GP4,A,B,C	16GP4,B	Glass Equivalent	5. Space permitt		
		16DP4A	6. Use single ma	agnet ion trap.	

Notes 2,4,5,6

59

TABLE I

RATED IMPEDANCE IN PLATE CIRCUIT FOR VARIOUS RECTIFIER TUBES

			RATED
TYPE	BASE	PLATE PINS	IMPEDANCE
5AX4	5т	4 and 6	50
5AZ4	5 T	4 and 6	50
5U4G	5T	4 and 6	170
5V4G	5L	4 and 6	100
5W4GT	5T	4 and 6	50
5Y3GT	5т	4 and 6	50
5¥4G	5Q	3 and 5	50
5Z3	4C	2 and 3	170
5Z4GT	5 L	4 and 6	50
6AX5GT	6S	3 and 5	50
6X4	5BS	1 and 6	150
6X5GT	6S	3 and 5	150
7Y4	5AB	3 and 6	150
7Z4	5AB	3 and 6	75

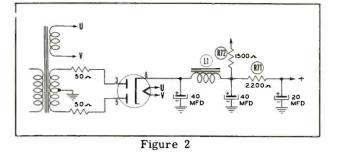
3. The current drain to be required.

4. The heating time of the rectifier tube compared to the rest of the equipment.

These problems will now be considered one at a time.

Filament Requirements -

It is important to compare the ratings of the proposed substitute tube to learn whether the present filament winding on the transformer will handle the additional current if any is required. If it is desired to substitute a cathode type six volt heater tube for a 5 volt one, it is necessary to determine whether the six volt heater winding will handle the added drain and also whether the present six volt winding is tied to ground or chassis at any point. If it is, then there might be a greater possibility of a



heater to cathode short developing than if the heater is floating. Table II indicates the filament requirements of several of the most commonly used rectifier tubes.

Voltage Drop in the Tube -

The voltage drop in the tube is a function of the spacing between the plate and cathode (or filament in directly heated tubes) and the current flowing. In a given circuit which requires a certain plate current drain, and a given power transformer, the relation between voltage drop and voltage applied to the filter is such that the total remains practically constant. Figure 3 illustrates this and "A" designates "tube drop" while "B" indicates that voltage available at the filter. The sum of "A" and "B" is constant for a given receiver application. Thus the greater the voltage drop ("A") in the tube, the lower the voltage across the filter ("B"). A comparison of these values in Table II for the original tube type and the proposed type will indicate whether this problem will be encountered if the substitution is made. If "B" is much higher than the rating of the filter capacitors, breakdowns might occur if that substitution were carried out.

Current Drain -

When a substitution is contemplated it is necessary to make certain that the proposed tube can supply the required current without being overloaded. Table II also gives a listing of current ratings for various types of rectifiers.

TA	BLE	II
----	-----	----

RECT. TUBE TYPE	VOLTAGE	FILAMENT CURRENT (AMPERES)	HEATING TIME	RATED PLATE CURRENT (MILLI-AMP.)	"A" TUBE DROP AT RATED PLATE CURRENT (VOLTS)	*'B'' *
5AZ4	5.0	2.0	FAST	125	60	Et-60
5U4G	5.0	3.0	FAST	225	58	Et-58
5V4G	5.0	2.0	SLOW	175	25	Et-25
5W4GT	5.0	2.0	FAST	110	50	Et-50
5Y3GT	5.0	2.0	FAST	125	60	Et-60
5Y4G	5.0	2.0	FAST	125	60	Et-60
5Z3	5.0	3.0	FAST	125	58	Et-58
5Z4GT	5.0	2.0	SLOW	125	20	Et-20
6X4	6.3	0.6	SLOW	70	22	Et-22
6X5GT	6.3	0.6	SLOW	70	22	Et-22
7Y4	6.3	0.5	SLOW	70	22	Et-22
7Z4	6.3	0.9	SLOW	100	40	Et-40
80	5.0	2.0	FAST	125	60	Et-60

* E_t indicates the voltage rating of one-half of the secondary (high voltage) winding. The expression E_t -60, as used in the text, actually means the maximum <u>DC</u> voltage developed minus the 60 volt drop in the tube. Transformer regulation, input filter capacitance, etc., represent variables so that it is impossible to translate the transformer AC voltage rating into the developed DC voltage value without these variables being known.

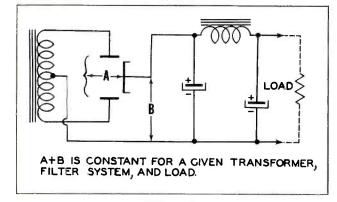


Figure 3

Heating Time -

The heating time of a directly heated tube is much less than that of an indirectly heated tube.

Continued from page 29 .

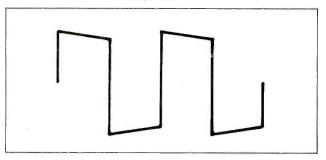


Fig. 9. Amplifier with Improvement in Low Frequency Response.

These waveforms were taken through amplifier stages employing RC (Resistance-Capacitance) coupling.

In general, a rounding off or dropping of the leading edge of the wavefront points toward deficiencies in high-frequency characteristics, while similar characteristics in the trailing edge can be interpreted as deficiencies in the amplifier's lowfrequency response.

At the time of writing, we have not completed our experiments in either the RC field, or more particularly, the inductively coupled systems. In terms of either accentuation or attenuation of bands of frequencies, inductively coupled systems are far more likely to have such characteristics.

Additionally, since inductively coupled systems have sharper resonance properties, we can expect to find greater tendencies toward oscillation in the waveform.

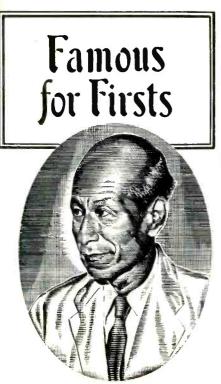
We intend to continue presenting the results of our experiments in this and allied fields as they are available. Again, we make no attempt to formulate conclusions from this material; rather, we are interested in providing a basis for study and experimentation on behalf of those interested in this activity. This means that voltage will be available at "B" in Figure 3 much sooner than with an indirectly heated tube. If the tubes constituting the load in the receiver are of the cathode or indirectly heated types then they will not require current for an appreciable time after there is voltage available at "B" if the rectifier tube is a fast heater. This would cause the voltage at "B" to increase considerably until the receiver began to draw current. This increase is due to the fact that with little or no current drain the voltage drop in the rectifier tube ("A") is very small. This may be further increased by the regulation of the power transformer. Thus if the substitution of a fast heating rectifier is contemplated, steps must be taken to prevent this voltage rise. Table II also includes a column indicating slow or fast heating time.

The substitution of rectifier tubes requires consideration of many factors. It is hoped that this article will help the reader to make intelligent substitutions so that receivers may be kept in operation.

PARTS LIST

Item	Description	
V1	Type 6SN7GT Tube	Sylvania or equivalent
V2	Type 6AL5 Tube	Sylvania or equivalent
V 3	Type 6SJ7 Tube	Sylvania or equivalent
R1	500K Ohm Input	(IRC Q13-133
	Control	(Clarostat AG060-Z, (FS-3
		(Centralab B-60
R2	100K Ohm Output	(IRC Q13-128
	Control	(Clarostat AG-51-Z,
		(FS-3
		(Centralab B-41
R3	4700 Ohm 1/2 Watt	IRC BTS - 4700
R4	220K Ohm 1/2 Watt	IRC BTS - 220K
R5	470K Ohm 1/2 Watt	IRC BTS - 470K
R6	27K Ohm $1/2$ Watt	IRC BTS - 27K
$\mathbf{R7}$	47K Ohm $1/2$ Watt	IRC BTS - 47K
R8	1.5 Megohm	
- 0	1/2 Watt	IRC BTS - 1.5 Meg.
R9	47K Ohm $1/2$ Watt	IRC BTS - 47K
R10	4700 Ohm 1/2 Watt	IRC BTS - 4700
C1	10 Mfd. @ 25 Volts	(Aerovox PRS 25/10
		(Cornell-Dubilier
		(BR102A
		(Sprague TVA-1204
C2,C3	3 16 Mfd. @ 450 Volts	(Aerovox PRS 450/16
		(Cornell-Dubilier
		(BR1645A
		(Sprague TVA-1707
C4	.05 Mfd. @ 600 Volts	(Aerovox P688-05
		(Cornell-Dubilier
		(PTE 6S5
		(Sprague 6TM-S5
T1	Interstage Transformer	(Stanco A-4155*
	1 to 3 pri. to sec.	(Merit A-2912*
	ratio	(Chicago IN-15*

* Additional cathode resistance may be necessary if DC resistance of primary winding is less than 1100 ohms. If resistance must be added, connect between low side of primary winding and ground, and bypass the resistor with 10 Mfd. @ 25 volt electrolytic capacitor.



DR. HIDETSUGU YAGI

World famous scientist, originator of the YAGI antenna principle.

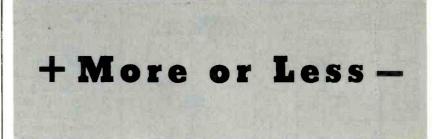
> Analyzing the Ward engineered YAGI TV antenna design, Dr. Yagi recently wrote:

"The low numerical value of Voltage Standing Wave Ratio as recorded is . . . the proof of the exact matching between circuit elements. In this regard, I highly esteem the excellent ability of your engineers."

Ward antennas are the result of over 20 years of design and production experience. Ward, the oldest and largest exclusive manufacturer, builds the world's finest antennas.



THE WARD PRODUCTS CORP. Division of The Gabriel Co. 1523 East 45th Street Cleveland 3, Ohio



On behalf of our entire organization, we would like to extend sincere thanks to all those who filled out and returned questionnaires enclosed with the first issue of PHOTOFACT INDEX and TECHNICAL DIGEST.

We appreciate how difficult it is to find time in a busy day, or in the rarely uninterrupted evenings, to give the amount of consideration to this matter, which is evident in the extent of the replies and in the detail of their contents.

You know a lot of firms in all sorts of endeavors use the questionnaire approach as a means of obtaining statistics. Immediately upon their return, someone is assigned to assemble an impressive stack, a posed photograph produced, and suitable figures turned over to the Sales and Advertising Departments for blow-up or dissemination.

If that were our objective in sending out the questionnaire, then we in no way would be worthy of the generous response which we have received from you.

We simply would like to go on record that we had a genuine interest in finding out what the service field needed in terms of technical help and general information . . . and it is gratifying to us to find the same genuineness in your expressions.

We can tell you quite frankly that it was our original intention (and will continue to be) to fashion PF INDEX and TECHNICAL DIGEST to the interests and desires of the service technician. The start you have given us should guarantee the success of this policy.

Just in case you're curious about how your preferences compare with nationwide averages of questionnaires analyzed to date, here is the ranking of the first ten subject treatments most requested. Also listed are their percentages of total requests.

	Subject	% of Total
1.	Test Instrument Applications	10.7
2.	TV Receiver Difficulties and General	
	Trouble Shooting	9.7
3.	General Circuit Analysis	8.3
4.	Short Cuts in Servicing	6.0
5.	Fringe Area Information (Antennas,	
	Boosters, Installations, Alignment)	6.0
6.	Waveform Analysis (What we see, not what	
	we should see)	4.86
7.	TV Receiver Conversions	4.3
8.	Latest Circuit Designs	4.3
9.	TV Tuner Data (more)	4.0
10.	Color Television	4.0

To all those who included some form of inquiry in the returned questionnaires, may I say that we are handling them just as fast as possible. If you have not already received a reply, you will in the near future. While we are attempting to make individual analyses of each questionnaire returned, and to forward individual replies, I believe you can realize the size of this undertaking and the consequent length of time that may be required to complete the job.

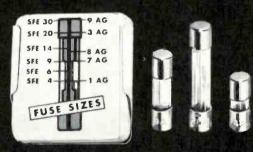
Please, then, let this column express an overall "Thank you" until individual correspondence can catch up.



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Conversion

TO:

