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# electronics today

INTERNATIONAL

June 1983 85p

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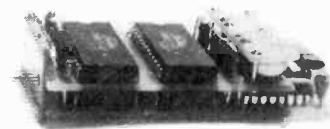
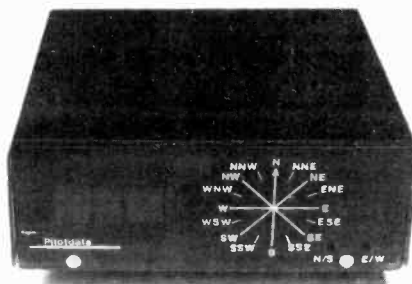
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## FEATURES

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- CONFIGURATIONS** . . . . . 59  
In his penultimate discourse on electronics, Ian Sinclair sheds some light on the field of opto-electronics.
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- ATOM KEYPAD** . . . . . 78  
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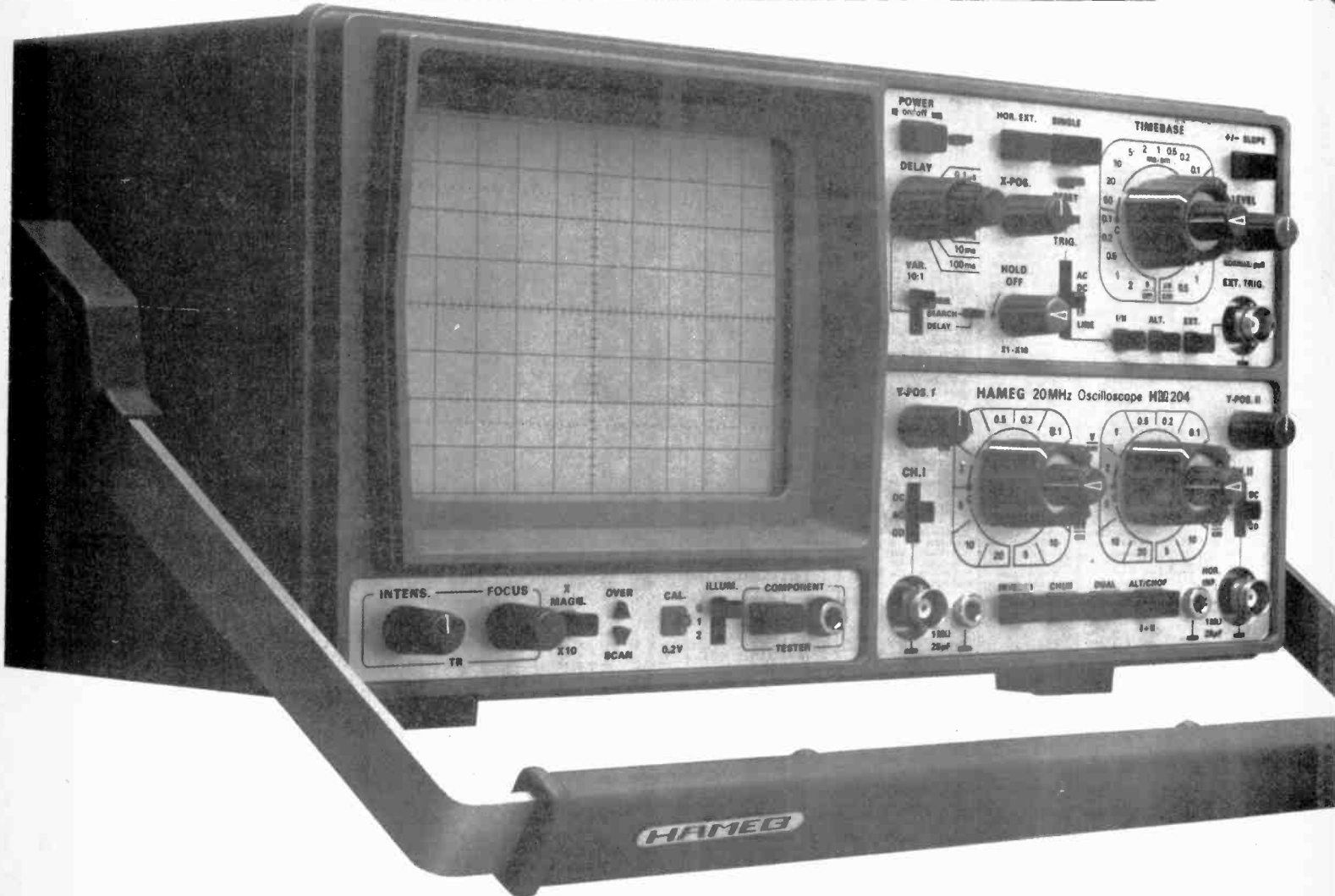




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NEXT  
MONTH



## Portable Induction Loop

Following the degree of interest that our feature on Induction Loops aroused, Vivian Capel has designed a practical system. Now there's no excuse for the hard-of-hearing to be left out of things anymore.

## Tech Tips Special

Eight pages of ingenious, novel, and money-saving designs from one of the country's largest design teams — yourselves.

## Microcomputer Output Driver

Don't let your micro just sit there, contentedly passing information back and forth inside itself — get it busy working for *you*, with this 16-channel output driver.

## TV Storage Scope

This little unit will enable you to utilise your TV screen to display oscilloscope waveforms. It's always seemed illogical

that you have to use two CRTs for different jobs, hasn't it — now you can use just the one. Not only that, but the unit will store waveforms too, all for under £100.

## Oscilloscope Survey

Just in case you're too lazy to build the TV Scope, we're taking a look at the options open if you want to buy one. And if you're too poor to afford a new scope, we'll be examining the state of the second-hand market too. If you're on the look-out for test equipment, don't miss the July issue of ETI.

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ON SALE JUNE 3rd

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Articles described here are in an advanced state of preparation. However, circumstances may dictate changes to the final contents.







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K2551	Central Alarm Unit	15.48

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With a microcomputer as an alternative keyboard the world is even greater adding bulk updating to viewdata computers or receiving telesoftware for implementation to any personal computer.

Even without the Prestel option, Telesoftware from the Teletext pages free!

The full features of Teletext, including subtitles are all included in the basic kit.

An attractive stylish case is available to complement the finished kit.

Basic Teletext Kit (no box) £130 + VAT P/P £2.50

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box by itself £14.95 + VAT P/P 75p

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A Prestel micro computer adaptor to give full autodialing to your micro computer. All the usual Prestel facilities are added via this unit, plus many more, and, can operate to any viewdata computer.

You can shop from home, bank transmitt messages and receive software, which means that the uses your micro can be put to are limitless.

The unit is not restricted to just the UK, for at least 28 countries use the Prestel viewdata format, so you can also mail-order from anywhere. The Prestel unit is suitable for most micro computers even the ZX-81, so at the push of a button, the technology of tomorrow is in your home today.

## ANTEX

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2532	3.50
2764 (200ns)	11.00
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# DIGEST

## Compressor/ Limiter Tips

Ian Martin, designer of the Compressor/Limiter project we ran last month, has written suggesting some setting up hints. He thinks it would be wise to check the operation of the DC side chain before inserting the VCA chips IC2, 5 and 7: if there was a fault in the side chain which resulted in no current being passed into the control pins on IC7 (pins 1 and 16), it's possible that the output of IC2 and IC5 could swing to the positive or negative supply rail and damage the LM13600 which follows. For this reason we suggest that these ICs are removed and the input currents to pins 1 and 16 of IC7 are measured. These should be in the region of 66 uA ( $V_{MIN}/7k5$ ), and  $V_{MIN}$  should be 0V5. Both these measurements should be made with no input signal present.

## New From ILP!

15 VA transformers — all fully encased in ABS plastic shells with easy fixing by an M4 bush at the base. ILP are planning to extend this facility throughout the year to cover transformers up to 120VA. ILP Electronics, Graham Bell House, Roper Close, Canterbury, Kent CT2 7EP.

## Two-way Tapping

Complete two-way 'live' telephone conversations can be recorded with the Ansafone 600. This microprocessor controlled unit switches from its usual role as a telephone answering machine to a two-way conversation recorder in less than a second. Ansafone, Lyon Way, Frimley Road, Camberley, Surrey.

## The Sinclair That Never Forgets

ROM-81 is a memory expansion unit for the ZX81 personal computer which enables the user to read useful routines and commonly used information, stored in U.V. erasable, programmable Read Only Memory. The unit is supplied without EPROMs as these are normally programmed and provided by the user.

Two 24 pin sockets allow either 2716 or 2732 EPROMs to be used. They can provide up to 8K of memory in 2K increments. The sockets are decoded to lie between 8K and 16K in the ZX81 memory map, which is just below the BASIC area. Separate 2K and 4k decoding is link selectable to make it possible to vacate locations occupied by other peripheral cards.

ROM-81 has additional circuits to allow the use of slow EPROMs. The most popular EPROMs have a maximum access time of 450 nS. This is too slow for the ZX81. A special Wait State circuit in ROM-81 automatically requests the CPU in the ZX81 to wait until data has been read. Wait States do very slightly decrease the speed of operation of the computer and affect precise calculations of delay loops. The key device has therefore been socketed. Removing it will prevent implementation of Wait States.

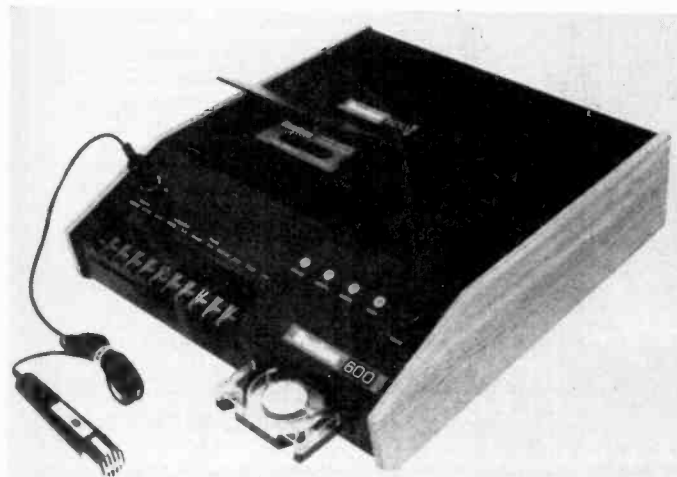
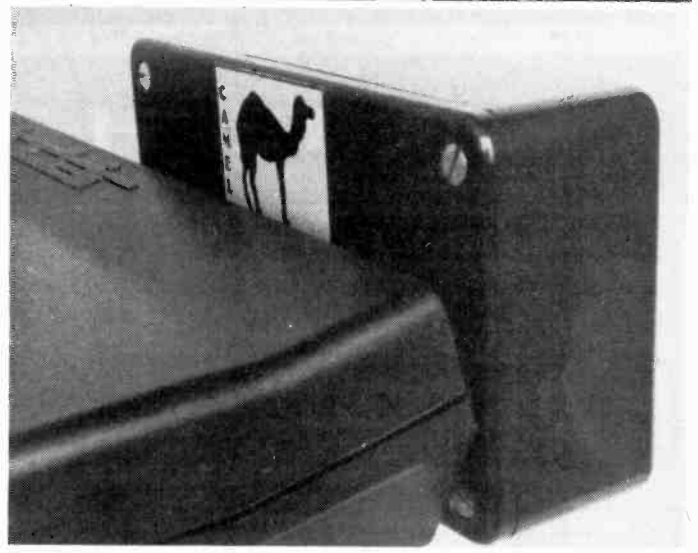
ROM-81 comes in a black ABS case with a screwed down cover for quick accessibility without vulnerability. It plugs on to the ZX81 with an adaptor at the rear of the box for further expansions. It is supplied with easy to follow User Notes which give the programs for data retrieval. (But what is the camel for? — Ed).



## Megamania Mangled

We did it! We penetrated the deep mysteries of what lies beyond the magic million of Megamania (new readers will not be aware that this is the best game in the Universe). After a finger-mangling fight with the Atari joystick during a game that

dragged on for an hour and a quarter, Peter "I'll beat it if it kills me" Green finally got to 999,999 with four spare lives, as evidenced by this photograph. The result was something of an anticlimax, as the game simply latches up. No fanfare, no screen message of congratulations, no dancing girls — the game just dies, its little software counter overloaded. Pity, that — now we can only score it  $9\frac{1}{2}$  instead of 10.



## New Catalogue From Toolmail

The new 1983 Toolmail 128-page full colour catalogue is now on sale, offering better value than ever before. With 600 new items and over 2,000 tools illustrated, the product range available must be larger than any specialist tool shop. Prices are still really competitive and all items are available for immediate delivery through Toolmail's efficient nationwide mail order service.

In addition, there are many new features in the 1983 Toolmail

catalogue. Of particular interest are the three £1 vouchers to be found on page 128. Valid until July 31, 1983 they are each redeemable on any order in excess of £10.

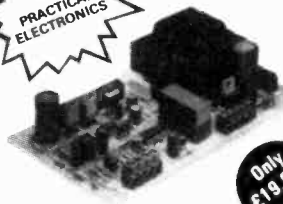
A further Toolmail venture for 1983 is the opening of the Sevenoaks Tool Room in March. Many of the tools in catalogue will be available there. Others may be ordered there through the regular Toolmail service for delivery by post. Customers may also redeem their Toolmail vouchers at the Sevenoaks Tool Room.

The catalogue is available in all major newsagents price £1. Copies are also available direct from Toolmail (1982) Limited, PO Box 46, Maidstone, Kent ME15 8EQ.

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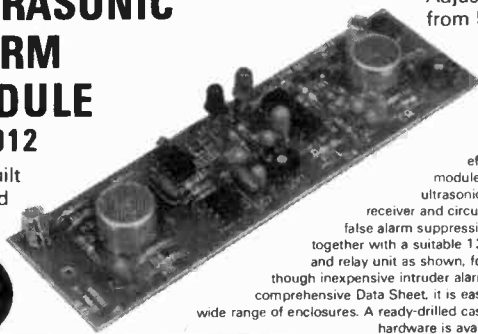
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- Screw connections for ease of installation
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Fully built  
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£10.95  
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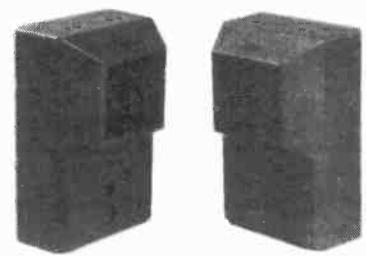
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## Power Supply & Relay Units PS 4012

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BC109C	13p	BD136	30p	MPSA06	18p	LM1458	70p	7812	33p
BC113/4	15p	BD138	35p	MPSA12	20p	LM1459	47p	7815	33p
BC115/6	18p	BD140	35p	MPSA14	22p	MC1455	15p	100MA T092	50p
BC117	18p	BD204	64p	MPSA42	23p	MC1458	30p	78L05	30p
BC119	18p	BD206	64p	MPSA43	23p	MC1496 L2	70p	78L12	30p
BC139	30p	BD222	56p	MPSA55	18p	MC3302	72p	78L15	30p
BC140	28p	BD239A	40p	MPSA63	22p	MC3401	30p		
BC142/3	30p	BD239C	40p	MPSA64	22p	MC3403	30p		
BC160	28p	BD240A	42p	MPSA92	24p	MC3456	23p		
BC161	32p	BD240C	48p	MPSA93	24p	NE529	42p		
BC169C	9p	BD241A	50p	TIP29A	30p	NE532	80p		
BC170	9p	BD242A	52p	TIP29B	30p	NE544	200p		
BC171	9p	BD243A	50p	TIP29C	30p	NE550	15p		
BC172	9p	BD234C	65p	TIP30A	40p	NE555	24p		
BC173	9p	BD244A	60p	TIP30B	30p	NE568	18p		
BC177	16p	BD244C	78p	TIP31A	33p	NE567	14p		
BC178C	16p	BF180	36p	TIP31B	33p	NE567	14p		
BC178	16p	BF181	30p	TIP31C	37p	NE568	14p		
BC182	16p	BF182	30p	TIP32A	33p	NE568	14p		
BC182L	16p	BF183	30p	TIP32B	33p	NE567	14p		
BC184	16p	BF184	30p	TIP32C	37p	NE570	37p		
BC184L	16p	BF185C	29p	TIP41	65p	NE571	35p		
BC212	12p	BF194	12p	TIP110	18p	NE594	22p		
BC212L	12p	BF195	12p	TIP111	35p	NE644	38p		
BC213	12p	BF196	12p	TIP115	35p	NE645	27p		
BC213L	12p	BF197	12p	TIP120	35p	SL490	20p		
BC214	10p	BF198	10p	TIP121	60p	SN76108	135p		
BC214L	10p	BF199	10p	TIP122	60p	SN76115	15p		
BC217	10p	BF244C	10p	TIP122	60p	SN76660	43p		
BC238/9	10p	BF245	10p	TIP27	80p	T8A120S	70p		
BC251	10p	BF256C	25p	TIP295	70p	T8A200	70p		
BC300	40p	BF357	30p	TIP305S	70p	T8A820	90p		
BC301/2	36p	BF258	30p	TIS44	20p	TDA1022	55p		
BC303/4	36p	BF336	30p	TIS90	24p	TDA1024	120p		
BC327	12p	BF337	30p	2N1613	30p	TDA2002	98p		
BC329	12p	BF338	40p	2N1611	30p	TDA2003	130p		
BC337	12p	BF457	30p	2N1711	40p	TDA2004	240p		
BC338	12p	BF458	32p	2N1893	30p	TDA2030	200p		
BC413C	10p	BF459	38p	2N2218A	25p	TL061	40p		
BC415C	10p	BF460	28p	2N2222	25p	TL061	40p		
BC416C	10p	BF461	28p	2N2222A	25p	TL081	40p		
BC477	23p	BF485	28p	2N2368	20p	TL355	85p		
BC478	23p	BF486	28p	2N2369A	20p	TL356	85p		
BC479	23p	BF487	23p	2N2369B	20p	TL357	85p		
BC545	9p	BF488	23p	2N2904A	20p	LM301	25p		
BC547B	9p	BF490	23p	2N2904B	20p	LM307	25p		
BC548B	9p	BF511	22p	2N2905A	20p	LM339	40p		
BC549C	9p	BF512	22p	2N2905B	20p	LM338	40p		
BC550C	9p	BF513	28p	2N2906A	20p	LM358	60p		
BC555	10p	BF584	23p	2N2906B	20p	LM359	60p		
BC557B	9p	BF605	14p	2N2907	20p	LM380	12p		
BC558B	9p	BF608	15p	2N2907A	20p	LM382	25p		
BC559C	9p	BF609	15p	2N3053	23p	LM384	12p		
BC560C	9p	BF610	15p	2N3054	98p	LM386	50p		
BCV70	18p	MJE310	48p	2N3055	48p	LM387	50p		
BCV71	18p	MJE311	48p	2N3056	110p	LM388	130p		
BCV72	15p	MJE520	80p	2N3372	120p	LM389	150p		
BD115	50p	MJE521	80p	2N3702	48p	LM393	15p		
BD131	45p	MJE2955	80p	2N3703	91p	LM709	32p		
				2N3704	80p	LM741	14p		

★ FAST RELIABLE SERVICE ★  
★ VERY COMPETITIVE ★

## COMTECH ELECTRONICS

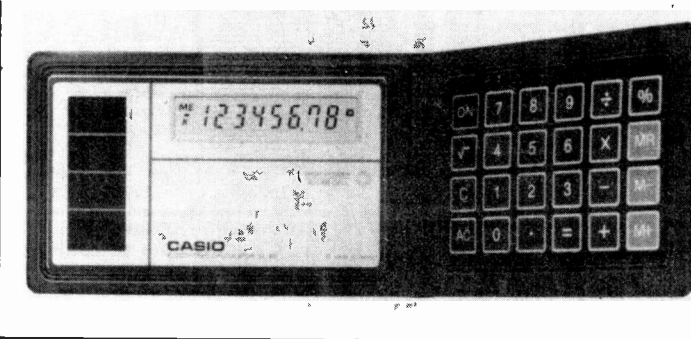
## MAIL ORDERS

205 STURDEE ROAD,  
LEICESTER LE2 9FY  
Telephone: (0533) 779578

## Centronics Interface For Spectrum

If you're frustrated with the limitations of the ZX printer or would like to utilize your Spectrum for business use you can now link your Spectrum to Centronics type printers with the use of a Kempston Centronics Interface.

A major feature is the recognition of LLIST and LPRINT by the interface. This allows programs to be listed directly from your Spectrum and also allows you to print-out direct from listings (BASIC only) without the need of special user calls. It is also possible to send out control codes to the printer giving the facility of different characters, i.e., condensed, expanded, etc. The interface is supplied cased and ready to use by simply plugging directly into a Centronics type printer, eg all Epsoms including MX-80 F/T III, Seikosha 100 A, OKI Microline 80 etc, and also includes driving



software which allows up to 128 characters per line (depending on printer type).

There is also a range of business software from Hilderbay Ltd which can be used in conjunction with Centronics type printers, i.e., accounts, stock control, etc. The interface complete with connecting lead will be priced at £45 including VAT (mail order £1 p&p) which includes a 12 month guarantee. Kempston (Micro) Electronics are at 180a Bedford Park, Kempston, Bedford MK42 8BL.



## Shorts

- The latest catalogue from Stotron Ltd has arrived in our offices; rather curiously it's dated 1982, but never mind, it's full of goodies. You can obtain your copy from Stotron Ltd, Haywood Way, Ivyhouse Lane, Hastings, East Sussex, TN35 4PL, telephone 0424 442160.

- Since MAC has been chosen as the UK's system for future direct satellite broadcasting (see ETI March 83), it's reassuring to know that it's been successfully demonstrated in Germany. About 100 telecommunications engineers saw Multiplex Analogue Component pictures which had been sent from the IBA Engineering Centre near Winchester, England, via the European Orbiting Test Satellite stationed 36,000 km above West Africa. The demonstrations included a direct comparison with conventional PAL-encoded colour signals.

- Gould Power Supplies UK, of Raynham Road, Bishop's Stortford, Herts, have expanded their Econoflex range of switch-mode power supplies with the addition of the EX24/5. Not surprisingly, this is a 24 V, 5 A output unit. The power output is floating and fully regulated to within 0.2%, protection circuitry is incorporated, and

the input can be 110 or 220 V AC.

- Hardly a month goes by without some new information from those busy chaps at OK Industries UK Ltd. This time they've produced a new brochure describing their range of PacTec enclosures, which are moulded from impact-resistant ABS in a range of sizes from small to medium. Various colours and accessories are available. Copies of the 22-page catalogue are available from OK at Dutton Lane, Eastleigh, Hants, SO5 4AA (telephone 0703 610944).

- Remember Zemco? They manufacture in-car microprocessor equipment and we've mentioned their products before in Digest. Anyway, the US parent company has bought out the UK operation in order to do their own marketing: the good news is that this move will result in a hefty 30% cut in prices, as well as a further financial investment in service facilities. Zemco UK's office is at 66 Earlsdon Street, Coventry CV5 6EJ (telephone 0203 79969), and readers interested in cruise controls and in-car computers are directed thence.

- Accelerate your Apple with the Accelerator II from Pete & Pam Computers, New Hall Hey Road, Rossendale, Lancs, BB4 6JG (telephone 0706 227011). For £299 you get a plug-in board con-

## Look, Mum, No Keys!

Casio's latest miniature calculator SL80 has no keys in the conventional sense. Instead it has flat, sensor-touch pads

which respond to finger pressure. The keyboard itself, forming part of an integrated protective wallet, is as large as the whole of a normal credit card calculator, so individual keypads are generously proportioned and well spaced making the calculator much easier to operate.

SL80 has no batteries either, saving the consumer the expense and inconvenience of having to replace a battery. Power comes from built-in solar cells, needing light intensity of only 50 lux. For the untrained 50 lux is the equivalent of a dull light for normal reading!

This tough, maintenance-free little instrument is coming into the shops at a recommended retail price of only £10.95. Casio Electronics Co Ltd, Unit 6, 1000 North Circular Road, London, NW2 7JD.

## Bite A Computer!

Husky, the world's first portable microcomputer, will be on show at this year's Hanover Fair, April 13-20, where it will be demonstrating its indestructible computing power to a deadly Amazonian piranha, immersed in a tank of water at UCSL Microsystems' stand (B5604) in CeBIT Hall 1 (but don't try typing anything in while it's there!).

Husky is now in use for data capture and portable data processing applications by scientific,

commercial and military users in Europe, the United States and other countries.

Husky's applications range from guided missile support to brewery stock-taking. Husky's large memory — up to 144K — allows it to be used independently for long periods. Husky is the first portable microcomputer to offer IBM's 2780 synchronous package, allowing direct communication with mainframe computers.

Husky is designed and manufactured in Britain by DVW Microelectronics, a member of the AIDCOM International group of companies.

taining a 6502C processor and 64K of memory which will boost the speed of the Apple II Plus from 1 MHz to 3.58 MHz. It can run all native Apple II software.

- Calling all radiologists, radiographers and clinicians — the second London Course in Whole Body Tomography (whatever that is!) will be held from 7-10 June 1983. Further information may be obtained from one of the organisers, Dr Janet Husband, CT Scanning Unit, Royal Marsden Hospital, Downs Road, Sutton, Surrey SM2 5PT (telephone 01-642 6011 ext. 496). The course is approved under HM67/27 for study leave purposes.

- A 50-page two colour catalogue is now available from STC Meridian, West Road, Harlow, Essex CM20 2BP. The range of products includes such components as picture tubes and microphones, buzzers, sounders, re-entrant horns, counters, timers and relays, as well as vacuum fluorescent displays, DC-to-DC converters, display drivers, infra-red diodes, phototransistors, LED and LCD displays, and printers.

- Flexible Switch Technology Ltd can now offer a low-profile QWERTY keyboard with an overall thickness of less than 2 mm. The keyboard is fully sealed and comes complete with connectors on two flexible tails. All

you have to do is plug them in to a suitable drive board. Two types are available: flat membrane switches or FST's new 'click' effect tactile buttons. You can have any colour you like as long as it's blue, black or white. Flexible Switch Technology are at Unit 31, Middlefield Industrial Estate, Sunderland Road, Sandy, Beds SG19 1RB (telephone 0767 80332).

- Sharp PC-1251 owners can now buy a spreadsheet program which allows as many as 26 columns, 100 rows, and over 200 'cells' on the sheet. You can either use it as a tool for solving the "what if" type of question, or as a mini-database with up to 100 names and telephone numbers. The program, called Easi-calc 1251, requires the CE125 printer-recorder and costs £14.95 including VAT from Elkan Electronics, 11 Bury New Road, Prestwich, Manchester M25 8JZ (telephone 061-798 7613).

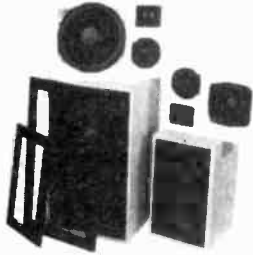
- Two new books have been added to the successful 'Understanding...' series from Texas Instruments. The titles are 'Understanding Electronic Control of Energy Systems' and 'Understanding Telephone Electronics', and each costs £3.95 plus £1.50 postage and packing from Texas Instruments Ltd, PO Box 50, Market Harborough, Leicestershire.



### MULLARD SPEAKER KITS

Purposefully designed 40 watt R.M.S. and 30 watt R.M.S. 8 ohm speaker systems recently developed by MULLARD'S specialist team in Belgium. Kits comprise Mullard woofer (8" or 5") with foam surround and aluminium voice coil. Mullard 3" high power domed tweeter. B.K.E. built and tested crossover based on Mullard circuit, combining low loss components, glass fibre board and recessed loudspeaker terminals. SUPERB SOUNDS AT LOW COST. Kits supplied in polystyrene packs complete with instructions. 8" 40W system — recommended cabinet size 240 x 216 x 445mm  
Price £14.90 each + £2.00 P & P.  
5" 30W system — recommended cabinet size 160 x 175 x 295mm  
Price £13.90 each + £1.50 P & P.

Designer approved flat pack cabinet kits, including grill fabric. Can be finished with iron on veneer or self adhesive vinyl etc.  
8" system cabinet kit £8.00 each + £2.50 P & P.  
5" system cabinet kit £7.00 each + £2.00 P & P.



### STEREO CASSETTE TAPE DECK MODULE

Comprising of a top panel and tape mechanism coupled to a record/play back printed board assembly. Supplied as one complete unit for horizontal installation into cabinet or console of own choice. These units are brand new, ready built and tested.

Features: Three digit tape counter. Autostop. Six piano type keys, record, rewind, fast forward, play, stop and eject. Automatic record level control. Main inputs plus secondary inputs for stereo microphones. Input Sensitivity: 100mV to 2V. Input Impedance: 68K. Output level: 400mV to both left and right hand channels. Output Impedance: 10K. Signal to noise ratio: 45dB. Wow and flutter: 0.1%. Power Supply requirements: 18V DC at 300mA. Connections: The left and right hand stereo inputs and outputs are via individual screened leads, all terminated with phono plugs (phono sockets provided). Dimensions: Top panel 5 1/2" x 11 1/2". Clearance required under top panel 2 1/2". Supplied complete with circuit diagram and connecting diagram. Attractive black and silver finish.  
Price £26.70 + £2.50 postage and packing.  
Supplementary parts for 18V D.C. power supply (transformer, bridge rectifier and smoothing capacitor) £3.50.



### LOUDSPEAKERS

15" 100 watt R.M.S. (HI-FI, P.A. DISCO, BASS GUITAR) Die cast chassis. 2" aluminium voice coil, white cone with aluminium centre dome. 8 ohm imp., Res. Freq. 20Hz., Freq. Resp. to 2.5KHz., Sens. 97dB (As photograph). Price: £32.00 + £3 carriage.

12" 100 watt R.M.S. (HI-FI) Die cast chassis. 2" aluminium voice coil. Black cone. 8 ohm imp., Res. Freq. 20Hz., Freq. Resp. to 4.5KHz., Sens. 95dB. (As photograph). Price: £23.50 + £3 carriage.

8" 50 watt R.M.S. (HI-FI, P.A.) 1 1/2" aluminium voice coil. White cone. 8 ohm imp. Res. Freq. 40Hz., Freq. Resp. to 6KHz. Sens. 92dB. Also available with black cone fitted with black metal protective grille. (As photograph). Price: White Cone £8.90, Black cone/grille £9.50 P&P £1.25.

12" 85 watt R.M.S. McKENZIE C1285GP (LEAD GUITAR, KEYBOARD, DISCO) 2" aluminium voice coil, aluminium centre dome, 8 ohm imp., Res. Freq. 45Hz., Freq. Resp. to 6.5KHz., Sens. 98dB. Price: £22.00 + £3 carriage.

12" 85 watt R.M.S. McKENZIE C1285TC (P.A., DISCO) 2" aluminium voice coil. Twin cone. 8 ohm imp., Res. Freq. 45Hz., Freq. Resp. to 14KHz. Price £22 + £3 carriage.

15" 150 watt R.M.S. McKENZIE C15 (BASS GUITAR, P.A.) 3" aluminium voice coil. Die cast chassis. 8 ohm imp., Res. Freq. 40Hz., Freq. Resp. to 4KHz. Price: £47 + £4 carriage.



### BK ELECTRONICS

Prompt Deliveries  
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Audio Equipment  
Test Equipment  
by  
Thandar  
and  
Leader



### OMP 80 LOUDSPEAKER

The very best in quality and value.

Ported tuned cabinet in hard-wearing black vynide with protective corners and carry handle. Built and tested, employing 10in British driver and Piezo tweeter. Spec: 80 watts RMS; 8 ohms; 45Hz-20Hz; Size: 20in x 15in x 12in. Weight: 30 pounds.

Price: £49.00 each  
£90 per pair

Carriage: £5 each £7 per pair

### PANTEC

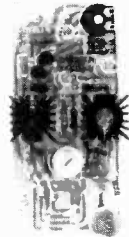
HOBBY KITS. Proven designs including glass fibre printed circuit board and high quality components complete with instructions.

FM MICROTRANSMITTER (BUG) 90/105MHz with very sensitive microphone. Range 100/300 metres. 5 x 46 x 14mm (9 volt) Price: £6.58.

DIGITAL THERMOMETER -9.9 C to +99.9 C LED display. Complete with sensor. 70 x 70 mm (9 volt) Price: £22.94

3 WATT FM TRANSMITTER / WATT 85/115MHz varicap controlled, professional performance. Range up to 3 miles 35 x 84 x 12 mm (12 volt) Price: £10.64

SINGLE CHANNEL RADIO CONTROLLED TRANSMITTER/RECEIVER 27MHz Range up to 500 metres. Double coded modulation. Receiver output operates relay with 2amp/240 volt contacts. Ideal for many applications. Receiver 90 x 70 x 22 mm 9/12 volt) Price: £14.38. Transmitter 80 x 50 x 15 mm (9/12 volt) Price £9.15. P&P All Kits +50p S.A.E. for complete list.

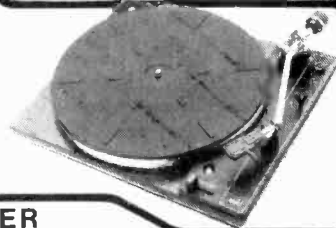


3 watt FM Transmitter

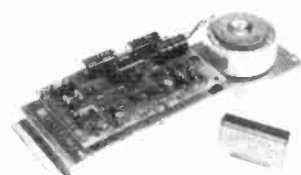
### BSR P256 TURNTABLE

P256 turntable chassis • S shaped tone arm • Belt driven • Aluminium platter • Precision calibrated counter balance • Anti-skate (bias device) • Damped cueing lever • 240 volt AC operation (Hz) • Cut out template supplied • Completely manual arm  
This deck has a completely manual arm and is designed primarily for disco and studio use where all the advantages of a manual arm are required.

Price £31.35 each. £2.50 P&P



### OMP POWER AMPLIFIER MODULE



New model.  
Improved specification

NEW OMP100 Mk.II POWER AMPLIFIER MODULE Power Amplifier Module complete with integral heat sink, toroidal transformer power supply and glass fibre p.c.b. assembly. Incorporates drive circuit to power a compatible LED Vu meter. New improved specification makes this amplifier ideal for P.A., Instrumental and Hi-Fi applications.

SPECIFICATION  
Output Power: — 110 watts R.M.S.  
Loads: — Open and short circuit proof 4/16 ohms  
Frequency Response: — 15Hz - 30KHz -3dB.  
T.H.D. — 0.01%  
S.N.R. (Unweighted): — -118dB -3.5dB  
Sensitivity for Max. Output — 500mV @ 10K.  
Price — £31.99 + £2.00 P&P  
Vu Meter Price — £7.00 + P&P

### PIEZO ELECTRIC TWEETERS - MOTOROLA

Join the Piezo revolution. The low dynamic mass (no voice coil) of a Piezo tweeter produces an improved transient response with a lower distortion level than ordinary dynamic tweeters. As a crossover is not required these units can be added to existing speaker systems of up to 100 watts (more if 2 put in series). FREE EXPLANATORY LEAFLETS SUPPLIED WITH EACH TWEETER.

TYPE 'A' (KSN2036A) 3" round with protective wire mesh, ideal for bookshelf and medium sized Hi-fi speakers. Price £4.29 each.

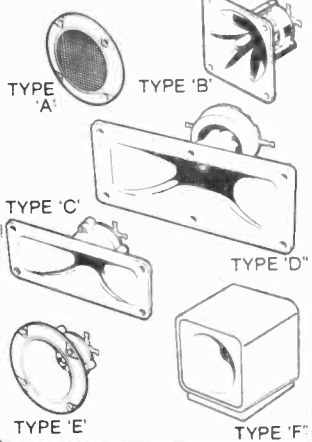
TYPE 'B' (KSN1005A) 3 1/2" super horn. For general purpose speakers, disco and P.A. systems etc. Price £4.99 each.

TYPE 'C' (KSN6016A) 2" x 5" wide dispersion horn. For quality Hi-fi systems and quality discos etc. Price £5.99 each.

TYPE 'D' (KSN1025A) 2" x 6" wide dispersion horn. Upper frequency response retained extending down to mid range (2KHz). Suitable for high quality Hi-fi systems and quality discos. Price £7.99 each.

TYPE 'E' (KSN1038A) 3 3/4" horn tweeter with attractive silver finish trim. Suitable for Hi-fi monitor systems etc. Price £4.99 each.

TYPE 'F' (KSN1057A) Cased version of type 'E'. Free standing satellite tweeter. Perfect add on tweeter for conventional loudspeaker systems. Price £10.75 each  
P&P 20p ea. (or SAE for Piezo leaflets).



### HOME PROTECTION SYSTEM

Better to be 'Alarmed' than terrified.  
Thandar's famous 'Minder' Burglar Alarm System. Superior microwave principle. Supplied as three units, complete with interconnection cable FULLY GUARANTEED.

Control Unit — Houses microwave radar unit, range up to 15 metres adjustable by sensitivity control. Three position, key operated fascia switch — off — test — armed 30 second exit and entry delay.

Indoor alarm — Electronic swept freq. siren. 104dB output.

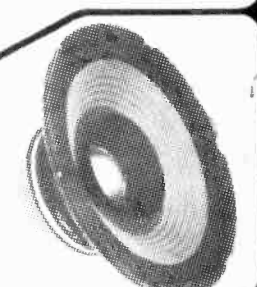
Outdoor Alarm — Electronic swept freq. siren 98dB output. Housed in a tamper-proof heavy duty metal case.

Both the control unit and outdoor alarm contain rechargeable batteries which provide full protection during mains failure. Power requirement 200/260 Volt AC 50/60Hz. Expandable with door sensors, panic buttons etc. Complete with instructions.

SAVE OVER £100 Usual price £228.85  
— B.K.E.'s BARGAIN PRICE £128.00 + £5.00 P&P  
S.A.E. for colour brochure.



12" 80 watt R.M.S. loudspeaker.  
A superb general purpose twin cone loudspeaker. 50 oz. magnet. 2" aluminium voice coil. Rolled surround. Resonant frequency 25Hz. Frequency response to 13KHz. Sensitivity 95dB Impedance 8ohm. Attractive blue cone with aluminium centre dome.  
Price £18.49 each + £3.00 P&P



## B.K. ELECTRONICS

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★ SAE for current lists. ★ Official orders welcome. ★ All prices include VAT. ★ Mail order only ★ All items packed (where applicable) in special energy absorbing PU foam. Callers welcome by prior appointment, please phone 0702-527572

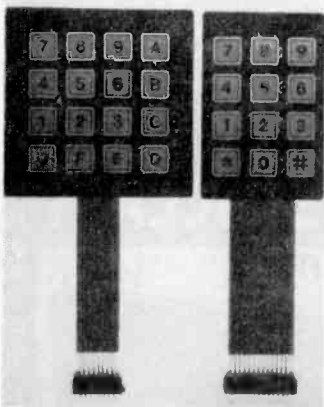


## ZX81 Music Board

We've just discovered (since the prototype was tested using a Memotech RAMpack) that it isn't possible to plug peripherals into the back of Sinclair's RAMpack. Have no fear, you can still use the Music Board: all the completed boards and kits sold by Petron Electronics will be modified so that the Sinclair memory can be plugged in behind the Music Board. Problem solved.

## ORIC Overload

Tangerine Users Group (TUG) have asked us to bring to the attention of ORIC-1 owners that they are receiving several thousand enquiries per month, thereby creating an overload condition in the organisation. Although they are making every effort to respond as quickly as possible, delays of up to 14 days are inevitable, during this period of reorganisation.



## Membrane Keypads From Velleman

Velleman have introduced a new range of membrane keypads, available with 12 keys (type KB12) or 16 keys (type KB16). Both versions are offered with standard legend or with blank keys to enable customer to print their own legend.

These multi-layer keyboards are manufactured by Velleman using high quality materials with the top layer being polycarbonate film which resists scratching, dust and water. Termination is by insulated flat cable and a suitable PCB connector with 2.54mm (0.1 inch) spacing is supplied. Ratings are 24 V and 25 mA maximum.

A data sheet with full technical specification is available upon request. Price including VAT and postage is £8.44 for both versions (1 off quantity) with discounts available for larger quantities. Velleman (UK) Limited, PO Box 30, St Leonards on Sea, East Sussex, TN37 7NL.



## High-Tech Pager

A new version of an economical radio paging system for on-site or wide-area use has been launched by Multitone Electronics, using technology currently unique to the company.

The system, called 'Readout', is a single or dual-format encoder (control unit) that can operate both analogue and digital pagers. This means that users with analogue equipment can protect past investment in their existing receivers and still upgrade their system to take advantage of the latest high-speed digital technology.

Up to 500 digital pagers can be added to an existing system without any need to replace the transmitter, aerial and associated wiring.

One of the key new features on FM systems is 'receiver out of range'. This prevents pager wearers from being out of contact without being aware of it. When taken out of range of the transmitter, a pager emits a low-frequency buzz warning.

Another new feature is transmitted steering, developed to enable users to add a new digital system to an analogue system on another radio frequency. The dual-format 'Readout' can steer calls for the analogue receivers to the existing AM transmitter and calls for the new digital receivers to a new FM transmitter. Multitone is the only manufacturer that can supply an encoder to provide this facility.

For further information contact: Multitone Communications Systems Ltd, 6-28 Underwood Street, London N1 7JT. Tel: 01-253 7611.

## New Crimson Factory

Due to successful and increasing trading Crimson Elektrik has opened a new factory to handle sales and production of all modules, kits and industrial amplifiers.

From the same date the Leicester factory will only be handling production and sales of the prestigious 600 series hifi amplifiers, available only through specialist retailers. All enquiries other than for the 600 series should now be addressed to: Crimson Elektrik, 500 King Street, Longton, Stoke-on-Trent ST3 1EZ.



THOSE JOKERS AT THE JOB CENTRE SAID I'D BE FIELD-TESTING COMPUTERS...

# BI-PAK

## "IRRESISTABLE RESISTOR BARGAINS"

Pak No.	Qty*	Description	Price
SX10	400	Mixed All type Resistors	£1
SX11	400	Pre-formed 1/4 watt Carbon Resistors	£1
SX12	200	1/4 watt Carbon Resistors	£1
SX13	200	1/4 watt Carbon Resistors	£1
SX14	150	1/4 watt Resistors 22 ohm 2m2 Mixed	£1
SX15	100	1 and 2 watt Resistors 22 ohm 2m2 Mixed	£1

Paks SX12-15 contain a range of Carbon Film Resistors of assorted values from 22 ohms to 2.2 meg. Save pounds on these resistor paks and have a full range to cover your projects  
\*Quantities approximate, count by weight

## "CAPABLE CAPACITOR PAKS"

Pak No.	Qty*	Description	Price
SX16	250	Capacitors Mixed Types	£1
SX17	200	Ceramic Capacitors Miniature Mixed	£1
SX18	100	Mixed Ceramics 1pf-56pf	£1
SX19	100	Mixed Ceramics 68pf-0.5m	£1
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\*Quantities approximate, count by weight



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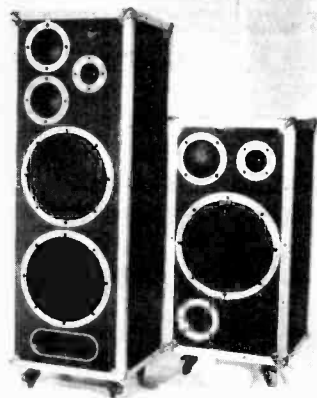
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## Yet More Speakers From Wilmslow

New from Wilmslow Audio, PROKITS provide a range of speakers for the home constructor which are ideal for small venue discos, public address, parties, etc. Combining Wharfedale's design expertise with Wilmslow Audio's kit know-how, they offer true hi-fi quality sound together with high sensitivity to make optimum use of the amplifier power available. The kits contain everything required for the construction of the speakers except adhesives. Only a

few simple tools are required. Six models are offered: E50PRO (100 watts), E70PRO (150 watts), E90PRO (200 watts), E50PRO SUPER (150 watts), E70PRO SUPER (175 watts), E90PRO SUPER (300 watts). The PRO SUPER range is fitted with special hand-built bass units and the crossover networks are fitted with thermal overload protection for the treble units and for the complete system. The photograph shows the E50PRO and the E90PRO.

Prices range from £218.95 per pair for the E50PRO to £425 per pair for the E90PRO SUPER. Details from Wilmslow Audio Ltd, 35/39 Church Street, Wilmslow, Cheshire SK9 1AS. Telephone 0625 529599.

## A New Name

Kenema Associates is a newly formed O.E.M. company now launching a comprehensive range of products for the Oric-1 computer. Included in this range is a 'Multipurpose/Personnel Records File'. This powerful software permits the mass storage and file handling of confidential information on personnel or products. The package is priced at £15.00.

Also supporting the range is an 'Oric-1 Keyboard Trainer' software package and a growing number of games, books, and accessories for the Oric-1 owner including Oric-1 American T-Shirts. Kenema Associates Ltd, 1 Marlborough Drive, Worle, Avon BS22 0DQ.

## Large Format Clock/Counter

A new range of digital counters and clocks using high brightness 3½ inch vacuum fluorescent character display tubes is now available from Greatch Electronics Limited. The digital counters, two, three and four digits accept BCD input with optional blanking and remote display data input control.

The clock model is available in a twelve or twenty-four hour ver-

sion with a mains frequency locked crystal controlled oscillator powered by a battery (rechargeable) during mains failure. (One of these may even be enough to help the Editor of a certain sister electronics magazine get to work on time!)

The display tube's brightness is adjustable and available in white, green, red and yellow. A variety of other colours can be provided by filtering and contrast enhancement provided to suit particular applications. Greatch Electronics Ltd, Hay Lane, Braintree, Essex.



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### DISPLAYS

DL304/307 - 75p

DL704 - 90p

DL500/507 - 85p

DL527/528 - £1.50

DL0727 (Orange) - £2.00

DL0747 (Orange) - £1.00

DL6304/307 (Green) - 80p

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4 Digit 0.5" Clock - £2.00

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DL1414 Intelligent - £12.50

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5 each, 4 colours, 3,5mm, Rectangular Gravestone Led

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### HOME LIGHTING KITS

These kits contain all necessary components and full instructions & are designed to replace a standard wall switch and control up to 300w. of lighting.

- TDR300K Remote Control Dimmer **£14.30**
- MK6 Transmitter for above **£ 4.20**
- TD300K Touchdimmer **£ 7.00**
- TDE/K Extension kit for 2-way switching for TD300K **£ 2.00**
- LD300K Rotary Controlled Dimmer **£ 3.50**



### HOME CONTROL CENTRE

This kit enables you to control up to 16 different appliances anywhere in the house from the comfort of your armchair. The transmitter... appliance addressed. The transmitter also includes a COMPUTER interface so you can programme your favourite micro (e.g. ZX81) to switch lights, heating, electric blanket, make your morning coffee, etc., automatically without rewiring your house. **JUST THINK OF THE POSSIBILITIES.** The kit includes all PCBs and components for one transmitter and two receivers, plus a pre-drilled box for the transmitter. Order as XK112. **£42.00**  
Additional Receivers XK111 **£10.00**

### "OPEN-SESAME"

The XK103 is a general purpose infra-red transmitter/receiver with one momentary (normally open) relay contact and two latched transistor output. Designed primarily for controlling motorised garage doors and two auxiliary outputs for drive/garage lights at a range of up to 40 ft. The unit also has numerous applications in the home for switching lights, TV, closing curtains, etc. Ideal for aged or disabled persons.  
The kit comprises a mains powered receiver, a four button transmitter, complete with pre-drilled box, requiring a 9V battery and one opto-isolated solid state switch kit for interfacing the receiver to mains appliances. As with all our kits, full instructions are supplied.

**ONLY £23.75**

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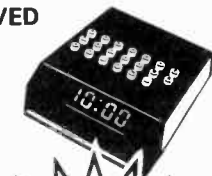
Based on the SAB0600 IC the kit is supplied with all components, including loudspeaker, printed circuit board, a pre-drilled box (95 x 71 x 35mm) and full instructions. Requires only a PP3 9V battery and push-switch to complete.  
**AN IDEAL PROJECT FOR BEGINNERS** Order as XK102 **£5.00**

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Now you can run your central heating, lighting, hi-fi system and lots more with just one programmable timer. At your selection it is designed to control four mains outputs independently, switching on and off at pre-set times over a 7 day cycle, e.g. to control your central heating (including different switching times for weekends), just connect it to your system programme and set it and forget it—the clock will do the rest.

- FEATURES INCLUDE:-**
- 0.5" LED 12 hour display.
  - Day of week, am/pm and output status indicators.
  - 4 zero voltage switched mains outputs.
  - 50/60Hz mains operation.
  - Battery backup saves stored programmes and continues time keeping during power failures. (Battery not supplied).
  - Display blanking during power failure to conserve battery power.
  - 16 programme time sets.
  - Powerful "Everyday" function enabling output to switch every day but use only one time set.
  - Useful "sleep" function—turns on output for one hour.
  - Direct switch control enabling output to be turned on immediately or after a specified time interval.
  - 20 function keypad for programme entry.
  - Programme verification at the touch of a button.

(Kit includes all components, PCB, assembly and programming instructions). ORDER AS CT5000



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Supplied with hand held plastic box. Requires 9V (PP3) battery. **£4.20**
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Mains powered with triac output to switch up to 500W at 240V ac. Range approx 20 ft. on/off or momentary control. **£9.00**  
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For use with MK18/MK12 transmitter receiver where only 4 channels are required. **£1.90**
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A mains powered LR receiver providing control signals to 10 on/off and 3 analogue circuits. May be used for controlling the volume of an amplifier, brightness of a lamp, etc. **£12.00**
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A mains powered LR Receiver providing up to 16 outputs for switching.

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This new design is based on the ICL7126 (a lower power version of the ICL7106 chip) and a 3 1/2 digit liquid crystal display. This kit will form the basis of a digital multimeter (only a few additional resistors and switches are required—details supplied), or a sensitive digital thermometer (-50°C to +150°C) reading to 0.1°C. The basic kit has a sensitivity of 200mV for a full scale reading, automatic polarity indication and an ultra low power requirement—giving a 2 year typical battery life from a standard 9V PP3 when used 8 hours a day, 7 days a week.

**Price £15.50**

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This value for-money kit features a bi-directional sequence speed of sequencing and frequency of direction change, being variable by means of potentiometers and incorporates a master dimming control. **£14.60**

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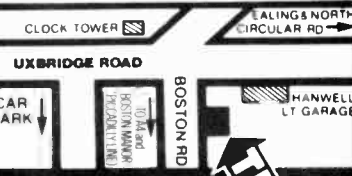
**DL3000K**  
This 3 channel sound to light kit features zero voltage switching, automatic level control and built in mic. No connections to speaker or amp required. No knobs to adjust simply connect to mains supply and lamps. (1Kw channel) **Only £11.95**

### MINI KITS

- MK1 ELECTRONIC THERMOSTAT**  
Uses LM331 IC to sense temperature (80°C max) and triac to switch heater (1KW). Mains powered. **£2.50**
- MK2 SOLID STATE RELAY**  
Switches 240V ac motors, lights, heaters from logic/computer circuits. Zero voltage switching, opto-isolated. Supplied without triac. **£2.50**
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- MK5 MAINS TIMER**  
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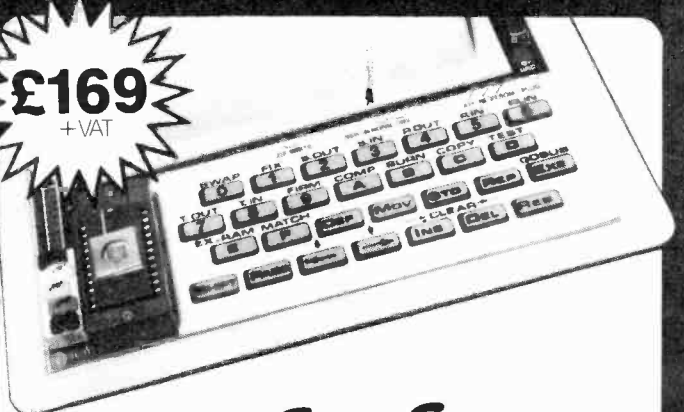
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# Dataman Designs

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# AUTOCOMPASS

Don't say we aren't good to you! Not only is this a unique and cheap-to-build project featuring some unusual techniques, but it also gives an insight into a little-known area of binary numbers. Design by Nigel Collier.

This novel project describes the operation and construction of an electronic compass specifically designed for use in automobiles. The unit can distinguish 16 different headings, and uses a dot matrix LED display to register the car's heading in nautical point and quarterpoint notation — for landlubbers this means N, NNE, NE, ENE and so on. The display dims automatically as the ambient light level drops.

The whole assembly, which is simply compensated for accurate readings inside the metal body of the car, is housed in an elegant matt black box, complete with a fixing bracket which enables it to be positioned almost anywhere in the car.

Finally, the sensor itself is designed using an apparently newly-discovered set of binary numbers!

## The Circuit

The unusual sensor, details of which are given later, generates a four-bit code, with each of the 16 possible numbers corresponding to a different heading. The remainder

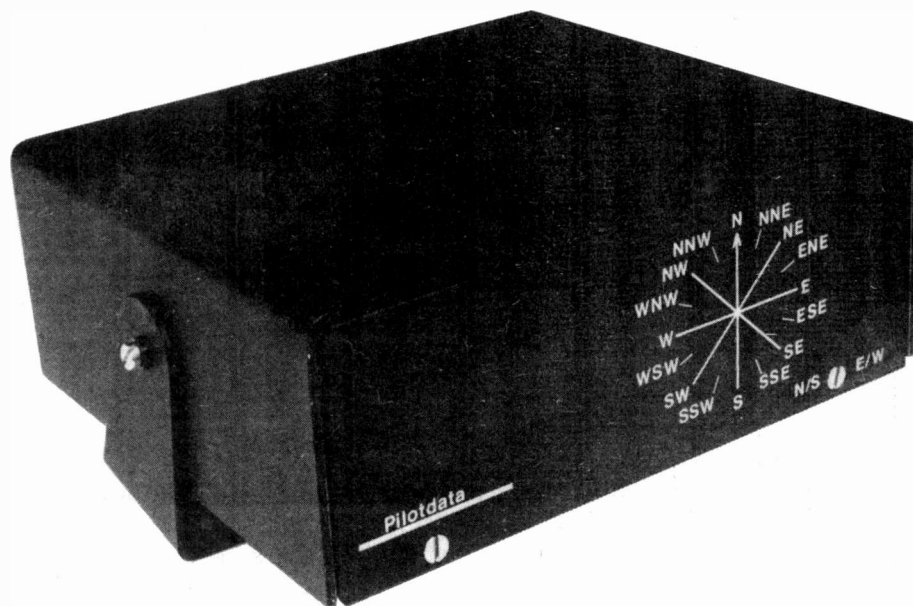
of the logic circuitry is based around a 2716 2K EPROM (at last, a non-computer-orientated use for an EPROM).

The EPROM contains the data necessary to give meaningful alphabetical displays as the perceived heading is updated. A listing of the data contained in the EPROM can be obtained from us on receipt of an s.a.e., for those amongst you with access to an EPROM programmer. Don't worry if you haven't, though; ready-blown EPROMs will be available (see Buylines).

The display consists of two 7 x 5 dot matrix displays type DL5735, manufactured by Litronix. These are arranged to form three adjacent display sections with areas of 5 x 5, 4 x 5 and 5 x 5.

## Sensor Design

The operation of the sensor can best be described by looking at Fig. 1. The central pillar, topped by the pivot needle, is made from transparent perspex and houses the incandescent light source. Balanced on the pivot needle is the compass



## PARTS LIST

Resistors (all 1/4W, 5% except where stated)

R1	1k0
R2-6,9	4k7
R7	10k
R8	not used
R10,34,35	470R
R11-15,	
21-25,33	2k7
R16-29,26-30	30R, 1/2W
R31,32	15R, 1W

Potentiometers

PR1,2	100R miniature horizontal preset, 1/4W rating
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Capacitors

C1	220nF polyester
C2	470uF 6V3 axial electrolytic
C3-6	47nF polyester
C7	2n0 polystyrene
C8	2u2 63 V axial electrolytic
C9	1n0 polystyrene

Semiconductors

IC1	2716 single rail EPROM (+5 V)
IC2	4017B
IC3,4	4001B
IC5	7472
IC6	7475
IC7,8	74LS367
IC9	7805
IC10	ULN2003
Q1-10	BC212L
Q11	BC182L
Q12-16	OP500 phototransistor
DISP1,2	DL5735 dot matrix LED display

Miscellaneous

LDR1	ORP12
L1-4	500 turns of 40 swg enamelled copper wire on 1M0 1W resistor (see text)
PCBs (see Buylines); sensor (see Buylines); case (see Buylines); mounting hardware, wire etc.	

## BUYLINES

A ready-built sensor, complete with phototransistors, is available for £8. The two double-sided PCBs, both with silk-screened component identification, are available for £6.50 the pair. Customer-supplied EPROMs will be programmed for 40p plus postage. Preprogrammed 2716 EPROMs are available for £3.00. An Autocompass case plus silk-screened perspex front panel is available for £4.00. Finally, a complete kit of parts, down to the last nut and bolt, will be supplied for £35.00 plus £1.50 postage and packing. All the above are available from Pilotdata Ltd, 2 Derwent Close, Wokingham, Berks.

mask — this is free to both rotate and tilt. A toroidal magnet, magnetised through its major diameter, is fixed to the mask. Thus the mask will always orientate itself in the same direction (with respect to the Earth's magnetic field) as the car's heading changes.

The mask pattern is designed such that the four phototransistors arranged around the circumference of the mask can unambiguously

encode the sensor heading. For those who are interested in this sort of thing, a full description of the coding method used is given in the box entitled 'Cyclic Binaries'.

This form of coding, rather than a standard four-bit coding for four phototransistors positioned in a vertical line, was necessary to stop

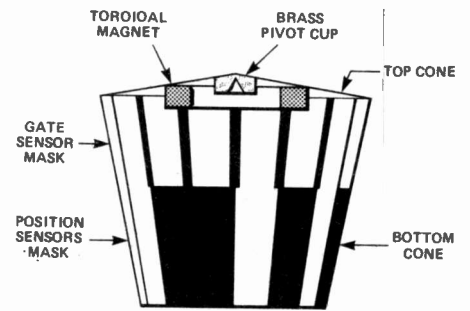
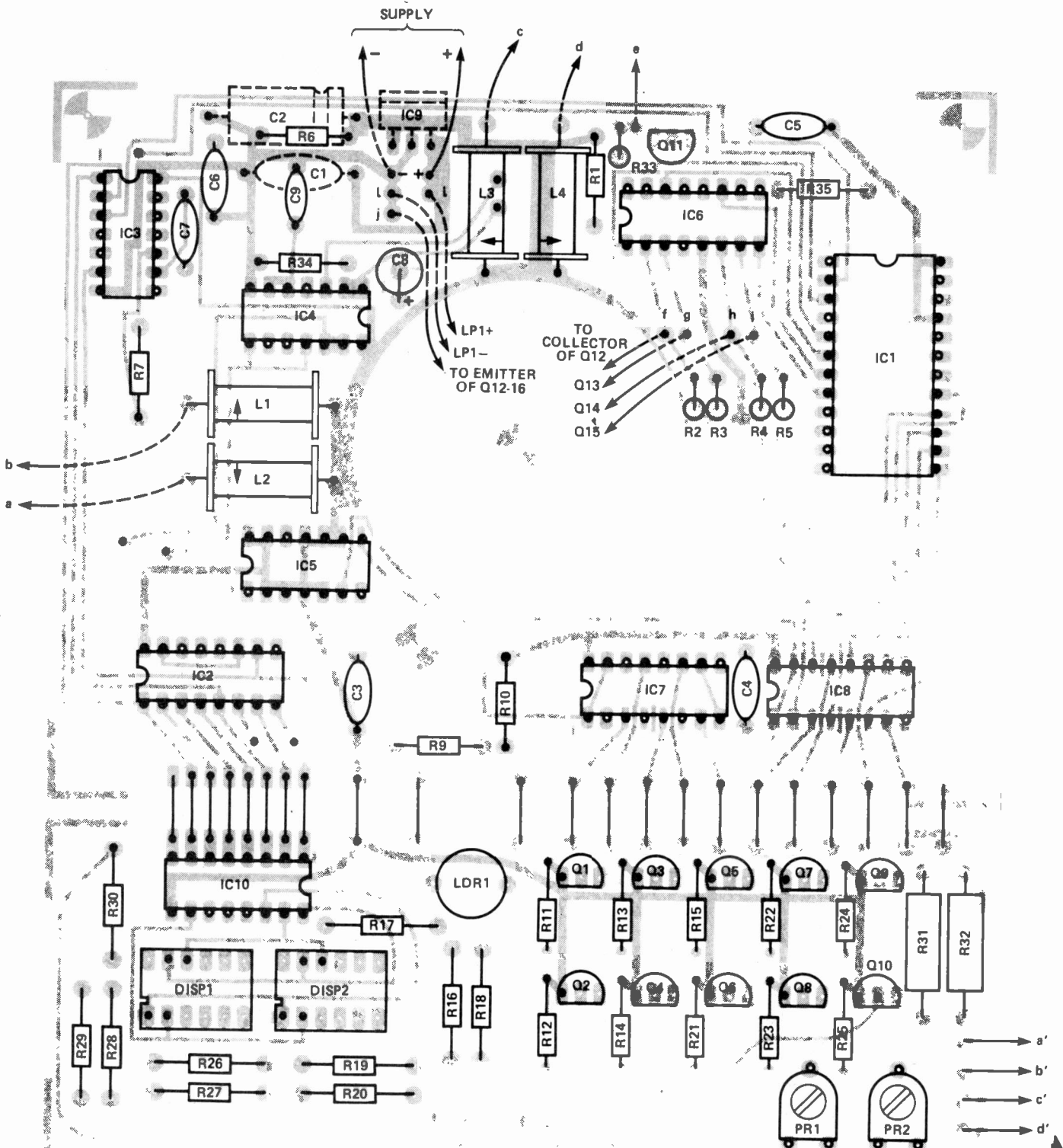


Fig. 1 (Right) The sensor mask.  
Fig. 2 (Below) Component overlay.



erroneous coding occurring as the mask tilts (as it is inclined to do as the car accelerates or corners). The fifth phototransistor, Q16, reads another part of the mask, and only allows the logic circuitry to update the sensor heading when the four 'position' phototransistors are nicely matched with a mask section.

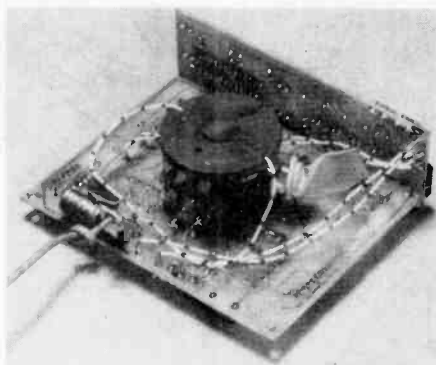
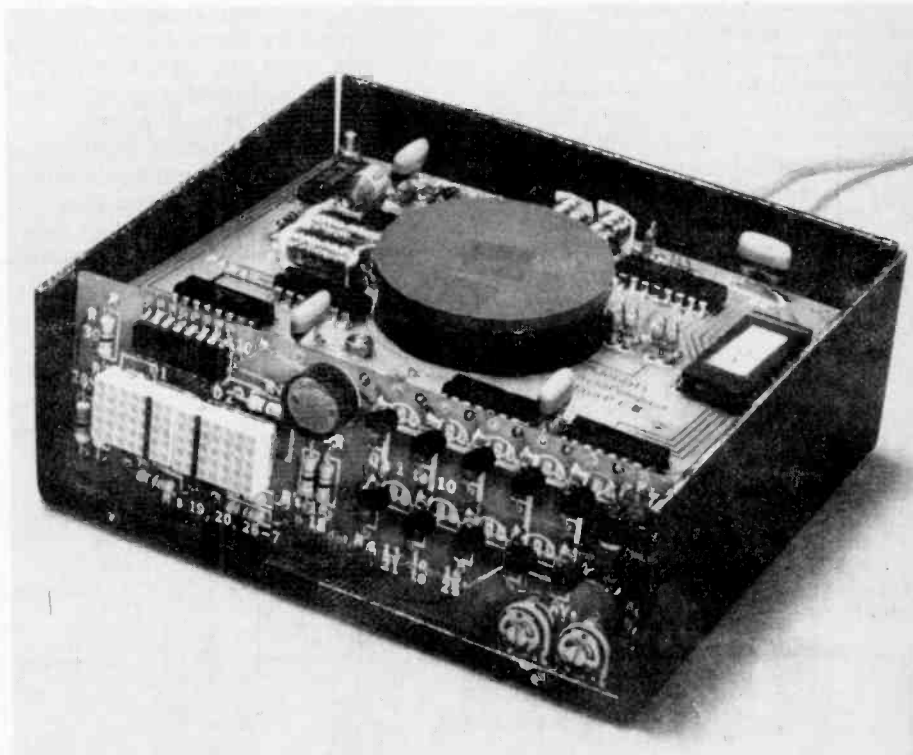
The whole assembly is fluid-damped to stop unnecessary oscillations. A somewhat incongruous loop of silicone tube is fixed onto the side of the housing. This is not a way of getting rid of spare balloons from the Christmas party — it is present to absorb any changes of volume of the damping fluid that occur with changes in temperature.

### Construction

Because of the difficulties involved in the making of the compass mask, it is not recommended that the average home constructor should attempt to build his (or her) own sensor from scratch. Ready-built and tested sensors are available from Pilotdata at the address given in Buylines. We doubt if many of our readers are keen on metal-bashing either, given that Pilotdata can supply the case assemblies ready-screened and painted, but dimensions are available from us if you send a stamped addressed envelope (A4-size).

From an electronic point of view there are few problems posed by the construction — however, you will need a soldering iron with a very fine bit.

The main PCB is double-sided but not plated-through. There are seven points where isolated tracks on opposite sides of the board need to be joined with a linking pin — these are shown with a dot on our overlay and a 'p' on the silk-screened legend on the manufactured boards. Any other component lead or IC pin indicated with a dot should also be soldered both sides. The 10 terminal pins for the off-board connections must be all pushed through from the component side of the PCB. The voltage regulator and its two associated capacitors, C1 and C2, are shown in dotted outline on the overlay and must be soldered onto the underside of the PCB. The leads of the voltage regulator should be carefully pre-bent such that the mounting hole of the regulator is 19 mm below the underside surface of the PCB and flush with the edge of the board when soldered in



These two internal shots should help with construction.

position. Using an epoxy adhesive, fix a 5/16" 6BA screw into the regulator mounting hole with the screw thread protruding out backwards from the body of the PCB.

The four coils are wound on 1 megohm, 1 W resistors. Each resistor is prepared by glueing onto each end a 9 mm square plastic or cardboard pad to retain the wire.

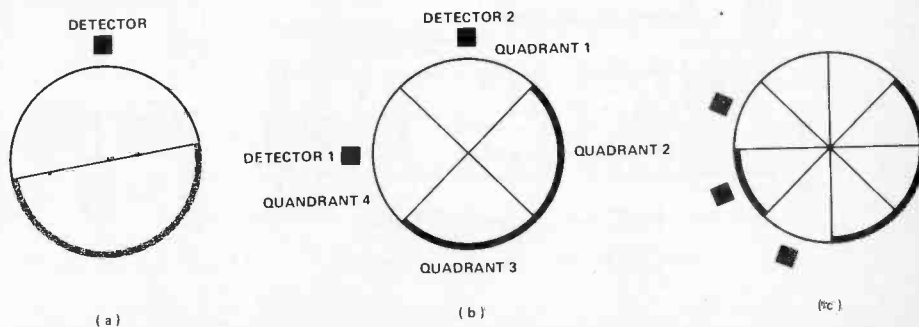


Fig. 3 Using cyclic binaries to detect shaft position.

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ZX81 PROGRAM TO GENERATE
16 BIT CYCLIC BINARIES

10 PRINT "ENTER NO. ";
20 INPUT N$
30 PRINT N$
40 LET P = 1
50 DIM A$(16,4)
60 LET A$(P) = N$(2)
70 LET N$(1) = N$(3)
80 LET N$(2) = N$(4)
90 LET N$(3) = N$(4)
100 LET N$(4) = "0"
110 FOR C = 1 TO P
120 IF N$(C) = A$(C) THEN GOTO 210
130 NEXT C
140 LET P = P + 1
150 LET A$(P) = N$
160 IF P < 16 THEN GOTO 70
170 FOR N = 1 TO P
180 PRINT N, A$(N)
190 NEXT N
200 STOP
210 LET N$(4) = "1"
220 FOR C = 1 TO P
230 IF N$(C) = A$(C) THEN GOTO 270
240 NEXT C
250 GOTO 140
260 LET P = P - 1
270 LET N$(C) = A$(P)
280 LET N$(4) = "1"
290 FOR C = 1 TO P
300 IF N$(C) = A$(C) THEN GOTO 260
310 NEXT C
320 GOTO 150

CHANGES TO GENERATE A 32 BIT CYCLIC BINARY

50 DIM A$(32,5)
95 LET N$(4) = N$(5)
100 LET N$(5) = "0"
160 IF P < 32 THEN GOTO 70
210 LET N$(5) = "1"
280 LET N$(5) = "1"

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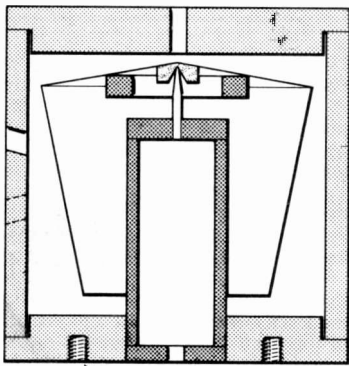


Fig. 4 (Above) A cross-section through the sensor housing.

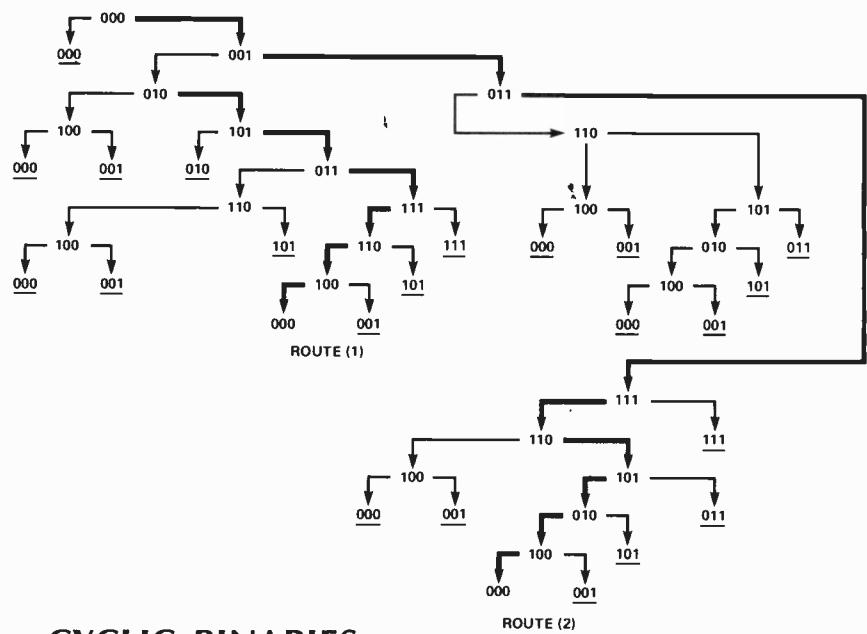


Fig. 5 (Right) Generating the cyclic binary for Fig. 3c.

## CYCLIC BINARIES

The mask pattern of the compass sensor owes its existence to a group of numbers which the author has called 'cyclic binaries'. So far as we are aware, this intriguing set has so far been ignored by the mathematicians, though if anyone knows otherwise, we would be delighted to hear from them.

Consider a cross-section of a shaft (Fig. 3a) whose circumference is half white and half black. A single detector is sufficient to determine which side of the shaft is nearest the detector, by detecting the local shaft colour.

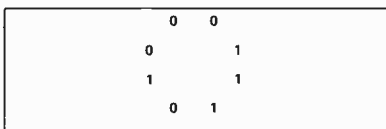
Using the same shaft, let us now introduce a second similar detector such that the angle subtended by the two detectors at the axis of the shaft is 90° (Fig. 3b). The position of a point on the shaft can now be narrowed down to one of four quadrants. This table gives the truth table for the four possibilities:

Q1	Q2	Quadrant
Black	Black	1
Black	White	4
White	Black	2
White	White	3

In these examples the detectors are obviously binary detectors: we can call black a 0 and white a 1. Thus in Fig. 3b the shaft's rotational position has been encoded as a two-bit number.

Can we extend this system to three numbers, so as to encode for eight sectors? We can, but the shaft's pattern now becomes a bit more complex (Fig. 3c). As in the case for one or two detectors, the pattern is unique (ignoring simple rotations and mirror images) in generating a separate three-bit number for each of the eight sectors. Note that the detectors now subtend angles of 45° to their neighbours.

Now the colouring of the three-sensor shaft pattern can be represented by an eight-bit binary number arranged in a circle thus:



I propose to write this number thus:

11100010<sub>C</sub>

The subscript C signifies that the binary is cyclic, ie its end is joined to its start. (Note that it doesn't matter at what point in the number you start, or in what direction you proceed: 10001110<sub>C</sub> is the same number).

Now for four detectors with 22.5° separation: the arrangement used in the Autocompass. The shaft pattern is now divided into 16 sectors: are there any patterns which yield a unique four-bit number for each of the discernable rotational positions? The answer is obviously yes, since one is used in the compass sensor. In fact there are six such numbers (not counting rotations and reflections), and these six can be split up into three pairs. Each pair consists of a number and its inverse (ie each 1 becomes a 0 and vice versa). The pairs are:

0110101111000010<sub>C</sub> and inverse  
1000011010111100<sub>C</sub> and inverse  
1000010111101001<sub>C</sub> and inverse

As far as we know, perfect cyclic binaries (ie ones that generate unique positional data for each of their sectors) can be found for any number of sensors. For the sake of interest, the table below shows perfect cyclic binaries for five, six and seven sensors.

FIVE BIT:

00000111110111001101011000101001<sub>C</sub>

SIX BIT:

00000011111101110011101011100011011010011001  
0110000101010001001<sub>C</sub>

SEVEN BIT:

000000011111101111001111010111000111011011  
101001110010111000011011001101010110100011001  
00110001011000001010100101000010010001<sub>C</sub>

How can perfect cyclic binaries be generated? Consider the case of determining the cyclic binary for three sensors. Figure 5 shows the search tree involved: we have arbitrarily started from '000'. If the shaft rotates one position from right to left, the new number will be '00?'. The least significant digit can be 0 or 1, but if it is 0, we have a repeat of the first number, so it must be a 1. The tree shows the continuation of this pro-

cess. Every underlined number is a repeat.

The completed tree shows two routes through all eight numbers and back to the starting number. Selecting the first digit of each number along the routes, we get:

00010111<sub>C</sub> (Route 1)  
00011101<sub>C</sub> (Route 2)

ie both routes yield the same number.

For larger cyclic binaries, drawing the tree out becomes impractical, but the search method is ideally suited to adaptation to a computer program. We've given a program here, in BASIC, which has been written for a ZX81 and will perform a tree search to generate 16-bit cyclic binaries — all you have to do to start it is enter a four-bit number.

The most obvious drawback of using cyclic binaries for determining the rotational position of a shaft is that an additional gating phototransistor is necessary. This would be superfluous if the cyclic binary was also Gray-coded. Gray-coded binary number sequences are special in that any two adjacent numbers in the sequence only differ by one binary digit. The following table shows a representative Gray coding sequence to illustrate the property:

000  
 001  
 011  
 010  
 110  
 111  
 101  
 100  
 000 etc.

Do 'super-perfect' Gray-coded cyclic binaries exist? We think not, but this judgement is based on a very limited investigation.

If you wish to search for superperfect cyclic binaries, here are a couple of hints. First, note that the detectors do not need to be adjacent (though any rotationally symmetrical arrangement of detectors will lead nowhere). Second, looking for cyclic Gray-coded sequences will be easier than looking for Gray-coded cyclic binaries (if you get the distinction!). Good luck.

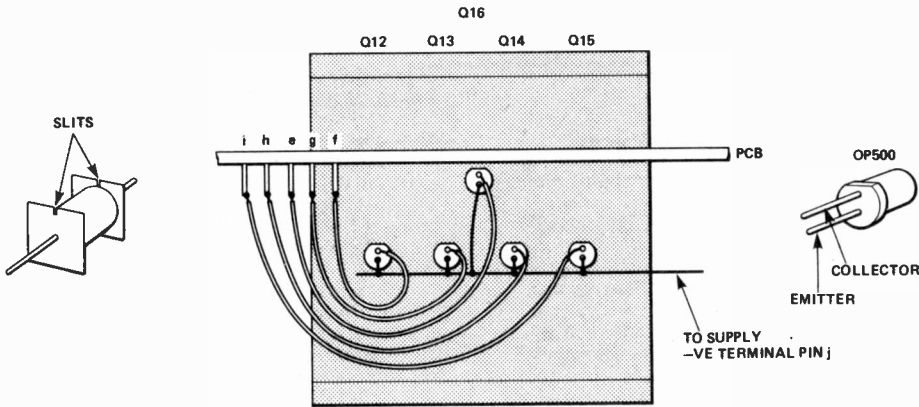


Fig. 6 (Above) Coil and sensor details.

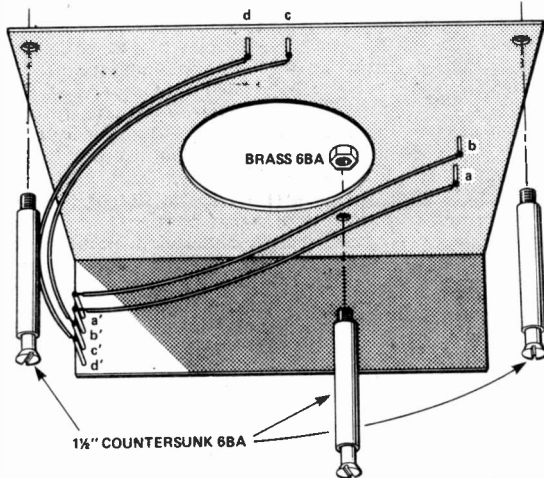


Fig. 7 (Left) The PCB fixing bolts fit here.

Q12-16 are phototransistors attached to the compass sensor. The voltage present at the collector of any one of the phototransistors is pulled up to a logic 1 level by the resistors R2-5 and R33, so long as the phototransistor in question is shaded from the sensor light source by the compass mask. Once exposed to light, a phototransistor will turn on and its collector voltage will drop to a logic 0.

IC6 is a four-bit bistable latch. So long as phototransistor Q16 is shaded, the latch outputs will present the inverse of the latch inputs. However, once Q16 is exposed to light, the latch will hold the last four-bit input, irrespective of any further changes in input. In other words, IC6 is a memory device, supplying IC1 with information concerning the sensor position at the last positional update. Table 1 shows the output of IC6 with reference to the sensor positions.

The rest of the logic circuit converts the four-bit output of IC6 to an alphanumeric display of the sensor heading. IC3c,d are configured with R7 and C7 as an astable multivibrator. With the values chosen, this generates a

Fig. 8 (Below left) Circuit operation shown frozen at one point in the multiplexing.

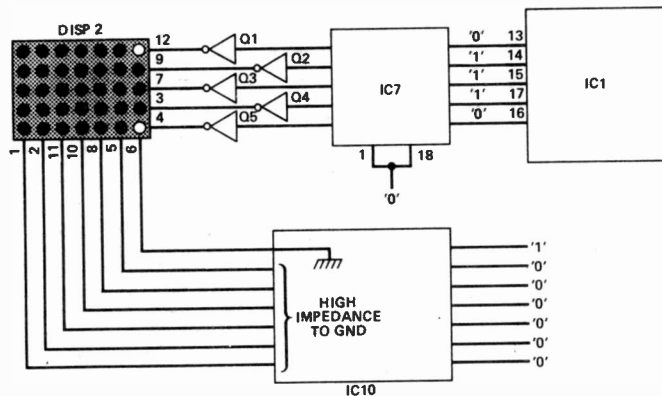


TABLE 1

IC6 OUTPUT VERSUS SENSOR HEADING					
SENSOR HEADING	A3 (PIN 1)	A2 (PIN 14)	A1 (PIN 11)	A0 (PIN 8)	DECIMAL
N	1	1	1	0	14
NNE	1	1	1	1	15
NE	0	1	1	1	7
ENE	1	0	1	1	11
E	1	1	0	1	13
ESE	0	1	1	0	6
SE	0	0	1	1	3
SSE	1	0	0	1	9
S	0	1	0	0	4
SSW	1	0	1	0	10
SW	0	1	0	1	5
WSW	0	0	1	0	2
W	0	0	0	1	1
WNW	0	0	0	0	0
NW	1	0	0	0	8
NNW	1	1	0	0	12

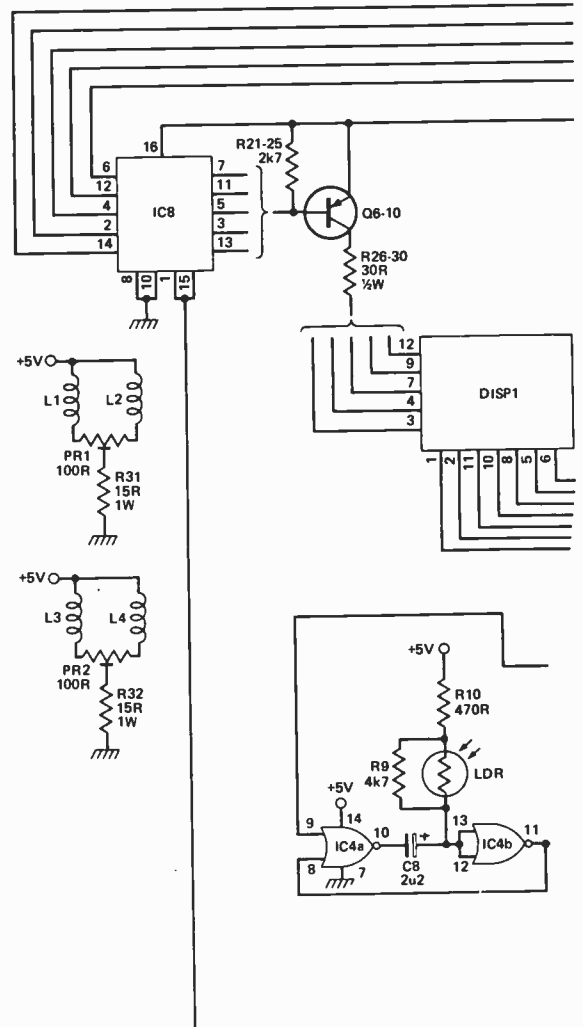


Fig. 9 Autocompass circuit diagram.



# PROJECT : Autocompass

## HOW IT WORKS

square wave of about 25 kHz, which is fed to the clock input of IC2.

IC2 is a 4017 counter-divider: only one of the chip's 10 outputs can be high at any one time. Every positive transition of the clock signal advances the high output to the next pin in sequence, from right to left in the circuit diagram. Once pin 6 is driven high it activates the monostable multivibrator configured around IC4a,b. The monostable is, in effect, a pulse stretcher. The output of the monostable (pin 11) is normally low. When pin 9 of IC4a goes high, the output simultaneously goes high. As IC4b pin 11 is connected to the 4017 clock inhibit (pin 13), this immediately stops any further clocking of the 4017's outputs.

The length of time that IC4b pin 11 remains high is dependent on the resistance of the LDR. The more light that falls on the LDR, the lower is its resistance, the quicker C8 charges, and the shorter is the resultant length of the output pulse. Once pin 11 returns low again, the 4017 is free to continue clocking on the next positive clock signal transition. When IC2 pin 11 goes high, it drives the reset pin (pin 15) high which

starts the whole cycle again by sending pin 3 high. To summarise, the 4017 outputs sweep from pin 3 to pin 5; then there is a delay with all the relevant outputs of the 4017 low, this delay being dependent on the ambient light level, before the cycle repeats. It can therefore be seen that the delay constitutes a display brightness control.

IC10 is a seven-stage Darlington driver. As each input of IC10 goes high, so the corresponding output is grounded, which in turn grounds the cathodes of the five LEDs in the relevant row of each display matrix. This constitutes one-half of the display multiplexing process.

IC1 is an EPROM which produces a specific eight-bit output at pins 9-11, 13-17 (only the latter five being used) depending on the state of the 11 address input pins. We have seen that pins 5-8 of IC1 depend on the sensor heading, being driven by the outputs of IC6. The remaining seven address lines define the position of the multiplexing cycle.

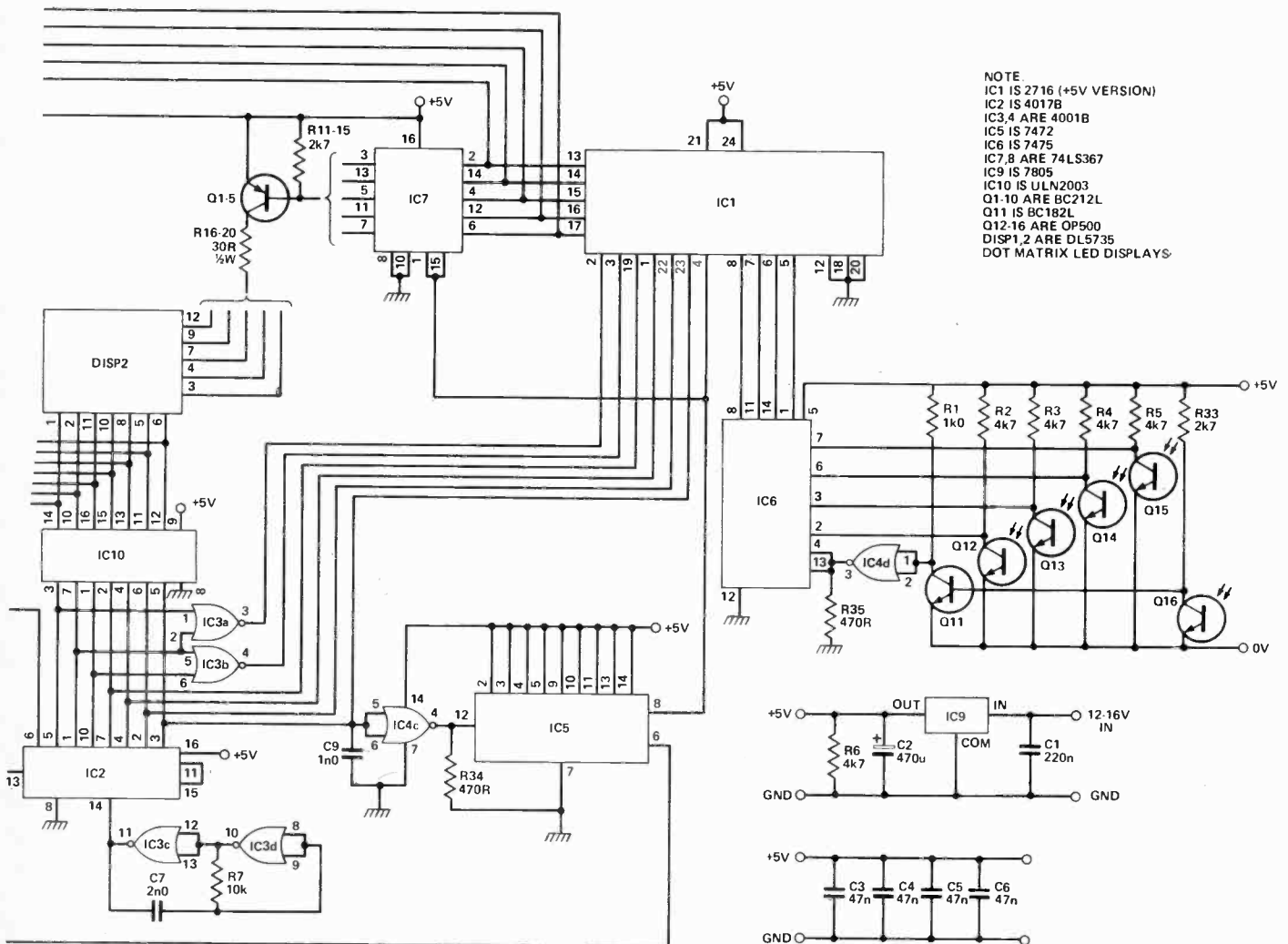
The two dot matrix displays together have 14 vertical columns: the 4017 cycles through seven states. Therefore

the multiplexing circuit must service one display at a time. This is organised by IC5, a flip-flop. Every time the 4017 starts a new cycle by sending pin 3 high, the outputs of IC5 (pins 8,6) change state. Pin 8 is always the inverse of pin 6. IC5 pin 8 addresses pin 4 of the EPROM: thus pin 4 high means 'multiplex display 1', while pin 4 low means 'multiplex display 2'.

The remaining six address lines of IC1 are driven by the outputs of IC2. The two NOR gates IC3a,3b compress the information on IC2 pins 10,1, and 5 to the two lines on pins 2 and 3 of IC1.

IC7 and IC8 are tristate buffers. They are gated on by the two outputs of IC5 such that only one is on at a time. Both receive the data output from IC1, and each in turn uses the data to supply current to the anodes of the LEDs in its display via the current amplifiers Q1-10.

To help show what is going on, Fig. 8 shows the multiplexing at one instant of time. Note that Q1-10 act as inverters, so a 0 at the output of the EPROM defines an LED which is lit. In this case the display could be showing the last column of, say, "NNE".



The wire is 40 gauge enamelled copper wire, and 500 turns are needed on each resistor — this won't take you as long as it might seem! Remember to scrape off the insulation from the free ends before attempting to solder them to the resistor leads. Each pair of coils consists of one which has been wound 'right-handed' and one which has been wound 'left-handed'; ie they are not capable of superimposition, one being the mirror-image of the other. Label the two types unambiguously.

The display has four terminal pins but does not require any linking pins: all through-board connections here are made by soldering the necessary component leads on both sides (again indicated by dots). After soldering the terminal pins, which again are inserted from the component side, solder in the LED dot matrix displays. Mount them so that their pins only protrude about 1/2 mm from the rear of the PCB: this allows sufficient clearance to solder to the tracks on the front of the PCB. Save the clipped leads from the resistors — they will be ideal later for joining the two PCBs together.

The boards are mounted so that the main PCB lies 10 mm below the top edge of the display PCB, and 1 mm free of the underside of the display PCB. The diagram shows the

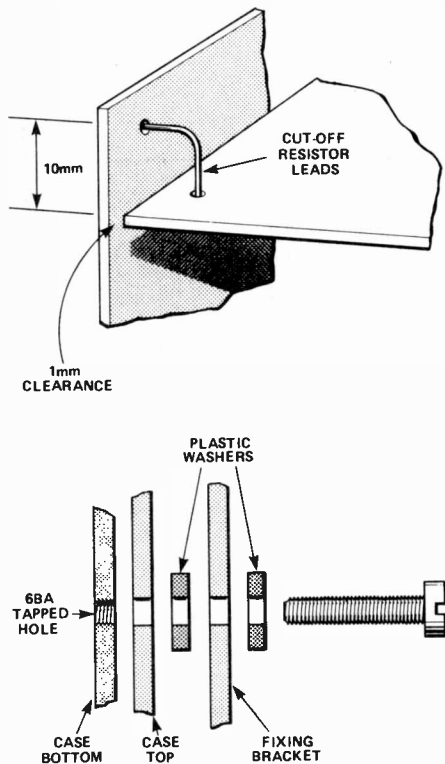


Fig. 10 It goes together like this . . .

method used to join the boards (Fig. 10).

The interwiring of the coils to the display PCB and the sensor to the main PCB is straightforward: just keep the wiring short and neat and remember that all the wiring is taken from the underside of the PCBs. Note that the phototransistors protrude from the sensor towards the back of the instrument.

The assembly of the two PCBs and the sensor fits snugly into the case bottom, being fastened in place with three 1 1/2" screws which have slide-on spacers. Two 6BA x 3/16" countersunk screws fix the sensor housing to the case bottom, and a further two attach the perspex fascia. The case top and fixing bracket are attached using two 6BA x 1/2" screws, as shown in Fig. 10. Note that the fixing bracket can either be fixed underneath the unit as shown in the photos, for dashboard mounting, or the other way up so that the compass may be hung from the car roof.

## Setting Up

When benchtesting the completed PCB assembly, remember to adequately heatsink the 5 V regulator. The display is multiplexed, and so the DC current through each LED can exceed 100 mA (though for a few microseconds only, of course). But, were the oscillator to be inoperative then it would be perfectly possible to blow up the display. So play it safe — temporarily disconnect the 5 V supply to the display PCB by desoldering the ninth board-linking pin from the right-hand side (viewed from the front) and check on an oscilloscope that the waveforms are what they should be (see Fig. 11).

If you don't have an oscilloscope, don't despair: an LED with a 200R resistor in series will suffice. Ground the appropriate end and touch the LED's anode end to the output pins of the 4017 (pins 1, 2, 3, 4, 5, 6, 7, 10). When operating correctly, the LED should glow very dimly (you may need to turn off the lights to detect it). If, on the other hand, one of the channels lights the LED brightly, with the remaining channels off, then you have a fault. However, with reasonable care in construction, the compass should work first time.

The only adjustment required is the setting of the bias coils to null out the car's magnetic field. Using a map (or a second compass outside the car), position the car so that it is pointing north. If the Autocompass

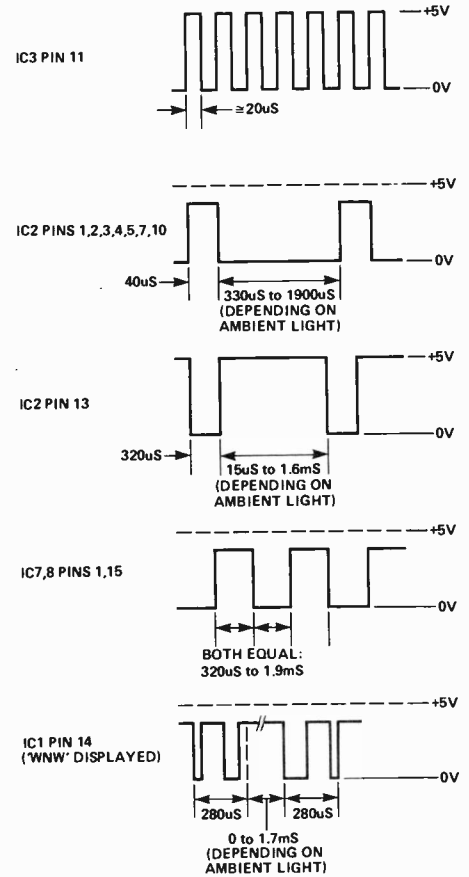


Fig. 11 Some test waveforms.

does not read north, it is because of the car's field component which lies at right angles to the car's long axis. We can nullify this component by rotating the pot labelled 'N/S' until the compass reads north.

Now the car must be positioned to point east. If the compass does not read east, it can only be because of the car's field component which lies parallel to the car's long axis. So adjust the 'E/W' pot to neutralise this component, and the compass is compensated.

## Installation

The Autocompass has been designed to be easily fixed into a car, but there are a few cautionary points that must be made. First, keep it away from any loudspeakers, as they possess strong magnets which can overcome even the compensation circuits. Second, ensure that the top of the box is as level as possible in its final position — the sensor will tolerate quite a bit of tilt, but if it is level at rest then its operational range will be enhanced. Finally, the display can consume about 800 mA with the display flat out, so an in-line fuse may be desirable.



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# LABORATORIES ON A CHIP

Silicon is turning out to have more uses than just a conventional chip material. It can be chemically machined to form a wide variety of structures, perhaps even miniature laboratories with built-in computers. Stephen McClelland explains.

**W**hat could be a new era of Lilliputian engineering is quietly unfolding thanks to a different kind of silicon chip development. Now, microscopic mechanical structures, some less than the thickness of a human hair, can be fabricated in silicon just like standard transistors and integrated circuits.

Silicon has been used for some time to create pressure and strain gauges but present techniques can produce nozzles, valves and sensors of all types. The manufacture of complex 3D mechanics in silicon is now being contemplated. Researchers at Stanford University, California — who dub these operations 'micromachining' — have even been able to place most of a gas chromatograph on a flat silicon wafer 5 cm in diameter. This sort of result indicates that complete electromechanical and electronic systems can be made in silicon less expensively and yet more accurately than conventional techniques would allow. The processing methods so well established in the integrated circuit industry will be capable of producing simultaneously large numbers of components in a silicon wafer with a consequent cost reduction.

All this is possible because of continuous improvement of silicon integrated circuit technology, particularly in the area of pattern definition and photolithography (the generation and transfer of a small enough mechanical pattern onto the silicon to be machined) and etching (the chemical dissolution of selected areas of silicon).

## Photolithography And Etching

Photolithography has been propelled forward because of demands made by high density electronic chips which are approaching VLSI (Very Large Scale Integration) complexity. At the moment it is possible to design patterns, 200 x or 500 x larger than life, which will eventually produce a minimum feature size of 2 or 3  $\mu\text{m}$  quite routinely. Next generation equipment will allow such features to be cut less than 1  $\mu\text{m}$ , making it possible to fabricate, easily, novel optical components such as diffraction gratings in silicon. But the key to micromachining has been the variety of etching techniques that carve detail on the silicon surface.

Some (isotropic) etchants merely dissolve silicon at equal rates in all directions, but some show anisotropic behaviour; that is, they preferentially etch only certain crystal planes of silicon. By etching faster in certain directions than others — and the relative difference in speed can be two orders of magnitude — predictable three-dimensional shapes can be cut.

What is cut depends on both the etchant used and the crystallographic orientation of silicon used. We describe

the crystallography of silicon numerically using Miller indices, which essentially state the orientation of a plane of a silicon atomic lattice by defining its intercepts with a hypothetical set of axes (Fig. 1). From the micromachining point of view the most important planes are the (100), (110) and (111) in the (cubic) silicon lattice. Etchants like potassium hydroxide or an ethylene diamine/pyrocatechol mixture essentially migrate much faster in the (100) direction than they do in the (111) direction because the packing density of silicon atoms is much lower in the 100 direction. The result when etching is that a V-shaped notch is formed in a (100) slice where the sides of the 'V' are the slower-etching (111) planes.

But the groove can be very accurately reproduced and its sidewalls will always make an angle of  $54.74^\circ$  with the surface of the silicon. Moreover the depth of the groove is directly related to the width of the surface opening etched, since etching effectively stops at the (111) planes which intersect the sides of the opening. Wafers are typically of the order of a few tenths of a millimetre thick and so it takes a few hours to etch a deep groove, or, if the surface detail is wide enough, a nozzle-shaped hole right through the slice.

## Nozzles, Valves And Beams

Dr. Ernest Bassous' group at IBM has patented a variety of nozzle structures, based essentially on this technique, which are intended for projecting very fine ink sprays in high resolution printers. Although accurate dimensioning can be achieved by simply etching right through the silicon as described above, Bassous has found that better

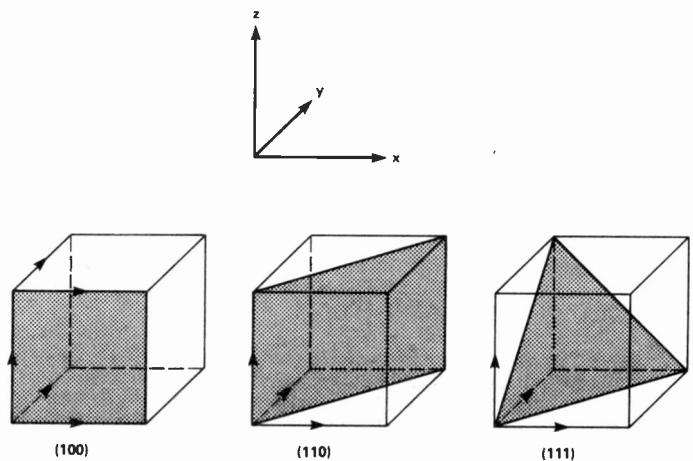


Fig. 1 The three basic index planes of silicon.

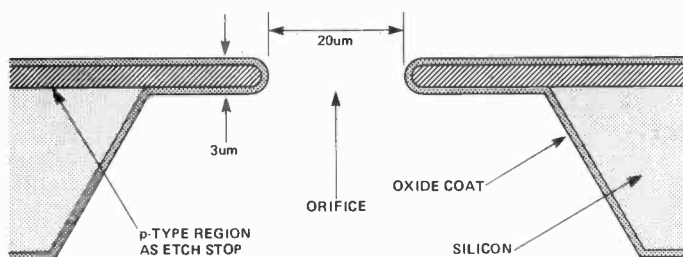


Fig. 2 The smallest nozzle in the world? IBM Research Labs made this silicon nozzle.

nozzles can be made by employing the natural resistance to etching of heavily-doped P-type silicon. A thin layer of this is formed at the back of the wafer, uniform except where orifices are required. The wafer is then etched anisotropically from its upper surface. Etching is terminated only by the thin P-barrier but punctures the slice completely in its unprotected regions. After cleaning and silicon dioxide regrowth the final structure is shown in Fig. 2. It has an orifice typically less than 20  $\mu\text{m}$  wide set in a membrane only 3  $\mu\text{m}$  thick.

Another IBM researcher, Kurt Petersen, takes membrane manufacture further. He allows the etch to deliberately undercut the overlying silicon dioxide layer and so produces an ultra-thin, springy, 'diving board' structure made entirely of oxide. The 1 mm thick membranes can be easily — but not irreversibly — bent by an electric field and IBM has used them as electronically controlled scanning mirrors to reflect illuminated data from a single character generator on to a ground glass screen for display purposes. IBM also foresees applications for them in high-isolation electromechanical switches.

A flexible structure is also the basis of Dr Lynn Roylance's miniature accelerometer to study heart wall motion. In the Stanford laboratories she made a 3 x 2 x 0.6 mm cantilevered beam unit entirely from etched silicon. The beam bends in response to applied acceleration with considerable sensitivity — it can detect an acceleration from 0.01 g to 50 g, with a 1% accuracy. The actual detecting elements for this kind of transducer are usually of a piezoresistive variety. This means that the resistance of a diffused element (usually p-type) changes when it is mechanically stressed. However, these resistive elements are usually quite sensitive to temperature changes, which is why they usually take the form of a Wheatstone Bridge circuit.

## Engine-eering

Silicon transducers, if they're small and light enough, can find their way into a whole spectrum of applications. They have been mounted on heart walls, on turbine components in aircraft and may well be shot into space on planetary probes. But most engineers believe that the traditional benefits of silicon processing (ie mass-produced, low cost, high precision techniques) will only really show when micromachining is adopted by a mass-production industry.

The US car industry could be just such a sponsor. Silicon sensors have been edging their way onto the Detroit production lines in a drive for higher fuel efficiency through better monitoring techniques in automobiles, although not as widely as predicted. William Wolber, a Michigan sensor specialist, describes the place of silicon components in the automobile industry as a 'useful addition' but warns that improvements in processing will continually be required if silicon is to be competitive. At present, the US industry wants to monitor a variety of

## MICROMACHINING SILICON

Standard silicon processing techniques basically involve one or more repeats of an oxidation-etch-diffusion cycle, outlined below:

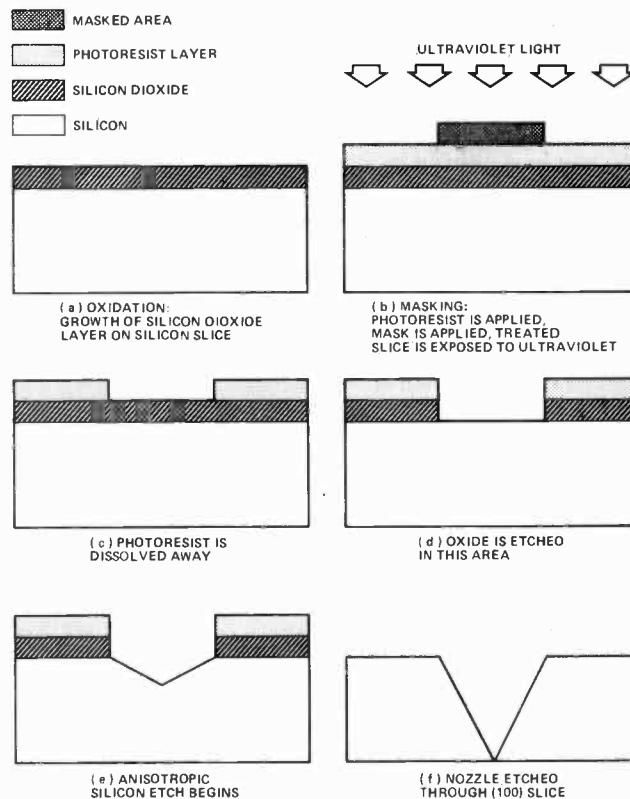
- Oxidation — A layer of silicon dioxide is grown on the silicon wafer by heating it in an oxygen stream. This layer will act as the pattern definition layer for the rest of the process (a).
- Etching — The wafer is now coated with photoresist (a light-sensitive compound) which is exposed to light via a master negative glass plate on which are detailed the features to be machined (b). Unexposed parts of the resist (underlying the negative) are then easily rinsed off, but the exposed, hardened resist remains. This serves to protect the underlying oxide from dissolution when the wafer is treated with hydrofluoric acid. Unprotected oxide is dissolved off leaving bare silicon (c,d).

If the wafer is now treated with an etching solution that dissolves silicon but not oxide, the silicon will be eaten away beneath the oxide window exclusively (e). If the etchant is *isotropic* the rate of etching will be the same in all directions. If the etchant is *anisotropic* it will etch in a preferential direction, eg potassium hydroxide solution will produce a V-groove in (100) orientated silicon and a vertical walled profile in the (110) direction (f).

For many devices, the process ends here or is cycled through again after a complete re-oxidation of silicon, depending on the profiles required. For more complicated structures (eg Dr. Bassous' membrane-nozzle described in the article) and to make electronic devices such as integrated circuits, gaseous impurities (eg boron compounds) are carefully allowed to diffuse into the silicon, through oxide windows like those made previously.

After the micromachining has been fully performed, the silicon wafer is split into individual silicon chips each containing a copy of the micromachined device.

Cross-section of silicon wafer, or slice, showing the procedure for cutting a simple nozzle through the wafer. Dimensions are not to scale.



# FEATURE: Micromachining

variables in the engine including air and coolant temperatures and fuel metering. The latter is derived from the determination of the manifold absolute pressure by silicon strain gauge techniques.

## Chromatography On A Wafer

Perhaps the most spectacular development to date, however, has been the fabrication of a gas chromatograph system on a single silicon wafer, by Dr Stephen Terry and his colleagues at Stanford. Gas chromatography, the separation of a mixed sample gas back into its components, can be broken down into three separate stages:

- sample injection from the outside world
- sample separation in a long thin column
- detection and quantity measurement of each individual component.

The sample gas is injected into the system mixed with a carrier gas, typically nitrogen. Separation occurs in the column and is determined by the relative migration rates of each component in the sample. These in turn, are influenced by both the carrier gas velocity and the relative adsorption/desorption parameters of the components, in a so-called stationary phase, a substance which lines the walls of the column. With a sufficiently long column, the individual components emerge as separate entities ready to be detected by a suitable transducer.

The Stanford instrument cleverly reduces the large-scale complexity required for such an instrument with micromachining. In particular, the separator column (which has to be long to achieve good separation) is coiled into a spiral groove 1.5 m long but only 200  $\mu\text{m}$  by 30  $\mu\text{m}$  in cross-section. It is sealed after etching with a Pyrex cover slip to make a closed capillary column. Once again, the etching is patterned through a grown silicon dioxide overcoat to define the spiral.

The components are detected by a simple, yet very effective, thermal conductivity sensor. This is essentially a nickel film resistor heated by an electric current. The temperature it actually reaches for a constant current depends on the thermal conductivity of the gas stream passing through the device.

The results of the chromatograph are impressive, especially when compared to larger laboratory instruments. It has been used so far to analyse hydrocarbon mixtures and does this very efficiently with a fast time constant — which means that sample peaks can be as short as one tenth of a second compared with the width of several seconds realized by standard laboratory instruments. This in turn means the instrument can provide better resolution of the sample.

## Material Benefits

But, of course, one need not be restricted to silicon for micromachining. With slight modification, other materials can be used as long as they are compatible in basic ways with the planar process for making chips. With a little imagination, the number of applications can match the number of different materials. And some are highly exotic. The gas chromatograph discussed above, for example, has been suggested for planetary probes and what are essentially entire chip-based 'laboratories' are now being actively researched. In the space of a few millimetres, such a chip could collect, treat and analyse a sample, even to the extent of heating or micro-refrigerating it through use of the Peltier effects. For ambitious experiments on space probes, the above could be combined with gas detectors, magnetic and electric field detectors and pressure and temperature detectors, all in silicon with the advantage of suitable signal processing electronics fabricated simultaneously.

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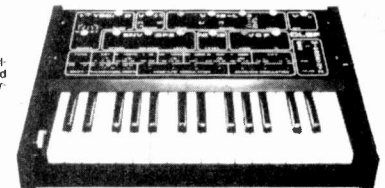
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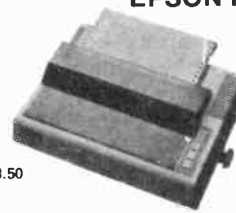
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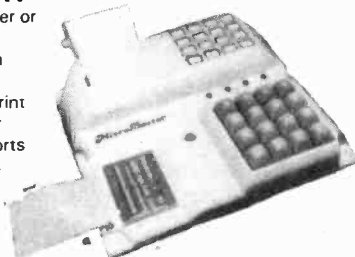
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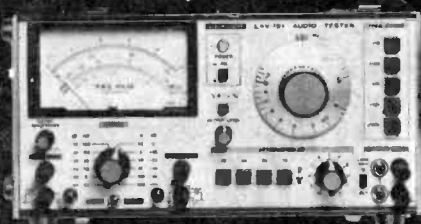
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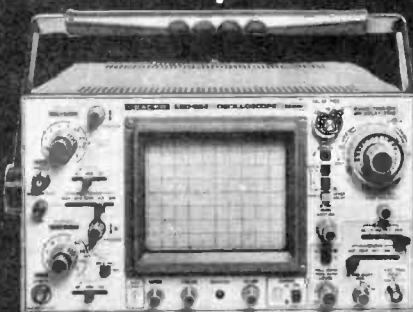


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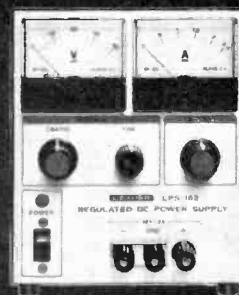
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# SWITCHED MODE POWER SUPPLY

There's a lot of unnecessary mystique surrounding the design and construction of switched mode power supplies: here's one anybody can build. Design by International Rectifier.

The advantages in using a switched mode power supply instead of a linear power regulator are well-known (if you *don't* know them, turn to Configurations in the April '83 ETI). However, the simplicity with which a linear regulator may be designed makes this configuration the most popular solution for low power applications.

The availability of low cost switching components (switching regulator ICs, magnetic components and high frequency MOSFET switches) is increasing the tendency for design

engineers to choose switched mode PSUs in their equipment. The design presented here is the work of International Rectifier and shows how such a supply may be designed quite simply by using IR HexFETs with their following desirable properties:

- no secondary breakdown
- high input impedance
- fast switching speed
- no current hogging
- no minority storage time

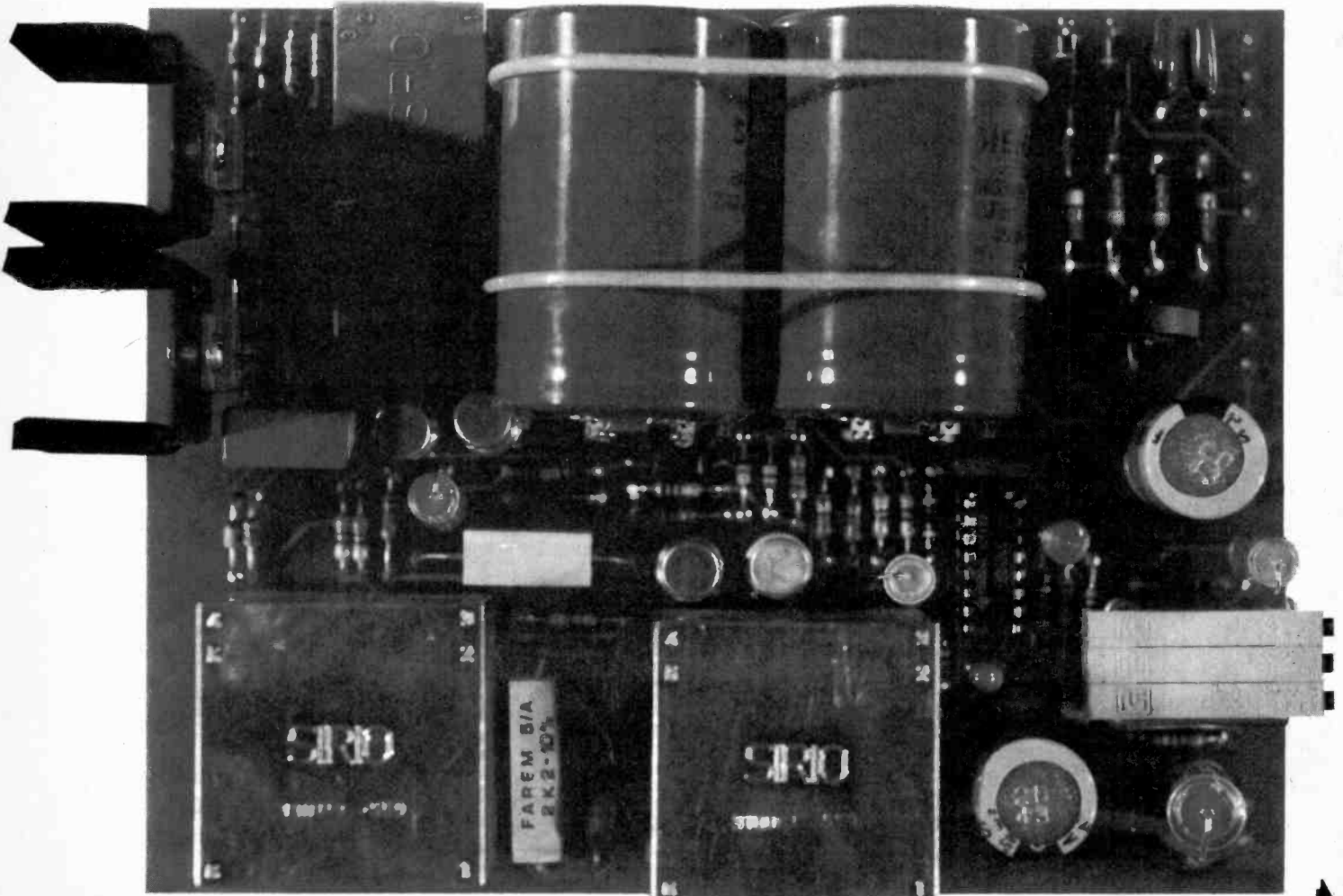
The project is built on a single double-sided PCB with plated-through holes, which carries all the components except

the rectifying diodes and the auxiliary power supply for the controller IC.

Acknowledgements are due to Ing. Bernadi, Ing. Cavalleri and the technicians of the D.T. Application Laboratory of Telettra-Vimercate for their contributions to the design of this power supply.

## System Description

This power supply uses a half-bridge configuration that can be made to operate from either 120 or 240 V AC (the half-bridge configuration was explained in more detail in the



Bird's eye view of the finished board.

Designer's Notebook feature on page 63 of the April '83 ETI).

The circuit uses two high voltage HexFETs to drive a transformer directly from the rectified mains at a frequency of 75 kHz. The transformer output is rectified and the output voltage is stabilised by changing the duty cycle of the switching waveform with a PWM inverter circuit. This function is performed by the SG3524 (IC1) and its external components.

The power supply for the control circuit (marked on the circuit diagram as the auxiliary voltage) is obtained from the mains using a 12 V, 3 W transformer and a bridge rectifier.

The HexFETs are 400 V devices in a TO-220 package and are driven by transformers so as to provide voltage isolation between the primary and secondary circuits — this is necessary because the controller IC is grounded at the load.

The power supply can be divided into four main sections:

- High power voltage stage
- Low voltage power stage
- Driving circuitry
- Control circuitry

Each of these parts will be looked at separately. Figure 1 shows the complete circuit diagram of the switched mode power supply.

### High Voltage Power Stage

The mains voltage is rectified by a bridge consisting of diodes D1-4 and is filtered by C3 and C4. C3, C4, and C5, together with the HexFETs Q1 and Q2 and the transformer T1 form the half-bridge inverter. The DC voltage across C3 and C4 is applied in alternate directions to the primary of T1 as Q1 and Q2 switch commutatively. The theoretical waveforms (both voltage and current) that appear across each HexFET are indicated in Fig. 2. The full DC voltage appears at the HexFET drain only when the MOSFET is completely cut off. A 400 V HexFET allows a 15% margin in voltage, giving a lower loss during the conduction time due to a lower on resistance.

The current flowing in the transformer primary is a square wave defined by:

$$I_p = \frac{I_o}{N} + I_\mu$$

where  $I_o$  is the output current,  $N$  the primary/secondary winding ratio and  $I_\mu$  the magnetising current. The working current for each HexFET at maximum loading is about 1 A so the choice of a device with a 400 V breakdown voltage and a current capability of 2 A is more than adequate. For this switched mode power supply the IRF722 HexFET is

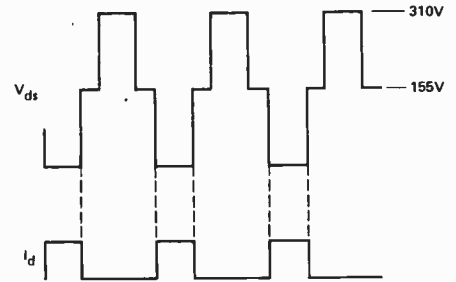


Fig. 2 Theoretical waveforms.

used, which is supplied in a plastic package.

The transformer is constructed using a Siemens pot core. The number of turns on the primary winding is defined by:

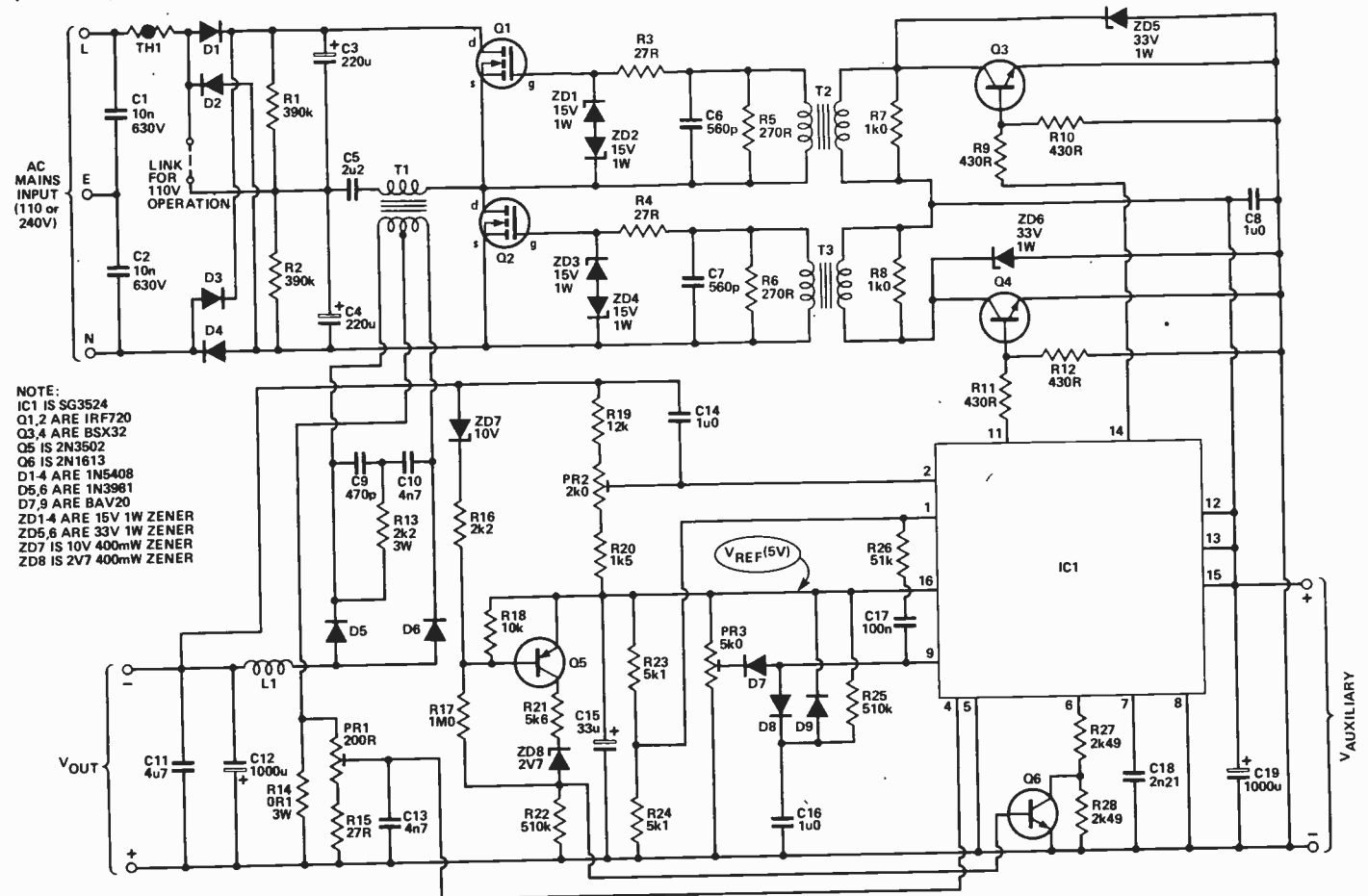
$$N_p = \frac{V_{dc}}{4B_m \cdot f \cdot A_e}$$

where  $V_{dc}$  is the peak voltage across C3 or C4,  $f$  the frequency of operation,  $B_m$  the peak operating flux density in webers/square meter and  $A_e$  is the effective core area in square meters.

The number of turns on the secondary winding is determined by this equation-

$$\frac{N_s}{N_p} = \frac{V_s}{V_p}$$

where  $V_s$  is the voltage on the secondary including all losses due to



NOTE:  
 IC1 IS SG3524  
 Q1,2 ARE IRF720  
 Q3,4 ARE BSX32  
 Q5 IS 2N3502  
 Q6 IS 2N1613  
 D1,4 ARE 1N5408  
 D5,6 ARE 1N3981  
 D7,9 ARE BAV20  
 ZD1,4 ARE 15V 1W ZENER  
 ZD5,6 ARE 33V 1W ZENER  
 ZD7 IS 10V 400mW ZENER  
 ZD8 IS 2V7 400mW ZENER

Fig. 1 Circuit diagram.

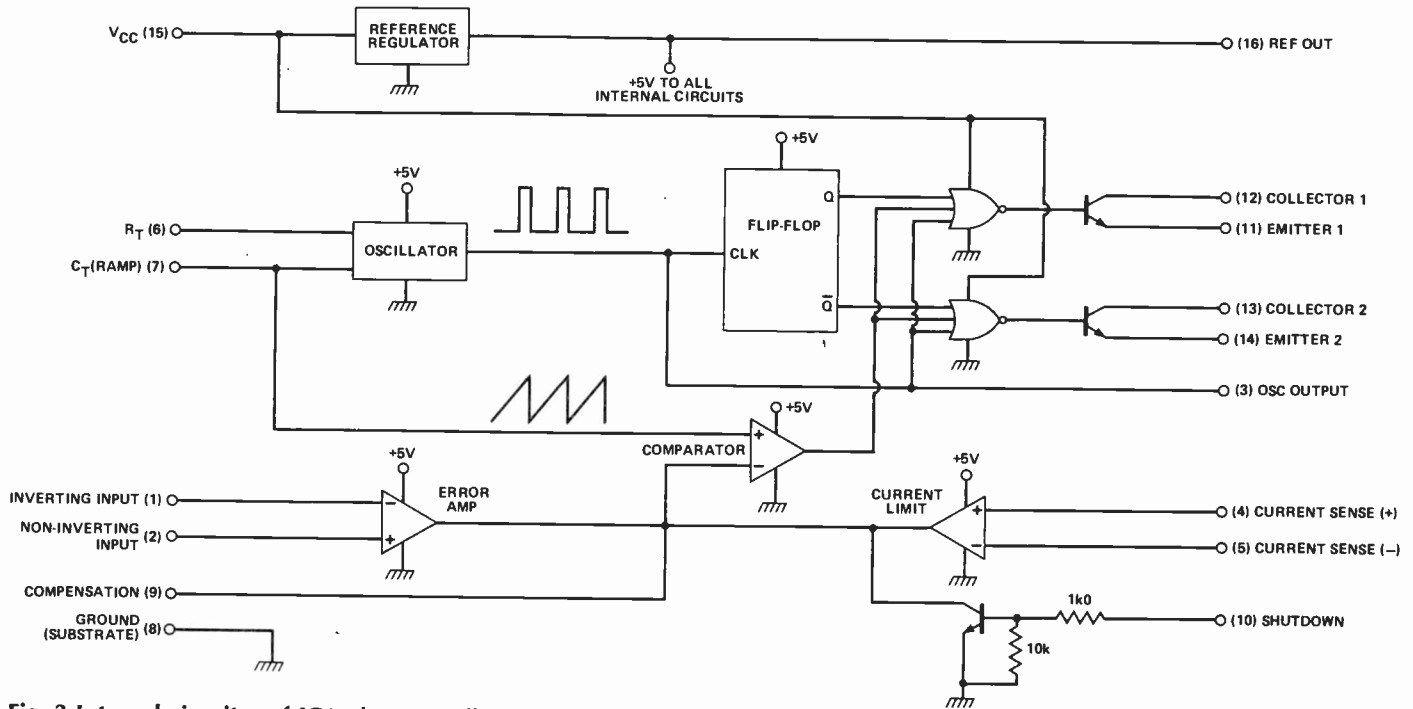


Fig. 3 Internal circuitry of IC1, the controller IC.

diodes, inductor, the current-sensing resistor and the dead time (see later), and  $V_p$  is the voltage on the primary when the system is working at the minimum allowable mains voltage and including all losses due to capacitor ripple and the input circuitry.

### Low Voltage Power Stage

The output voltage at the secondary of T1 is rectified by the high speed diodes D5 and D6 and filtered by an LC section (L1, C12). The inductor value, L, must be chosen higher than the critical value so that current circulation will be stopped, i.e:

$$L > L_c = \frac{V_o \cdot T_{off}}{2 \cdot I_o \cdot \min}$$

$V_o$  and  $I_o \cdot \min$  are, respectively, the output voltage and the minimum output current (500 mA in this design). The capacitor value is determined by the following equation:

$$LC = \frac{V_o}{\Delta V_o \cdot 8 \cdot f^2 (\eta - 1)}$$

where  $\Delta V_o$  is the desired variation in output and  $\eta$  is the ratio between the on and off switching times. To give an output with very low dynamic impedance, the value of C12 is chosen several times higher than the calculated value.

The additional components R13, C9, C10 and C11 are included to damp high-frequency ripple.

### Driving Circuitry

The HexFETs are driven by the two transformers T2 and T3. The primary of T2 or T3 loads a small switching

transistor Q3 or Q4, which is driven directly by the output of the controller IC. To clamp the over-voltage generated by T2 or T3 during the turn-off of Q3 or Q4, the zener diodes ZD5 and ZD6 are connected in parallel with the transformers on the secondaries of the transformers.

Two zeners are connected back-to-back forming a clamp which limits any gate-to-source voltage spikes to  $\pm 15$  V. The gate series resistors R3 and R4 damp any unacceptable oscillations.

### Control Circuitry

The control circuitry consists of a switching regulator IC (an SG3524) and a few other components. The IC provides all the functions necessary to produce a stabilised DC output voltage with a limited DC output current by controlling the duty cycle of the switching power MOSFETs. The control circuitry may be broken down into the following functions:

- Oscillator
- Dead time
- Soft start
- Short circuit protection
- DC output voltage stabilisation.

### Oscillator

The frequency of the oscillator is set by R27 and C18 at 150 kHz. Figure 4 shows the relationship between resistance, capacitance, and the oscillator period. The timing resistor R27 is connected to ground through R28, which is shunted by transistor Q6. This configuration allows the frequency to be halved by the automatic short-circuit protection, as described later.

### Dead Time

To avoid the simultaneous conduction of two HexFETs during the transition time, the IC generates a blanking pulse. The width of this dead time is controlled by the value of the timing resistor CT (C18) as shown in Fig. 5. To adjust this width an external circuit consisting of PR3 and D7 is added.

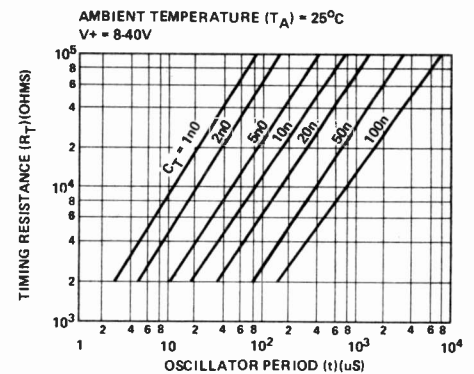


Fig. 4 Oscillator parameters.

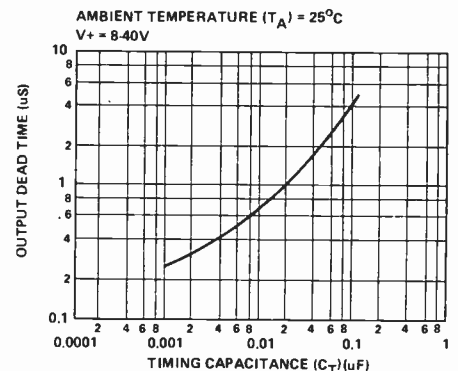
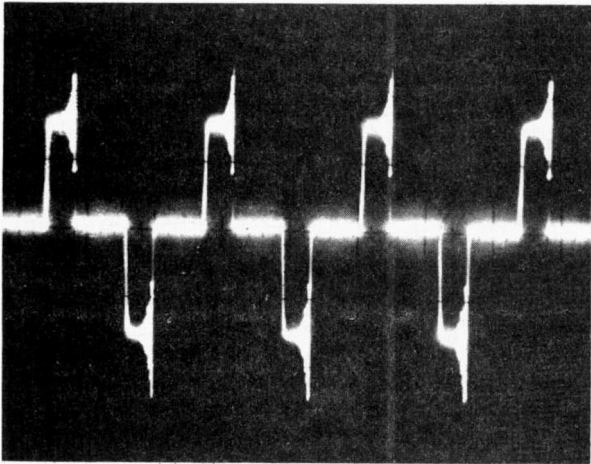
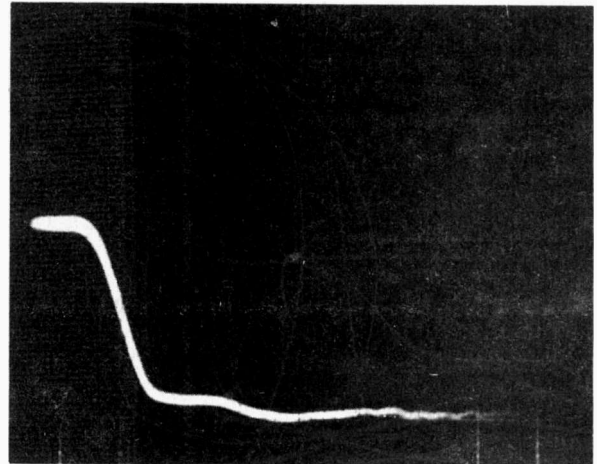


Fig. 5 Setting the dead time.



The drain voltage waveform (50V/div and 5 uS/div).



The drain voltage turn-on waveform (50 V/div and 50 nS/div).

### Soft Start

A few components (C16, D8, D9 and R25) provide a low duty cycle initially so that at start-up, a high current flow into the HexFET and the consequent possibility of saturating the transformer core is avoided.

### Short Circuit Protection

This operation is performed by sensing the current flowing in the load via R14. The voltage developed across the resistor is fed back into the current limit sense amplifier included in the IC. The current limit can be adjusted by means of the preset potentiometer PR1. R14 *must* be non-inductive to prevent any possibility of instability.

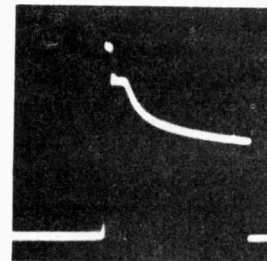
To protect the switched mode power supply from a short circuit on the secondary, the circuitry built round Q5,6 and some other components is included. The function of this circuit is to change the value of RT (by adding R28 to R27) and thus halving the frequency of operation. In fact R28 is normally shunted by Q6, but when the output voltage drops lower than 5 V Q6

is switched off. By halving the frequency, the conduction angle is increased to allow a better performance by the IC.

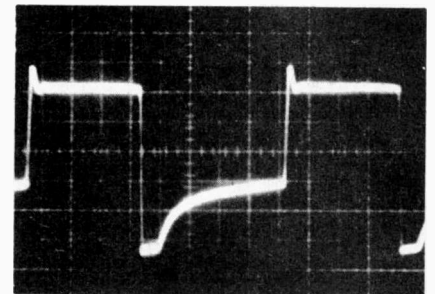
### DC Output Stabilisation

The control circuit is grounded at the load and the voltage sense for the feedback is direct. Through PR2, R19 and R20, a fraction of the output voltage is fed back into the internal error amplifier of the IC. The error amplifier is working in the common mode configuration (Fig. 5) and the gain at open loop is equalised with the network R26 and C17 to compensate for roll-off due to the LC filter on the output.

It's inevitable when travelling a little off the beaten path of circuit design that components will be required which are not readily available. Such is the case for the switched mode power supply and consequently we are arranging for a supplier to stock them. Prices and addresses will be included in next month's article, which deals with the construction of this project.



Signal at the collector of Q3, Q4.



To verify the correct functioning of the unit, check that this waveform is present at the gate of the HexFET.

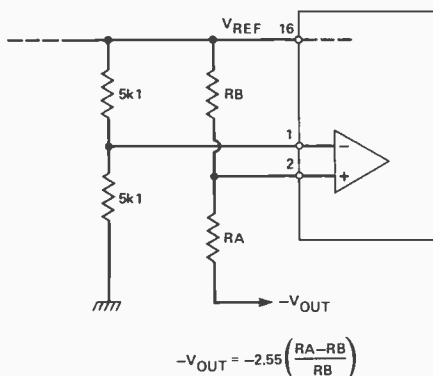


Fig. 6 Error amp operation.

## SPECIFICATIONS

Input voltage range:	110 or 220 V $\pm 10\%$
Output voltage:	12 V
Output current:	4 A
Line regulation:	0.01% at 4 A for +10% -20% line variation
Load regulation:	$\pm 1\%$ from 500 mA to 4 A
100 kHz ripple:	10 mVpp 220 V input 4 A output
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It'll take you no time at all to realise how clever Richard and Steven were to design the Jupiter Ace around FORTH. And even less time to realise what a silly price £89.95 is to charge for it.

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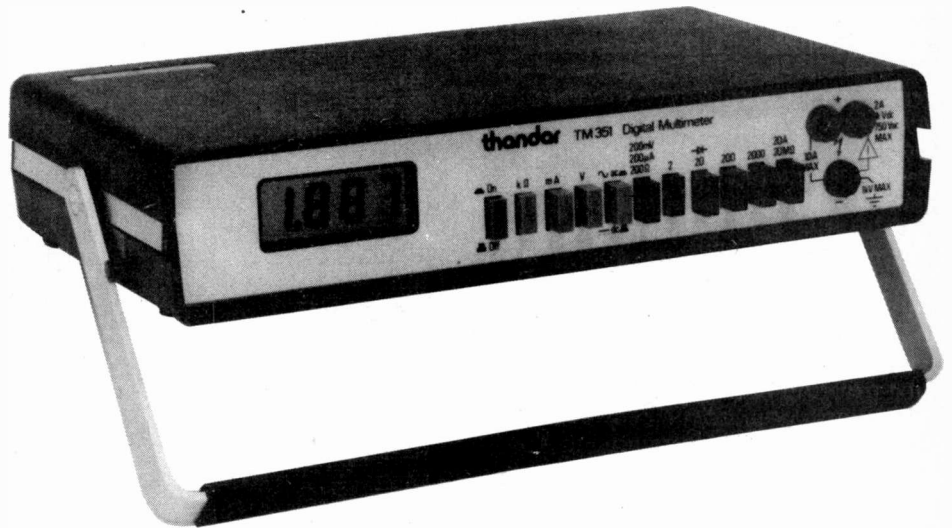
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# BUYER'S GUIDE TO

We've been getting testy with test gear, and in particular DMMs. Here are specs on over 30, and info on more besides. You'll find

The price of DMMs has fallen so much that you can now buy a reasonable quality instrument for the same price as a relatively cheap moving coil multimeter. So why doesn't everybody use DMMs and put the moving coil meter manufacturers out of business? Firstly, the cheap end of the market is likely to be dominated by moving coil meters for some time to come because the market is too small (unlike that, for example, of watches) for economics of scale to reduce the prices drastically. Secondly, and to an extent following on from the first reason, most electronics engineers and hobbyists are more familiar with moving coil meters and so have a certain preference towards them. And finally, there are some occasions when a moving coil meter can be far more useful than a digital — for instance, when you're measuring a voltage that's varying by around 10% per second, when you'd find yourself completely bedazzled by the flicker of digits on a DMM. Most engineers prefer to have a choice of the meter

The Thandar TM 351



for the job, so don't throw away your old test meter just because you're going to buy a nice new DMM.

## Multi-tudes Of Meters

In the table, we've listed all the test meters we were able to get information on, up to a maximum price of £100 (excluding VAT). Note that we have listed some meters as similar to those in the table. There were rather more meters than we bargained for, and we dread to think how many more leaflets will have arrived between our writing and your reading this! Some comments on the table will be helpful.

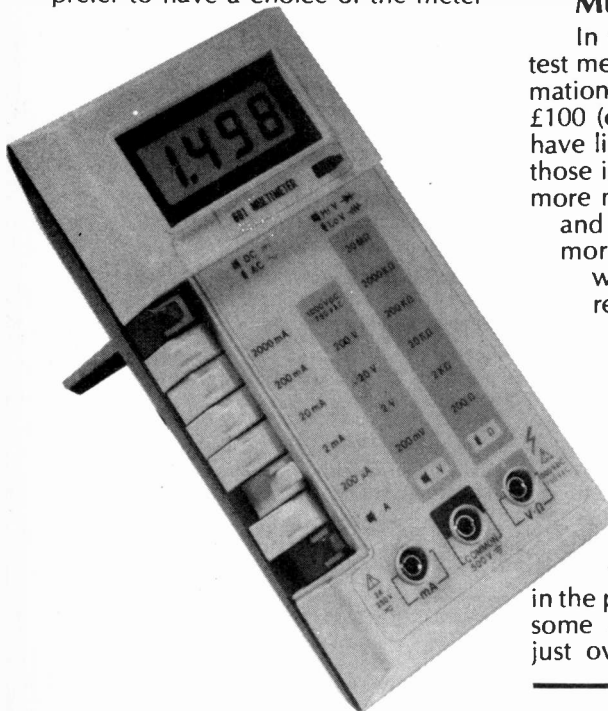
All the test meters in the table are 3½ digit types, ie the maximum scale reading is 1999.

You can obtain meters with more digits, but we did not come across any in the price range, though we did find some 4½-digit hand-held meters at just over £100. So, for example, a

200mV voltage range on our meters will have a maximum reading of, in fact 199.9mV, and will be, at best, able to distinguish (or resolve) voltages that are 0.1mV different.

## Accuracies

The accuracies quoted in the table are for guidance only. For two reasons we cannot quote exactly



The HC (well, sometimes) 601 (left) and the MIC-3300A (right): both these meters (and their relatives) turn up at a variety of sources.

# TEST EQUIPMENT

information on the latest in other types of test gear on page 48 too. Next month, oscilloscopes.

comparable figures for different meters: different manufacturers quote their error specifications in different ways, though most use the method we have adopted of quoting a percentage of the reading plus a number of digits (some quote a percentage of full scale as well as or instead of the number of digits); and different ranges of the same quantity may have different errors. For example, Keithley quote an error of 1% plus 1 digit for the 2, 20 and 200mA DC ranges of the 130, yet 2% plus 1 digit for the 2A and 10A ranges. Also, some go further and quote different errors for different frequencies on the AC ranges. So, given the space available we can do no more than give a very broad indication, and we strongly recommend that once you have narrowed down the field to those instruments you are most interested in, you should get full specs from the manufacturers or agents. You should also note that some meters have accuracy guarantees, ie, provided that the meter is used within specified temperature and humidity, and is not abused, then the meter should stay within the specifications for at least a year. Therefore it's likely that these meters will be fairly conservatively specified.

The 2000 range of Avometers: from l to r, the Digiminor 2000, the 2001 and the Vehicle Test 2002. The photograph really is the right way up.



The Elemic Digital 9: the 8 and 10 are very similar in looks

You should decide what accuracy you actually need. It isn't that often that it's necessary to measure quantities to 1%, in fact, for most purposes, 5% is perfectly adequate. However, what you might need to do is to distinguish between quantities that differ by, say, 1%, and this is a relatively straightforward task for virtually any DMM.

Note that for maximum DC and AC voltage ranges, the maximum volts is always less than the maximum reading that the display is capable of. The maximum for AC is specified as volts RMS, very often 700V RMS, which will mean a peak voltage of 1000V for a sinusoidal waveform.

## Features

Most of the features are self-explanatory, though you may be a little surprised that feature F, folds for protection, really does mean that the

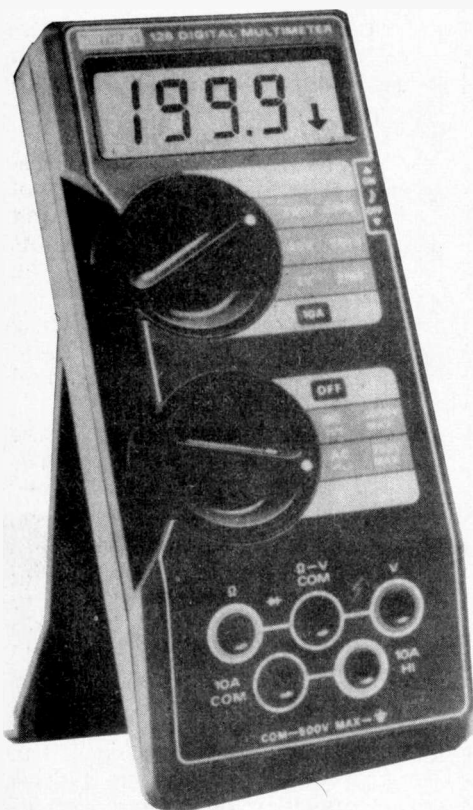
# DIGITAL MULTIMETERS

**Voltage DC Ranges**  
 (number of ranges in brackets)  
**Accuracy on DC Volts**  
 (number of ranges in brackets)  
**Voltage AC Ranges**  
 (number of ranges in brackets)  
**Accuracy on AC Volts**  
 (number of ranges in brackets)  
**Frequency Range (Hz)**  
**Current DC Ranges**  
 (number of ranges in brackets)  
**Accuracy on DC Current**  
 (number of ranges in brackets)  
**Current AC Ranges**  
 (number of ranges in brackets)  
**Accuracy on AC Current**  
 (number of ranges in brackets)

Model	Voltage DC Ranges (brackets)	Accuracy on DC Volts (brackets)	Voltage AC Ranges (brackets)	Accuracy on AC Volts (brackets)	Frequency Range (Hz)	Current DC Ranges (brackets)	Accuracy on DC Current (brackets)	Current AC Ranges (brackets)	Accuracy on AC Current (brackets)
Avometer DA211	200mV to 1000V (5) ✓	0.8+1 ✓	200V, 1000V (2) ✓	1.2+10	45 to 500	200uA to 10A (5) ✓	1.2+2	-	-
Avometer DA212	200mV to 1000V (5)	0.25+1	200mV to 750V (5)	0.75+3	45 to 500	200uA to 1A (5)	0.75+5	200uA to 1A (5)	1.2+4
AVO Digmilor 2000	2V to 1000V (4)	0.25+1	2V to 1000V (4)	1.0+3	40 to 450	20mA, 2A (2)	1.0+1	20mA, 2A	2.0+3
Avometer 2001	200mV to 1000V (5)	0.25+1	200mV to 1000V (5)	1.0+3	40 to 1k	200uA to 10A (6)	0.75+1	200uA to 10A (6)	1.5+3
BBC MA2D	200mV to 650V (5)	1.0+1	200mV to 650V (5)	1.5+3	NS	2mA to 10A* (5)	1.0+1	2mA to 10A* (5)	1.5+3
BBC MA3D	200mA to 650V (5)	1.0+1	200mV to 650V (5)	1.5+2	15 to 4k	2mA to 10A* (5)	1.0+1	2mA to 10A* (5)	1.5+2
Beckman T110	200mV to 1000V (5)	0.25+1	200mV to 750V (5)	1.0+3	NS	200uA to 10A (6)	1.0+1	200uA to 10A (6)	1.5+3
Beckman TECH 310	200mV to 1500V (5)	0.25+1	200mV to 1000V (5)	0.75+3	45 to 2k	200uA to 10A (6)	0.75+1	200uA to 10A (6)	1.5+3
Eagle TS 500	2V to 1000V (4)	1.0+1	200V, 500V (2)	2.0+1	NS	2mA to 200mA (2)	1.5+1	-	-
<b>Eagle TS 750</b>	200mV to 1000V (5) ✓	0.8+1 ✓	200mV to 700V (5) ✓	1.0+2 ✓	NS	200uA to 10A (16) ✓	1.5+1 ✓	200uA to 10A (6) ✓	3.0+2 ✓
Eagle TS 3500	200mV to 1000V (5)	1.0+1	200mV+600V (5)	1.0+1	NS	20mA to 1A (3) (+10A with shunt)	1.0+1	20mA to 1A (3) (+10A with shunt)	1.0+1
Elemic Digital 10	200mV to 1000V (5)	0.2+1	200mV to 1000V (5)	0.5+1	10 to 5k	200uA to 10A* (6)	0.2+1	200uA to 10A*(6)	0.5+1
Fluke 8022B	200mV to 1000V (5)	0.25+1	200mV to 750 (5)	1.0+3	45 to 450	2mA to 2A (4)	0.75+1	2mA to 2A (4)	2.0+3
Hansen HD30/B	200mV to 1000V (5) ✓	1.35+1	2V to 600V (4) ✓	2.3+1	40 to 500	200mA (1) ✗	2.0+1	200mA (1)	3.0+1
Hansen HD31	200mV to 1000V (5)	0.8+1	2V to 750V (4)	1.2+1	40 to 500	200mA, 10A (2) ✗	1.8+1	200mA, 10A (2)	2.0+1
HC601	200mV to 1000V (5) ✓	1.0+2	200mV to 750V (5) ✓	2.0+5	45 to 400	200uA to 2A (5) ✓	2.0+1	200uA to 2A (5) ✓	2.0+5
HC703	200mV to 1000V (5)	1.0+2	200mV to 750V (5)	0.6+3	45 to 1k	200uA to 2A (5) ✗	0.25+1	200uA to 2A (5) ✗	2.0+2
Keithley 128	2V to 1000V (4)	0.5+1	2V to 750V (4)	1.0+1	45 to 500	10A (1)	1.5+1	10A (1)	2.0+1
Keithley 130	200mV to 1000V (5)	0.5+1	200mV to 750V (5)	1.0+5	45 to 500	2mA to 10A (5)	2.0+1	2mA to 10A (5)	3.0+5
Lascar DP2010	2V to 500V (4f) ✓	1.0+1	2V to 500V (4) ✓	2.0+5	NS	2mA to 2A (4) ✓	3.0+1	2mA to 2A (4) ✓	4.0+1
Lascar LMM100	200mV to 1000V (5)	0.1+1	200mV to 1000V (5)	0.75+5	NS	200uA to 2A (5)	0.25+1	200uA to 2A (5)	1.0+5
MDS D350	200mV to 1000V (5)	0.75+1	2V to 600V (4)	1.0+2	40 to 500	200mA (1) (20A with shunt)	1.0+1	200mA (1) (20A with shunt)	1.25+1
MIC-3300A	200mV to 1000V (5) ✓	0.8+1	200V, 750V (2) ✗	1.2+3	40 to 500	200uA to 10A (6) ✓	0.8+1	- ✗	-
<b>MIC-6000Z</b>	200mV to 1000V (5) ✓	0.5+1 ✓	200mV to 750V (5) ✓	1.0+2 ✓	40 to 600	200uA to 10A (6) ✓	0.8+1 ✓	2mA, 200mA, 10A (3) ✓	1.5+1 ✓
Micronta 22-191	200mV to 1000V (5) ✓	1.0+1	2V to 500V (4) ✓	1.5+2	45 to 1k	2mA to 200mA (3) ✗	2.5+1	2mA to 200mA (3) ✗	2.5+2
Micronta 22-192	200mV to 1000V (5)	1.5+2	2V to 500V	1.5+2	45 to 1k	200mA (1)	2.5+1	200mA (1)	2.5+2
Pantec Pan 2001	200mV to 1000V (5)	0.2+1	200mV to 750V (5)	0.5+1	NS	200uA to 10A (6)	0.2+1	200uA to 10A (6)	0.5+1
Sabtronics 2015A	200mV to 1000V (5)	0.1+1	200mV to 1000V (5)	0.5+1	40 to 40k	200uA to 10A (6)	0.3+3	200uA to 10A (6)	0.5+1
Sabtronics 2035A	200mV to 1000V (5)	0.1+1	200mV to 1000V (5)	0.3+2	NS	200uA to 2A (5)	0.3+1	200uA to 2A (5)	0.25+1
Sifam DMM 2200B	2V to 1000V (4) ✓	0.3+2	2V to 1000V (4) ✓	0.4+2	45 to 700	2mA to 2A (4) ✓	0.7+2	2mA to 2A (4) ✓	1.0+3
Sifam DMM 2500	200mV to 1000V (5)	0.3+2	200mV to 1000 (5)	0.4+2	45 to 700	2mA to 2A (4)	0.7+2	2mA to 2A (4)	1.0+3
Thandar TM 351	200mV to 1000V (5)	0.1+1	200mV to 750V (5)	0.5+2	50 to 1k	200uA to 10A (6)	0.3+1	200uA to 10A (6)	1.0+2
Thandar TM 354	2V to 1000V (4)	0.75+1	200V, 500V (2)	1.0+2	50 to 500	2mA to 2A (4)	1.0+1	-	-
Trio DL-705	2V to 1000V (4)	0.5+2	2V to 1000V (4)	1.0+2	40 to 1k	20mA, 200mA (2)	1.0+2	-	-

AC Current Rating + digits	Frequency Range (Hz) for Stated Accuracy	Maximum Voltage Burden on Current Ranges	Ohms Ranges, in Ohms (number of ranges in brackets)	Accuracy on Ohms Ranges	Features (see Key)	Type of DMM/Type of Display	Similar Models (see Key)	Price, Excluding VAT	Accessories Supplied (see Key)	
—	300mV (10A 500mV)	2k to 2M (4)	x	1.0+2	BDHMNO	hand	—	bt	56.50	Avometer DA211
45 to 1k	250mV (1A 600mV)	200 to 20M (5)		NS	KLMO	hand/LCD	—	bt	79.90	Avometer DA212
40 to 2k	500mV	2k to 20M (3)		1.0+1	BCDGMNOS	hand/LCD	—	bkt	69.40	AVO Digimino 2000
40 to 450	500mV	200 to 20M (2)		0.25+1	BCDGMNOS	hand/LCD	Vehicle Test 2002	bkt	85.40	Avometer 2001
NS	NS	2k to 20M (5)		2.0+1	BKMNO	hand/LCD	MA1D	t	68.00	BBC MA2D
15 to 4k	NS	2k to 20M (5)		2.0+1	BFKMO	hand/LCD	—	t	99.00	BBC MA3D
NS	750mV	200 to 20M		0.5+3	BCKLM	hand/LCD	T100	bt	59.00	Beckman T110
45 to 400	700mV	200 to 20M		0.5+1	BDKMNS	hand/LCD	—	bt	95.00	Beckman Tech 310
—	NS	2k to 2M (4)		1.5+1	BGMO	hand/LCD	—	ct	29.95	Eagle TS500
NS	NS	200 to 20M (6)	✓	1.0+1	BKMOS	hand/LCD	—	ct	49.95	Eagle TS750
NS	NS	2k to 20M (5)		1.0+1	KMPS	hand/LCD	—	cst	74.95	Eagle TS3500
10 to 5k	200mV	2k to 20M (5)		0.2+1	BDKMO	hand/LCD	Digital 8, Digital 9	bct	83.00	Elemic Digital 10
45 to 450	250mV (2A 900mV)	200 to 20M (6)		2.0+1	ABDHMNO	hand/LCD	8021B	bt	99.00	Fluke 8022B
40 to 500	NS	200 to 2M (5)		1.35+1	CDKOR	hand/LCD	HD30	bft	38.70	Hansen HD30/B
40 to 500	NS	200 to 20M (5)		0.8+1	CDKOR	hand/LCD	—	bft	51.26	Hansen HD31
45 to 400	250mV (2A 700mV)	200 to 20M	≈	0.6+4	ABDHLMOS	hand/LCD	HC6010	bt	34.00	HC601
NS	250mV (2A 700mV)	200 to 20M		0.2+3	ABDHLMOS	hand/LCD	HC7030	bt	44.00	HC703
45 to 500	300mV	200 to 20M (4)		0.5+1	ABCDKMN	hand/LCD	135A	bt	119.00**	Keithley 128
45 to 500	300mV (2A 700mV)	200 to 20M (5)		0.5+1	AKMO	hand/LCD	131	t	99.00**	Keithley 130
NS	NS	2k to 2M (4)	x	1.0+1	BDGMO	hand/LCD	DP2010K	t	27.40	Lascar DP2010
NS	NS	200 to 20M (5)		0.2+1	ABHO	bench/LCD	LMM1001	t	75.20	Lascar LMM1000
40 to 500	300mV (2V with S)	200 to 2M (5)		0.75+3	CHLOR	hand/LCD	—	bcst	69.00	MDS D350
—	350mV	200 to 20M (6)	✓	0.8+1	ABDKMNOT	hand/LCD	—	ft	42.00	MIC-3300A
NS	350mV	200 to 20M (6)	≈	0.5+1	ABCDKMNO	hand/LCD	—	ft	46.00	MIC-6000Z
NS	NS	200 to 20M (6)	≈	2.0+1	CDKMOS	hand/LCD	—	t	47.78	Micronta 22-192
NS	NS	200 to 2M (5)		2.0+1	CDKORS	hand/LCD	—	t	64.95	Micronta 22-192
NS	NS	200 to 20M (6)		0.5+1	BKMO	hand/LCD	—	t	79.83	Pantec Pan 2001
NS	NS	200 to 20M (6)		0.1+1	HLMO	bench/LCD	2010A	t	83.00	Sabtronics 2015 A
NS	NS	200 to 20M (6)		0.25+1	BHLMO	hand/LCD	—	t	62.00	Sabtronics 2035A
45 to 700	200mV	2k to 20M (5)	x	0.3+2	BKMO	hand/LCD	—	bft	43.43	Sifam DMM 2200B
45 to 700	200mV	200 to 20M (6)		0.3+2	BHMO	bench/LCD	—	bt	66.04	Sifam DMM 2500
50 to 1k	NS	200 to 20M (6)		0.2+1	DHMO	bench/LCD	TM353, TM355	bt	99.00	Thandar TM351
—	300mV (2A 900mV)	2k to 2M (6)	x	0.75+1	BDGMO	hand/LCD	—	t	39.95	Thandar TM354
—	NS	2k to 20M (6)		0.5+2	HO(R)	bench/LED	—	bft	92.00	Trio DL-705

meter folds in two down the centre. Meters with low power ohms ranges can be useful for measuring resistance values with the components still in circuit because the test voltage they apply is 0V5 or less, which is too low to turn on transistor or diode junctions. Not all manufac-



Keithley's 128 — the 135 has an extra digit.

turers specify the maximum test voltage that their meters will apply, so we may have missed some meters that apply less than 0V5 on all resistance ranges (feature N).

### Prices

Note that these do not include VAT. Before ordering, you should check the price with the suppliers — it may be possible to get a better price than we've quoted by shopping around.

### Similar Brands

These are brands that share a similar specification, with differences from the meter in the table as detailed below:

**AVO Vehicle Test 2002** (£97.00): limited basic ranges, but wide selection of special accessories specially for automotive testing;

**BBC MA1D** (£54.00): same ranges as MA2D (except no 10A current ranges), lower accuracy;

**Beckman T100** (£49.00): same ranges as T110 (except no continuity buzzer), lower accuracy;

**Elemic Digital 9** (£58.95): same ranges as Digital 10, lower accuracy;

**Digital 8** (£52.50): as Digital 9 but no 10A ranges;

**Hansen HD30** (£36.48): as HD30B but no continuity buzzer;

**HC6010** (£37), **HC7030** (£47): as HC601, HC703 but with 10A ranges;

**Keithley 135** (£225): similar to 128 but with 4½ digits and higher accuracy;

**Keithley 131** (£139): similar to 130 but with higher accuracy;

**Lascar DP2010K** (£19.95 special offer): kit version of DP2010;

**Lascar LMM1001** (£79.95): as MM100 but with higher accuracy;

**Sabtronics 2010A** (£71.00): as 2015A but with LED display;

**Thandar TM353** (£75.00): similar to TM351, but lower accuracy, no 10A range and rotary switch range selection;

**Thandar TM355** (£75.00): similar to TM351, but lower accuracy and LED display.

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137/139 Sandgate Road, Folkestone, Kent CT20 2DE; and: House of Instruments Ltd, Clifton Chambers, 62 High Street, Saffron Walden, Essex, CB10 1EE

**Beckman** available through: Farnell Electronic Components Ltd, Canal Road, Leeds, LS12 2TU; and, Audio Electronics, Cubegate Ltd, 301 Edgware Road, London W2 1BN

**Eagle International** Precision Centre, Heather Park Drive, Wembley, HA0 1SU

**Elemic** available through: Black Star Ltd, 9A Crown Street, St Ives, Huntingdon, Cambs PE17 4EB

**Fluke (GB) Ltd** Colonial Way, Watford, Herts WD2 4TT

**Hansen** available through: Audio Electronics, Cubegate Ltd

**HC601, HC703, etc** These are available through a number of suppliers, and are often sold as 'own brands', with or without the HC prefix. However, you should check that the specs are the same as those we quote which are for the House of Instruments Ltd version.

**Keithley Instruments Ltd** 1 Boulton Road, Reading, Berkshire RG2 0NL

**Lascar Electronics Ltd** Module House, Whiteparish, Salisbury, Wiltshire SP5 2SJ

**MDS** Micro-Data Systems, Coach Mews, St Ives, Huntingdon, Cambs PE17 4BN

**MIC** available from: House of Instruments Ltd

**Micronta** available through Tandy high street shops.

**Pantec Pan** available through: Audio Electronics, Cubegate Ltd.

**Sabtronics** available through: Black Star Ltd; and, Stotron Ltd, Haywood Way, Ivy House Lane, Hastings, East Sussex TN35 4PL. (Stotron also sell the Taisei DM2350, which bears a remarkable similarity to the MDS D350, though the specs are slightly lower, for £48.00).

**Sifam Limited** Woodland Road, Torquay, Devon TQ2 7AY

**Thandar Electronics Limited** London Road, St Ives, Huntingdon, Cambs PE17 4HJ

**Trio** available from: House of Instruments Ltd

### Key to DMM Features

- A = accuracy guarantee
- B = single battery operation
- C = continuity beeper
- D = diode test facility
- F = folds for protection
- G = slider switch range/mode selection
- H = push-button range/mode selection
- K = rotary switch range/mode selection
- L = low power ohms range
- M = manual ranging
- N = low power ohms on all ranges
- O = overload protection
- R = autoranging
- S = stand
- T = transistor test facility

### Key to accessories supplied

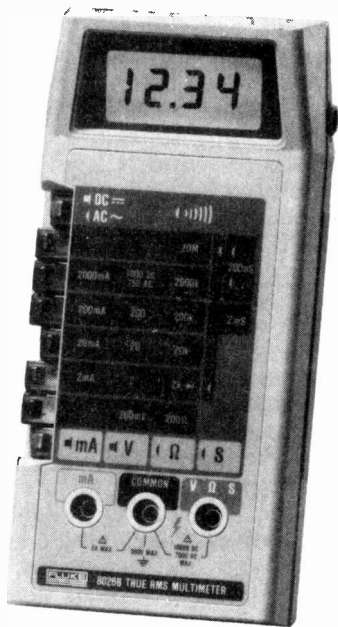
- b = battery
- c = carry case
- f = sapre fuse
- k = test clips
- s = high-current shunt
- t = test leads
- \* = 20A for limited period
- \*\* = likely to be discounted
- NS = not specified

### Addresses

**AVO** Thorn EMI Instruments Limited, Archcliffe Road, Dover, Kent CT17 9EN

**BBC** available through: John Minster Instruments Ltd,

## Fluke Hand-held True RMS DMM.



Fluke has expanded its best-selling range of 3½ digit hand-held digital multimeters with the introduction of a version with true RMS measurement capability. Fluke pioneered true RMS measurement on hand-held DMMs with their recently launched 4½ digit family and have now introduced this useful facility to their 3½ digit range.

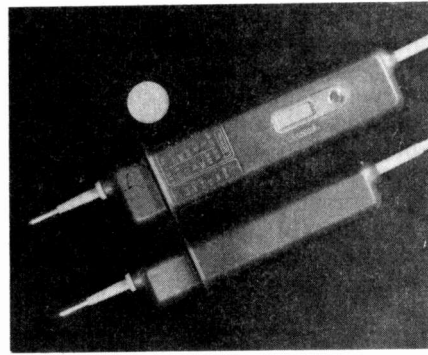
Called the 8026B, this new model provides all the facilities and performance of the 8020B series models such as 0.1% basic DC accuracy, high speed continuity beeping for open/short continuity testing, and conductance which allows high resistance measurements from 20 Mohms to 10,000 Mohms to be made.

True RMS capability is most useful when measuring non-sinusoidal waveforms as in digital equipment, e.g. modems, terminals and monitors as well as in motor and thyristor circuits, noise measurement etc. Conventional meters using only averaging techniques can introduce errors of up to 30% or more in such applications.

Other features of the 8026A, which is priced at £180 + VAT, include a heavy duty 600V dual-fuse system to protect against high-energy input signals, a very rugged design, non-skid rubber feet and a tilt bail for ease of use in the field. The new instrument comes with a two-year guarantee and labour warranty and a one-year calibration cycle. For further information, contact Fluke (GB) Limited, Colonial Way, Watford, Herts. WD2 4TT. Tel: 0923 40511.

## New Source for Digicheck

The Steinel Digi-Check, a voltmeter and ohmmeter of a rather novel design, is now available from Electronic and Computer Workshop Ltd, 171 Broomfield Road, Chelmsford, Essex CM1 1RY.



## Temperature Measurement From Your DMM



Anyone who has access to a digital multimeter can now use it as a versatile wide range temperature measuring instrument using standard type K thermo-couples, by adding the DVM/TC Interface Unit, which costs £36 (+ £1.35 p&p + VAT) from the makers.

This new device, at considerably lower cost than a dedicated instrument, has a temperature range of -50C to 1100°C and incorporates automatic cold junction compensation. Thermocouples are attached through a miniature compensated socket. A basic thermocouple and mating plug are supplied as standard with the instrument. The output of 1mV per degree centigrade is via a 0.75 metre coiled lead fitted with 4mm plugs. Long term stability is claimed to be excellent and the low battery drain allows it to be used for continuous monitoring if necessary.

Full details are available from the manufacturers Graham Bell Instrumentation, PO Box 230, 39 Derbyshire Lane, Sheffield, S8 0TH. Tel: 0742 582370.

## Enhanced Logic Analyzer

Two optional enhancements for their Model 632 Logic Analyzer announced by Zicon Instruments of Norwich.

The Model 632 is designed as a high performance, low cost instrument for use in troubleshooting synchronous logic systems. It has sixteen data channels, two qualifiers and a clock input which, coupled with its integral hexadecimal display, allows it to be used "free-standing" in the analysis of microprocessor systems. The unit also provides oscilloscope outputs which are used to generate the familiar timing diagram.

In its standard form, the Model 632 is capable of capturing 128 16-bit words at speeds up to 4MHz with 64 words pre-trigger.

Data can be recorded using either edge of the external clock; and a clock qualifier allows (for example) only-memory reads or only-data outputs to be recorded. The trigger word is set by four thumbwheel switches in hexadecimal code with disable/enable selection for each switch. A 1-X-0 trigger qualifier expands the trigger word to 17 bits or it may function as an external trigger.

The stored data can be examined in one of two ways. The integral 4-digit hexadecimal display shows one word at a time with its position in memory relative to the trigger word indicated by a 2-digit cursor display extending from -64 to +63. A timing diagram output is also provided which generates a 16-channel display on any general purpose oscilloscope with trigger and cursor position markers.

An optional 12 MHz memory speed (Option 11) is now available on this model. As a further enhancement (Option 12), Zicon Instruments have designed and produced a dedicated IEEE-488 bus data pod for use with Model 632. This probe connects directly to the bus and permits the monitoring of bus activity in either state or timing format with 8 input lines switchable between handshake and management lines or flying leads for user connection.

Current prices are: Model 632 £1095; Option 11 £295; Option 12 £295. Zicon Analyzers are distributed by STC Instruments Ltd of Harlow and Elex Systems of Bracknell. Zicon Instruments Ltd, 23 Meteor Close, Airport Industrial Estate, Norwich NR8 6HQ. Tel: Norwich (0603) 400083.

## High-Current Clamps For Beckman Multimeters

A range of three current clamps is now available to extend the AC and DC current ranges of all Beckman Instruments' multimeters. The ranges are 10 to 150A RMS AC (model CT-231, £19.00 inc. VAT), 10 to 1000A RMS AC (model CT-232, £98.50 inc. VAT) and 0 to 600A AC and DC (model CT-233, £69.00 inc. VAT, available June).

With the multimeter in the 200mA AC position, the CT-231 displays the current directly in amperes. With the CT-232, the multimeter is set in the 2A AC position and the current reading is multiplied by 1000. Depending on DC or AC measurement, the multimeter using the CT-232 is set to either 200mV or 2V DC or AC position, the readings being directly in amperes and amperes/1000 respectively. (The clamps should be usable with other brands of DMM — Ed.). Beckman Instruments Ltd, Mylen House, 11 Wagon Lane, Sheldon, Birmingham B26 3DU. Tel: 021-742 7761.

## Low Cost Capacitance Meter



A portable LCD digital capacitance meter costing £49.50 plus VAT marketed by Semiconductor Supplies, Sutton, has a 0.1 pF to 2000 uF range covering virtually all electronics engineering applications. A 3½ digit display is provided.

The DM 6013 is supplied with an instruction manual, alligator test leads and a spare 0.2A fuse. For further information contact Semiconductor Supplies International, Dawson House, 128/130 Carshalton Road, Sutton, Surrey SM1 4RS. Tel: 01-643 1126.

## National Agreement

Wessex Electronics Ltd of Bristol has been appointed by National Panasonic (UK) Ltd to market National measuring and testing equipment.

Among the products Wessex will be distributing is a 20 MHz, 32 channel logic analyser. The VP-3620A is a dedicated unit catering to a wide range of industrial needs. It contains a logic state analysing function for self-contained microprocessor operation analysis and a logic timing function for the evaluation of peripheral operations. With its multifunction capability, National's logic analyser is also able to display the traced data in disassembled mnemonic mode corresponding to each microprocessor in use.

The National Panascope will also be available from Wessex. The new oscilloscope (VP-5512A) offers DC to 100 MHz with an alternate sweep function. There is four channel, eight trace capability so that a large number of signals can be simultaneously observed. The Panascope features an advanced dome-meshed CRT and an auto-fix circuit for easy triggering. Ideal for field or lab use, the Panascope has a tv sync separation circuit for video

signals and variable hold off function for trigger stabilisation.

Advanced technology utilised in the Panascope, such as reduced number of parts for greater reliability and epoxy circuit boards for heat and shock resistance are also featured in the logic control series oscilloscopes (VP-5520B and VP-5530B) manufactured by National and to be marketed by Wessex. Both the VP-5520B and the VP-5530B offer multi-trace capability and facilitate accurate signal measurement up to 200 MHz and 300 MHz respectively.

Making up the initial complement of National products to be marketed by Wessex Electronics is a new generation A4 digital graphic plotter. Offering six colour graphics and plotting speeds up to 400mm/sec, the plotter (VP-6801A) allows great versatility in the creation of graphs and line definitions. The VP-6801A features a high level of intelligence to simplify the external programming required to generate complex graphs, shapes and alphanumerics. Computer interfaces available include both GP-IB and RS-232C, thus allowing the VP-6801A to be used as a computer graphics device as well as for instrumentation graphics. Wessex Electronics Limited, 114-116 North Street, Downend, Bristol, BS16 5SE. Tel: 0272 571404.

National's VP-3620A Logic Analyser



## Platinum Digital Thermometer

A pocket-sized, digital thermometer, with +/- 0.1 degrees C accuracy over an extended temperature range from -200 to +800 degrees C, is announced by Ancom Ltd. The Ancom BLR-800 offers a one degree C resolution, very good repeatability and a stability figure of 0.01 degrees C per degree C, with a wide range of standard platinum RTD temperature sensor probes. These include models for measurement of air temperature, liquid immersion, surface

temperature, hypodermic insertion and fast response types. The sensors are available with conformity to British and DIN standards in grades A, B, C, D and E and are simply connected to the instrument by a three-way plug/socket.

The BLR-800 uses a low power 3½ digit LCD for data display, which also displays a low battery sign.

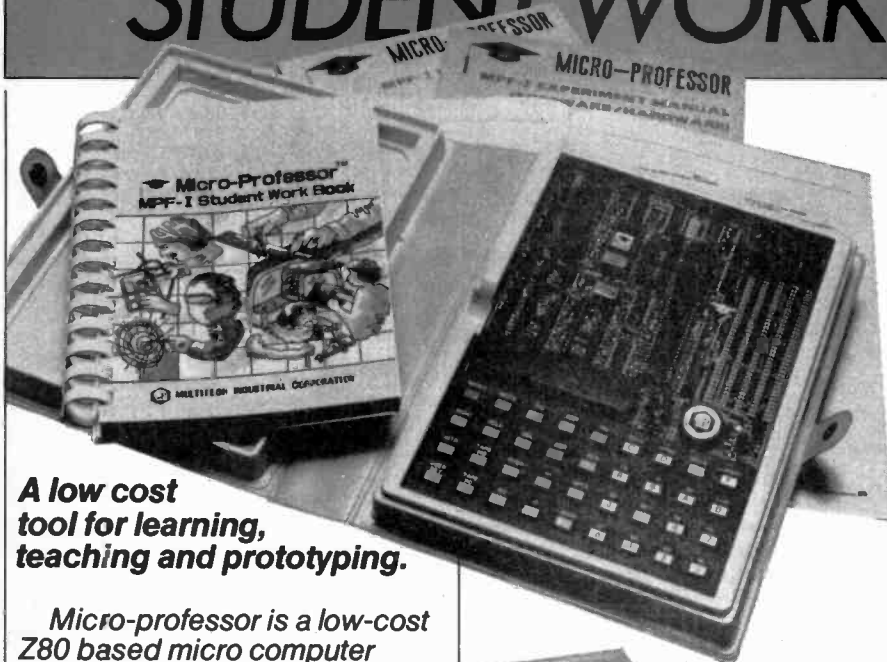
Calibration is very simply carried out by the user, using boiling water and melting ice references.

For further information please contact Ancom Limited, Devonshire Street, Cheltenham GL50 3LT, Tel: 0242 513861.

ETI



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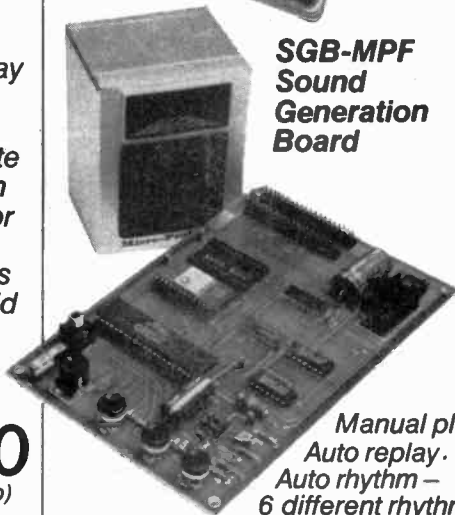
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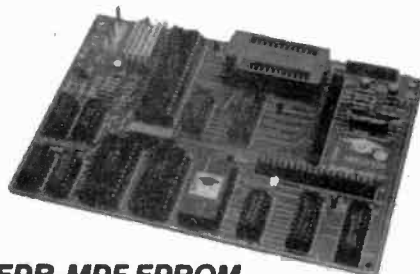
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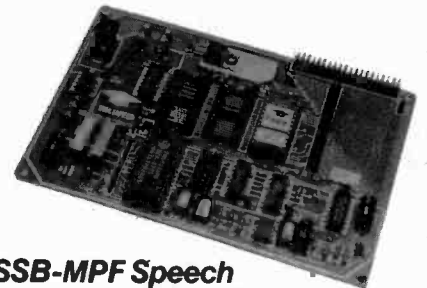
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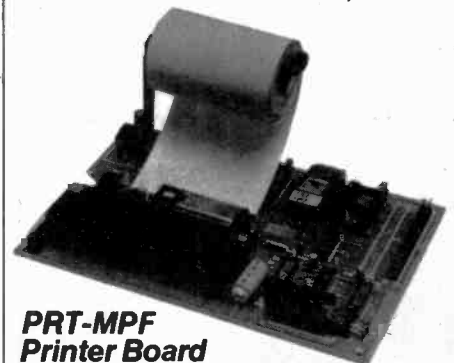


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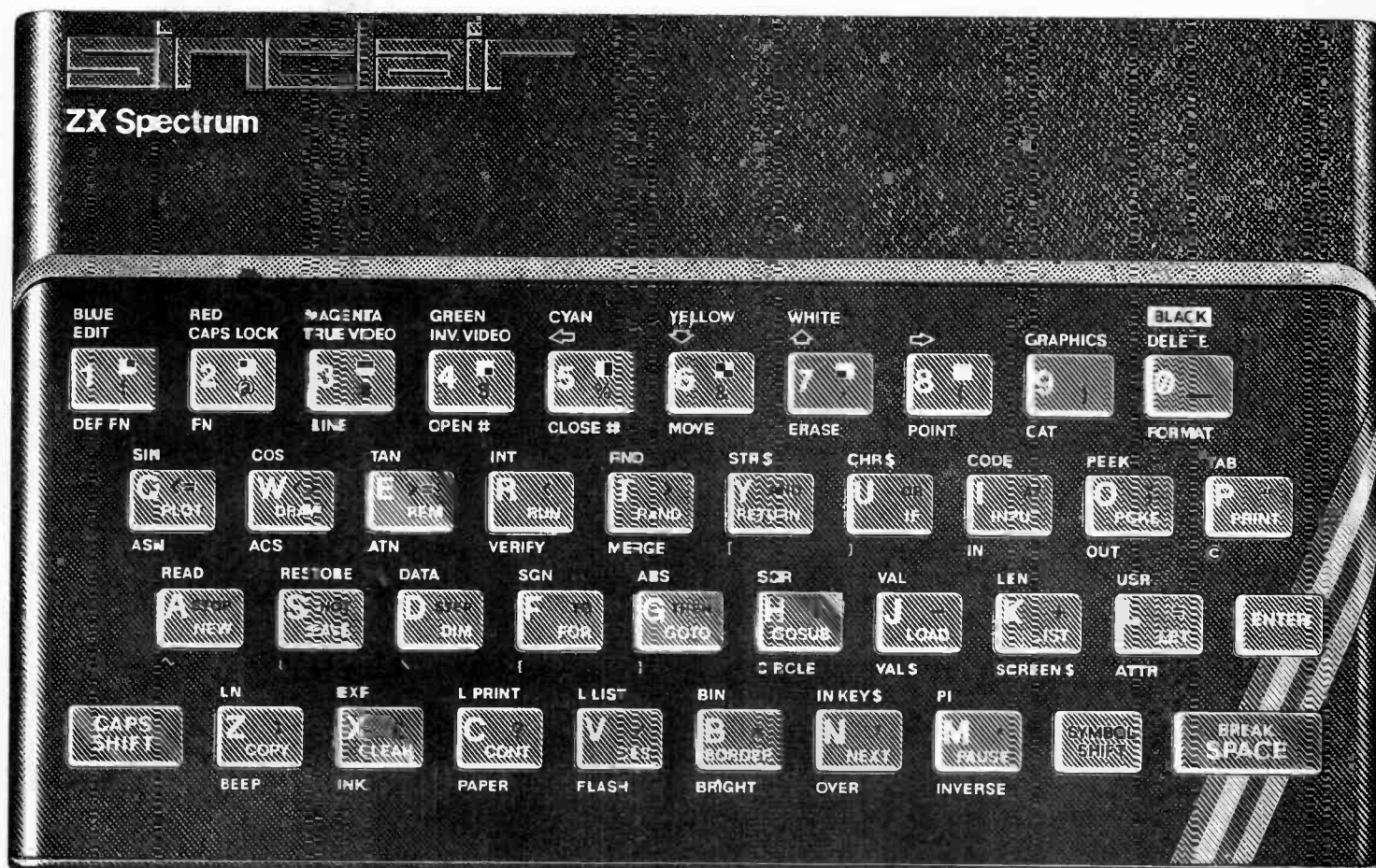
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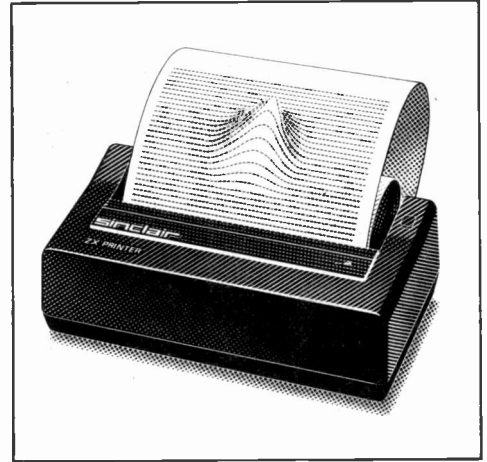


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# PSEUDOROM

EPROMs never forget but they're a damned nuisance to reprogram. RAMs are easy to overwrite but they lose their contents on power-down. ETI, naturally, has combined the best of both worlds. Design by Phil Walker.

The ETI PseudoROM now offers the home constructor a device having the capacity of the larger ROM and EPROM chips, but with the programmability of RAM, which can be inserted into the existing ROM socket of your microcomputer. This makes it possible to develop large and complex operating software in one module which only occupies the same physical board area as the eventual ROM or EPROM (if used).

Another advantage is that the access time of the module is only 50 nS or so slower than the memories it contains (if you're using both the 74HC32 and the 74HC138). This will be somewhat less than the normal 450 nS spec of

many EPROM devices advertised, and could enable you to run your system at a higher clock frequency (processor permitting), so increasing throughput.

## The Circuit

The circuit used in this project is essentially very straightforward and consists of four 2K by 8 bit CMOS RAM chips, connected to an address decoder to make an effective 8K by 8 bit memory. By means of a small battery and some extra circuitry this sizeable chunk of memory will retain its data even when unplugged from its socket. To make it even more useful the unit is constructed such that it can be used as four areas of 2K by 8 read/write memory or alternatively four sections of 2K, two sections of 4K or a single section of 8K by 8 read-only memory. All these modes are selectable by the integral switches, except that 8K by 8 is only available when inserted into a 28 pin socket.

The point of this versatility is that the unit can be programmed in an existing RAM socket (or one can be provided) at any time and then transferred to an EPROM or ROM socket without losing data (as long as the write protect switch is used). This means that quite large sections of operating systems or special software can be modified and tested without the delays of erasing and reprogramming EPROMs or losing data in system crashes.

## 2-4-8K

The only difference between the 4K and 8K versions of this unit lie in the type of DIL plug used, the size of socket into which it is plugged and the numbering of the input/output pins on the circuit diagram. For memory simulation up to 4K only 24 pins are needed, while for 8K all 28 pins are required. The easiest way of inserting the unit correctly is to identify the 0 V pin and make sure it goes into the 0 V pin position in

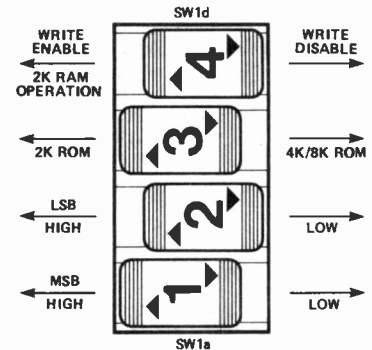
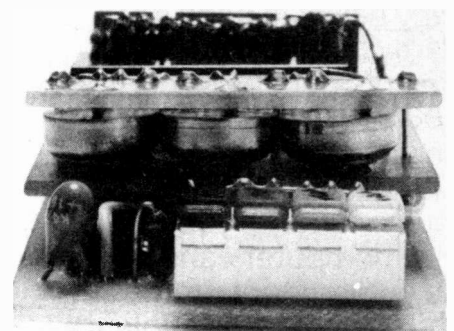


Fig. 1 Switch operations for the PseudoROM.

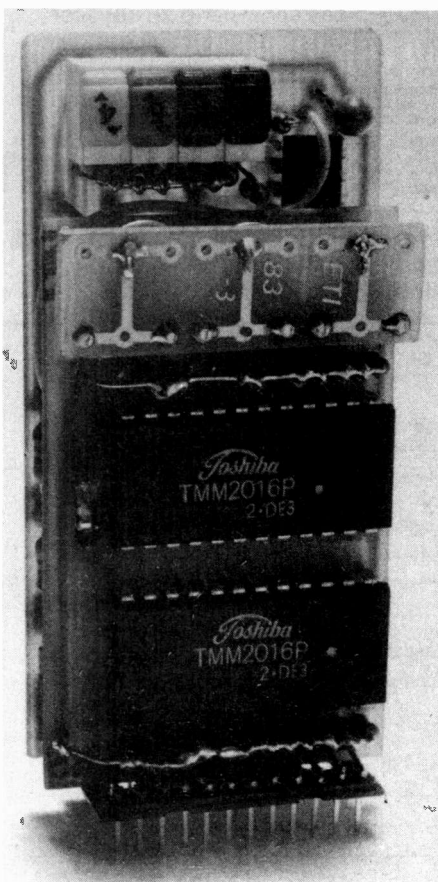
the socket used. For 24 pin sockets this is pin 12, for 28 pin sockets it is pin 14. Note that for 28 pin plugs and sockets the equivalent functions will occur on pin numbers having a value of 2 greater than those of 24 pin sockets, with the exception of +5 V on pin 28 and A12 on pin 2 (this latter address line goes to the connection marked (2\*) on the circuit diagram).



View from the top.

## BUYLINES

The 74HC series logic ICs which are required for this project are fairly new to the hobbyist scene and appear so far only to be stocked by Ambit International, 200 North Service Road, Brentwood, Essex CM14 4SG. The DIL switch shouldn't cause any problems but if you do have difficulty Maplin stock a suitable type. The mercury cells should be obtainable from most high street chemists and photographic shops, while the PCBs are available using our PCB Service order form on page 95.



Here's a fine, upstanding project . . .

## HOW IT WORKS

The essential part of the circuit consists of IC1-4. These are 6116 CMOS 2K x 8 RAMs, which have the property of consuming very little power when not in use. However, for this to happen pin 18 (CS) of each device must be at a logic high level and preferably within a few per cent of the supply voltage. Also, the remaining device pins should be held close to either supply rail. R1-19 hold the address and data pins of the RAMs close to 0 V (via R29) when the unit is unplugged, or to the positive supply rail when in position.

ZD1, R28,30,31 and Q1 sense the presence of a suitable external power supply. When this is found the inputs to IC5a go low which causes its output to go low also. This then enables IC5b, IC5c and IC6 to operate. IC5c acts as a buffer for the OE control (pin 20), while IC5b buffers the WE function (normally pin 21). IC6 is used both to buffer the CS input and to use the two most significant address lines to select which RAM device is to be activated. SW1a and SW1b with their associated resistors provide default addresses to IC6 when direct inputs are not available. SW1d selects either ROM or RAM type operation while SW1c selects 2K or 4K/8K ROM simulation.

D1 isolates the memory power supply from that of the host machine when the main power is off and is a germanium device for minimum voltage drop. Likewise, D2 isolates the backup battery from the main supply when not in backup mode. When the external supply fails, the control signals to the RAM ICs are all forced to logic HIGH and the battery will be able to maintain the data in the RAMs for a long time.

## Construction

Examine the overlay diagram and photographs very carefully . . . The unit consists of three PCBs mounted one on top of another. These must be assembled correctly and put together in the right order as there may be no second chance. We suggest that you follow the procedure here so that it goes together correctly, but read it carefully before starting.

Start with the smallest PCB: this is the battery connector. Take three thin brass shims about 0.1" (2.5 mm) wide by 0.6" (15 mm) long and solder each to the central bar of each pattern on the PCB such that they overlap one edge by 0.4" (10 mm) or so. Bend the overlapping length round the edge of the PCB so that it lies near the non-track side. This will form the positive contact for each cell. Alternatively, any springy material to hand may be soldered into the centre hole to do the job. In fact we

used contacts from a piece of edge connector.

Next solder six lengths of 20 swg copper (or paper clip) wire into the holes nearest the free end of the battery connector described above. These wires will connect to the next board down and complete the battery holder.

The middle-sized PCB should be assembled next: this will form the middle of the sandwich. Be very careful to get the right value resistors in the right holes. There are nineteen 47k resistors and one 1k0 resistor on the board, all of which mount vertically. Insert them into their holes and solder into position as close as possible to the board. DO NOT CUT ANY OF THE LEADS OFF.

Using two 2½" (60 mm) lengths of 22 swg wire, connect the free ends of each line of resistors together and pass the remainder of the wire through the hole shown on the overlay; solder into position but

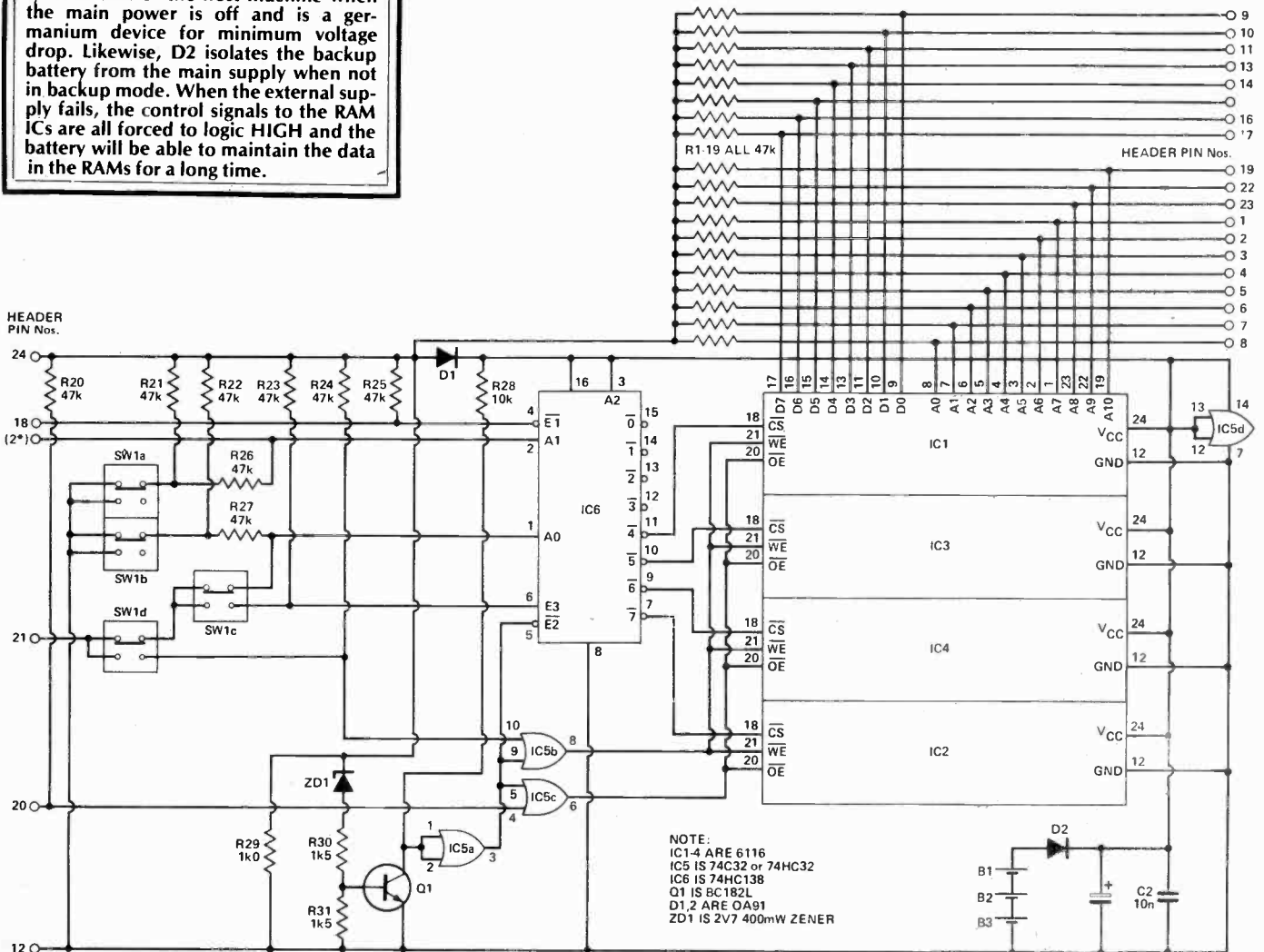


Fig. 2 Complete circuit diagram for the PseudoROM. All the header pin numbers refer to the 24-pin version: add two to all the numbers for the 28-pin version except for the pin marked (\*). This is left unconnected on the 24-pin version.

DO NOT CUT OFF. There are seven more links to be made to the lower board which should consist of 1" (25 mm) lengths of 22 swg wire. A useful tip — squeeze the end of the wire with pliers to help it stay in the hole when soldering. Now mount D2 on the PCB making sure that the polarity is correct.

The next step is to assemble the battery connector onto this board. First insert some 22 swg tinned copper wire into the three holes in the middle PCB which lie in the centre of the cell positions. Solder them into position and crop off close to the track side and to about  $\frac{1}{4}$ " (6 mm) on the component side. Bend this short wire towards the edge of the PCB and flatten it (... GENTLY ...). These form the negative battery contacts. Now take the small PCB and insert the six wires from it into the corresponding holes in the medium-sized PCB. Put three of the specified mercury cells in position and adjust the two PCBs until the cells are held reasonably firmly in their correct positions. Solder the link wires into position and crop them close to the PCB; then remove the cells.

Solder 12  $\frac{1}{2}$ " (12 mm) lengths of 22 swg wire (14 lengths for a 28-pin plug) into the row of holes on the edge of the PCB (starting at the R29 end) such that the wire projects on the component side. Now carefully insert the two memory devices into the PCB and solder them into position. **MAKE ABSOLUTELY SURE THEY ARE THE RIGHT WAY ROUND and TAKE FULL PRECAUTIONS AGAINST STATIC DAMAGE.** Crop the IC leads close to the foil side.

The largest PCB can now be assembled. First insert all the

### PARTS LIST

#### Resistors (all $\frac{1}{4}$ W, 5%)

R1-27	47k
R28	10k
R29	1k0
R30,31	1k5

#### Capacitors

C1	4u7 16 V tantalum
C2	10nF disc ceramic

#### Semiconductors

IC1-4	6116
IC5	74C32 or 74HC32
IC6	74HC138
Q1	BC182L
D1,2	OA91
ZD1	2V7 400 mW zener

#### Miscellaneous

SW1	quad SPDT DIL switch
B1-3	PX675 mercury cell
PCBs (see Buylines); 24 pin (or 28 pin)	
DIL header plug; wire, thin brass shim or other contact material.	

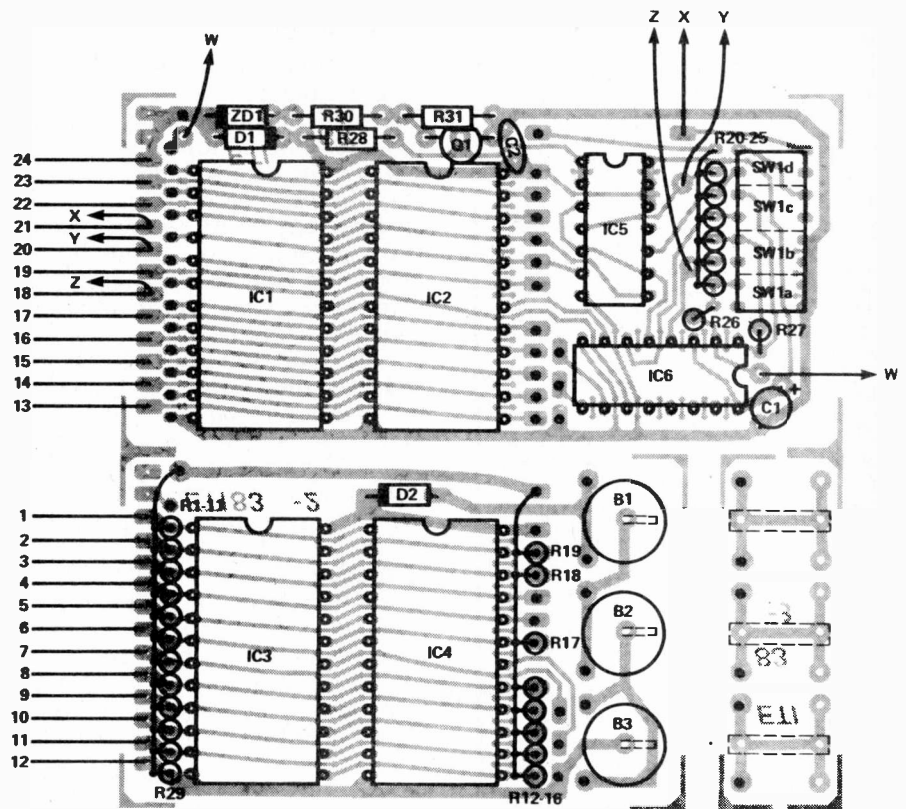


Fig. 3 Component overlay for the PseudoROM. Remember to separate the three boards before soldering!

resistors, diodes, capacitors and transistor. Note that six of the resistors (all 47k) are mounted vertically as before with their free ends connected together and to a hole in the PCB. This time, however, the leads can be cropped close to the PCB. Next fit the DIL switch SW1, the two 74 series CMOS devices, and the remaining two memories. REMEMBER THAT ALL THE ICs ARE SENSITIVE TO STATIC. Make sure that they are the correct way round before soldering them in. Finally on this PCB, use four lengths of THIN insulated wire to make the long links shown on the overlay (W-W, X-X, Y-Y, Z-Z).

Having reached this stage it would be advisable to recheck all the solder joints on all the boards with a magnifying glass for accidental blobs, splashes or other faults.

From now on it will be virtually impossible to rectify any constructional errors. So check again!

Take the smaller PCB assembly and crop the wires projecting on the foil side so that the three links nearest the battery holder are virtually full length but the longest is at the edge and the shortest is farthest from it. Next crop the two lines of resistor leads and links so that the longest are about  $\frac{3}{4}$ " (18

mm) and the shortest are about  $\frac{1}{4}$ " (12 mm), graded evenly across the board width. This is done to make the next operation easier.

Now the tricky bit. With great care feed the wires from the smaller PCB assembly into the corresponding holes in the larger PCB. Make sure that the wires are straight when you do this and don't allow the long links to cross each other on top of the memories. It should be possible to get the two PCBs down to about  $\frac{1}{4}$ " (6 mm) apart. If you can't do this it may be that the transistor or capacitor is in the way. Rectify this before carrying on. The transistor must be very close to the board. If this is not the problem one of the links may be bent. This is a fault, as when properly assembled all the links between boards are straight.

When this stage has been reached successfully, solder all the inter-board links and crop off the excess wire.

The end is now in sight. Examine the connector end of the assembly you have made and offer up the 24 or 28 pin plug you wish to use. It may be necessary to file a little material off one of the PCBs so that the plug will lie square against the two PCBs when the pins are soldered in position. Do this now if necessary.

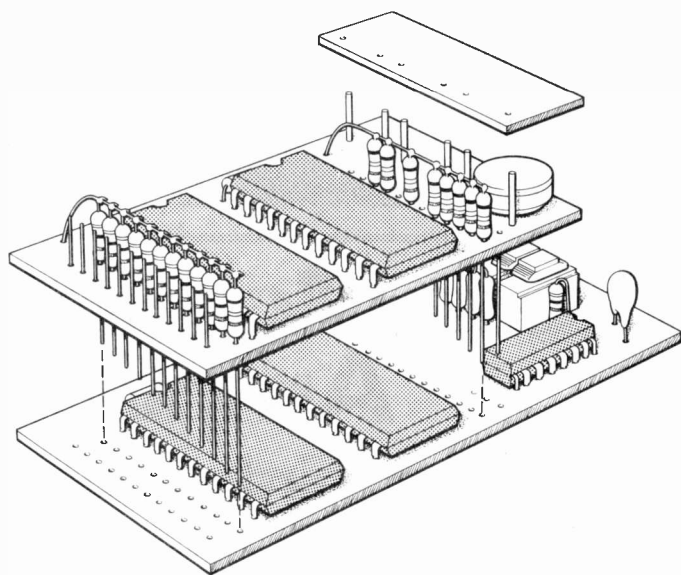
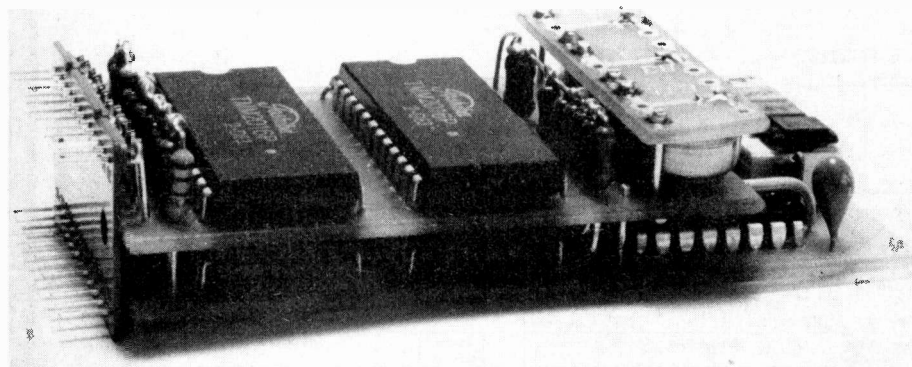


Fig. 4 An exploded view to help you through the traumas of construction. Compare with the photo below.

If there is a large blob of solder on any of the three pads where the links have been soldered already, remove some of it with a solder sucker or piece of fluxed braid. This is to allow the plug pins to seat down onto the PCB foil. Align the tags of the plug labelled with pins 13 to 24 (or 15 to 28 for the larger plug) against the pads along the edge of the largest PCB in the assembly. Solder them carefully to the pads such that pin 24 connects to the innermost of the three joined pads (or pin 28 joins to the outermost in a 28 pin plug). This should cause the remaining pins to be forced against the link wires coming from the middle PCB. With

any luck they will be in the right place to be soldered but check first. Pin 12 (or 14 on 28-way plug) should lie next to R29 (1k $\Omega$ ). If all is correct then solder them into position and crop off any excess wire.

The assembly should now be complete and after inserting the mercury cells is ready for use. These cells should be inserted with the flat or outer part of the case nearest to the small PCB. This is the POSITIVE terminal. The rounded stud is negative and goes to the middle PCB. Make sure the batteries are pushed well in so that they locate against the pairs of supporting struts and do not touch each other.



The assembled project is neat and compact — at least it should be if you've done things correctly.

Otherwise one battery will be shorted out.

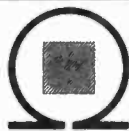
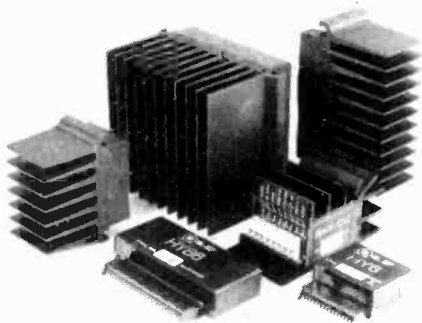
## Using the PseudoROM

It is recommended that the unit is only inserted or removed from its socket when the equipment using it is switched off. To make use of the unit the data must first be written into it in 2K blocks. For this a 24 pin socket must be provided with a R/W signal on pin 21 and other signals as for a standard 2K EPROM. SW1d should be set to connect the pin 21 input through to IC5b and SW1a and SW1b set to select the 2K segment to be used. Data can now be entered into the unit. Change the SW1a and SW1b settings to load up the other segments as required, remembering that SW1a is the MSB and SW1b the LSB of the address.

When placed in a suitable 24 or 28 pin ROM socket, SW1d should be set so that pin 21 (23) is no longer connected to IC5b: this prevents accidental overwriting of the data. SW1c can now be used to select either 2K ROM simulation (connecting to IC6 pin 6) or 4K in a 24 pin socket and 8K in a 28 pin socket (connecting to IC6 pin 1). Whichever mode is selected, the segment select addresses selected by SW1a and SW1b will be overridden by the relevant external address lines. This saves much extra switching but anomalies may occur if the 2K ROM mode is selected in a 28 pin socket.

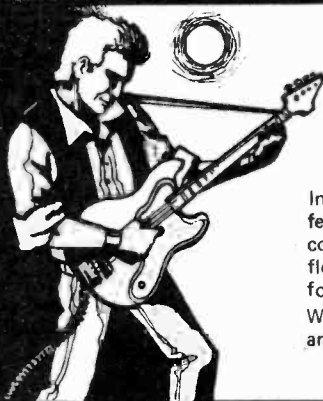
Note that this module is only suitable for use in EPROM sockets with single rail power supplies and the CE or CS signal on pin 18 (20 for a 28 pin socket) and the OE signal on pin 20 (22 for a 28 pin socket). Pin 21 (23 for a 28 pin socket) must be logic high level: A11 address or WE as appropriate for the mode used. Unfortunately this rules out some devices such as the three-rail TMS2716, the TMS2532, and under some circumstances the TMS2516. If in doubt, consult the data sheets for your particular device and circuit. **Warning:** DO NOT TRY TO PROGRAM THIS UNIT IN AN EPROM PROGRAMMER — the high voltage will damage it permanently. And take care when handling the device out of its socket — once assembled it's unlikely that any of the CMOS chips will be blown by static, but if it should happen, your guess is as good as ours how to replace them.

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HY60	30	4-8	0.015%	<0.006%	$\pm 25$	76 x 68 x 40	240	£9.55
HY6060	30 x 30	4-8	0.015%	<0.006%	$\pm 25$	120 x 78 x 40	420	£18.69
HY124	60	4	0.01%	<0.006%	$\pm 25$	120 x 78 x 40	410	£20.75
HY128	60	8	0.01%	<0.006%	$\pm 35$	120 x 78 x 40	410	£20.75
HY244	120	4	0.01%	<0.006%	$\pm 35$	120 x 78 x 50	520	£25.47
HY248	120	8	0.01%	<0.006%	$\pm 50$	120 x 78 x 50	520	£25.47
HY364	180	4	0.01%	<0.006%	$\pm 45$	120 x 78 x 100	1030	£38.41
HY368	180	8	0.01%	<0.006%	$\pm 60$	120 x 78 x 100	1030	£38.41

Protection: Full load line. Slew Rate: 15V/ $\mu$ s. Risettime: 5 $\mu$ s. S/N ratio: 100db. Frequency response (-3dB) 15Hz - 50KHz. Input sensitivity: 500mV rms. Input Impedance: 100K  $\Omega$ . Damping factor: 400Hz > 400.

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HY73	Guitar pre amp	Two Guitar (Bass Lead) and Mic + separate Volume Bass Treble + Mix	20mA	£15.36
HY78	Stereo pre amp	As HY66 less tone controls	20mA	£14.20

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#### MOSFET MODULES

Module Number	Output Power Watts rms	Load Impedance $\Omega$	DISTORTION T.H.D. Typ at 1KHz	I.M.D. 60Hz/7KHz 4:1	Supply Voltage Typ	Size mm	WT gms	Price inc. VAT
MOS128	60	4-8	<0.005%	<0.006%	1.45	120 x 78 x 40	420	£18.43
MOS248	120	4-8	<0.005%	<0.006%	1.55	120 x 78 x 50	520	£18.99
MOS364	180	4	<0.005%	<0.006%	1.55	120 x 78 x 100	1020	£36.33

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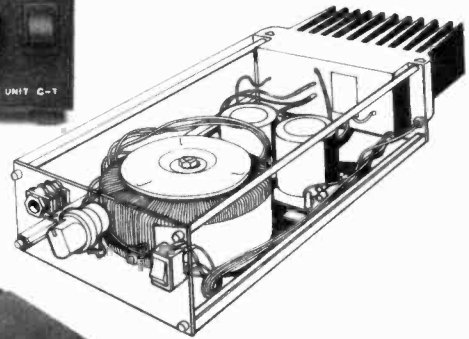
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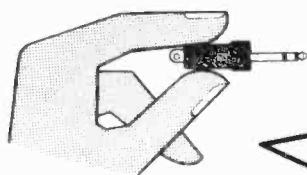
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# CONFIGURATIONS

Ian Sinclair has seen the light! Now he wants to illuminate the rest of us. In case you hadn't already guessed, this month's topic is opto-electronics.

Opto-electronics is a word that hadn't been thought of a few years ago, but which is now used to describe a set of devices that are important enough to merit a part of this series all to themselves. An opto-electronic device is one which makes use of light as part of its electronic function, so this label includes all varieties of devices that convert light signals into electrical signals or the other way round.

The simplest opto-electronic devices of the electricity-to-light type are the familiar LEDs. Familiar they may be, but even experienced engineers are not always aware of their eccentricities. Like any other diode the LED has an anode and a cathode, and passes current in the forward bias direction; this is when the light is emitted. What is not nearly so well known is that the peak reverse voltage of these diodes is very low; if you get an LED the wrong way round in a circuit, it's usually curtains for the LED when the voltage is switched on. A typical value of peak reverse voltage is 3 V, so practically any circuit that will operate the LED when it is connected the right way round (Fig. 1) will blow it up if it happens to be the wrong way round.

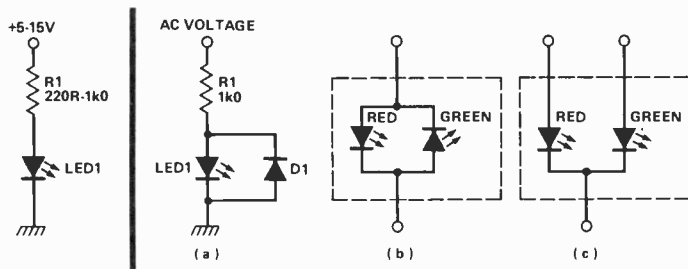


Fig. 1 (Right) The basic LED operating circuit. A current-limiting resistor must always be used unless the output resistance of the driving circuit is high.

Fig. 2 (Left) LED operation. (a) For use with AC, a silicon diode must be connected across the LED terminals as shown. (b) The two-colour LED uses two LED junctions connected in opposite directions. (c) The tri-colour LED uses separate LEDs with a common cathode connection.

In addition, the forward voltage across the LED is very much higher than the 0V6 that we merrily assume for a silicon diode. For gallium arsenide, the material used for many types of LEDs, the forward voltage is more like 2V1 to 2V4, so that LEDs are of little use in very low-voltage circuits — they won't, for example, work from a 1V5 cell.

## Current Affairs

One of the major snags about LEDs is that they consume a surprising amount of current. Manufacturers quote 'adequate' light output for red LEDs with 5 to 25 mA, and for the green/yellow varieties with 10 to 40 mA. This wouldn't be missed in a circuit operating at 5 V, 2.5 amps, but it can be quite a drain on battery equipment, often considerably more than all the CMOS ICs in a circuit intended for battery operation.

LEDs can be used with AC supplies providing there is a diode connected in reverse across each LED (to prevent excessive reverse voltage) as well as the usual current limiting resistor (Fig. 2a). Bi-colour LEDs consist of a package of two LEDs in one casing, connected in inverse parallel so that current in one direction will give a light of one colour, while the other colour is achieved by reversing the current (Fig. 2b). In this circuit, one LED protects the other against reverse voltage. Tri-colour indicators (Fig. 2c) use two diodes with a common cathode connection and separate anode leads, so that three colours can be indicated, one in each lead, plus yellow when both LED sections are activated. Personally, for indicating when mains voltage is on, I much prefer the old-fashioned neon.

## On Display

When it comes to digit displays, LED types have quite a lot of competition. The traditional seven-segment display (Fig. 3) comes as a common anode or a common-cathode type (Fig. 4), and each type needs a separate limiting resistor in each driver lead. The normal method of use is to connect the display to a decoder chip such as the 7448 or 7447, which in turn takes the digital information in as BCD signals — four bits per digit. The snag again is the current consumption, 10-20 mA per segment, which means that displaying a figure '8' uses 7 x 20 mA — 140 mA just to

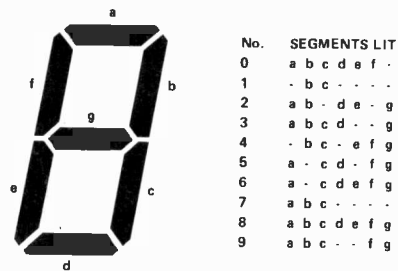


Fig. 3 Layout of the seven-segment display, with segment guide. An eighth segment, the decimal point, is often added.

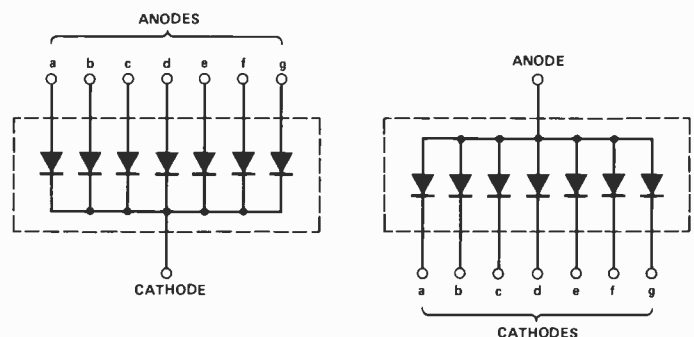


Fig. 4 Internal connections for common anode and common cathode displays. Whichever type is used, there must be a separate limiting resistor for each segment.

display one digit! While mains-powered equipment isn't too upset by this size of current, the LED seven-segment display did not last very long in battery-powered devices, even when multiplexing was used. Multiplexing means that only one digit at a time is activated, the digits being switched on in sequence fast enough to present the appearance of all the digits being illuminated at once.

Oddly enough, the forward voltage for the segments of an LED seven-segment display tends to be lower than for diodes, around 1V3 to 1V7. At temperatures above about 25°C, the maximum current has to be reduced by 0.3 mA per degree to avoid over-dissipation of the junction in each segment.

One competitive display that seems to be much less well-known is the filament seven-segment display. This can use as little as 5 mA per segment, and looks surprisingly bright — it can be driven by a decoder directly with no limiting resistors, and for many purposes is superior to LED displays. The usual reason for preferring solid-state displays is long life, but the quoted life of more than 100,000 hours for the filament type of display is pretty competitive, and some LED displays are notorious for short life — one frequent candidate for replacement in my experience is the display used in the old KIM microprocessor units.

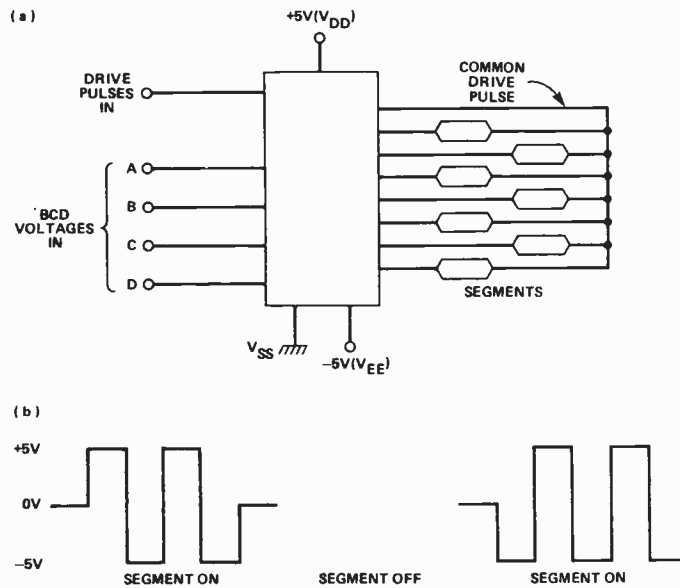
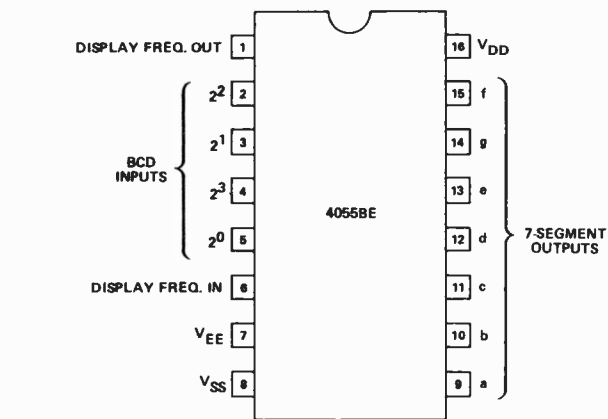


Fig. 5 Driving LCD displays. The common lead of the LCD display must not be earthed; it has to be returned to the driver IC. The waveform (b) applied is AC with no trace of DC.

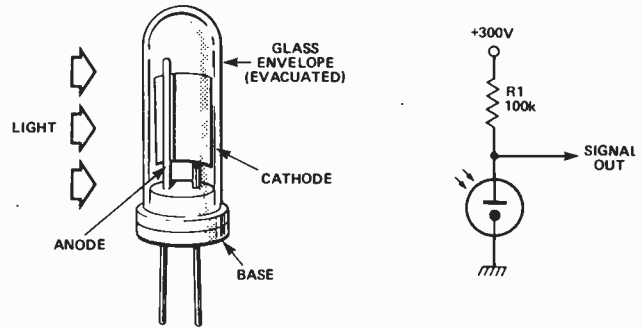


Fig. 6 The vacuum photocell, and a typical circuit arrangement.

### Oldies But Goodies

The two older types of displays which are also worth considering are the electron-beam type and the gas-discharge type. The electron beam display uses a miniature cathode wire to emit electrons, which will then be attracted to any positive anode. The anodes are coated with phosphors (similar to the phosphors used in cathode ray tubes), and any anode which is positive to the cathode by a sufficient voltage will glow. A 24-40 V supply is needed, which usually means the use of an inverter when low-voltage batteries are used, as in calculators. The display is easy to read, and uses less current than the LED type — I still prefer a calculator using this type of display to one using the more-common LCD display.

The gas-discharge display is an older type which uses the principle of the neon light — ionisation of a low-pressure gas in an electric field. Like all gas-discharge, this needs a high operating voltage, around 150-250 V, but the operating current is very low: only 0.7 mA per segment in a typical application. The display is very bright, and is worth considering for mains-operated equipment whose display has to be viewed under difficult illumination conditions, such as alternate brightness and darkness. A driver IC is available nowadays — in times past (dare we say the Dark Ages?), the major handicap of using this type of display was the lack of suitable driver transistors.

### Liquid Light

Last among the displays, of course, there is the LCD. A good LCD can give a dense black indication against a light grey background, is clearly visible in bright light, and reasonably visible even in low illumination conditions. There's a lot of variation between displays, however, even from the same manufacturer, and some are poor, with low contrast and very slow response to changing digits. Prices also vary considerably — one catalogue I have lists the price of a calculator-size display as being twice as much as I would have to pay for a complete calculator using a similar display!

Operating conditions for these displays are very different from those of other types of displays, because they have to be operated from high-frequency AC supplies. For this reason, displays either come with all the necessary circuitry for generating their driver pulses built in, or they can be used with a standard chip intended for this purpose. It's particularly important not to apply DC to the segments of an LCD display, because this can kill the display very rapidly.

### On The Receiving End

Moving to the other end of the opto-electronics business, we find the photocells. Vacuum photocells and

# FEATURE: Configurations

photomultipliers are rather specialised, and we'll only touch briefly on these types. They rely on photocathodes, surfaces which emit electrons into a vacuum when they are struck by light. The anode which collects the electrons (Fig. 6) must be at a fairly high voltage (100-500 V), and the currents are small: microamps rather than milliamps. Photomultipliers obtain greater sensitivity and increased output by using secondary multiplication, meaning that the electrons from the cathode (Fig. 7) are accelerated to surfaces, called dynodes, which will release electrons each time an electron strikes the surface. If each of these multipliers releases two to five electrons for each striking electron, spectacular gain can be achieved which, unlike amplification of signals by conventional methods, is practically noise-free.

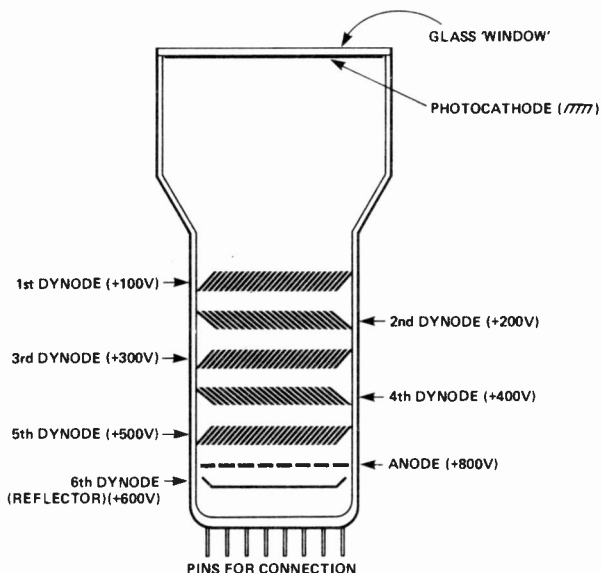


Fig. 7 Cross-section of a photomultiplier, used for detection of very low light levels.

The more familiar solid-state light-to-electrical-signal devices that we use are the solid-state photodetectors, of which the most commonly used is the cadmium sulphide cell. The ORP12 is the standard device of this type, often called an LDR (light dependent resistor). The cell consists of a strip of cadmium sulphide whose resistance decreases as light falls on it. The resistance in the dark is high, up to 10M, and the resistance can fall as low as 100R in bright sunlight. A less well-known aspect of these cells is that they can withstand a fairly high voltage, around 100 V; subject to their dissipation limit of 200 mW, meaning that you might need a limiting resistor connected in series. The cadmium sulphide cell is a slow-acting device, needing about 350 mS for the resistance to fall on exposure to light, and around 75 mS for the resistance to rise again when the light is shut off. The response to different colours is generally similar to that of the human eye, but the cadmium sulphide is much more sensitive to red and infrared, which is why its use in cameras is now less common than it was some 10 years ago.

## Fun With Photodiodes

Other light detectors need some degree of amplification. Photodiodes are diodes of fairly conventional construction, with a transparent window over the junction,

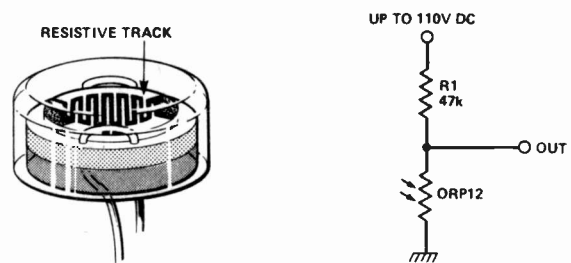


Fig. 8 The photoresistive cell or light-dependent resistor (LDR).

which are used reverse-biased. For such a diode, the reverse leakage current increases as the intensity of light on the junction is increased. This current is small, ranging from around 1 nA in darkness to almost 1 mA in very bright light, so that amplification is usually necessary, as in the circuit of Fig. 9. The response time is about 250 nS, so that the op-amps shown in Fig. 9 would have to be replaced by a transistor circuit, using high-speed switching transistors, if you wanted to use the photodiode for high-speed signals. Combined photodiode/op-amp packages can be bought for medium-speed applications.

The old-style phototransistor, which was a transistor formed with a window above the base-collector junction, is a thing of the past: what is now called a phototransistor is a combination of silicon photodiode and transistor in one package. This combines a sensitivity that is much greater than that of a photodiode alone with a good fast response time, giving typically a 2 MHz bandwidth. This is particularly useful for receiver use in light-beam transmission systems.

Last in our catalogue for the month come the optoisolators, which consist of a combination of LED and phototransistor in a single package. These components are embedded in clear plastic, which allows light transmission but which is a good electrical insulator. It's easy to achieve isolation to at least 4 kV, with reasonable signal transmission. For an ordinary isolator, the output signal will be about 20 per cent of the amplitude of the input, but when a Darlington phototransistor is used, the output can be three times or more the amplitude of the input. It's just the device I was looking for 25 years ago when I wanted to modulate the grid of a cathode-ray tube which was working at -4 kV!

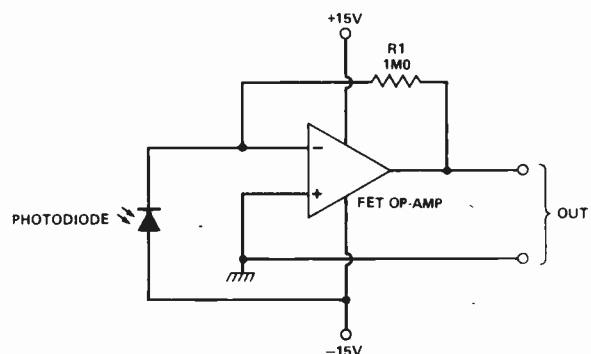
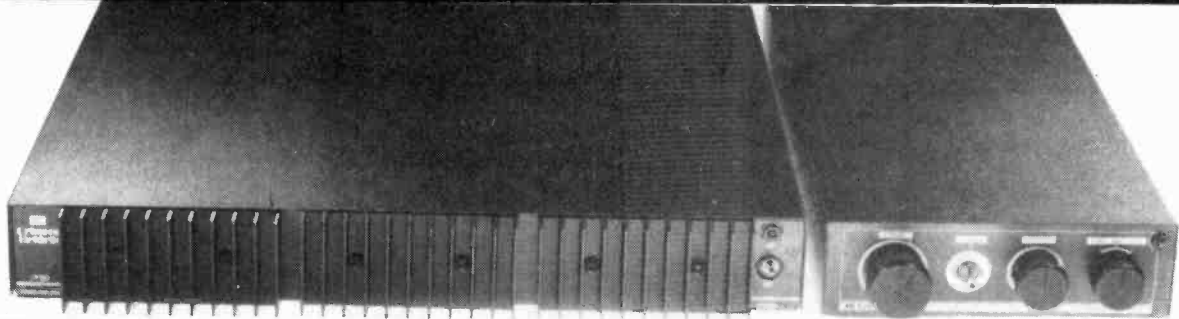


Fig. 9 Using a photodiode in conjunction with a FET op-amp. The FET type is needed because of the very high impedance of the photodiode circuit.

ETI

# The Reference Point for Kit Amplifiers from now on . . . . .



## HIFI STEREO AMPLIFIER KITS

From one of Britain's leading esoteric amplifier manufacturers comes an exciting new package of stereo amplifier kits, designed to offer all the advantages of true high fidelity but without the usual price penalty.

These new kits offer the choice of moving magnet or moving coil inputs, 40 to 100 watts per channel, in fact, everything that made the previous models so popular is included but with added style, easier construction and a full two year warranty.

The new range consists of The CK 1040 Stereo Pre Amplifier, The CK 1040 WPC Power Amplifier, The CK 1100 WPC Power Amplifier.

### CK 1010

This kit contains all the necessary parts to build a complete pre-amp. The main PCB is ready assembled and tested therefore construction is simply a matter of point to point wiring and mechanical assembly of the connections and controls to the pre punched chassis.

The CK 1010 takes its DC supply from the CK 1040, 1100 or, if using a different power amplifier a PSK power supply kit. Inputs for disc, tuner and tape are provided and an optional add-on moving coil input can be fitted to extend its versatility. (M.C.K)

### CK 1040

This is a nominal 40 watt per channel power amplifier kit which features our dual power supply and the DC output for the CK 1010. All components such as heatsinks, wire and connectors are included and protection is provided from short circuit outputs.

### CK 1100

Similar to the CK 1040 this model provides a nominal 100 watts per channel with extra heatsinking and thermal cutouts are provided as standard. When correctly assembled these kits are guaranteed for two years.

"It would seem then that Crimson have maintained their position at the top of the commercial kit-build field. There is no oriental amplifier I know of that can better the sound of this combination overall at any price and only a few - such as the KA-1000 (£500+) - are of comparable standard. . . . I can say no more than that for £250 it (CK 1010/MC2K/1100) is a bargain and one that becomes the reference point for kit amplifiers from now on."

**PRICES** CK 1010 - RRP £92.00; CK 1040 - RRP £121.00; CK 1100 - RRP £151.00; MC2K - RRP £25.00; PSK - RRP £20.00

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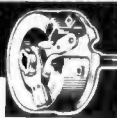


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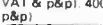
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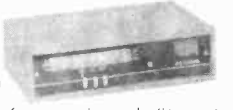
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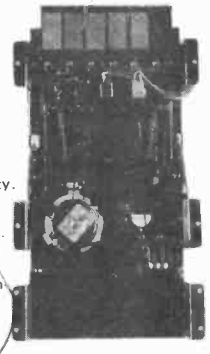
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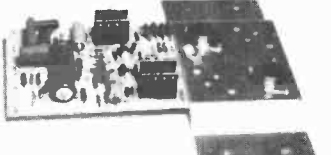


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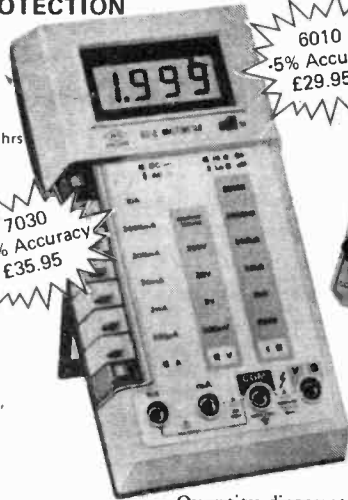
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# IMMERSIBLE HEATER

This simple-to-construct project has done sterling service as a temperature controller in a common fish tank heater but has wide application. Design by Jonathan Scott.

If this project did not run directly on 240 V we would call it a simple project. It contains only one IC, a triac and very little else, yet it is a full zero-crossing switch system capable of controlling up to 1500 W of heating power that may be employed to regulate the temperature of a room, a fish tank or a bowl of yoghurt-in-the-making.

At this point, let us stress that the PCB is operating at live mains potential as the IC is designed to run directly on the mains without transformer isolation, so be careful. You should never touch any of the components while the circuit is plugged in, nor try to adjust the setpoint preset pot.

The triac must be of 400 or more volts rating, and it must be of at least 4 A current rating as the PCB is designed to take the type which comes in a bolt-on package, and these start at 4 A. This over-rating, if you are only planning to have around 150 W of heater, as we did, allows you to dispense with a heatsink. There is not reason why you shouldn't have a higher rating still if the thing physically fits in place. With a heatsink, the type we found most common at the suppliers (and used) is rated at 6 A and will thus run up to 1500 W.

The thermistor you use will depend upon how much you wish to pay, how robust a component you need, and how you wish to connect it thermally to the load, as well as the temperature you are going for. We used both a cheap rod type, and a type G23. The latter is a small, but strong, glass bead type with a wide temperature range — but a £5-plus price tag! The G23 is a professional type, but reliable and predictable. As we point out later, a failure of the thermistor is not likely to cause anything but a cooling of the load, so unless you are willing to pay the price and need the reliability, use a cheap type. It is only necessary to have a resistance of 2k $\Omega$  to 60k $\Omega$  at the regulated temperature. We used a type having a resistance of 47k $\Omega$  at 25°C, costing only 40p or so.

Once you have obtained the thermistor, it is necessary to select R4 and PR1 to suit. These must together be able to equal the resistance of the thermistor at the temperature you are going for. Now, the G23 has a resistance of 2k $\Omega$  at 20°C and a temperature coefficient of so many ohms-per-degree. As we were going for 20-25 degrees in that model, we chose 820R for R4 and 5k $\Omega$  for PR1, ensuring that we could reach 2k $\Omega$  or a bit less. In general, if the thermistor is a cheap type, you had best measure its resistance while at the temperature you need. If it comes with specifications it will be possible to calculate the resistance at a given temperature.

So, having selected these components, you should assemble them on the PCB in accordance with the overlay diagram. Take care to get the polarity of the IC, capacitor and triac correct. Note that the metal tag of the triac is the MT2 connection, and the lead to the load is taken via a lug on the

bolt holding it to the PCB.

The final piece of constructional detail is the mounting of the thermistor. It is important to have it in close thermal contact with the item you are trying to heat. In our constructional example we are regulating the temperature of the liquid surrounding the tube within which the controller and the heater are immersed. It is thus only possible to regulate the tube temperature, as we cannot put the (live) sensor in the liquid. The sensor was pressed against the tube and seated in a blob of thermal compound of the type used for mounting transistors on their heatsinks. This meant that it was held at the temperature of the outside as much as possible, rather than at the temperature of the heater and controller themselves.

As the design of the housing is largely up to the individual constructor, there is very little to say about the physical makeup of the project. If you are copying the format shown in the pictures,

## THE ZERO CROSSING SWITCH TECHNIQUE

In normal phase control switching the switching element, an SCR or triac, is triggered into conduction at some time during each half cycle. The moment of switching is varied, so that the duration of the applied voltage pulse, which corresponds to the fraction of the cycle left at the point of triggering, is varied. This is a simple and direct method which does vary the applied power fairly smoothly (see Fig. 1). Unfortunately, the sudden application of voltage tends to produce a lot of electromagnetic radiation due to the sudden current change in the load circuit. This is responsible for a lot of radio frequency interference, or RFI.

In ZCS the switching element is only allowed to change to the conducting

state when the supply voltage is crossing the zero-voltage point — hence the name. This means that there are no sharp voltage transitions across the load, and so no RFI. The penalty paid is that only whole half cycles are applied to the load. The system is thus not readily applicable to lighting applications as the lights flicker badly due to the relatively long periods between conduction and isolation (see Fig. 2).

For applications where the system has sufficient inertia, such as heating, this system is by far the superior technique. As the required functions are available within a single, relatively cheap IC it is practicable to build a ZCS system with almost the same ease as a proportional system.

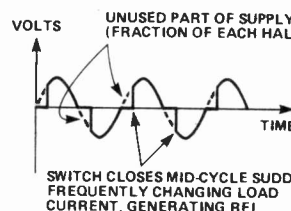


Fig. 1

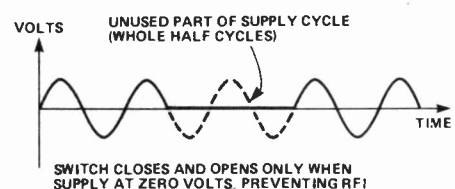


Fig. 2

## BUYLINES

This simple project uses very few parts, most of which are commonly stocked items. However, to choose a suitable thermistor it will be necessary to check in the mail order companies' catalogues; they don't bother putting these items in their magazine ads. Maplin stock the G23, plus a suitable selection of cheap rod thermistors. We got our fish tank heater in the local pet shop, but if you do this it's advisable to check the various types in the shop, making sure that you can get into it and that there's room for the PCB. The PCB can be obtained from the PCB Service using the order form on page 95.

hours of undisturbed operation and measure the liquid temperature with a thermometer. Adjust the preset pot (mains unplugged!) and repeat. If you measured the thermistor beforehand, while at the correct temperature, you can adjust the pot/R4 combination to give you that resistance straight away and there should be no need to make a second adjustment of the pot. If at some later stage you wish to change the set point, Silastic is easy to peel away and you can reseal the tube when the adjustment is made and checked.

We do not advise you to leave an access hole to permit adjustment of the pot because, firstly, someone might try and do that with the power on (poof!) or more likely the liquid will find a way of invading the tube and quietly ruin the components.

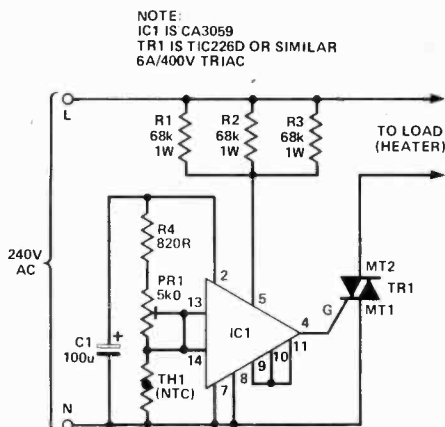


Fig. 3 The circuit. Simple, what?

assembly is rather straightforward. The only point to note is that you will have to take particular care to see that the tube is sealed against the accidental entry of liquid. Here, Silastic or a similar silicone sealant comes in very useful.

It is necessary to set the preset pot before sealing up the immersible tube. This is done by carefully setting the system up with water coming up to the point on the tube above where the sensor is located. Then, allow a couple of

## HOW IT WORKS

Most of the functions of this temperature controller are contained inside the IC, so let's take a look at the zero-voltage switch IC first.

Three zero-voltage switches are made by RCA — the CA3058, CA3059 and CA3079. They are all designed to control a thyristor in a variety of AC power switching applications for AC input voltages of 24, 230, 230 and 277 V at 50, 60 and 400 Hz. Each incorporates four functional blocks as follows (refer to the block diagram here):

- Limited-Power Supply — permits operation directly from an AC line.
- Directional On/Off Sensing Amplifier — tests the condition of external sensors or command signals. Hysteresis or proportional-control capability may easily be implemented in this section.
- Zero-Crossing Detector — synchronises the output pulses of the circuit at the time when the AC cycle is at zero voltage point; thereby eliminating radio-frequency interference (RFI) when used with resistive loads.
- Triac Gating Circuit — provides high-current pulses to the gate of the power controlling thyristor.

In addition, the CA3058 and CA3059 provide the following important auxiliary functions:

- A built-in protection circuit that may be actuated to remove drive from the triac if the sensor opens or shorts.
- Thyristor firing may be inhibited through the action of an internal diode gate connected to Terminal 1.
- High power DC comparator operation is provided by overriding the action of the zero-crossing detector. This is accomplished by connecting pin 12 to pin 7. Gate current to the thyristor is continuous when pin 13 is positive with respect to pin 9.

Because the CA3079 does not incorporate the built-in protection circuit, the CA3058 or CA3059 have been specified for this project. If the project is used to control a fish tank heater, one doesn't want to boil one's finny friends in the event of a thermistor failure!

Now we know what's inside the IC, how is it put to work in the circuit?

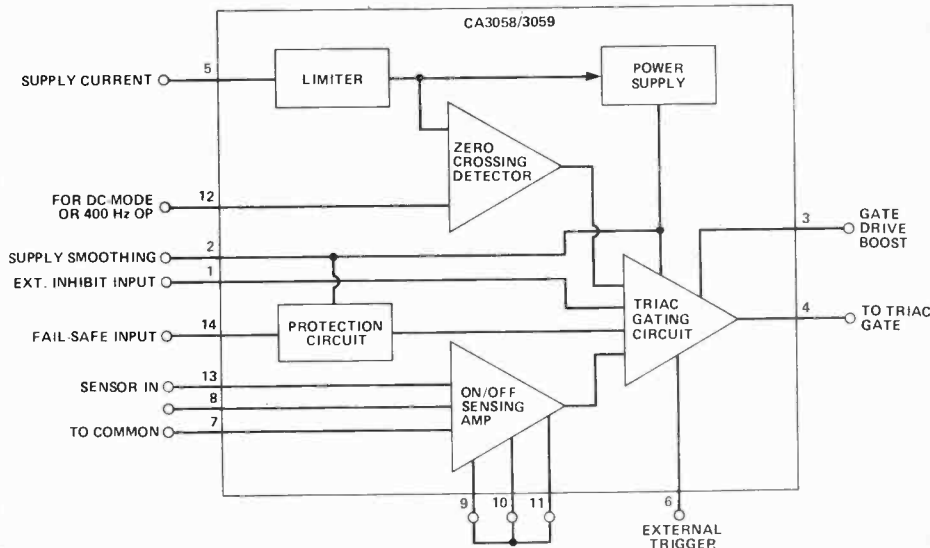
Initially, consider the triac to be turned off. Some current flows into pin 5 of the IC and this is limited by R1-3 and rectified within the IC to provide about 8 V DC for the operation of the circuit. Capacitor C1 smooths this supply. Inside the IC are a number of separate sub-circuits centered on a comparator ('ON/OFF SENSING AMP'). Connection of pins 9, 10 and 11 uses internal resistors to establish half supply rail (about 4 V) as one of the levels to be compared. When the voltage on pin 13 exceeds half rail potential the comparator activates a circuit which turns the triac on at the next supply zero, and each subsequent zero until the voltage falls below half rail.

Clearly then, PR1/R4 must be selected so that they add up to the resistance of the sensing thermistor at the temperature for which it is desired to regulate. Thus, when the temperature reaches the preset point, the voltage across TH1 corresponds to half rail potential on pin 13.

Pin 14 allows the protection circuit to detect when TH1 goes either open circuit or short circuit by looking at the voltage at the junction of R4 and TH1. If this voltage nears the DC supply rail or the local common (N), there has been a failure, and the firing of the triac is inhibited until the condition is removed.

The supply dropping resistors R1-3 are used instead of a single resistor purely for size considerations. All that is required is that they deliver 10 to 50 milliamps to the IC's rectifier-regulator.

The sensing thermistor must be a negative temperature coefficient type (NTC), as its resistance must drop with increasing temperature in order to reduce the voltage on pin 13 as the temperature is brought towards the set-point. There is sufficient excess supply current to allow it to draw at least one milliamp if necessary. Thus, any of the common small bead types with a few kilohms of resistance at the setpoint may be used. The total permissible sensor resistance range is 2k $\Omega$  to 100k $\Omega$ .



# PROJECT: Immersible Heater

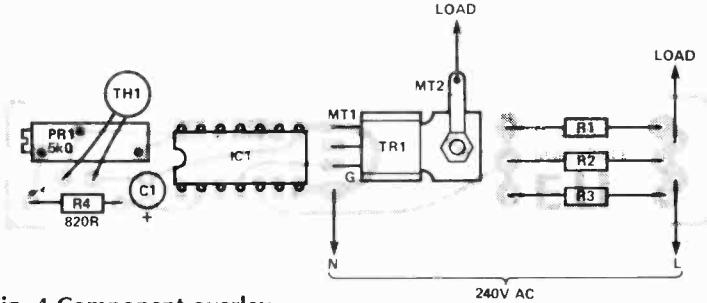
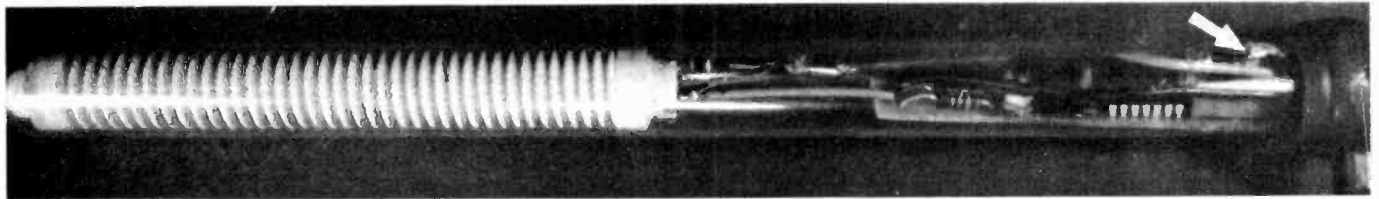


Fig. 4 Component overlay.



Naked and clothed. At top is the naked PCB (approximately life size) to compare with the component overlay. The lower picture shows the completed immersible temperature controller from a common fish tank heater. The arrow shows the positioning of the thermistor sensor.



ETI

## PARTS LIST

Resistors (all  $\frac{1}{2}W$ , 5% except where stated)  
**R1-3**          68k 1 W  
**R4**               820R (see text)

Potentiometer  
**PR1**            5k $\Omega$  multiturn cermet preset

Capacitor  
**C1**               100 $\mu$ F 16 V tantalum

Semiconductors  
**IC1**              CA3059  
**TR1**              6 A, 400 V triac eg TIC226D or similar

Miscellaneous  
**TH1**              thermistor to suit application eg G23 glass bead type (see text)

PCB (see Buylines): solder lug; hookup wire; heating element etc.

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748	35	LM13600	105	AY5-2376	590	LS04	12	LS158	27
9400CJ	345	MC1496	68	MC1488	55	LS05	12	LS160	30
AY-3-1270	710	MC3340	120	MC1489	55	LS08	12	LS161	35
AY-3-8910	370	MF10CN	350	MM5303	625	LS09	12	LS162	35
AY-3-8912	540	ML924	195	MM5307	1250	LS10	12	LS163	35
CA3048	60	NE529	225	MM58174	700	LS11	12	LS164	40
CA3080	65	NE531	135	TMS6011	365	LS12	12	LS165	50
CA3089	190	NE544	180	ULN2003	75	LS14	22	LS168	80
CA3090AQ	370	NE555	18	8T26	99	LS*5	12	LS170	70
CA3130E	85	NE556	45	8T28	120	LS20	12	LS173	47
CA3140E	38	NE565	110	8T95	90	LS21	12	LS174	36
CA3161E	100	NE566	140	8T97	80	LS22	12	LS175	36
CA3189	200	NE567	100	81LS95	80	LS27	14	LS181	87
CA3240E	110	NE570	370	81LS96	80	LS28	12	LS183	105
ICL7108	680	NE571	370	81LS97	80	LS30	12	LS190	35
ICL7811	95	RC4136	55	81LS98	85	LS32	13	LS191	35
ICL7821	180	RC4568	45	6522	310	LS37	14	LS192	35
ICL7822	180	SL490	250	6532	440	LS38	14	LS193	36
ICL8038	290	SL78477	380	6821	110	LS40	12	LS194	32
ICL8211A	150	SP8629	250	6845	650	LS42	28	LS195	43
ICM7224	775	TBA120S	70	6847	650	LS44	35	LS196	43
ICM7555	80	TBA800	75	6850	110	LS47	35	LS197	45
LF353	85	TBA810	95	6852	250	LS48	40	LS221	50
LF356	90	TBA820	70	6875	485	LS49	50	LS240	55
LM10	325	TBA950	220	8155	350	LS51	14	LS241	55
LM301A	24	TDA1008	310	8212	110	LS54	14	LS242	55
LM311	70	TDA1022	480	8216	100	LS55	14	LS243	55
LM318	120	TDA1024	115	8224	110	LS74	16	LS244	55
LM324	30	TL061	60	8226	250	LS76	16	LS245	70
LM334Z	90	TL062	60	8228	220	LS78	17	LS251	28
LM335Z	120	TL064	95	8243	270	LS83	33	LS257	30
LM339	45	TL071	25	8250	865	LS85	35	LS258	29
LM348	60	TL072	45	8251	250	LS86	19	LS259	32
LM358	55	TL074	95	8253	400	LS90	22	LS266	18
LM377	165	TL081	24	8255	225	LS92	25	LS273	53
LM380	65	TL082	45	8257	400	LS93	22	LS279	30
LM381	120	TL084	90	8259	395	LS95	36	LS283	38
LM382	110	TL170	49	8279	385	LS109	21	LS290	40
LM384	130	UA2240	115	8832	250	LS112	20	LS293	40
LM386	65	ULN2003	75	9602	220	LS113	20	LS365	27
LM387	120	ULN2004	75	Z80ACTC	260	LS114	21	LS366	27
LM393	95	XR206	285	Z80ADART	775	LS123	34	LS367	27
LM711	60	ZN414	79	Z80ADMA	975	LS125	24	LS368	27
LM725	325	ZN423	130	Z80AP10	270	LS126	25		
LM733	69	ZN424	130	ZN425E8	320	LS132	29	<b>MICRO-MINI 100V CERAM PLATE CAPS</b>	
LM747	60	ZN425E	340	ZN426E8	320	LS136	23	1pF 10nF	7
LM1458	40	ZN426E	290	ZN427E8	575	LS139	24		
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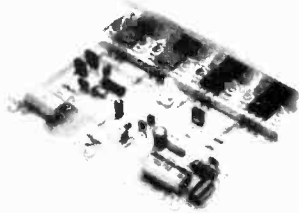
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 Gain  $\times$  23  
 Rin 30K  
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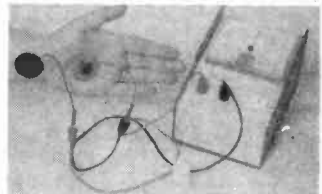
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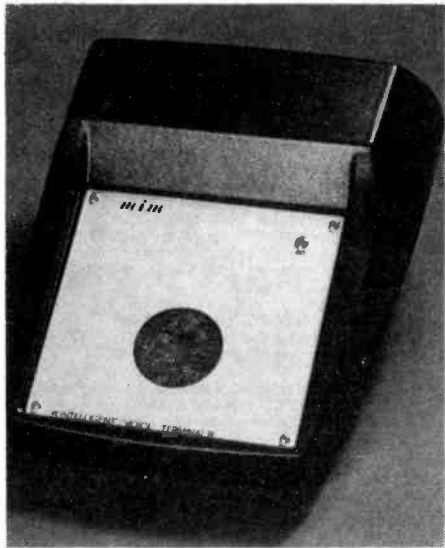
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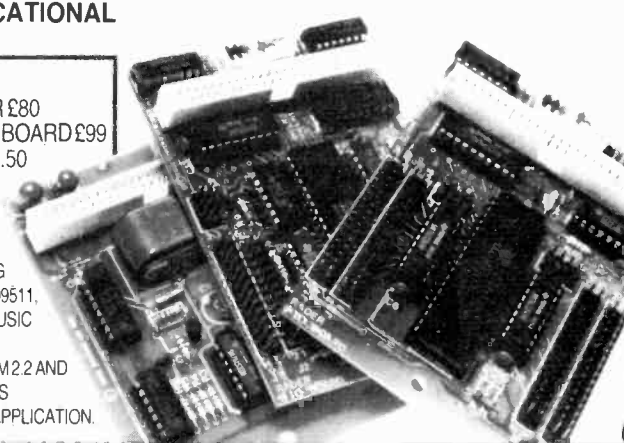
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# DATA SHEET

## M108/M208 SINGLE CHIP ORGAN

Given the great popularity of the Victory organ that we've been featuring in ETI since February, it isn't surprising that there have been many enquiries about further information on the SGS-Ates chips that form the heart of the project. This Data Sheet should clear up most of the questions.

The M108/M208 is a single chip organ featuring solo and/or accompaniment modes. It accepts 61 keys arranged in a 12 x 6 matrix; a scanning cycle takes 576 microseconds and all keys pressed are accepted. There are two keyboard formats: either 61 keys (solo) or 24 plus 37 (M108), 17 plus 44 (M208) keys (accompaniment and solo) with the possibility of automatic chords of the 'accompaniment' section. There are internal anti-bounce circuits.

A top octave synthesizer is incorporated for the generation of three 'footages'. More than one chip can be employed with synchronisation through the reset input. There are separate analogue outputs (for each foot) for solo, accompaniment and bass sections (square wave of 50% duty cycle) with the average value constant. Each section also has 'key down' and 'trigger' outputs. Sustain is provided for the last keys released in the solo section.

There are several choices of operating mode in the accompaniment section:

- Manual, with or without memorisation of the selected keys (free chords with alternate bass),
- Automatic, with or without memorisation of the selected key (priority to the left for automatic

chords and bass arpeggio). When in automatic mode there are again several possibilities:

- Major or minor third
- With or without seventh.

The chip operates from a standard single supply of +12 V  $\pm 5\%$ , with a low dissipation of less than 600 mW. All inputs are protected from electrostatic discharges.

### General Characteristics

The characteristics of the M208 are similar to those of the M108; the only difference is the keyboard split, which is 24+37 for the M108 and 17+44 for the M208 when used in 'accompaniment + solo' mode.

The circuit comprises:

- Two pins for clock input; one for the matrix scanning, the other for the incorporated Top Octave Synthesizer (TOS); by connecting both the clock inputs to the same matrix scanning clock (1000.12 kHz), the three footages generated are 16', 8' and 4'.

- Six inputs from the octave bars (keyboard and control scanning).

- Three multiplexed data inputs for addressing the bass selection. These inputs normally come from the outputs of an external logic memory (negative or positive logic with control inside the chip).

- Eight signal outputs divided by section: three for the solo section (16', 8', 4'), four for the accompaniment section (16' or root, 8' or 3rd, 4' or 5th, 8th/7th according to operating mode), one for the bass.

- Twelve outputs for the matrix scanning.

- Five trigger and key down outputs:  $\overline{KPS}$  (key pressed solo),  $\overline{TDS}$  (trigger decay solo),  $\overline{KPA}$  (key pressed accompaniment),  $\overline{NPA}$  (pitch present in accompaniment outputs),  $\overline{TDB}$  (trigger decay bass) respectively. These outputs, in conjunction with an external time constant, allow the formation of the envelope of the sustain and percussion effects. The duration of the trigger pulses is approximately 9 milliseconds

- One input (reset) to synchronize the device or more than one device (with the same keyboard scanning and using a single contact per key). The reset action, provided by an

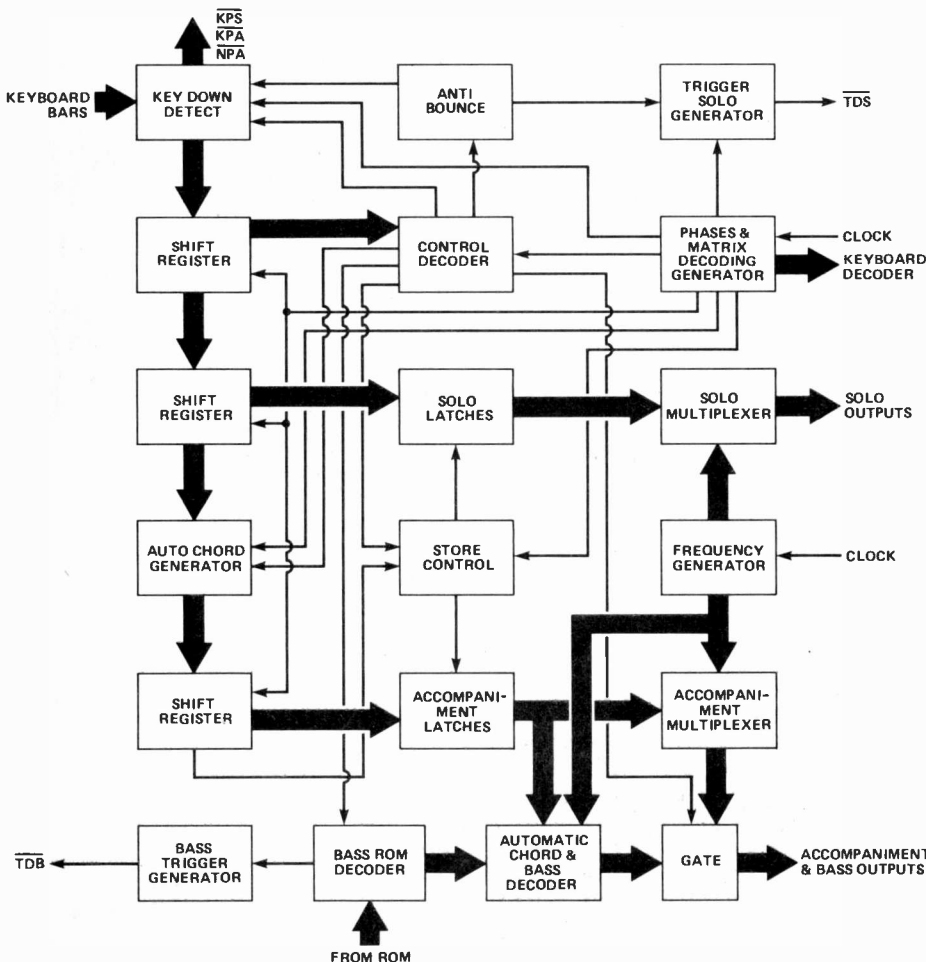


Fig. 1 Block diagram of the IC.

external circuit, is of the power-on reset (high active) type and its duration must be approximately 0.5 milliseconds.

- One TEST pin (in use it must be connected to  $V_{DD}$ ).
- Two supply pins.

### Features

The main feature of this chip is the possibility of forming the keyboard either with 61 keys (only 'solo' without automatism) or separating it into two sections ('accompaniment + solo') with the possibility of chord and bass automatic in the first section.

- The '61/24 + 37' (17 + 44) control chooses the keyboard operating mode, ie the whole keyboard dedicated to 'solo' or 24 (17) keys dedicated to 'accompaniment' and 37 (44) to 'solo'.

- The 'Man/Auto' control, which operates only in case of 'accompaniment and solo', chooses the manual or the automatic accompaniment.

- The 'Sust OFF/Sust ON' allows the storage of the 'solo' section and handles the whole keyboard or 37 (44) keys depending on the operating mode.

- The 'Latch/Latch' similarly allows the storage of the 'accompaniment + solo' only.

- The '3rd+/3rd-' which operates only in case of 'accompaniment + solo' and 'automatic' changes the automatic chord generated from major to minor or vice versa.

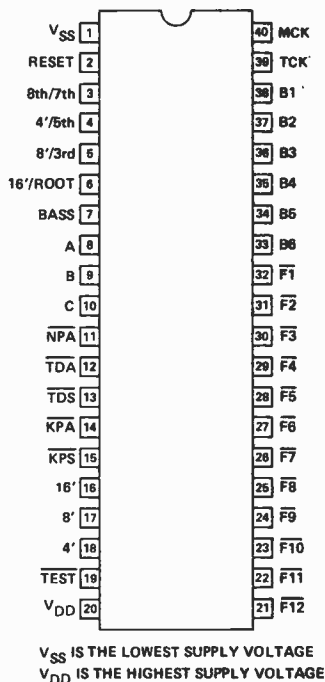


Fig. 2 Pin connections for the M108/M208.

## MATRIX ORGANISATION (KEYBOARD AND CONTROLS)

M108/208 Matrix outputs	M108/208 Octave bar inputs					
	B1	B2	B3	B4	B5	B6
$\overline{F1}$	C <sub>1</sub>	C <sub>2</sub>	C <sub>3</sub>	C <sub>4</sub>	C <sub>5</sub>	C <sub>6</sub>
$\overline{F2}$	C <sub>1</sub> #	C <sub>2</sub> #	C <sub>3</sub> #	C <sub>4</sub> #	C <sub>5</sub> #	7th OFF/7th ON
$\overline{F3}$	D <sub>1</sub>	D <sub>2</sub>	D <sub>3</sub>	D <sub>4</sub>	D <sub>5</sub>	3rd+/3rd-
$\overline{F4}$	D <sub>1</sub> #	D <sub>2</sub> #	D <sub>3</sub> #	D <sub>4</sub> #	D <sub>5</sub> #	Sust. OFF/Sust. ON
$\overline{F5}$	E <sub>1</sub>	E <sub>2</sub>	E <sub>3</sub>	E <sub>4</sub>	E <sub>5</sub>	Latch/Latch
$\overline{F6}$	F <sub>1</sub>	F <sub>2</sub>	F <sub>3</sub>	F <sub>4</sub>	F <sub>5</sub>	Man/Auto
$\overline{F7}$	F <sub>1</sub> #	F <sub>2</sub> #	F <sub>3</sub> #	F <sub>4</sub> #	F <sub>5</sub> #	61/24 + 37 (17 + 44)
$\overline{F8}$	G <sub>1</sub>	G <sub>2</sub>	G <sub>3</sub>	G <sub>4</sub>	G <sub>5</sub>	Antibounce ON/Antibounce OFF
$\overline{F9}$	G <sub>1</sub> #	G <sub>2</sub> #	G <sub>3</sub> #	G <sub>4</sub> #	G <sub>5</sub> #	ROM Low/ROM High
$\overline{F10}$	A <sub>1</sub>	A <sub>2</sub>	A <sub>3</sub>	A <sub>4</sub>	A <sub>5</sub>	
$\overline{F11}$	A <sub>1</sub> #	A <sub>2</sub> #	A <sub>3</sub> #	A <sub>4</sub> #	A <sub>5</sub> #	
$\overline{F12}$	B <sub>1</sub>	B <sub>2</sub>	B <sub>3</sub>	B <sub>4</sub>	B <sub>5</sub>	

C<sub>1</sub> is the first key on the left, C<sub>6</sub> is the last key on the right of the keyboard.

- The '7th OFF/7th ON' adds the seventh to the automatic chord generated.

- The 'Antibounce ON/Antibounce OFF' disables the antibounce circuit which is usually enabled.

- The 'ROM Low/ROM High' selects between ROMs with return to '1' (Low active) or with return to '0' (High active). Usually the chip is enabled for ROMs with return to '1' (Low active).

### 'Solo' Operation

In this case the chip recognizes the whole keyboard as 'solo' and does not read the controls which concern the 'accompaniment + solo' operation. The chip identifies all the keys pressed and transfers to the outputs of each section the analogue sum of corresponding pitches. The outputs are current generators with average value constant, therefore it is sufficient to connect the pins to one load and send the signals on to the filters.

In the case of 'Sustain OFF' each new key pressed or released is accepted or deleted in a time less than 576 microseconds. In the case of 'Sustain ON' the chip has a different operation according to whether the new key (keys) is pressed or released: each new key pressed is always accepted in a time less than 576 microseconds, whereas each key released is deleted with a delay of 73 milliseconds and only if there are still keys pressed. In fact, if after the 73 milliseconds there are no keys pressed, the last key (or keys) released remains stored until new keys are pressed. In this mode it is possible to have Sustain, with external envelope shaping, for the last keys (or key) released. The pitch

envelope is controlled by a DC signal KPS (any key pressed), and there is also an AC signal TDS (trigger decay solo) which provides a pulse whenever a key is pressed. An appropriate antibounce circuit, inside the chip, solves the problems associated with the keyboard contacts.

### 'Solo + Accompaniment' Operation

In this case the chip identifies the 'accompaniment' on the first 24 (17) keys on the left, and the 'solo' on the remaining 37 (44) keys and reads all the controls which concern the 'accompaniment' section. The 'solo' function is identical to '61 keys' mode, but for the 'accompaniment' section there are two possibilities:

#### Manual

The chip identifies which keys are pressed in the 'accompaniment' section, and transfers to the 'accompaniment' outputs the analogue sum of the corresponding pitches. The 'accompaniment' section is fully independent of the 'solo' section and the signals (if there is no 'latch') remain at the output only while the keys are pressed even if there is 'sustain on'.

The 'bass' section gives at the bass output an alternating bass bet-

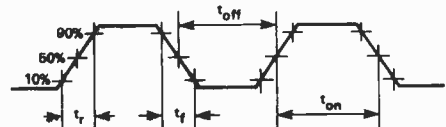


Fig. 3 The input clock waveform.



ween the first on the left and the first on the right of the keys pressed in the 'accompaniment' section; the pitch switching timing is dependent on an external ROM (three bits). The 'accompaniment' control stores the last keys released and the output signals, including the bass output, remain until new keys are pressed. The TDB (trigger decay bass) output gives a pulse corresponding to every output change; there are also two DC signals, KPA (any key pressed accompaniment) and NPA (pitches in output accompaniment) relative only to the 'accompaniment' section. The first of these signals (analogous to KPS) concerns the keyboard and does not consider the 'latch' condition. The second on the contrary concerns the 'accompaniment' output and considers the 'latch' condition.

## Automatic

The chip recognizes in the 'accompaniment' section only the first on the left of the keys pressed and, according to the setting of the following controls, produces a major or minor chord with or without seventh only the 4' footage but with separated outputs for root, third, fifth and eighth (or seventh if the chord is with seventh).

The bass section gives the bass arpeggio among root, third, fourth, fifth, sixth, seventh and eighth with pitch switching dependent on an external ROM (3 bits). In automatic mode the two octaves of the 'accompaniment' section inside the chip are connected in parallel both for the chord and for the bass; therefore by pressing any one of the two keys of the same note the chip generates the same chord.

The 'latch' control stores the major chord and the bass pitches (until new keys are pressed); the modification of the chord stored (from major to minor, addition of seventh) is always possible by operating the proper controls: by releasing these controls the chord becomes major again. It is possible to delete the stored pitches both in manual and in 'automatic' mode by a latch control signal.

Once again there are  $\overline{KPA}$ ,  $\overline{NPA}$ , and  $\overline{TDB}$  information; however the TDB pulse, which normally appears at each arrival of the ROM codes, does not appear if there are no pitches in the 'accompaniment' (and bass) outputs or, in the case of alternate bass (in manual mode) if the codes indicate conditions of indifference.

Parameter	Test conditions	Min.	Typ.	Max.	Unit
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## RECOMMENDED OPERATING CONDITIONS

$V_{DD}$	Highest supply voltage	11.4	12	12.6	V
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**STATIC ELECTRICAL CHARACTERISTICS** (Positive logic,  $V_{DD} = +10$  to  $+14$  V,  $V_{SS} = 0$  V,  $T_{amb} = 0$  to  $50^\circ\text{C}$  unless otherwise specified)

## INPUT SIGNALS

$V_{IH}$	Input high voltage	Note 1	$V_{DD}-1$		$V_{DD}$	V
		Note 2	4		18	V
		Note 3	$V_{DD}-2$		$V_{DD}$	V
$V_{IL}$	Input low voltage	Note 1	$V_{SS}$		$V_{SS}+1$	V
		Note 2	$V_{SS}$		$V_{SS}+0.6$	V
		Note 3	$V_{SS}$		$V_{SS}+2$	V
$I_{LI}$	Input leakage current	$V_I = 14$ V $T_{amb} = 25^\circ\text{C}$			10	$\mu\text{A}$

## LOGIC SIGNAL OUTPUTS

$R_{ON}$	Output resistance with respect to $V_{SS}$			300	500	ohms
$R_{ON}$	Output resistance with respect to $V_{DD}$	$V_{OUT} = V_{DD}-1$ (driver off)		15	25	kilohms
$V_{OH}$	Output high voltage		$V_{DD}-0.4$		$V_{DD}$	V
$V_{OL}$	Output low voltage			$V_{SS}+0.2$	$V_{SS}+0.4$	V

## POWER DISSIPATION

$I_{DD}$	Supply current	$T_{amb} = 25^\circ\text{C}$		30	45	mA
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## ANALOGUE SIGNAL OUTPUTS (the external load must be connected to $V_{DD}/2$ )

$I_{OH}$	Output current with respect to $V_{DD}/2$	Outputs loaded with 1k $\Omega$ resistor versus $V_{DD}/2$	35	50	70	$\mu\text{A}$
$I_{OL}$	Output current with respect to $V_{SS}$	Outputs loaded with 1k $\Omega$ resistor versus $V_{DD}/2$	-35	-50	-70	$\mu\text{A}$
Note 1: Refers only to the clock inputs						
Note 2: Refers only to the inputs from the external memory						
Note 3: Refers only to the reset input.						

## DYNAMIC ELECTRICAL CHARACTERISTICS

### MASTER CLOCK INPUT

$f_i$	Input clock frequency			1000.12		kHz
$t_r, t_f$	Input clock rise and fall time 10% to 90%	1000.12 kHz			40	nS
$t_{on}, t_{off}$	Input clock ON and OFF times	1000 kHz		500		nS

### TOP OCTAVE SYNTHESISER CLOCK INPUT

$f_i$	Input clock frequency		100	1000.12	2500	kHz
$t_r, t_f$	Input clock rise and fall times 10% to 90%	1000.12 kHz			40	nS
$t_{on}, t_{off}$	Input clock ON and OFF times	2000 kHz		250		nS

### TDS and TDB OUTPUTS

$t_{on}$	Pulse duration	1000 kHz		9.216		mS
$t_r, t_f$	Outputs rise and fall times 10% to 90%	1000 kHz		100		nS

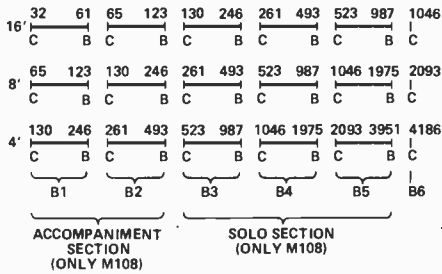


Fig. 4 The frequency range of each octave (16', 8', 4' footages).

## ABSOLUTE MAXIMUM RATINGS

$V_{DD}$	Source supply voltage with respect to $V_{SS}$ (GND) pin voltage	-0.3 to +20 V
$V_i$	Input voltage with respect to $V_{SS}$	-0.3 to +20 V
$I_O$	Output current at any pin	3 mA
$T_{op}$	Operating temperature	0 to 50°C

Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other condition above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

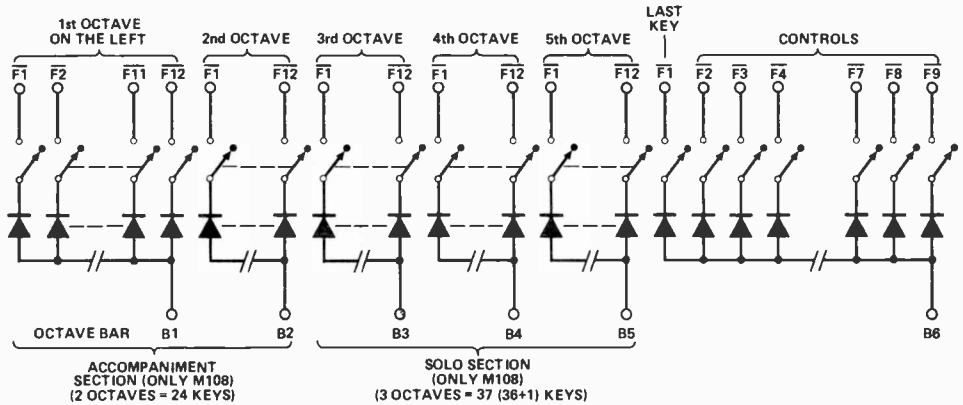


Fig. 5 Connection of the keyboard and control switches.

NOTE: THE SWITCH 'OPEN' CORRESPONDS TO 'KEY NOT PRESSED' OR 'CONTROL IN THE FIRST CONDITION' (SEE THE TABLE 'MATRIX ORGANISATION')

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0	0	1	7th	---
0	0	0	8th	---

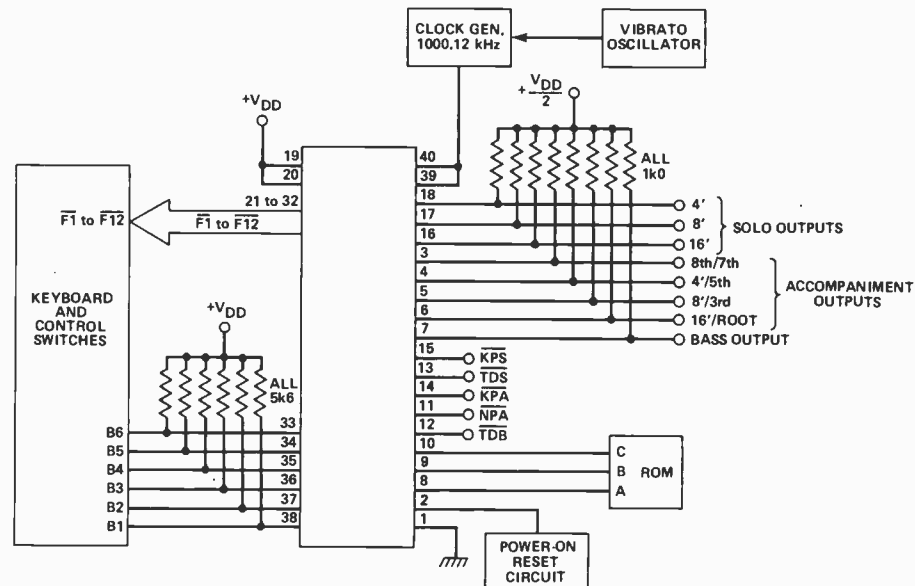


Fig. 6 A typical application.



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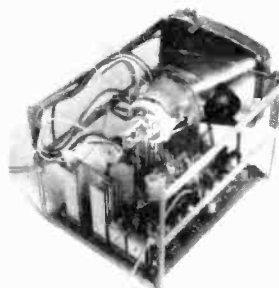
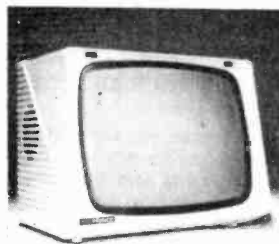
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Although these articles are being prepared for the next issue, circumstances may alter the final content.

# TOROIDALS

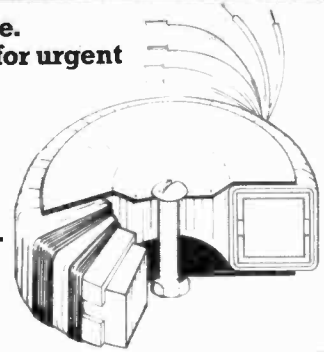
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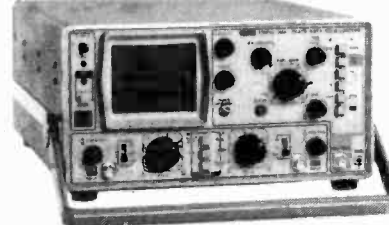
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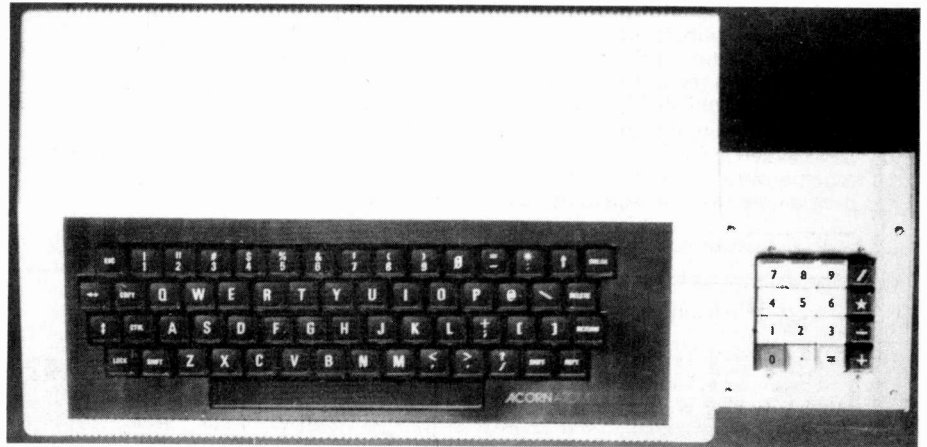
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# ATOM CALCULATOR PAD



And now for some Atomic fusion. Nothing to do with nuclear reactors (that's still at the breadboarding stage): this is an add-on numeric keypad. Design by Patrick Squire.

The addition of a numeric pad to the standard QWERTY keyboard greatly increases the ease of use of a microcomputer, both for programs containing extensive numerical statements and for joystick operations, where the layout of the numeric keys corresponds directly to the direction of movement of an object being controlled on the VDU. Anyone accustomed to using machines both with and without numeric pads will appreciate the advantages of the extra keys. An additional drawback of the QWERTY keyboard is that the +, \* and = operations are obtained by shifted keys. This can be very frustrating and lead to errors when entering programs.

For all these reasons it is desirable to be able to add a basic calculator pad, containing the numbers 0-9, the decimal point, equals sign and the four arithmetic operators, to machines not originally supplied with such a facility. Unfortunately such an accessory is not generally available for most micros. This article describes a calculator pad designed specifically for the Acorn Atom, although the general principles employed are capable of straightforward modification to other machines which use a similar system of scanning the key matrix. The whole unit can be constructed for under £20, but it does involve soldering to the masterboard of the Atom, so construction should not be undertaken by the fainthearted.

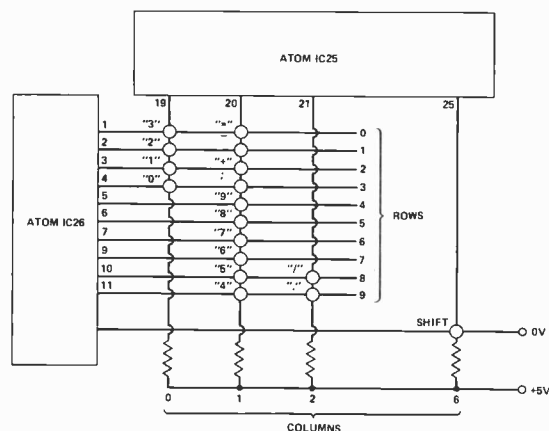


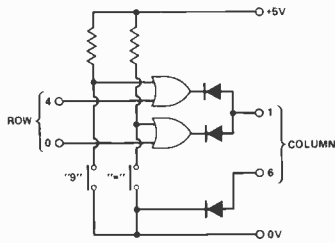
Fig. 1 Part of the Atom keyboard matrix. Each row is connected to one or more columns by a single keyboard switch. The rows are driven by the 3-to-10 decoder IC26. The columns are read by the PIA IC25.

## Matrix Scanning

In order to understand the operation of the unit described here, or in order to modify it for other machines or to incorporate alternative functions, it is essential to understand how the Atom reads a depressed key. Figure 1 shows the appropriate section of the Atom circuit. It can be seen that each character key connects a particular row (numbered 0 to 9) with a particular column (numbered 0 to 6). For example, key '9' connects row 4 with column 1. The SHIFT key works in a slightly different way, so I will first describe the operation of reading an unshifted key.

The Atom operating system includes a machine code routine

that scans the keyboard. Normally all the rows 0-9 are held at logic high. The columns are connected by pull-up resistors to the + 5 V rail, so they are also logic high. In this state the depression of a character key has no effect, since it merely connects two lines both in the same state. When the scan commences, each row is successively driven low by IC26. The routine then interrogates each column to see if it is low. Consider the state when row 4 has been set low. If no key has been pressed the columns will all be high. If key 9 is pressed, however, column 1 will be connected to row 4, which is low. Column 1 will thus be pulled low and the Atom will know from the



**Fig. 2** The principle of the keyboard extension. The unshifted key '9' drives column 1 low when row 4 is also low. The shifted key '=' drives column 1 low when row 1 is low; it also pulls column 6 low to give the shift operation.

information stored in ROM that the combination 'row 4 low and column 1 low' means that key '9' has been pressed. A similar process works for all the unshifted keys.

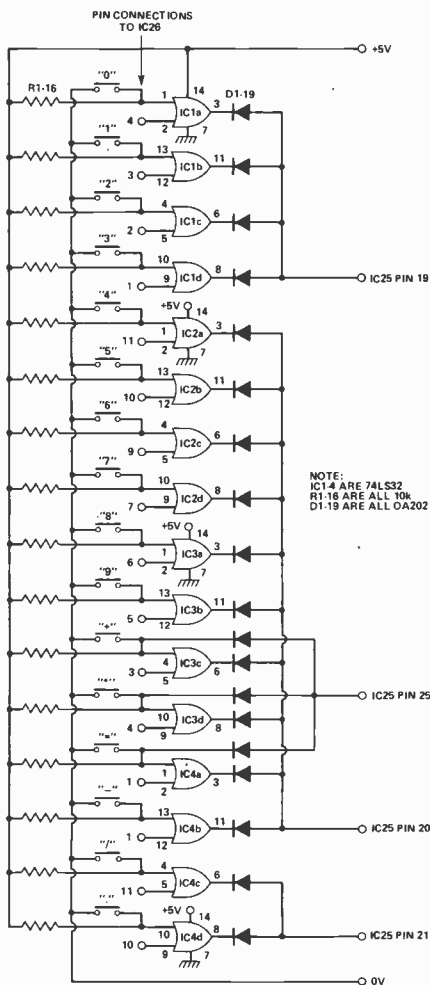
For the shifted keys use is made of column 6. Figure 1 shows that this is normally high, but that if the SHIFT key is pressed it pulls the line low by direct connection to the 0 V line. The key scan routine therefore also interrogates column 6 and, according to its state, decides whether a dual-role key, such as '=/\_' is to be interpreted as shifted or unshifted.

## Circuit Principles

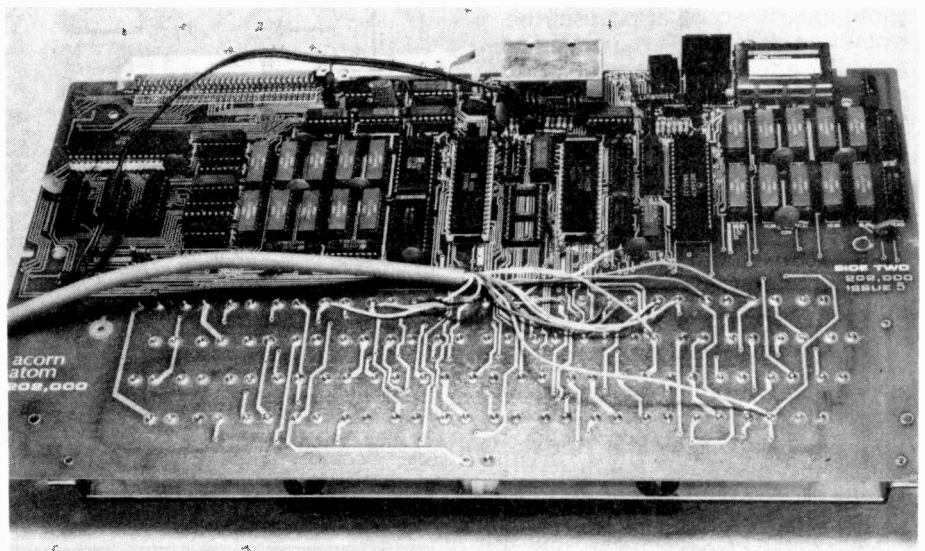
The direct way of extending the ATOM keyboard would be to parallel each keyboard switch by a corresponding key on the pad.

However, this would involve having 16 unconnected switches and 32 connecting wires. This would be a very inefficient system and would not be able to cope with shifted characters as single-key operations. The system employed in this design employs an OR gate for each key, and is illustrated in Fig. 2 for two keys, one the unshifted '9', the other the shifted '='.

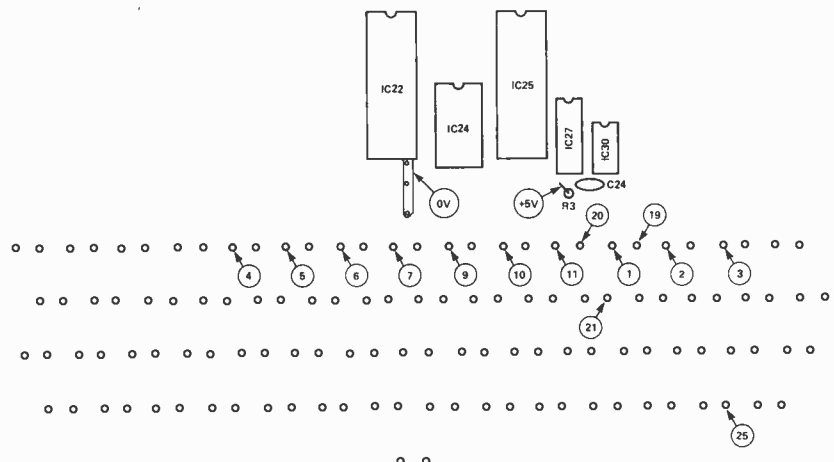
Consider first the unshifted '9' key. When the Atom drives pin 5 of IC26 low (this corresponds to row 4 — see Fig. 1) one input of the upper OR gate is driven low. So long as key '9' remains open, however, the other input of this gate remains high. The OR action then maintains the output high, so pin 20 of IC25 (column 1) remains high. If, on the other hand, key '9' is closed both



**Fig. 3** The complete circuit of the keypad extension.



Wiring up to the Atom — the diagram below refers.



**Fig. 4** A diagram of the Atom PCB (rearside of the keyboard) showing take-off points for the wiring. The numbers refer to the pins on IC25 and IC26 (as Fig. 1 shows, this numbering is unambiguous), and correspond to the numbers on the PCB.

## BUYLINES

Nothing at all unusual in this project and most suppliers should have the components in stock. One source for the case we used is BI-PAK Semiconductors, PO Box 6, 63a High Street, Ware, Herts SG12 9AG (telephone 0920 3182/3442): they also stock the semiconductors and resistors. The order number for the case is 148. The PCB can be ordered using our PCB Service order form on page 95.

## PARTS LIST

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R1-16 10k

Semiconductors  
IC1-4 74LS32  
D1-19 OA202

Miscellaneous  
PCB (see Buylines); 16-switch keypad matrix (see text);  $\frac{1}{2}$  meter 15-core plus screen cable; case to suit.

inputs to the OR gate are low, and the output is low. The fourth possibility (pin 5 of IC26 high, '9' low) also results in a high output, so pressing '9' will not have any effect unless row 4 is also low, as required. The purpose of the diodes is to isolate the outputs of the various OR gates. In the absence of the diodes, when one gate tried to go low it would be pulled up by all the other gates connected to the same line. It would also upset the operation of IC26. The diodes have a similar effect to open-collector outputs. However, open-collector OR gates are not generally available.

Now consider the operation of the shifted '=' key. The logic of the lower OR gate is identical to that just described, but in addition a diode connection to pin 25 of IC25 pulls column 6 low, thus simulating the effect of the combined operation of the SHIFT and '= / -' keys on the Atom keyboard. The complete circuit is shown in Fig. 3. Note that for 16 keys, the 16 OR gates conveniently occupy four quad OR gate ICs (type 74LS32).

## Construction

The most expensive item is the keyboard itself. It should contain 16 single-poll switches with one side commoned to the 0 V rail. The author used a very cheap but entirely satisfactory unit from Watford Electronics Ltd. Perusal of the advertising columns should reveal a source of a suitable alternative for less than £10. Avoid decoded types, which are connected in a matrix fashion unsuitable for this application. As a last resort you could make your own from the readily available keypad switches.

The circuit should be constructed on a printed circuit board. Veroboard can be used, but is much more difficult to wire up and looks messy with so many connections. We've designed a double-sided PCB for this project — with care this shouldn't cause grave difficulties to home constructors,

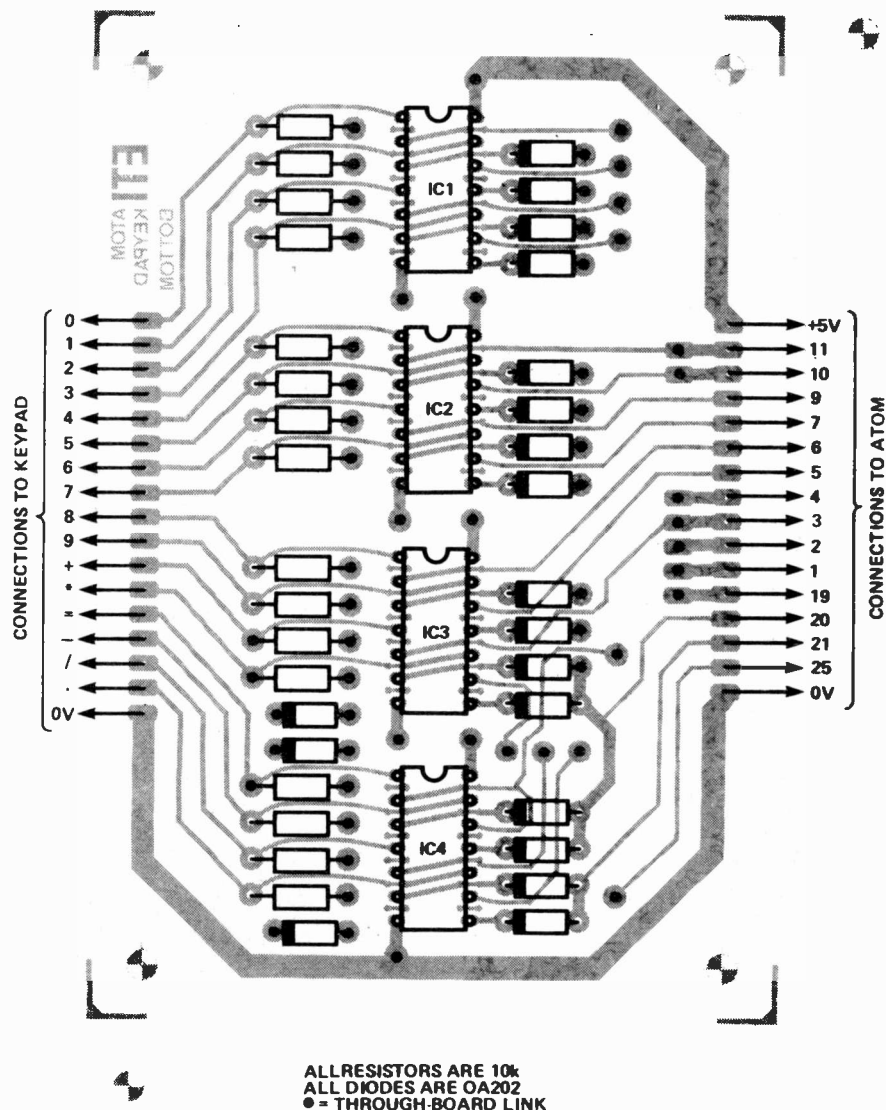


Fig. 5 Component overlay for the project. Remember to solder both sides of the board at each point indicated by a dot.

and finished boards will be on sale through our PCB Service. To avoid the cost of plated-through holes, links between the two sides of the PCB are required: these are shown by dots on the overlay. If the position has a component lead, solder it on both sides of the board — otherwise fit a through-board link.

Now comes the tricky bit. The problem is to find suitable places on the Atom masterboard to pick up connections to the key matrix. After an hour or so of probing with a multimeter we came up with the connection diagram of Fig. 5. We don't need to distinguish between the sets of pins for IC25 and IC26, incidentally, as the two have no numbers in common.

All the keyboard matrix connections can be made to the pins of the existing keyboard

switches. The supply connections are somewhat trickier. The 0 V is taken from a large area near IC22 (it will be necessary to scrape away the green varnish in order to solder the lead), while the +5 V is taken from the top end of R3. The photograph will hopefully make things clear.

The number of connections required between the Atom and the PCB is 16, including the 0 V and +5 V rails. These are conveniently provided by 15 core and earth cable, of which a length of half a meter is ample. You will have to make a hole in the Atom case to bring the cable through.

The pad is mounted in a sloping fronted Bim case, which conveniently matches the shape of the Atom case, and helps to give a uniform appearance to the finished product.

ETI

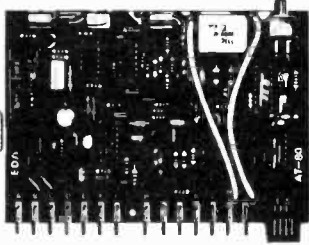
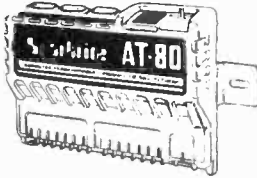
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Step-by-step fully illustrated assembly and fitting instructions are included together with circuit descriptions. Highest quality components are used throughout.

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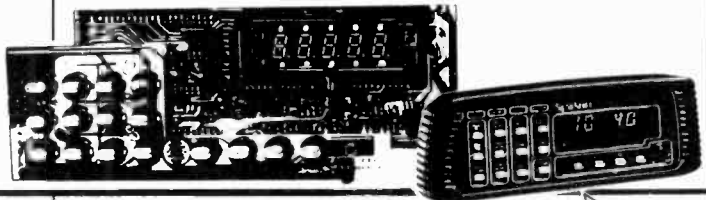


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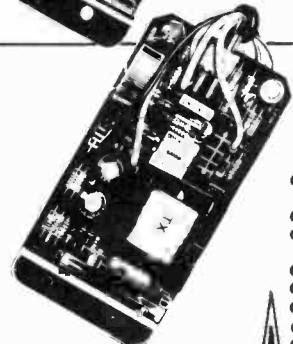
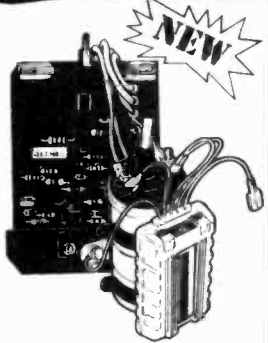


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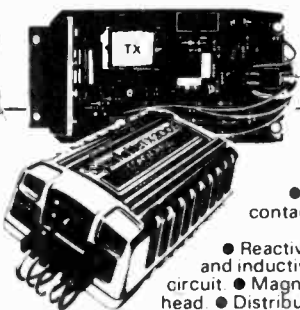
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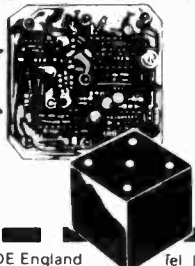


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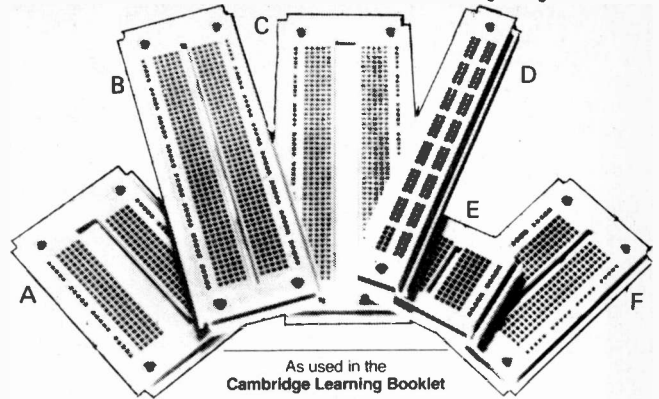
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40 type-written pages of description, specification, price lists etc. are yours for the asking (a 25p stamp and/or SAE is a help, but not essential), or telephone if you prefer. You'll have to live with your computer for a long time, so make the effort and find out all about Interak now, a couple of minutes is all it takes to ask for a leaflet!

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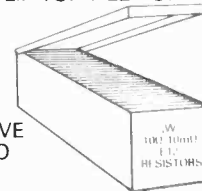
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# TECH TIPS

## Scope Bargraph Unit

Graeme Durant, Selby

This circuit is designed to be used in conjunction with any ordinary oscilloscope which has an X-deflection input, and allows it to be used as a bargraph display. The screen has 10 useable columns, thus making it suitable for use with the ETI Spectrum Analyst published recently.

The heart of the circuit is IC1, an LM3914 bargraph driver. The input to this, pin 5, is connected to a sawtooth generator running at about 1 kHz, formed around Q1. Q1 is a constant current generator supplying 5 mA and charging a 330nF capacitor to create a linear sweep. As the voltage on this capaci-

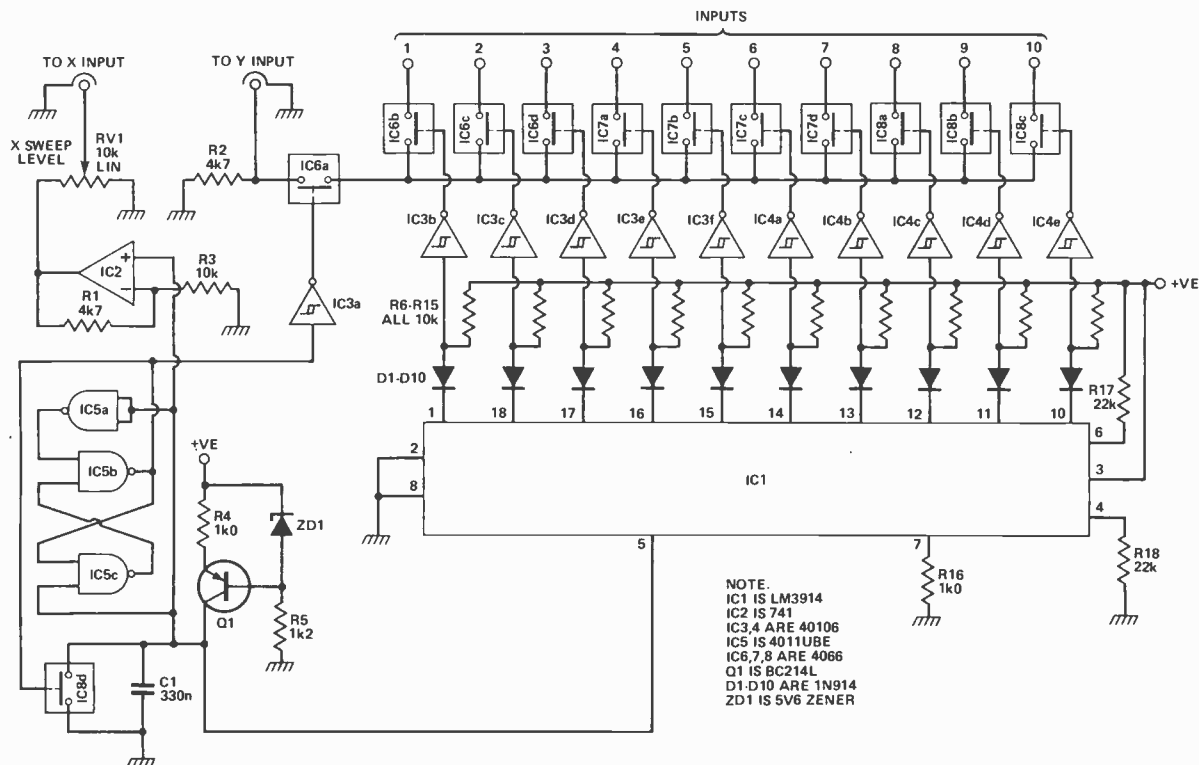
tor reaches the upper CMOS threshold, about two-thirds supply, a latch formed by IC5b and c is triggered by IC5a. This rapidly discharges the capacitor through IC8d. When the voltage has dropped to the lower CMOS level, about one-third supply, the latch is reset and the capacitor starts to charge up again. Thus a linear sawtooth waveform is produced.

This is buffered by IC2 and fed out to drive the X amplifier in the scope. However, as this sweep also drives a bargraph IC which has its upper and lower limits set to be similar to the two CMOS switching levels, the 10 outputs go low, one at a time, in sequence. These outputs are used to drive a multiplexing system: a set of 10 analogue switches (IC6b to IC8c). These are driven via inverting Schmitt triggers, diodes and pull-up resistors due to the limited drive cap-

ability of IC1 at logic 1.

The multiplexed output is sent to the scope's Y input via another analogue switch, which is normally on, but cut off while the sweep capacitor discharges so as to blank out the 'flyback'. Alternatively, the 'Z modulation' input of the scope could be used if one is available.

In use, the internal sweep generator in the scope is turned off and the circuit is connected. It is recommended that a regulated supply of 15 V is used so as to provide adequate X output drive. The X sweep level is adjusted until a suitable width of display is produced (this being a horizontal line at the present), which should be moved to the bottom of the screen. Now the inputs to the scope may be connected and the Y sensitivity of the scope adjusted to give a good display.



Tech-Tips is an ideas forum and is not aimed at the beginner. We regret we cannot answer queries on these items. ITI is prepared to consider circuits or ideas submitted by readers for this page. All items used will be paid for at a competitive rate.

Drawings should be as clear as possible and the text should be typed. Text and drawings must be on separate sheets. Circuits must not be subject to copyright. Items for consideration should be sent to ETI TECH-TIPS, Electronics Today International, 145 Charing Cross Road, London WC2H 0EE.

## Improving Crossover Performance (Zobel network)

J. P. Macauley, Crawley

When designing crossover networks, problems are encountered due to the reactive nature of the speakers. This problem can cause ringing, and more importantly the roll-offs will not occur at the frequency that the normal design calculations would indicate. It is pointless to design a crossover on the assumption that the impedance remains at a constant value throughout the operating range.

In fact a moving coil speaker looks like an inductance and a resistance in series. Because of the inductance the impedance of the speaker increases with frequency. It is not unusual to find that the impedance presented by a nominally 8 ohm speaker is nearer 15 ohms at 3 kHz.

This problem can be overcome by shunting the speakers in the pro-

posed system with a series resistor-capacitor combination. If the values of these components are correctly chosen the speaker will look like a nearly pure resistive load to the crossover. Obviously this ensures that the latter rolls the response of the speakers on and off at the calculated frequencies without problems.

The circuit is shown in Fig. 1, and the component values are determined as follows. Feed the speakers with a sine wave at the desired crossover frequency via a 100R resistor (see Fig. 2). Measure the output voltage at the output of the generator ( $V_1$ ) and the voltage across the speaker terminals ( $V_2$ ). The impedance of the speaker  $Z$  can now be determined by the equation

$$Z = \frac{100 \times V_2}{V_1 - V_2}$$

The DC resistance is now measured across the speaker terminals and  $R$  is made equal to this.  $C$  can now be determined from the equation

$$C = \frac{Z - R}{2\pi f R^2}$$

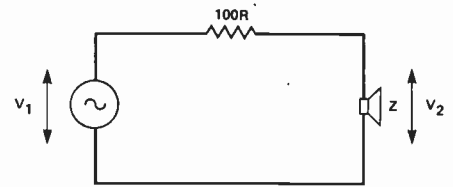


Fig. 1.

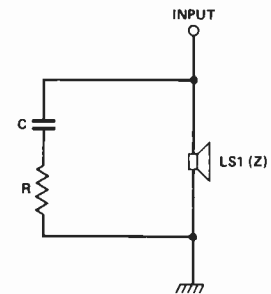


Fig. 2.

where  $f$  is the crossover frequency in Hertz and  $C$  is given in Farads. The speaker and network now present a resistive load equal to  $R$  and the crossover can safely be calculated on this assumption. **ETI**

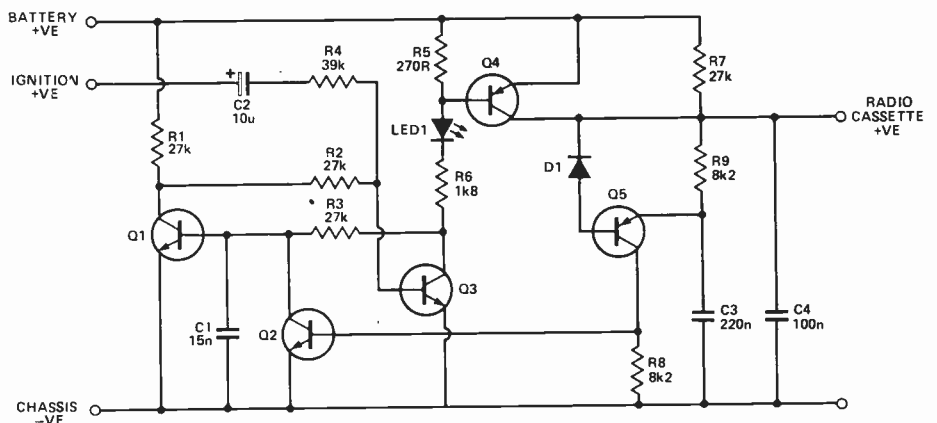
## Car Radio Latch

A. Miller, Loughborough

When fitting a car radio or cassette player into a car, one problem is deciding which side of the ignition to connect the supply lead. If it's connected to the ignition side the keys must be in to use the radio, a potential hazard if children are left listening. On the other hand, if it's connected to the battery side you have to remember to turn off the radio every time you leave the car.

The answer is simple — you connect to both using the circuit shown here. Normally the radio is left switched on and it will go on and off with the ignition. But if the ignition is off, switching the radio off and then on again also turns the radio on.

The circuit consists of a latch using Q1 and Q3, which controls a driver stage Q4. The LED indicates the state of the latch and is optional, but it doesn't consume any extra power since without it, the power would only be dissipated in R6. C2 serves to trigger the latch on and off with the ignition, and R4 prevents false triggering during starting. If the radio goes off after starting R4 should be increased, and if the radio



fails to go on and off with the ignition, R4 should be reduced.

When the latch is in the 'off' state, a small current passes through R7 to the radio. While the radio is on, C3 and C4 will remain discharged, but if the radio is off, C3 and C4 charge to the full battery voltage. If the radio is switched on, C4 rapidly discharges through the radio leaving C3 to discharge via Q5 and D1 and produce a current in R8. This turns on Q2, triggering the latch to supply power to the radio. C1 ensures reliable triggering.

Q1, 2, 3 and 5 are all general-purpose transistors, such as the BC108/BC158

types, and Q4 is a power Darlington with at least 2 A rated collector current. No heatsink should be necessary for Q4, as it is always either off or in saturation. D1 is a general-purpose diode such as the 1N4148. R4 is the only component with a critical value and may need adjusting as mentioned earlier. All the component values are those used in the prototype and any similar values should work. The quiescent power consumption is either 2 mA or 10 mA depending on the state of the latch, but if the vehicle is to be left standing for longer than a fortnight the unit (or the battery) should be disconnected. **ETI**

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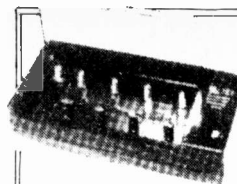
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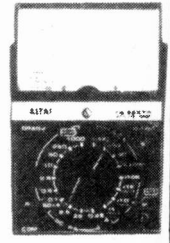
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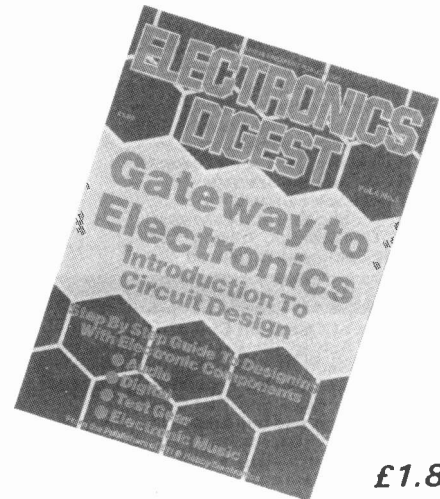
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# HYCONOMISER

Although not electronic, we believe that Vivian Capel's report on the Hyconomiser will interest ETI readers as it could affect something very near to their hearts — their wallets! After all, the more cash you save, the more you can spend on electronic goodies.

**P**etrol-saving gadgets have abounded over recent years, and even before that when the stuff was much cheaper. Savings were marginal with most, some made no difference and a few actually increased consumption! It was with great scepticism, then, that I greeted the announcement of yet another, especially as it cost the not insignificant sum of £47.50. Even today you can buy quite a few gallons of petrol for that.

The ad claimed savings of up to 20%. That phrase 'up to' of course is the let-out; 5% or even 2% could be said to qualify. However, I was impressed by the fact that all the listed suppliers and fitters were well-known main-agent garages that had reputations to lose. Fitting, incidentally, must be by a trained mechanic, and this is included in the price.

Still suspicious, I phoned all the garages and questioned the mechanics. All were enthusiastic except a Volkswagen dealer who was having troubles, which I believe have since been resolved.

So what actually is it and what does it do? Much of the wastage in a petrol engine is due to large drops of fuel which are too big to be fully burnt before they are ejected from the exhaust. Some, because of their size and mass do not move quickly enough to make it around the sharp bends of the inlet ducts, hit the walls and run down the cylinders to dilute the oil. This is aggravated at part-throttle settings by the throttle-plate, which diverts the flow toward the manifold walls. A further factor is the low-pressure region set up under the throttle-plate which tends to deflect the fuel toward the front cylinders, thus giving uneven distribution and power loss. And finally in the fuel saga, uneven atomisation produces regions of uneven mixture in the combustion chambers around the plug tips.

A formidable collection of ills, curable you might think only by completely redesigning the petrol engine! Not so, for the Hyconomiser claims to overcome them all. How? By sonic pulses.

## Blowing In The Wind

It consists of a plastic cylinder containing three plastic balls. A controlled air flow passes through the cylinder making the balls resonate at three different frequencies from 850 Hz to 2.5 kHz. The pulsed air flow is injected through a tube into the intake manifold below the carburettor. There the highest frequency produces a cone of microturbulence that smashes the fuel droplets into a fine mist with particles smaller than 20 microns.

The mixture still has to be distributed evenly to all cylinders, and evenly within each cylinder. This is the job of the middle and lower frequencies which set up areas of turbulence throughout the fuel inlet system.

The results are less consumption, more power, smoother running and, of interest to the environmentalists, less exhaust pollution. Engine and plugs also run

cleaner. Air/fuel ratio is maintained closely to the optimum 14.6:1 stoichiometric region.

Well, it sounds good but does it work? I decided to take a chance and had one fitted. Beforehand I made a careful check on my consumption in my elderly 1964 Wolseley 16/60, and for mixed town and country travel found it was 23 mpg. After fitting, I repeated the test with the same journeys. In view of the 'up to' phrase I would have been happy with 10% improvement: so I was very pleased to achieve 28 mpg, a 21% increase. Later, on a single long-distance run I did 34 mpg, but had no previous figure for this run to compare with.

I discussed this later with Hyco's managing director Tom Pearse. I learned that performances in excess of 20% were not at all uncommon, and I saw a report by one of the main-agent garages of before-and-after tests on various more recent model cars at several fixed speeds. Some were considerably higher, although this would not be a fair test of everyday motoring. Pressed for a minimum expected figure, Mr Pearse said he would be surprised if a normal production car did not attain 15% improvement.

As to cleaning the engine, I found it does this too. The makers stipulate that cars with old engines should be returned to the fitter after about 150 miles for re-tuning the device. It is in fact individually tuned for resonance with the particular engine it is fitted to. The turbulence breaks up the carbon deposits and thereby enlarges the combustion area, causing the device to de-tune. Once re-tuned the engine remains carbon-free and there is no further need of tuning.

The fitters did not inform me of this, and some thousand miles later my consumption had climbed and the device was completely off tune. Re-tuning restored optimum economy once more.

I can also confirm that the performance of my car is more sprightly although I have not pursued this too far, as my driving is definitely economy-orientated.

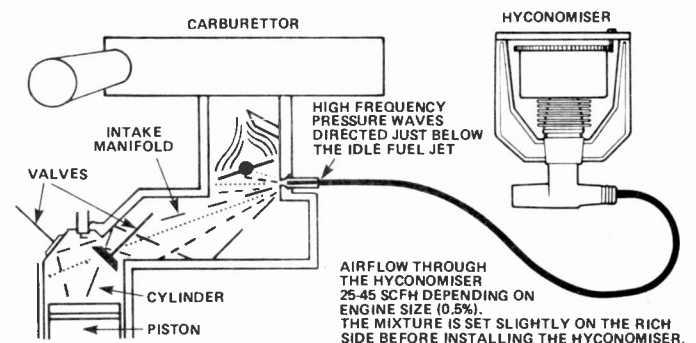


Fig. 1 Schematic diagram showing the operation of the Hyconomiser. By acoustically atomising the petrol, improved combustion and air/fuel ratio results.

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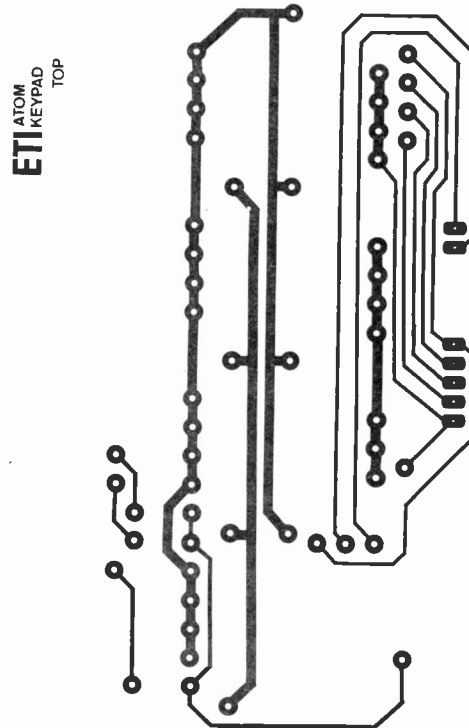
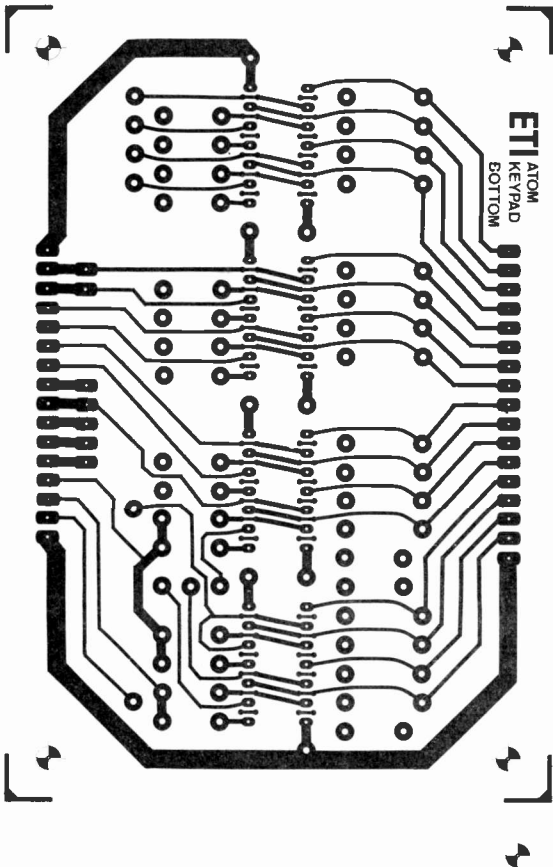
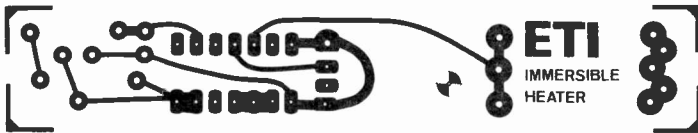
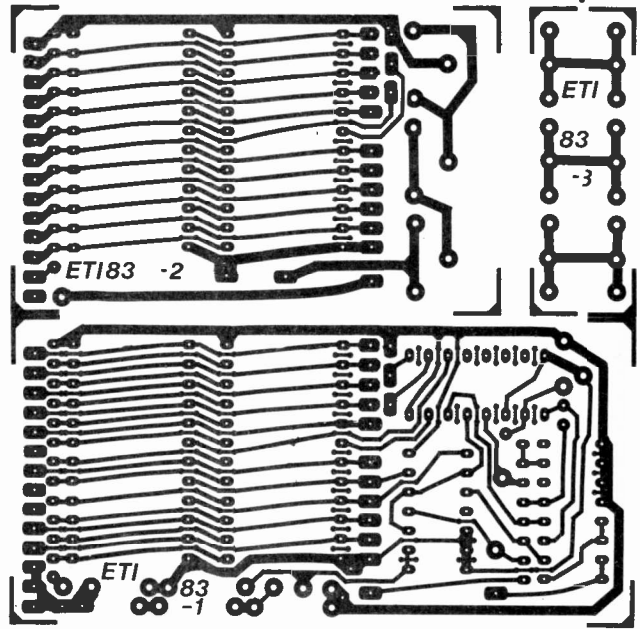
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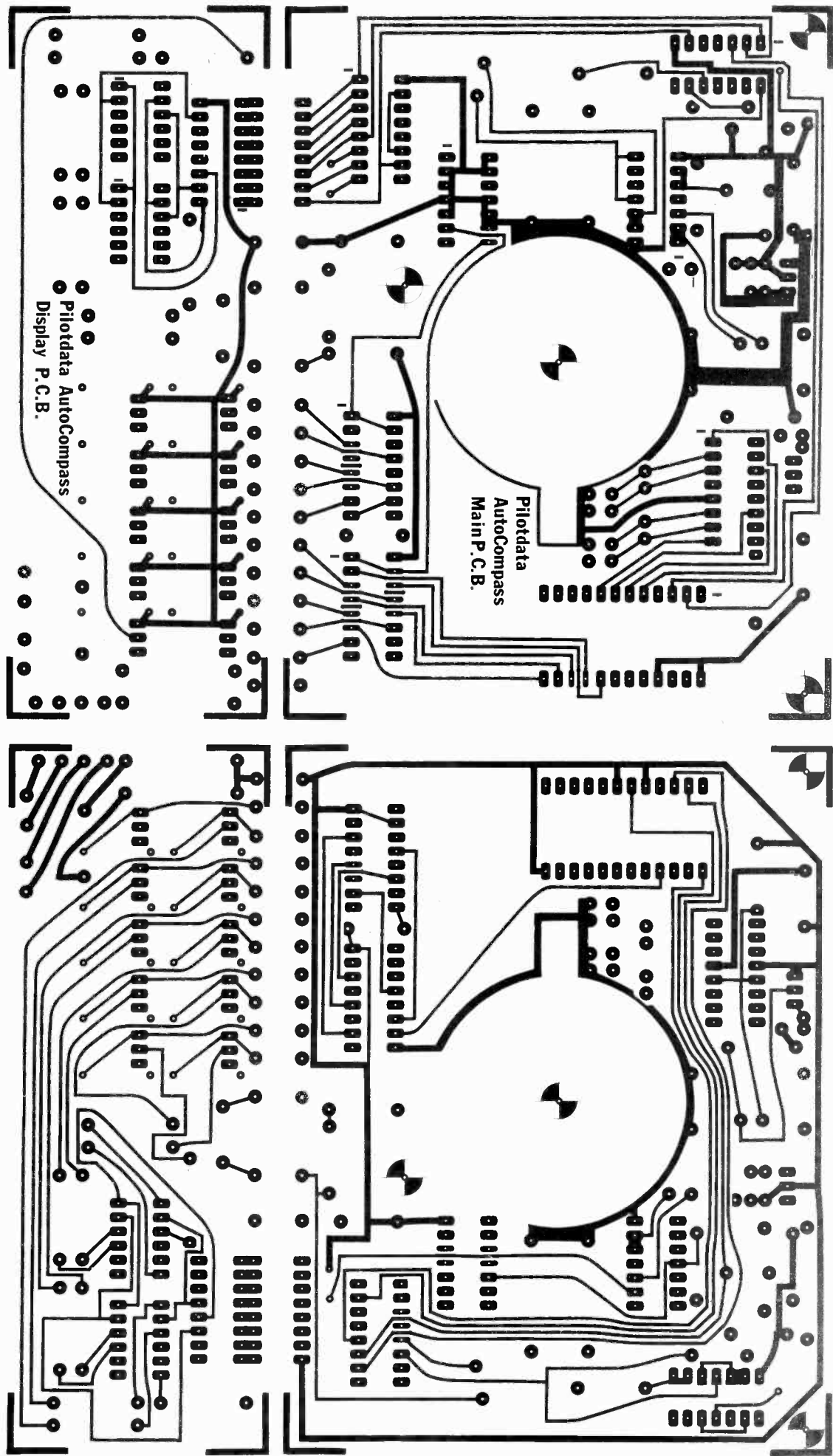
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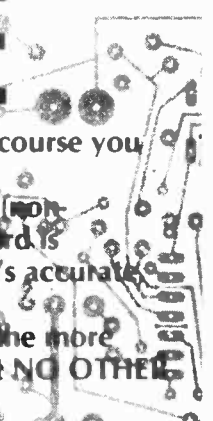


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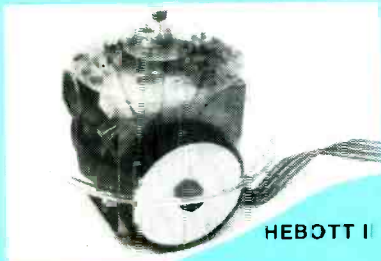
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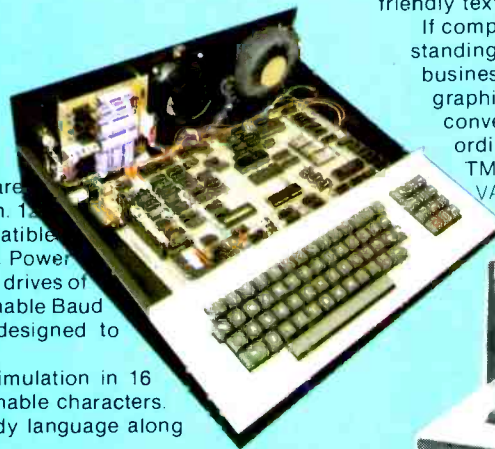
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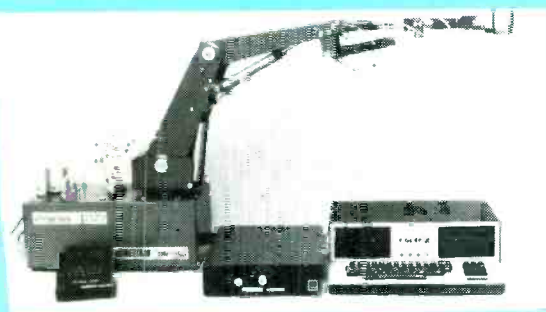
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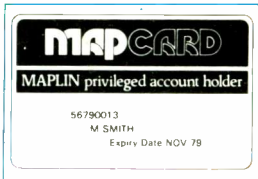
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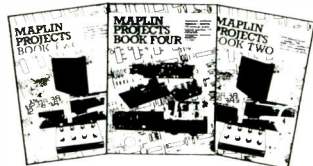
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